



 **TEXAS  
INSTRUMENTS**

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# **Data Transmission Circuits**

**Line Drivers, Receivers, Transceivers, UARTs**

*Data Book*

*Data Book*

**Data Transmission Circuits**  
**Line Drivers, Receivers, Transceivers,**  
**UARTs**

**1993**

**1993**

**Linear Products**

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***Data Transmission Circuits  
Data Book***

***Line Drivers, Receivers, Transceivers, UARTs***



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## INTRODUCTION

In the 1993 *Data Transmission Circuits Data Book*, the Linear Products Division of Texas Instruments presents technical information on various products for electronic media and electronic devices.

The Texas Instruments data transmission circuits represent technologies from classic bipolar through Advanced Low-Power Schottky (ALS), IMPACT™, LinBiCMOS™, CMOS, and BiMOS processes. The ALS and IMPACT oxide-isolated technologies provide the data transmission family with improved speed-power characteristics. LinBiCMOS technology features a step-function improvement in impedance, speed, power dissipation, and threshold stability.

This data book provides information on the following types of products:

- Data line drivers
- Data line receivers
- Data line transceivers
- Asynchronous communication elements (UARTs)

The data transmission line drivers, receivers, and transceivers, which support many popular data transmission standards, can connect electronic devices and systems at high data rates over significant cable lengths. The UARTs can control the sending and receipt of data through serial asynchronous data links.

Among new products offered by TI in the 1993 *Data Transmission Circuits Data Book* are numerous LinBiCMOS circuits for EIA RS-485, EIA-232, and IEEE 802.3 10BaseT data transmission standards. New packaging includes the shrink small-outline package (SSOP and DB) as well as surface-mount packages for popular mature products.

The data book is organized for quick location of a data sheet. The sequence is alphanumeric except for the SN prefixed parts; these data sheets are in base part number order, i.e., SN75ALS176 is located next to the SN75176B. The alphanumeric index provides a quick method of locating the data sheet for a known part number and indicates new products in this edition. The selection guide is grouped by industry standard and includes key features and the standard device footprint of the products in each category. The cross-reference guide lists other manufacturers' devices with the suggested TI replacement. Ordering information and mechanical data are in the last section of the data book.

An applications section has been added in this edition of the data book. This section is a reprint of material developed for the popular Linear Applications seminar series presented around the world. In this section are answers to the most commonly asked questions regarding data line circuits and applications.

While this data book offers design and specification data only for data transmission products, complete technical data for any TI semiconductor product is available from your nearest TI Field Sales Office, local authorized TI distributor, or by writing directly to:

Texas Instruments Incorporated  
LITERATURE RESPONSE CENTER  
P.O. Box 809066  
DALLAS, TEXAS 75380-9066

or telephone the Texas Instruments Literature Response number: 1-800-477-8924.

We sincerely believe the new 1993 *Data Transmission Circuits Data Book* will be a valuable addition to your collection of technical literature.



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<b>Line Drivers, Receivers, Transceivers</b>	<b>2</b>
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# DATA TRANSMISSION CIRCUITS SELECTION GUIDE

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				SN65173	2-549	
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1/1	25/45	5.4	SN75176	SN55LBC176	2-639	
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	11.5/18	50		TL3695	2-967	
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	22/37	55		SN65176B	2-603	
	60/35			SN75176B	2-603	
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	22/35	70		SN75177	SN75177B	2-649
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	22/37	50	MC34051	SN75ALS1178	2-925	
	35/35	110		SN751178	2-917	
3/3	13/19	90	SN75ALS170	SN75ALS170	2-505	
		22/37	72	SN75ALS171	SN75ALS171	2-519
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4/0	22	55	AM26LS31	SN65ALS172A	2-543	
	65	60		SN75ALS172A	2-543	
	22	55	SN75172	2-535		
	65	60	MC3487	SN75ALS174A	2-571	
				SN75174	2-563	
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	26.4/30.7		SN75LBC978	SN75LBC978	2-887	

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	300		uA9637	SN75146	2-337	
	25	uA9637A		2-985		
	85	50	uA9639	uA9639C	2-995	
0/4	30	14	AM26LS32	AM26C32	2-9	
	27	24		AM26LS32A	2-21	
		SN75ALS173		2-557		
	22	35		SN75ALS193	2-783	
				SN75ALS197	2-817	
	35	70		AM26LS32A	2-21	
				AM26LS33A	2-21	
				SN55173	2-549	
			SN65173	2-549		
	27	27	MC3486	SN75ALS175	2-587	
	22	35		SN55ALS195	2-805	
				SN75ALS195	2-805	
	35	70		SN75ALS199	2-839	
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	22/37	50		SN75C1167	2-909	
	35/35	110		SN75ALS1177	2-925	
	12/27	9	MC34051	SN751177	2-917	
	22/37	50		SN65C1168	2-909	
	35/35	110		SN75C1168	2-909	
				SN75ALS1178	2-925	
	3/3	13/19	90	SN75ALS170	SN75ALS170	2-505
		22/37	72	SN75ALS171	SN75ALS1711	2-943
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4/0	12	3	AM26LS31	AM26C31	2-3	
	14	45		SN55ALS192	2-773	
	22	55		SN75ALS192	2-773	
	65	60		SN65ALS172A	2-543	
	20	80		SN75ALS172A	2-543	
				SN75172	2-535	
				AM26LS31	2-13	
	14	45	MC3487	SN55ALS194	2-795	
	22	55		SN75ALS194	2-795	
	65	60		SN65ALS174A	2-571	
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			SN75189	2-751	
			MC1489A	2-751	
			SN55189A	2-751	
			SN75189A	2-751	
			SN65C189	2-759	
			SN75C189	2-759	
	SN65C189A	2-759			
	5 V or 12 V	SN75154	SN75154	2-361	
1/1	±5 V	SN75155	SN75155	2-369	
2/0	±12 V	SN75150	SN75150	2-343	
		uA9636	uA9636AC	2-979	
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			LT1081	2-59	
		MAX232	MAX232	2-71	
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		MC145406	SN65C1406	2-935	
			SN75C1406	2-935	
3/5		5 V	SN75C185	SN65C185	2-713
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		SN75LBC187	SN75LBC187	2-731	
4/0	±9 V	MC1488	MC1488	2-737	
			SN55188	2-737	
			SN75188	2-737	
			SN65C188	2-743	
	±5 V	LT1030	SN75C188	2-743	
			LT1030C	2-47	
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		SN75C198	2-829		
4/4	±12 V, 5 V	SN75186	SN75186	2-721	
	±5 V	SN75C1154	SN65C1154	2-901	
			SN75C1154	2-901	
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# DATA TRANSMISSION CIRCUITS SELECTION GUIDE

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	2/2	SN75ALS085	SN75ALS085	2-145
		SN75LBC086	SN75LBC086	2-161
IEEE 896.1 (Futurebus)	4/4	DS3893	SN75ALS053	2-97
		DS3897	SN55ALS057	2-105
		SN75ALS057	SN75ALS057	2-105
		DS3896	SN55ALS056	2-105
IEEE 488 (GPIB)	8/8	SN75160	SN75ALS056	2-105
			SN75ALS056	2-105
			SN75ALS160	2-413
		SN75161	SN75160B	2-405
			SN75ALS160	2-413
			SN55ALS161	2-437
		SN75162	SN75161B	2-423
			SN75ALS161	2-437
			SN75162B	2-423
		SN75163	SN75ALS162	2-449
			SN75163B	2-459
			SN75ALS163	2-467
SN75164	SN75164B	2-475		
	SN75ALS164	2-485		
SN75165	SN75ALS165	2-495		
IBM 360/370	0/3	N8T24	SN75124	2-267
	0/8	SN75128	SN75128	2-291
		SN75129	SN75129	2-291
	2/0	N8T23	SN75123	2-259
			SN75ALS123	2-263
			N8T23	2-259
	3/3	SN751730	SN751730	2-951
	4/0	MC3481	SN75126	2-279
			SN75ALS126	2-285
		MC3485	SN75130	2-297
SN75ALS130			2-303	
EIA RS-423	0/4	AM26LS32	AM26C32	2-9
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			SN75173	2-549
			SN75ALS173	2-557
			SN75ALS193	2-783
			SN75ALS197	2-817
		MC3486	MC3486	2-87
			SN55ALS195	2-805
			SN75ALS195	2-805
			SN65175	2-577
	0/2	SN75157	SN75157	2-379
			SN75146	2-337
		uA9637	uA9637AC	2-985
		uA9639	uA9639C	2-995
	2/0	uA9636	uA9636AC	2-979

## general purpose

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				SN55182	2-695
	Differential, $-3 < V_{ICM} < 3$ V	25 mV	SN75107	SN75207	2-851
				SN75207B	2-851
				SN55107A	2-171
				SN55107B	2-171
				SN75107A	2-171
				SN75107B	2-171
				SN75207	2-851
				SN75207B	2-851
				SN55108A	2-171
				SN55108B	2-171
	Differential, $-15 < V_{ICM} < 15$ V	1000 mV	SN75115	SN55115	2-217
				SN75115	2-217
	Single Ended	Adjustable		SN75140	2-329
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# DATA TRANSMISSION CIRCUITS SELECTION GUIDE

## general purpose (continued)

DRIVERS/RECEIVERS PER PACKAGE	TYPE OF LINE CIRCUIT	DRVR (RL)/ RVR ( $V_{th}$ )	FOOTPRINT	DEVICE TYPE	PAGE NUMBER
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				SN75122	2-253
				N8T14	2-253
0/4	ECL to TTL With Latch	ECL	DP8480	DP8480	2-39
	Differential, $-3 < V_{ICM} < 3 V$	25 mV	MC3450	MC3450	2-75
			MC3550	MC3550	2-75
			MC3452	MC3452	2-75
			MC3552	MC3552	2-75
1/1	Differential, Wired-OR	300 $\Omega$ /200 mV	SN75076B	SN65076B	2-135
	Differential, $-15 < V_{ICM} < 15 V$	100 $\Omega$ /1000 mV	SN75116	SN65076B	2-135
			SN75116	SN55116	2-227
	SN75117		SN75117	2-227	
	SN75118		SN75118	2-227	
	SN75119		SN75119	2-227	
Differential, $0 < V_{ICM} < 6 V$					
2/0	Differential, Voltage Mode	100 $\Omega$	DS8830	DS8830	2-705
				SN55183	2-705
				SN75183	2-705
2/0	Single Ended, Emitter Follower	50 $\Omega$	N8T13	SN55121	2-241
				SN75121	2-241
				SN75ALS121	2-247
				N8T13	2-259
	Differential, Current Mode	NA	SN75109	SN55109A	2-187
			SN75109	SN75109A	2-187
			SN75110	SN55110A	2-187
			SN75110	SN75110A	2-187
			SN75112	SN75112	2-187
	Differential, Voltage Mode	100 $\Omega$	SN75113	SN55113	2-197
			SN75113	SN75113	2-197
			SN75114	SN55114	2-209
			SN75114	SN75114	2-209
4/0	TTL to ECL With Latch	10K ECL	DP8481	DP8481	2-43
	Differential, Current Mode	NA	MC3453	MC3453	2-83
			MC3453	MC3553	2-83
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			N8T26	SN75136	2-311
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			SN75138	SN75138	2-317



# DATA TRANSMISSION CIRCUITS SELECTION GUIDE

## controllers

DESCRIPTION	FUNCTION	PRODUCT FEATURES	DEVICE TYPE	PACKAGE	PAGE NUMBER
UART†	Single ACE Without FIFO‡	Programmable Baud Generation	TL16C450	FN, N	3-3
UART†	Single ACE With Parallel Port and Without FIFO‡	Programmable Interface Characteristics	TL16C451	FN	3-27
UART†	Dual ACE With Parallel Port and Without FIFO‡	Programmable Interface Characteristics	TL16C452	FN	3-27
UART†	Single ACE With FIFO‡	Functional Upgrade of the 16C450	TL16C550A	FN, N	3-49

† UART – Universal Asynchronous Receivers/Transmitters

‡ FIFO – First In First Out

# DATA TRANSMISSION CIRCUITS CROSS-REFERENCE GUIDE

Texas Instruments makes no warranty as to the information furnished and the buyer assumes all risk in the use thereof. No liability is assumed for damages resulting from the use of the information contained herein.

Manufacturers are arranged in alphabetical order.

AMD	SUGGESTED TI REPLACEMENT	PAGE NO.
AM26LS31	AM26LS31	2-13
AM26LS32	AM26LS32A	2-21
AM26LS32A	AM26LS32A	2-21
AM26LS33	AM26LS33A	2-21
AM26S10	AM26S10	2-31
AM26S11	AM26S11	2-31
26LS31	AM26LS31	2-13
26LS32	AM26LS32A	2-21
AT&T	SUGGESTED TI REPLACEMENT	PAGE NO.
41LF	AM26LS32A	2-21
41LG	AM26LS31	2-13
LINEAR TECHNOLOGY	SUGGESTED TI REPLACEMENT	PAGE NO.
LT1030	LT1030C	2-47
LT1039	LT1039	2-53
LT1080	LT1080	2-59
LT1081	LT1081	2-59
MAXIM	SUGGESTED TI REPLACEMENT	PAGE NO.
MAX232	MAX232	2-71
MAX241	SN75LBC241	2-859
MOTOROLA	SUGGESTED TI REPLACEMENT	PAGE NO.
AM26LS31	AM26LS31	2-13
AM26LS32	AM26LS32A	2-21
MC1488	SN75188	2-737
MC1489	SN75189	2-751
MC1489A	SN75189A	2-751
MC26S10	AM26S10	2-31
MC3450	MC3450	2-75
MC3452	MC3452	2-75
MC3453	MC3453	2-83
MC3481	SN75126	2-279
MC3485	SN75130	2-297



# DATA TRANSMISSION CIRCUITS CROSS-REFERENCE GUIDE

MOTOROLA	SUGGESTED TI REPLACEMENT	PAGE NO.
MC3486	MC3486	2-87
MC3487	MC3487	2-93
MC3488A	uA9636A	2-979
MC34050	SN751177	2-917
MC34051	SN751178	2-917
MC6880A	SN75136	2-311
MC75107	SN75107A	2-171
MC75107	SN75107B	2-171
MC75108	SN75108A	2-171
MC75108	SN75108B	2-171
MC75127	SN75127	2-273
MC75128	SN75128	2-291
MC75129	SN75129	2-291
MC75S110	SN75110A	2-187
MC8T26A	SN75136	2-311
SN75172	SN75172	2-535
SN75173	SN75173	2-549
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NATIONAL SEMICONDUCTOR	SUGGESTED TI REPLACEMENT	PAGE NO.
DM26LS32	AM26LS32A	2-21
DP8480	DP8480	2-39
DP8481	DP8481	2-43
DS1488	SN75188	2-737
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DS14C241	SN75LBC241	2-859
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DS16F95	SN95176B	2-615
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DS26C32A	AM26C32	2-9
DS26F31	SN75ALS192	2-773
DS26F32	SN75ALS193	2-783
DS26LS31	AM26LS31	2-13
DS26LS32	AM26LS32A	2-21
DS26LS33	AM26LS33A	2-21
DS26LS33A	AM26LS33A	2-21
DS26S10	AM26S10	2-31
DS26S11	AM26S11	2-31
DS3486	MC3486	2-87

## DATA TRANSMISSION CIRCUITS CROSS-REFERENCE GUIDE

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DS3487	MC3487	2-93
DS34F86	SN75ALS195	2-805
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DS35F86	SN55ALS195	2-805
DS35F87	SN55ALS194	2-795
DS3603	SN75107A	2-171
DS3603	SN75107B	2-171
DS3603	SN75108A	2-171
DS3603	SN75108B	2-171
DS3650	MC3450	2-75
DS3652	MC3452	2-75
DS3695	SN75176B	2-603
DS3695	TL3695	2-967
DS3695A	SN75176B	2-603
DS3695A	TL3695	2-967
DS3697	SN75177B	2-649
DS36F95	SN75ALS176	2-627
DS55107	SN55107B	2-171
DS55108	SN55108B	2-171
DS55110A	SN55110A	2-187
DS55113	SN55113	2-197
DS55121	SN55121	2-241
DS55122	SN55122	2-253
DS55173	SN55173	2-549
DS75107	SN75107A	2-171
DS75107A	SN75107B	2-171
DS75108	SN75108A	2-171
DS75108A	SN75108B	2-171
DS75113	SN75113	2-197
DS75114	SN75114	2-209
DS75115	SN75115	2-217
DS75121	SN75121	2-241
DS75123	SN75123	2-259
DS75124	SN75124	2-267
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DS75176B	SN75176B	2-603
DS75176B	TL3695	2-967
DS75176BT	SN65176B	2-603
DS7820A	SN55182	2-695
DS7830	SN55183	2-705
DS8820	DS8820A	2-695
DS8830	DS8830	2-705
DS8830	SN75182	2-695
DS8832	SN75183	2-705

# DATA TRANSMISSION CIRCUITS CROSS-REFERENCE GUIDE

NATIONAL SEMICONDUCTOR	SUGGESTED TI REPLACEMENT	PAGE NO.
DS96110A	SN75110A	2-187
DS9614	SN55114	2-209
DS9615	SN55115	2-217
DS96172	SN75172	2-535
DS96173	SN75173	2-549
DS96174	SN75174	2-563
DS96175	SN75175	2-577
DS96177	SN75177B	2-649
DS9636A	uA9636AC	2-979
DS9637A	uA9637AC	2-985
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DS9639A	uA9639C	2-995
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DS96F175	SN75ALS175	2-587
MC145406	SN75C1406	2-935
uA26LS32	AM26LS32A	2-21
uA9636A	uA9636AC	2-979
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uA9639	uA9639C	2-995

SIGNETICS	SUGGESTED TI REPLACEMENT	PAGE NO.
AM26LS31	AM26LS31	2-13
AM26LS32	AM26LS32A	2-21
AM26LS33	AM26LS33A	2-21
MC1488	SN75188	2-737
MC1489	MC1489	2-751
MC1489A	MC1489A	2-751

<b>General Information</b>	<b>1</b>
<b>Line Drivers, Receivers, Transceivers</b>	<b>2</b>
<b>Universal Async Receivers/Transmitters</b>	<b>3</b>
<b>Explanation of Logic Symbols</b>	<b>4</b>
<b>Applications</b>	<b>5</b>
<b>Mechanical Data</b>	<b>6</b>

## **2 Line Drivers, Receivers, Transceivers**

# AM26C31C, AM26C31I QUADRUPLE DIFFERENTIAL LINE DRIVERS

SLLS103B – D3636, DECEMBER 1990 – REVISED JANUARY 1993

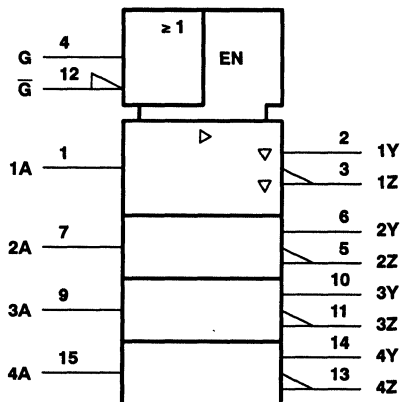
- Meets EIA Standard RS-422-A and CCITT Recommendation V.11
- Low Power,  $I_{CC} = 100 \mu A$  Typ
- Operates From a Single 5-V Supply
- High Speed,  $t_{PLH} = t_{PHL} = 7 \text{ ns}$  Typ
- Low Pulse Distortion ( $t_{sk(p)} = 0.5 \text{ ns}$  Typ)
- High Output Impedance in Power-Off Conditions
- Direct Replacement for National Semiconductor DS26C31
- Improved Replacement for AM26LS31
- ESD Protection Exceeds 2000 V Per MIL-STD-883C, Method 3015

## description

The AM26C31C and AM26C31I are quadruple complementary-output line drivers designed to meet the requirements of EIA Standard RS-422-A and CCITT V.11. The 3-state outputs have high-current capability for driving balanced lines such as twisted-pair or parallel-wire transmission lines, and they provide a high-impedance state in the power-off condition. The enable function is common to all four drivers and offers the choice of an active-high or active-low enable input. BiCMOS circuitry reduces power consumption without sacrificing speed.

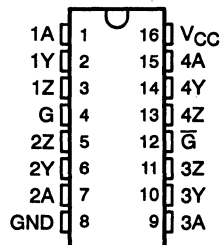
The AM26C31C is characterized for operation from 0°C to 70°C, and the AM26C31I is characterized for operation from -40°C to 85°C.

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## D OR N PACKAGE (TOP VIEW)

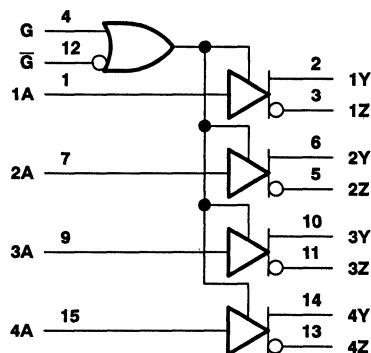


## FUNCTION TABLE (each driver)

INPUT A	ENABLES		OUTPUTS	
	G	G-bar	Y	Z
H	H	X	H	L
L	H	X	L	H
H	X	L	H	L
L	X	L	L	H
X	L	H	Z	Z

H = high level      X = irrelevant  
L = low level      Z = high impedance (off)

## logic diagram (positive logic)



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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# AM26C31C, AM26C31I QUADRUPLE DIFFERENTIAL LINE DRIVERS

SLLS103B - D3636, DECEMBER 1990 - REVISED JANUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	-0.5 V to 7 V
Input voltage range, $V_I$	-0.5 V to $V_{CC} + 0.5$ V
Output voltage range, $V_O$	-0.5 V to 7 V
Input or output clamp current, $I_{IK}$ or $I_{OK}$	$\pm 20$ mA
Output current, $I_O$	$\pm 150$ mA
$V_{CC}$ current	200 mA
GND current	-200 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : AM26C31C	0°C to 70°C
AM26C31I	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values, except differential output voltage  $V_{OD}$ , are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.5	5	5.5	V
High-level input voltage, $V_{IH}$		2			V
Low-level input voltage, $V_{IL}$				0.8	V
High-level output current, $I_{OH}$				-20	mA
Low-level output current, $I_{OL}$				20	mA
Operating free-air temperature, $T_A$	AM26C31C		0	70	°C
	AM26C31I		-40	85	

# AM26C31C, AM26C31I QUADRUPLE DIFFERENTIAL LINE DRIVERS

SLLS103B – D3636, DECEMBER 1990 – REVISED JANUARY 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>O</sub> = -20 mA	2.4	3.4		V
V <sub>OL</sub>	Low-level output voltage	I <sub>O</sub> = 20 mA		0.2	0.4	V
V <sub>OD</sub>	Differential output voltage	R <sub>L</sub> = 100 Ω	2	3.1		V
Δ V <sub>OD</sub>	Change in magnitude of differential output voltage‡	R <sub>L</sub> = 100 Ω			±0.4	V
V <sub>OC</sub>	Common-mode output voltage	R <sub>L</sub> = 100 Ω			3	V
Δ V <sub>OC</sub>	Change in magnitude of common-mode output voltage‡	R <sub>L</sub> = 100 Ω			±0.4	V
I <sub>I</sub>	Input current	V <sub>I</sub> at V <sub>CC</sub> , V <sub>IH</sub> , V <sub>IL</sub> , or GND			±1	μA
I <sub>O(off)</sub>	Driver output current with power off	V <sub>CC</sub> = 0, V <sub>O</sub> = 6 V			100	μA
		V <sub>CC</sub> = 0, V <sub>O</sub> = -0.25 V			-100	μA
I <sub>OS</sub>	Driver output short-circuit current	V <sub>O</sub> = 0	-30		-150	mA
I <sub>OZ</sub>	Off-state (high-impedance state) output current	V <sub>O</sub> = 2.5 V			20	μA
		V <sub>O</sub> = 0.5 V			-20	μA
I <sub>CC</sub>	Quiescent supply current	I <sub>O</sub> = 0, V <sub>I</sub> = 0 V or 5 V			100	μA
		I <sub>O</sub> = 0, V <sub>I</sub> = 2.4 V or 0.5 V, See Note 2		1.5	3	mA
C <sub>i</sub>	Input capacitance			6		pF

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

‡ Δ|V<sub>OD</sub>| and Δ|V<sub>OC</sub>| are the changes in magnitude of V<sub>OD</sub> and V<sub>OC</sub>, respectively, that occur when the input is changed from a high level to a low level.

NOTE 2: Measured per input. All other inputs are at 0 V or 5 V.

## switching characteristics, V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	See Figures 1 and 2, S1 is open		7	12	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output			7	12	ns
t <sub>sk(p)</sub>	Pulse skew ( t <sub>PLH</sub> - t <sub>PHL</sub>  )			0.5	4	ns
t <sub>rD</sub> , t <sub>fD</sub>	Differential output rise and fall times	See Figures 1 and 4, S1 is open		5	10	ns
t <sub>pZH</sub>	Output enable time to high level	See Figures 1 and 3, S1 is closed		10	19	ns
t <sub>pZL</sub>	Output enable time to low level			10	19	ns
t <sub>pHZ</sub>	Output disable time from high level			7	16	ns
t <sub>pLZ</sub>	Output disable time from low level			7	16	ns
C <sub>pd</sub>	Power dissipation capacitance (see Note 3)	No load		100		pF

NOTE 3: C<sub>pd</sub> is used to estimate the switching losses according to P<sub>D</sub> = C<sub>pd</sub> V<sub>CC</sub><sup>2</sup> f where P<sub>D</sub> is in watts, C<sub>pd</sub> is in farads, V<sub>CC</sub> is in volts, and f is in hertz.

# AM26C31C, AM26C31I QUADRUPLE DIFFERENTIAL LINE DRIVERS

SLLS103B – D3636, DECEMBER 1990 – REVISED JANUARY 1993

## PARAMETER MEASUREMENT INFORMATION

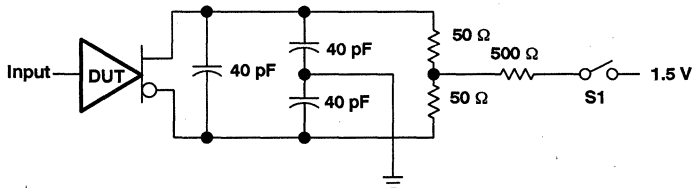


Figure 1. Test Circuit

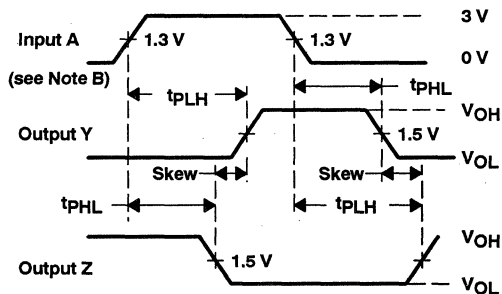


Figure 2. Propagation Delay Times and Skew Waveforms

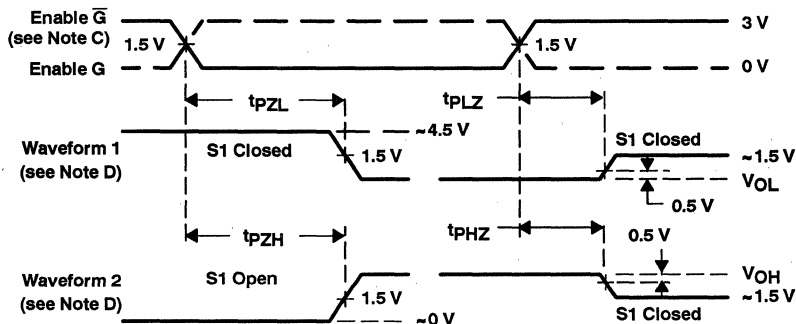


Figure 3. Enable and Disable Time Waveforms

- NOTES: A. All input pulses are supplied by generators having the following characteristics:  $PRR \leq 1 \text{ MHz}$ ,  $Z_O \sim 50 \Omega$ ,  $t_r \leq 15 \text{ ns}$ , and  $t_f \leq 6 \text{ ns}$ .  
 B. When measuring propagation delay times and skew, switch S1 is open.  
 C. Each enable is tested separately.  
 D. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.

# AM26C31C, AM26C31I QUADRUPLE DIFFERENTIAL LINE DRIVERS

SLLS103B – D3636, DECEMBER 1990 – REVISED JANUARY 1993

## PARAMETER MEASUREMENT INFORMATION

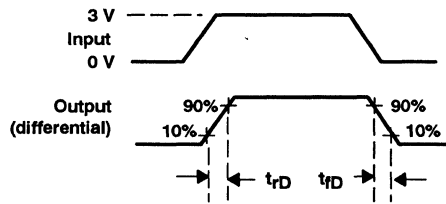


Figure 4. Differential Output Rise and Fall Times

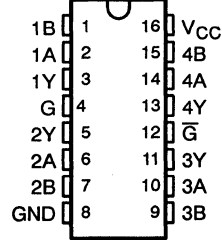


# AM26C32C, AM26C32I QUADRUPLE DIFFERENTIAL LINE RECEIVERS

SLLS104B - D3634, DECEMBER 1990 - REVISED DECEMBER 1992

- Meets EIA Standards RS-422-A, RS-423-A, and CCITT Recommendation V.11
- Low Power,  $I_{CC} = 9 \text{ mA Typ}$
- $\pm 7\text{-V}$  Common-Mode Range With  $\pm 200\text{-mV}$  Sensitivity
- Input Hysteresis . . . 60 mV Typical
- $t_{pd} = 19 \text{ ns (Typ)}$
- Operates From a Single 5-V Supply
- 3-State Outputs
- Input Fail-Safe Circuitry
- Improved Replacement for AM26LS32

D, N, OR NS† PACKAGE  
(TOP VIEW)



† The NS package is available in left-ended taped and reeled (order device AM26C32CNSLE).

## description

The AM26C32C and AM26C32I are quadruple differential line receivers for balanced and unbalanced digital data transmission. The enable function is common to all four receivers and offers a choice of active-high or active-low input. Three-state outputs permit connection directly to a bus-organized system. Fail-safe design ensures that if the inputs are open, the outputs will always be high.

The AM26C32 is manufactured using a BiCMOS process, which is a combination of bipolar and CMOS transistors. This process provides the high voltage and current of bipolar with the low power of CMOS to reduce the power consumption to about one-fifth that of the standard AM26LS32 while still maintaining ac and dc performance.

The AM26C32C is characterized for operation from 0°C to 70°C, and the AM26C32I is characterized from -40°C to 85°C.

FUNCTION TABLE  
(each receiver)

DIFFERENTIAL INPUT	ENABLES		OUTPUT
	G	$\bar{G}$	
$V_{ID} \geq V_{T+}$	H	X	H
	X	L	H
$V_{T-} \leq V_{ID} \leq V_{T+}$	H	X	?
	X	L	?
$V_{ID} \leq V_{T-}$	H	X	L
	X	L	L
X	L	H	Z

H = high level, L = low level, X = irrelevant  
Z = high impedance (off), ? = indeterminate

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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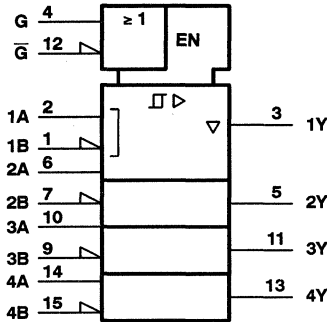
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# AM26C32C, AM26C32I QUADRUPLE DIFFERENTIAL LINE RECEIVERS

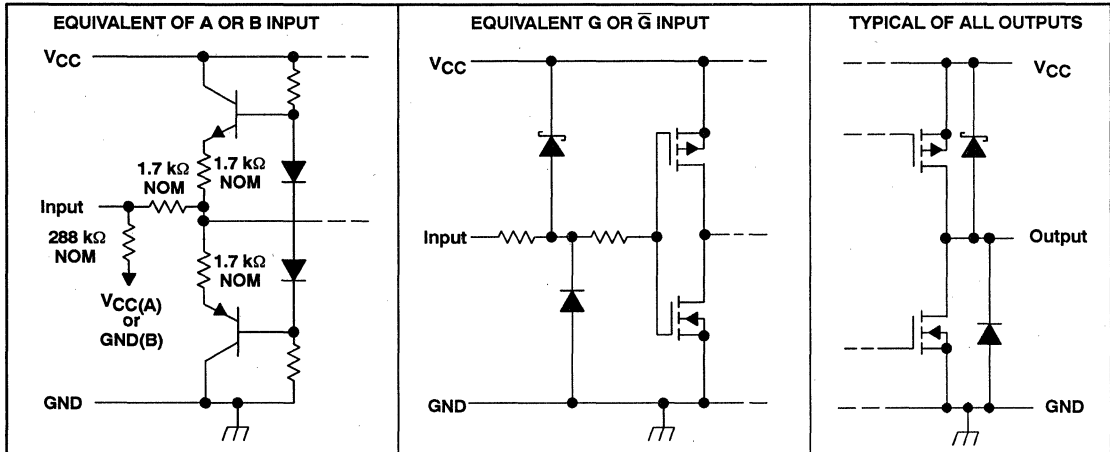
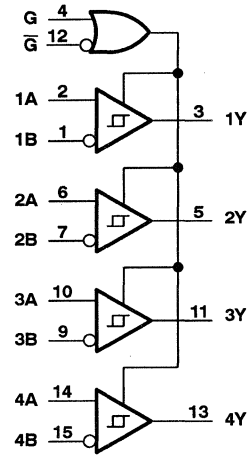
SLLS104B – D3634, DECEMBER 1990 – REVISED DECEMBER 1992

**logic symbol†**



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

**logic diagram (positive logic)**



**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, V <sub>CC</sub> (see Note 1)	7 V
Input voltage range, V <sub>i</sub> : A or B inputs	-11 V to 14 V
G or G-bar inputs	7 V
Output voltage, V <sub>O</sub>	7 V
Output current, I <sub>O</sub>	±25 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T <sub>A</sub> : AM26C32C	0°C to 70°C
AM26C32I	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values, except differential output voltage, V<sub>OD</sub>, are with respect to network ground terminal. Currents into the device are positive and currents out of the device are negative.

# AM26C32C, AM26C32I

## QUADRUPLE DIFFERENTIAL LINE RECEIVERS

SLLS104B – D3634, DECEMBER 1990 – REVISED DECEMBER 1992

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/ $^\circ\text{C}$	608 mW	494 mW
N	1150 mW	9.2 mW/ $^\circ\text{C}$	736 mW	598 mW
NS	625 mW	5.0 mW/ $^\circ\text{C}$	400 mW	325 mW

### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.5	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$				0.8 V
Common-mode input voltage, $V_{IC}$				$\pm 7$ V
High-level output current, $I_{OH}$				-6 mA
Low-level output current, $I_{OL}$				6 mA
Operating free-air temperature, $T_A$	AM26C32C		0	70
	AM26C32I		-40	85

### electrical characteristics over recommended ranges of $V_{CC}$ , $V_{IC}$ , and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Differential input high-threshold voltage	$V_O = V_{OH\text{ min}}$ , $I_{OH} = -440 \mu\text{A}$			0.2	V
$V_{T-}$ Differential input low-threshold voltage	$V_O = 0.45 \text{ V}$ , $I_{OL} = 8 \text{ mA}$	-0.2‡			V
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )			60		mV
$V_{IK}$ Enable input clamp voltage	$V_{CC} = 4.5 \text{ V}$ , $I_I = -18 \text{ mA}$			-1.5	V
$V_{OH}$ High-level output voltage	$V_{ID} = 200 \text{ mV}$ , $I_{OH} = -6 \text{ mA}$	2.7			V
$V_{OL}$ Low-level output voltage	$V_{ID} = -200 \text{ mV}$ , $I_{OL} = 6 \text{ mA}$		0.2	0.3	V
$I_{OZ}$ Off-state (high-impedance-state) output current	$V_O = V_{CC}$ or GND		$\pm 0.5$	$\pm 5$	$\mu\text{A}$
$I_I$ Line input current	$V_I = 10 \text{ V}$ , Other input at 0 V			1.5	mA
	$V_I = -10 \text{ V}$ , Other input at 0 V			-2.5	
$I_{IH}$ High-level enable current	$V_I = 2.7 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$ Low-level enable current	$V_I = 0.4 \text{ V}$			-100	$\mu\text{A}$
$r_i$ Input resistance	One input to ac ground		17		k $\Omega$
$I_{CC}$ Supply current	$V_{CC} = 5.5 \text{ V}$	All outputs disabled	9	14	mA
		All outputs enabled	12		





# AM26C32C, AM26C32I QUADRUPLE DIFFERENTIAL LINE RECEIVERS

SLLS104B - D3634, DECEMBER 1990 - REVISED DECEMBER 1992

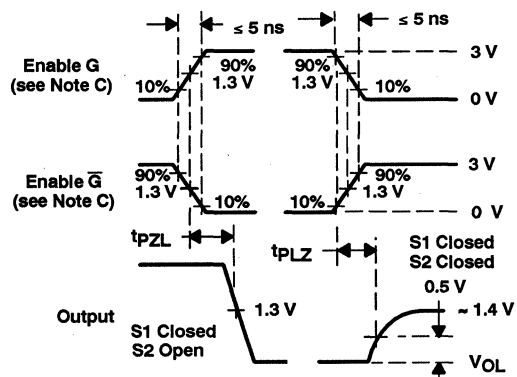
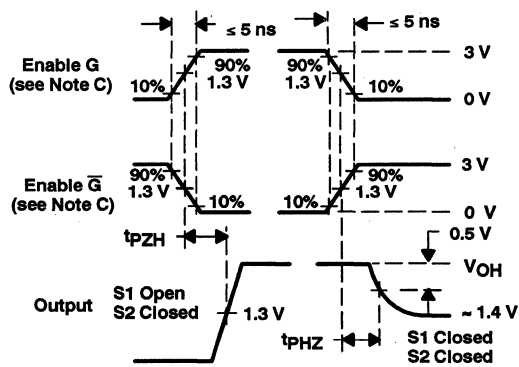
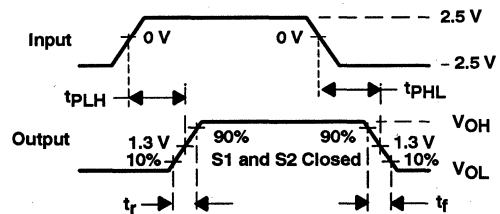
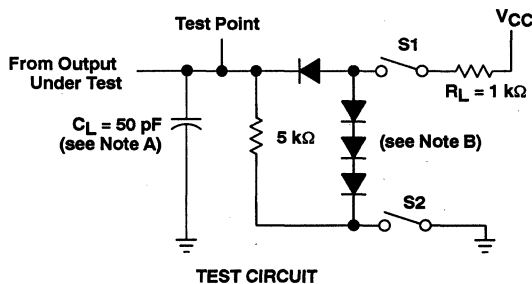
switching characteristics over recommended ranges of  $V_{CC}$ ,  $V_{IC}$ , and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	See Figure 1	10	19	30	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output		10	19	30	ns
$t_f$ Output fall time		4	9	ns	
$t_r$ Output rise time		4	9	ns	
$t_{PZH}$ Output enable time to high level		13	22	ns	
$t_{PZL}$ Output enable time to low level		13	22	ns	
$t_{PHZ}$ Output disable time from high level		13	22	ns	
$t_{PLZ}$ Output disable time from low level		13	22	ns	

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ , and  $V_{IC} = 0$ .

‡ The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold levels only.

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A.  $C_L$  includes probe and jig capacitance.  
 B. All diodes are 1N3064 or equivalent.  
 C. Enable G is tested with  $\bar{G}$  high;  $\bar{G}$  is tested with G low.

Figure 1. Test Circuit and Voltage Waveforms

TEXAS  
INSTRUMENTS

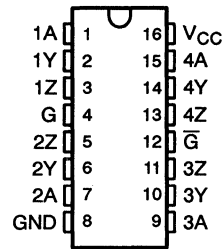
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# AM26LS31C QUAD DIFFERENTIAL LINE DRIVER

SLLS114 – D2433, JANUARY 1979 – REVISED MAY 1990

- Meets EIA Standard RS-422-A
- Operates From a Single 5-V Supply
- TTL Compatible
- Complementary Outputs
- High Output Impedance in Power-Off Conditions
- Complementary Output Enable Inputs

D OR N PACKAGE  
(TOP VIEW)



## description

The AM26LS31C is a quad complementary-output line driver designed to meet the requirements of EIA Standard RS-422-A Federal Standard 1020. The 3-state outputs have high-current capability for driving balanced lines such as twisted-pair or parallel-wire transmission lines, and they provide a high-impedance state in the power-off condition. The enable function is common to all four drivers and offers the choice of an active-high or active-low enable input. Low-power Schottky circuitry reduces power consumption without sacrificing speed.

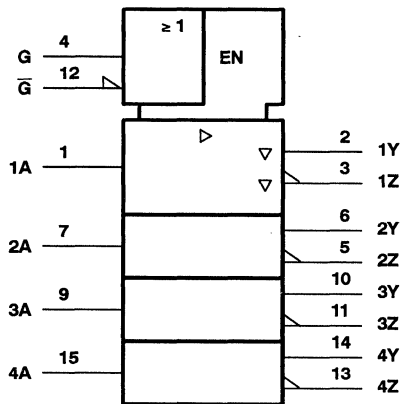
The AM26LS31C is characterized for operation from 0°C to 70°C.

FUNCTION TABLE  
(each driver)

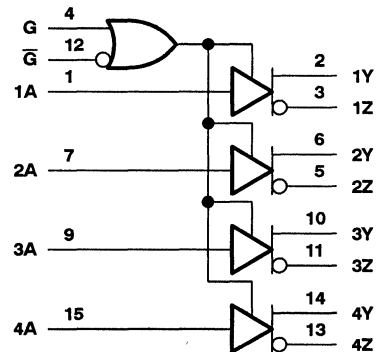
INPUT A	ENABLES		OUTPUTS	
	G	$\bar{G}$	Y	Z
H	H	X	H	L
L	H	X	L	H
H	X	L	H	L
L	X	L	L	H
X	L	H	Z	Z

H = high level      X = irrelevant  
L = low level      Z = high impedance (off)

## logic symbol†



## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

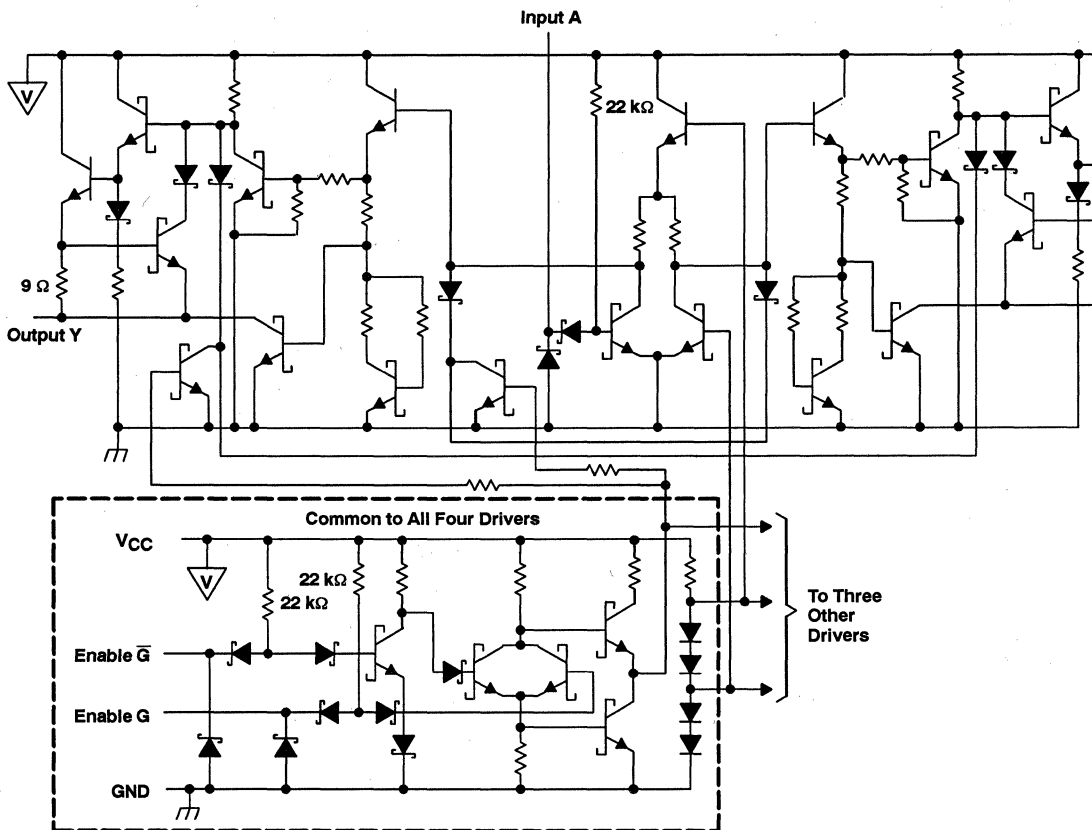
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# AM26LS31C QUAD DIFFERENTIAL LINE DRIVER

SLLS114 - D2433, JANUARY 1979 - REVISED MAY 1990

## schematic (each driver)



All resistor values are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	7 V
Output offstate voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values, except differential output voltage  $V_{OD}$ , are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

# AM26LS31C QUAD DIFFERENTIAL LINE DRIVER

SLLS114 – D2433, JANUARY 1979 – REVISED MAY 1990

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-20	mA
Low-level output current, $I_{OL}$			20	mA
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$V_{CC} = 4.75$ V, $I_I = -18$ mA			-1.5	V
$V_{OH}$	High-level output voltage	$V_{CC} = 4.75$ V, $I_{OH} = -20$ mA	2.5			V
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.75$ V, $I_{OL} = 20$ mA			0.5	V
$I_{OZ}$	Off-state (high-impedance state) output current	$V_{CC} = 4.75$ V, $V_O = 0.5$ V			-20	$\mu$ A
		$V_{CC} = 4.75$ V, $V_O = 2.5$ V			20	
$I_I$	Input current at maximum input voltage	$V_{CC} = 5.25$ V, $V_I = 7$ V			0.1	mA
$I_{IH}$	High-level input current	$V_{CC} = 5.25$ V, $V_I = 2.7$ V			20	$\mu$ A
$I_{IL}$	Low-level input current	$V_{CC} = 5.25$ V, $V_I = 0.4$ V			-0.36	$\mu$ A
$I_{OS}$	Short-circuit output current‡	$V_{CC} = 5.25$ V	-30		-150	mA
$I_{CC}$	Supply current	$V_{CC} = 5.25$ V, All outputs disabled		32	80	mA

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .

‡ Not more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

## switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

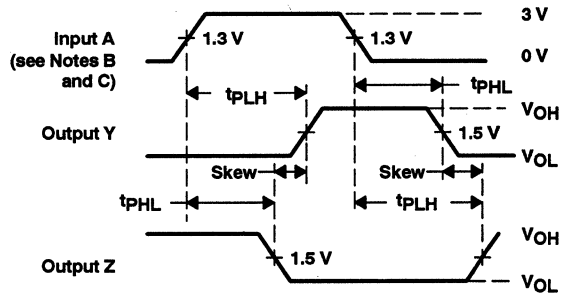
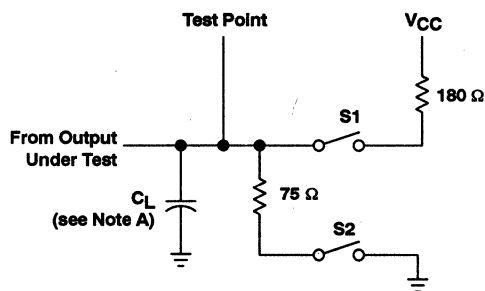
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$C_L = 30$ pF, S1 and S2 open, See Figure 1		14	20	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output			14	20	ns
	Output-to-output skew			1	6	ns
$t_{PZH}$	Output enable time to high level	$C_L = 30$ pF, $R_L = 75$ $\Omega$ , See Figure 1		2.5	40	ns
$t_{PZL}$	Output enable time to low level	$C_L = 30$ pF, $R_L = 180$ $\Omega$ , See Figure 1		37	45	ns
$t_{PHZ}$	Output disable time from high level	$C_L = 10$ pF, S1 and S2 closed, See Figure 1		21	30	ns
$t_{PLZ}$	Output disable time from low level			23	35	ns



# AM26LS31C QUAD DIFFERENTIAL LINE DRIVER

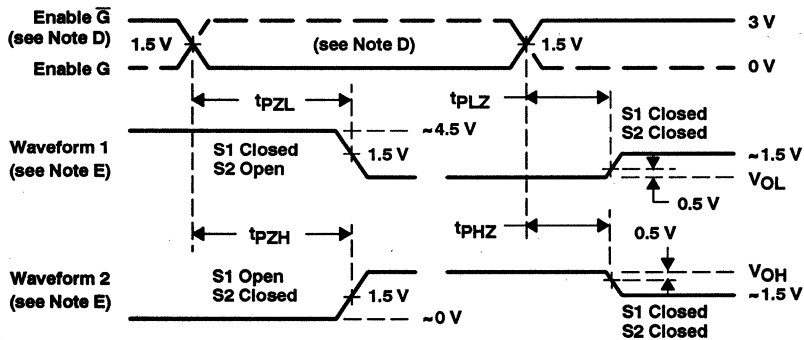
SLLS114 – D2433, JANUARY 1979 – REVISED MAY 1990

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT

PROPAGATION DELAY TIMES AND SKEW



ENABLE AND DISABLE TIME WAVEFORMS

- NOTES: A.  $C_L$  includes probe and jig capacitance.
- B. All input pulses are supplied by generators having the following characteristics: PRR  $\leq$  1 MHz,  $Z_O \approx 50 \Omega$ ,  $t_r \leq 15$  ns, and  $t_f \leq 6$  ns.
- C. When measuring propagation delay times and skew, switches S1 and S2 are open.
- D. Each enable is tested separately.
- E. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.

Figure 1. Test Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

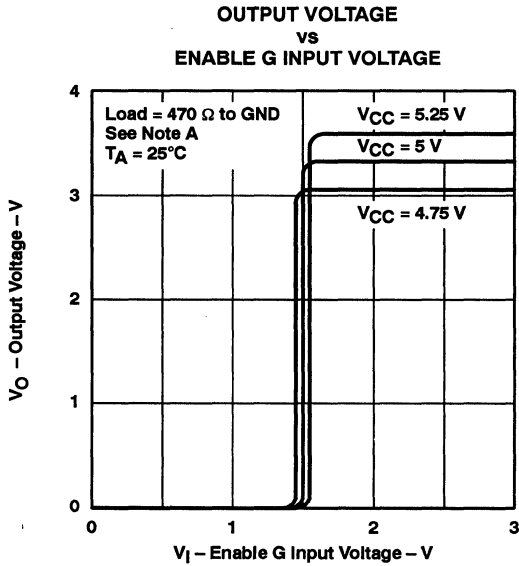


Figure 2

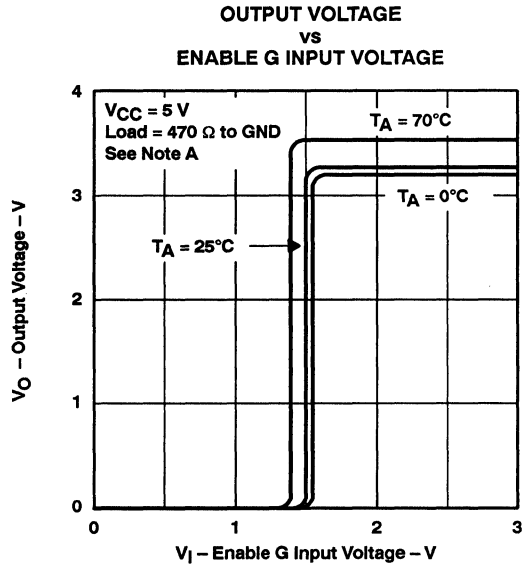


Figure 3

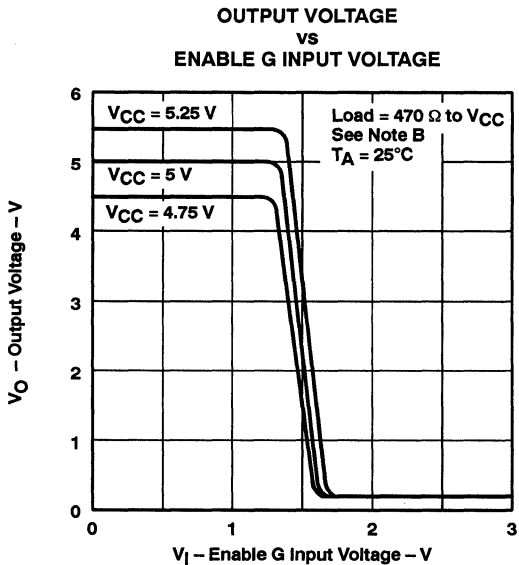


Figure 4

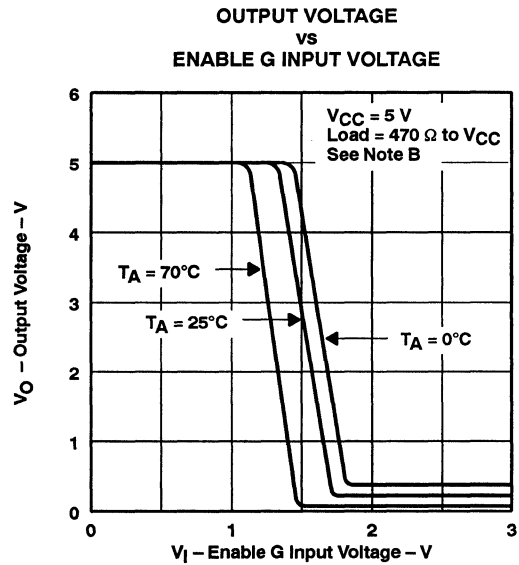


Figure 5

NOTES: A. The A input is connected to  $V_{CC}$  during the testing of the Y outputs and to ground during testing of the Z outputs.  
 B. The A input is connected to ground during the testing of the Y outputs and to  $V_{CC}$  during the testing of the Z outputs.

# AM26LS31C QUAD DIFFERENTIAL LINE DRIVER

SLLS114 - D2433, JANUARY 1979 - REVISED MAY 1990

## TYPICAL CHARACTERISTICS

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

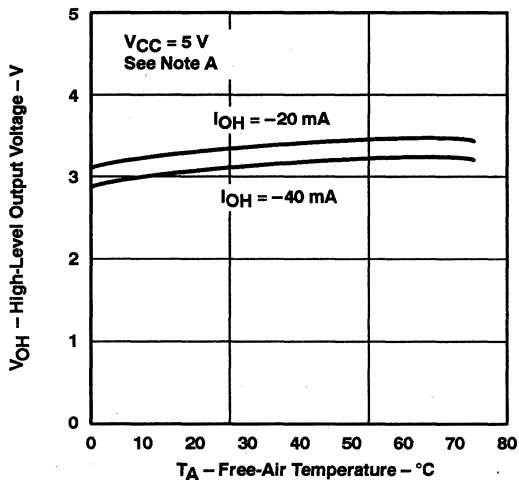


Figure 6

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

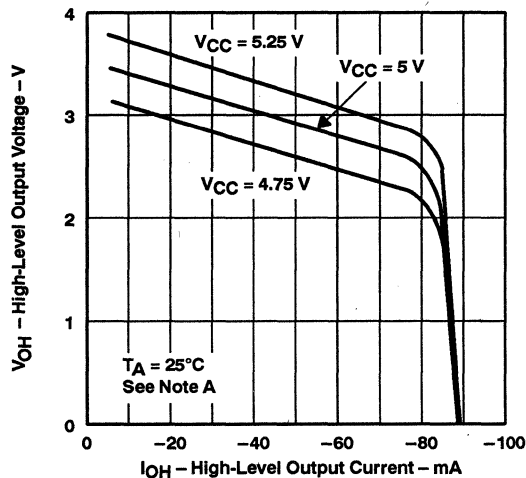


Figure 7

LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

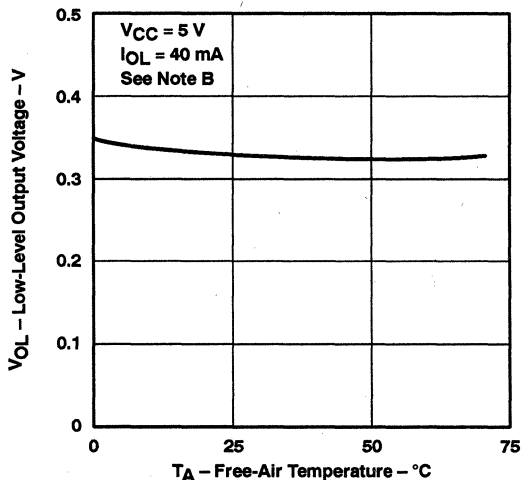


Figure 8

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

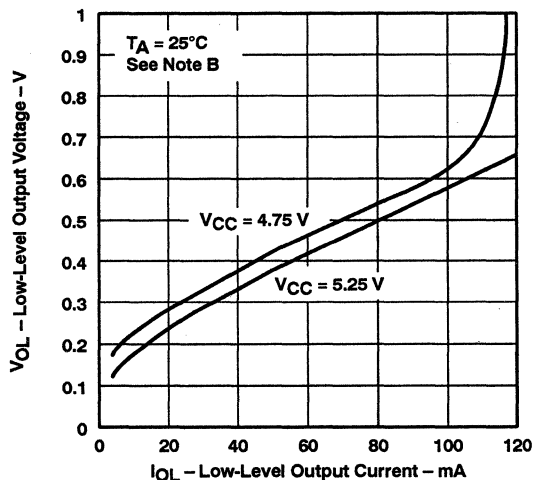


Figure 9

NOTES: A. The A input is connected to V<sub>CC</sub> during the testing of the Y outputs and to ground during testing of the Z outputs.  
B. The A input is connected to ground during the testing of the Y outputs and to V<sub>CC</sub> during the testing of the Z inputs.

TEXAS  
INSTRUMENTS

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TYPICAL CHARACTERISTICS

Y OUTPUT VOLTAGE  
 vs  
 DATA INPUT VOLTAGE

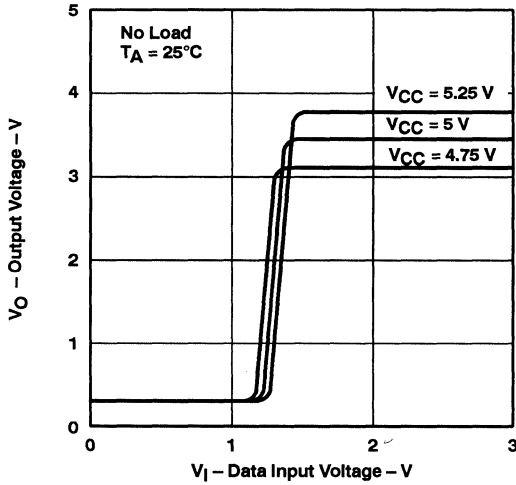


Figure 10

Y OUTPUT VOLTAGE  
 vs  
 DATA INPUT VOLTAGE

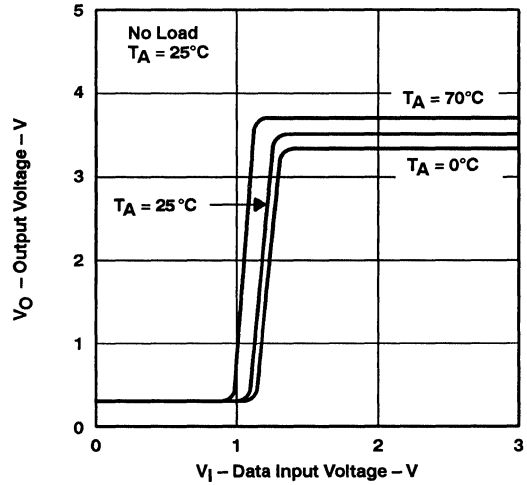


Figure 11



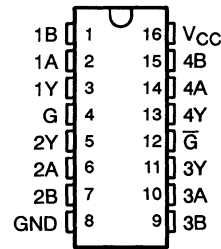


# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

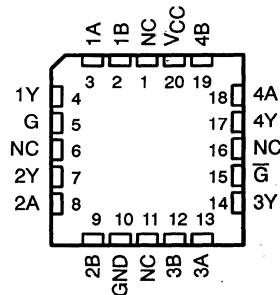
SLLS115A – D2434, OCTOBER 1980 – REVISED MARCH 1993

- **AM26LS32A Meets EIA Standards RS-422-A and RS-423-A**
- **AM26LS32A Has  $\pm 7$ -V Common-Mode Range With  $\pm 200$ -mV Sensitivity**
- **AM26LS32A Has  $\pm 15$ -V Common-Mode Range With  $\pm 500$ -mV Sensitivity**
- **Input Hysteresis . . . 50 mV Typical**
- **Operates From a Single 5-V Supply**
- **Low-Power Schottky Circuitry**
- **3-State Outputs**
- **Complementary Output Enable Inputs**
- **Input Impedance . . . 12 k $\Omega$  Min**
- **Designed to Be Interchangeable With Advanced Micro Devices AM26LS32 and AM26LS33**

AM26LS32AC, AM26LS33AC . . . D OR N PACKAGE  
AM26LS32AM, AM26LS33AM . . . J PACKAGE  
(TOP VIEW)



AM26LS32AM, AM26LS33AM . . . FK PACKAGE  
(TOP VIEW)



## description

The AM26LS32A and AM26LS33A are quad line receivers for balanced and unbalanced digital data transmission. The enable function is common to all four receivers and offers a choice of active-high or active-low input. The 3-state outputs permit connection direct to a bus-organized system. Fail-safe design ensures that if the inputs are open, the outputs will always be high.

Compared to the AM26LS32 and the AM26LS33, the AM26LS32A and AM26LS33A incorporate an additional stage of amplification to improve sensitivity. The input impedance has been increased resulting in less loading of the bus line. The additional stage has increased propagation delay; however, this will not affect interchangeability in most applications.

The AM26LS32AC and AM26LS33AC are characterized for operation from 0°C to 70°C. The AM26LS32AM and AM26LS33AM are characterized for operation over the full military temperature range of -55°C to 125°C.

NC—No internal connection

FUNCTION TABLE  
(each receiver)

DIFFERENTIAL A - B	ENABLES		OUTPUT Y
	G	G-bar	
$V_{ID} \geq V_{TH}$	H	X	H
	X	L	H
$V_{TL} \leq V_{ID} \leq V_{TH}$	H	X	?
	X	L	?
$V_{ID} \leq V_{TL}$	H	X	L
	X	L	L
X	L	H	Z
Open	H	X	H
	X	L	H

H = high level, L = low level, ? = indeterminate;  
X = irrelevant, Z = high impedance (off)

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

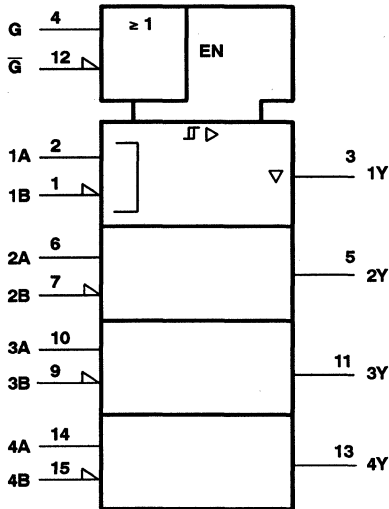
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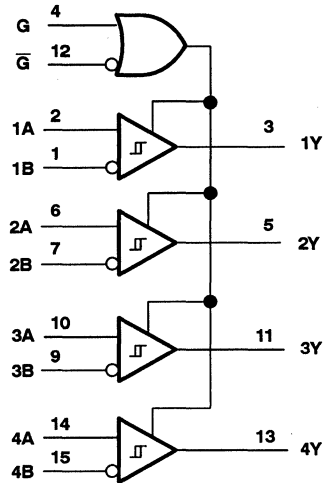
# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A - D2434, OCTOBER 1980 - REVISED MARCH 1993

## logic symbol†



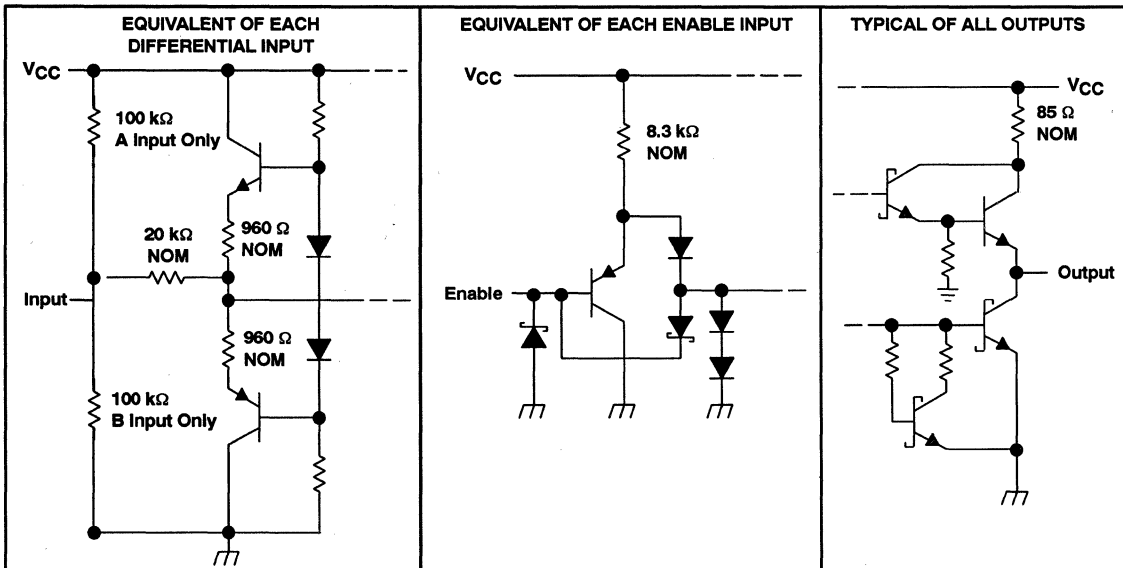
## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for D, J, and N packages.

## schematics of inputs and outputs



# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A - D2434, OCTOBER 1980 - REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

		AM26LS32AC AM26LS33AC	AM26LS32AM AM26LS33AM	UNIT
Supply voltage, $V_{CC}$ (see Note 1)		7	7	V
Input voltage, any differential input		$\pm 25$	$\pm 25$	V
Differential input voltage (see Note 2)		$\pm 25$	$\pm 25$	V
Continuous total power dissipation		See Dissipation Rating Table		
Operating free-air temperature range		0 to 70	-55 to 125	$^{\circ}\text{C}$
Storage temperature range		-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package	260		$^{\circ}\text{C}$
Case temperature for 60 seconds	FK package		260	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J package	300	300	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.  
 2. Differential voltage values are at the noninverting (A) input terminals with respect to the inverting (B) input terminals.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW	—
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
J	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW	—

## recommended operating conditions

		AM26LS32AC AM26LS33AC			AM26LS32AM AM26LS33AM			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$		4.75	5	5.25	4.5	5	5.5	V
High-level input voltage, $V_{IH}$		2			2			V
Low-level input voltage, $V_{IL}$					0.8			V
Common-mode input voltage, $V_{IC}$	AM26LS32AC, AM26LS32AM				$\pm 7$			V
	AM26LS33AC, AM26LS33AM				$\pm 15$			
High-level output current, $I_{OH}$					-440			$\mu\text{A}$
Low-level output current, $I_{OL}$					8			mA
Operating free-air temperature, $T_A$		0		70	-55		125	$^{\circ}\text{C}$

# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A – D2434, OCTOBER 1980 – REVISED MARCH 1993

electrical characteristics over recommended ranges of  $V_{CC}$ ,  $V_{IC}$ , and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = V_{OHmin}$ , $I_{OH} = -440 \mu A$	AM26LS32A			0.2	mV
			AM26LS33A			0.5	
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.45 V$ , $I_{OL} = 8 mA$	AM26LS32A	-0.2‡			V
			AM26LS33A	-0.5‡			
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )				50		mV
$V_{IK}$	Enable input clamp voltage	$V_{CC} = MIN$ , $I_I = -18 mA$				-1.5	V
$V_{OH}$	High-level output voltage	$V_{CC} = MIN$ , $V_{I(\bar{G})} = 0.8 V$ , $V_{ID} = 1 V$ , $I_{OH} = -440 \mu A$	'32AC, '33AC	2.7			V
			'32AM, '33AM	2.5			
$V_{OL}$	Low-level output voltage	$V_{CC} = MIN$ , $V_{I(\bar{G})} = 0.8 V$ , $V_{ID} = -1 V$	$I_{OL} = 4 mA$			0.4	V
			$I_{OL} = 8 mA$			0.45	
$I_{OZ}$	Off-state (high-impedance-state) output current	$V_{CC} = MAX$	$V_O = 2.4 V$			20	$\mu A$
			$V_O = 0.4 V$			-20	
$I_I$	Line input current	$V_I = 15 V$ , Other input at $-10 V$ to $15 V$				1.2	mA
		$V_I = -15 V$ , Other input at $-15 V$ to $10 V$				-1.7	
$I_{I(EN)}$	Enable input current	$V_I = 5.5 V$				100	$\mu A$
$I_{IH}$	High-level enable current	$V_I = 2.7 V$				20	$\mu A$
$I_{IL}$	Low-level enable current	$V_I = 0.4 V$				-0.36	mA
$r_i$	Input resistance	$V_{IC} = -15 V$ to $15 V$ , One input to ac ground			12	15	k $\Omega$
$I_{OS}$	Short-circuit output current§	$V_{CC} = MAX$			-15	-85	mA
$I_{CC}$	Supply current	$V_{CC} = MAX$ , All outputs disabled			52	70	mA

† All typical values are at  $V_{CC} = 5 V$ ,  $T_A = 25^\circ C$ , and  $V_{IC} = 0$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold levels only.

§ Not more than one output should be shorted to ground at a time, and duration of the short circuit should not exceed one second.

## switching characteristics $V_{CC} = 5 V$ , $T_A = 25^\circ C$

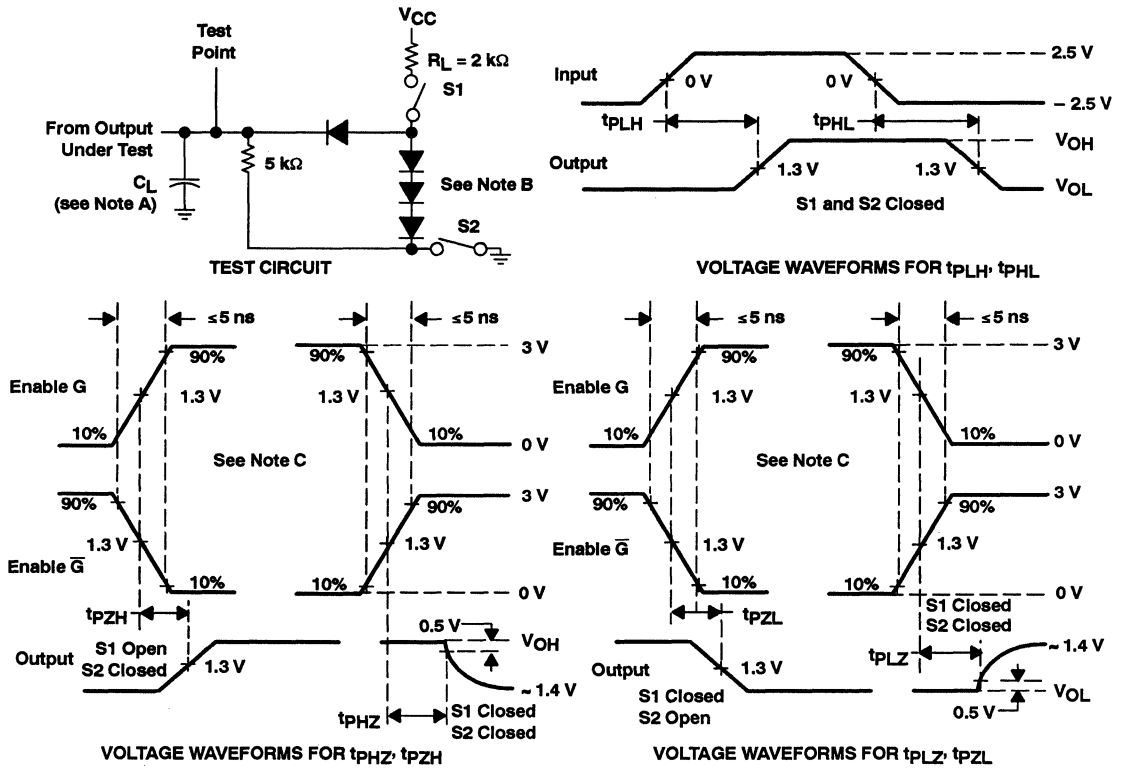
PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$C_L = 15 pF$ , See Figure 1		20	35		ns
$t_{PHL}$	Propagation delay time, high-to-low-level output				22	35	ns
$t_{PZH}$	Output enable time to high level	$C_L = 15 pF$ , See Figure 1		17	22		ns
$t_{PZL}$	Output enable time to low level				20	25	ns
$t_{PHZ}$	Output disable time from high level	$C_L = 5 pF$ , See Figure 1		21	30		ns
$t_{PLZ}$	Output disable time from low level				30	40	ns



# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A – D2434, OCTOBER 1980 – REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A.  $C_L$  includes probe and jig capacitance.  
 B. All diodes are 1N3064 or equivalent.  
 C. Enable  $G$  is tested with  $\bar{G}$  high;  $\bar{G}$  is tested with  $G$  low.

Figure 1

# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A - D2434, OCTOBER 1980 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

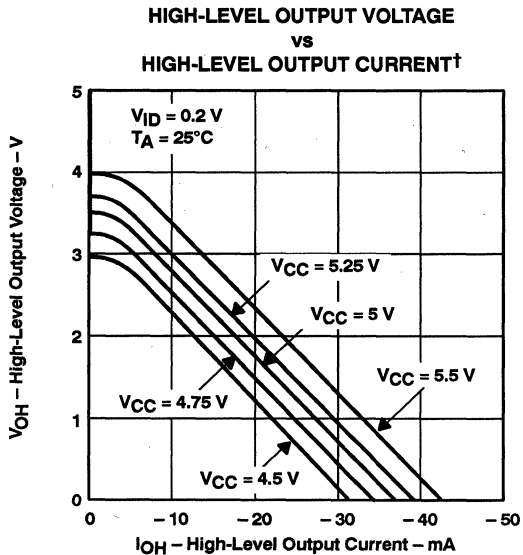


Figure 2

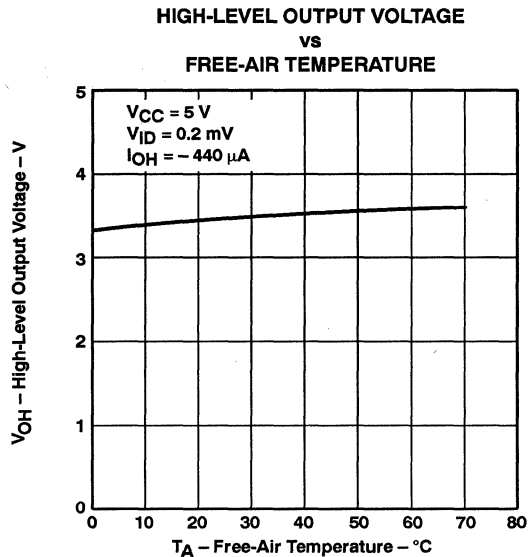


Figure 3

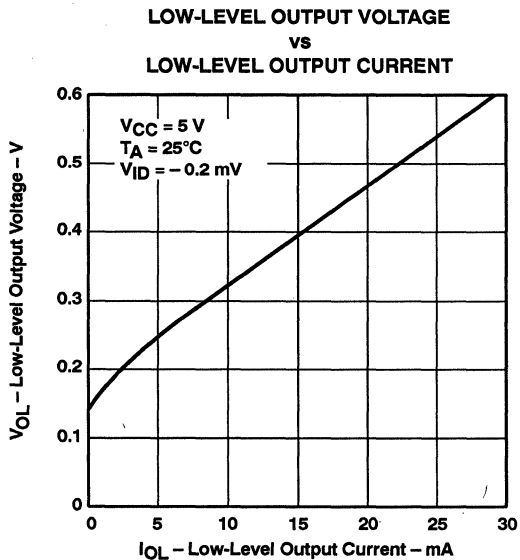


Figure 4

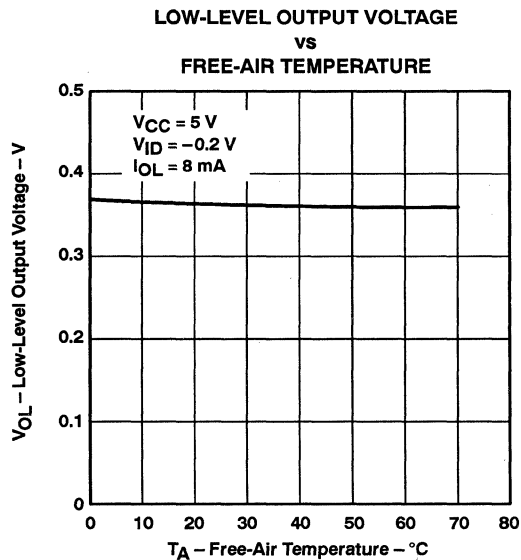


Figure 5

†  $V_{CC} = 5.5\text{ V}$  and  $V_{CC} = 4.5\text{ V}$  applies to M suffix devices only.

# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A - D2434, OCTOBER 1980 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

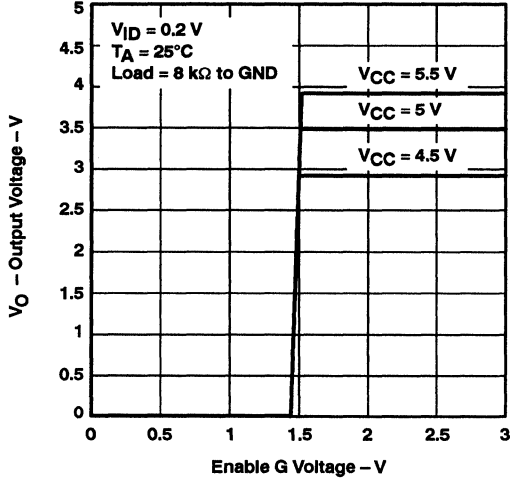


Figure 6

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

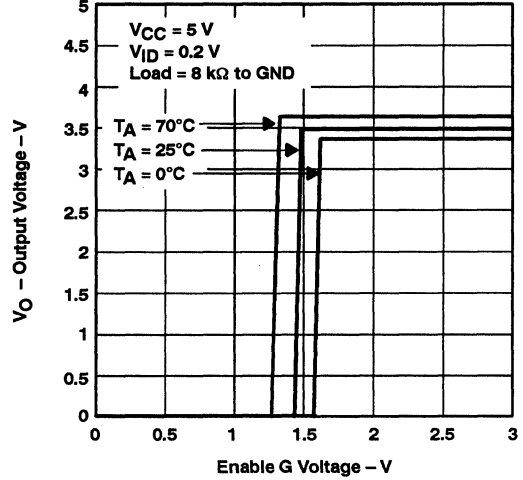


Figure 7

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

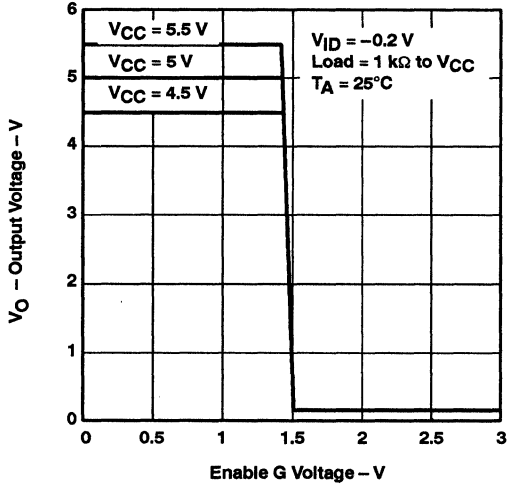


Figure 8

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

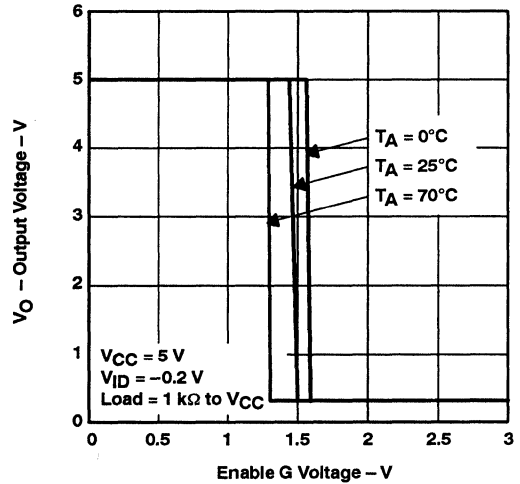


Figure 9





# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A - D2434, OCTOBER 1980 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

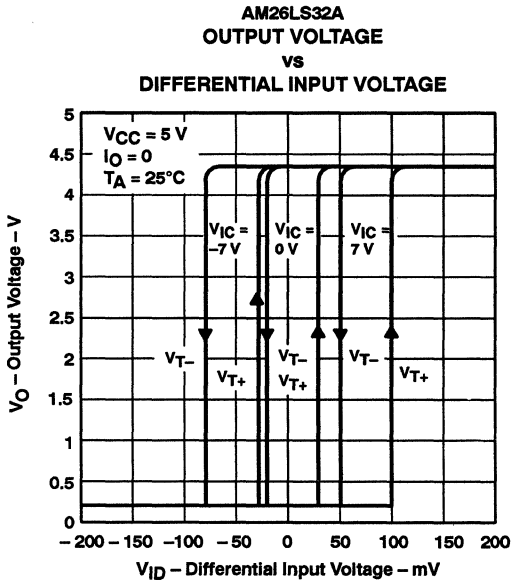


Figure 10

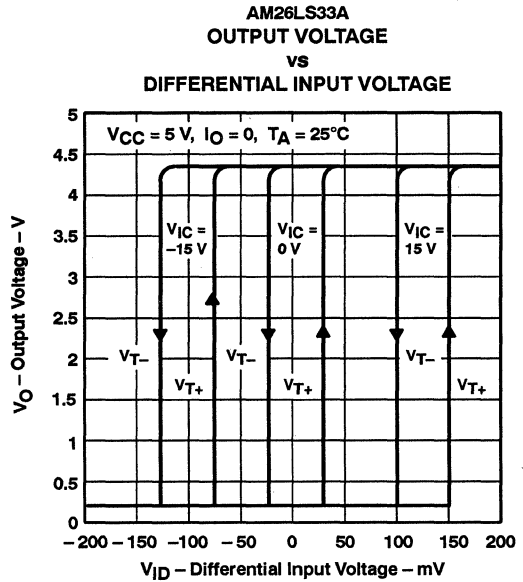


Figure 11

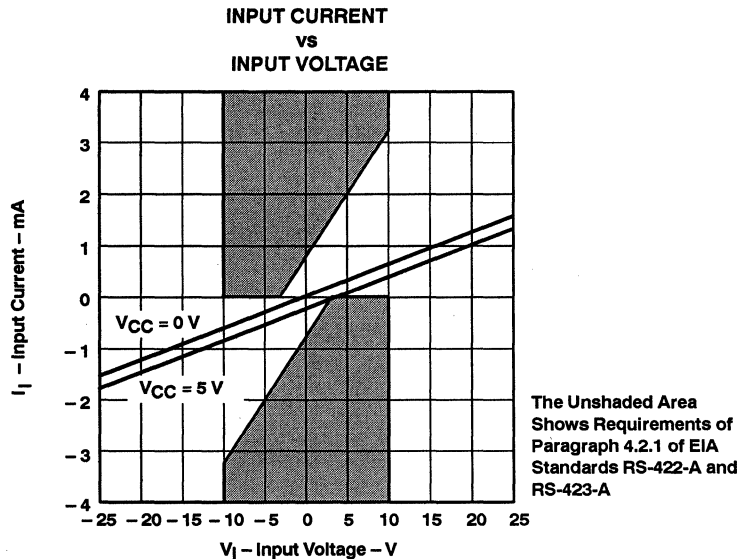
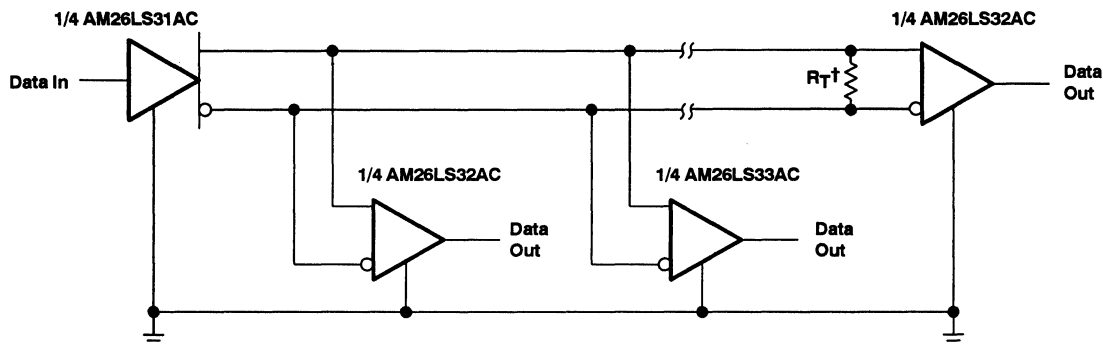


Figure 12

# AM26LS32AC, AM26LS33AC, AM26LS32AM, AM26LS33AM QUAD DIFFERENTIAL LINE RECEIVERS

SLLS115A - D2434, OCTOBER 1980 - REVISED MARCH 1993

## APPLICATION INFORMATION



†  $R_T$  equals the characteristic impedance of the line.

Figure 13. Circuit With Multiple Receivers

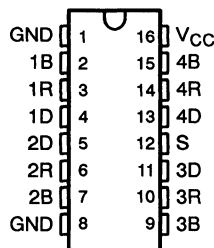


# AM26S10C, AM26S11C QUADRUPLE BUS TRANSCEIVERS

SLLS116A - D2298, JANUARY 1977 - REVISED JANUARY 1993

- Schottky Circuitry for High Speed, Typical Propagation Delay Time . . . 12 ns
- Drivers Feature Open-Collector Outputs for Party-Line (Data Bus) Operation
- Driver Outputs Can Sink 100 mA at 0.8 V Maximum
- PNP Inputs for Minimal Input Loading
- Designed to Be Interchangeable With Advanced Micro Devices AM26S10 and AM26S11

D OR N PACKAGE  
(TOP VIEW)



## description

The AM26S10C and AM26S11C are quadruple bus transceivers utilizing Schottky-diode-clamped transistors for high speed. The drivers feature open-collector outputs capable of sinking 100 mA at 0.8 V maximum. The driver and strobe inputs use pnp transistors to reduce the input loading.

The driver of the AM26S10C is inverting; the driver of the AM26S11C is noninverting. Each device has two ground connections for improved ground current-handling capability. For proper operation, the ground pins should be tied together.

The AM26S10C and AM26S11C are characterized for operation over the temperature range of 0°C to 70°C.

## Function Tables

AM26S10C  
(transmitting)

INPUTS		OUTPUTS	
S	D	B	R
L	H	L	H
L	L	H	L

AM26S11C  
(transmitting)

INPUTS		OUTPUTS	
S	D	B	R
L	H	H	L
L	L	L	H

AM26S10C AND AM26S11C  
(receiving)

INPUTS			OUTPUT
S	B	D	R
H	H	X	L
H	L	X	H

H = high level, L = low level, X = irrelevant

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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**INSTRUMENTS**

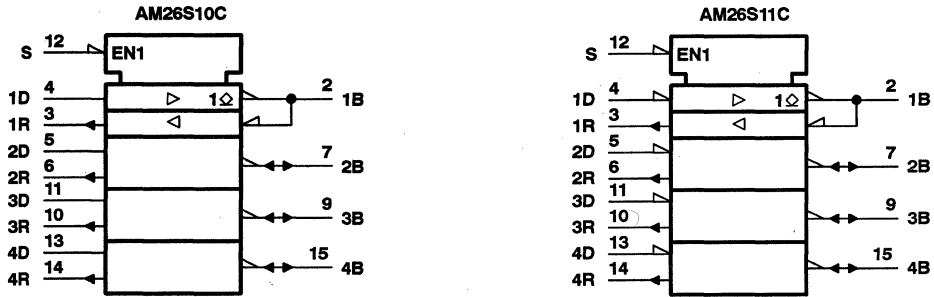
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# AM26S10C, AM26S11C QUADRUPLE BUS TRANSCEIVERS

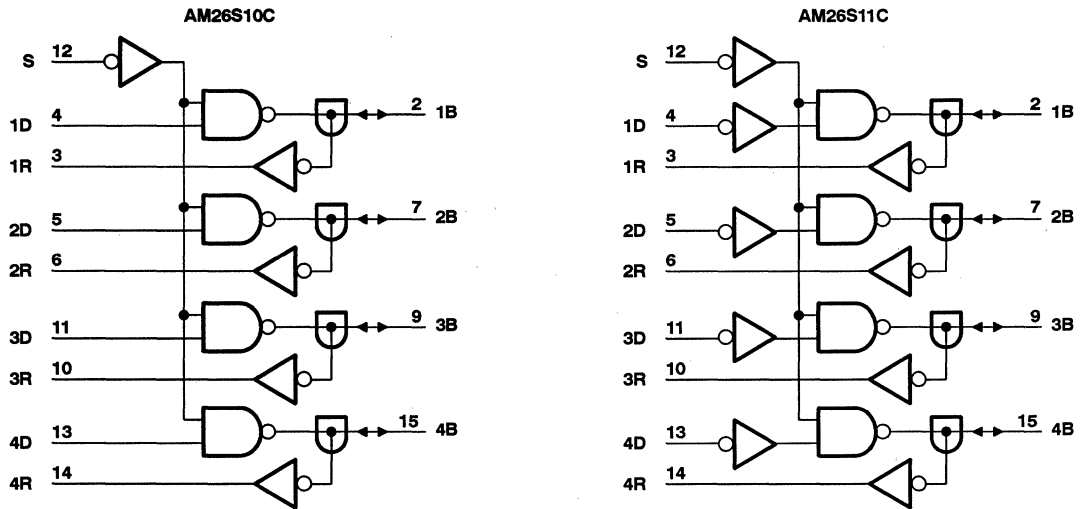
SLLS116A - D2298, JANUARY 1977 - REVISED JANUARY 1993

## logic symbols†



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

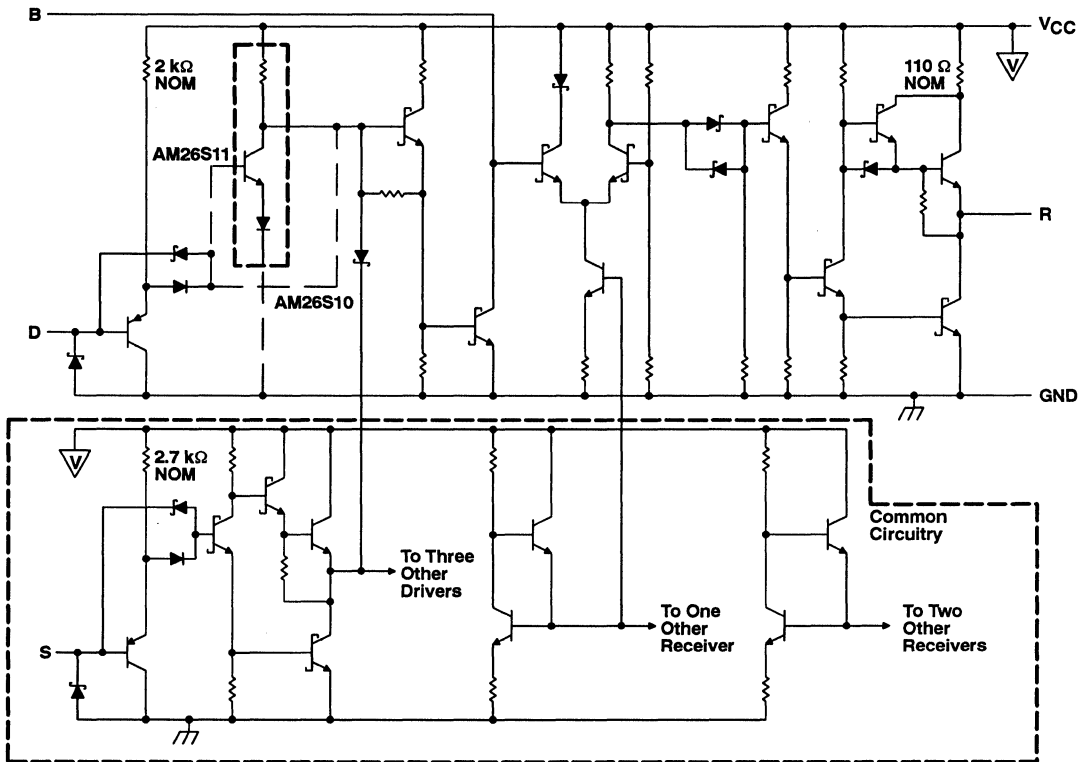
## logic diagrams (positive logic)



# AM26S10C, AM26S11C QUADRUPLE BUS TRANSCEIVERS

SLLS116A - D2298, JANUARY 1977 - REVISED JANUARY 1993

## schematic (each transceiver)



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# AM26S10C, AM26S11C QUADRUPLE BUS TRANSCEIVERS

SLLS116A – D2298, JANUARY 1977 – REVISED JANUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	–0.5 V to 7 V
Driver or strobe input voltage range, $V_I$	–0.5 V to 5.5 V
Bus voltage range, driver output off, $V_O$	–0.5 V to 5.25 V
Driver or strobe input current range, $I_I$	–30 mA to 5 mA
Driver output current, $I_O$	200 mA
Receiver output current, $I_O$	30 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminals connected together.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	D or S	2			V
	B	2.25			
Low-level input voltage, $V_{IL}$	D or S	0.8			V
	B	1.75			
Receiver high-level output current, $I_{OH}$		–1			mA
Low-level output current, $I_{OL}$	Driver	100			mA
	Receiver	20			
Operating free-air temperature, $T_A$		0	70		°C

# AM26S10C, AM26S11C QUADRUPLE BUS TRANSCEIVERS

SLLS116A – D2298, JANUARY 1977 – REVISED JANUARY 1993

## electrical characteristics over recommended operating free-air temperature range

PARAMETER			TEST CONDITIONS		MIN	TYP†	MAX	UNIT	
V <sub>IK</sub>	Input clamp voltage	D or S	V <sub>CC</sub> = 4.75 V, I <sub>I</sub> = -18 mA				-1.2	V	
V <sub>OH</sub>	High-level output voltage	R	V <sub>CC</sub> = 4.75 V, I <sub>OH</sub> = -1 mA	V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V	2.7	3.4		V	
V <sub>OH</sub>	Low-level output voltage	R	V <sub>CC</sub> = 4.75 V, V <sub>IL</sub> = 0.8 V	V <sub>IH</sub> = 2 V,			I <sub>OL</sub> = 20 mA	0.5	V
		B			I <sub>OL</sub> = 40 mA	0.33	0.5		
					I <sub>OL</sub> = 70 mA	0.42	0.7		
					I <sub>OL</sub> = 100 mA	0.51	0.8		
I <sub>O(off)</sub>	Off-stage output current	B	V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V	V <sub>CC</sub> = 5.25 V, V <sub>O</sub> = 0.8 V			-50	μA	
		V <sub>CC</sub> = 5.25 V, V <sub>O</sub> = 4.5 V			100				
		V <sub>CC</sub> = 0, V <sub>O</sub> = 4.5 V			100				
I <sub>IH</sub>	High-level input current	D S	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 2.7 V				30 20	μA	
I <sub>I</sub>	Input current at maximum input voltage	D or S	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 5.5 V				100	μA	
I <sub>IL</sub>	Low-level input current	D	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 0.4 V				-0.54	mA	
		S					-0.36		
I <sub>OS</sub>	Short-circuit output current‡	R	V <sub>CC</sub> = 5.25 V		-18		-60	mA	
I <sub>CC</sub>	Supply current		V <sub>CC</sub> = 5.25 V, Strobe at 0 V, No load, All driver outputs low			45	70 80	mA	

† All typical values are at T<sub>A</sub> = 25°C and V<sub>CC</sub> = 5 V.

‡ Not more than one output should be shorted to ground at a time, and duration of the short circuit should not exceed one second.

## switching characteristics, V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	AM26S10C			AM26S11C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PLH</sub>	D	B	See Figure 1	10		15	12		19	ns
t <sub>PHL</sub>				10		15	12		19	
t <sub>PLH</sub>	S	B		14		18	15		20	ns
t <sub>PHL</sub>				13		18	14		20	
t <sub>PLH</sub>	B	R		10		15	10		15	ns
t <sub>PHL</sub>				10		15	10		15	
t <sub>TLH</sub>		B		4		10	4		10	ns
t <sub>THL</sub>				2		4	2		4	

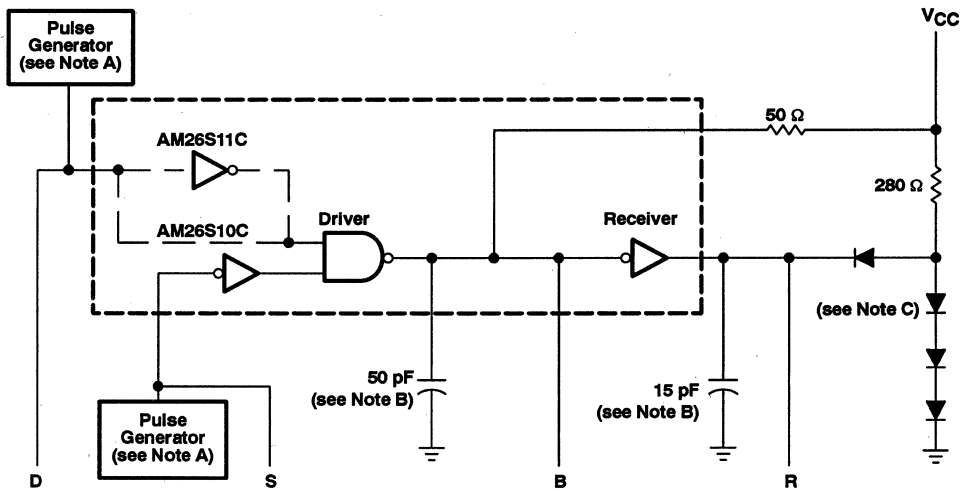




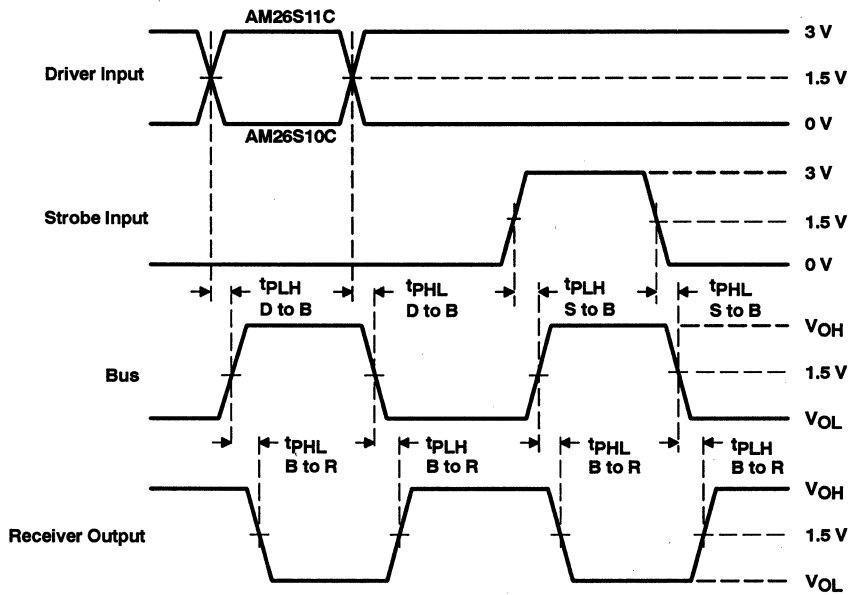
# AM26S10C, AM26S11C QUADRUPLE BUS TRANSCEIVERS

SLLS116A - D2298, JANUARY 1977 - REVISED JANUARY 1993

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

- NOTES: A. The pulse generators have the following characteristics:  $Z_0 = 50 \Omega$ ,  $t_r = 10 \pm 5$  ns.  
 B. Includes probe and jig capacitance.  
 C. All diodes are 1N916 or equivalent.

Figure 1. Test Circuit and Voltage Waveforms

# AM26S10C, AM26S11C QUADRUPLE BUS TRANSCEIVERS

SLLS116A - D2298, JANUARY 1977 - REVISED JANUARY 1993

## APPLICATION INFORMATION

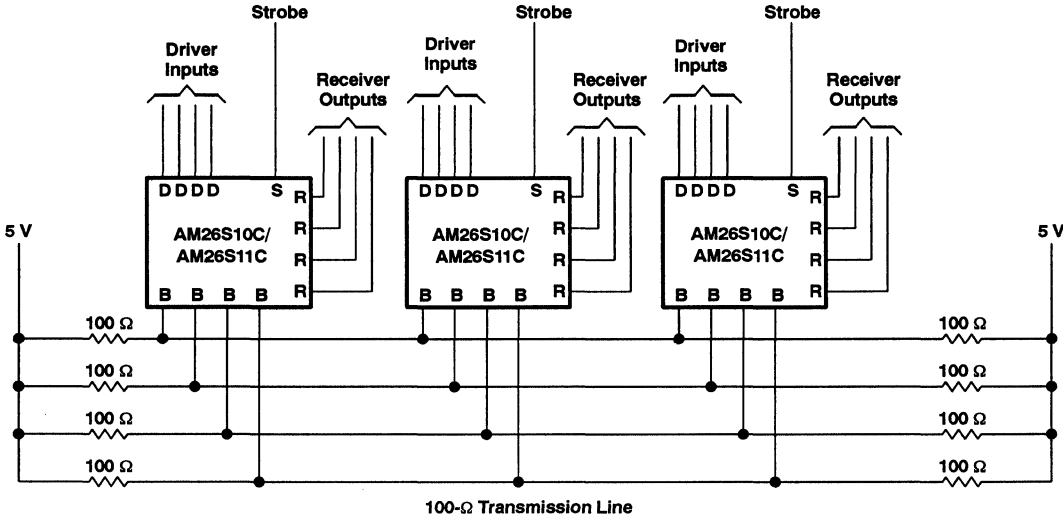


Figure 2. Party-Line System





# DP8480

## 10K ECL-TO-TTL LEVEL TRANSLATOR WITH LATCH

SLLS035B – D3058, NOVEMBER 1987 – REVISED FEBRUARY 1993

- ECL Control Inputs
- 3-State Outputs
- 10K ECL Input Compatible
- Direct Replacement for National Semiconductor DP8480

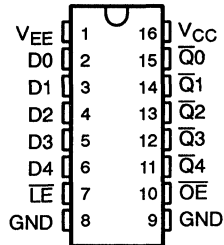
### description

This circuit translates ECL-input levels to TTL-output levels and provides an inverting transparent latch. The 3-state outputs are designed to drive highly capacitive loads. All inputs operate at ECL levels.

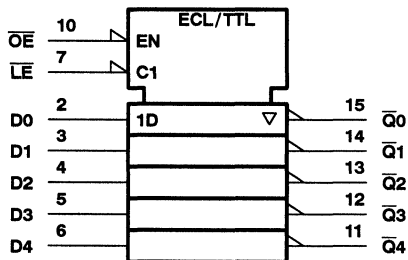
If latch enable ( $\overline{LE}$ ) is low, the latches are transparent and the  $\overline{Q}$  outputs follow the complement of the D inputs. If  $\overline{LE}$  is high, the outputs are latched. If output enable ( $\overline{OE}$ ) is high, the outputs are in the high-impedance state, as they are during power up and power down.

The DP8480 is characterized for operation from 0°C to 75°C.

N PACKAGE  
(TOP VIEW)



### logic symbol

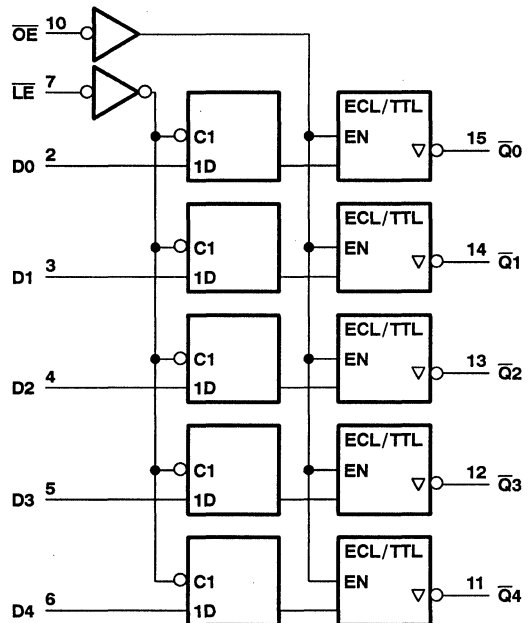


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

FUNCTION TABLE  
(each latch/translator)

$\overline{OE}$	$\overline{LE}$	D	$\overline{Q}$
H	X	X	Z
L	L	L	H
L	L	H	L
L	H	X	$Q_0$

### logic diagram



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

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# DP8480

## 10K ECL-TO-TTL LEVEL TRANSLATOR WITH LATCH

SLLS035B – D3058, NOVEMBER 1987 – REVISED FEBRUARY 1993

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$	7 V
Supply voltage, $V_{EE}$	-8 V
Input voltage range, $V_I$	0 V to $V_{EE}$
Output voltage, $V_O$	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 75°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 75^\circ\text{C}$ POWER RATING
N	1150 mW	9.2 mW/°C	690 mW

### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.5	V
Supply voltage, $V_{EE}$	-4.68	-5.20	-5.72	V
High-level Input voltage, $V_{IH}$ (see Note 1)	$T_A = 0^\circ\text{C}$		-840	mV
	$T_A = 25^\circ\text{C}$		-810	
	$T_A = 75^\circ\text{C}$		-720	
Low-level Input voltage, $V_{IL}$ (see Note 1)	$T_A = 0^\circ\text{C}$		-1490	mV
	$T_A = 25^\circ\text{C}$		-1475	
	$T_A = 75^\circ\text{C}$		-1450	
Pulse duration, $\overline{LE}$ low, $t_w$ (see Figure 1)		5		ns
Setup time, data before $\overline{LE}\uparrow$ , $t_{SU}$ (see Figure 1)		3		ns
Hold time, data after $\overline{LE}\uparrow$ , $t_H$ (see Figure 1)		3		ns
Operating free-air temperature, $T_A$		0	75	°C

NOTE 1: The algebraic convention, in which the least positive (most negative) value is designated minimum, is used in this data sheet for logic levels only.

### electrical characteristics over recommended ranges of supply voltages and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}$ High-level output voltage	$I_{OH} = -10$ mA	$V_{CC} - 2$			V
$V_{OL}$ Low-level output voltage	$I_{OL} = 12$ mA		0.2	0.5	V
$I_{IH}$ High-level input current	$V_{IH} = V_{IH}$ max		75	350	μA
$I_{IL}$ Low-level input current	$V_{IL} = V_{IL}$ min		50	85	μA
$I_{OSH}$ High-state short-circuit output current	$V_{OSH} = 0$ , See Note 2	-70	-150		mA
$I_{OSL}$ High-state short-circuit output current	$V_{OSL} = 2.5$ V, See Note 2	70	150		mA
$I_{OZ}$ High-impedance state output current	$V_O = 0$ to 5 V		±1	±50	μA
$I_{CC}$ Supply current from $V_{CC}$	Outputs open, Inputs = $V_{IL}$		16	35	mA
$I_{EE}$ Supply current from $V_{EE}$	Outputs open, Inputs = $V_{IL}$		-30	-50	mA

† Typical values are at  $V_{CC} = 5$  V,  $V_{EE} = -5.2$  V,  $T_A = 25^\circ\text{C}$ .

NOTE 2: During testing of  $I_{OSH}$  or  $I_{OSL}$ , only one output should be tested at a time and the current should be limited to a maximum of ±120 mA.



# DP8480 10K ECL-TO-TTL LEVEL TRANSLATOR WITH LATCH

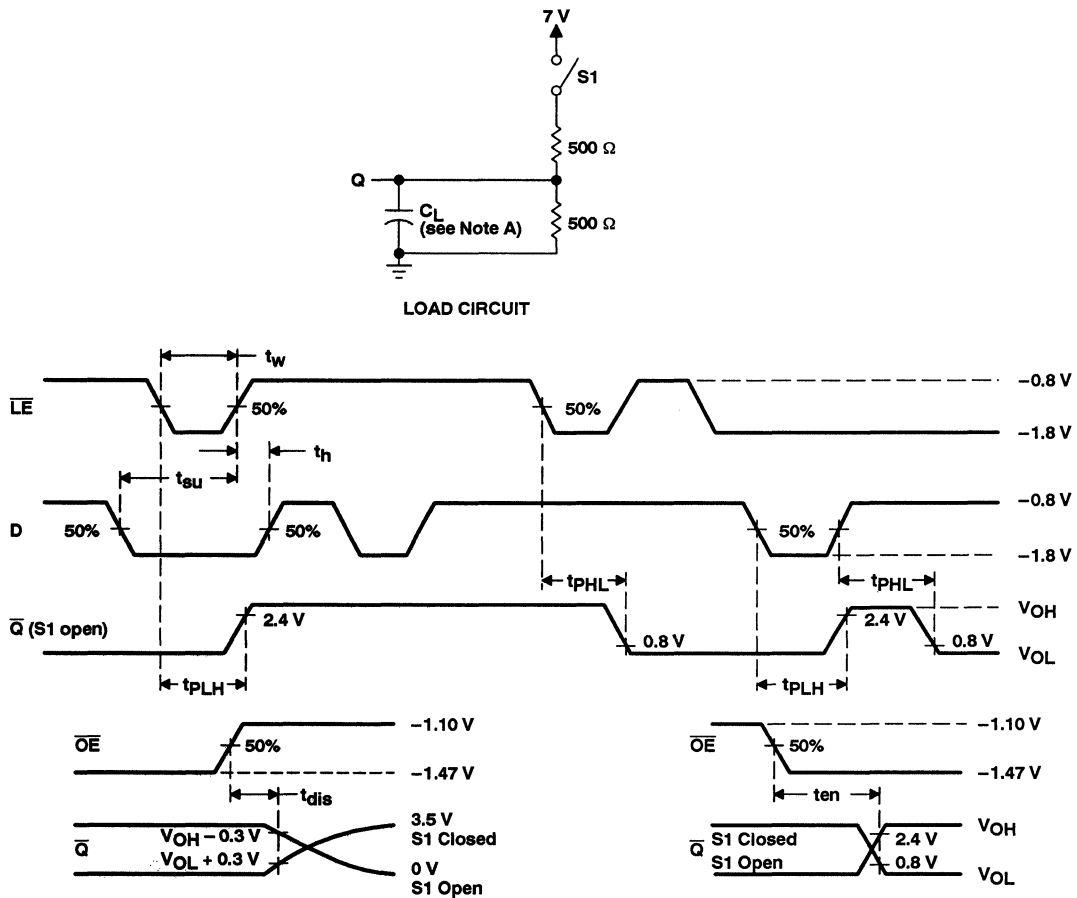
SLLS035B – D3058, NOVEMBER 1987 – REVISED FEBRUARY 1993

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output from $\overline{LE}$ input	$C_L = 50$ pF, See Figure 1	4	10	15	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output from $\overline{LE}$ input		4	11	15	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output from D input		3.5	10	15	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output from D input		3.5	11	15	ns
$t_{en}$ Output enable time from $\overline{OE}$ input		6	12	25	ns
$t_{dis}$ Output disable time from $\overline{OE}$ input		4.5	8	22	ns

† Typical values are at  $V_{CC} = 5$  V,  $V_{EE} = -5.2$  V,  $T_A = 25^\circ\text{C}$ , and with all channels switched simultaneously.

## PARAMETER MEASUREMENT INFORMATION



NOTES: A.  $C_L$  includes probe and jig capacitance.  
B. ECL input rise times and fall time are  $2\text{ ns} \pm 0.2\text{ ns}$  from 20% to 80%.

**Figure 1. Load Circuit and Voltage Waveforms**



# DP8481 10K TTL-TO-ECL LEVEL TRANSLATOR WITH LATCH

SLLS036B – D3059, NOVEMBER 1987 – REVISED FEBRUARY 1993

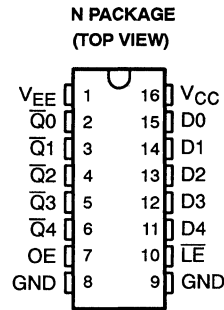
- ECL Control Inputs
- 10K ECL Compatible
- Propagation Delay . . . 4 ns Typ
- Direct Replacement for National Semiconductor DP8481

## description

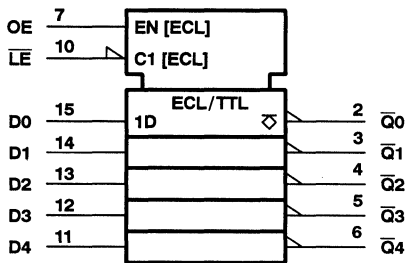
This circuit translates TTL-input levels to ECL-output levels and provides a 5-bit transparent latch. The outputs are gated by output enable (OE) and can be wire-OR connected. The latch enable ( $\overline{LE}$ ) and OE inputs are ECL.

If latch enable ( $\overline{LE}$ ) is low, the latches are transparent and the  $\overline{Q}$  outputs follow the complement of the D inputs. If  $\overline{LE}$  is high, the outputs are latched. If output enable (OE) is low, the outputs are forced to the low level.

The DP8481 is characterized for operation from 0°C to 75°C.



## logic symbol†

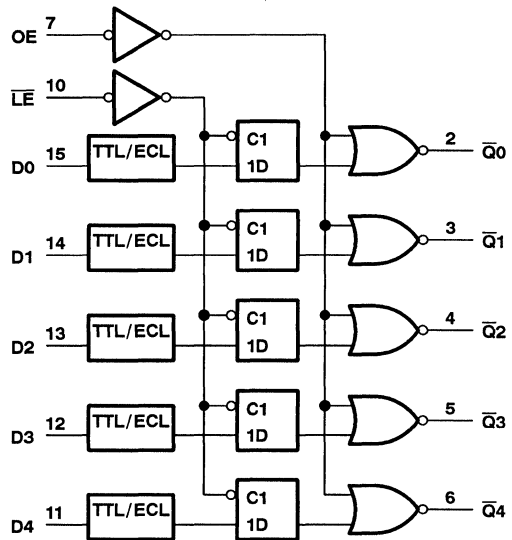


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

**FUNCTION TABLE**  
(each latch/translator)

OE	$\overline{LE}$	D	$\overline{Q}$
H	L	H	L
H	L	L	H
H	H	X	$Q_0$
L	X	X	L

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.





# DP8481 10K TTL-TO-ECL LEVEL TRANSLATOR WITH LATCH

SLLS036B – D3059, NOVEMBER 1987 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ .....	7 V
Supply voltage, $V_{EE}$ .....	-8 V
Input voltage range, $V_I$ : OE or $\overline{LE}$ input .....	0 V to $V_{EE}$
D inputs .....	-1 V to 5.5 V
Output current, $I_O$ .....	-50 mA
Continuous total dissipation .....	See Dissipation Rating Table
Operating temperature range, $T_A$ .....	0°C to 75°C
Storage temperature range .....	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 75^\circ\text{C}$ POWER RATING
N	1150 mW	9.2 mW/°C	690 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT	
Supply voltage, $V_{CC}$	4.5	5	5.5	V	
Supply voltage, $V_{EE}$	-4.68	-5.20	-5.72	V	
High-level input voltage, $V_{IH}$ (TTL-level D inputs)	2			V	
Low-level input voltage, $V_{IL}$ (TTL-level D inputs)			0.8	V	
High-level input voltage, $V_{IH}$ (ECL-level OE and $\overline{LE}$ inputs) (see Note 1)	$T_A = 0^\circ\text{C}$		-1145	-840	mV
	$T_A = 25^\circ\text{C}$		-1105	-810	
	$T_A = 75^\circ\text{C}$		-1045	-720	
Low-level input voltage, $V_{IL}$ (ECL-level OE and $\overline{LE}$ inputs) (see Note 1)	$T_A = 0^\circ\text{C}$		-1870	-1490	mV
	$T_A = 25^\circ\text{C}$		-1850	-1475	
	$T_A = 75^\circ\text{C}$		-1830	-1450	
Pulse duration, $\overline{LE}$ low, $t_w$ (see Figure 1)		5		ns	
Setup time, $t_{su}$		Data before $\overline{LE}\uparrow$ (see Figure 1)	5		ns
		Data before OE $\uparrow$ (see Note 2 and Figure 1)	5.5		
Hold time, data after $\overline{LE}\uparrow$ , $t_h$ (see Figure 1)		1		ns	
Operating free-air temperature, $T_A$	0		75	°C	

NOTES: 1. The algebraic convention, in which the least positive (most negative) value is designated minimum, is used in this data sheet for logic levels only.

2. This setup time applies when operating in the transparent mode ( $\overline{LE}$  is low) and it is necessary that valid data be available at the output immediately after the outputs are enabled.

# DP8481

## 10K TTL-TO-ECL LEVEL TRANSLATOR WITH LATCH

SLLS036B – D3059, NOVEMBER 1987 – REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of supply voltages and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	D0–D4 $I_I = -12 \text{ mA}$	-0.8	-1.2		V
$I_{IH}$	High-level input current	D0–D4 $V_I = 2.5 \text{ V}$		1	40	$\mu\text{A}$
		OE, $\overline{LE}$ $V_I = -0.8 \text{ V}$			200	
$I_{IL}$	Low-level input current	D0–D4 $V_I = 0.5 \text{ V}$		-50	-200	$\mu\text{A}$
		OE, $\overline{LE}$ $V_I = -1.8 \text{ V}$			150	
$V_{OH}$	High-level output voltage (see Notes 1 and 3)	$V_{EE} = -5.2 \text{ V}, T_A = 0^\circ\text{C}$	-1000		-840	mV
		$V_{EE} = -5.2 \text{ V}, T_A = 25^\circ\text{C}$	-960		-810	
		$V_{EE} = -5.2 \text{ V}, T_A = 75^\circ\text{C}$	-900		-720	
$V_{OH(C)}$	Critical high-level output voltage (see Notes 1 and 3)	$V_{EE} = -5.2 \text{ V}, T_A = 0^\circ\text{C}$		-1020		mV
		$V_{EE} = -5.2 \text{ V}, T_A = 25^\circ\text{C}$		-980		
		$V_{EE} = -5.2 \text{ V}, T_A = 75^\circ\text{C}$		-920		
$V_{OL}$	Low-level output voltage (see Notes 1 and 3)	$V_{EE} = -5.2 \text{ V}, T_A = 0^\circ\text{C}$	-1870		-1665	mV
		$V_{EE} = -5.2 \text{ V}, T_A = 25^\circ\text{C}$	-1850		-1650	
		$V_{EE} = -5.2 \text{ V}, T_A = 75^\circ\text{C}$	-1830		-1625	
$V_{OL(C)}$	Critical low-level output voltage (see Notes 1 and 3)	$V_{EE} = -5.2 \text{ V}, T_A = 0^\circ\text{C}$			-1645	mV
		$V_{EE} = -5.2 \text{ V}, T_A = 25^\circ\text{C}$			-1630	
		$V_{EE} = -5.2 \text{ V}, T_A = 75^\circ\text{C}$			-1605	
$I_{CC}$	Supply current from $V_{CC}$	$V_{CC} = 5.5 \text{ V}$			20	mA
$I_{EE}$	Supply current from $V_{EE}$	$V_{EE} = -5.7 \text{ V}$			-90	mA

NOTES: 1. The algebraic convention, in which the least positive (most negative) value is designated one minimum, is used in this data sheet for logic levels only.

3.  $V_{OH}$  and  $V_{OL}$  are tested using the outer-limit values  $V_{IH}$  max and  $V_{IL}$  min. The critical values  $V_{OH(C)}$  and  $V_{OL(C)}$  are tested using the inner-limit values  $V_{IH}$  min and  $V_{IL}$  max. The latter values ensure the noise margins of 155-mV high and 125-mV low associated with 10K ECL.

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

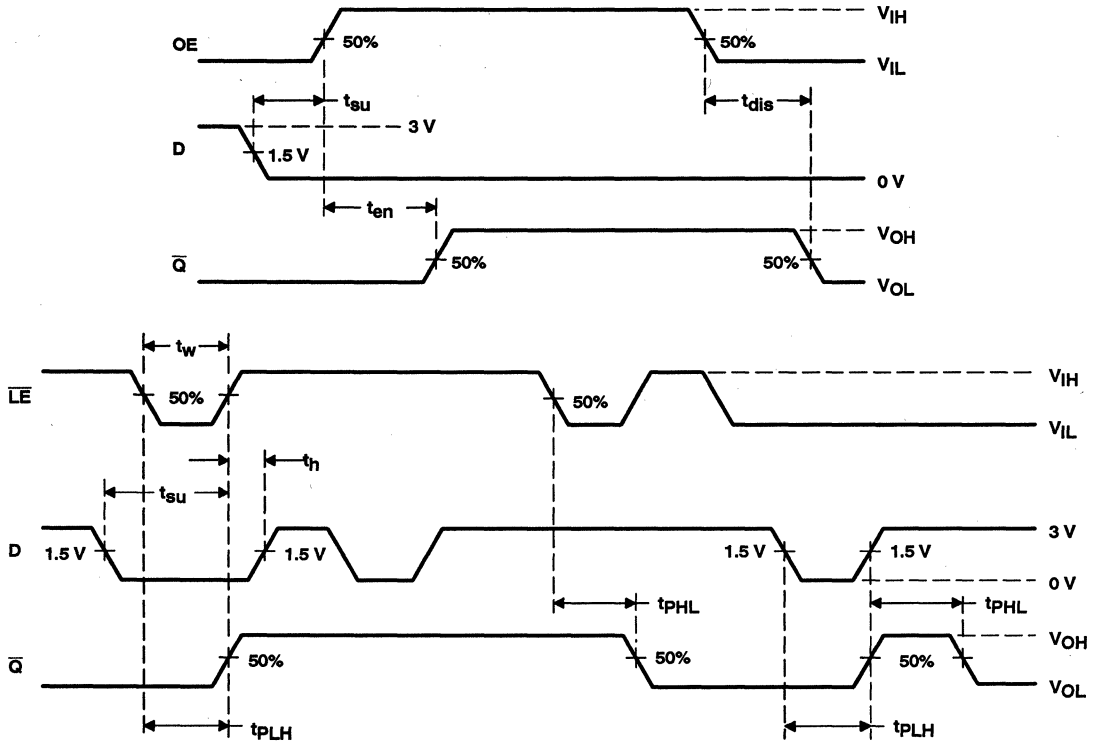
PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output from $\overline{LE}$ input	$R_L = 50 \Omega$ to $-2 \text{ V}$ , See Figure 1	1.5	4	6	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output from $\overline{LE}$ input		1.5	4	6	ns
$t_{PLH}$	Propagation delay time, low-to-high-level output from D input		2.5	4	7.5	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output from D input		2.5	4	7.5	ns
$t_{en}$	Output enable time from OE input		1	3	4	ns
$t_{dis}$	Output disable time from OE input		1	3	4	ns

† Typical values are at  $V_{CC} = 5 \text{ V}$ ,  $V_{EE} = -5.2 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

**DP8481**  
**10K TTL-TO-ECL LEVEL TRANSLATOR WITH LATCH**

SLLS036B - D3059, NOVEMBER 1987 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



NOTE A: ECL input rise and fall times at OE and  $\overline{LE}$  are  $2 \text{ ns} \pm 0.2 \text{ ns}$  from 20% to 80%. TTL input rise and fall times at D inputs are 3 ns maximum measured between 10% and 90%.

**Figure 1. Switching Time Waveforms**

# LT1030C QUAD LOW-POWER LINE DRIVER

SLLS048C - D3297, APRIL 1989 - REVISED MARCH 1993

- Low Supply Voltage . . .  $\pm 5\text{ V}$  to  $\pm 15\text{ V}$
- Supply Current . . . 500  $\mu\text{A}$  Typ
- Zero Supply Current When Shut Down
- Outputs Can Be Driven  $\pm 30\text{ V}$
- Output Open When Off (3-State)
- 10-mA Output Drive
- Output of Several Devices Can Be Paralleled
- Meets ANSI/EIA-232-D-1986 Specifications (Revision of EIA Std RS-232-C)
- Designed to Be Interchangeable With Linear Technology LT1030

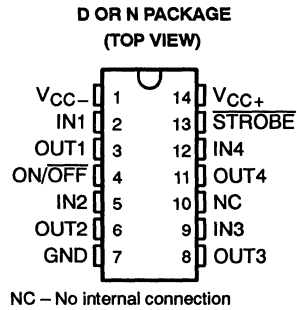
## description

The LT1030C is an EIA-232 line driver that operates over a  $\pm 5\text{-V}$  to  $\pm 15\text{-V}$  supply voltage range on low supply current. The device can be shut down to zero supply current. Current limiting fully protects the outputs from externally-applied voltages of  $\pm 30\text{ V}$ . Since the output swings to within 200 mV of the positive supply and to within 1 V of the negative supply, supply voltage requirements are minimized.

A major advantage of the LT1030C is the high-impedance output state when the device is off or powered down. This feature allows several different drivers on the same bus.

The device can be used as an EIA-232 driver, micropower interface, or level translator, among others.

The LT1030C is characterized for operation from  $0^\circ\text{C}$  to  $70^\circ\text{C}$ .

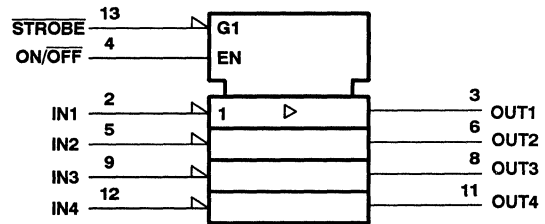


## AVAILABLE OPTIONS

TA	PACKAGE	
	SMALL OUTLINE (D)	PLASTIC DIP (N)
0°C to 70°C	LT1030CD	LT1030CN

The D package is available taped and reeled. Add the suffix R to the device type (i.e., LT1030CDR).

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## Terminal Functions

PIN	NO	DESCRIPTION
GND	7	Ground pin
IN1, IN2, IN3, IN4	2, 5, 9, 12	Logic inputs. Operate properly on TTL or CMOS levels. Output valid from $V_I = V_{CC-} + 2\text{ V}$ to 15 V. Connect to 5 V when not used.
ON/OFF	4	Shuts down entire circuit. Cannot be left open. For normally on operation, connect between 5 V and 10 V. If $V_{IL}$ is at or near 0.8 V, significant settling time may be required.
OUT1, OUT2, OUT3, OUT4	3, 6, 8, 11	Line driver outputs
STROBE	13	Forces all outputs low. Drive with 3 V. Strobe terminal input impedance is approximately 2 k $\Omega$ to GND. Leave open when not used.
VCC+	14	Positive supply
VCC-	1	Negative supply

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



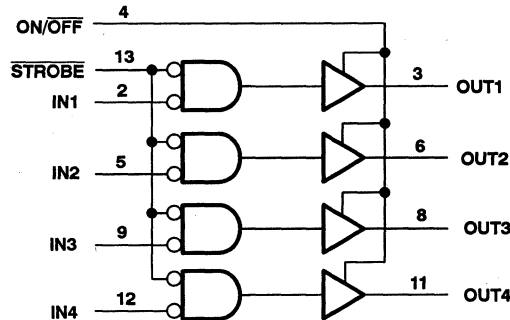
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# LT1030C QUAD LOW-POWER LINE DRIVER

SLLS048C – D3297, APRIL 1989 – REVISED MARCH 1993

## logic diagram



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC+}$ (see Note 1)	0 V to 15 V
Supply voltage range, $V_{CC-}$	0 V to -15 V
Input voltage range, logic inputs, $V_I$	$V_{CC-}$ to 25 V
Input voltage range at ON/OFF, $V_I$	0 V to 12 V
Output voltage range (any output)	$V_{CC+} - 30$ V to $V_{CC-} + 30$ V
Duration of output short circuit to $\pm 30$ V at (or below) 25°C (see Note 2)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values, except differential voltages, are with respect to GND.

2. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	5	15	V
Supply voltage, $V_{CC-}$	-5	-15	V
High-level input voltage, $V_{IH}$ (see Note 3)	2	15	V
Low-level input voltage, $V_{IL}$ (see Note 3)		0.8	V
Operating free-air temperature, $T_A$	0	70	°C

NOTE 3: These  $V_{IH}$  and  $V_{IL}$  specifications apply only for inputs IN1–IN4. For operating levels for ON/OFF, see Figure 2.



# LT1030C QUAD LOW-POWER LINE DRIVER

SLLS048C - D3297, APRIL 1989 - REVISED MARCH 1993

**electrical characteristics over operating free-air temperature range,  $V_{CC\pm} = \pm 5\text{ V}$  to  $\pm 15\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OM+}$	Maximum positive peak output voltage swing	$I_O = -2\text{ mA}$ , $T_A = 25^\circ\text{C}$	$V_{CC+} - 0.3$	$V_{CC+} - 0.1$		V
$V_{OM-}$	Maximum negative peak output voltage swing	$I_O = 2\text{ mA}$ , $T_A = 25^\circ\text{C}$		$V_{CC-} + 0.9$	$V_{CC-} + 1.4$	V
$I_{IH}$	High-level input current	$V_I \geq 2\text{ V}$ , $T_A = 25^\circ\text{C}$		2	20	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I \leq 0.8\text{ V}$ , $T_A = 25^\circ\text{C}$		10	20	$\mu\text{A}$
$I_I$	Input current, ON/OFF	$V_I = 0$		-0.1	-10	$\mu\text{A}$
		$V_I = 5\text{ V}$		30	65	
$I_O$	Output current	$T_A = 25^\circ\text{C}$	5	12		mA
$I_{OZ}$	Off-state output current	$V_O = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$ , ON/OFF at 0.4 V		$\pm 2$	$\pm 100$	$\mu\text{A}$
$I_{CC}$	Supply current (all outputs low)	$V_I \geq 2.4\text{ V}$ , $I_O = 0$		500	1000	$\mu\text{A}$
$I_{CC(off)}$	Off-state supply current	ON/OFF at 0.4 V			10	$\mu\text{A}$
		ON/OFF at 0.1 V		10	150	

**operating characteristics,  $V_{CC\pm} = \pm 5\text{ V}$  to  $\pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
SR	Driver slew rate	$R_L = 3\text{ k}\Omega$ , $C_L = 51\text{ pF}$	4	15	30	V/ $\mu\text{s}$

† All typical values are at  $V_{CC\pm} = \pm 12\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

# LT1030C QUAD LOW-POWER LINE DRIVER

SLLS048C - D3297, APRIL 1989 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE SWING  
vs  
OUTPUT CURRENT

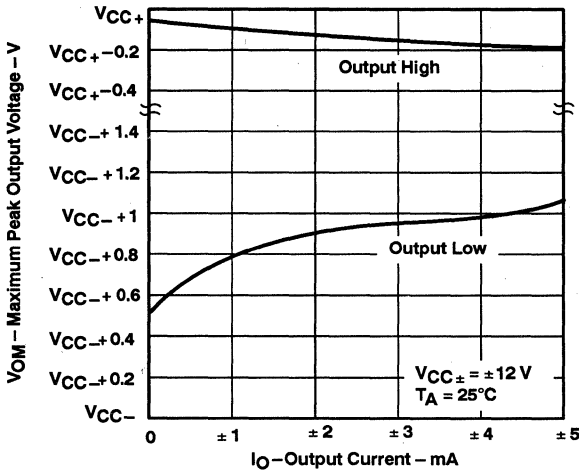


Figure 1

ON/OFF TERMINAL VOLTAGE  
vs  
FREE-AIR TEMPERATURE

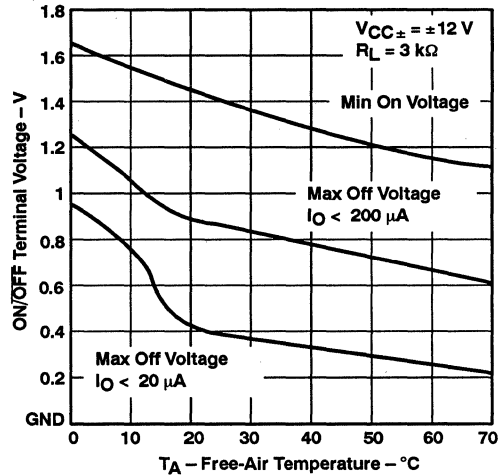


Figure 2

MAXIMUM PEAK OUTPUT VOLTAGE SWING  
vs  
FREE-AIR TEMPERATURE

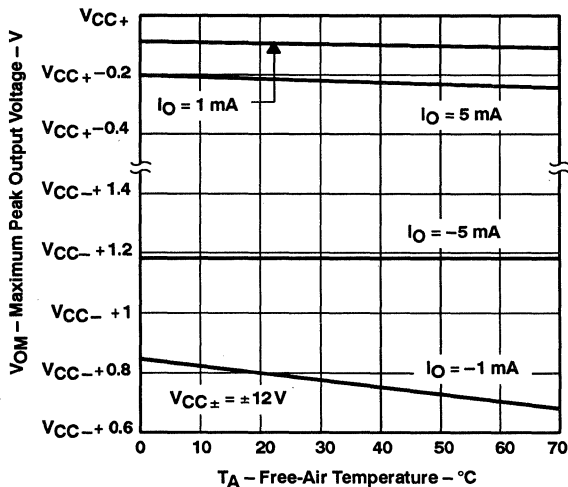


Figure 3

ON/OFF TERMINAL CURRENT  
vs  
ON/OFF TERMINAL VOLTAGE

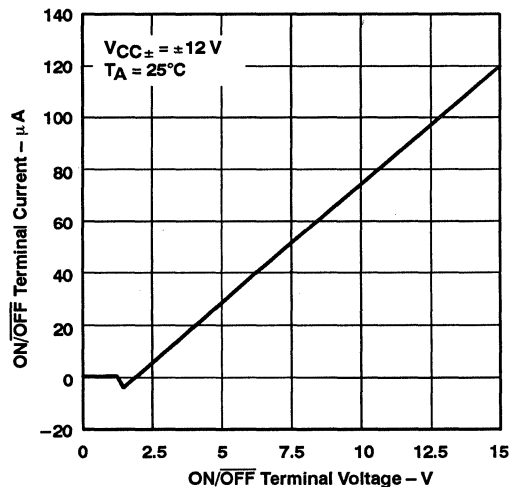


Figure 4



TYPICAL CHARACTERISTICS

OUTPUT CURRENT LIMIT  
 vs  
 FREE-AIR TEMPERATURE

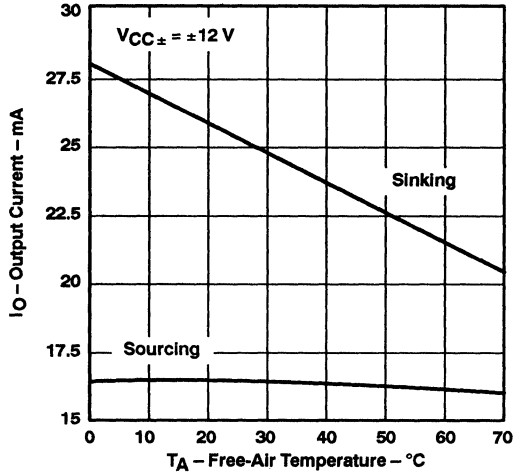


Figure 5

OFF-STATE OUTPUT CURRENT  
 vs  
 FREE-AIR TEMPERATURE

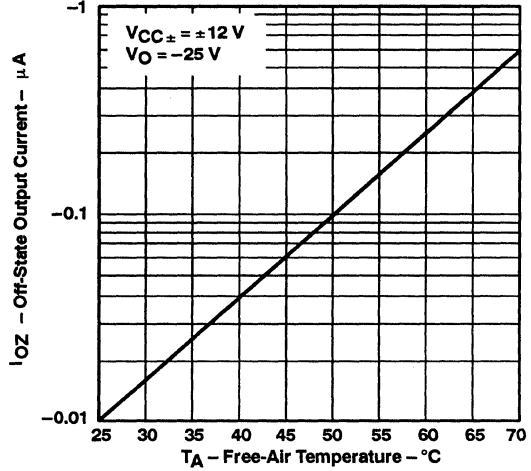


Figure 6

OFF-STATE SUPPLY CURRENT  
 vs  
 FREE-AIR TEMPERATURE

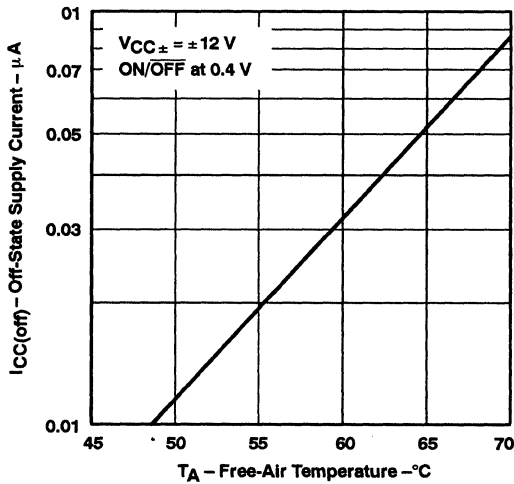


Figure 7

SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE

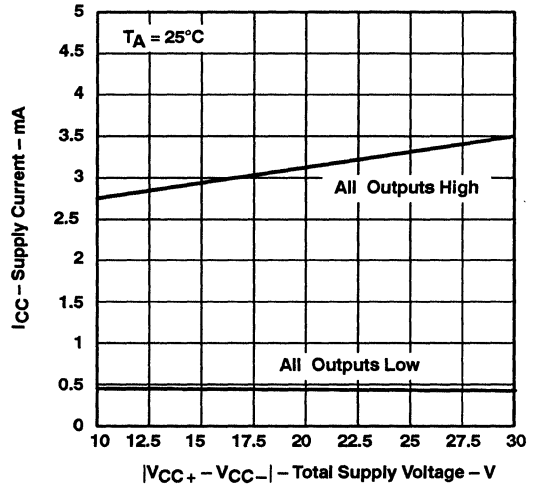


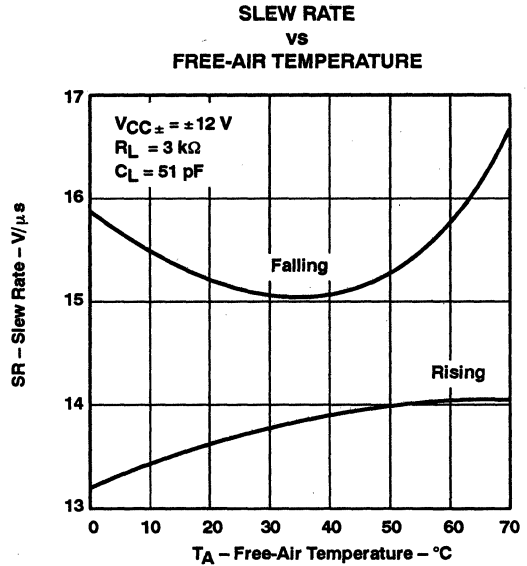
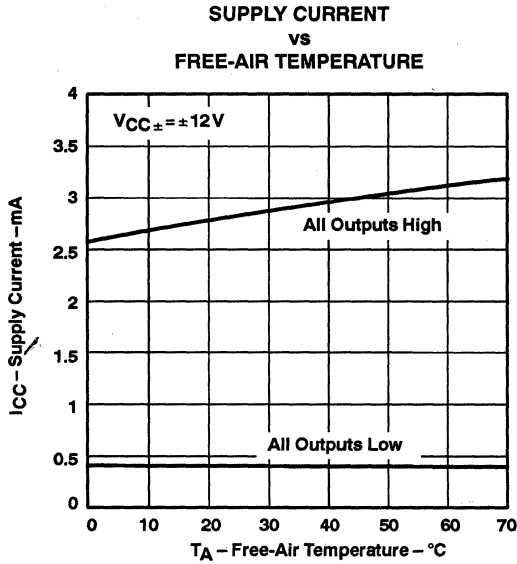
Figure 8



# LT1030C QUAD LOW-POWER LINE DRIVER

SLLS048C - D3297, APRIL 1989 - REVISED MARCH 1993

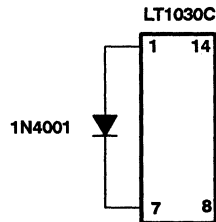
## TYPICAL CHARACTERISTICS



## APPLICATION INFORMATION

### forward biasing the substrate

As with other bipolar integrated circuits, forward biasing the substrate diode can cause problems. The LT1030C will draw high current from  $V_{CC+}$  to GND if  $V_{CC-}$  is open circuited or pulled above ground. If this is possible, connecting a diode from  $V_{CC-}$  to GND will prevent the high-current state. Any low-cost diode can be used (see Figure 11).



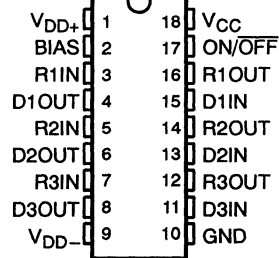
**Figure 11. Connecting a Diode From  $V_{CC-}$  to GND**

# LT1039 TRIPLE EIA-232 LINE TRANSCEIVER

SLLS105B – D3627, FEBRUARY 1991 – REVISED JANUARY 1992

- Meets All EIA-232-D (Revision of RS-232-C) Specifications
- Three Independent Drivers and Receivers Per Package
- EIA-232 Inputs and Outputs Withstand  $\pm 30$  V
- 3-State Outputs
- All Outputs Are Short-Circuit Protected
- Virtually Zero Supply Current When Shutdown
- Output of Several Devices Can Be Paralleled
- Operates From  $\pm 5$ -V to  $\pm 15$ -V Supplies
- Designed to Be Interchangeable With Linear Technology LT1039

DW OR N PACKAGE  
(TOP VIEW)



## description

The LT1039 is a triple EIA-232 line transceiver designed to meet the requirements of Standard EIA-232-D. All outputs are fully protected against an overload or short to ground. A major advantage of the LT1039 is high-impedance output states when the device is off or powered down. This feature allows several different devices to be connected together on the same bus.

The bias pin provides a receiver to be kept alive when the LT1039 is shutdown (ON/OFF = low).

The LT1039 is characterized for operation from 0°C to 70°C.

### AVAILABLE OPTIONS

T <sub>A</sub>	PACKAGE	
	SMALL OUTLINE (DW)	PLASTIC DIP (N)
0°C to 70°C	LT1039CDW	LT1039CN

The DW package is available taped and reeled. Add the suffix R to the device type (i.e., LT1039CDWR).

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

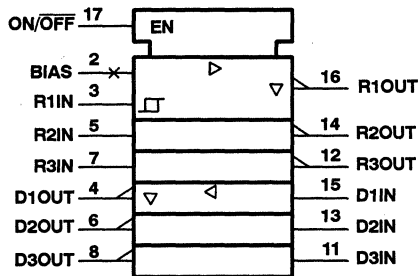
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# LT1039 TRIPLE EIA-232 LINE TRANSCEIVER

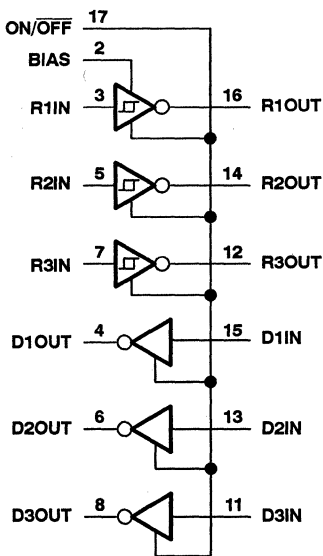
SLLS105B - D3627, FEBRUARY 1991 - REVISED JANUARY 1992

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram



**LT1039**  
**TRIPLE EIA-232 LINE TRANSCEIVER**

SLLS105B – D3627, FEBRUARY 1991 – REVISED JANUARY 1992

**Terminal Functions**

PIN		DESCRIPTION
NAME	NO.	
BIAS	2	Keeps receiver 1 alive while the LT1039 is in the shutdown mode. Leave BIAS open when not in use.
D1IN, D2IN, D3IN	15, 13, 11	Line driver inputs. Operate properly on TTL or CMOS levels. Output valid from $V_I = (V_{DD-}) + (2 \text{ to } 15 \text{ V})$ . Connect to 5 V when not used.
D1OUT, D2OUT, D3OUT	4, 6, 8	Line driver outputs
GND	10	Ground
ON/OFF	17	Shuts down entire circuit. Cannot be left open. If $V_{IL}$ is at or near 0.8 V, significant settling time may be required.
R1IN, R2IN, R3IN	3, 5, 7	Receiver inputs. Input impedance is normally 30 k $\Omega$ . Accepts EIA-232 voltage levels and has 0.4 V of hysteresis to provide noise immunity.
R1OUT, R2OUT, R3OUT	16, 14, 12	Receiver outputs with TTL/CMOS voltage levels
V <sub>DD+</sub>	1	Positive supply voltage for driver
V <sub>DD-</sub>	9	Negative supply voltage for driver
V <sub>CC</sub>	18	5-V supply voltage for receivers

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage range, V <sub>DD+</sub> (see Note 1)	0 V to 15 V
Supply voltage range, V <sub>DD-</sub>	0 V to -15 V
Supply voltage, V <sub>CC</sub>	7 V
Input voltage range, driver input	V <sub>DD-</sub> to 25 V
receiver input	$\pm 30 \text{ V}$
Input voltage range, ON/OFF	0 V to 12 V
Output voltage range, driver output	V <sub>DD+</sub> - 30 V to V <sub>DD-</sub> + 30 V
Duration of output short circuit at (or below) T <sub>A</sub> = 25°C (to $\pm 30 \text{ V}$ , see Note 2)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, T <sub>A</sub>	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values, except differential voltages, are with respect to the GND terminal.

2. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	T <sub>A</sub> $\leq$ 25°C POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
DW	1025 mW	8.2 mW/°C	656 mW
N	1150 mW	9.2 mW/°C	736 mW



# LT1039

## TRIPLE EIA-232 LINE TRANSCEIVER

SLLS105B – D3627, FEBRUARY 1991 – REVISED JANUARY 1992

### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{DD+}$	5	12	15	V
Supply voltage, $V_{DD-}$	-5	-12	-15	V
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$ (see Note 3)	2			V
Low-level input voltage, $V_{IL}$ (see Note 3)			0.8	V
Operating free-air temperature, $T_A$	0		70	°C

NOTE 3:  $V_{IH}$  and  $V_{IL}$  specifications apply only for inputs D1IN to D3IN.

### DRIVER SECTION

electrical characteristics over recommended operating free-air temperature range,  $V_{DD\pm} = \pm 11.4$  V to  $\pm 12.6$  V, ON/OFF at 2.5 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OM+}$ Maximum positive peak output voltage swing	Load = 3 k $\Omega$ to GND	$V_{DD+} - 0.4$	$V_{DD+} - 0.4$		V
$V_{OM-}$ Maximum negative peak output voltage swing		$V_{DD-} + 1.0$	$V_{DD-} + 1.5$		
$I_{IH}$ High-level input current	$V_I \geq 2$ V		1	20	$\mu$ A
$I_{IL}$ Low-level input current	$V_I \leq 0.8$ V		5	20	$\mu$ A
	$V_I = 0$			15	
	$V_I = 5$ V			80	
	Sourcing current, $V_O = 0$	5	15		
	Sinking current, $V_O = 0$	-5	-15		
$I_{OZ}$ Off-state output current	$V_O = \pm 18$ V, $V_I = 0$ , ON/OFF at 0.4 V		$\pm 10$	$\pm 200$	$\mu$ A
$I_{CC}$ Supply current	$I_O = 0$		4	8	mA
$I_{CC(off)}$ Off-state supply current	ON/OFF at 0.4 V		1	100	$\mu$ A
SR Slew rate	$R_L = 3$ k $\Omega$ , $C_L = 51$ pF	4	15	30	V/ $\mu$ s

† All typical values are at  $V_{DD\pm} = \pm 12$  V,  $T_A = 25^\circ\text{C}$ .

**RECEIVER SECTION**

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature, ON/OFF at 2.5 V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T-}$ Negative-going input threshold voltage		0.5	1.3		V
$V_{T+}$ Positive-going input threshold voltage			1.7	2.8	V
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )		0.1	0.4	1	V
$V_{OL}$ Low-level output voltage	$I_O = -1.6$ mA		0.4	0.5	V
$V_{OH}$ High-level output voltage	$I_O = 160$ $\mu$ A	3.5	4.8		V
$I_{OS}$ Short-circuit output current	Sinking current, $V_O = V_{CC}$	-10			mA
	Sourcing current, $V_O = 0$	0.5	1		
$I_{OZ}$ Off-state (high-impedance state) output current	$V_O = 0$ to $V_{CC}$ , $V_I = 0$ , ON/OFF at 0.4 V		$\pm 1$	$\pm 10$	$\mu$ A
$I_{CC}$ Supply current	$I_O = 0$		4	7	mA
$I_{CC(off)}$ Off-state supply current	ON/OFF at 0.4 V		1	100	$\mu$ A

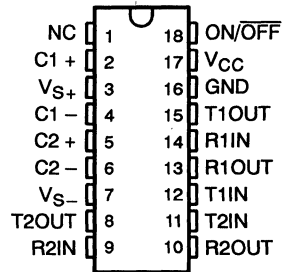
† All typical values ground are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .



**LT1080, LT1081**  
**ADVANCED LOW-POWER**  
**5-V DUAL EIA-232 DRIVER/RECEIVERS**  
 SLLS050A - D3121, SEPTEMBER 1989 - REVISED FEBRUARY 1991

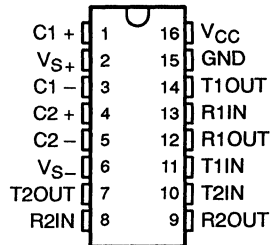
- Operates With Single 5-V Power Supply
- Generates  $\pm 9$ -V Supply Voltages With Only 1- $\mu$ F Capacitors
- Designed to Avoid Latch-Up
- CMOS Comparable Low Power . . . 60 mW
- Features Superior to CMOS:
  - Improved Speed . . . Operates Over 64K Baud
  - Improved Protection . . . Outputs Can Be Forced to  $\pm 30$  V Without Damage
  - 3-State Outputs Are at High Impedance When Off
- Power Additional EIA-232 Drivers . . . 10 mA
- 1- $\mu$ A Supply Current in Shutdown
- Available With or Without Shutdown
- Suitable for ANSI/EIA-232-D-1986 Applications (Revision of EIA Std RS-232-C)
- Designed to Be Interchangeable With Linear Technology LT1080 and LT1081

LT1080C, LT1080I . . . DW OR N PACKAGE  
(TOP VIEW)



NC - No internal connection

LT1081C, LT1081I . . . DW OR N PACKAGE  
(TOP VIEW)



**description**

The LT1080 and LT1081 are dual driver/receivers that include a charge pump to supply EIA-232 voltage levels from a single 5-V supply. These interface-optimized devices are designed to avoid latch-up and provide a realistic balance between CMOS levels of power dissipation and real-world requirements for ruggedness. The driver outputs are fully protected against overload and can be shorted to  $\pm 30$  V. Unlike CMOS devices, the advanced architecture of the LT1080 and LT1081 does not load the signal line when shut down or when the power is off. Both the receiver and EIA-232 outputs are put into a high-impedance state. An advanced output stage allows driving higher capacitive loads at higher speeds with exceptional resistance to ESD.

Applications for these devices include portable computers, battery-powered EIA-232 systems, power-supply generators, terminals, and modems.

The LT1080C and LT1081C are characterized for operation from 0°C to 70°C. The LT1080I and LT1081I are characterized for operation from -40°C to 85°C.

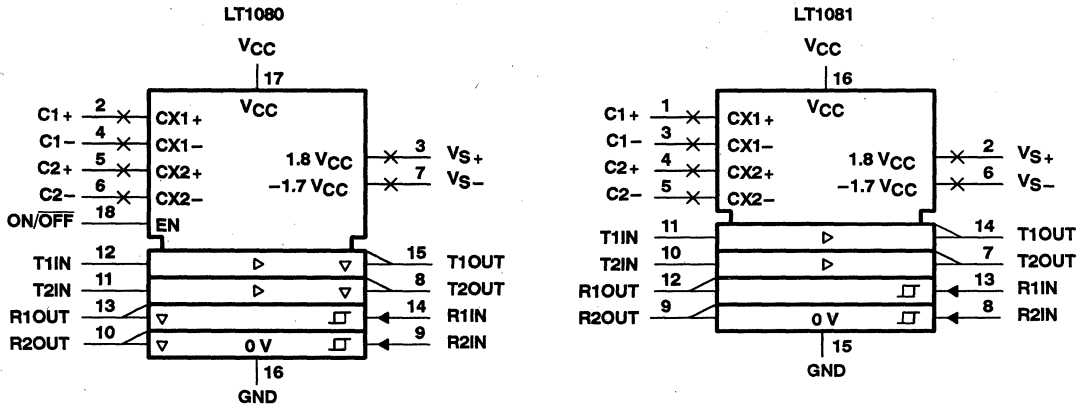
PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.





**LT1080, LT1081**  
**ADVANCED LOW-POWER**  
**5-V DUAL EIA-232 DRIVER/RECEIVERS**  
 SLLS050A - D3121, SEPTEMBER 1989 - REVISED FEBRUARY 1991

**logic symbols†**



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

**Terminal Functions**

PIN	NAME	NO.†	I/O	DESCRIPTION
C1 +	2	[1]		External capacitor C1
C1 -	4	[3]		
C2 +	5	[4]		External capacitor C2
C2 -	6	[5]		
GND	16	[15]		Ground pin
ON/OFF	18		I	Controls the operation mode and is TTL/CMOS compatible. A low logic level places the device in the shutdown mode, which reduces input supply current to near zero and places both driver and receiver outputs in a high-impedance state. This input is not available on the LT1081.
R1IN	14	[13]	I	Receiver input. Accepts EIA-232 voltage levels ( $\pm 30$ V) and has hysteresis to provide noise immunity.
R2IN	9	[8]	I	Same as R1IN
R1OUT	13	[12]	O	Receiver output with TTL/CMOS voltage levels. Output is in a high-impedance state when in the shutdown mode or when $V_{CC} = 0$ to allow bus operation. Fully short-circuit protected to GND or $V_{CC}$ with power on, power off, or in the shutdown mode.
R2OUT	10	[9]	O	Same as R1OUT
T1IN	12	[11]	I	EIA-232 driver input pin. Input is TTL/CMOS compatible. Unused inputs should be tied to $V_{CC}$ .
T2IN	11	[10]	I	Same as T1IN
T1OUT	15	[14]	O	Driver output with EIA-232 voltage levels. Outputs are in a high-impedance state when in the shutdown mode or when $V_{CC} = 0$ to allow bus operation. Fully short-circuit protected to GND or $V_{CC}$ with power on, power off, or in the shutdown mode.
T2OUT	8	[7]	O	Same as T1OUT
$V_{S+}$	3	[2]		Positive supply for EIA-232 drivers. Requires an external capacitor (1- $\mu$ F) for charge storage.
$V_{S-}$	7	[6]		Negative supply for EIA-232 drivers. Requires an external capacitor (1- $\mu$ F) for charge storage.
$V_{CC}$	17	[16]		Input supply pin. Supply current drops to near zero in the shutdown mode. Driver and receiver outputs are in a high-impedance state when $V_{CC} = 0$ .

† Pin numbers in brackets are for the LT1081.



**LT1080, LT1081**  
**ADVANCED LOW-POWER**  
**5-V DUAL EIA-232 DRIVER/RECEIVERS**  
SLLS050A - D3121, SEPTEMBER 1989 - REVISED FEBRUARY 1991

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	6 V
Positive output supply voltage	12 V
Negative output supply voltage	-12 V
Input voltage range: Driver	$\pm 12$ V
Receiver	$\pm 30$ V
ON/OFF	GND to 12 V
Output voltage range: T1OUT, T2OUT	$V_{S-} + 30$ V to $V_{S+} - 30$ V
R1OUT, R2OUT	-0.3 V to $V_{CC} + 0.3$ V
Duration of output short circuit at (or below) 25°C: $V_{S+}$	30 s
$V_{S-}$	30 s
Driver or receiver output	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: LT1080C, LT1081C	0°C to 70°C
LT1080I, LT1081I	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
DW	1025 mW	8.2 mW/°C	656 mW	533 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

**recommended operating conditions**

	LT1080C, LT1081C			LT1080I, LT1081I			UNIT
	MIN	TYP	MAX	MIN	TYP	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.5	5	5.5	V
High-level input voltage, $V_{IH}$ (T1IN, T2IN)	2		5.5	2		5.5	V
Low-level input voltage, $V_{IL}$ (T1IN, T2IN)			0.8			0.8	V
Operating free-air temperature, $T_A$	0		70	-40		85	°C



**LT1080, LT1081**  
**ADVANCED LOW-POWER**  
**5-V DUAL EIA-232 DRIVER/RECEIVERS**  
 SLLS050A – D3121, SEPTEMBER 1989 – REVISED FEBRUARY 1991

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature,  $V_I = 3\text{ V}$  (unless otherwise noted)

**driver section**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OM+}$ Positive-peak output voltage	$R_L = 3\text{ k}\Omega$ to GND	5	7.5		V
$V_{OM-}$ Negative-peak output voltage	$R_L = 3\text{ k}\Omega$ to GND	-5	-6.5		V
$I_{IH}$ High-level input current	$V_I = 2\text{ V}$ to 5.5 V		5	20	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_I = 0.8\text{ V}$		-5	-20	$\mu\text{A}$
$I_{OHS}$ Output short-circuit current (sourcing)	$V_I = 0$ , $V_O = 0$ , $T_A = 25^\circ\text{C}$	-7	12		mA
$I_{OLS}$ Output short-circuit current (sinking)	$V_I = 2\text{ V}$ , $V_O = 0$ , $T_A = 25^\circ\text{C}$	7	-12		mA
$I_{OZ}$ High-impedance output current	Shutdown mode, $V_O = \pm 30\text{ V}$			100	$\mu\text{A}$

**receiver section**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going input threshold voltage	LT1080C, LT1081C		1.7	2.4	V
	LT1080I, LT1081I		1.7	3	
$V_{T-}$ Negative-going input threshold voltage	LT1080C, LT1081C	0.8	1.3		V
	LT1080I, LT1081I	0.2	1.3		
$V_{hys}$ Input hysteresis		0.1	0.4	1	V
$r_i$ Input resistance	$T_A = 25^\circ\text{C}$	3	5	7	k $\Omega$
$V_{OH}$ High-level output voltage	$I_{OH} = -160\text{ }\mu\text{A}$ , $V_{CC} = 5\text{ V}$	3.5	4.8		V
$V_{OL}$ Low-level output voltage	$I_{OL} = 1.6\text{ mA}$		0.2	0.4	V
$I_{OSH}$ Output short-circuit current (sourcing)	$V_I = 3\text{ V}$ , $V_O = 0$ , $T_A = 25^\circ\text{C}$	-0.6	-1		mA
$I_{OSL}$ Output short-circuit current (sinking)	$V_I = 3\text{ V}$ , $V_O = V_{CC}$ , $T_A = 25^\circ\text{C}$	10	20		mA
$I_{OZ}$ High-impedance output current	Shutdown mode, $V_O = 0$ to $V_{CC}$			10	$\mu\text{A}$

**power supply section,  $V_{CC} = 5\text{ V}$ , driver outputs low (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{S+}$ Supply output voltage	$I_S = 0$ , $T_A = 25^\circ\text{C}$	8	9		V
	$I_S = 10\text{ mA}$ , $T_A = 25^\circ\text{C}$	7	8		
	$I_S = 15\text{ mA}$ , $T_A = 25^\circ\text{C}$	6.5	7.5		
$V_{S-}$ Supply output voltage	$I_S = 0$ , $T_A = 25^\circ\text{C}$	-7.5	-8.5		V
	$I_S = 10\text{ mA}$ , $T_A = 25^\circ\text{C}$	-5.5	-6.5		
	$I_S = 15\text{ mA}$ , $T_A = 25^\circ\text{C}$	-5	-6		
$I_{CC}$ Supply current			10	22	mA
$I_{CC(off)}$ Off-state supply current (LT1080)	ON/OFF at 0.4 V			100	$\mu\text{A}$
$I_i$ Input current (ON/OFF) (LT1080)	$V_I = 5\text{ V}$			80	$\mu\text{A}$
	$V_I = 0\text{ V}$			-15	

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

**operating characteristics over recommended range of supply voltage,  $V_I = 3\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Driver slew rate	$R_L = 3\text{ k}\Omega$ to 7 k $\Omega$ , $C_L = 560\text{ pF}$ , See Note 2	4	15	30	V/ $\mu\text{s}$
$t_r(\text{supply})$ Supply rise time (see Note 3)	$C_1 - C_4 = 1\text{ }\mu\text{F}$		1		ms

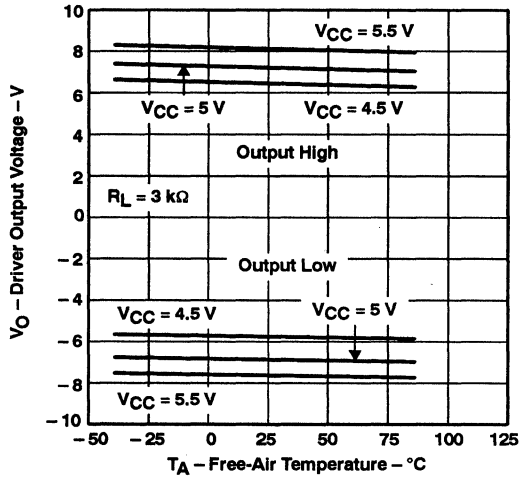
NOTES: 2. Meets EIA-232-D specifications for capacitive loads greater than 560 pF.

3. Time from either shutdown input ON/OFF (LT1080) goes active high or  $V_{CC}$  power on (LT1081) until the output voltages reach  $V_{S+} \geq 6\text{ V}$  and  $V_{S-} \leq -6\text{ V}$ .



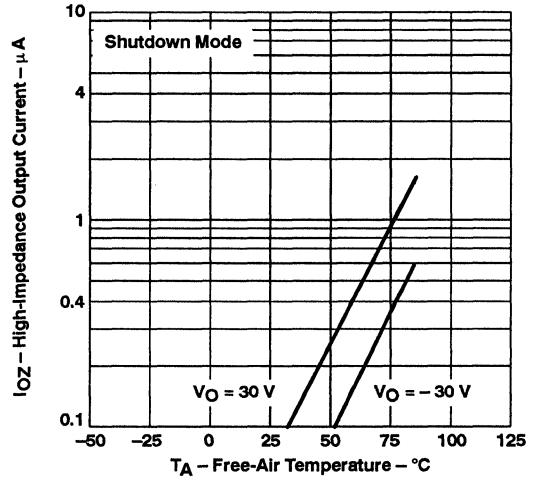
**TYPICAL CHARACTERISTICS†**

**HIGH- AND LOW-LEVEL DRIVER OUTPUT VOLTAGE vs FREE-AIR TEMPERATURE**



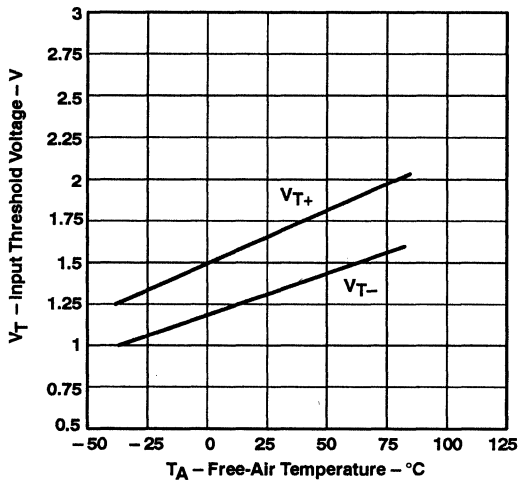
**Figure 1**

**HIGH-IMPEDANCE DRIVER OUTPUT CURRENT vs FREE-AIR TEMPERATURE**



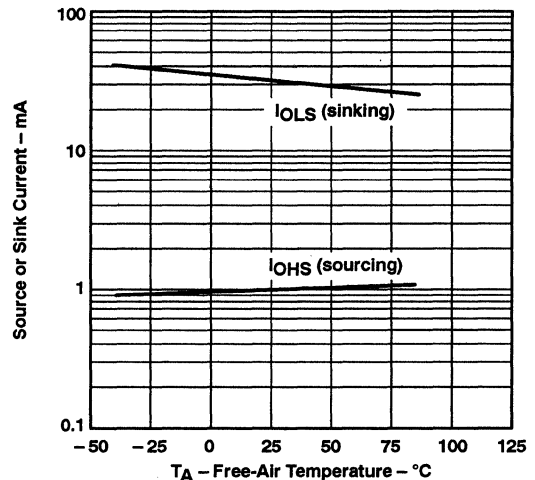
**Figure 2**

**POSITIVE-GOING AND NEGATIVE-GOING RECEIVER INPUT THRESHOLD VOLTAGE vs FREE-AIR TEMPERATURE**



**Figure 3**

**RECEIVER SOURCE CURRENT AND SINK CURRENT vs FREE-AIR TEMPERATURE**



**Figure 4**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

SUPPLY OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

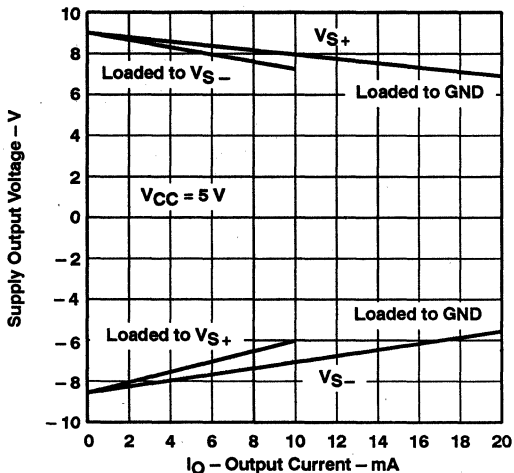


Figure 5

SUPPLY OUTPUT VOLTAGE  
 vs  
 ELAPSED TIME

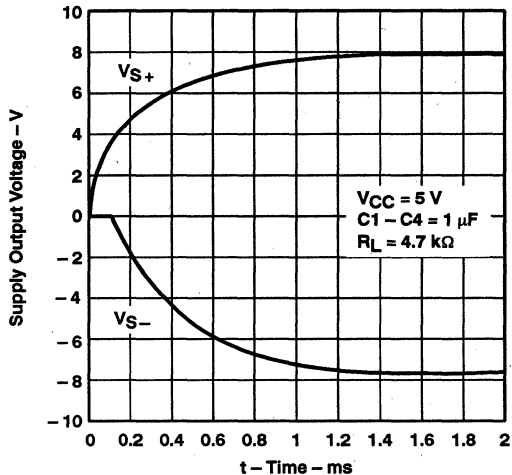


Figure 6

ON/OFF INPUT CURRENT  
 vs  
 INPUT VOLTAGE

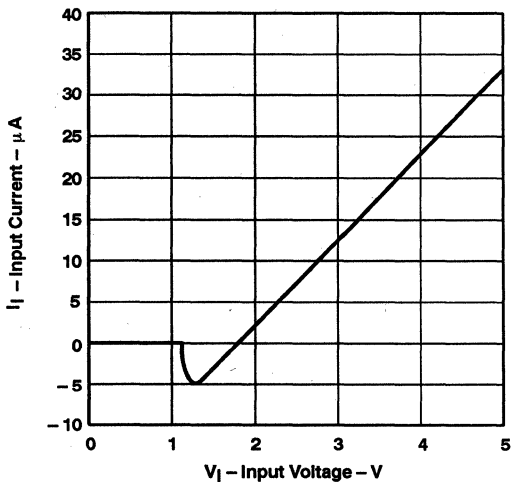


Figure 7

ON/OFF INPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

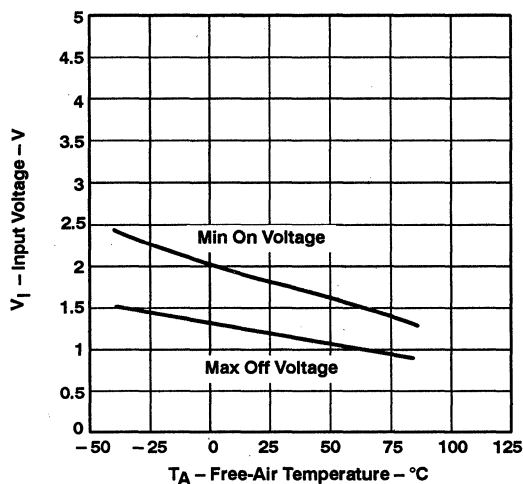


Figure 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

OFF-STATE SUPPLY CURRENT  
 vs  
 FREE-AIR TEMPERATURE

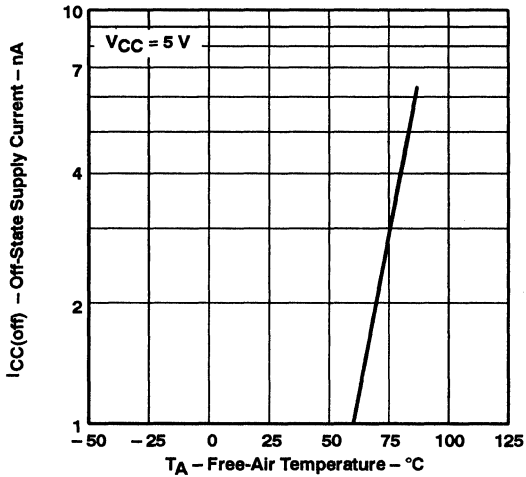


Figure 9

DRIVER and RECEIVER  
 PULSE RESPONSE

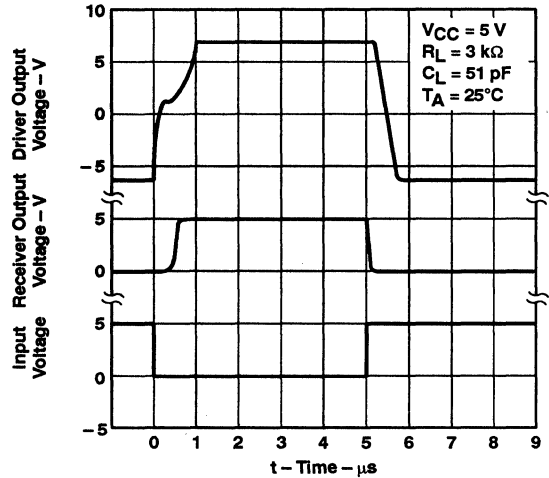


Figure 10

ON/OFF TO DRIVER OUTPUT  
 PULSE RESPONSE

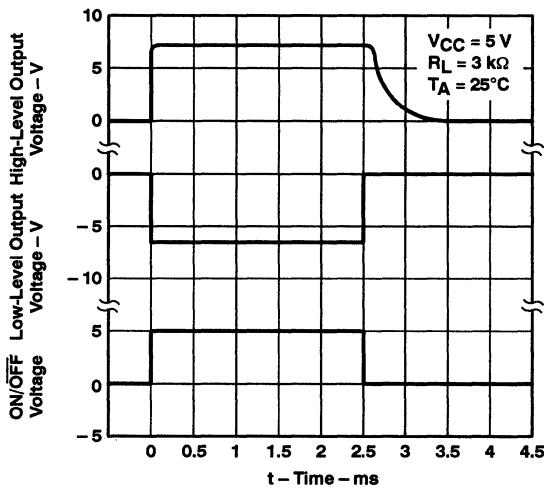


Figure 11

ON/OFF TO RECEIVER  
 PULSE RESPONSE

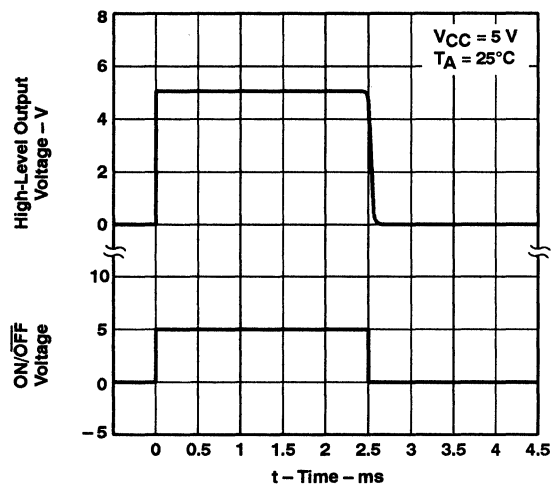
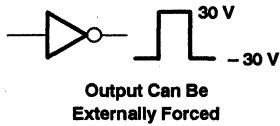


Figure 12

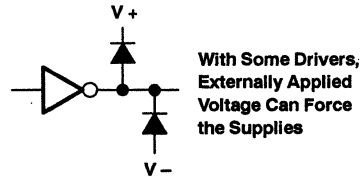
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**APPLICATION INFORMATION**

The driver output stage of the LT1080 offers significantly improved protection over older bipolar and CMOS designs (see Figures 13 and 14). In addition to limiting current, the driver output can be externally forced to  $\pm 30$  V without damage, excessive current flow, or supply disruption. Some drivers have diodes connected between the outputs and the supplies, allowing externally applied voltages to cause excessive supply voltage to develop.

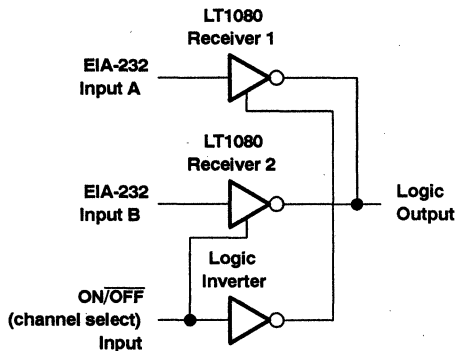


**Figure 13. LT1080/LT1081 Driver**

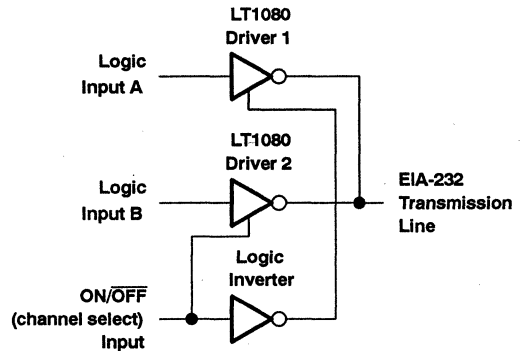


**Figure 14. Older EIA-232 Drivers and CMOS Drivers**

Placing the LT1080 in the shutdown mode (pin 18 low) puts both the driver and receiver outputs in a high-impedance state. This allows for bus operation and transceiver applications (see Figures 15–17). The shutdown mode also drops input supply current ( $V_{CC}$ ) to near zero for power-conscious systems.

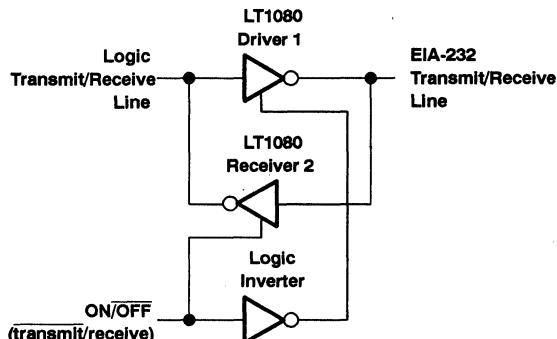


**Figure 15. Sharing a Receiver Line**



**Figure 16. Sharing a Transmitter Line**

**APPLICATION INFORMATION**

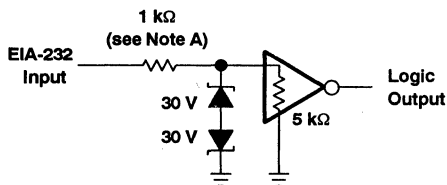


**Figure 17. Transceiver**

To protect against receiver input overloads in excess of  $\pm 30$  V, a voltage clamp can be placed on the data line and still maintain EIA-232 compatibility (see Figure 18).

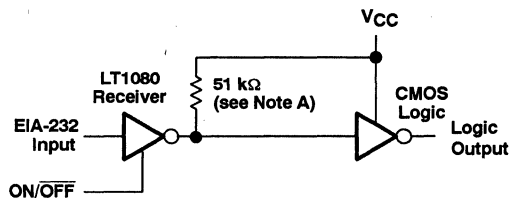
When driving CMOS logic from a receiver that will be used in the shutdown mode and when there is no other active receiver on the line, a  $51\text{-k}\Omega$  resistor can be placed from the logic input to  $V_{CC}$  to force a definite logic level when the receiver output is in a high-impedance state (see Figure 19).

The generated driver supplies ( $V_{S+}$  and  $V_{S-}$ ) may be used to power external circuitry such as other EIA-232 drivers or operational amplifiers (see Figure 20). They should be loaded with care, since excessive loading can cause the generated supply voltages to drop, causing the EIA-232 driver output voltages to fall below EIA-232 requirements. See Figure 5 for a comparison of generated supply voltage versus supply current.



NOTE A: A PTC thermistor will allow continuous overload of greater than  $\pm 100$  V.

**Figure 18. Input Overload Protection**

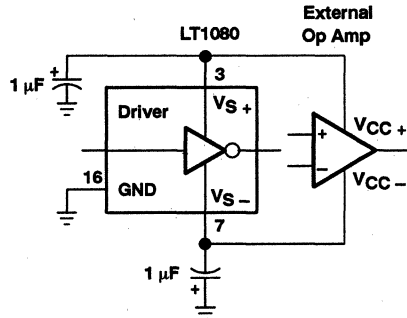


NOTE A: Forces logic input state when  $V_I$  is low.

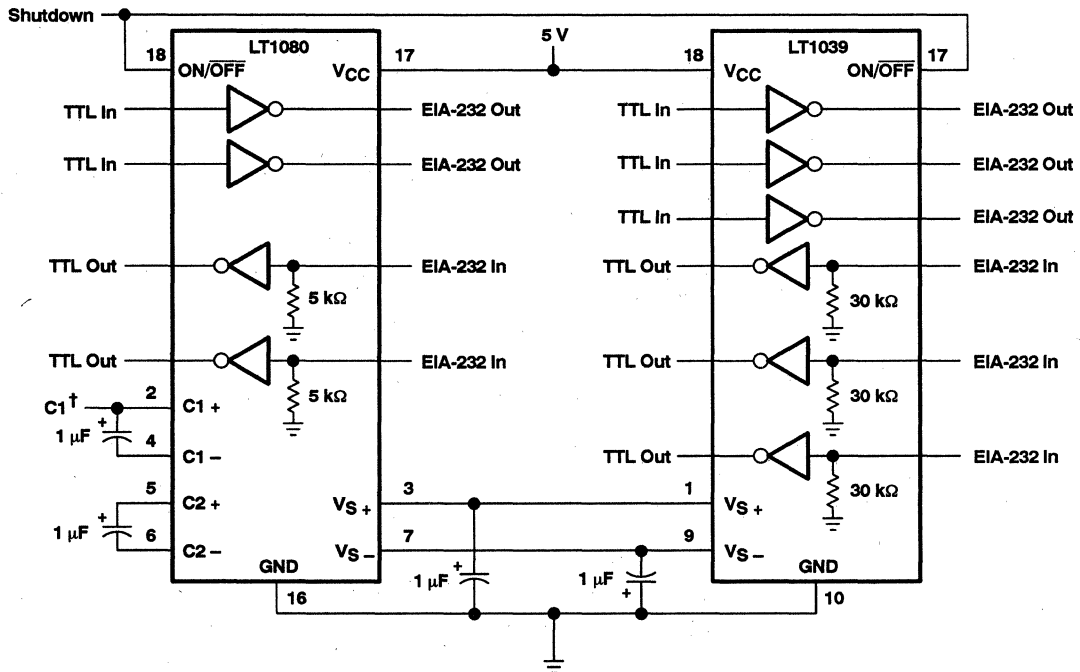
**Figure 19. Forcing a Definite Logic Level**



**APPLICATION INFORMATION**



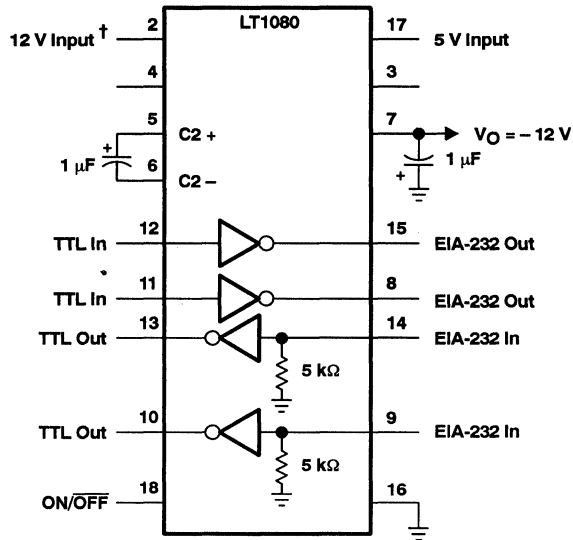
**Figure 20. Powering External Circuitry**



† In applications in which a separate second positive supply is available (such as 5 V and 12 V), the 12-V supply may be connected to pin 2 and C1 deleted. The power circuitry will then invert the 12-V supply. The 5-V supply is still needed to power the biasing circuitry and receivers.

**Figure 21. Supporting an LT1039 (Triple Driver/Receiver)**

**APPLICATION INFORMATION**



† C1+ used on LT1081

**Figure 22. Operating With 5 V and 12 V**



# MAX232, MAX2321 DUAL EIA-232 DRIVER/RECEIVER

SLLS047C – D3120, FEBRUARY 1989 – REVISED MARCH 1993

- Operates With Single 5-V Power Supply
- LinBICMOS™ Process Technology
- Two Drivers and Two Receivers
- ±30-V Input Levels
- Low Supply Current . . . 8 mA Typ
- Meets ANSI/EIA-232-D-1986 Specifications (Revision of EIA Standard RS-232-C)
- Designed to be Interchangeable With Maxim MAX232
- Applications
  - EIA-232 Interface
  - Battery-Powered Systems
  - Terminals
  - Modems
  - Computers
- ESO Protection Exceeds 2000 V Per MIL-STD-883, Method 3015

## description

The MAX232 is a dual driver/receiver that includes a capacitive voltage generator to supply EIA-232 voltage levels from a single 5-V supply. Each receiver converts EIA-232 inputs to 5-V TTL/CMOS levels. These receivers have a typical threshold of 1.3 V and a typical hysteresis of 0.5 V, and can accept ±30-V inputs. Each driver converts TTL/CMOS input levels into EIA-232 levels. The driver, receiver, and voltage-generator functions are available as cells in the Texas Instruments LinASIC™ library.

The MAX232 is characterized for operation from 0°C to 70°C. The MAX2321 is characterized for operation from -40°C to 85°C.

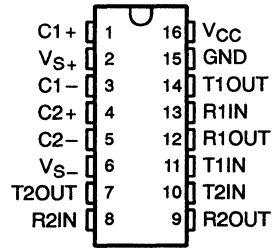
## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Input supply voltage range, $V_{CC}$ (see Note 1)	-0.3 V to 6 V
Positive output supply voltage range, $V_{S+}$	$V_{CC} - 0.3$ V to 15 V
Negative output supply voltage range, $V_{S-}$	-0.3 V to -15 V
Input voltage range: Driver	-0.3 V to $V_{CC} + 0.3$ V
Receiver	±30 V
Output voltage range: T1OUT, T2OUT	$V_{S-} - 0.3$ V to $V_{S+} + 0.3$ V
R1OUT, R2OUT	-0.3 V to $V_{CC} + 0.3$ V
Short-circuit duration: T1OUT, T2OUT	unlimited
Operating free-air temperature range: MAX232	0°C to 70°C
MAX2321	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

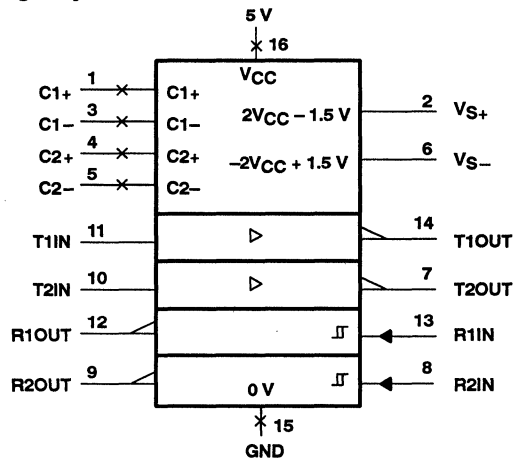
NOTE 1: All voltage values are with respect to network ground terminal.

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D, DW, OR N PACKAGE  
(TOP VIEW)



## logic symbol



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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# MAX232, MAX232I

## DUAL EIA-232 DRIVER/RECEIVER

SLLS047C - D3120, FEBRUARY 1989 - REVISED MARCH 1993

### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.5	V
High-level input voltage, $V_{IH}$ (T1IN, T2IN)	2			V
Low-level input voltage, $V_{IL}$ (T1IN, T2IN)			0.8	V
Receiver input voltage, R1IN, R2IN			$\pm 30$	V
Operating free-air temperature, $T_A$	0		70	$^{\circ}\text{C}$

### electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT		
$V_{OH}$	High-level output voltage	T1OUT, T2OUT	$R_L = 3\text{ k}\Omega$ to GND	5	7	V		
		R1OUT, R2OUT	$I_{OH} = -1\text{ mA}$	3.5				
$V_{OL}$	Low-level output voltage‡	T1OUT, T2OUT	$R_L = 3\text{ k}\Omega$ to GND		-7	V		
		R1OUT, R2OUT	$I_{OL} = 3.2\text{ mA}$		0.4			
$V_{T+}$	Receiver positive-going input threshold voltage	R1IN, R2IN	$V_{CC} = 5\text{ V}$ , $T_A = 25^{\circ}\text{C}$	1.7	2.4	V		
$V_{T-}$	Receiver negative-going input threshold voltage	R1IN, R2IN	$V_{CC} = 5\text{ V}$ , $T_A = 25^{\circ}\text{C}$	0.8	1.2	V		
$V_{hys}$	Input hysteresis	R1IN, R2IN	$V_{CC} = 5\text{ V}$	0.2	0.5	1	V	
$r_i$	Receiver input resistance	R1IN, R2IN	$V_{CC} = 5$ , $T_A = 25^{\circ}\text{C}$	3	5	7	$\text{k}\Omega$	
$r_o$	Output resistance	T1OUT, T2OUT	$V_{S+} = V_{S-} = 0$ , $V_O = \pm 2\text{ V}$	300			$\Omega$	
$I_{OS}^{\S}$	Short-circuit output current	T1OUT, T2OUT	$V_{CC} = 5.5\text{ V}$ , $V_O = 0$		$\pm 10$		mA	
$I_{IS}$	Short-circuit input current	T1IN, T2IN	$V_I = 0$			200	$\mu\text{A}$	
$I_{CC}$	Supply current		$V_{CC} = 5.5\text{ V}$ , $T_A = 25^{\circ}\text{C}$			8	10	mA

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^{\circ}\text{C}$ .

‡ The algebraic convention, in which the least positive (most negative) value is designated minimum, is used in this data sheet for logic voltage levels only.

§ Not more than one output should be shorted at a time.

### switching characteristics, $V_{CC} = 5\text{ V}$ , $T_A = 25^{\circ}\text{C}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}(R)$	Receiver propagation delay time, low-to-high-level output	See Figure 2		500		ns
$t_{PHL}(R)$	Receiver propagation delay time, high-to-low-level output	See Figure 2		500		ns
SR	Driver slew rate	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Figure 3			30	$\text{V}/\mu\text{s}$
SR(tr)	Driver transition region slew rate	See Figure 4		3		$\text{V}/\mu\text{s}$



# MAX232, MAX232I DUAL EIA-232 DRIVER/RECEIVER

SLLS047C - D3120, FEBRUARY 1989 - REVISED MARCH 1993

## APPLICATION INFORMATION

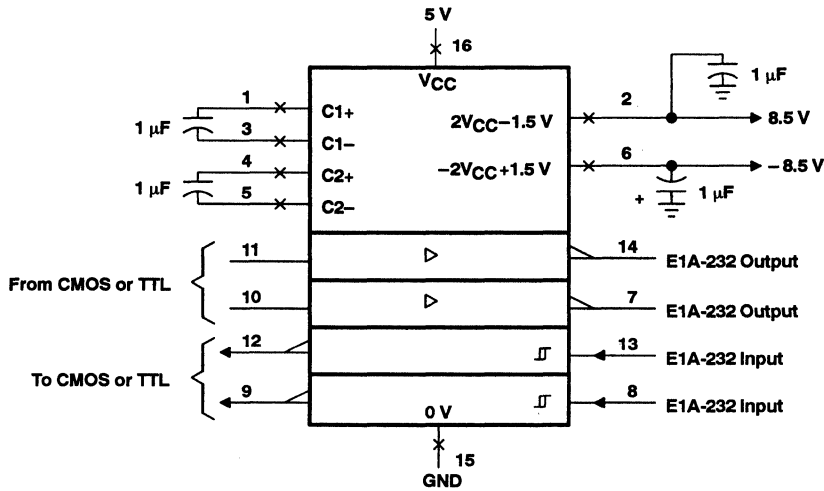
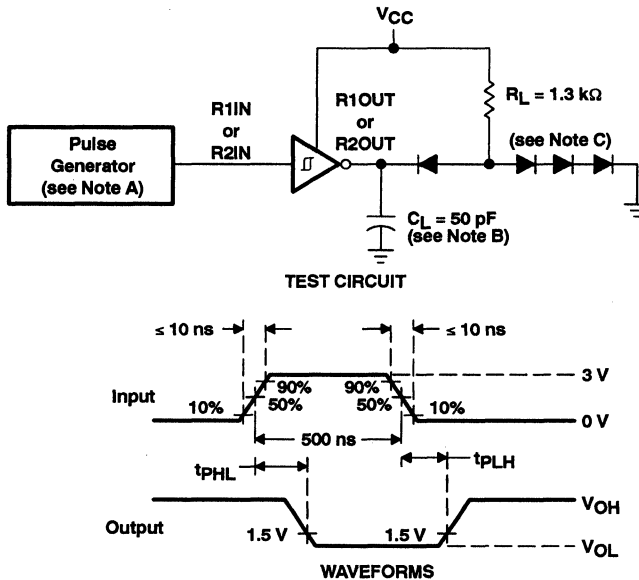


Figure 1. Typical Operating Circuit

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ , duty cycle  $\leq 50\%$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.

Figure 2. Receiver Test Circuit and Waveforms for  $t_{PHL}$  and  $t_{PLH}$  Measurement

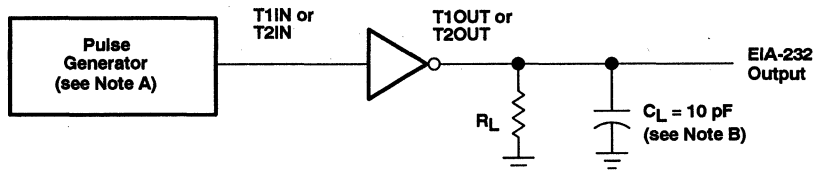
TEXAS  
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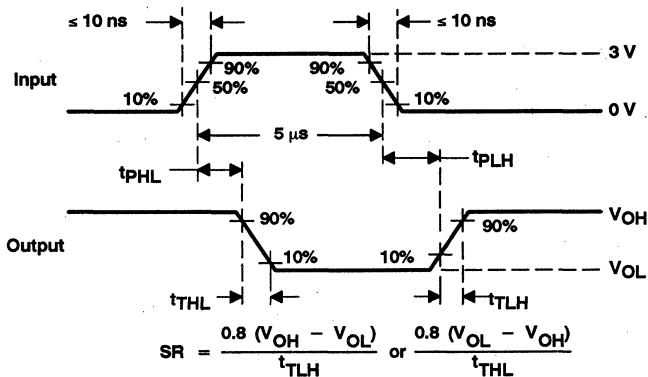
# MAX232, MAX232I DUAL EIA-232 DRIVER/RECEIVER

SLLS047C - D3120, FEBRUARY 1989 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



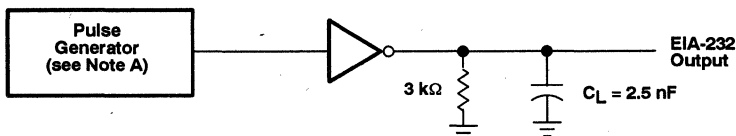
TEST CIRCUIT



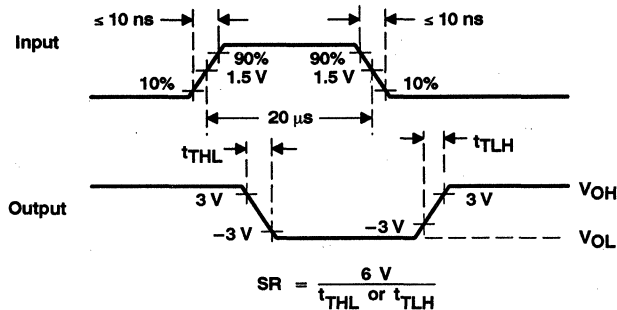
WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ , duty cycle  $\leq 50\%$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 3. Driver Test Circuit and Waveforms for  $t_{PHL}$  and  $t_{PLH}$  Measurement (5- $\mu$ s Input)



TEST CIRCUIT



WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ , duty cycle  $\leq 50\%$ .

Figure 4. Test Circuit and Waveforms for  $t_{THL}$  and  $t_{TLH}$  Measurement (20- $\mu$ s Input)

# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS012B - D3006, FEBRUARY 1986 - REVISED FEBRUARY 1993

- Four Independent Receivers With Common Enable Input
- High Input Sensitivity . . . 25 mV Max
- High Input Impedance
- MC3450 and MC3550 Have 3-State Outputs
- MC3452 Has Open-Collector Outputs
- Glitch-Free Power-Up/Power-Down Operation

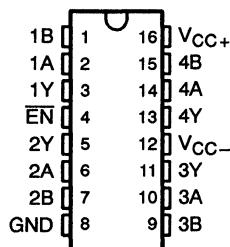
## description

The MC3450, MC3550, MC3452, and MC3552 are quad differential line receivers designed for use in balanced and unbalanced digital data transmission. The MC34/3550 and MC34/3552 are the same except that the MC3450 and MC3550 have 3-state outputs whereas the MC3452 and MC3552 have open-collector outputs, which permit the wire-AND function with similar output devices. The 3-state and open-collector outputs permit connection directly to a bus-organized system.

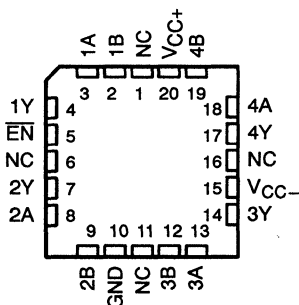
The MC3450, MC3550, MC3452, and MC3552 are designed for optimum performance when used with either the MC3453 or MC3553 quad differential line driver or SN75109A, SN75110A, and SN75112 dual differential drivers.

The MC3450 and MC3452 are characterized for operation from 0°C to 70°C. The MC3550 and MC3552 are characterized for operation over the full military temperature range of -55°C to 125°C.

MC3450, MC3452 . . . D OR N PACKAGE  
MC3550, MC3552 . . . J PACKAGE  
(TOP VIEW)



MC3550, MC3552 . . . FK PACKAGE  
(TOP VIEW)



NC—No internal connection

**THE MC3452 IS NOT  
RECOMMENDED FOR NEW DESIGN**

FUNCTION TABLE

DIFFERENTIAL INPUTS A - B	ENABLE EN	OUTPUT Y
$V_{ID} \geq 25 \text{ mV}$	L	H
$-25 \text{ mV} < V_{ID} < 25 \text{ mV}$	L	?
$V_{ID} \leq 25 \text{ mV}$	L	L
X	H	Z

H = high level, L = low level, ? = indeterminate,  
X = impedance (off)

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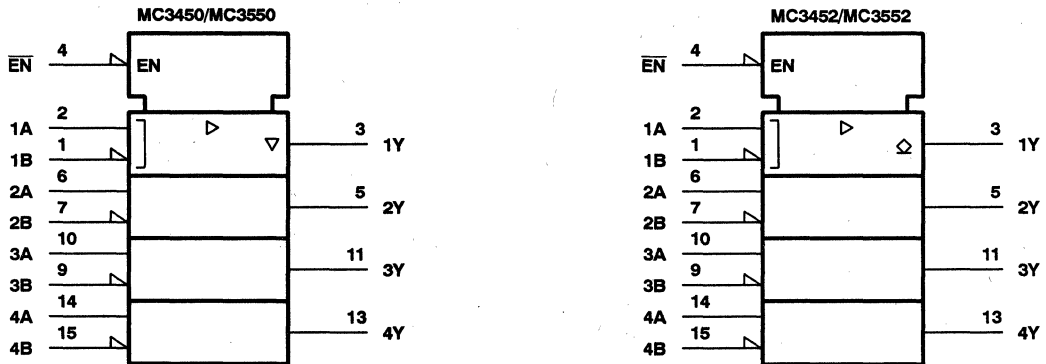
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# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

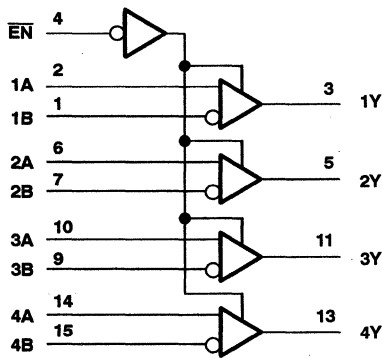
SLLS012B - D3006, FEBRUARY 1986 - REVISED FEBRUARY 1993

## logic symbols†



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

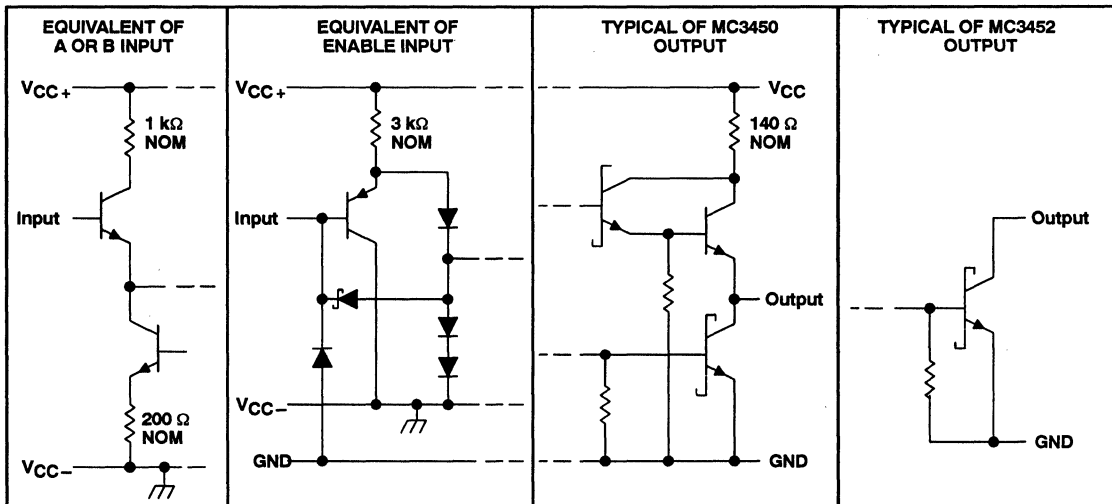
## logic diagram (positive logic)



# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS012B - D3006, FEBRUARY 1986 - REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	7 V
Supply voltage, $V_{CC-}$	-7 V
Differential input voltage (see Note 2)	$\pm 6$ V
Common-mode input voltage (see Note 3)	$\pm 5$ V
Enable input voltage	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential input voltage is measured at the noninverting input with respect to the corresponding inverting input.  
 3. Common-mode input voltage is the average of the voltages at the A and B inputs.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—

**TEXAS**  
**INSTRUMENTS**

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# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

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## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$	$T_A \geq 25^\circ\text{C}$	4.5	5	5.5	V
	$T_A < 25^\circ\text{C}$	4.75	5	5.5	
Supply voltage, $V_{CC-}$	$T_A \geq 25^\circ\text{C}$	-4.5	-5	-5.5	V
	$T_A < 25^\circ\text{C}$	-4.75	-5	-5.5	
High-level enable input voltage, $V_{IH}$		2			V
Low-level enable input voltage, $V_{IL}$		0.8			V
Low-level output current, $I_{OL}$		-16			mA
Differential input voltage, $V_{ID}$ (see Note 4)		-5†		5	V
Common-mode input voltage, $V_{IC}$ (see Note 4)		-3†		3	V
Input voltage range, any different input to GND		-5†		3	V
Operating free-air temperature, $T_A$	MC3450, MC3452	0		70	°C
	MC3550, MC3552	-55		125	

† The algebraic convention, in which the less positive (more negative) limit is designated minimum, is used in this data sheet for common-mode input voltage.

NOTE 4: The recommended combinations of input voltages fall within the shaded area of Figure 1.

### RECOMMENDED COMBINATIONS OF INPUT VOLTAGES

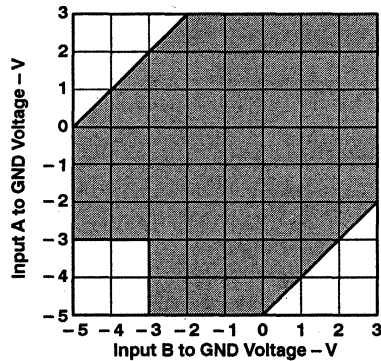


Figure 1

# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

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**electrical characteristics over recommended operating free-air temperature range,  $V_{CC\pm} = \text{MAX}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MC3450, MC3550		MC3452, MC3552		UNIT
			MIN	TYP†	MAX	MIN	
$V_{OH}$	High-level output voltage	$V_{CC\pm} = \pm 4.75 \text{ V}$ , $V_{ID} = 25 \text{ mV}$ , $\overline{EN}$ at 0.8 V, $I_{OH} = -400 \mu\text{A}$ $V_{IC} = -3 \text{ V to } 3 \text{ V}$	2.4				V
$V_{OL}$	Low-level output voltage	$V_{CC\pm} = \pm 4.75 \text{ V}$ , $V_{ID} = -25 \text{ mV}$ , $\overline{EN}$ at 2 V, $I_{OL} = 16 \text{ mA}$ , $V_{IC} = -3 \text{ V to } 3 \text{ V}$		0.5		0.5	V
$I_{OH}$	High-level output current	$V_{CC\pm} = \pm 4.75 \text{ V}$ , $V_{OH} = 5.25 \text{ V}$				250	$\mu\text{A}$
$I_{IH}$	High-level input current	A inputs	$V_{ID} = -2 \text{ V}$		30	75	$\mu\text{A}$
		B inputs	$V_{ID} = -2 \text{ V}$		30	75	$\mu\text{A}$
		$\overline{EN}$	$V_{IH} = 2.4 \text{ V}$		40		$\mu\text{A}$
		$\overline{EN}$	$V_{IH} = 5.25 \text{ V}$		1		$\text{mA}$
$I_{IL}$	Low-level input current	A inputs	$V_{ID} = 2 \text{ V}$		-10		$\mu\text{A}$
		B inputs	$V_{ID} = 2 \text{ V}$		-10		$\mu\text{A}$
		$\overline{EN}$	$V_{IL} = 0.4 \text{ V}$		-1.6		$\text{mA}$
$I_{OZ}$	High-impedance state output current	$V_O = 2.4 \text{ V}$			40		$\mu\text{A}$
		$V_O = 0.4 \text{ V}$			-40		
$I_{OS}$	Short-circuit output current‡	$V_{ID} = 25 \text{ mV}$ , $V_O = 0$ , $\overline{EN}$ at 0.8 V	-18	-70			$\text{mA}$
$I_{CCH+}$	Supply current from $V_{CC+}$ , outputs high	A inputs at GND, B inputs at 3 V, $\overline{EN}$ at 3 V			60		$\text{mA}$
$I_{CCH-}$	Supply current from $V_{CC-}$ , outputs high				-30		$\text{mA}$

† All typical values are at  $V_{CC+} = 5 \text{ V}$ ,  $V_{CC-} = -5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ Not more than one output should be shorted at a time.

## switching characteristics, $V_{CC\pm} = \pm 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MC3450, MC3550		MC3452, MC3552		UNIT
				MIN	TYP†	MAX	MIN	
$t_{PLH}$	A and B	Y	$C_L = 50 \text{ pF}$ , See Figure 2	17		25		ns
			$C_L = 15 \text{ pF}$ , See Figure 2			19		
$t_{PHL}$	A and B	Y	$C_L = 50 \text{ pF}$ , See Figure 2	17		25		ns
			$C_L = 15 \text{ pF}$ , See Figure 2			19		
$t_{PZH}$	$\overline{EN}$	Y	$C_L = 50 \text{ pF}$ , See Figure 2			21		ns
$t_{PZL}$	$\overline{EN}$	Y				27		
$t_{PHZ}$	$\overline{EN}$	Y	$C_L = 15 \text{ pF}$ , See Figure 3			18		ns
$t_{PLZ}$	$\overline{EN}$	Y				29		
$t_{PLH}$	$\overline{EN}$	Y	$C_L = 15 \text{ pF}$ , See Figure 4			25		ns
$t_{PHL}$	$\overline{EN}$	Y	$C_L = 15 \text{ pF}$ , See Figure 4			25		ns

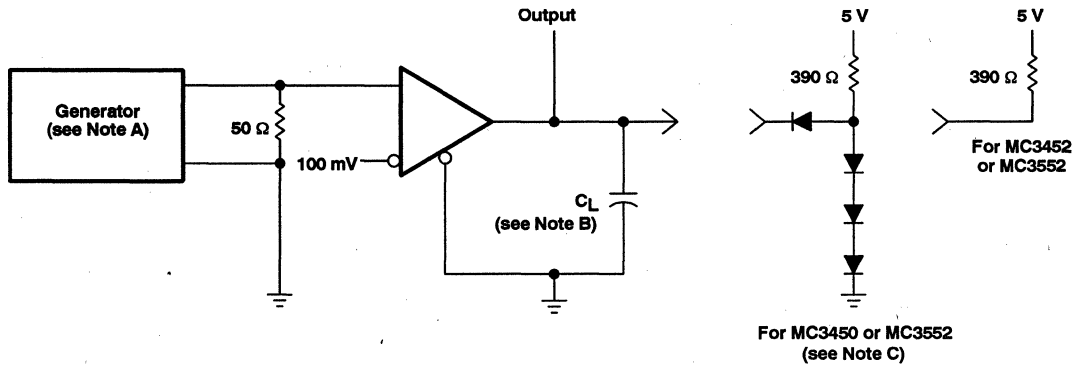
† All typical values are at  $V_{CC+} = 5 \text{ V}$ ,  $V_{CC-} = -5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .



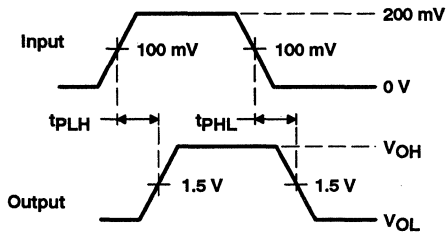
# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS012B - D3006, FEBRUARY 1986 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



### TEST CIRCUIT



### VOLTAGE WAVEFORMS

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle = 50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns.  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N916 or equivalent.

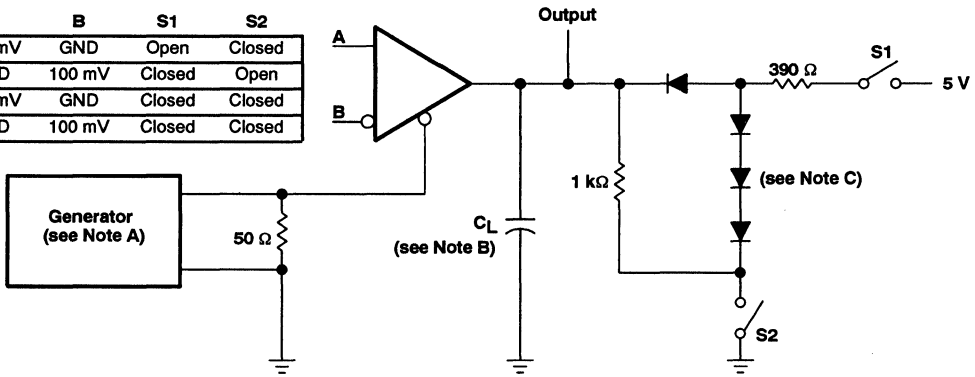
Figure 2. Test Circuit and Voltage Waveforms

# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

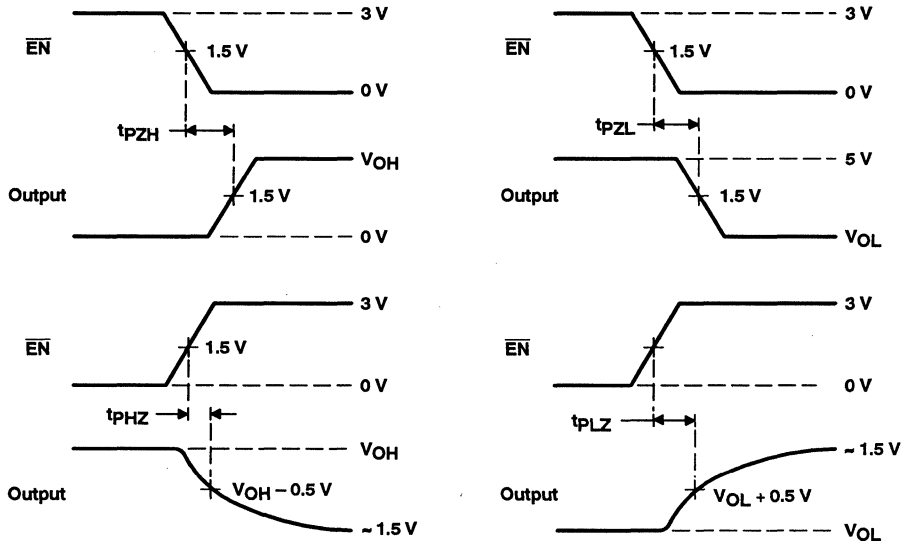
SLLS012B - D3006, FEBRUARY 1986 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

	A	B	S1	S2
$t_{pZH}$	100 mV	GND	Open	Closed
$t_{pZL}$	GND	100 mV	Closed	Open
$t_{pHZ}$	100 mV	GND	Closed	Closed
$t_{pLZ}$	GND	100 mV	Closed	Closed



TEST CIRCUIT



VOLTAGE WAVEFORMS

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle = 50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns.  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N916 or equivalent.

Figure 3. MC3450 and MC3550 Test Circuit and Voltage Waveforms

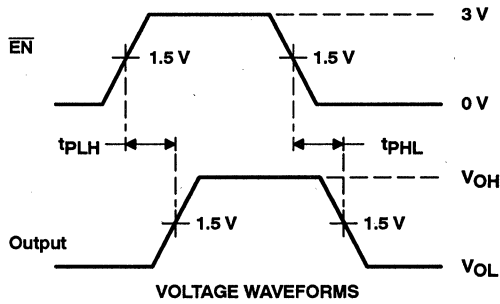
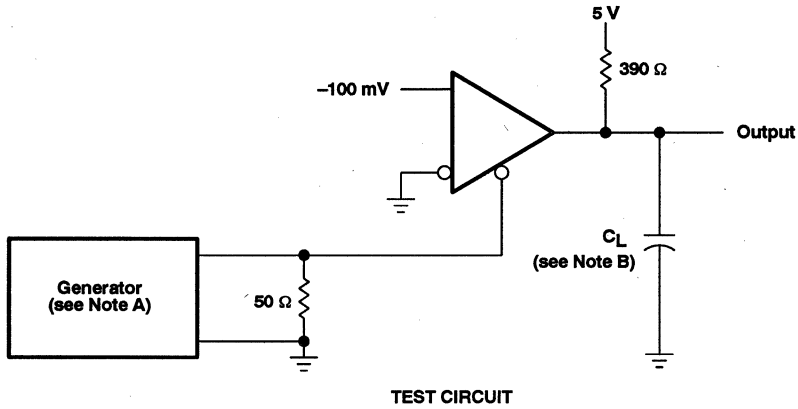
TEXAS  
INSTRUMENTS

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# MC3450, MC3452, MC3550, MC3552 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS012B – D3006, FEBRUARY 1986 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1\text{ MHz}$ , duty cycle = 50%,  $t_r \leq 6\text{ ns}$ ,  $t_f \leq 6\text{ ns}$ .  
B.  $C_L$  includes probe and jig capacitance.

**Figure 4. MC3452 and MC3552 Test Circuit and Voltage Waveforms**

# MC3453, MC3553 QUAD LINE DRIVERS WITH COMMON ENABLE

SLLS119A – D3000, FEBRUARY 1986 – REVISED FEBRUARY 1993

- Similar to a Dual Version of SN55/75110A Line Driver
- Improved Stability Over Supply Voltage and Temperature Ranges
- Constant-Current Outputs
- High Output Impedance
- High Common-Mode Output Voltage Range –3 V to 10 V
- Glitch-Free Power-Up/Power-Down Operation
- TTL-Input Compatibility
- Common-Enable Circuit
- Designed to Be Interchangeable With Motorola MC3453 and Military-Temperature-Range Version of MC3553

## description

The MC3453 and MC3553 feature four line drivers with a common-enable input. When the ENABLE input is high, a constant output current is switched between each pair of output terminals in response to the logic level at that channel input. When the ENABLE is low, all channel outputs are nonconductive (transistors biased to cutoff). This minimizes loading in party-line systems where a large number of drivers share the same line.

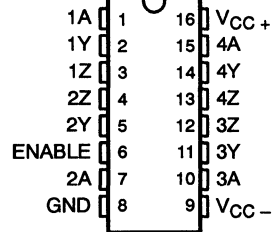
The driver outputs have a common-mode voltage range of –3 V to 10 V, allowing common-mode voltages on the line without affecting driver performance.

All outputs should be maintained within the recommended common-mode output voltage range to ensure that the channels do not interact with each other. To minimize power dissipation, all unused outputs should be grounded.

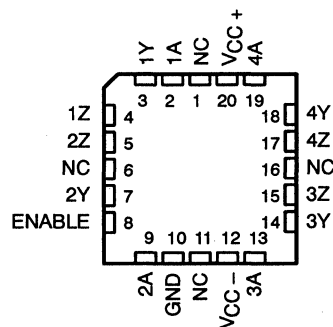
All inputs are diode clamped and are designed to satisfy TTL-system requirements. The inputs are tested at 2 V for high-logic-level input conditions and 0.8 V for low-logic-level input conditions. These tests ensure 400 mV of noise margin when interfaced with Series 54/74 TTL.

The MC3453 is characterized for operation from 0°C to 70°C. The MC3553 is characterized for operation over the full military temperature range of –55°C to 125°C.

D, J, OR N... PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC—No internal connection

FUNCTION TABLE

LOGIC INPUT	ENABLE INPUT	OUTPUT CURRENT	
		Z	Y
H	H	On	Off
L	H	Off	On
H	L	Off	Off
L	L	Off	Off

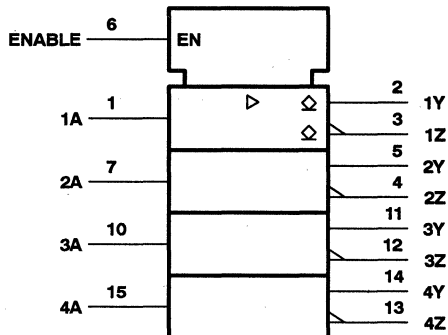
L = low logic level, H = high logic level



# MC3453, MC3553 QUAD LINE DRIVERS WITH COMMON ENABLE

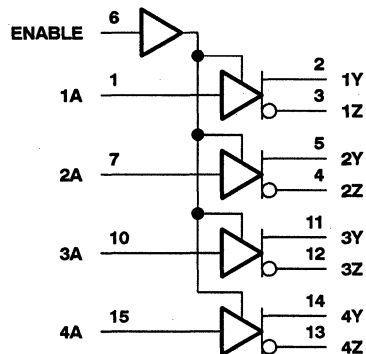
SLLS119A – D3000, FEBRUARY 1986 – REVISED FEBRUARY 1993

## logic symbol†

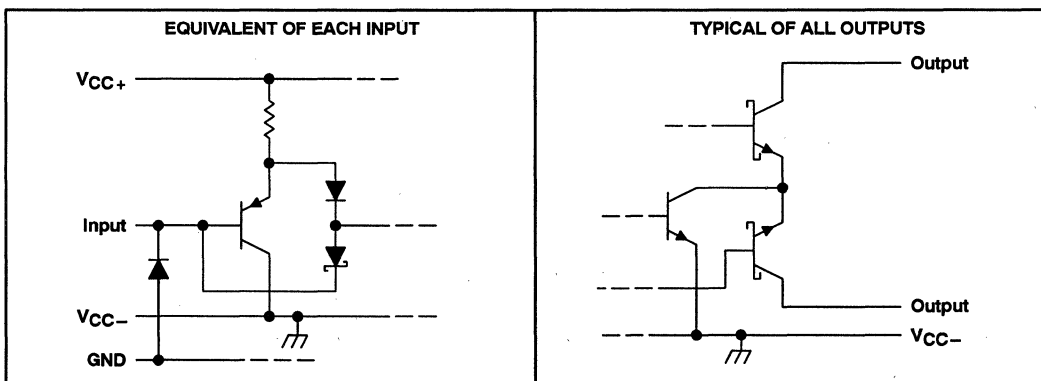


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	7 V
Supply voltage, $V_{CC-}$	-7 V
Input voltage (any input)	5.5 V
Output voltage range (any output)	-5 V to 12 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range:	
MC3453	0°C to 70°C
MC3553	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

NOTE 1: All voltage values are with respect to network ground terminal.

# MC3453, MC3553 QUAD LINE DRIVERS WITH COMMON ENABLE

SLLS119A - D3000, FEBRUARY 1986 - REVISED FEBRUARY 1993

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/ $^\circ\text{C}$	608 mW	—
FK	1375 mW	11.0 mW/ $^\circ\text{C}$	880 mW	275 mW
J	1375 mW	11.0 mW/ $^\circ\text{C}$	880 mW	275 mW
N	1150 mW	9.2 mW/ $^\circ\text{C}$	736 mW	—

### recommended operating conditions

		MIN	TYP	MAX	UNIT
Supply voltage, $V_{CC+}$	$T_A \geq 25^\circ\text{C}$	4.5	5	5.5	V
	$T_A < 25^\circ\text{C}$	4.75	5	5.5	
Supply voltage, $V_{CC-}$	$T_A \geq 25^\circ\text{C}$	-4.5	-5	-5.5	V
	$T_A < 25^\circ\text{C}$	-4.75	-5	-5.5	
High-level input voltage, $V_{IH}$		2		5.5	V
Low-level input voltage, $V_{IL}$		0		0.8	V
Common-mode output voltage range	$V_{OCR+}$	0		10	V
	$V_{OCR-}$	0		-3	V
Operating free-air temperature, $T_A$	MC3453	0		70	$^\circ\text{C}$
	MC3553	-55		125	

### electrical characteristics over recommended operating free-air temperature range, $V_{CC\pm} = \text{MAX}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†			MIN	TYP‡	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -12 \text{ mA}$			-0.9		-1.5	V
$I_{O(on)}$	On-state output current	$V_{CC+} = \text{MAX}$		$V_{CC-} = \text{MAX}$		11	15	mA
		$V_{CC+} = \text{MIN}$		$V_{CC-} = \text{MIN}$		6.5	11	
$I_{O(off)}$	Off-state output current	$V_{CC+} = \text{MIN}$		$V_{CC-} = \text{MIN}$ , $V_O = 10 \text{ V}$			100	$\mu\text{A}$
$I_{IH}$	High-level input current	$V_I = 2.4 \text{ V}$					40	$\mu\text{A}$
		$V_I = V_{CC+} \text{ max}$					1	
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$					-1.6	mA
$I_{CC+}$	Supply current from $V_{CC+}$	A inputs at 0.4 V	ENABLE at 2 V			33	50	mA
			ENABLE at 0.4 V			33	50	
$I_{CC-}$	Supply current from $V_{CC-}$	A inputs at 0.4 V	ENABLE at 2 V			-68	-90	mA
			ENABLE at 0.4 V			-31	-40	

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at  $V_{CC+} = 5 \text{ V}$ ,  $V_{CC-} = -5 \text{ V}$ , and  $T_A = 25^\circ\text{C}$ .

### switching characteristics, $V_{CC+} = 5 \text{ V}$ , $V_{CC-} = -5 \text{ V}$ , $R_L = 50 \Omega$ , $C_L = 40 \text{ pF}$ , $T_A = 25^\circ\text{C}$

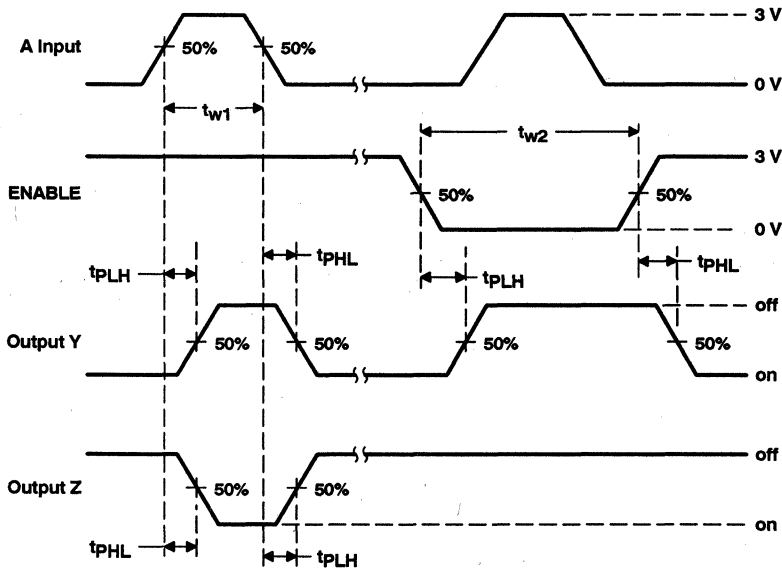
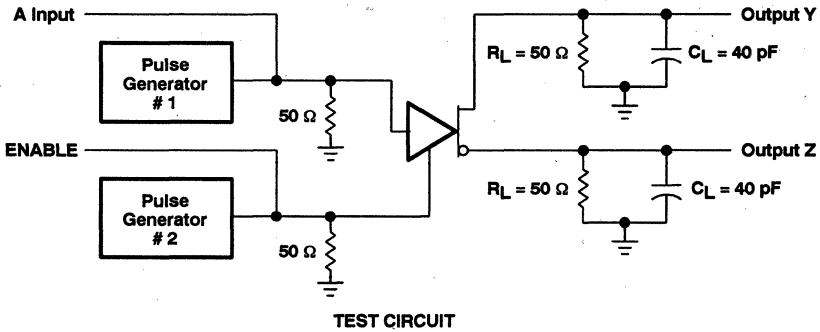
PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high level output	A	Y or Z	See Figure 1		9	15	ns
$t_{PHL}$	Propagation delay time, high-to-low level output	A	Y or Z			7	15	ns
$t_{PLH}$	Propagation delay time, low-to-high level output	ENABLE	Y or Z			14	25	ns
$t_{PHL}$	Propagation delay time, high-to-low level output	ENABLE	Y or Z			15	25	ns



# MC3453 QUADRUPLE LINE DRIVER WITH COMMON ENABLE

SLLS119 – D3000, FEBUARY 1986 – REVISED JULY 1990

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The pulse generators have the following characteristics:  $Z_O = 50 \Omega$ ,  $t_r = t_f = 10 \pm 5 \text{ ns}$ ,  $t_{w1} = 200 \text{ ns}$ ,  $\text{PRR} \leq 1 \text{ MHz}$ ,  $t_{w2} = 1 \mu\text{s}$ ,  $\text{PRR} \leq 500 \text{ kHz}$ .  
 B.  $C_L$  includes probe and jig capacitance.

**Figure 1. Test Circuit and Voltage Waveforms**

# MC3486

## QUAD DIFFERENTIAL LINE RECEIVER WITH 3-STATE OUTPUTS

SLLS097A - D2434, JUNE 1980 - REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A and RS423-A and Federal Standards 1020 and 1030
- 3-State, TTL-Compatible Outputs
- Fast Transition Times
- Operates From Single 5-V Supply
- Designed to Be Interchangeable With Motorola MC3486

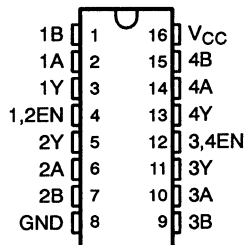
### description

The MC3486 is a monolithic quad differential line receiver designed to meet the specifications of EIA Standards RS-422-A and RS-423-A and Federal Standards 1020 and 1030. The MC3486 offers four independent differential-input line receivers that have TTL-compatible outputs. The outputs utilize 3-state circuitry to provide a high-impedance state at any output when the appropriate output enable is at a low logic level.

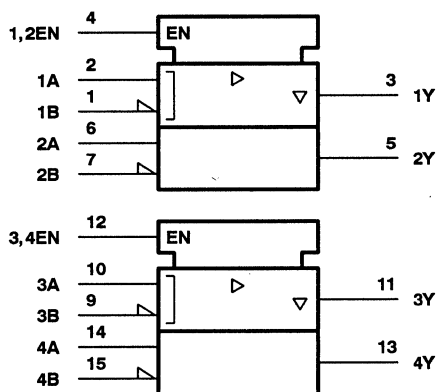
The MC3486 is designed for optimum performance when used with the MC3487 quadruple differential line driver. It is supplied in a 16-pin package and operates from a single 5-V supply.

The MC3486 is characterized for operation from 0°C to 70°C.

D OR N PACKAGE  
(TOP VIEW)



### logic symbol†



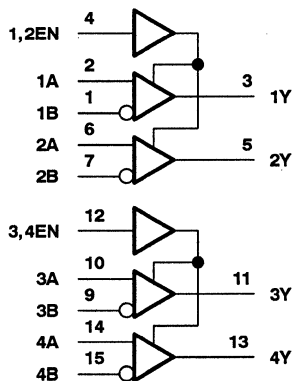
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

FUNCTION TABLE  
(each receiver)

DIFFERENTIAL INPUTS A-B	ENABLE	OUTPUT Y
$V_{ID} \approx 0.2 V$	H	H
$-0.2 V < V_{ID} < 0.2 V$	H	?
$V_{ID} \approx -0.2 V$	H	L
Irrelevant	L	Z

H = high level, L = low level, Z = high-impedance (off),  
? = indeterminate

### logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

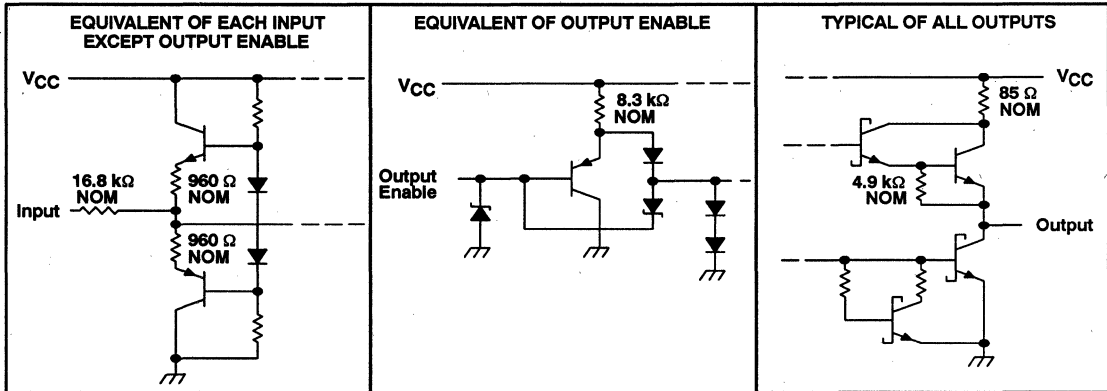
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# MC3486 QUAD DIFFERENTIAL LINE RECEIVER WITH 3-STATE OUTPUTS

SLLS097A—D2434, JUNE 1980—REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	8 V
Input voltage, A or B inputs	$\pm 15$ V
Differential input voltage (see Note 2)	$\pm 25$ V
Enable input voltage	8 V
Low-level output current	50 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential-input voltage, are with respect to network ground terminal.  
2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Differential input voltage, $V_{ID}$			$\pm 6$	V
High-level enable input voltage, $V_{IH}$	2			V
Low-level enable input voltage, $V_{IL}$			0.8	V
Operating free-air temperature, $T_A$	0		70	°C

TEXAS  
INSTRUMENTS

**MC3486**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS097A – D2434, JUNE 1980 – REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
$V_{T+}$	Differential-input high-threshold voltage	$V_O = 2.7\text{ V}$ , $I_O = -0.4\text{ mA}$		0.2	V
$V_{T-}$	Differential-input low-threshold voltage	$V_O = 0.5\text{ V}$ , $I_O = -8\text{ mA}$	-0.2†		V
$V_{IK}$	Enable-input clamp voltage	$I_I = -10\text{ mA}$		-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 0.4\text{ V}$ , $I_O = -0.4\text{ mA}$ , See Note 3 and Figure 1		2.7	V
$V_{OL}$	Low-level output voltage	$V_{ID} = -0.4\text{ V}$ , $I_O = 8\text{ mA}$ , See Note 3 and Figure 1		0.5	V
$I_{OZ}$	High-impedance-state output current	$V_{IL} = 0.8\text{ V}$ , $V_{ID} = -3\text{ V}$ , $V_O = 2.7\text{ V}$		40	$\mu\text{A}$
		$V_{IL} = 0.8\text{ V}$ , $V_{ID} = 3\text{ V}$ , $V_O = 0.5\text{ V}$		-40	
$I_{IB}$	Differential-input bias current	$V_{CC} = 0\text{ V}$ or $5.25\text{ V}$ , Other inputs at $0\text{ V}$	$V_I = -10\text{ V}$	-3.25	mA
			$V_I = -3\text{ V}$	-1.5	
			$V_I = 3\text{ V}$	1.5	
			$V_I = 10\text{ V}$	3.25	
$I_{IH}$	High-level enable input current	$V_I = 5.25\text{ V}$		100	$\mu\text{A}$
		$V_I = 2.7\text{ V}$		20	
$I_{IL}$	Low-level enable input current	$V_I = -0.5\text{ V}$		-100	$\mu\text{A}$
$I_{OS}$	Short-circuit output current	$V_{ID} = 3\text{ V}$ , $V_O = 0$ , See Note 4	-15	-100	mA
$I_{CC}$	Supply current	$V_{IL} = 0$		85	mA

† The algebraic convention, in which the least positive (most negative) limit is designated as minimum, is used in this data sheet for threshold voltages only.

NOTES: 3. Refer to EIA Standards RS-422-A and RS-423-A for exact conditions.

4. Only one output should be shorted at a time.

**switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 15\text{ pF}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PHL}$	Propagation delay time, high-to-low level output	See Figure 2		28	35	ns
$t_{PLH}$	Propagation delay time, low-to-high level output			27	30	ns
$t_{PZH}$	Output enable time to high level	See Figure 3		13	30	ns
$t_{PZL}$	Output enable time to low level			20	30	ns
$t_{PHZ}$	Output disable time from high level			26	35	ns
$t_{PLZ}$	Output disable time from low level			27	35	ns



**MC3486**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS097A - D2434, JUNE 1980 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**

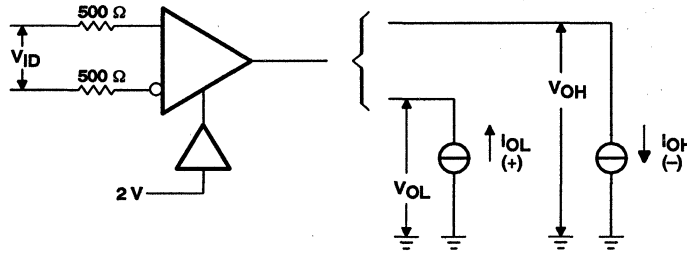


Figure 1.  $V_{OH}$ ,  $V_{OL}$

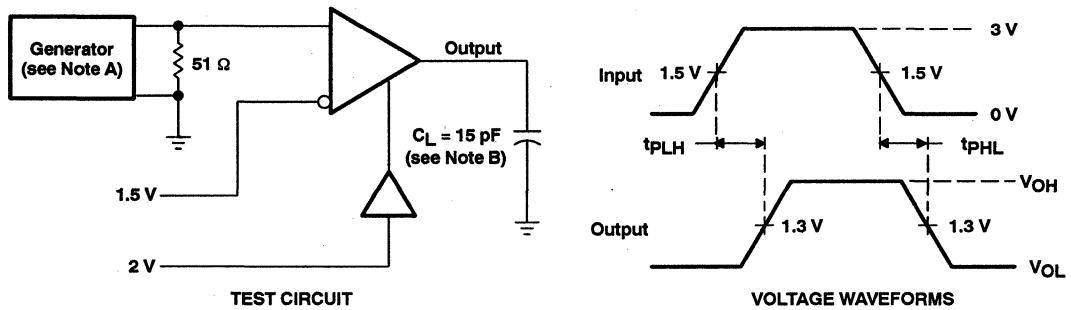


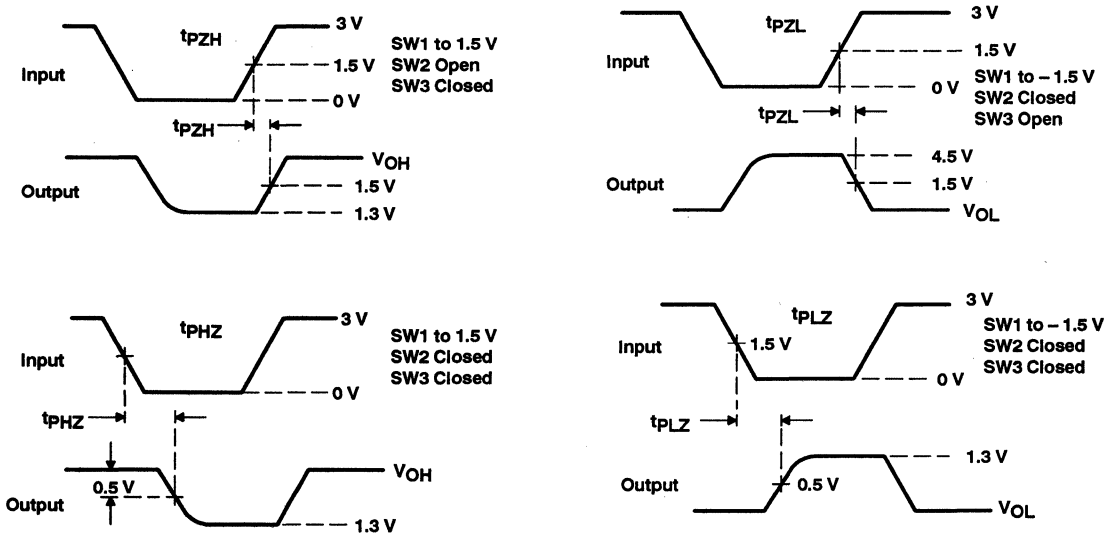
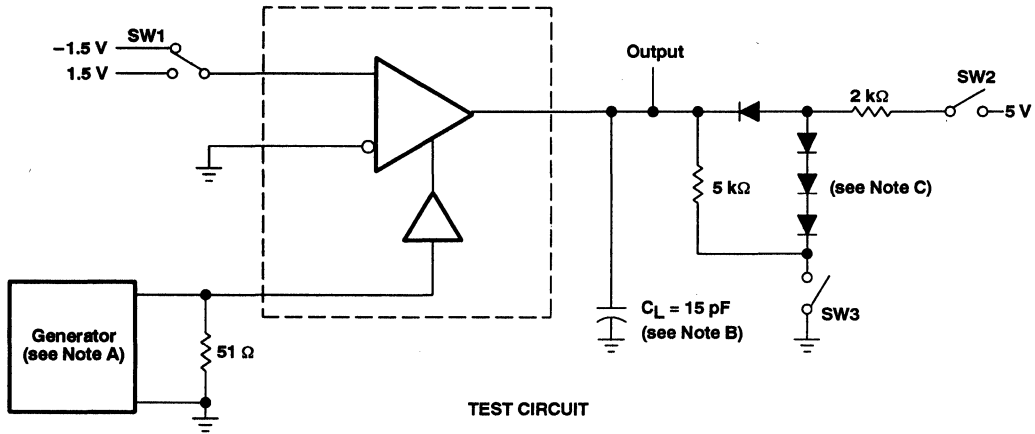
Figure 2. Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1 \text{ MHz}$ , duty cycle = 50%,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ .  
 B.  $C_L$  includes probe and stray capacitance.

**MC3486**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS097A - D2434, JUNE 1980 - REVISED FEBRUARY 1983

**PARAMETER MEASUREMENT INFORMATION**



**Figure 3. Test Circuit and Voltage Waveforms**

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle = 50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns.  
 B.  $C_L$  includes probe and stray capacitance.  
 C. All diodes are 1N916 or equivalent.



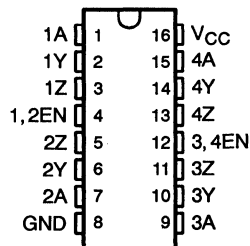


# MC3487 QUAD DIFFERENTIAL LINE DRIVER WITH 3-STATE OUTPUTS

SLLS098 - D2578, MAY 1980 - REVISED SEPTEMBER 1986

- Meets EIA Standard RS-422-A and Federal Standard 1020
- 3-State, TTL-Compatible Outputs
- Fast Transition Times
- High-Impedance Inputs
- Single 5-V Supply
- Power-Up and Power-Down Protection
- Designed to Be Interchangeable With Motorola MC3487

D, J, OR N PACKAGE  
(TOP VIEW)



## description

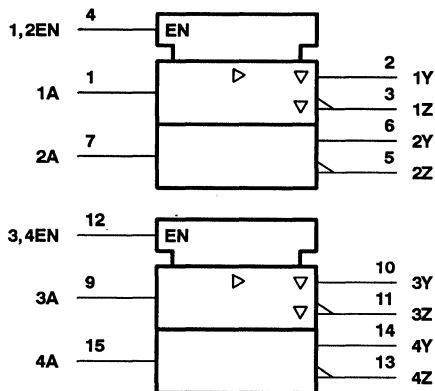
The MC3487 offers four independent differential line drivers designed to meet the specifications of EIA Standard RS-422-A and Federal Standard 1020. Each driver has a TTL-compatible input buffered to reduce current and minimize loading.

The driver outputs utilize 3-state circuitry to provide high-impedance states at any pair of differential outputs when the appropriate output enable is at a low logic level. Internal circuitry is provided to ensure a high-impedance state at the differential outputs during power-up and power-down transition times, provided the output enable is low. The outputs are capable of source or sink currents of 48 mA.

The MC3487 is designed for optimum performance when used with the MC3486 quadruple line receiver. It is supplied in a 16-pin dual-in-line package and operates from a single 5-V supply.

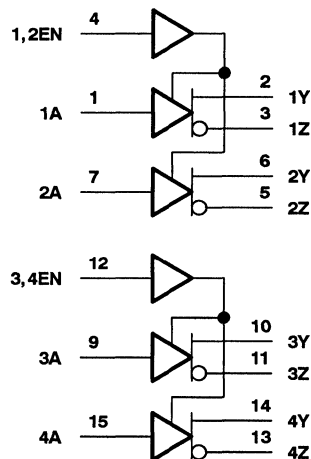
The MC3487 is characterized for operation from 0°C to 70°C.

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



# MC3487 QUAD DIFFERENTIAL LINE DRIVER WITH 3-STATE OUTPUTS

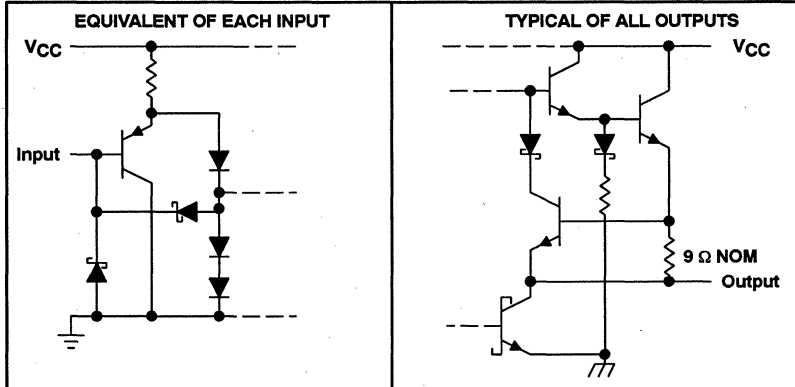
SLLS098 - D2578, MAY 1980 - REVISED SEPTEMBER 1986

FUNCTION TABLE  
(each driver)

INPUT	OUTPUT ENABLE	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	High-impedance	High-impedance

H = TTL high level, L = TTL low level, X = irrelevant

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	8 V
Input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C

NOTE 1: All voltage values, except differential output voltage,  $V_{OD}$ , are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
J	1025 mW	8.2 mW/°C	656 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
Operating free-air temperature, $T_A$	0		70	°C

TEXAS  
INSTRUMENTS

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**MC3487**  
**QUAD DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS098 – D2578, MAY 1980 – REVISED SEPTEMBER 1986

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS			MIN	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$				-1.5	V
$V_{OH}$	High-level output voltage	$V_{IL} = 0.8 \text{ V}$ ,	$V_{IH} = 2 \text{ V}$ ,	$I_{OH} = -20 \text{ mA}$	2.5		V
$V_{OL}$	Low-level output voltage	$V_{IL} = 0.8 \text{ V}$ ,	$V_{IH} = 2 \text{ V}$ ,	$I_{OL} = 48 \text{ mA}$		0.5	V
$ V_{OD} $	Differential output voltage	$R_L = 100 \Omega$ ,	See Figure 1		2		
$\Delta V_{OD} $	Change in magnitude of differential output voltage†	$R_L = 100 \Omega$ ,	See Figure 1			$\pm 0.4$	V
$V_{OC}$	Common-mode output voltage‡	$R_L = 100 \Omega$ ,	See Figure 1			3	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage†	$R_L = 100 \Omega$ ,	See Figure 1			$\pm 0.4$	V
$I_O$	Output current with power off	$V_{CC} = 0$	$V_O = 6 \text{ V}$		100		$\mu\text{A}$
			$V_O = -0.25 \text{ V}$		-100		
$I_{OZ}$	High-impedance-state output current	Output enables at 0.8 V	$V_O = 2.7 \text{ V}$		100		$\mu\text{A}$
			$V_O = 0.5 \text{ V}$		-100		
$I_I$	Input current at maximum input voltage	$V_I = 5.5 \text{ V}$				100	$\mu\text{A}$
$I_{IH}$	High-level input current	$V_I = 2.7 \text{ V}$				50	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0.5 \text{ V}$				-400	$\mu\text{A}$
$I_{OS}$	Short-circuit output current§	$V_I = 2 \text{ V}$			-40	-140	mA
$I_{CC}$	Supply current (all drivers)	Outputs disabled				105	mA
		Outputs enabled, No load				85	

†  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

‡ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

§ Only one output at a time should be shorted, and duration of the short circuit should not exceed one second.

**switching characteristics over recommended range of operating free-air temperature,  $V_{CC} = 5 \text{ V}$**

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high level output	$C_L = 15 \text{ pF}$ ,	See Figure 2		20	ns
$t_{PHL}$	Propagation delay time, high-to-low level output				20	ns
	Skew				6	ns
$t_D$	Differential-output transition time	$C_L = 15 \text{ pF}$ ,	See Figure 3		20	ns
$t_{PZH}$	Output enable time to high level	$C_L = 50 \text{ pF}$ ,	See Figure 4		30	ns
$t_{PZL}$	Output enable time to low level				30	ns
$t_{PHZ}$	Output disable time from high level				25	ns
$t_{PLZ}$	Output disable time from low level				30	ns

**PARAMETER MEASUREMENT INFORMATION**

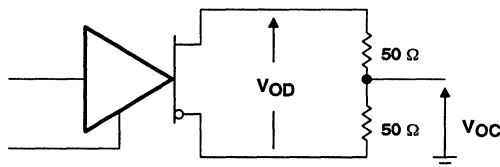
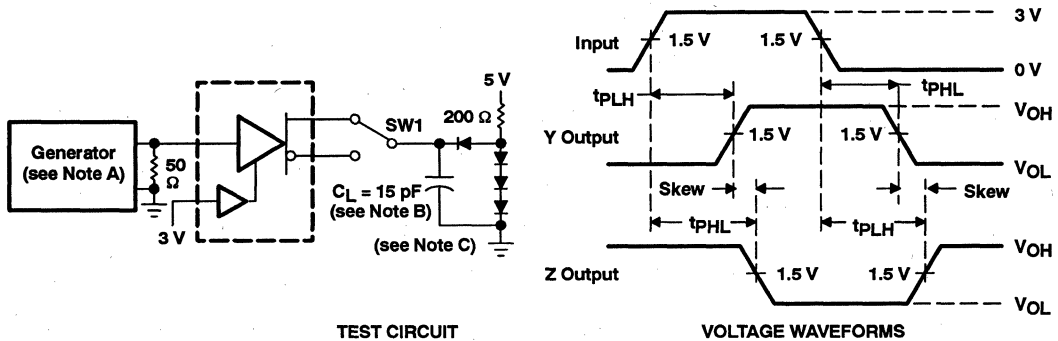


Figure 1. Differential and Common-Mode Output Voltages

**MC3487**  
**QUAD DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS098 - D2578, MAY 1980 - REVISED SEPTEMBER 1986

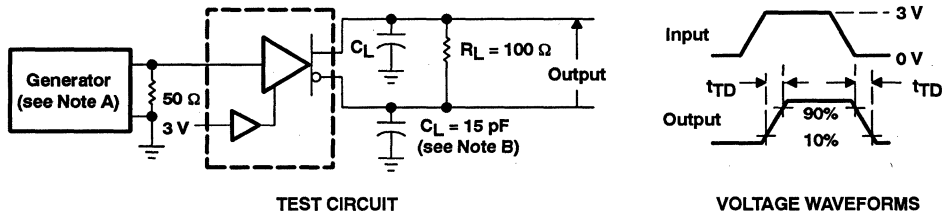
**PARAMETER MEASUREMENT INFORMATION**



TEST CIRCUIT

VOLTAGE WAVEFORMS

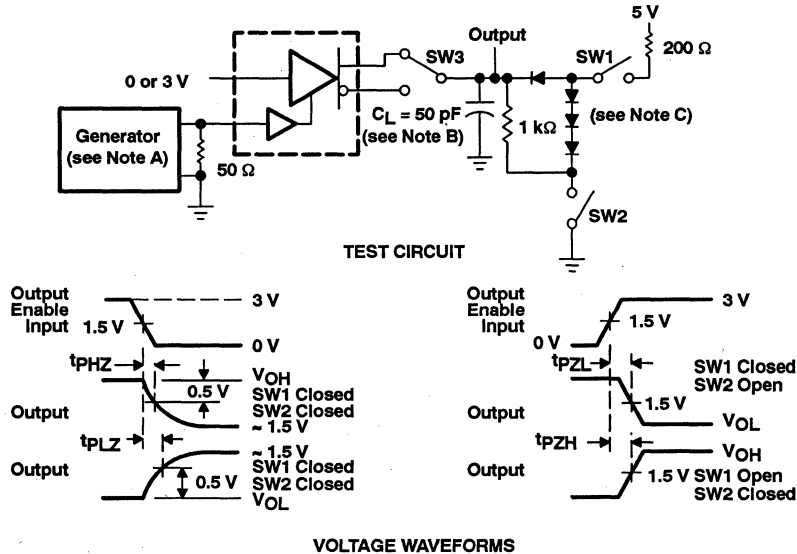
**Figure 2. Test Circuit and Voltage Waveforms**



TEST CIRCUIT

VOLTAGE WAVEFORMS

**Figure 3. Test Circuit and Voltage Waveforms**



TEST CIRCUIT

VOLTAGE WAVEFORMS

**Figure 4. Driver Test Circuit and Voltage Waveforms**

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5$  ns,  $t_f \leq 5$  ns, PRR  $\leq 1$  MHz, duty cycle = 50%,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and stray capacitance.  
 C. All diodes are 1N916 or 1N3064.

# SN75ALS053 QUAD FUTUREBUS TRANSCEIVER

SLLS034B - D3077, JANUARY 1988 - REVISED SEPTEMBER 1989

- High-Speed Quad Transceiver
- Fully Compatible With IEEE Std. 896.1 - 1987 Futurebus Requirements
- Drives Load Impedances as Low as 10  $\Omega$
- High-Speed Advanced Low-Power Schottky Circuits
- Low Power Dissipation . . . 81 mW Max per Channel
- High-Impedance PNP Inputs
- BTL™ Logic Level 1-V Bus Swing Reduces Power Consumption
- Low Bus-Port Capacitance
- Power-Up/Power-Down Protection (Glitch Free)
- Open-Collector Driver Outputs Allows Wired-OR Connections
- Multiple Bus Channel Ground Returns to Reduce Channel Noise Interference
- Designed to Be a Faster, Lower Power Functional Equivalent of National DS3893

## description

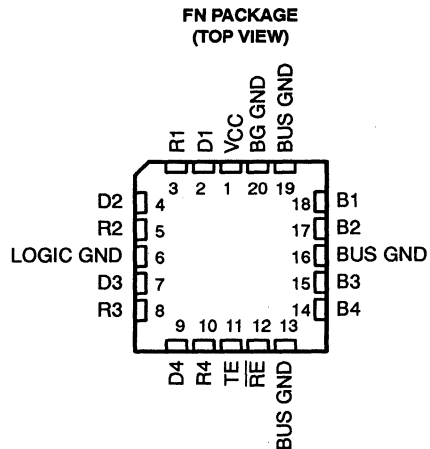
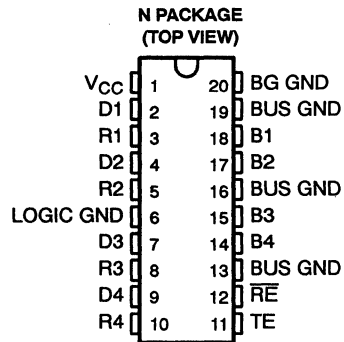
The SN75ALS053 is a four-channel, monolithic, high-speed, advanced low-power Schottky device designed for two-way data communication in a densely populated backplane. The SN75ALS053 has independent driver input (Dn) and receiver output (Rn) pins and separate driver and receiver disables. This transceiver is designed for use in high-speed bus systems and is similar to the SN75ALS057 transceiver except that the trapezoidal feature has been eliminated to speed up the propagation delays.

These transceivers feature open-collector driver outputs, each with a series Schottky diode to reduce capacitive loading to the bus. By using a 2-V pullup on the bus, the output signal swing will be approximately 1 V, which reduces the power necessary to drive the bus load capacitance. The driver outputs are capable of driving an equivalent dc load of as low as 10  $\Omega$ .

The receivers have a precision threshold set by an internal bandgap reference to give accurate input thresholds over  $V_{CC}$  and temperature variations.

These transceivers are compatible with Backplane Transceiver Logic (BTL™) technology at significantly reduced power dissipation per channel.

The SN75ALS053 is characterized for operation from 0° to 70°C.



BTL is a trademark of National Semiconductor Corporation.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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# SN75ALS053 QUAD FUTUREBUS TRANSCEIVER

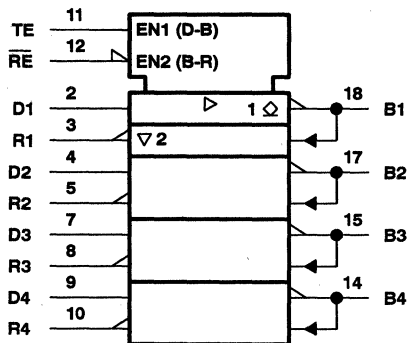
SLLS034B - D3077, JANUARY 1988 - REVISED SEPTEMBER 1989

**FUNCTION TABLE  
TRANSMIT/RECEIVE**

CONTROLS		CHANNELS	
TE	RE	D→B	B→R
L	L	D	R
L	H	D	D
H	L	T	R
H	H	T	D

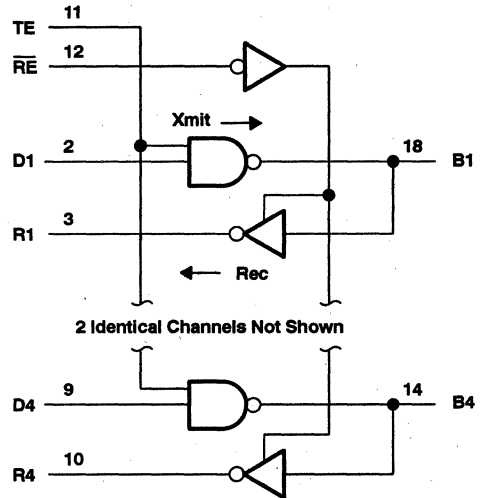
H = high level, L = low level, R = receive,  
T = transmit, D = disable  
Direction of data transmission is from Dn to  
Bn, direction of data reception is from Bn to Rn.

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

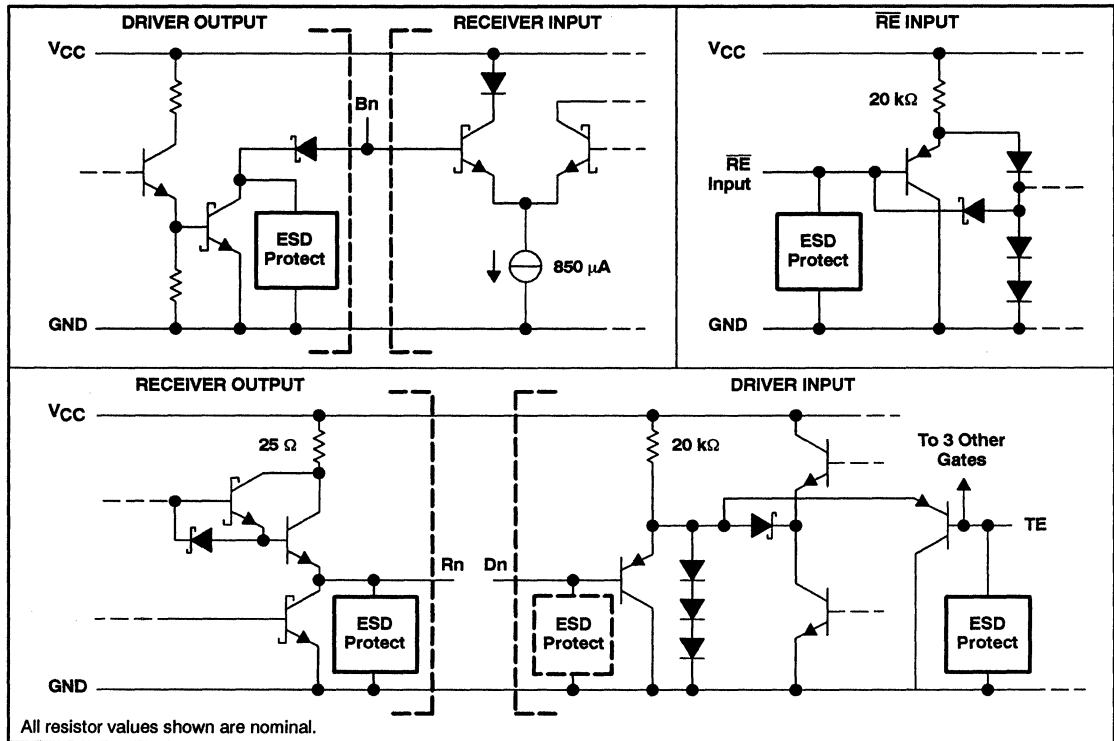
## logic diagram (positive logic)



# SN75ALS053 QUAD FUTUREBUS TRANSCEIVER

SLLS034B – D3077, JANUARY 1988 – REVISED SEPTEMBER 1989

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	6 V
Control input voltage	5.5 V
Driver input voltage	5.5 V
Driver output voltage	2.5 V
Receiver input voltage	2.5 V
Receiver output voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 10 seconds: FN package	260°C
Lead temperature 1,6 mm (1/16 in) from case for 10 seconds: N package	260°C

NOTE 1: Voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
FN	1400 MW	11.2 MW/°C	896 MW
N	1150 MW	9.2 MW/°C	736 mW

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# SN75ALS053 QUAD FUTUREBUS TRANSCEIVER

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## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level driver and control input voltage, $V_{IH}$	2			V
Low-level driver and control input voltage, $V_{IL}$			0.8	V
Bus termination voltage	1.9		2.1	V
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{IK}$	Input clamp voltage at Dn, DE, or $\overline{RE}$	$I_I = -18$ mA			-1.5	V
$V_T$	Receiver input threshold voltage at Bn		1.426		1.674	V
$V_{OH}$	High-level output voltage at Rn	Bn at 1.2 V, $\overline{RE}$ at 0.8 V, $I_{OH} = -1$ mA	2.5			V
$V_{OL}$	Low-level output voltage	Rn			0.5	V
		Bn	Dn at 2.4 V, $V_L = 2$ V, TE at 2.4 V, $R_L = 10 \Omega$ See Figure 1,	0.75	1.2	
$I_{IH}$	High-level input current	Dn, TE or $\overline{RE}$	$V_I = V_{CC}$		40	$\mu$ A
		Bn	$V_I = 2$ V, Dn at 0.8 V, $V_{CC} = 0$ or 5.25 V, TE at 0.8 V		100	
$I_{IL}$	Low-level input current at Dn, TE or $\overline{RE}$	$V_I = 0.4$ V			-400	$\mu$ A
$I_{OS}$	Short-circuit output at Rn	Rn at 0 V, Bn at 1.2 V, $\overline{RE}$ at 0.8 V	-70		-200	mA
$I_{CC}$	Supply current				65	mA
$C_{O(B)}$	Driver output capacitance	$V_{CC} = 5$ V, $T_A = 25^\circ$ C		6.5		pF

# SN75ALS053 QUAD FUTUREBUS TRANSCEIVER

SLLS034B – D3077, JANUARY 1988 – REVISED SEPTEMBER 1989

## switching characteristics over recommended ranges of supply voltage and operating free-air temperature

### driver

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time low-to-high-level output	Dn	Bn	TE at 3 V, V <sub>L</sub> = 2 V, See Figure 2	2	7	ns
t <sub>PHL</sub>	Propagation delay time high-to-low-level output				2	7	
t <sub>PLH</sub>	Propagation delay time low-to-high-level output	Dn	Bn	Dn at 3 V, V <sub>L</sub> = 2 V, See Figure 2	2	7	ns
t <sub>PHL</sub>	Propagation delay time high-to-low-level output				2	7	
t <sub>TLH</sub>	Transition time, low-to-high-level output	Dn	Bn	TE at 3 V, V <sub>L</sub> = 2 V, See Figure 2	0.5	5	ns
t <sub>THL</sub>	Transition time, high-to-low-level output				0.5	5	
Skew between driver channels †		Dn	Bn	TE at 3 V, V <sub>L</sub> = 2 V	1		ns

### receiver

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Bn	Rn	$\overline{RE}$ at 0.3 V, TE at 0.3 V	2	8	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output				2	8	
t <sub>PLZ</sub>	Output disable time from low level	$\overline{RE}$	Rn	Bn at 2 V, TE at 0.3 V, V <sub>L</sub> = 5 V, C <sub>L</sub> = 5 pF, R <sub>L1</sub> = 500 Ω, See Figure 4	6		ns
t <sub>PZL</sub>	Output enable time to low level	$\overline{RE}$	Rn	Bn at 2 V, TE at 0.3 V, V <sub>L</sub> = 5 V, C <sub>L</sub> = 5 pF, R <sub>L1</sub> = 500 Ω, See Figure 4	12		ns
t <sub>PHZ</sub>	Output disable time from high level	$\overline{RE}$	Rn	Bn at 1 V, TE at 0.3 V, V <sub>L</sub> = 0, C <sub>L</sub> = 5 pF, R <sub>L1</sub> = 500 Ω, See Figure 4	6		ns
t <sub>PZH</sub>	Output enable time to high level	$\overline{RE}$	Rn	Bn at 1 V, TE at 0.3 V, V <sub>L</sub> = 0, C <sub>L</sub> = 5 pF, R <sub>L1</sub> = 500 Ω, See Figure 4	12		ns
Skew between receiver channels †		Bn	Rn	$\overline{RE}$ at 0.3 V, TE at 0.3 V	1		ns

† Skew is the difference between the propagation delay time (t<sub>PLH</sub> or t<sub>PHL</sub>) of one receiver channel and that same propagation delay time of any other receiver channel. It applies for both t<sub>PLH</sub> and t<sub>PHL</sub>.



# SN75ALS053 QUAD FUTUREBUS TRANSCEIVER

SLLS034B - D3077, JANUARY 1988 - REVISED SEPTEMBER 1989

## PARAMETER MEASUREMENT INFORMATION

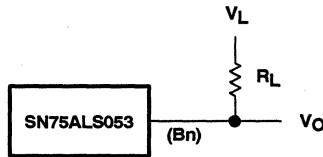
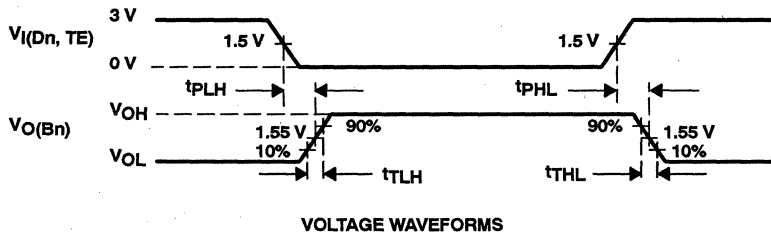
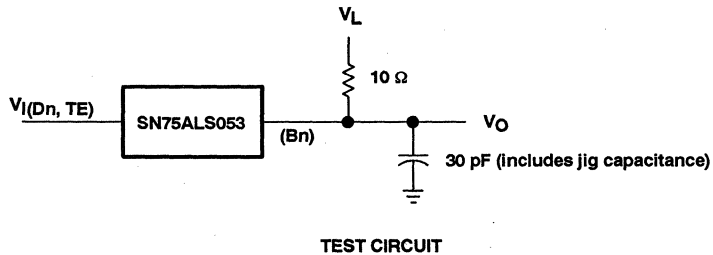


Figure 1. Driver Low-Level-Output-Voltage Test Circuit



NOTE:  $t_r = t_f \leq 5$  ns from 10% to 90%

Figure 2. Driver Test Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION

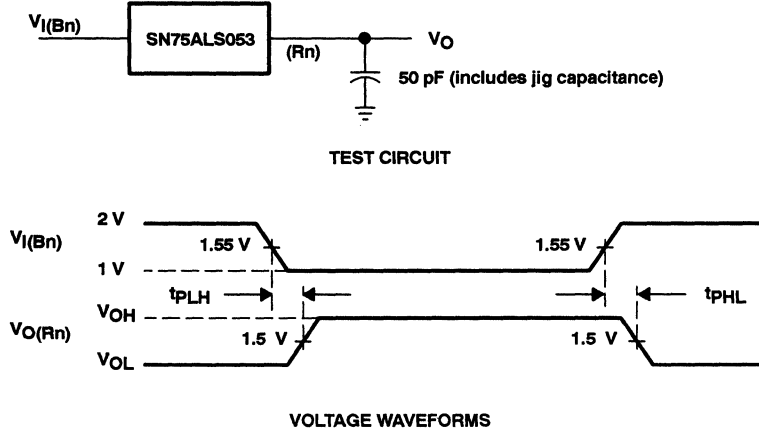


Figure 3. Receiver Test Circuit and Voltage Waveforms

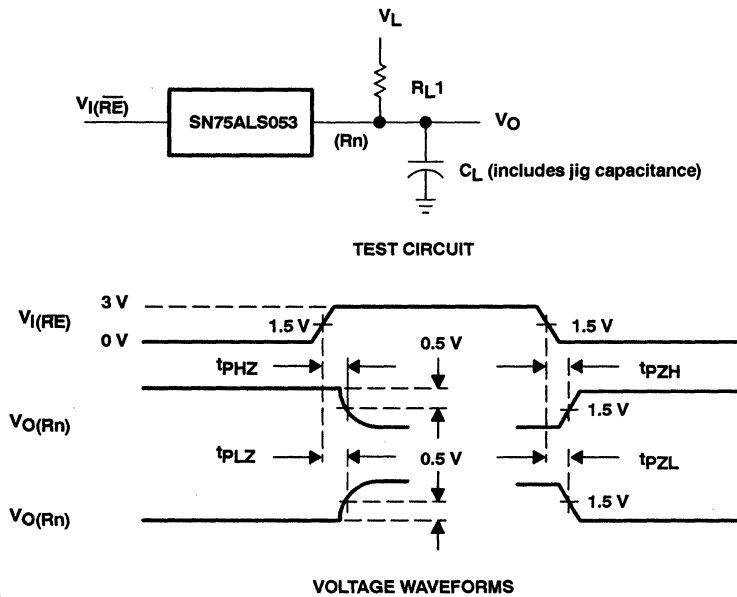


Figure 4. Test Circuit and Voltage Waveforms From  $\overline{RE}$  to Rn



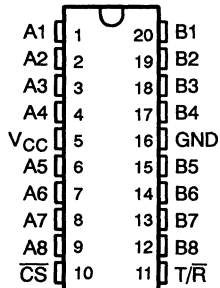
# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D – D3025, AUGUST 1987 – REVISED MARCH 1993

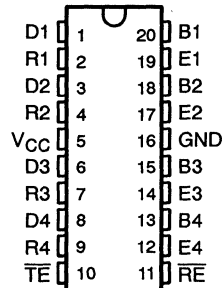
## SUITABLE FOR IEEE STANDARD 896 APPLICATIONS†

- SN55ALS056 and SN75ALS056 are Octal Transceivers
- SN55ALS057 and SN75ALS057 are Quad Transceivers
- High-Speed Advanced Low-Power Schottky Circuitry
- Low Power Dissipation:  
SN55' Devices ... 60 mW/Channel Max  
SN75' Devices ... 52.5 mW/Channel Max
- High-Impedance PNP Inputs
- Logic Level 1-V Bus Swing Reduces Power Consumption
- Trapezoidal Bus Output Waveform Reduces Noise Coupling to Adjacent Lines
- Power-Up/Power-Down Protection (Glitch Free)
- Open-Collector Driver Outputs Allow Wired-OR Connections
- Designed to Be a Faster, Lower Power Functional Equivalent of National DS3896, DS3897

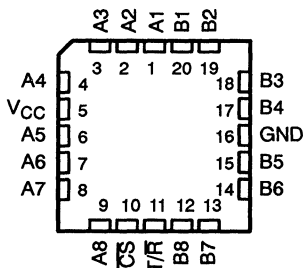
SN55ALS056 ... J OR W PACKAGE  
SN75ALS056 ... DW OR N PACKAGE  
(TOP VIEW)



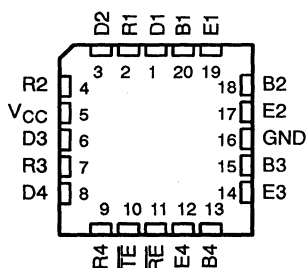
SN55ALS057 ... J OR W PACKAGE  
SN75ALS057 ... DW OR N PACKAGE  
(TOP VIEW)



SN55ALS056 ... FK PACKAGE  
(TOP VIEW)



SM55ALS057 ... FK PACKAGE  
(TOP VIEW)



† The transceivers are suitable for IEEE Standard 896 applications to the extent of the operating conditions and characteristics specified in this data sheet. Certain limits contained in the IEEE specification are not met or cannot be tested over the entire military temperature range.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

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## description

The SN55ALS056 and SN75ALS056 are 8-channel, monolithic, high-speed, advanced low-power Schottky devices designed for 2-way data communication in a densely populated backplane. The SN55ALS057 and SN75ALS057 are 4-channel versions with independent driver input (Dn) and receiver output (Rn) pins and a separate driver disable for each driver (En).

These transceivers feature open-collector driver outputs with a series Schottky diode to reduce capacitive loading to the bus. By using a 2-V pullup termination on the bus, the output signal swing will be approximately 1 V, which reduces the power necessary to drive the bus load capacitance. The driver outputs generate trapezoidal waveforms that reduce crosstalk between channels. The drivers are capable of driving an equivalent dc load as low as 18.5  $\Omega$ . The receivers have internal low-pass filters to further improve noise immunity.

The SN55ALS056 and SN55ALS057 are characterized over the full military operating range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75ALS056 and SN75ALS057 are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

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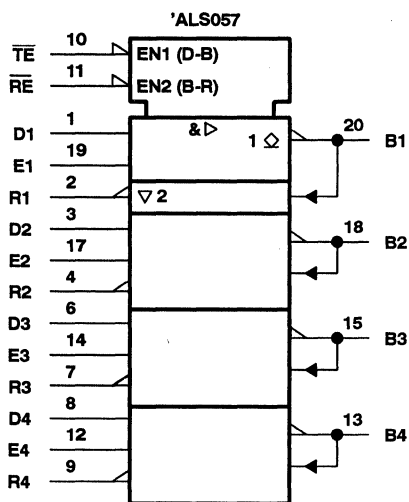
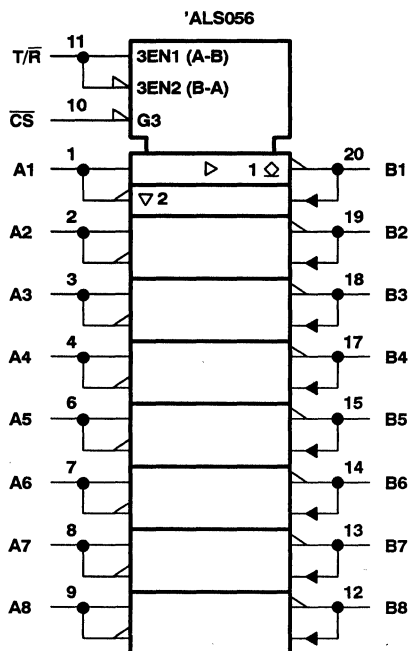
TEXAS  
INSTRUMENTS



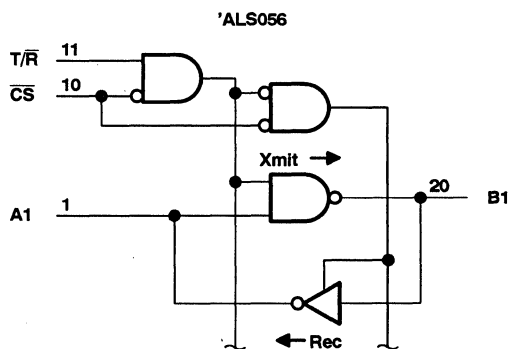
# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

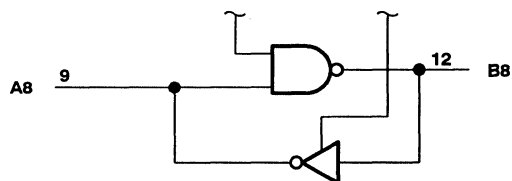
## logic symbols†



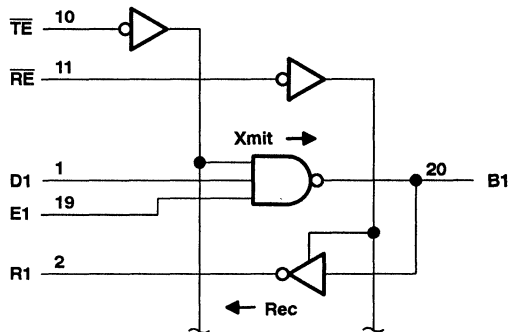
## logic diagrams (positive logic)



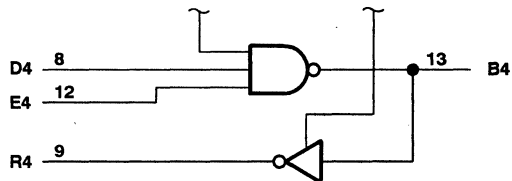
6 Identical Channels Not Shown



'ALS057



2 Identical Channels Not Shown



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.



# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

## Function Tables

'ALS056  
TRANSMIT/RECEIVE

CONTROLS		CHANNELS
CS	T/R	A ↔ B
L	H	T (A → B)
L	L	R (B → A)
H	X	D

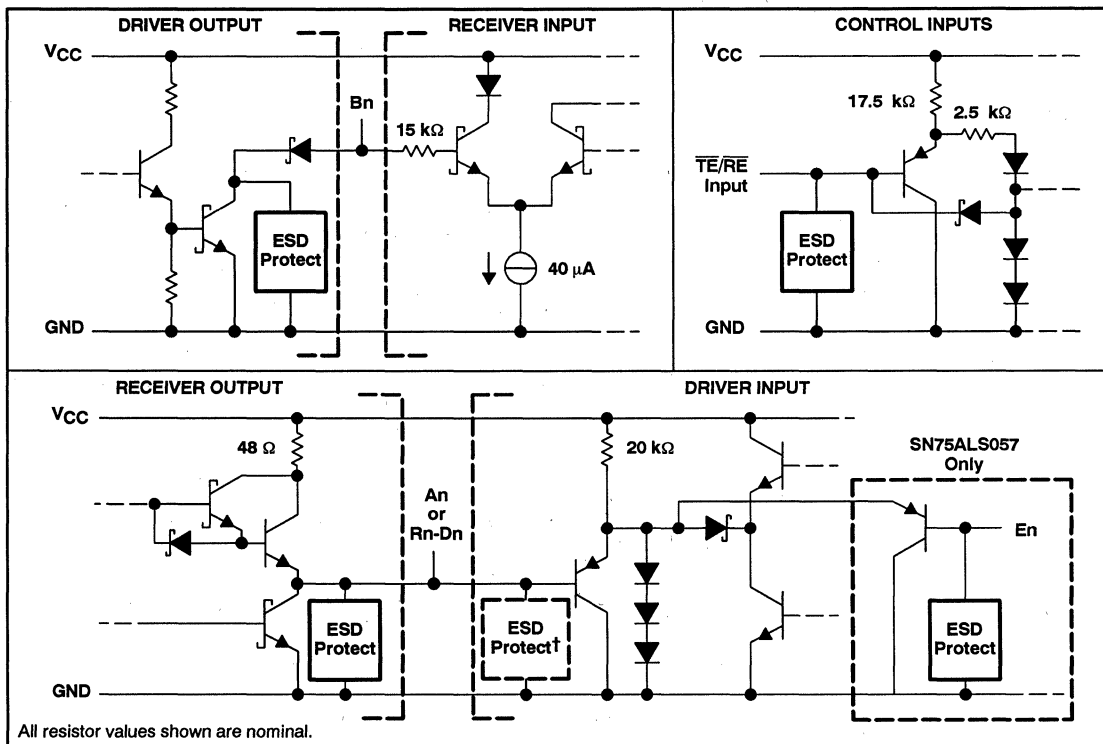
'ALS057  
TRANSMIT/RECEIVE

CONTROLS			CHANNELS	
TE	RE	En	D → B	B → R
L	L	L	D	R
L	L	H	T	R
L	H	L	D	D
L	H	H	T	D
H	L	X	D	R
H	H	X	D	D

H = high level, L = low level, R = receive, T = transmit, D = disable, X = irrelevant

Direction of data transmission is from An to Bn for the 'ALS056 and from Dn to Bn for the 'ALS057. Direction of data reception is from Bn to An for the 'ALS056 and from Bn to Rn for the 'ALS057. Data transfer is inverting in both directions.

## schematics of inputs and outputs



† Additional ESD protection is on the 'ALS057, which has separate receiver output and driver input pins.

# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D – D3025, AUGUST 1987 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	6 V
Control input voltage	5.5 V
Driver input voltage	5.5 V
Driver output voltage	2.5 V
Receiver input voltage	2.5 V
Receiver output voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN55ALS05_	-55°C to 125°C
SN75ALS05_	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	300 °C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: DW or N package	260 °C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	300 °C

NOTE 1: Voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
DW	1025 mW	8.2 mW/°C	656 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—
J	1375 mW	11.0 mW/°C	880 mW	275 mW
W	1000 mW	8.0 mW/°C	640 mW	200 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	SN55ALS05_	4.5	5	5.5	V
	SN75ALS05_	4.75	5	5.25	
High-level driver and control input voltage, $V_{IH}$		2			V
Low-level driver and control input voltage, $V_{IL}$		0.8			V
Bus termination voltage		1.9	2.1		V
Operating free-air temperature, $T_A$	SN55ALS05_	-55	125		°C
	SN75ALS05_	0	70		



# SN55ALS056, SN75ALS056 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

## SN55ALS056 electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage at An, T/R, or $\overline{CS}$	$V_{CC} = 4.5\text{ V}$ ,	$I_I = -18\text{ mA}$			-1.5	V
$V_T$	Receiver input threshold voltage at Bn	$V_{CC} = 5\text{ V}$ ,	$T_A = 25^\circ\text{C}$	1.45		1.65	V
		$V_{CC} = 5\text{ V}$ ,	$T_A = -55^\circ\text{C to } 125^\circ\text{C}$	1.4		1.7	
$V_{OH}$	High-level output voltage at An	$V_{CC} = 4.5\text{ V}$ , $\overline{CS}$ at 0.8 V, $I_{OH} = -400\ \mu\text{A}$	Bn at 1.2 V, T/R at 0.8 V,	2.4			V
$V_{OL}$	Low-level output voltage	An	$V_{CC} = 4.5\text{ V}$ , $\overline{CS}$ at 0.8 V, $I_{OL} = 16\text{ mA}$	Bn at 2 V, T/R at 0.8 V,		0.5	V
		Bn	$V_{CC} = 4.5\text{ V}$ , $\overline{CS}$ at 0.8 V, See Figure 1	An at 2 V, T/R at 2 V,	0.75	1.2	
$I_{IH}$	High-level input current	An, T/R or $\overline{CS}$	$V_I = V_{CC} = 5.5\text{ V}$			40	$\mu\text{A}$
		Bn	$V_{CC} = 5.5\text{ V}$ , An at 0.8 V,	$V_I = 2\text{ V}$ , T/R at 0.8 V		100	
$I_{IL}$	Low level input current at An, T/R, or $\overline{CS}$	$V_{CC} = 5.5\text{ V}$ ,	$V_I = 0.4\text{ V}$			-400	$\mu\text{A}$
$I_{OS}$	Short-circuit output current at An	$V_{CC} = 5.5\text{ V}$ , Bn at 1.2 V, T/R at 0.8 V	An at 0 V, $\overline{CS}$ at 0.8 V,	-35		-125	mA
$I_{CC}$	Supply current	$V_{CC} = 5.5\text{ V}$				85	mA
$C_{O(B)}$	Driver output capacitance					4.5	pF

## SN75ALS056 electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage at An, T/R, or $\overline{CS}$	$I_I = -18\text{ mA}$				-1.5	V
$V_T$	Receiver input threshold voltage at Bn			1.426		1.674	V
$V_{OH}$	High-level output voltage at An	Bn at 1.2 V, T/R at 0.8 V,	$\overline{CS}$ at 0.8 V, $I_{OH} = -400\ \mu\text{A}$	2.4			V
$V_{OL}$	Low-level output voltage	An	Bn at 2 V, T/R at 0.8 V, $I_{OL} = 16\text{ mA}$			0.5	V
		Bn	An at 2 V, T/R at 2 V, $R_L = 18.5\ \Omega$ , See Figure 1	$\overline{CS}$ at 0.8 V, $V_L = 2\text{ V}$ , See Figure 1	0.75	1.2	
$I_{IH}$	High-level input current	An, T/R or $\overline{CS}$	$V_I = V_{CC}$			40	$\mu\text{A}$
		Bn	$V_I = 2\text{ V}$ , An at 0.8 V,	$V_{CC} = 0\text{ or } 5.25\text{ V}$ , T/R at 0.8 V		100	
$I_{IL}$	Low level input current at An, T/R, or $\overline{CS}$	$V_I = 0.4\text{ V}$				-400	$\mu\text{A}$
$I_{OS}$	Short-circuit output current at An	An at 0 V, $\overline{CS}$ at 0.8 V,	Bn at 1.2 V, T/R at 0.8 V	-40		-120	mA
$I_{CC}$	Supply current					75	mA
$C_{O(B)}$	Driver output capacitance					4.5	pF

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .



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# SN55ALS057, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

## SN55ALS057 electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage at Dn, En, $\overline{TE}$ , or $\overline{RE}$	$V_{CC} = 4.5\text{ V}$ ,	$I_I = -18\text{ mA}$			-1.5	V
$V_T$	Receiver input threshold voltage at Bn	$V_{CC} = 5\text{ V}$ ,	$T_A = 25^\circ\text{C}$	1.45		1.65	V
		$V_{CC} = 5\text{ V}$ ,	$T_A = -55^\circ\text{C}$ to $125^\circ\text{C}$	1.4		1.7	
$V_{OH}$	High-level output voltage at Rn	$V_{CC} = 4.5\text{ V}$ ,	Bn at 1.2 V, $RE$ at 0.8 V, $I_{OH} = -400\ \mu\text{A}$	2.4			V
$V_{OL}$	Low-level output voltage	Rn	$V_{CC} = 4.5\text{ V}$ , $RE$ at 0.8 V, Bn at 2 V, $I_{OL} = 16\text{ mA}$			0.5	V
		Bn	$V_{CC} = 4.5\text{ V}$ , Dn at 2 V, En at 2 V, $\overline{TE}$ at 0.8 V, See Figure 1	0.75		1.2	
$I_{IH}$	High-level input current	Dn, En, $\overline{TE}$ , or $\overline{RE}$	$V_I = V_{CC} = 5.5\text{ V}$			40	$\mu\text{A}$
		Bn	$V_{CC} = 5.5\text{ V}$ , Dn at 0.8 V, $\overline{TE}$ at 0.8 V, $V_I = 2\text{ V}$ , En at 0.8 V			100	
$I_{IL}$	Low-level input current at Dn, En, $\overline{TE}$ , or $\overline{RE}$	$V_{CC} = 5.5\text{ V}$ ,	$V_I = 0.4\text{ V}$			-400	$\mu\text{A}$
$I_{OS}$	Short-circuit output current at Rn	$V_{CC} = 5.5\text{ V}$ , Bn at 1.2 V,	Rn at 0 V, $RE$ at 0.8 V	-435		-125	mA
$I_{CC}$	Supply current	$V_{CC} = 5.5\text{ V}$				45	mA
$C_{O(B)}$	Driver output capacitance					4.5	pF

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

## SN75ALS057 electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage at Dn, En, $\overline{TE}$ , or $\overline{RE}$	$I_I = -18\text{ mA}$				-1.5	V
$V_T$	Receiver input threshold voltage at Bn			1.43		1.65	V
$V_{OH}$	High-level output voltage at Rn	Bn at 1.2 V, $I_{OH} = -400\ \mu\text{A}$	$\overline{RE}$ at 0.8 V,	2.4			V
$V_{OL}$	Low-level output voltage	Rn	Bn at 2 V, $I_{OL} = 16\text{ mA}$ , $\overline{RE}$ at 0.8 V,			0.5	V
		Bn	Dn at 2 V, $\overline{TE}$ at 0.8 V, $R_L = 18.5\ \Omega$ , $V_I = 2\text{ V}$ , See Figure 1	En at 2 V, $V_L = 2\text{ V}$ ,	0.75		
$I_{IH}$	High-level input current	Dn, En, $\overline{TE}$ , or $\overline{RE}$	$V_I = V_{CC}$			40	$\mu\text{A}$
		Bn	$V_I = 2\text{ V}$ , Dn at 0.8 V, $\overline{TE}$ at 0.8 V, $V_{CC} = 0$ or $5.25\text{ V}$ , En at 0.8 V			100	
$I_{IL}$	Low-level input current at Dn, En, $\overline{TE}$ , or $\overline{RE}$	$V_I = 0.4\text{ V}$				-400	$\mu\text{A}$
$I_{OS}$	Short-circuit output current at Rn	Rn at 0 V, $\overline{RE}$ at 0.8 V	Bn at 1.2 V,	-40		-120	mA
$I_{CC}$	Supply current					40	mA
$C_{O(B)}$	Driver output capacitance					4.5	pF

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .



# SN55ALS056, SN75ALS056 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

## SN55ALS056 driver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	T <sub>A</sub> †	MIN	TYP‡	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-input level	An	Bn	CS at 0.8 V, V <sub>L</sub> = 2 V, R <sub>L2</sub> = 500 Ω, C <sub>L</sub> = 50 pF, See Figure 2	25°C			10	ns
t <sub>PHL</sub> Propagation delay time high-to-low-input level				Full range		40		
t <sub>PLH</sub> Propagation delay time, low-to-high-input level	CS	Bn	An and T/R at 2 V, V <sub>L</sub> = 2 V, R <sub>L1</sub> = 18 Ω, C <sub>L</sub> = 50 pF, See Figure 2	25°C			18	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-input level				Full range		30		
t <sub>PLH</sub> Propagation delay time, low-to-high-input level	T/R	Bn	CS at 0.8 V, V <sub>L</sub> = 2 V, R <sub>L1</sub> = 18 Ω, C <sub>L</sub> = 50 pF, See Figure 3	25°C			18	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-input level				Full range		37		
t <sub>PLH</sub> Propagation delay time, low-to-high-input level	An	Bn	CS at 0.8 V, V <sub>L</sub> = 2 V, R <sub>L1</sub> = 18 Ω, C <sub>L</sub> = 50 pF, See Figure 2	25°C	1	3	8	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-input level				Full range	1	33		
t <sub>PLH</sub> Propagation delay time, low-to-high-input level	An	Bn	CS at 0.8 V, V <sub>L</sub> = 2 V, R <sub>L1</sub> = 18 Ω, C <sub>L</sub> = 50 pF, See Figure 2	25°C	1	3	10	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-input level				Full range	1	13		

## SN75ALS056 driver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP§	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-output level	An	Bn	CS at 0.8 V, V <sub>L</sub> = 2 V, R <sub>L2</sub> not connected, See Figure 2			19	ns
t <sub>PHL</sub> Propagation delay time high-to-low-output level						18	
t <sub>PLH</sub> Propagation delay time, low-to-high-output level	CS	Bn	An and T/R at 2 V, V <sub>L</sub> = 2 V, R <sub>L1</sub> = 18 Ω, C <sub>L</sub> = 30 pF, See Figure 2			24	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-output level						20	
t <sub>PLH</sub> Propagation delay time, low-to-high-output level	T/R	Bn	V <sub>I(An)</sub> = 5 V, R <sub>L1</sub> = 18 Ω, R <sub>L2</sub> not connected, See Figure 3			25	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-output level						35	
t <sub>TLH</sub> Transition time, low-to-high-level output	An	Bn	CS at 0.8 V, V <sub>L</sub> = 2 V, R <sub>L1</sub> = 18 Ω, R <sub>L2</sub> not connected, See Figure 2	1	3	11	ns
t <sub>THL</sub> Transition time, high-to-low-level output				1	3	6	

† Full range is -55°C to 125°C.

‡ Typical values are at V<sub>CC</sub> = 5 V.

§ All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

TEXAS  
INSTRUMENTS

# SN75ALS056 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANCEIVER

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

**SN55ALS056 receiver**

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	T <sub>A</sub> †	MIN	MAX	UNIT	
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Bn	An	$\overline{CS}$ at 0.8 V, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF, T/ $\overline{R}$ at 0.8 V, R <sub>L2</sub> = 500 Ω, See Figure 4	25°C		20	ns	
t <sub>PHL</sub> Propagation delay time, high-to-low-level output				Full range		22		
				25°C		18		
				Full range		20		
t <sub>PLZ</sub> Output disable time from low level	$\overline{CS}$	An	Bn at 2 V, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF, See Figure 5, T/ $\overline{R}$ at 0.8 V, R <sub>L2</sub> = 500 Ω, V <sub>L</sub> = 5 V,	25°C		20	ns	
t <sub>PZL</sub> Output enable time to low level				Full range		22		
				25°C		13		
				Full range		14		
t <sub>PHZ</sub> Output disable time from high level	$\overline{CS}$	An	Bn at 0.8 V, V <sub>L</sub> = 0, R <sub>L1</sub> = R <sub>L2</sub> = 500 Ω, See Figure 5, T/ $\overline{R}$ at 0.8 V, C <sub>L</sub> = 50 pF, See Figure 5	25°C		12	ns	
t <sub>PZH</sub> Output enable time to high level				Full range		13		
				25°C		14		
				Full range		22		
t <sub>PLZ</sub> Output disable time from low level	T/ $\overline{R}$	An	Bn at 2 V, V <sub>L</sub> = 5 V, R <sub>L2</sub> = 500 Ω, See Figure 3, $\overline{CS}$ at 0.8 V, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF,	25°C		17	ns	
t <sub>PZL</sub> Output enable time to low level				Full range		20		
				25°C		25		
				Full range		40		
t <sub>PHZ</sub> Output disable time from high level	T/ $\overline{R}$	An	Bn at 0.8 V, V <sub>L</sub> = 0, R <sub>L2</sub> = 500 Ω, See Figure 3, $\overline{CS}$ at 0.8 V, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF,	25°C		12	ns	
t <sub>PZH</sub> Output enable time to high level				Full range		13		
				25°C		15		
				Full range		22		
t <sub>w(NR)</sub> Receiver noise rejection pulse duration	Bn	An	V <sub>L</sub> = 5 V, R <sub>L2</sub> = 500 Ω, See Figure 6, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF,	25°C	4		ns	
				Full range	2			

† Full range is -55°C to 125°C.



# SN75ALS056 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANCEIVER

SLLS028D - D3025, AUGUST 1987 - REVISED MARCH 1993

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

## SN75ALS056 receiver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	Bn	An	$\overline{CS}$ at 0.8 V, $T/\overline{R}$ at 0.8 V, $R_{L1} = 390 \Omega$ , $R_{L2} = 1.6 k\Omega$ , $C_L = 30 pF$ , See Figure 4		18	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output					18	
$t_{PLZ}$ Output disable time from low level	$\overline{CS}$	An	Bn at 2 V, $T/\overline{R}$ at 0.8 V, $V_L = 5 V$ , $R_{L1} = 390 \Omega$ , $R_{L2}$ not connected, See Figure 5		18	ns
$t_{PZL}$ Output enable time to low level	$\overline{CS}$	An	Bn at 2 V, $T/\overline{R}$ at 0.8 V, $V_L = 5 V$ , $R_{L1} = 390 \Omega$ , $R_{L2} = 1.6 k\Omega$ , See Figure 5		15	ns
$t_{PHZ}$ Output disable time from high level	$\overline{CS}$	An	Bn at 0.8 V, $T/\overline{R}$ at 0.8 V, $V_L = 0$ , $R_{L1} = 390 \Omega$ , $R_{L2}$ not connected, See Figure 5		8	ns
$t_{PZH}$ Output enable time to high level	$\overline{CS}$	An	Bn at 0.8 V, $T/\overline{R}$ at 0.8 V, $V_L = 0$ , $R_{L1}$ not connected, $R_{L2} = 1.6 k\Omega$ , See Figure 5		17	ns
$t_{PLZ}$ Output disable time from low level	$T/\overline{R}$	An	$\overline{CS}$ at 0.8 V, $V_I(Bn) = 2 V$ , $V_L = 5 V$ , $R_{L1} = 390 \Omega$ , $R_{L2}$ not connected, $C_L = 15 pF$ , See Figure 3		20	ns
$t_{PZL}$ Output enable time to low level	$T/\overline{R}$	An	$\overline{CS}$ at 0.8 V, $V_I(Bn) = 2 V$ , $V_L = 5 V$ , $R_{L1} = 390 \Omega$ , $R_{L2} = 1.6 k\Omega$ , $C_L = 30 pF$ , See Figure 3		40	ns
$t_{PHZ}$ Output disable time from high level	$T/\overline{R}$	An	$\overline{CS}$ at 0.8 V, $V_I(Bn) = 0$ , $V_L = 0$ , $R_{L1} = 390 \Omega$ , $R_{L2}$ not connected, $C_L = 15 pF$ , See Figure 3		17	ns
$t_{PZH}$ Output enable time to high level	$T/\overline{R}$	An	$\overline{CS}$ at 0.8 V, $V_I(Bn) = 0$ , $V_L = 0$ , $R_{L1}$ not connected, $R_{L2} = 1.6 k\Omega$ , $C_L = 30 pF$ , See Figure 3		15	ns
$t_w(NR)$ Receiver noise rejection pulse duration	Bn	An	$\overline{CS}$ at 0.8 V, $T/\overline{R}$ at 0.8 V, $R_{L1} = 390 \Omega$ , $R_{L2} = 1.6 k\Omega$ , $C_L = 30 pF$ , See Figure 6		3	ns



# SN55ALS057, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

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**switching characteristics over recommended ranges of supply voltages and operating free-air temperature**

### SN55ALS057 driver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	T <sub>A</sub> †	MIN	TYP‡	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Dn or En	Bn	$\overline{TE}$ at 0.8 V, $\overline{RE}$ at 2 V, $V_L = 2$ V, $R_{L1} = 18$ $\Omega$ , $R_{L2} = 500$ $\Omega$ , $C_L = 50$ pF, See Figure 2	25°C			10	ns
				Full range			27	
t <sub>PHL</sub> Propagation delay time, high-to-low-level output				25°C			12	ns
				Full range			15	
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	$\overline{TE}$	Bn	Dn, En, $\overline{RE}$ at 2 V, $V_L = 2$ V, $R_{L1} = 18$ $\Omega$ , $R_{L2} = 500$ $\Omega$ , $C_L = 50$ pF, See Figure 2	25°C			10	ns
Full range						27		
t <sub>PHL</sub> Propagation delay time, high-to-low-level output				25°C			17	ns
				Full range			19	
t <sub>TLH</sub> Transition time, low-to-high-level output	Dn or En	Bn	$\overline{RE}$ at 2 V, $V_L = 2$ V, $R_{L1} = 18$ $\Omega$ , $R_{L2} = 500$ $\Omega$ , $C_L = 50$ pF, See Figure 2	25°C	1	3	8	ns
t <sub>THL</sub> Transition time, high-to-low-level output				Full range	1		33	ns
				25°C	1	3	10	
				Full range	1		13	

† Full range is –55°C to 125°C.

‡ Typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

### SN75ALS057 driver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Dn or En	Bn	$\overline{TE}$ at 0.8 V, $\overline{RE}$ at 2 V, $V_L = 2$ V, $R_{L1} = 18$ $\Omega$ , $R_{L2}$ not connected, $C_L = 30$ pF, See Figure 2			19	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-level output						18	
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	$\overline{TE}$	Bn	Dn, En, $\overline{RE}$ at 2 V, $V_L = 2$ V, $R_{L2}$ not connected, $R_{L1} = 18$ $\Omega$ , $C_L = 30$ pF, See Figure 2			24	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-level output						20	
t <sub>TLH</sub> Transition time, low-to-high-level output	Dn or En	Bn	$\overline{RE}$ at 2 V, $V_L = 2$ V, $\overline{TE}$ at 0.8 V, $R_{L1} = 18$ $\Omega$ , $R_{L2}$ not connected, $C_L = 30$ pF, See Figure 2	1	3	11	ns
t <sub>THL</sub> Transition time, high-to-low-level output				1	3	6	

† All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.





# SN55ALS057, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

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**switching characteristics over recommended ranges of supply voltages and operating free-air temperature (continued)**

## SN55ALS057 receiver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	T <sub>A</sub> †	MIN	MAX	UNIT	
t <sub>PLH</sub>	Bn	Rn	$\overline{RE}$ at 0.8 V, V <sub>L</sub> = 5 V, R <sub>L2</sub> = 500 Ω, See Figure 4	$\overline{TE}$ at 2 V, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF,	25°C	20	ns	
t <sub>PHL</sub>					Full range	22		
t <sub>PLZ</sub>	$\overline{RE}$	Rn	Bn at 2 V, V <sub>L</sub> = 5 V, R <sub>L2</sub> = 500 Ω, See Figure 5	$\overline{TE}$ at 2 V, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF,	25°C	15		ns
t <sub>PZL</sub>					Full range	17		
t <sub>PHZ</sub>	$\overline{RE}$	Rn	Bn at 0.8 V, V <sub>L</sub> = 0 V, R <sub>L2</sub> = 500 Ω, See Figure 5	$\overline{TE}$ at 2 V, R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF,	25°C	14	ns	
t <sub>PZH</sub>					Full range	15		
t <sub>w(NR)</sub>	Bn	Rn	V <sub>L</sub> = 5 V, R <sub>L2</sub> = 500 Ω, See Figure 6	R <sub>L1</sub> = 500 Ω, C <sub>L</sub> = 50 pF,	25°C	4		ns
					Full range	2		

† Full range is -55°C to 125°C.

## SN75ALS057 receiver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT	
t <sub>PLH</sub>	Bn	Rn	$\overline{RE}$ at 0.8 V, R <sub>L1</sub> = 390 Ω, See Figure 4	$\overline{TE}$ at 2 V, R <sub>L2</sub> = 1.6 kΩ, C <sub>L</sub> = 30 pF,	V <sub>L</sub> = 5 V, C <sub>L</sub> = 30 pF,	18	ns
t <sub>PHL</sub>						18	
t <sub>PLZ</sub>	$\overline{RE}$	Rn	Bn at 2 V, C <sub>L</sub> = 5 pF, R <sub>L2</sub> not connected,	$\overline{TE}$ at 2 V, R <sub>L1</sub> = 390 Ω, See Figure 5	V <sub>L</sub> = 5 V,	18	ns
t <sub>PZL</sub>	$\overline{RE}$	Rn	Bn at 2 V, C <sub>L</sub> = 30 pF, See Figure 5	$\overline{TE}$ at 2 V, R <sub>L1</sub> = 390 Ω,	V <sub>L</sub> = 5 V, R <sub>L2</sub> = 1.6 kΩ,	15	ns
t <sub>PHZ</sub>	$\overline{RE}$	Rn	Bn at 0.8 V, C <sub>L</sub> = 5 pF, R <sub>L2</sub> not connected,	$\overline{TE}$ at 2 V, R <sub>L1</sub> = 390 Ω, See Figure 5	V <sub>L</sub> = 0,	17	ns
t <sub>PZH</sub>	$\overline{RE}$	Rn	Bn at 0.8 V, C <sub>L</sub> = 30 pF, R <sub>L2</sub> = 1.6 kΩ,	$\overline{TE}$ at 2 V, R <sub>L1</sub> not connected, See Figure 5	V <sub>L</sub> = 0,	17	ns
t <sub>w(NR)</sub>	Bn	Rn	$\overline{TE}$ at 2 V, R <sub>L1</sub> = 390 Ω, See Figure 6	$\overline{RE}$ at 0.8 V, R <sub>L2</sub> = 1.6 kΩ,	V <sub>L</sub> = 0, C <sub>L</sub> = 30 pF,	3	ns



# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

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**switching characteristics over recommended ranges of supply voltages and operating free-air temperature (continued)**

### SN55ALS057 driver plus receiver

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	T <sub>A</sub> †	MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Dn	Rn	$\overline{RE}$ at 0.8 V, $\overline{TE}$ at 0.8 V, $V_L = 2$ V, $R_{L1} = 500 \Omega$ , $R_{L2} = 500 \Omega$ , $C_L = 50$ pF, See Figure 7	25°C		25	ns
					Full range		35	
25°C					25			
Full range					44			
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output							

† Full range is –55°C to 125°C.

### SN75ALS057 driver plus receiver

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Dn	Rn	$\overline{RE}$ at 0.8 V, $\overline{TE}$ at 0.8 V, $R_{L2} = 1.6$ k $\Omega$ , $C_L = 30$ pF, See Figure 7		40	ns
						40	
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output						

## PARAMETER MEASUREMENT INFORMATION

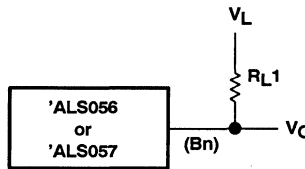
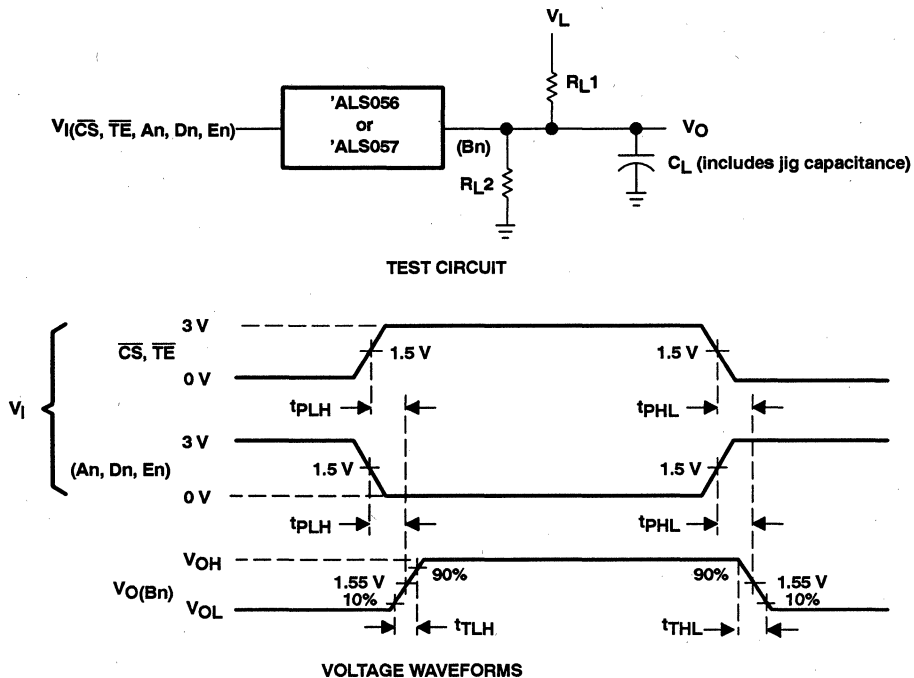


Figure 1. Driver Low-Level-Output-Voltage Test Circuit

# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION



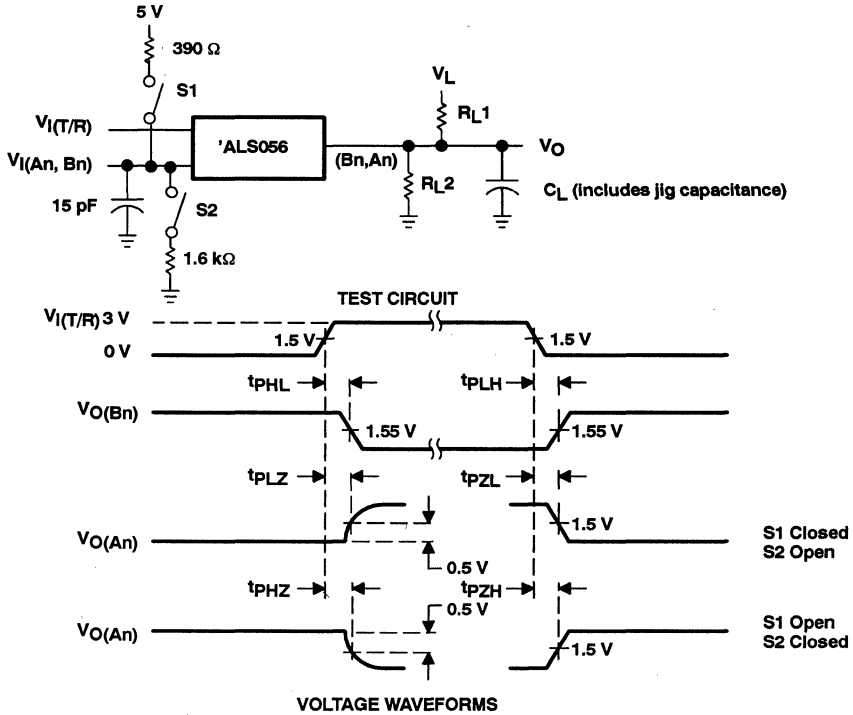
NOTE:  $t_r = t_f \leq 5$  ns from 10% to 90%

Figure 2. Driver Test Circuit and Voltage Waveforms

# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

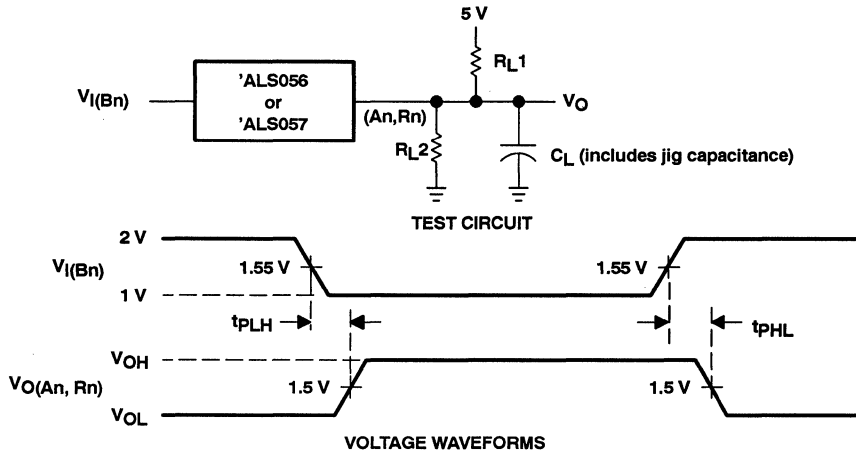
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## PARAMETER MEASUREMENT INFORMATION



NOTE:  $t_r = t_f \leq 5$  ns from 10% to 90%

Figure 3. Propagation Delay From T/R to An or Bn Test Circuit and Voltage Waveforms



NOTE:  $t_r = t_f \leq 10$  ns from 10% to 90%

Figure 4. Receiver Test Circuit and Voltage Waveforms

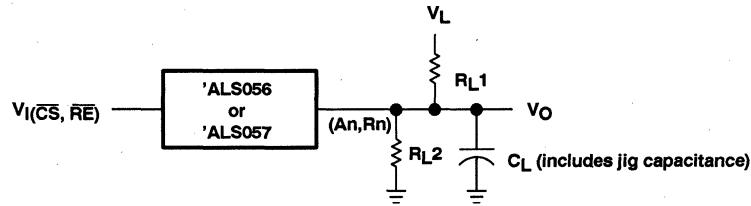
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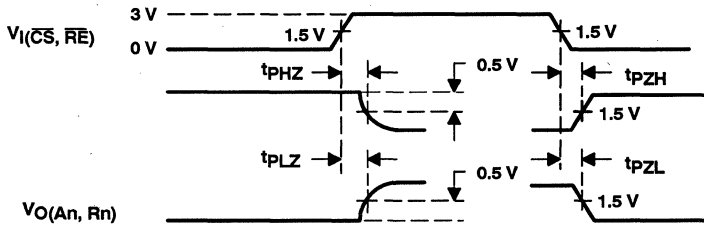
# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION



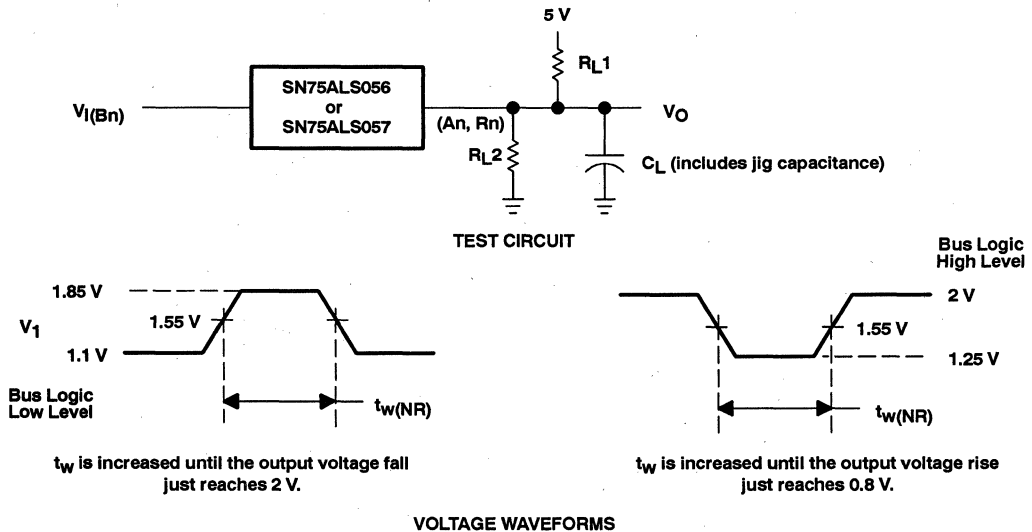
TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTE:  $t_r = t_f \leq 5$  ns from 10% to 90%

Figure 5. Propagation Delay From  $\overline{CS}$  to An or  $\overline{RE}$  to Rn Test Circuit and Voltage Waveforms



$t_w$  is increased until the output voltage fall just reaches 2 V.

$t_w$  is increased until the output voltage rise just reaches 0.8 V.

VOLTAGE WAVEFORMS

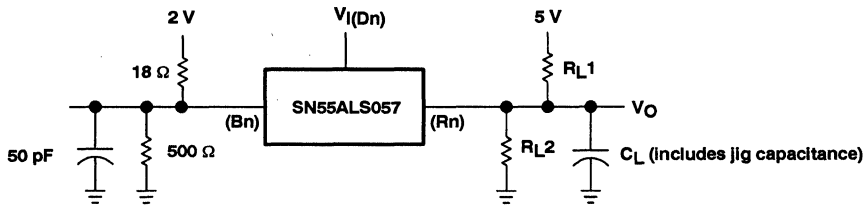
NOTE:  $t_r = t_f \leq 2$  ns from 10% to 90%

Figure 6. Receiver Noise Immunity Test Circuit and Voltage Waveforms

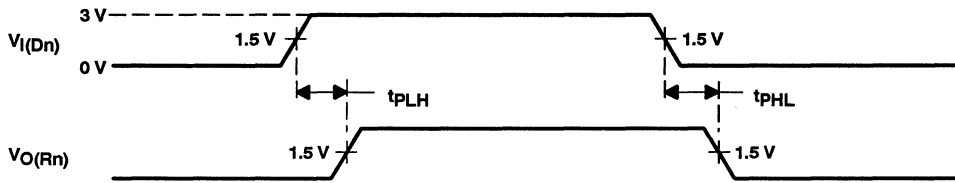
# SN55ALS056, SN55ALS057, SN75ALS056, SN75ALS057 TRAPEZOIDAL-WAVEFORM INTERFACE BUS TRANSCEIVERS

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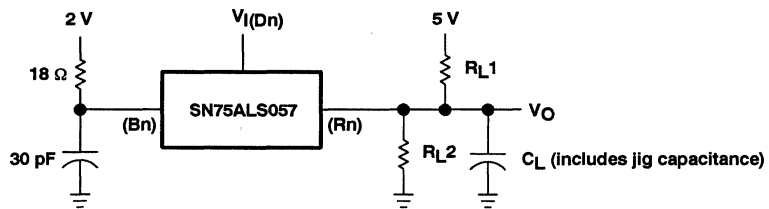
## PARAMETER MEASUREMENT INFORMATION



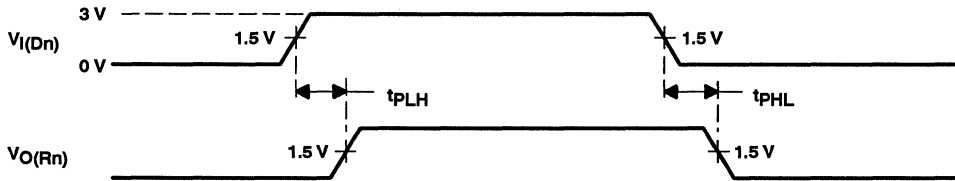
TEST CIRCUIT



VOLTAGE WAVEFORMS



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTE:  $t_r = t_f \leq 5$  ns from 10% to 90%

Figure 7. Driver Plus Receiver Delay Times Test Circuits and Voltage Waveforms

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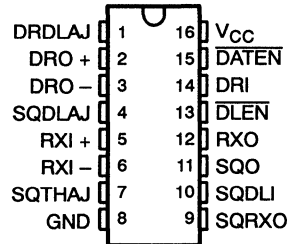


# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

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- IEEE 802.3 1BASE5 Driver and Receiver
- On-Chip Receiver Squelch With Adjustable Threshold
- Adjustable Squelch Delay
- Direct TTL-Level Squelch Output
- Squelch Circuit Allows for External Noise Filtering
- Two Driver-Enable Options
- On-Chip Start-of-Idle Detection and Disable
- Driver Provides 2 V Minimum into a 50-Ω Differential Load to Allow for Use With Doubly-Terminated Lines and Multipoint Architectures
- On-Chip Driver Slew-Rate Control for Very Closely Matched Output Rise and Fall Times

N PACKAGE  
(TOP VIEW)



### Function Tables

DRIVER

INPUTS			OUTPUTS	
DRI	DATEN	DLEN	DRO +	DRO -
L	L	X	L	H
H	L	X	H	L
X	H	H	Z	Z
H	H	L	H†	L†
L	H	L	L‡	H‡

RECEIVERS§

CONDITION	INPUTS		OUTPUTS	
	RXI +	RXI -	RXO	SGO
No active signal ¶	X	X	H	H
Active signal ¶	L	H	L	L
	H	L	H	L

† This condition is valid during the time period set by DRDLAJ following a rising transition on DRI. Following this, if no subsequent positive transition occurs on DRI, the outputs will go to the high-impedance state.

‡ This condition is valid if it occurs within the enable time set by DRDLAJ after a rising transition on DRI. Otherwise, the outputs will be in the high-impedance state.

§ Pins 9 and 10 are tied together.

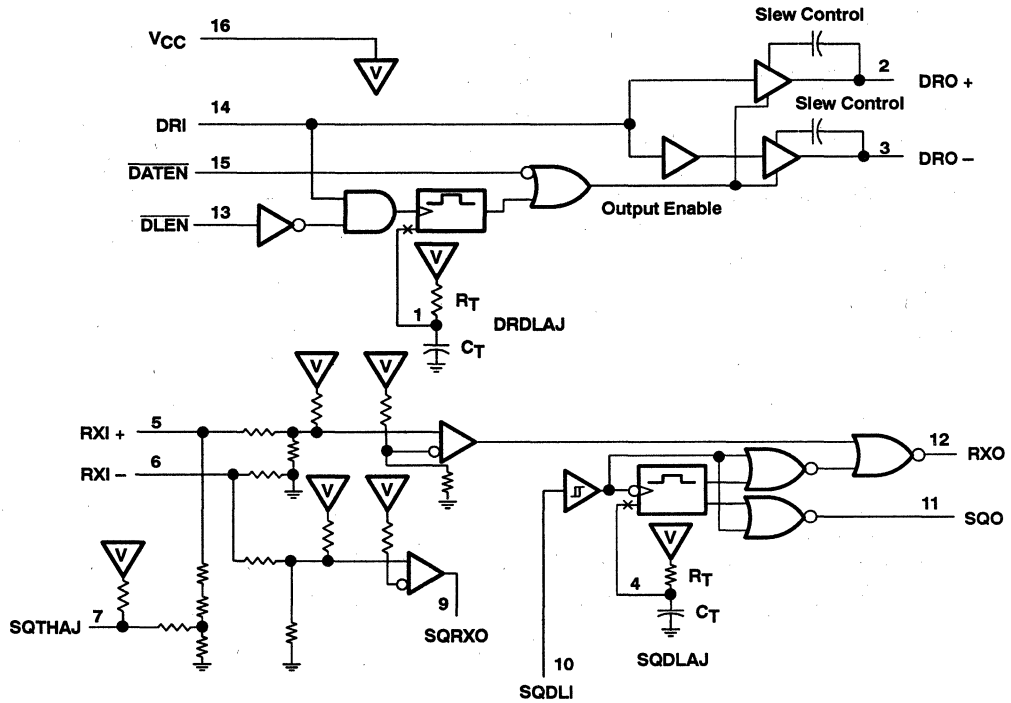
¶ An active signal is one that has an amplitude greater than the threshold level set by SQTHAJ.



# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

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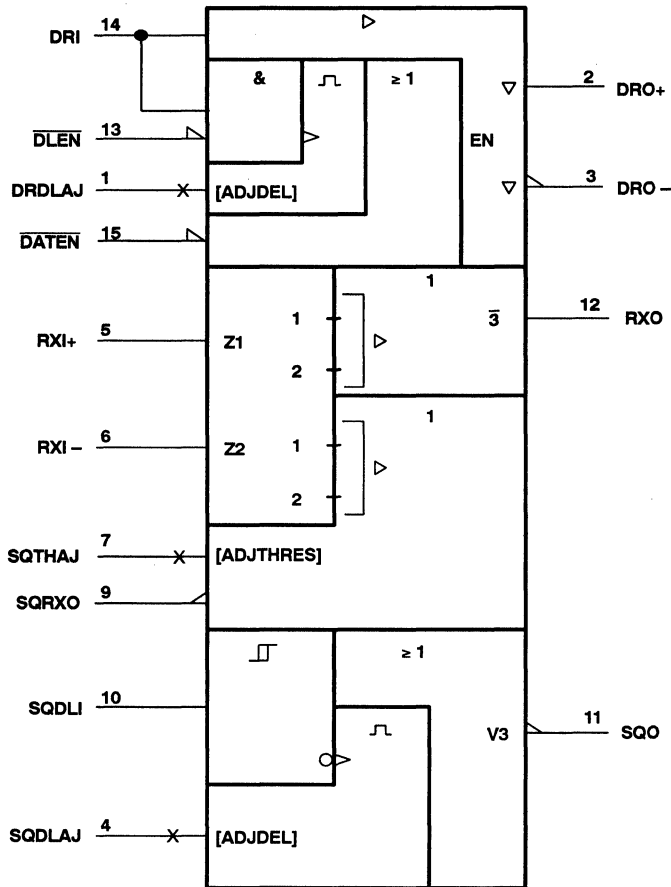
## logic diagram (positive logic)



# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

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## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## description

The SN75061 is a single-channel driver/receiver pair designed for use in IEEE 802.3, 1BASE5 applications as well as other general data communications circuits. The SN75061 offers the system designer both a driver and a receiver that are easily configured for use with a variety of controllers and data encoder/decoders.

The receiver features a full analog squelch circuit with an adjustable threshold and a programmable squelch delay. Internal nodes of the squelch circuitry are brought out to external connections to allow for the insertion of noise filtering circuitry of the designer's choice.

As with the receiver, the driver offers a variety of implementation options. Driver enabling may be controlled directly by an external logic input or by use of an on-chip one-shot that is retrigged as long as data is being sent to the driver. The driver will then automatically go to the high-impedance state when end-of-packet occurs. The driver features internal slew-rate control for optimal matching of rise and fall times allowing for reduction of driver-induced jitter.

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# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

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## receiver

The SN75061 receiver implements full analog squelch functions by integrating both a separate, parallel squelch receiver with an externally programmable threshold, and a programmable one-shot. The output of the squelch receiver and the input to the high-level dc-triggered one-shot are brought out to external connections. These pins can be shorted for direct implementation or used for the insertion of noise-filtering circuitry of the implementer's design. The receiver one-shot can be effectively bypassed by applying a high logic level to SQDLI. The squelch threshold may be set externally by applying an external voltage set to a level that is  $-2$  times the desired threshold voltage. If SQTHAJ is left open, the squelch receiver will default to its internal preset value of  $-600$  mV. The receiver also outputs a high logic squelch signal when there is no active data present at the receiver inputs. When no data is present on the transmission line, the receiver output assumes a high level. The unsquelch duration is set externally with an R-C combination at SQDLAJ.

## driver

The driver offers a variety of implementation options. Driver enabling may be controlled directly by an active-low external logic input on DATEN or by use of another on-chip one-shot that retriggers with positive-going transitions on the driver input line. If no positive transition occurs within the pulse duration set by an external R-C combination, the one-shot times out and the driver is automatically put into a high-impedance state. When operating in the delay-enable mode, the 2-bit-time high-level start-of-idle pulse prescribed by IEEE 802.3 1BASE5 causes the one-shot to time out and automatically place the driver outputs in the high-impedance state. This delay time is also adjustable for use in other applications. The driver implements an output slew-rate control that is internally set for nominally 40 mV/ns. (This is roughly a 100-ns peak-to-peak differential transition time.) The driver outputs are capable of driving a 50- $\Omega$  differential load with a minimum output level of 2 V. Short-circuit output current is greater than 100 mA.

## Terminal Functions

PIN		DESCRIPTION
NAME	NO.	
DATEN	15	Driver data enable. When low, driver outputs are in an active state. When high, the driver outputs are in a high-impedance state if DLEN is also high.
DLEN	13	Driver delay enable. When this signal is low and DATEN is high, the driver outputs are active for a period of time set by DRDLAJ after a positive-going transition on DRI. If there is no active data on DRI, the outputs are in a high-impedance state.
DRDLAJ	1	Driver delay adjust is a connection for the external R-C combination that determines the duration of the driver output active state after a positive transition on DRI when DLEN is low and DATEN is high.
DRI	14	Driver data input
DRO+	2	Noninverting driver output
DRO-	3	Inverting driver output
GND	8	Ground. Common for all voltages
RXI+	5	Noninverting receiver input
RXI-	6	Inverting receiver input
RXO	12	Main receiver input
SQDLAJ	4	Squelch delay adjust is a connection for an external R-C combination that determines the duration of the receiver unsquelch after a negative-going transition on SQDLI.
SQDLI	10	Squelch delay input is the input to the one-shot that controls the duration of the receiver unsquelch period. The main receiver output remains unsquelched as long as SQDLI is held high. Timing of the unsquelch period begins on the high-to-low transition of SQDLI.
SQO	11	Squelch output is high while the receiver is squelched.
SQRXO	9	Squelch receiver output is high only when the differential receiver input exceeds the threshold set by SQTHAJ.
SQTHAJ	7	Squelch receiver threshold adjust. The voltage at this input determines the threshold of the squelch receiver in a ratio of $-2$ , SQTHAJ to threshold. If left open, the squelch receiver threshold defaults to $-600$ mV.
VCC	16	Supply-voltage input

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# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$	7 V
Input voltage (any logic input)	7 V
Receiver differential input voltage	$\pm 25$ V
Receiver input voltage	$\pm 15$ V
Driver output voltage	-0.5 V to 15 V
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 1)	1150 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: For operation above 25°C free-air temperature, derate to 736 mW at 70°C at the rate of 9.2 mW/°C.

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Driver high-level input voltage, $V_{IH}$	2			V
Driver low-level input voltage, $V_{IL}$			0.8	V
Receiver common-mode input voltage, $V_{IC}$ (see Note 2)	-2.5		5	V
Driver high-level output current, $I_{OH}$			-150	mA
Driver low-level output current, $I_{OL}$			150	mA
External timing resistance, $R_{ext}$	5		260	k $\Omega$
External timing capacitance, $C_{ext}$	No restriction			
Operating free-air temperature, $T_A$	0		70	°C

NOTE 2: The algebraic convention, in which the less-positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode input voltage  $V_{IC}$  and threshold levels  $V_{T+}$  and  $V_{T-}$ .

## electrical characteristics over recommended operating free-air and supply voltage range (unless otherwise noted)

### driver

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA			-1.5	V
$V_{OD}$	Differential-output voltage	$R_L = 50 \Omega$	2	2.4	3.3	V
		$R_L = 115 \Omega$			3.65	
$\Delta V_{OD}$	Change in differential-output voltage for a change in logic input state				50	mV
$I_{IH}$	High-level input current	$V_I = 2.4$ V			20	$\mu$ A
$I_{IL}$	Low-level input current	$V_I = 0.5$ V	$\pm 100$		-35	$\mu$ A
$I_{OS}$	Short-circuit output current	$V_O = 0$ or 6 V, $V_I = 0.8$ V or 2.5 V			$\pm 300$	mA
$I_{OZ}$	High-impedance output current	$V_{CC} = 5.25$ V	$V_{OC} = 10$ V		100	$\mu$ A
			$V_{OC} = 0$		-100	

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .



# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

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**electrical characteristics over recommended operating free-air and supply voltage range (unless otherwise noted) (continued)**

## receiver

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage, squelch delay	$I_I = -18 \text{ mA}$				-1.5	V
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7 \text{ V}$ ,	$I_O = -0.4 \text{ mA}$			50	mV
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5 \text{ V}$ ,	$I_O = 16 \text{ mA}$	-50‡			mV
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )				50		mV
$V_{IC}$	Common-mode input voltage					5	V
$V_{OH}$	High-level output voltage	RXO	$V_{CC} = 4.75 \text{ V}$ , SQDLAJ at 0.8 V	$I_{OH} = -400 \mu\text{A}$ ,	2.7		V
		SQO			2.7	3.5	
		SQRXO	$V_{CC} = 4.75 \text{ V}$ , $V_{ID(RXI)} = -0.7 \text{ V}$ ,	$I_{OH} = -20 \mu\text{A}$ , SQDLAJ open	2.7	4.65	
$V_{OL}$	Low-level output voltage	RXO	$V_{CC} = 4.75 \text{ V}$ , SQDLAJ at 2 V	$I_{OL} = 8 \text{ mA}$	0.45		V
				$I_{OL} = 16 \text{ mA}$	0.5		
		SQO		$I_{OL} = 8 \text{ mA}$	0.35	0.5	
				$I_{OL} = 16 \text{ mA}$	0.5		
		SQRXO		$I_{OL} = 8 \text{ mA}$	0.45		
				$I_{OL} = 16 \text{ mA}$	0.5		
$I_{IH}$	High-level input current	SQDLI	$V_I = 2.4 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$	Low-level input current	SQDLI	$V_I = 0.5 \text{ V}$			-35	$\mu\text{A}$
$I_{OS}$	Short-circuit output current	RXO	$V_{CC} = 5.25 \text{ V}$ ,	$V_O = 0$	-15	-85	mA
		SQO			-15	-100	
		SQRXO			$V_{CC} = 5 \text{ V}$ ,	$V_O = 0$	
$r_i$	Input resistance			10			k $\Omega$
$V_{T-(sq)}$	Squelch preset threshold voltage	$V_{CC} = 5 \text{ V}$ , SQTHAJ open	$V_{IC} = 1.5 \text{ V to } 3.5 \text{ V}$	-525	-600	-675	mV
			$V_{IC} = -2.5 \text{ V to } 1.5 \text{ V}$ or 3.5 V to 5 V	-500		-700	mV
Ratio of SQTHAJ input voltage to actual squelch threshold voltage		SQTHAJ at 200 mV to 4 V		-1.9		-2.1	

## driver and receiver

$I_{CC}$	Supply current	$V_{CC} = 5.25 \text{ V}$ , No load	Driver outputs disabled,	70			mA
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† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode input voltage  $V_{IC}$  and threshold levels  $V_{T+}$  and  $V_{T-}$ .

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switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

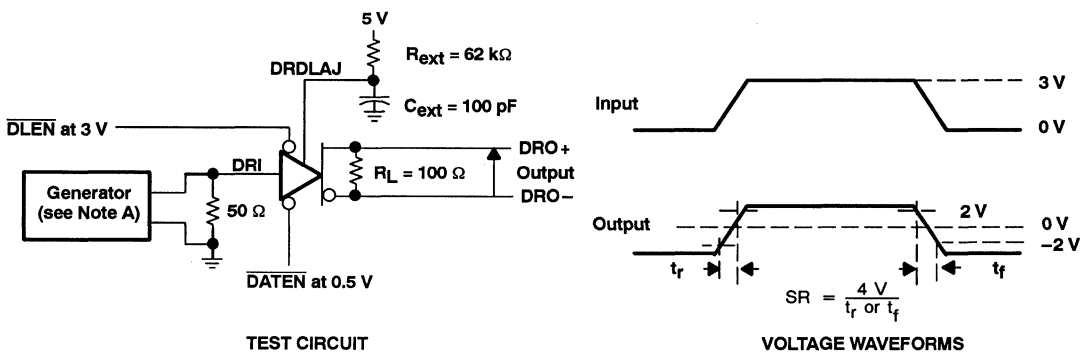
**driver**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
SR	Differential-output slew rate	$V_O = -2\text{ V to } 2\text{ V}$ , $R_L = 100\ \Omega$ (differential), See Figure 1	28	40	52	mV/ns	
$t_{dD}$	Differential-output delay time ( $t_{dD+}$ and $t_{dD-}$ )	$C_1 = 15\text{ pF}$ , $R_L = 100\ \Omega$ (differential), See Figure 2			160	ns	
$t_{dD+} - t_{dD-}$	Differential-output delay time difference	$R_L = 100\ \Omega$ (differential), See Figure 2			5	ns	
$t_{PHZ}$	Disabled time from $\overline{\text{DATEN}}$	See Figure 3, 4, and 5			220	ns	
$t_{PLZ}$					300	ns	
$t_{PHZ}$	Enable time from $\overline{\text{DATEN}}$				220	ns	
$t_{PLZ}$					290	ns	
$t_{PZH}$	Enable time from $\overline{\text{DLEN}}$				250	ns	
$t_{w(en)}$	Enable pulse duration time (with DLEN low)	$C_{ext} = 100\text{ pF}$ , See Figure 6	$R_{ext} = 62\text{ k}\Omega$ ,	2	2.5	3	$\mu\text{s}$

**receiver**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
$t_{en(RX)}$	Receiver enable time	Squelch off, See Figure 7		117		ns	
$t_{PLH}$	Propagation delay time, low-to-high level output	Squelch off, See Figure 8		20	35	ns	
$t_{PHL}$	Propagation delay time, high-to-low level output	Squelch off, See Figure 8		22	35	ns	
$t_d(\text{unsq})$	Unsquench delay time	$C_{ext} = 50\text{ pF}$ , See Figure 9	$R_{ext} = 51\text{ k}\Omega$ ,	1	1.2	1.45	$\mu\text{s}$
		$C_{ext} = 15\text{ pF}$ , See Figure 9	$R_{ext} = 6.8\text{ k}\Omega$ ,			180	ns

**PARAMETER MEASUREMENT INFORMATION**



NOTE A: The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1\text{ MHz}$ , duty cycle  $\leq 50\%$ ,  $t_r \leq 6\text{ ns}$ ,  $t_f \leq 6\text{ ns}$ ,  $Z_0 = 50\ \Omega$ .

**Figure 1. Driver Slew Rate Test Circuit and Voltage Waveforms**

# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

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## PARAMETER MEASUREMENT INFORMATION

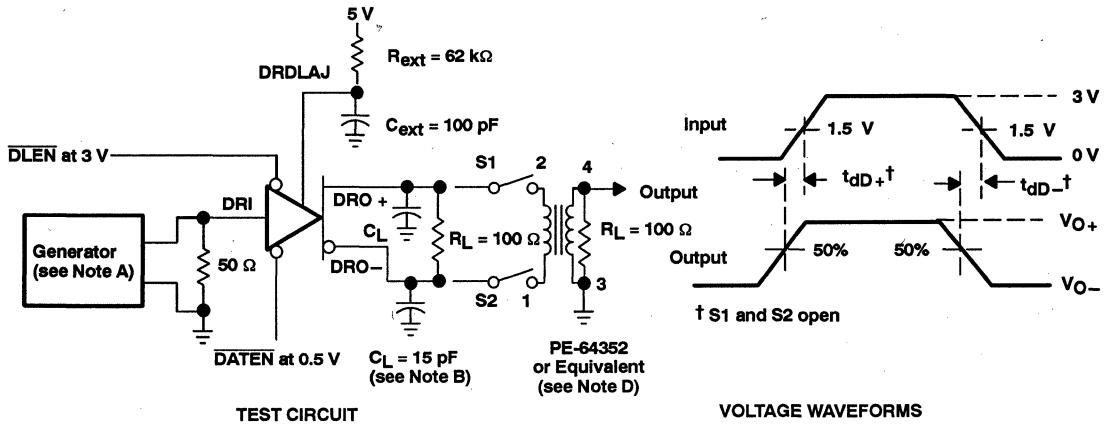


Figure 2. Driver Differential Delay Times Test Circuit and Voltage Waveforms

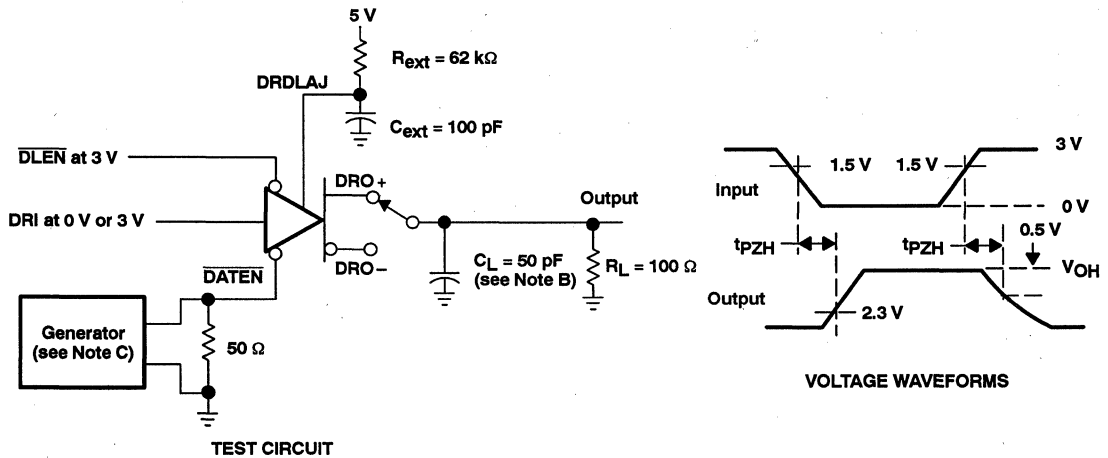


Figure 3. Driver Enable and Disable Time Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .
- B.  $C_L$  includes probe and jig capacitance.
- C. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  500 kHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .
- D. When measuring differential-output delay time difference, switches S1 and S2 are closed. (Isolation transformer from Pulse Engineering P/N PE-64352).

PARAMETER MEASUREMENT INFORMATION

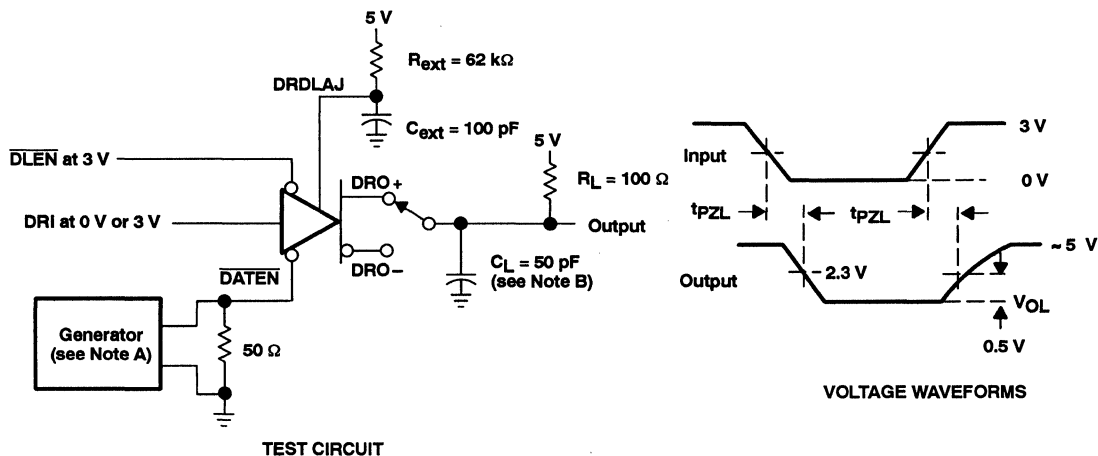


Figure 4. Driver Enable and Disable Time Test Circuit and Voltage Waveforms

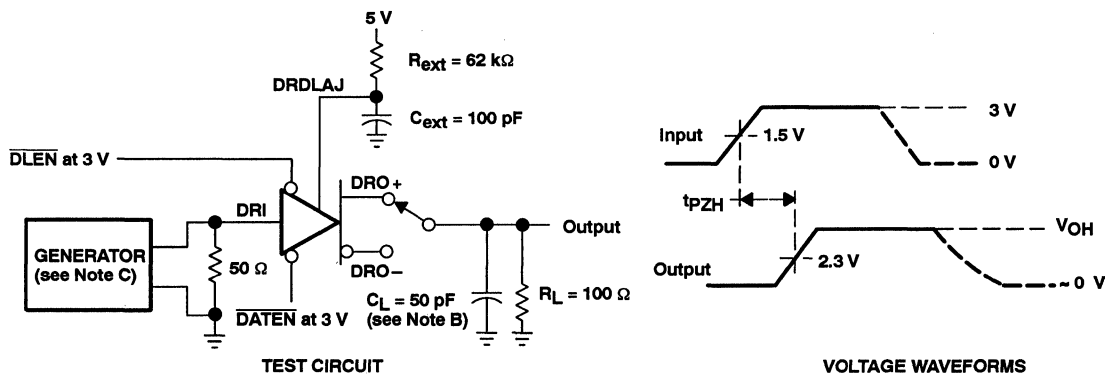


Figure 5. Enable Times From Delay Enable Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  200 kHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .



# SN75061 DRIVER/RECEIVER PAIR WITH SQUELCH

SLLS026C – D2959, JANUARY 1987 – REVISED JULY 1990

## PARAMETER MEASUREMENT INFORMATION

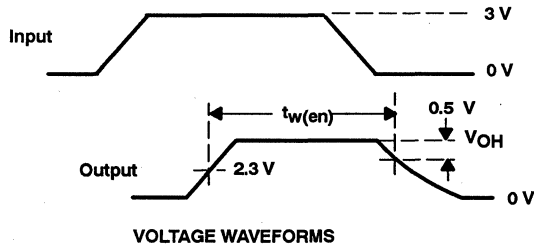
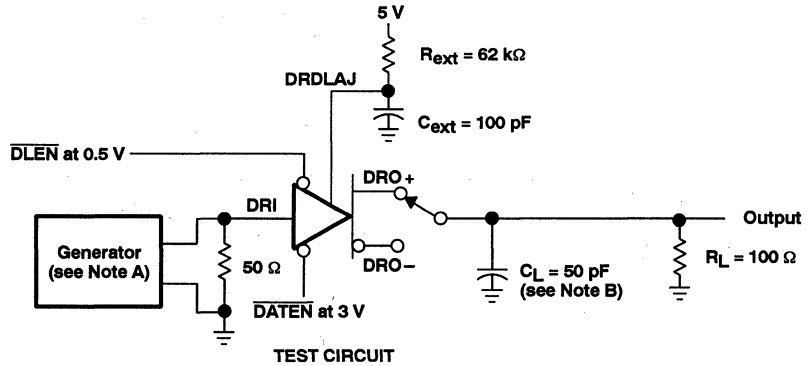


Figure 6. Enable Pulse Duration With Delay Enable Low Test Circuit and Voltage Waveforms

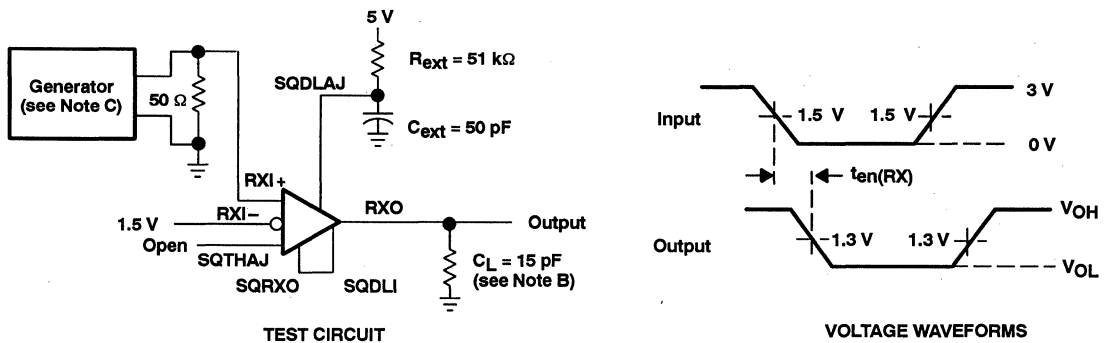


Figure 7. Receiver Enable (Unsquench) Time Test Circuit and Voltage Waveforms

- NOTES: D. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  200 kHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .
- E.  $C_L$  includes probe and jig capacitance.
- F. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  500 kHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .



PARAMETER MEASUREMENT INFORMATION

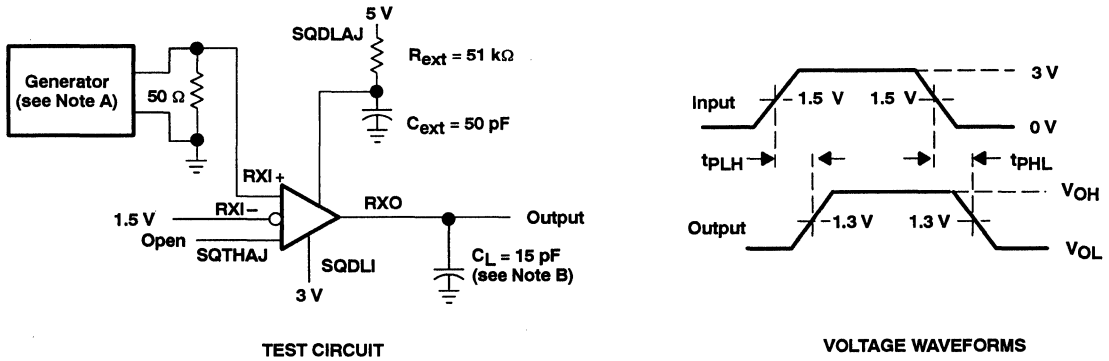


Figure 8. Receiver Propagation Delay Time Test Circuit and Voltage Waveforms

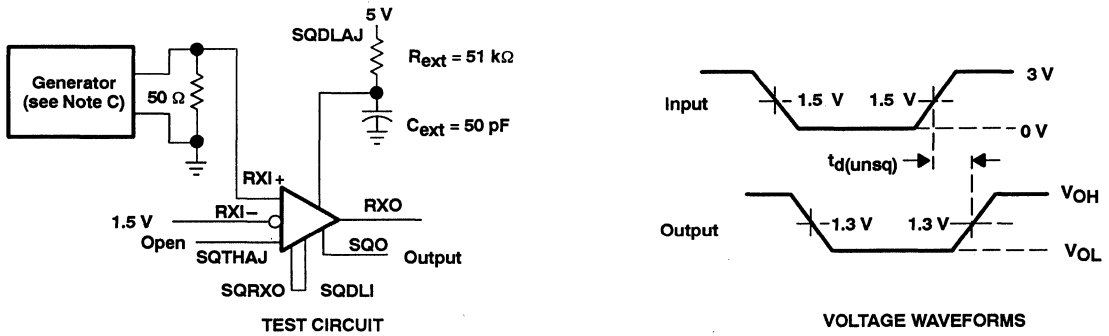


Figure 9. Unsquelch Duration Time Test Circuit and Voltage Waveforms

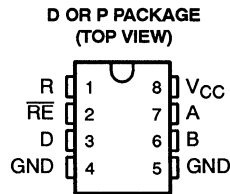
- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .
- B.  $C_L$  includes probe and jig capacitance.
- C. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  100 kHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .



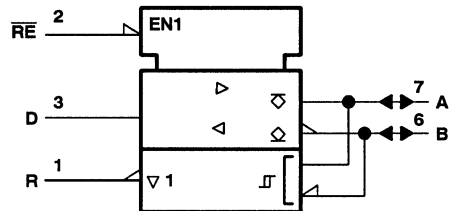
# SN65076B, SN75076B DIFFERENTIAL BUS TRANSCEIVERS

SLLS061 – D3407, JANUARY 1990

- Bidirectional Transceiver
- Designed for Multipoint Transmission in Noisy Environments Such as Automotive Applications
- 3-State Driver and Receiver Outputs
- Individual Driver and Receiver Enables
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capability . . .  $\pm 10$  mA Max
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Impedance . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From Single 5-V Supply
- Low Power Requirements



logic symbol†



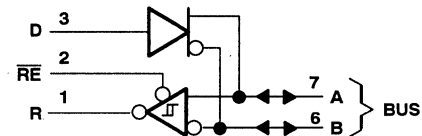
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## description

The SN65076B and SN75076B differential bus transceivers are monolithic integrated circuits designed for bidirectional data communication on multipoint bus transmission lines. They are designed for noisy environments, where a low-impedance termination to ground is required.

The SN65076B and SN75076B combine a differential line driver and a differential input line receiver, both of which operate from a single 5-V power supply. The receiver has an active-low enable. The driver differential outputs and the receiver differential inputs are connected internally to form differential input/output (I/O) bus ports that are designed to offer minimum loading to the bus whenever the driver is disabled or  $V_{CC} = 0$ . These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

logic diagram (positive logic)



## Function Tables

DRIVER

INPUT D	OUTPUTS	
	A	B
H	H	L
L	L <sup>†</sup>	H <sup>†</sup>

† These levels assume that the open-collector outputs (A) and the open-emitter outputs (B) are connected to a pullup and pulldown resistor, respectively.

RECEIVER

DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_D \geq 0.2$ V	L	L
$-0.2$ V < $V_D < 0.2$ V	L	?
$V_D \leq -0.2$ V	L	H
X	H	Z

H = high level, L = low level, ? = indeterminate;  
X = irrelevant, Z = high impedance (off)

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

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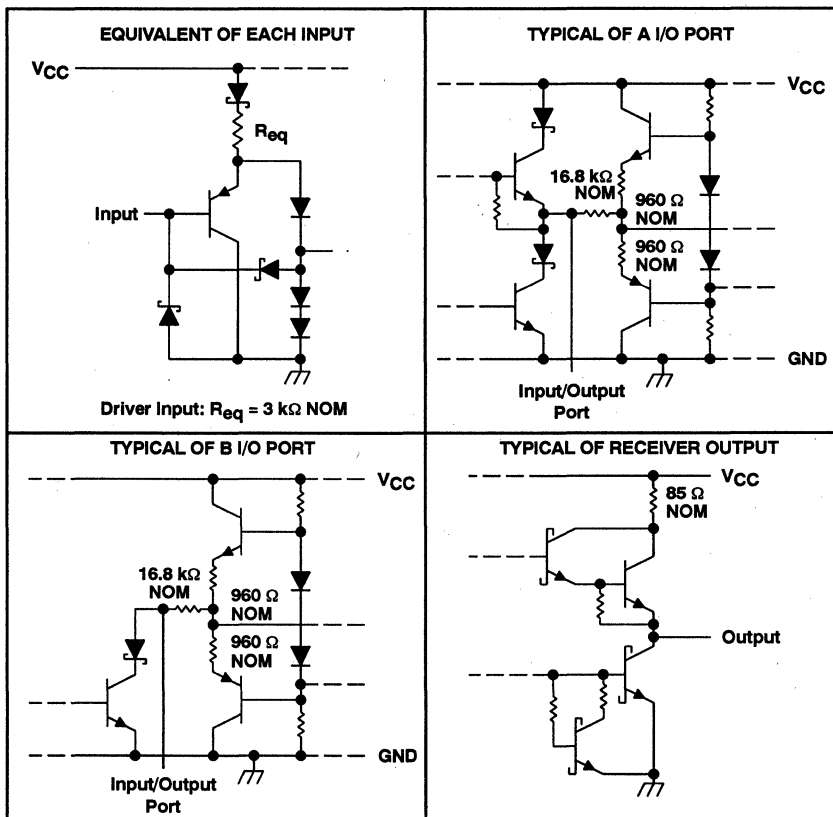
# SN65076B, SN75076B DIFFERENTIAL BUS TRANSCEIVERS

SLLS061 - D3407, JANUARY 1990

## description (continued)

The driver is designed to handle loads up to 10 mA of sink and source current. The driver features positive- and negative-current limiting and thermal shutdown for protection from line fault conditions. Thermal shutdown is designed to occur at a junction temperature of approximately 150°C in the P package and 170°C in the D package. The receiver features a minimum input impedance of 12 kΩ, an input sensitivity of ±200 mV, and a typical input hysteresis of 50 mV.

The SN65076B is characterized for operation from -40°C to 105°C and the SN75076B is characterized for operation from 0°C to 70°C.



# SN65076B, SN75076B DIFFERENTIAL BUS TRANSCEIVERS

SLLS061 – D3407, JANUARY 1990

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	–10 V to 15 V
Enable input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65076B	–40°C to 105°C
SN75076B	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds	260°C

NOTE 1: All voltage values, except differential input/output bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 105^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	261 mW
P	1100 mW	8.8 mW/°C	702 mW	396 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT	
Supply voltage, $V_{CC}$		4.75	5	5.25	V	
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$					12	V
					–7	
High-level input voltage, $V_{IH}$	D and $\overline{RE}$	2			V	
Low-level input voltage, $V_{IL}$	D and $\overline{RE}$				0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)					±12	V
High-level output current, $I_{OH}$	Driver (A)				–10	mA
	Receiver				–400	µA
Low-level output current, $I_{OL}$	Driver (B)				10	mA
	Receiver				8	
Operating free-air temperature, $T_A$	SN65076B	–40			105	°C
	SN75076B	0			70	

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.



# SN65076B, SN75076B DIFFERENTIAL BUS TRANSCEIVERS

SLLS061 – D3407, JANUARY 1990

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$		-1.5	V
$V_O$	Output voltage	$V_I = 2 \text{ V}$ , $I_O = 0$	0	6	V
$V_{OD1}$	Differential output voltage	$I_O = 0$	1.5	6	V
$V_{OD2}$	Differential output voltage	See Figure 1	1.5	5	V
$I_O$	Output current	$V_I = 0.8 \text{ V}$		1	mA
		$V_O = 12 \text{ V}$ $V_O = -7 \text{ V}$		-0.8	
$I_{IH}$	High-level input current	$V_I = 2.4 \text{ V}$		20	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$		-400	$\mu\text{A}$
$I_{OS}$	Short-circuit output current	$V_O = -7 \text{ V}$		-250	mA
		$V_O = 0$		-150	
		$V_O = V_{CC}$		250	
		$V_O = 12 \text{ V}$		250	
$I_{CC}$	Supply current (total package)	No load		30	mA

switching characteristics,  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{on}$	Differential-output turn-on time	See Figure 3		60	90	ns
$t_{off}$	Differential-output turn-off time			75	110	ns

RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
V <sub>T+</sub> Positive-going input threshold voltage	V <sub>O</sub> = 2.7 V, I <sub>O</sub> = -0.4 mA			0.2	V
V <sub>T-</sub> Negative-going input threshold voltage	V <sub>O</sub> = 0.5 V, I <sub>O</sub> = 8 mA	-0.2‡			V
V <sub>hys</sub> Hysteresis (V <sub>T+</sub> - V <sub>T-</sub> )			50		mV
V <sub>IK</sub> Enable-input clamp voltage	I <sub>I</sub> = -18 mA			-1.5	V
V <sub>OH</sub> High-level output voltage	V <sub>ID</sub> = -200 mV, See Figure 2, I <sub>OH</sub> = -400 μA	2.7			V
V <sub>OL</sub> Low-level output voltage	V <sub>ID</sub> = -200 mV, See Figure 2, I <sub>OL</sub> = 8 mA			0.45	V
I <sub>OZ</sub> High-impedance-state output current	V <sub>O</sub> = 0.4 V to 2.4 V			±20	μA
I <sub>I</sub> Line input current	Other input = 0 V, V <sub>I</sub> = 12 V, V <sub>I</sub> = -7 V, See Note 3			1 -0.8	mA
I <sub>IH</sub> High-level enable-input current	V <sub>IH</sub> = 2.7 V			20	μA
I <sub>IL</sub> Low-level enable-input current	V <sub>IL</sub> = 0.4 V			-100	μA
r <sub>i</sub> Input resistance			12		kΩ
I <sub>OS</sub> Short-circuit output current		-15		-85	mA
I <sub>CC</sub> Supply current (total package)	No load			30	mA

† All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for threshold voltage levels only.

NOTE 3: This applies for both power on and power off.

switching characteristics, V<sub>CC</sub> = 5 V, C<sub>L</sub> = 15 pF, T<sub>A</sub> = 25°C

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high level output	V <sub>ID</sub> = 0 to 3 V, See Figure 4		21	35	ns
t <sub>PHL</sub> Propagation delay time, high-to-low level output			23	35	ns
t <sub>PZH</sub> Output enable time to high level	See Figure 5		10	20	ns
t <sub>PZL</sub> Output enable time to low level			12	20	ns
t <sub>PHZ</sub> Output disable time from high level	See Figure 5		20	35	ns
t <sub>PLZ</sub> Output disable time from low level			17	25	ns



# SN65076B, SN75076B DIFFERENTIAL BUS TRANSCEIVERS

SLLS061 – D3407, JANUARY 1990

## PARAMETER MEASUREMENT INFORMATION

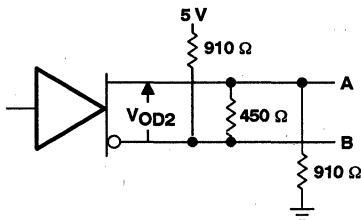


Figure 1. Driver  $V_{OD2}$

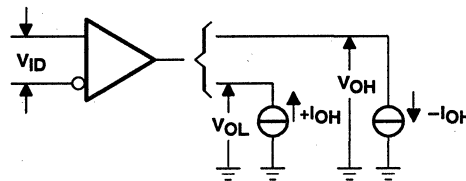
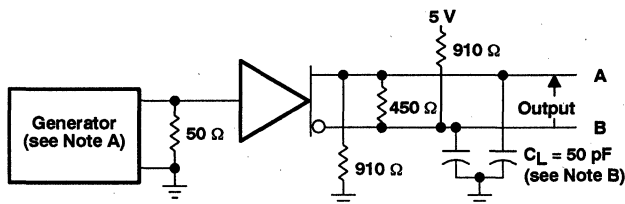
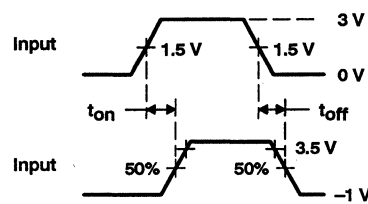


Figure 2. Receiver  $V_{OH}$  and  $V_{OL}$

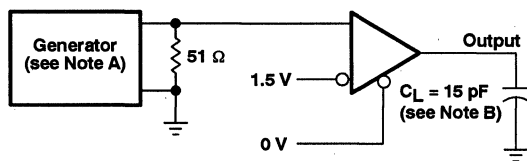


TEST CIRCUIT

Figure 3. Driver Differential-Output Delay Times

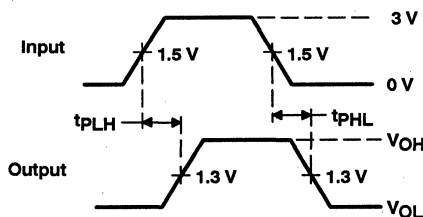


VOLTAGE WAVEFORMS



TEST CIRCUIT

Figure 4. Receiver Test Circuit and Voltage Waveforms Propagation Delay Times

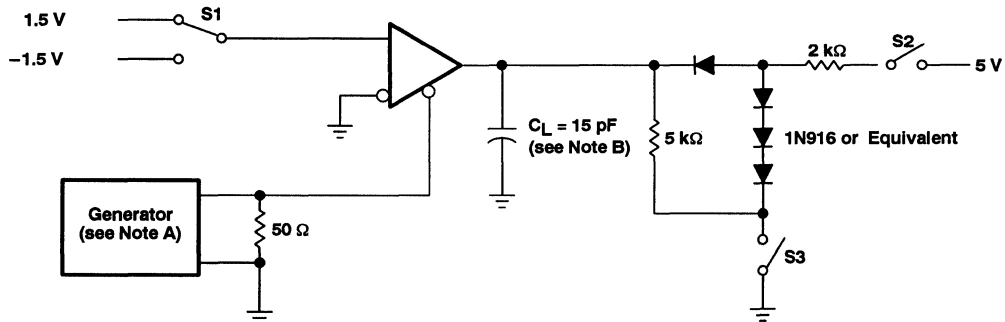


VOLTAGE WAVEFORMS

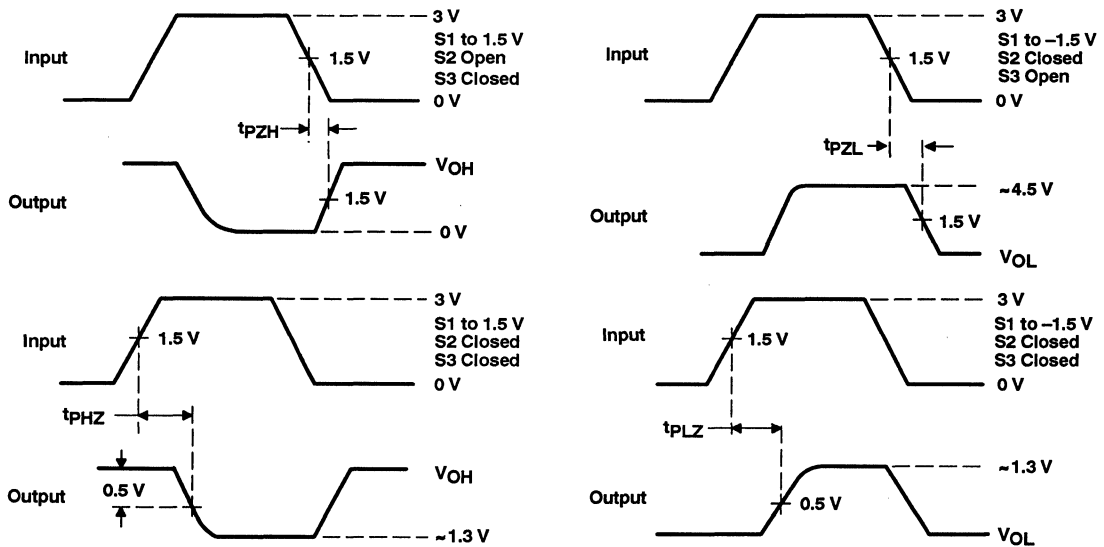
NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  500 kHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

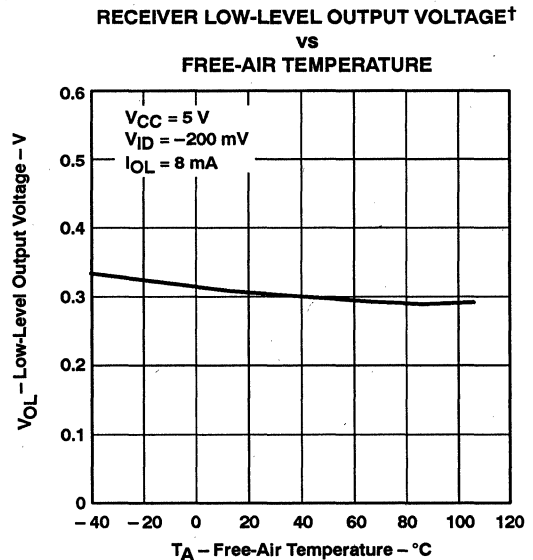
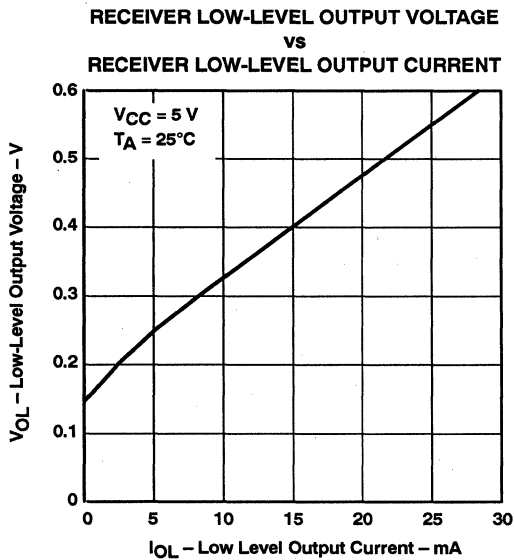
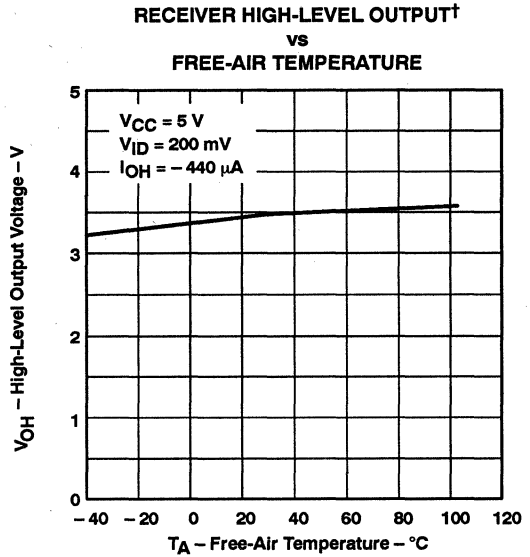
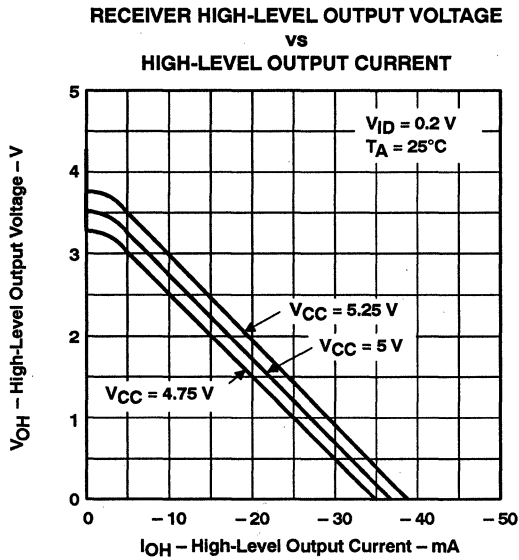
Figure 5. Receiver Output Enable and Disable Times

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 500 \text{ kHz}$ , 50% duty cycle,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ ,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

# SN65076B, SN75076B DIFFERENTIAL BUS TRANSCEIVERS

SLLS061 – D3407, JANUARY 1990

## TYPICAL CHARACTERISTICS



† Only the 0°C to 70°C portion of the curve applies for the SN75076B.

TYPICAL CHARACTERISTICS

RECEIVER OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE

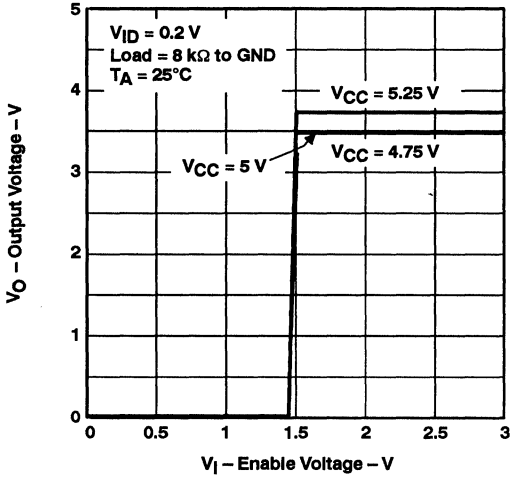


Figure 10

RECEIVER OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE

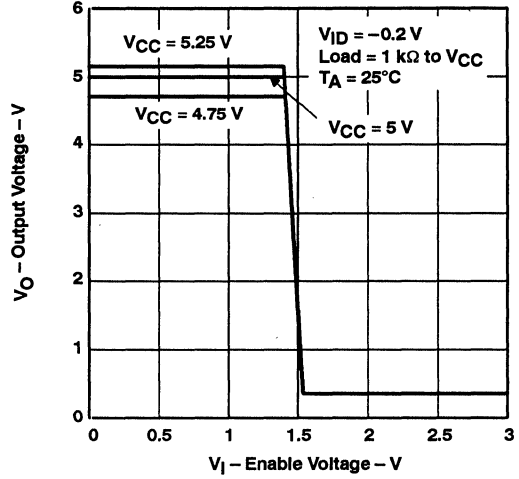


Figure 11

APPLICATION INFORMATION

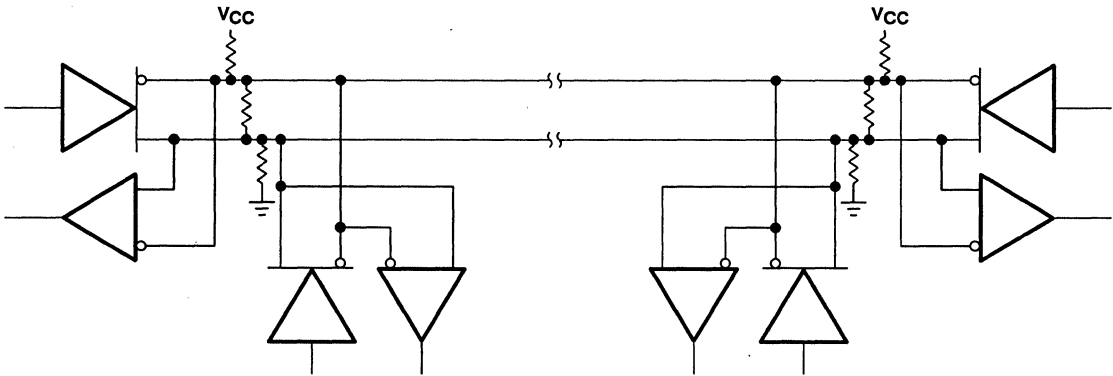


Figure 12. Typical Application Circuit



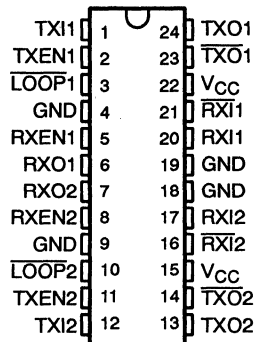
# SN75ALS085

## LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A – D3279, APRIL 1989 – REVISED FEBRUARY 1993

- **Compatible With IOS 8802.3:1989 and ANSI/IEEE Std 802.3-1988**
- **Interdevice Loop-Back Paths for System Testing**
- **Squelch Function Implemented on the Receiver Inputs**
- **Drivers Will Drive a Balanced 78-Ω Load**
- **Transformer Coupling Not Required in System**
- **Power-Up/Power-Down Protection (Glitch Free)**
- **Isolated Ground Pins for Reduced Noise Coupling**
- **Fault-Condition Protection Built into the Device**
- **Driver Inputs Are Level-Shifted ECL Compatible**

**DW OR NT PACKAGE  
(TOP VIEW)**



### description

The SN75ALS085 is a monolithic, high-speed, advanced low-power Schottky, dual-channel driver/receiver device designed for use in the AUI of ANSI/IEEE Std 802.3-1988. The two drivers on the device drive a 78-Ω balanced, terminated twisted-pair transmission line up to a maximum length of 50 meters. In the off (idle) state, the drivers maintain minimal differential output voltage on the twisted-pair line and, at the same time, remain within the required output common-mode range.

With the driver enable (TXEN) high, upon receiving the first falling edge into the driver input, the differential outputs will rise to full-amplitude output levels within 25 ns. The output amplitude is maintained for the remainder of the packet. After the last positive packet edge is transmitted into the driver, the driver will maintain a minimum of 70% full differential output for a minimum of 200 ns, then decay down to a minimum level for the reset (idle) condition within 8 μs. Disabling the driver by taking the driver enable low will also force the output into the idle condition after the normal 8-μs timeout. While operating, the driver is able to withstand a set of fault conditions and not suffer damage due to the faults being applied. The drivers power up in the idle state to ensure that no activity is placed on the twisted-pair cable that could be interpreted as network traffic.

The line receiver squelch function interfaces to a differential twisted-pair line terminated external to the device. The receiver squelch circuit allows differential receive signals to pass through as long as the input amplitude and pulse duration are greater than the minimum squelch threshold. This ensures a good signal-to-noise ratio while the data path is active and prevents system noise from causing false data transitions during line shutdown and line-idle conditions. The RXO outputs default to a high level and the RXEN outputs default to a low level while the squelch function is blocking the data path through the receiver (idle). The line receiver squelch will become active within 50 ns when the input squelch threshold is exceeded. RXEN will be driven high when the squelch circuit is allowing data to pass through the receiver. The receiver squelch circuit can also withstand a set of fault conditions while operating without causing permanent damage to the device.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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# SN75ALS085

## LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

### description (continued)

The purpose of the loop functions is to provide a means by which system data path verification can be done to isolate faulty interfaces and assist in network diagnosis. The LOOP pins are TTL compatible and must be held high for normal operation. When  $\overline{\text{LOOP1}}$  is taken low, the output of driver 1 (TXO1) immediately goes into the idle state. Also, the input to receiver 1 is ignored and a path from TXI1 to RXO1 is established. When  $\overline{\text{LOOP1}}$  is taken back high, driver 1 and receiver 1 revert back to their normal operation. When  $\overline{\text{LOOP2}}$  is taken low, a similar data path is established between TXI1 and RXO2. TXEN1 must be high for the loop functions to operate and TXEN1 can be used to gate the loop function if desired. During loop operation, the respective receiver enable output (RXEN) will reflect the status of TXEN1.

### Function Tables

RECEIVER -  $\overline{\text{LOOP}} = \text{H}$

RXI	PREVIOUS RXEN	OUTPUTS	
		RXEN	RXO
$V_{ID} = 1315 \text{ mV to } -175 \text{ mV}, t_W < 25 \text{ ns}$	L	L	H
$V_{ID} = -275 \text{ mV to } -1315 \text{ mV}, t_W > 50 \text{ ns}$	X	H	L
$V_{ID} = 318 \text{ mV to } 1315 \text{ mV}, t_W < 142 \text{ ns}$	H	H	H
$V_{ID} = 318 \text{ mV to } 1315 \text{ mV}, t_W > 187 \text{ ns}$	X	L	H

DRIVER -  $\overline{\text{LOOP}} = \text{H}$

TXI	TXEN	PREVIOUS TXO	OUTPUT TXO
L	L	Idle	Idle
H	L	Idle	Idle
↓	H	Idle	L
L	H	Active	L
$H < 260 \mu\text{s}$	H	Active	H
$H > 8 \mu\text{s}$	H	Active	Idle
L	$L > 8 \mu\text{s}$	Active	Idle
$H < 260 \text{ ns}$	$L > 8 \mu\text{s}$	Active	Idle
$H < 260 \text{ ns}$	$L < 260 \text{ ns}$	Active	H
$H > 8 \mu\text{s}$	$L < 260 \text{ ns}$	Active	Idle
L	$L < 260 \text{ ns}$	Active	L

$H = V_I \geq V_T \text{ max}, L = V_I \leq V_T \text{ min}$

### LOOP

INPUTS						OUTPUTS				
LOOP1	LOOP2	TXI1	TXEN1	RXI1	RXI2	RXO1	RXO2	RXEN1	RXEN2	TXO1
L	L	L	H	X	X	L	L	H	H	Idle
L	L	H	H	X	X	H	H	H	H	Idle
L	L	X	L	X	X	H	H	L	L	Idle
L	H	L	H	X	Normal	L	Normal	H	Normal	Idle
L	H	H	H	X	Normal	H	Normal	H	Normal	Idle
L	H	X	L	X	Normal	H	Normal	L	Normal	Idle
H	L	L	H	Normal	X	Normal	L	Normal	H	Idle
H	L	H	H	Normal	X	Normal	H	Normal	H	Idle
H	L	X	L	Normal	X	Normal	H	Normal	L	Idle
H	H	Normal	Normal	Normal	Normal	Normal	Normal	Normal	Normal	Normal

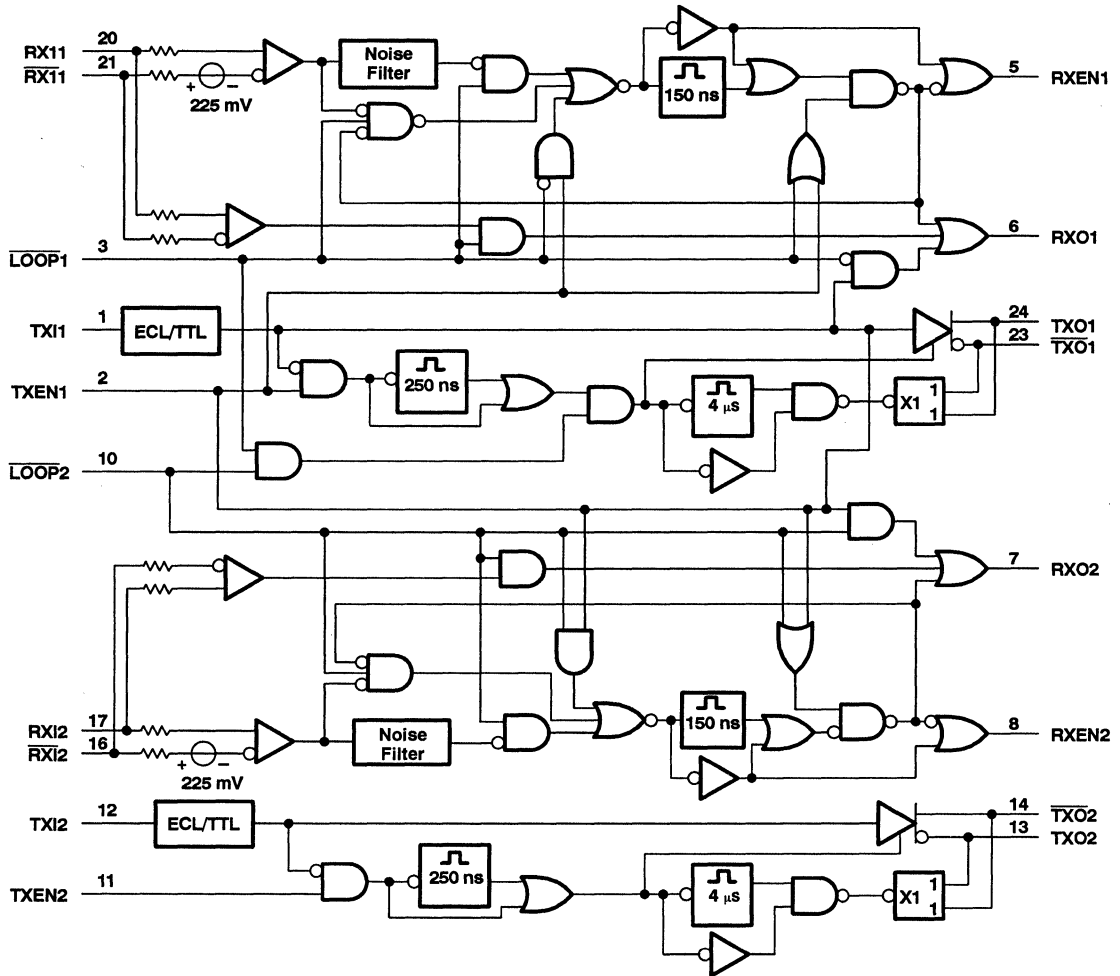
H = high level, L = low level, X = don't care



# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

## logic diagram (positive logic)



**TEXAS**  
**INSTRUMENTS**

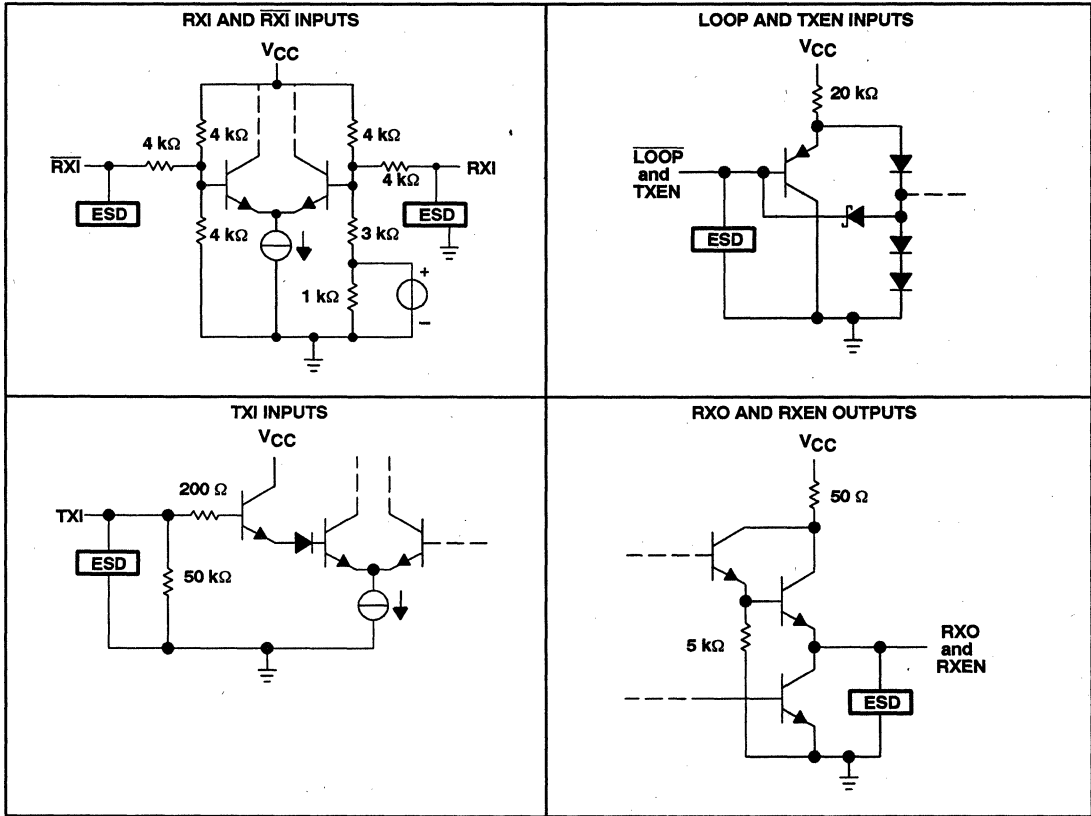
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# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

## schematics of inputs and outputs



# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	6 V
TXI and $\overline{LOOP}$ input voltage	5.5 V
TXO and $\overline{TXO}$ output voltage	16 V
RXI and $\overline{RXI}$ input voltage	16 V
RXO and RXEN output voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1350 mW	10.8 mW/°C	864 mW
NT	1250 mW	10.0 mW/°C	800 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode voltage at RXI inputs, $V_{IC}$	1		4.2	V
Differential voltage between RXI inputs, $V_{ID}$	± 318		± 1315	mV
High-level input voltage, $\overline{LOOP}$ and TXEN, $V_{IH}$	2			V
Low-level input voltage, $\overline{LOOP}$ and TXEN, $V_{IL}$			0.8	V
High-level output current, RXO and RXEN, $I_{OH}$			-0.4	mA
Low-level output voltage, RXO and RXEN, $I_{OL}$			16	mA
Setup time, driver mode, TXEN high before TXI↓, $t_{su1}$ (see Figure 8)	10			ns
Setup time, loop mode, $\overline{LOOP}$ low before TXEN↑, $t_{su2}$ (see Figure 10)	15			ns
Setup time, loop mode, TXEN high before TXI↓, $t_{su3}$ (see Figure 10)	10			ns
Hold time, loop mode, TXEN high after TXI↑, $t_{h1}$ (see Figure 9)	10			ns
Hold time, loop mode, $\overline{LOOP}$ low after TXEN↓, $t_{h2}$ (see Figure 9)	15			ns
Operating free-air temperature, $T_A$	0		70	°C



# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A – D3279, APRIL 1989 – REVISED FEBRUARY 1993

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT		
$V_{IK}$	Clamp voltage at all inputs	$I_I = -18 \text{ mA}$			-1.5	V		
$V_T$	Driver input (TXI) threshold voltage	$T_A = 0^\circ\text{C}$	$V_{CC} = 4.75 \text{ V}$	3.202	3.752	V		
			$V_{CC} = 5 \text{ V}$	3.389	3.998			
			$V_{CC} = 5.25 \text{ V}$	3.577	4.244			
		$T_A = 25^\circ\text{C}$	$V_{CC} = 4.75 \text{ V}$	3.213	3.797	V		
			$V_{CC} = 5 \text{ V}$	3.400	4.043			
			$V_{CC} = 5.25 \text{ V}$	3.588	4.289			
		$T_A = 70^\circ\text{C}$	$V_{CC} = 4.75 \text{ V}$	3.239	3.849	V		
			$V_{CC} = 5 \text{ V}$	3.426	4.095			
			$V_{CC} = 5.25 \text{ V}$	3.614	4.341			
$V_{IDT}$	Receiver differential input threshold voltage				-275	mV		
$V_{OC}$	Driver output (TXO) common-mode voltage	Idle	TXEN at 0.8 V, LOOP2 at 2 V, See Figure 1	LOOP1 at 2 V, See Figure 1	1	4.2	V	
		Active	TXEN at 2 V, LOOP2 at 2 V, See Figure 1	LOOP1 at 2 V, TXI at 3.2 V, See Figure 1	1	4.2		
		Active	TXEN at 2 V, LOOP2 at 2 V, See Figure 1	LOOP1 at 2 V, TXI at 4.4 V, See Figure 1	1	4.2		
$V_{OD}$	Driver output (TXO) differential voltage	Idle	TXEN at 0.8 V, LOOP2 at 2 V, See Figure 1	LOOP1 at 2 V, See Figure 1	$\pm 40$		mV	
		Active	TXEN at 2 V, LOOP2 at 2 V, See Figure 1	LOOP1 at 2 V, TXI at 3.2 V, See Figure 1	-600	1315		
		Active	TXEN at 2 V, LOOP2 at 2 V, See Figure 1	LOOP1 at 2 V, TXI at 4.4 V, See Figure 1	600	1315		
$V_{OH}$	High-level output voltage	RXO, RXEN	$I_{OH} = -0.4 \text{ mA}$		2.4		V	
$V_{OL}$	Low-level output voltage	RXO, RXEN	$I_{OL} = 16 \text{ mA}$			0.5	V	
$I_{IH}$	High-level input current	TXEN, LOOP	$V_I = 2 \text{ V}$			20	$\mu\text{A}$	
		TXI	$V_I = 4.5 \text{ V}$			400		
		RXI, RXI	$V_{ID} = -0.5 \text{ V}$ ,	$V_{IC} = 1 \text{ V to } 4.2 \text{ V}$				1000
$I_{IL}$	Low-level input current	TXEN, LOOP	$V_I = 0.8 \text{ V}$			-200	mA	
		TXI	$V_I = 3.1 \text{ V}$			100		
			$V_I = 0.3 \text{ V}$			4		10
		RXI, RXI	$V_{ID} = 0.5 \text{ V}$ ,	$V_{IC} = 1 \text{ V to } 4.2 \text{ V}$				1000
$I_{OD}$	Driver differential output current	Idle	TXEN at 0.8 V, LOOP2 at 2 V,	LOOP1 at 2 V, See Figure 2		$\pm 4$	mA	
$I_{OS}$	Short-circuit output current†	RXO, RXEN	$V_O$ at 0 V, RXI at 2 V	RXI at 3 V,		-40	-150	mA
$I_{CC}$	Supply current	LOOP2 at 2 V, TXI at 4.5 V,		TXEN at 2 V, Outputs open		225	mA	

† Not more than one output should be shorted at a time, and the duration of the test should not exceed 1 second.



# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted) (continued)**

PARAMETER	TEST CONDITIONS†	MIN	MAX	UNIT
Driver fault condition current	TXO shorted to $\overline{\text{TXO}}$ , Current measured in short		150	mA
	TXO at 0 V, $\overline{\text{TXO}}$ is open, Current measured at TXO		150	
	TXO is open, $\overline{\text{TXO}}$ at 0, Current measured at $\overline{\text{TXO}}$		150	
	TXO at 0 V, $\overline{\text{TXO}}$ at 0 V, Current measured at TXO and $\overline{\text{TXO}}$		150	
	TXO at 16 V, $\overline{\text{TXO}}$ is open, Current measured at TXO		150	
	TXO is open, $\overline{\text{TXO}}$ at 16 V, Current measured at TXO		150	
	TXO at 16 V, $\overline{\text{TXO}}$ at 16 V, Current measured at TXO and $\overline{\text{TXO}}$		150	
Receiver fault condition current	RXI shorted to $\overline{\text{RXI}}$ , Current measured in short		10	mA
	RXI at 0 V, $\overline{\text{RXI}}$ is open, Current measured at RXI		3	
	RXI is open, $\overline{\text{RXI}}$ at 0 V, Current measured at RXI		3	
	RXI at 0 V, $\overline{\text{RXI}}$ at 0 V, Current measured at RXI and $\overline{\text{RXI}}$		3	
	RXI at 16 V, $\overline{\text{RXI}}$ at open, Current measured at RXI		10	
	RXI at open, $\overline{\text{RXI}}$ at 16 V, Current measured at $\overline{\text{RXI}}$		10	
	RXI at 16 V, $\overline{\text{RXI}}$ at 16 V, Current measured at RXI and $\overline{\text{RXI}}$		10	

† Fault conditions should be measured on only one channel at a time.



# SN75ALS085

## LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A – D3279, APRIL 1989 – REVISED FEBRUARY 1993

switching characteristics over recommended ranges of supply voltage and operating free-air temperature

### driver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
$t_{PLH}$	TXI	TXO, $\overline{TXO}$	TXEN at 2 V, See Figure 3		15	ns
$t_{PHL}$	TXI	$\overline{TXO}$ , TXO	TXEN at 2 V, See Figure 3		15	ns
$t_{PIL}$	TXI	TXO, $\overline{TXO}$	TXEN at 2 V, See Figure 4		25	ns
$t_{PIL}$	TXEN	TXO, $\overline{TXO}$	TXI at 3.2 V, See Figure 5		25	ns
$t_w$		TXO, $\overline{TXO}$	TXEN at 2 V, See Figure 6	260	8000	ns
$V_{OD(U)}$	TXI	TXO, $\overline{TXO}$	TXEN at 2 V, See Figure 6		-100	mV
$t_{skew}$	TXI	TXO, $\overline{TXO}$	TXEN at 2 V, See Figure 3		$\pm 3$	ns
$t_r$			TXEN at 2 V, See Figure 3	1	5	ns
$t_f$			TXEN at 2 V, See Figure 3	1	5	ns

### receiver

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
$t_{PLH}$	$\overline{RXI}$ , RXI	R XO	$V_{IC} = 1\text{ V to }4.2\text{ V}$ , See Figure 11		15	ns
$t_{PHL}$	$\overline{RXI}$ , RXI	R XO	$V_{IC} = 1\text{ V to }4.2\text{ V}$ , See Figure 10		15	ns
$t_{PLH}$	$\overline{RXI}$ , RXI	R XEN	$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = -500\text{ mV}$ , See Figure 12		55	ns
$t_{PHL}$	$\overline{RXI}$ , RXI	R XEN	$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = 500\text{ mV}$ , See Figure 12	142	181	ns
$t_{skew}$	$\overline{RXI}$ , RXI	R XO	$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = 500\text{ mV}$ , See Figure 10		$\pm 3$	ns
$t_w$			$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = -175\text{ mV}$ , See Figure 11	25		ns
$t_w$			$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = -275\text{ mV}$ , See Figure 11		50	ns
$t_{r1}$			$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = \pm 500\text{ mV}$ , See Figure 10	1	8	ns
$t_{r2}$			$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = \pm 500\text{ mV}$ , See Figure 12	1	8	ns
$t_{f1}$			$V_{IC} = 1\text{ V to }4.2\text{ V}$ , $V_{ID} = \pm 500\text{ mV}$ , See Figure 10	1	8	ns
$t_{f2}$			$V_{IC} = 2.5\text{ V}$ , $V_{ID} = \pm 500\text{ V}$ , See Figure 12	1	8	ns
$t_{valid}$			See Figure 10	-10	15	ns



# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

**loop**

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	TXI	RXO	LOOP at 0.8 V, See Figure 13	TXEN at 2 V,	30 ns
t <sub>PHL</sub>	Propagation delay time, high-to-low level output	TXI	RXO	LOOP at 0.8 V, See Figure 13	TXEN at 2 V,	30 ns
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	TXEN	RXEN	LOOP at 0.8 V, See Figure 14		50 ns
t <sub>PHL</sub>	Propagation delay time, high-to-low level output	TXEN	RXEN	LOOP at 0.8 V, See Figure 14		50 ns

### PARAMETER MEASUREMENT INFORMATION

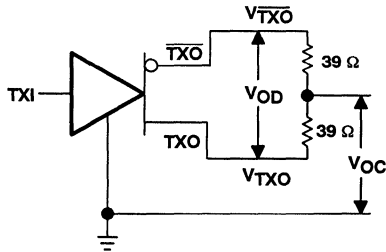


Figure 1

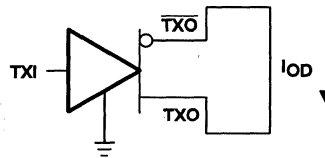
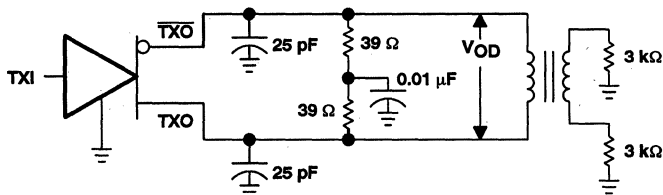


Figure 2

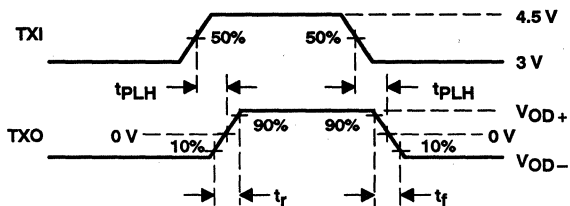
# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



### TEST CIRCUIT



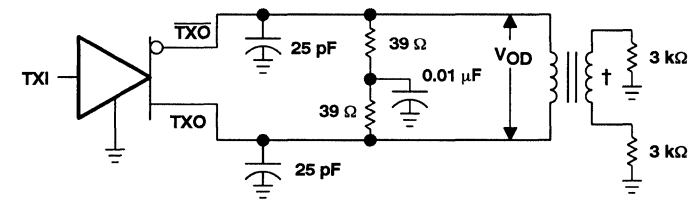
### VOLTAGE WAVEFORMS

### TRANSFORMER SPECIFICATIONS

Turns Ratio	1:1
Magnetizing Inductance	26 to 30 $\mu$ H
Winding Resistance	0.6 $\Omega$ Max
Rise Time 10% to 90%	5 ns Max
Interwinding Capacitance	25 pF
Leakage Inductance	0.25 $\mu$ H Max
Inductive Q	1250 Min

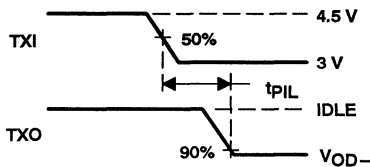
Figure 3. Test Circuit and Voltage Waveforms

**PARAMETER MEASUREMENT INFORMATION**



† See Figure 3

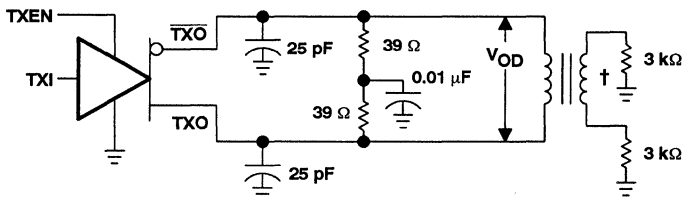
**TEST CIRCUIT**



**VOLTAGE WAVEFORMS**

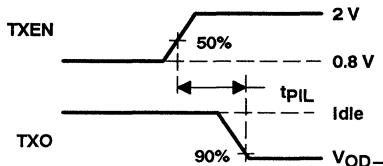
NOTE: Input  $t_r \leq 5$  ns;  $t_f \leq 5$  ns

**Figure 4. Test Circuit and Voltage Waveforms**



† See Figure 3

**TEST CIRCUIT**



**VOLTAGE WAVEFORMS**

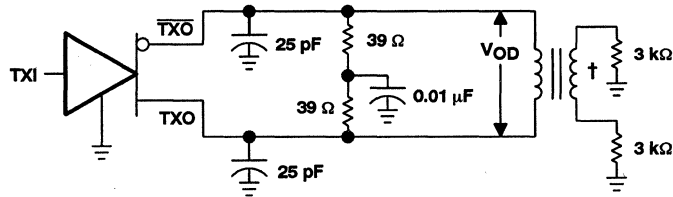
**Figure 5. Test Circuit and Voltage Waveforms**



**SN75ALS085**  
**LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER**

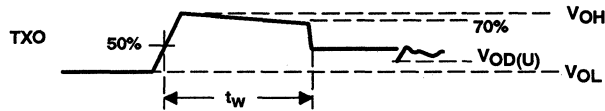
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**PARAMETER MEASUREMENT INFORMATION**



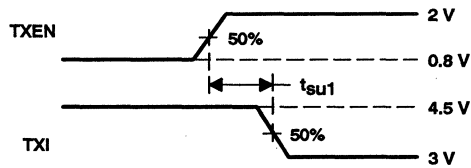
† See Figure 3

**TEST CIRCUIT**

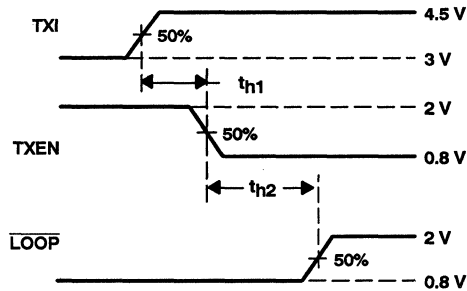


**VOLTAGE WAVEFORMS**

**Figure 6. Test Circuit and Voltage Waveforms**



**Figure 7**



**Figure 8**

NOTE: Input  $t_r \leq 5$  ns;  $t_f \leq 5$  ns

# SN75ALS085 LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

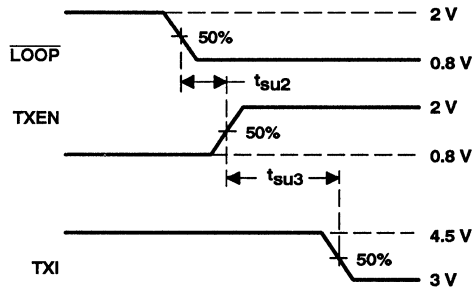
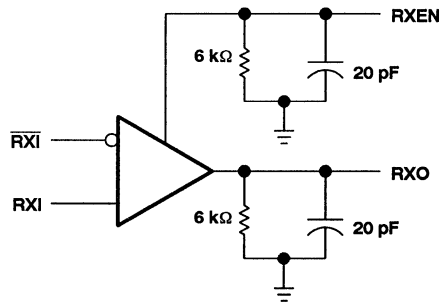


Figure 9



TEST CIRCUIT

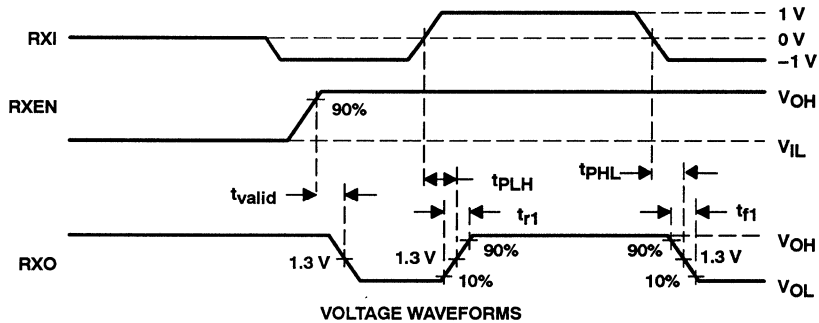


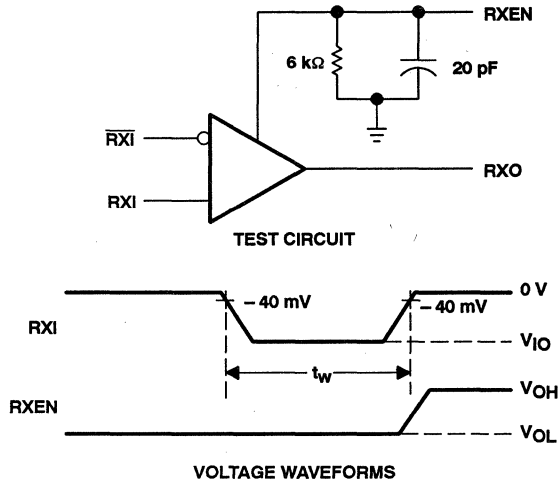
Figure 10. Test Circuit and Voltage Waveforms

NOTE: Input  $t_r \leq 5$  ns;  $t_f \leq 5$  ns

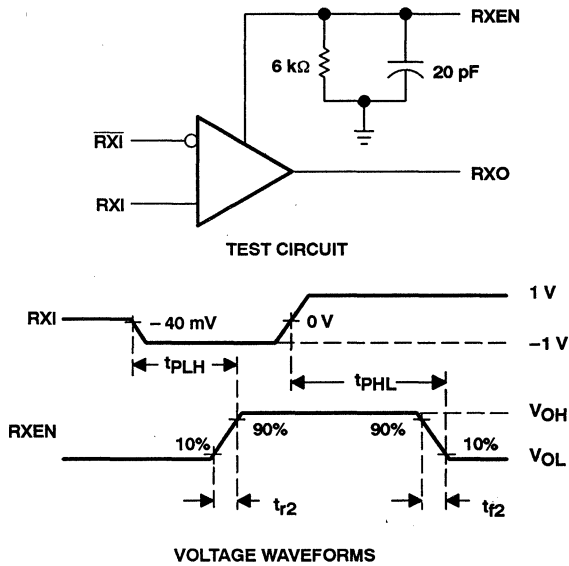
**SN75ALS085**  
**LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER**

SLLS054A - D3279, APRIL 1989 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



**Figure 11. Test Circuit and Voltage Waveforms**



**Figure 12. Test Circuit and Voltage Waveforms**

NOTE: Input  $t_r \leq 5$  ns;  $t_f \leq 5$  ns

**SN75ALS085**  
**LAN ACCESS UNIT INTERFACE DUAL DRIVER/RECEIVER**

SLLS054A – D3279, APRIL 1989 – REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**

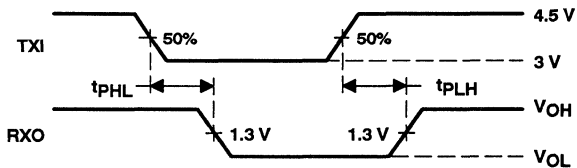


Figure 13

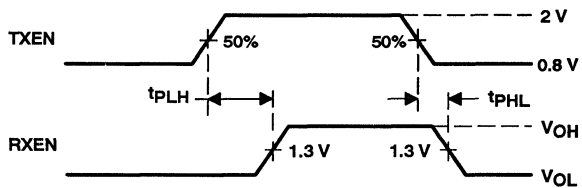


Figure 14

NOTE: Input  $t_r \leq 5$  ns;  $t_f \leq 5$  ns

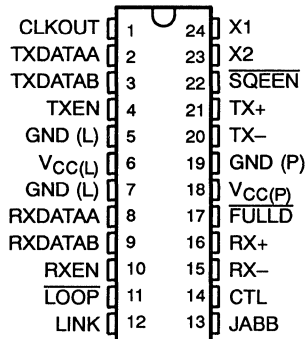


**SN75LBC086**  
**DIFFERENTIAL I/O DRIVER/RECEIVER PAIR**  
**WITH SQUELCH, JABBER CONTROL, AND COLLISION DETECTION**

SLLS120A—D3525, JUNE 1991—REVISED SEPTEMBER 1991

- **Compliant With IEEE STD 802.3I,  
Type 10BASE-T**
- **Differential (Twisted-Pair) I/O  
Driver/Receiver**
- **High-Speed Receiver . . .  $t_{pd} = 50$  ns Max**
- **Receiver Squelch Circuit Integrity Improved  
With Noise Filter**
- **Jabber Control Prevents Network Lockup**
- **Collision Detection for Multiple-User  
Networks**
- **Data Link Integrity Monitored With Link  
Test Pulse**
- **Externally Addressable Test Register  
Controls Signal Quality Error Testing**
- **CMOS and Raised ECL Compatible**
- **24-Pin, 300-mil Dual-In-Line Package**

**DW PACKAGE**  
(TOP VIEW)



**description**

The SN75LBC086 is a single-channel differential driver/receiver interface device for the medium attachment unit (MAU) used in 10-MHz twisted-pair Ethernet applications. The device uses a 5-V supply and is designed to interface with two pairs of telephone-grade twisted-pair cables coupled through isolation transformers. The functional components of the device include a differential receiver and driver, receiver squelch with noise filter, jabber controls, collision detection, data link monitor, and signal quality error (SQE) testing. The LinBiCMOS™ process technology is used in the device design to ensure analog precision, low power, and high-speed operation.

The device contains an elaborate receiver-squelch circuit† that provides an improved level of noise rejection by qualifying the incoming signal stream with three different criteria. First, the signal is compared to a set threshold voltage level. Then, the pulse duration is compared to a set time window. Last, the signal must follow a set pattern of positive and negative pulses before the circuit finally opens the receiver channel to the incoming data packet.

The jabber control is designed to prevent a defective controller from locking up the network by limiting the data packet transmission time to 20 to 30 ms. When a packet length exceeds 20 to 30 ms, the driver is turned off for about 600 ms. The driver-enable input must be made inactive by the controller during this period before the jabber control will release the driver. The JABB output is active (high) when a jabber condition exists.

Collision detection is used to arbitrate access to the multiuser network. This detection is done logically by monitoring the receive line for a valid signal during a driver transmission. When a collision is detected, this device informs the controller with an active-high CTL output. After a valid packet transmission, the device also performs a signal quality error test causing the CTL output to go active (high). This test is disabled when the SQUEEN input goes inactive (high).

The device tests data-link integrity during the idle state by periodically driving the driver line with a unipolar pulse called a link-test pulse. The receiver looks for this link-test pulse on the receive line. A failed line link is indicated by a high-impedance state at the LINK output. This output drives an LED for monitoring if needed.

† Embodies technology covered by one or more Digital Equipment Corporation Patents.  
 LinBiCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date.  
 Products conform to specifications per the terms of Texas Instruments  
 standard warranty. Production processing does not necessarily include  
 testing of all parameters.



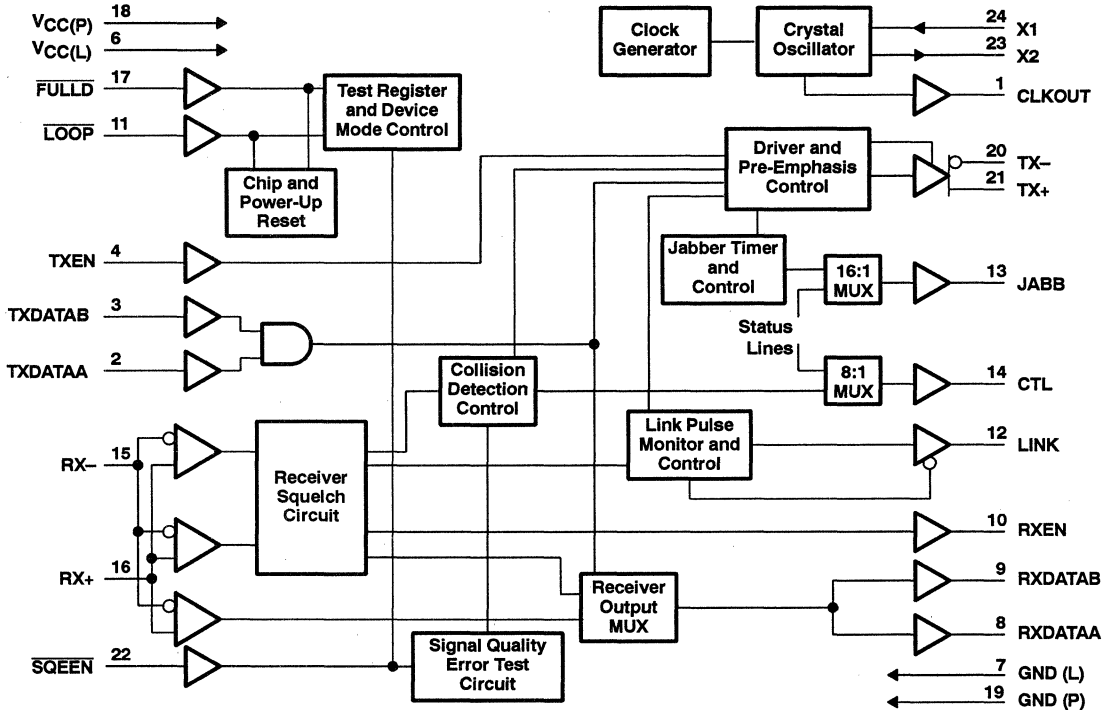
**SN75LBC086**  
**DIFFERENTIAL I/O DRIVER/RECEIVER PAIR**  
**WITH SQUELCH, JABBER CONTROL, AND COLLISION DETECTION**

SLLS120A - D3525, JUNE 1991 - REVISED SEPTEMBER 1991

**description (continued)**

An internal test register is externally controlled with inputs  $\overline{FULLD}$  and  $\overline{LOOP}$  to select the device testing mode. When in the test mode, serial test-mode control patterns are clocked into the test register through input  $\overline{SQEEN}$ . These control patterns select various modes to test the internal circuits.

**functional block diagram**



**SN75LBC086**  
**DIFFERENTIAL I/O DRIVER/RECEIVER PAIR**  
**WITH SQUELCH, JABBER CONTROL, AND COLLISION DETECTION**  
 SLLS120A – D3525, JUNE 1991 – REVISED SEPTEMBER 1991

**Terminal Functions**

NAME	PIN LEVEL	NO.	I/O	DESCRIPTION
CLKOUT	CMOS	1	O	Clock output. This 10-MHz buffered clock is provided for driving other interface devices.
CTL	CMOS	14	O	Control. In normal mode, CTL high indicates a collision. In test mode, status lines are muxed out.
FULLD	TTL	17	I	Full-duplex mode. When active (low), the device is placed in the full-duplex operating mode for simple point-to-point communication applications. In the full-duplex mode, the receiver and driver are both active with collision detection disabled. After LOOP and FULLD go active (low), in that order, a device reset is initiated and while both are active (low), test select data is clocked into the test register using a 100-ns clock at the X1 input. This terminal is held inactive (high) due to an internal pullup resistor.
GND (L)	GROUND	5 7		Logic grounds. These terminals provide a ground return for the CMOS core logic.
GND (P)	GROUND	19		Power ground. This provides a ground return for the input and output buffers, driver (transmitter), and receiver circuits.
JABB	CMOS	13	O	Jabber control. When a jabber condition exists during normal mode operation, this signal goes active (high) to report jabber-control status to the controller. In the test mode, this provides a multiplexed signal for internal timer and counter functions.
LINK	CMOS	12	O	Link status. This 3-state output indicates the status of the receiver and interface link. When driving an LED (with anode to resistor to VCC), a high-impedance level indicates a failed link and the LED is off. A momentary high level indicates the device is receiving valid data and the LED is blinking on and off. A continuous low level indicates the device is receiving valid link pulses but no data, and the LED is on.
LOOP	TTL	11	I	Loop-back mode. When the device is in the normal operating mode (not test mode) and LOOP is active (low), the driver (transmit) data is directed to the receive data path to put the device in the loop-back mode and the driver is turned off. After LOOP and FULLD go active (low), in that order, a device reset is initiated and while both are active (low), test select data is clocked into the test register using a 100-ns clock at the X1 input. This terminal is held inactive (high) due to an internal pullup resistor.
RX+		16	I	Differential receiver inputs
RX-		15	I	
RXEN	CMOS	10	O	Receiver squelch status. This provides squelch status information to the controller. When active (high), this signal indicates that the data path is valid or open from the receive channel through the device. An inactive (low) indicates that the receive channel is squelched or closed. This signal is capable of driving an LED monitor.
RXDATAA	CMOS	8	O	Received-data serial outputs. These provide a choice of logic levels and serial data either from the differential receiver input (RX+ and RX-) or data from the controller (TXDATAA or TXDATAB) when in the loop-back mode. When the receiver is idle, these output levels are normally high. These terminals are held inactive (high) due to an internal pullup resistor.
RXDATAB	ECL	9	O	
SQEN	TTL	22	I	Signal-quality error-test enable. In normal operating mode, this enables the SQE test function performed at the end of a data packet transmission. In the test mode, SQEN is used (with X1 clock) as a serial data input port to load test patterns or selections into the test register. This terminal is held inactive (high) due to an internal pullup resistor.
TX+		21	O	Differential driver outputs
TX-		20	O	
TXEN	TTL	4	I	Transmitter (driver) enable. When TXEN is active (high), serial data at the TXDATA inputs starts and stops the driver. When TXEN is inactive (low), the driver begins transmitting an idle signal independent of the TXDATA inputs.
TXDATAA	CMOS	2	I	Transmit-data inputs. A choice of logic-level inputs provide Manchester-encoded serial data to the driver. Internal pullup resistors are included.
TXDATAB	ECL	3	I	
VCC(L)	SUPPLY	6		VCC logic power supply. This provides power to the CMOS core logic.
VCC(P)	SUPPLY	18		VCC power supply. This provides power to the input and output buffers, drivers, and receivers.
X1	CMOS	24	I	Crystal input/output. X1 provides an input from an external 10-MHz crystal or another external clock source if the crystal is disconnected. X2 provides an oscillator output.
X2		23	O	



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### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	-0.5 V to 7 V
Input voltage range at any input, $V_I$	-0.5 V to 5.5 V
Output voltage range at any output, $V_O$	-0.5 V to 7 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to device ground pins GND(L) and GND(P) shorted together.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1350 mW	10.8 mW/°C	864 mW

### recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level output voltage, $V_{IH}$	TXDATAA, X1	3.15			V
	TXDATAB (see Figure 1)	$T_A = 0^\circ\text{C}$	$0.984V_{CC} - 0.922$	$0.984V_{CC} - 0.763$	
		$T_A = 25^\circ\text{C}$	$0.984V_{CC} - 0.877$	$0.984V_{CC} - 0.727$	
		$T_A = 70^\circ\text{C}$	$0.984V_{CC} - 0.825$	$0.984V_{CC} - 0.645$	
TXEN, LOOP, FULLD, SQUEEN	2				
Low-level output voltage, $V_{IL}$	TXDATAA, X1	0.8			V
	TXDATAB (see Figure 1)	$T_A = 0^\circ\text{C}$	$0.75V_{CC} - 0.590$	$0.750V_{CC} - 0.375$	
		$T_A = 25^\circ\text{C}$	$0.75V_{CC} - 0.550$	$0.750V_{CC} - 0.350$	
		$T_A = 70^\circ\text{C}$	$0.75V_{CC} - 0.531$	$0.750V_{CC} - 0.324$	
TXEN, LOOP, FULLD, SQUEEN	0.8				
Differential input voltage, $V_{ID}$		0.586		2.8	V
Common-mode input voltage, $V_{IC}$		1.8		3.2	V
Operating free-air temperature, $T_A$		0		70	°C



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electrical characteristics over recommended operating free-air temperature and supply voltage range (unless otherwise noted)

**drivers**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
VOH	High-level output voltage	CLKOUT, RXDATAA, RXEN, JABB, CTL	IOH = -12 mA		3.7	V
		RXDATAB	See Figure 1	TA = 0°C	0.984 VCC - 0.922	0.984 VCC - 0.763
			TA = 25°C	0.984 VCC - 0.877	0.984 VCC - 0.727	
				TA = 70°C	0.984 VCC - 0.825	0.984 VCC - 0.645
VOL	Low-level output voltage	CLKOUT, RXDATAA, RXEN, JABB, CTL	IOL = 16 mA		0.5	V
		RXDATAB	See Figure 1	TA = 0°C	0.750 VCC - 0.590	0.750 VCC - 0.375
			TA = 25°C	0.750 VCC - 0.550	0.750 VCC - 0.350	
			TA = 70°C	0.750 VCC - 0.531	0.750 VCC - 0.324	
		LINK	IOL = 12 mA		0.5	V
VOD	Differential-output voltage (peak)	See Figure 2	2.2		2.8	V
VOD	Differential-output voltage (step)	See Figure 2	1.53		1.982	V
	Common-mode driver impedance	TX+, TX-	2	5	8	Ω

**receivers**

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
IIH	High-level input current	TXDATAA, TXEN, LOOP, FULLD, SQUEEN	VI = 5.25 V		20	μA
		X1			100	
		TXDATAB	VIH = MAX		400	
IIL	Low-level input current	TXDATAA, TXEN, LOOP, FULLD, SQUEEN	VI = 0		-20	μA
		X1			-100	
		TXDATAB	VIL = MIN		-400	

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

**drivers and receivers**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ICC	Supply current	VCC(L), VCC(P)	VCC(L) = 5.25 V, VCC(P) = 5.25 V		180	mA



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#### switching characteristics

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{pd1}$	Propagation delay time	RX+, RX-	RXEN	See Figure 4			5 bit times	
$t_{pd2}$	Propagation delay time at startup	RX+, RX-	RXDATAA or RXDATAB high	See Figure 4		75		ns
$t_{sk(o)}$	Output skew time	RXEN high	RXDATAA or RXDATAB low	See Figure 4		±10		ns
$t_{pd3}$	Propagation delay time after startup	RX+, RX-	RXDATAA or RXDATAB high	See Figure 4		50		ns
$t_{sk(p)}$	Pulse skew time ( $t_{PLH4} - t_{PHL4}$ )	RX+, RX-	RXDATAA or RXDATAAB	See Figure 4		2		ns
$t_{pd4}$	Propagation delay time	RX+, RX-	RXEN low	See Figure 5	155		250	ns
$t_{pd5}$	Propagation delay time	TXDATA or TXDATAB	TX+, TX-	See Figure 6			75	ns
$t_{sk(p)}$	Pulse skew time ( $t_{PLH5} - t_{PHL5}$ )	TXDATAA or TXDATAB	TX+, TX-	See Figure 6		2		ns
$t_{pd6}$	Propagation delay time in loop mode	TXDATAA or TXDATAB	RXDATAA, RXDATAB	See Figure 7			50	ns
$t_{pd7}$	Propagation delay time in loop mode	TXEN high	RXEN high	See Figure 7			50	ns
$t_{pd8}$	Propagation delay time in loop mode	LOOP low	RXEN low	See Figure 7			30	ns
$t_{pd10}$	Propagation delay time	TXEN low	RXEN low	See Figure 8			350	ns
$t_{pd11}$	Propagation delay time	TXEN low	TX+, TX- high	See Figure 8			50	ns
$t_{p1}$	Precompensation pulse duration		TX+, TX-	See Figure 6	45		55	ns
$t_{p2}$	Receiver link-beat minimum pulse duration			See Figure 9	80		120	ns
$t_{en1}$	Enable time	TXDATAA or TXDATAB	TX+, TX-	See Figure 6			75	ns
$t_{en2}$	Enable time	TXEN	TX+, TX-	See Figure 6			75	ns
$t_{dis1}$	Disable time, caused by TXDATAA or TXDATAB high or TXEN low	TX+, TX- high	TX+, TX- at 585-mV level	See Figure 8	250			ns
$t_{pd12}$	Propagation delay time to looped RXEN	TXEN high	RXEN high	See Figure 6			100	ns
$t_{pd13}$	Propagation delay time for looped back data	TXDATAA or TXDATAB	RXDATAA or RXDATAB	See Figure 6			75	ns

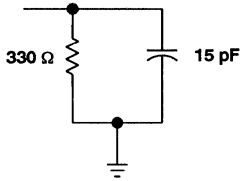
#### timing requirements

	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Setup time, test mode, SQUEEN before X1↑, $t_{su1}$	See Figure 10	30			ns
Setup time, test mode, LOOP low before FULLD↓, $t_{su2}$	See Figure 10	25			ns
Hold time, test mode, SQUEEN after X1↑, $t_{h1}$	See Figure 10	25			ns

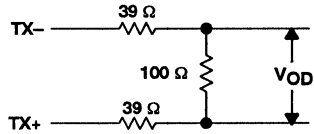


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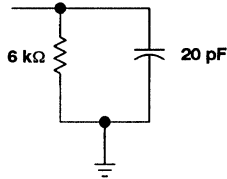
**PARAMETER MEASUREMENT INFORMATION**



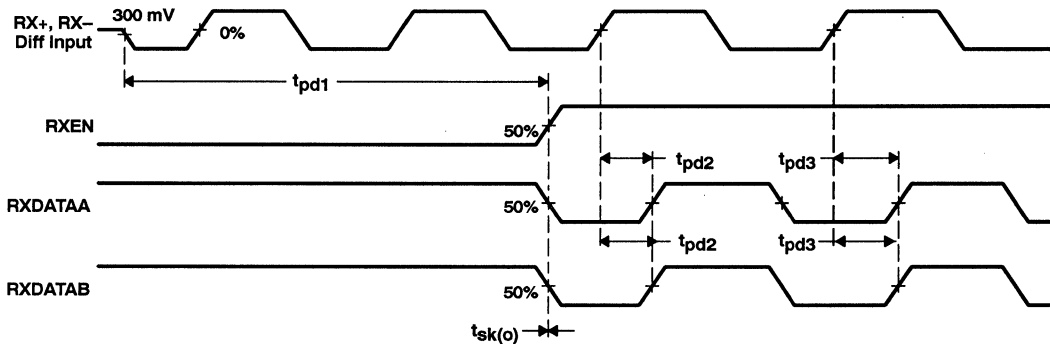
**Figure 1. ECL Load Circuit**



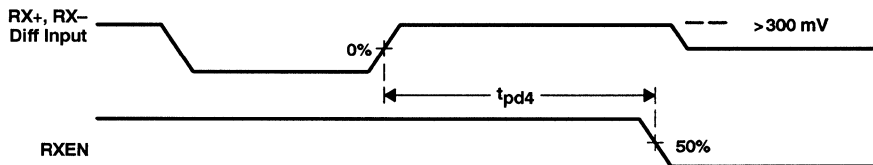
**Figure 2. Differential Load Circuit**



**Figure 3. CMOS Load Circuit**



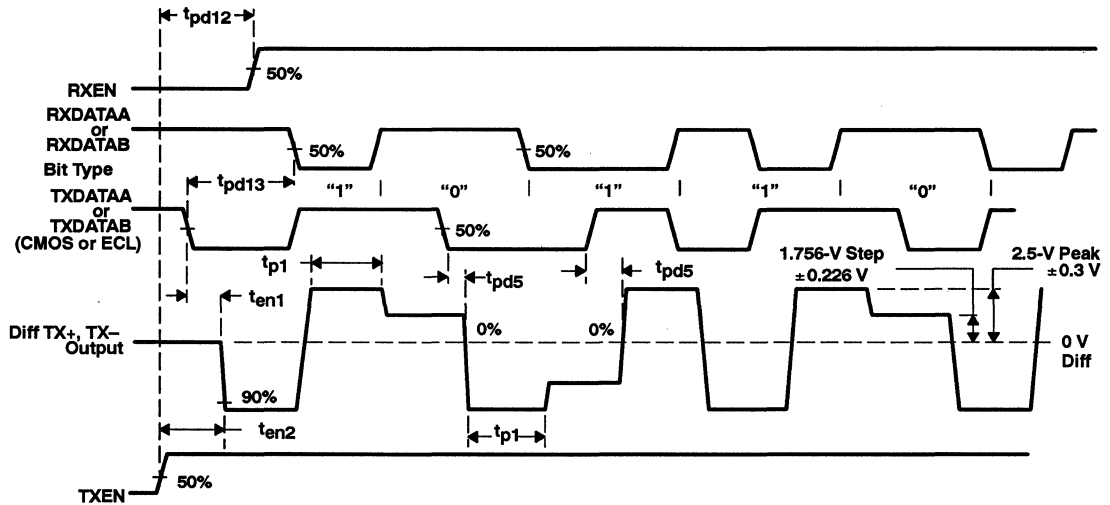
**Figure 4. Receiver Startup Waveforms**



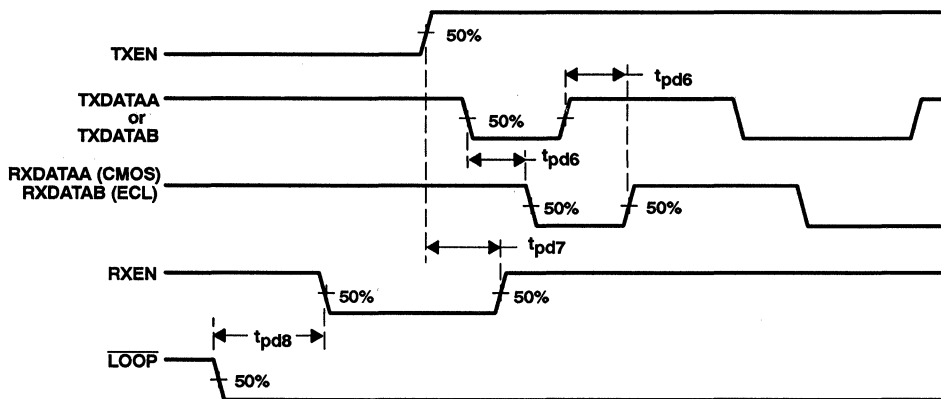
**Figure 5. Receiver Shutdown Waveforms**

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**PARAMETER MEASUREMENT INFORMATION**



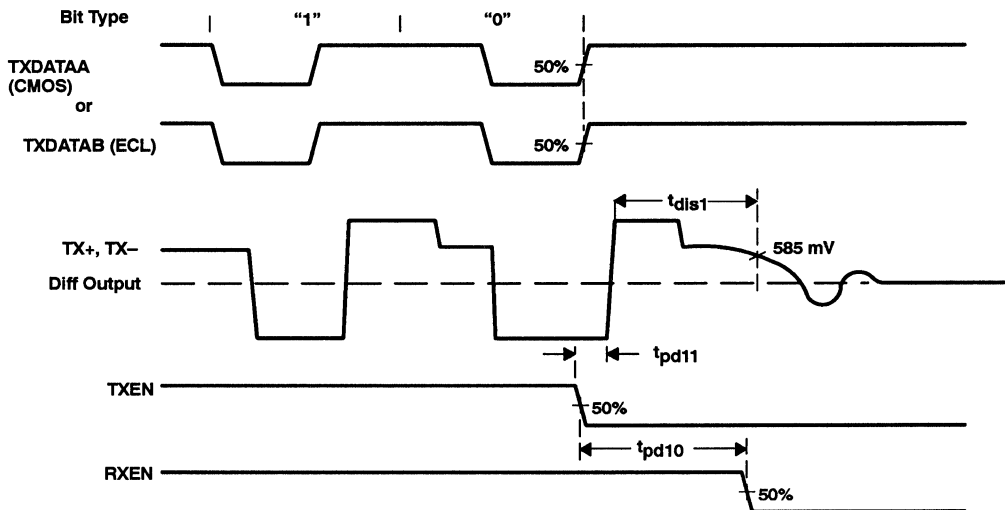
**Figure 6. Driver Startup Waveforms**



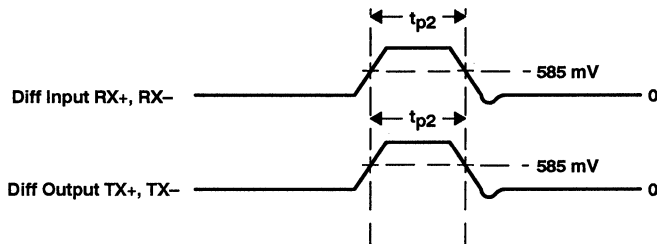
**Figure 7. Propagation Delay Waveforms in Loop Mode**

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**PARAMETER MEASUREMENT INFORMATION**



**Figure 8. Driver Shutdown Waveforms**



**Figure 9. Link Beat Pulse Duration Waveform**

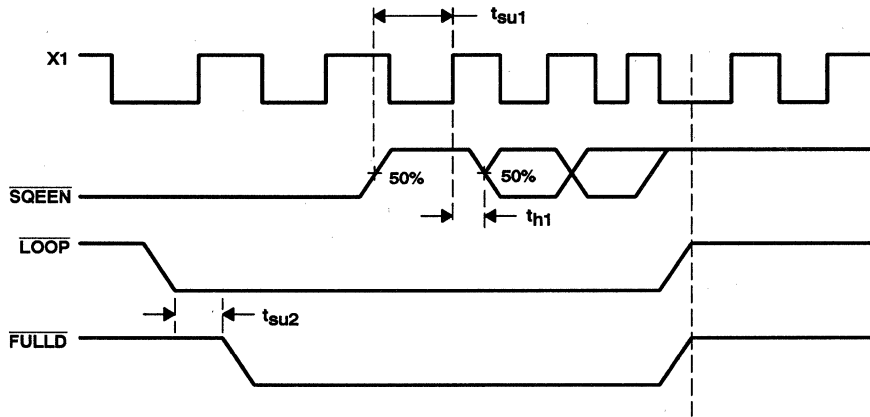
**SN75LBC086**

**DIFFERENTIAL I/O DRIVER/RECEIVER PAIR**

**WITH SQUELCH, JABBER CONTROL, AND COLLISION DETECTION**

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**PARAMETER MEASUREMENT INFORMATION**



**Figure 10. Setup and Hold Time Waveforms**

# SN55107A, SN55107B, SN55108A, SN55108B SN75107A, SN75107B, SN75108A, SN75108B DUAL LINE RECEIVERS

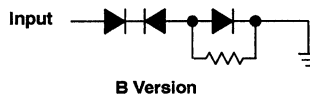
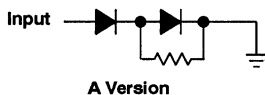
SLLS069A – D2304, JANUARY 1977 – REVISED JANUARY 1993

- High Speed
- Standard Supply Voltage
- Dual Channels
- High Common-Mode Rejection Ratio
- High Input Impedance
- High Input Sensitivity
- Differential Common-Mode Input Voltage Range of  $\pm 3$  V
- Strobe Inputs for Receiver Selection
- Gate Inputs for Logic Versatility
- TTL Drive Capability
- High dc Noise Margin
- '107A and '107B Have Totem-Pole Outputs
- '108A and '108B Have Open-Collector Outputs
- B Versions Have Diode-Protected Input for Power-Off Condition

## description

These circuits are TTL-compatible high-speed line receivers. Each is a monolithic dual circuit featuring two independent channels. They are designed for general use as well as such specific applications as data comparators and balanced, unbalanced, and party-line transmission systems. These devices are unilaterally interchangeable with and are replacements for the SN55107, SN55108, SN75107, and SN75108, but offer diode-clamped strobe inputs to simplify circuit design.

The essential difference between the A and B versions can be seen in the schematics. Input-protection diodes are in series with the collectors of the differential-input transistors of the B versions. These diodes are useful in certain party-line systems that may have multiple  $V_{CC+}$  power supplies and may be operated with some of the  $V_{CC+}$  supplies turned off. In such a system, if a supply is turned off and allowed to go to ground, the equivalent input circuit connected to that supply would be as follows:

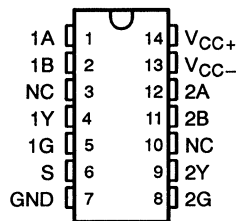


This would be a problem in specific systems that might possibly have the transmission lines biased to some potential greater than 1.4 V.

The SN55107A, SN55107B, SN55108A, and SN55108B are characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75107A, SN75107B, SN75108A, and SN75108B are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

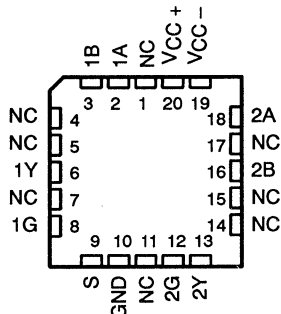
SN55107A, SN55107B, SN55108A,  
SN55108B ... J OR W PACKAGE  
SN75107A, SN75107B, SN75108A,  
SN75108B ... D, J, OR N PACKAGE

(TOP VIEW)



SN55107A, SN55107B, SN55108A,  
SN55108B ... FK PACKAGE

(TOP VIEW)



NC – No internal connection

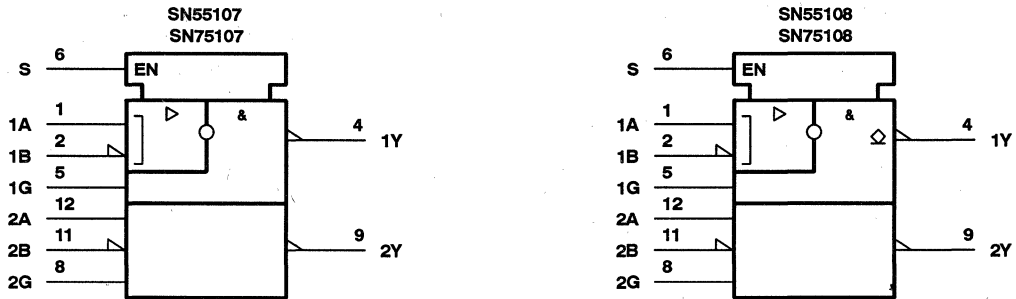
**THE SN75108B IS NOT  
RECOMMENDED FOR NEW DESIGN**



**SN55107A, SN55107B, SN55108A, SN55108B  
SN75107A, SN75107B, SN75108A, SN75108B  
DUAL LINE RECEIVERS**

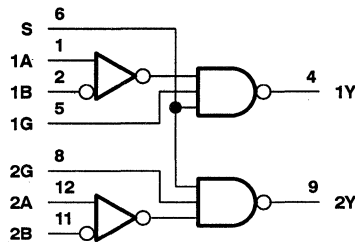
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**logic symbols†**



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12. Pin numbers shown are for the D, J, N, and W packages.

**logic diagram (positive logic)**



**FUNCTION TABLE**

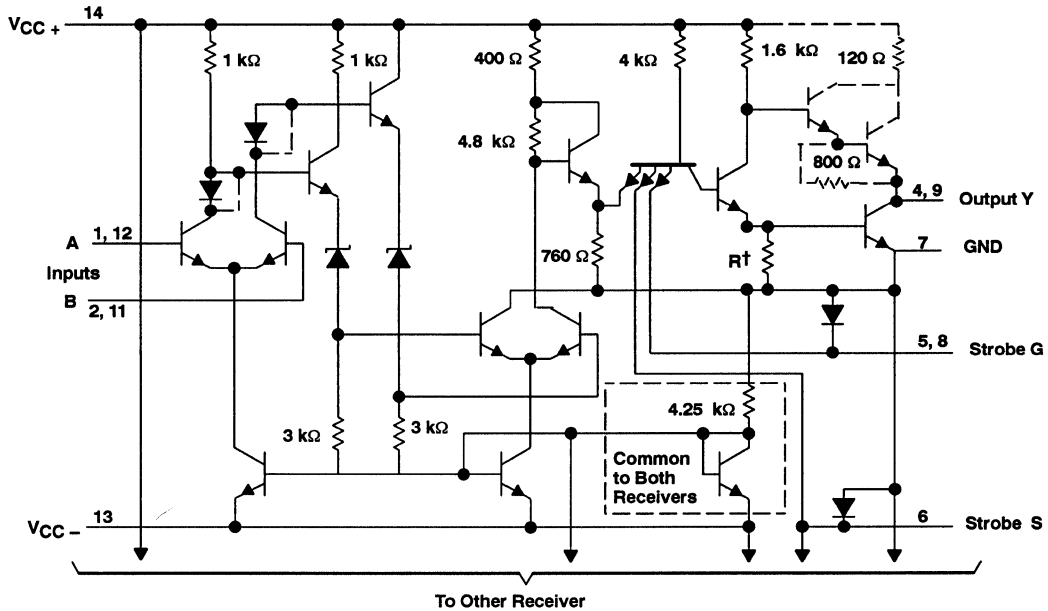
DIFFERENTIAL INPUTS A - B	STROBES		OUTPUT Y
	G	S	
$V_{ID} \geq 25 \text{ mV}$	X	X	H
$-25 \text{ mV} < V_{ID} < 25 \text{ mV}$	X	L	H
	L	X	H
	H	H	Indeterminate
$V_{ID} \leq -25 \text{ mV}$	X	L	H
	L	X	H
	H	H	L

H = high level, L = low level, X = irrelevant

SN55107A, SN55107B, SN55108A, SN55108B  
 SN75107A, SN75107B, SN75108A, SN75108B  
**DUAL LINE RECEIVERS**

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**schematic (each receiver)**



Pin numbers shown are for D, J, N, and W packages.

† R = 1 kΩ for '107A and '107B, 750 Ω for '108A and '108B.

NOTES: 1. Resistor values shown are nominal.

2. Components shown with dashed lines in the output circuitry are applicable to the '107A and '107B only. Diodes in series with the collectors of the differential input transistors are short circuited on '107A and '108A.

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC+}$ (see Note 3)	7 V
Supply voltage, $V_{CC-}$	-7 V
Differential input voltage (see Note 4)	±6 V
Common-mode input voltage (see Note 5)	±5 V
Strobe input voltage	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN55'	-55°C to 125°C
SN75'	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or W package	260°C

NOTES: 3. All voltage values, except differential voltages, are with respect to network ground terminal.

4. Differential voltage values are at the noninverting (A) terminal with respect to the inverting (B) terminal.

5. Common-mode input voltage is the average of the voltages at the A and B inputs.



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**SN55107A, SN55107B, SN55108A, SN55108B**  
**SN75107A, SN75107B, SN75108A, SN75108B**  
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**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J (SN5510_A,B)	1375 mW	11.0 mW/°C	880 mW	275 mW
J (SN7510_A,B)	1025 mW	8.2 mW/°C	656 mW	—
N	1150 mW	9.2 mW/°C	736 mW	—
W	1000 mW	8.0 mW/°C	640 mW	200 mW

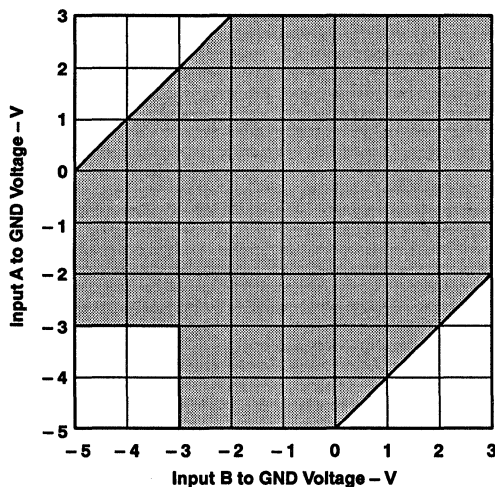
**recommended operating conditions (see Note 6)**

	SN55107A, SN55107B SN55108A, SN55108B			SN75107A, SN75107B SN75108A, SN75108B			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
	Supply voltage, $V_{CC+}$	4.5	5	5.5	4.75	5	
Supply voltage, $V_{CC-}$	-4.5	-5	-5.5	-4.75	-5	-5.25	V
High-level input voltage between differential inputs, $V_{IDH}$ (see Note 7)	0.025		5	0.025		5	V
Low-level input voltage between differential inputs, $V_{IDL}$ (see Note 7)	-5†		-0.025	-5†		-0.025	V
Common-mode input voltage, $V_{IC}$ (see Notes 7 and 8)	-3†		3	-3†		3	V
Input voltage, any differential input to GND (see Note 8)	-5†		3	-5†		3	V
High-level input voltage at strobe inputs, $V_{IH(S)}$	2		5.5	2		5.5	V
Low-level input voltage at strobe inputs, $V_{IL(S)}$	0		0.8	0		0.8	V
Low-level output current, $I_{OL}$			-16			-16	mA
Operating free-air temperature, $T_A$	-55		125	0		70	°C

† The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for input voltage levels only.

- NOTES: 6. When using only one channel of the line receiver, the strobe G of the unused channel should be grounded and at least one of the differential inputs of the unused receiver should be terminated at some voltage between -3 V and 3 V.  
 7. The recommended combinations of input voltages fall within the shaded area of the figure shown.  
 8. The common-mode voltage may be as low as -4 V provided that the more positive of the two inputs is not more negative than -3 V.

**RECOMMENDED COMBINATIONS  
OF INPUT VOLTAGES**



**TEXAS**   
**INSTRUMENTS**

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**SN55107A, SN55107B, SN55108A, SN55108B  
SN75107A, SN75107B, SN75108A, SN75108B  
DUAL LINE RECEIVERS**

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**electrical characteristics over recommended free-air temperature range (unless otherwise noted)**

PARAMETER		TEST CONDITIONS†		'107A, '107B			'108A, '108B			UNIT
				MIN	TYP‡	MAX	MIN	TYP‡	MAX	
V <sub>OH</sub>	High-level output voltage	V <sub>CC±</sub> = MIN, V <sub>I(L)(S)</sub> = 0.8 V, V <sub>IDH</sub> = 25 mV, I <sub>OH</sub> = -400 μA, V <sub>IC</sub> = -3 V to 3 V		2.4						V
V <sub>OL</sub>	Low-level output voltage	V <sub>CC±</sub> = MIN, V <sub>I(H)(S)</sub> = 2 V, V <sub>IDL</sub> = -25 mV, I <sub>OL</sub> = 16 mA, V <sub>IC</sub> = -3 V to 3 V		0.4			0.4			V
I <sub>IH</sub>	High-level input current	A	V <sub>CC±</sub> = MAX	V <sub>ID</sub> = 5 V	30	75	30	75	μA	
		B		V <sub>ID</sub> = -5 V	30	75	30	75		
I <sub>IL</sub>	Low-level input current	A	V <sub>CC±</sub> = MAX	V <sub>ID</sub> = -5 V	-10		-10		μA	
		B		V <sub>ID</sub> = 5 V	-10		-10			
I <sub>IH</sub>	High-level input current into 1G or 2G	V <sub>CC±</sub> = MAX, V <sub>I(H)(G)</sub> = 2.4 V		40		40		μA		
		V <sub>CC±</sub> = MAX, V <sub>I(H)(G)</sub> = MAX V <sub>CC+</sub>		1		1		mA		
I <sub>IL</sub>	Low-level input current into 1G or 2G	V <sub>CC±</sub> = MAX, V <sub>I(L)(G)</sub> = 0.4 V		-1.6		-1.6		mA		
		V <sub>CC±</sub> = MAX, V <sub>I(H)(S)</sub> = 2.4 V		80		80		μA		
V <sub>CC±</sub> = MAX, V <sub>I(H)(S)</sub> = MAX V <sub>CC+</sub>		2		2		mA				
I <sub>IL</sub>	Low-level input current into S	V <sub>CC±</sub> = MAX, V <sub>I(L)(S)</sub> = 0.4 V		-3.2		-3.2		mA		
		V <sub>CC±</sub> = MIN, V <sub>OH</sub> = MAX V <sub>CC+</sub>		250				μA		
I <sub>OS</sub>	Short-circuit output current§	V <sub>CC±</sub> = MAX		-18	-70			mA		
I <sub>CCH+</sub>	Supply current from V <sub>CC+</sub> , outputs high	V <sub>CC±</sub> = MAX, T <sub>A</sub> = 25°C		18	30	18	30	mA		
I <sub>CCH-</sub>	Supply current from V <sub>CC-</sub> , outputs high	V <sub>CC±</sub> = MAX, T <sub>A</sub> = 25°C		-8.4	-15	-8.4	-15	mA		

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at V<sub>CC+</sub> = 5 V, V<sub>CC-</sub> = -5 V, T<sub>A</sub> = 25°C.

§ Not more than one output should be shorted at a time.

**switching characteristics, V<sub>CC±</sub> = ±5 V, T<sub>A</sub> = 25°C, R<sub>L</sub> = 390 Ω (see Figure 1)**

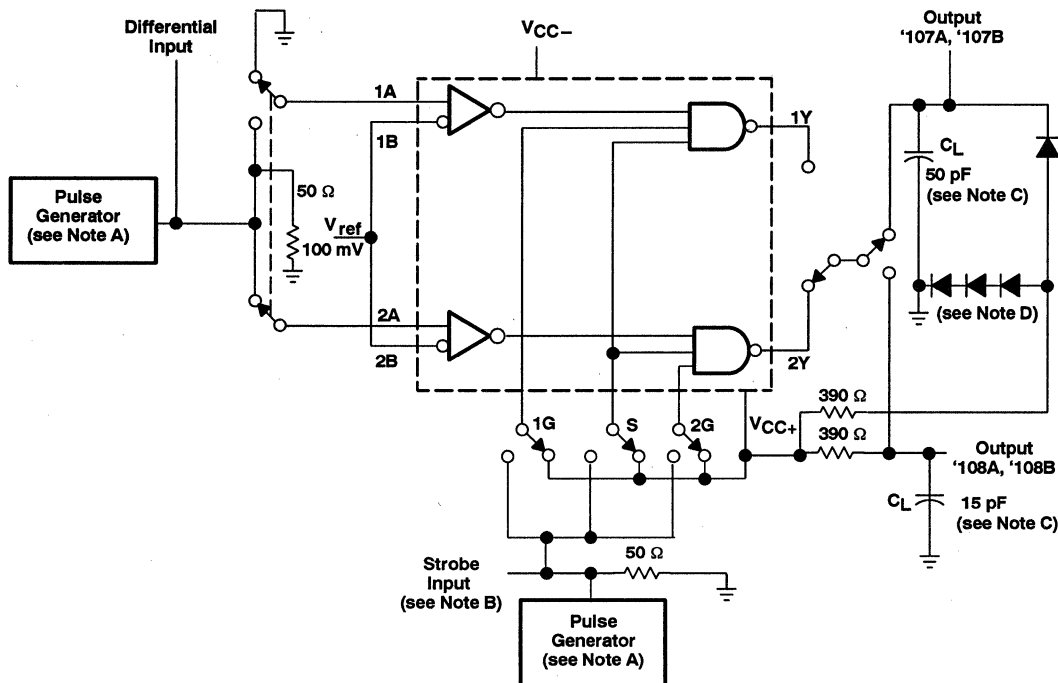
PARAMETER		TEST CONDITIONS	'107A, '107B			'108A, 108B			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PLH(D)</sub>	Propagation delay time, low-to-high-level output, from differential inputs A and B	C <sub>L</sub> = 50 pF	17		25				ns
		C <sub>L</sub> = 15 pF				19	25		
t <sub>PHL(D)</sub>	Propagation delay time, high-to-low-level output, from differential inputs A and B	C <sub>L</sub> = 50 pF	17		25				ns
		C <sub>L</sub> = 15 pF				19	25		
t <sub>PLH(S)</sub>	Propagation delay time, low-to-high-level output, from strobe input G or S	C <sub>L</sub> = 50 pF	10		15				ns
		C <sub>L</sub> = 15 pF				13	20		
t <sub>PHL(S)</sub>	Propagation delay time, high-to-low-level output, from strobe input G or S	C <sub>L</sub> = 50 pF	8		15				ns
		C <sub>L</sub> = 15 pF				13	20		



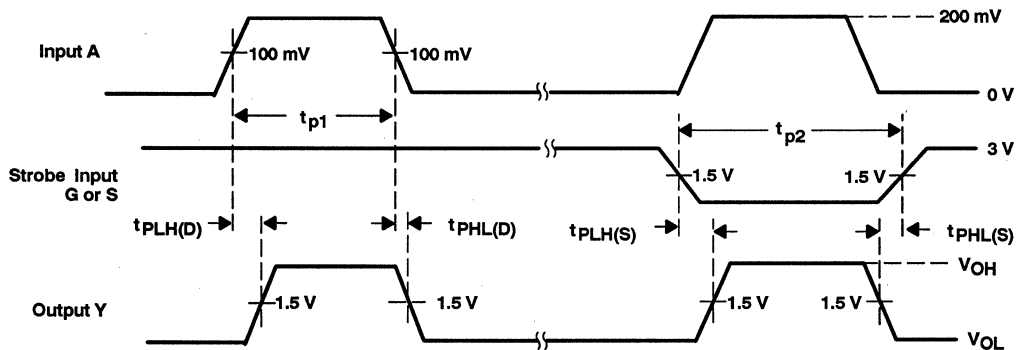
**SN55107A, SN55107B, SN55108A, SN55108B  
SN75107A, SN75107B, SN75108A, SN75108B  
DUAL LINE RECEIVERS**

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**PARAMETER MEASUREMENT INFORMATION**



**TEST CIRCUIT**



**VOLTAGE WAVEFORMS**

- NOTES: A. The pulse generators have the following characteristics:  $Z_0 = 50 \Omega$ ,  $t_r = 10 \pm 5 \text{ ns}$ ,  $t_f = 10 \pm 5 \text{ ns}$ ,  $t_{pd1} = 500 \text{ ns}$ ,  $\text{PRR} \leq 1 \text{ MHz}$ ,  $t_{pd2} = 1 \mu\text{s}$ ,  $\text{PRR} \leq 500 \text{ kHz}$ .
- B. Strobe input pulse is applied to Strobe 1G when inputs 1A-1B are being tested, to Strobe S when inputs 1A-1B or 2A-2B are being tested, and to Strobe 2G when inputs 2A-2B are being tested.
- C.  $C_L$  includes probe and jig capacitance.
- D. All diodes are 1N916.

**Figure 1. Test Circuit and Voltage Waveforms**

TYPICAL CHARACTERISTICS†

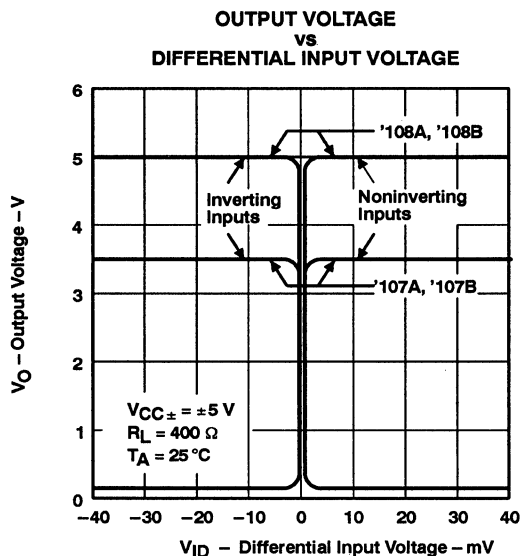


Figure 2

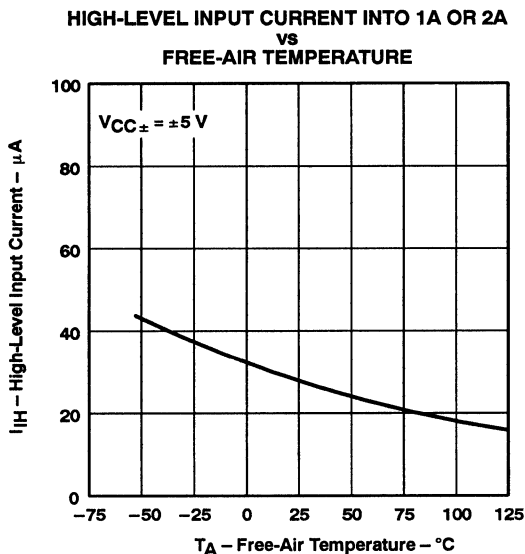


Figure 3

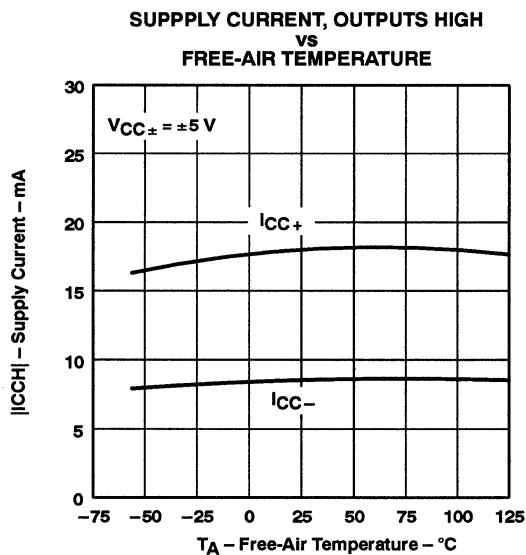


Figure 4

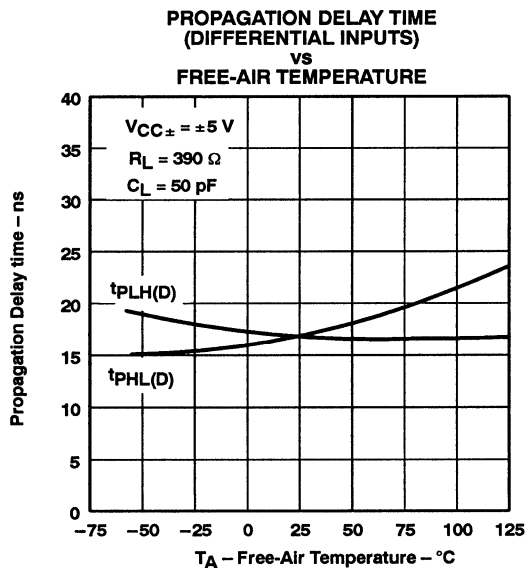


Figure 5

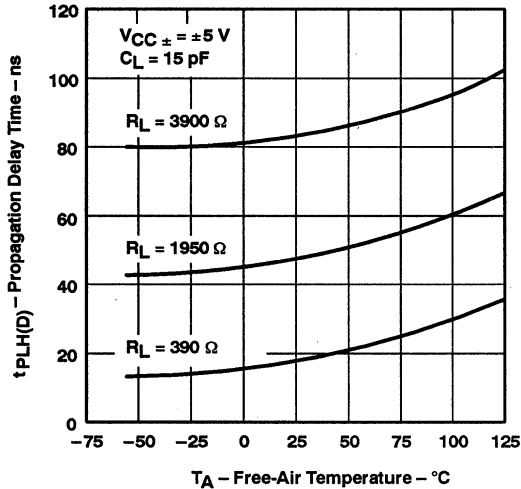
† Values below 0°C and above 70°C apply to SN55' only.

**SN55107A, SN55107B, SN55108A, SN55108B  
SN75107A, SN75107B, SN75108A, SN75108B  
DUAL LINE RECEIVERS**

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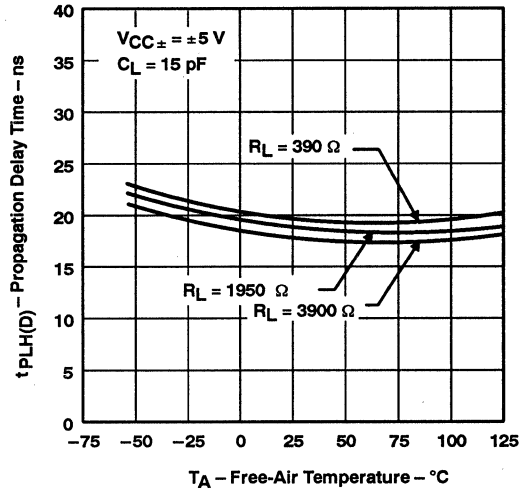
**TYPICAL CHARACTERISTICS†**

**PROPAGATION TIME, LOW-TO-HIGH LEVEL  
(DIFFERENTIAL INPUTS)  
vs  
FREE-AIR TEMPERATURE**



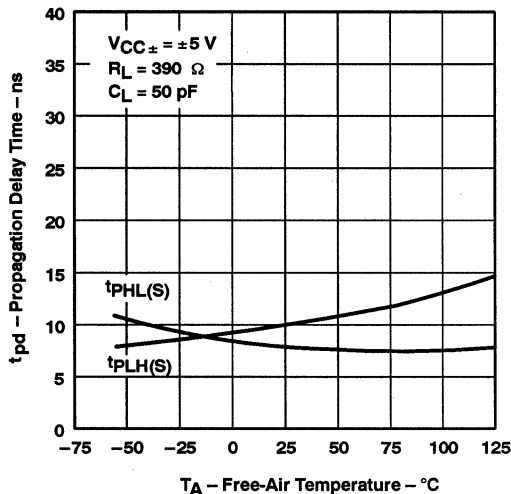
**Figure 6**

**PROPAGATION TIME, LOW-TO-HIGH LEVEL  
(DIFFERENTIAL INPUTS)  
vs  
FREE-AIR TEMPERATURE**



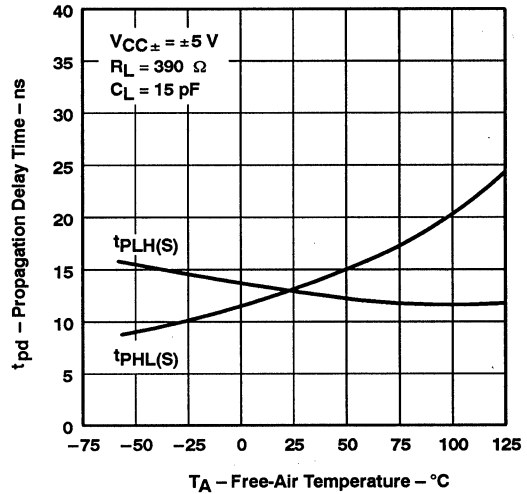
**Figure 7**

**'108A, '108B  
PROPAGATION DELAY TIME (STROBE INPUTS)  
vs  
FREE-AIR TEMPERATURE**



**Figure 8**

**'108A, '108B  
PROPAGATION DELAY TIME (STROBE INPUTS)  
vs  
FREE-AIR TEMPERATURE**



**Figure 9**

† Values below 0°C and above 70°C apply to SN55' only.



**APPLICATION INFORMATION**

**basic balanced-line transmission system**

The '107A, '107B, '108A, and '108B dual line circuits are designed specifically for use in high-speed data transmission systems that utilize balanced, terminated transmission lines such as twisted-pair lines. The system operates in the balanced mode, so noise induced on one line is also induced on the other. The noise appears common mode at the receiver input terminals where it is rejected. The ground connection between the line driver and receiver is not part of the signal circuit so that system performance is not affected by circulating ground currents.

The unique driver-output circuit allows terminated transmission lines to be driven at normal line impedances. High-speed system operation is ensured since line reflections are virtually eliminated when terminated lines are used. Crosstalk is minimized by low signal amplitudes and low line impedances.

The typical data delay in a system is approximately  $30 + 1.3 L$  ns, where L is the distance in feet separating the driver and receiver. This delay includes one gate delay in both the driver and receiver.

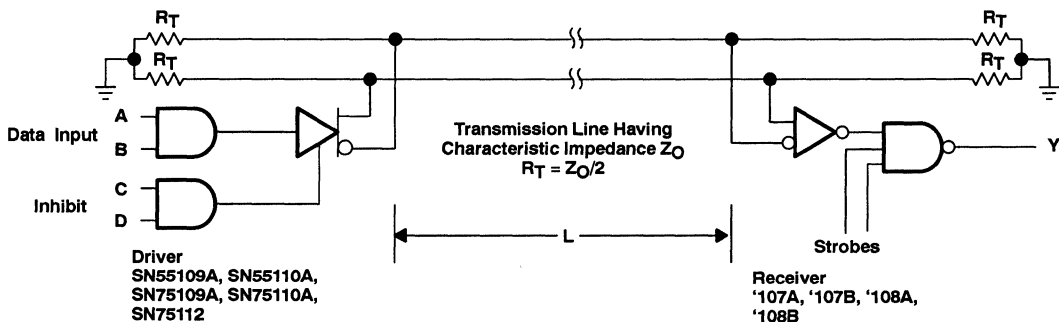
Data is impressed on the balanced-line system by unbalancing the line voltages with the driver output current. The driven line is selected by appropriate driver-input logic levels. The voltage difference is approximately:

$$V_{DIFF} \approx 1/2 I_{O(on)} \cdot R_T$$

High series line resistance will cause degradation of the signal. The receivers, however, will detect signals as low as 25 mV (or less). For normal line resistances, data may be recovered from lines of several thousand feet in length.

Line-termination resistors ( $R_T$ ) are required only at the extreme ends of the line. For short lines, termination resistors at the receiver only may prove adequate. The signal amplitude will then be approximately:

$$V_{DIFF} \approx I_{O(on)} \cdot R_T$$



**Figure 10. Typical Differential Data Line**

**data-bus or party-line system**

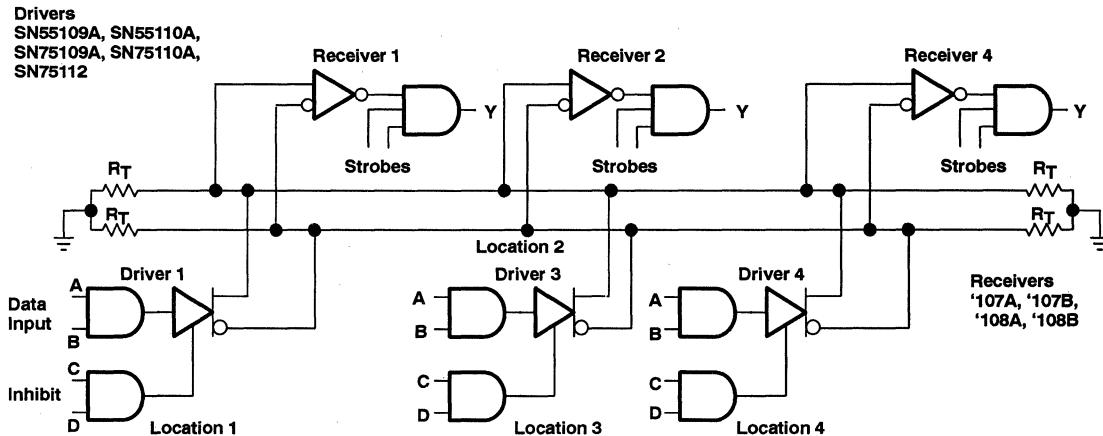
The strobe feature of the receivers and the inhibit feature of the drivers allow these dual line circuits to be used in data-bus or party-line systems. In these applications, several drivers and receivers may share a common transmission line. An enabled driver transmits data to all enabled receivers on the line while other drivers and receivers are disabled. Data is thus time multiplexed on the transmission line. The device specifications allow widely varying thermal and electrical environments at the various driver and receiver locations. The data-bus system offers maximum performance at minimum cost.



**SN55107A, SN55107B, SN55108A, SN55108B**  
**SN75107A, SN75107B, SN75108A, SN75108B**  
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**APPLICATION INFORMATION**



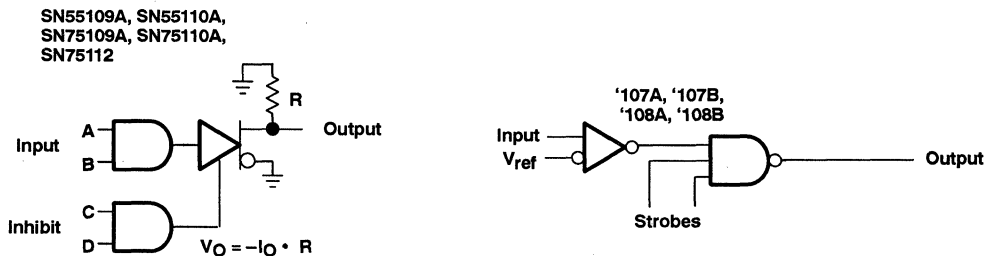
**Figure 11. Typical Differential Party Line**

**unbalanced or single-line systems**

These dual-line circuits may also be used in unbalanced or single-line systems. Although these systems do not offer the same performance as balanced systems for long lines, they are adequate for very short lines where environmental noise is not severe.

The receiver threshold level is established by applying a dc reference voltage to one receiver input terminal. The signal from the transmission line is applied to the remaining input. The reference voltage should be optimized so that signal swing is symmetrical about it for maximum noise margin. The reference voltage should be in the range of  $-3\text{ V}$  to  $3\text{ V}$ . It can be provided by a voltage supply or by a voltage divider from an available supply voltage.

A single-ended output from a driver may be used in single-line systems. Coaxial or shielded line is preferred for minimum noise and crosstalk problems. For large signal swings, the high output current (typically 27 mA) of the SN75112 is recommended. Drivers may be paralleled for higher current. When using only one channel of the line drivers, the other channel should be inhibited and/or have its outputs grounded.

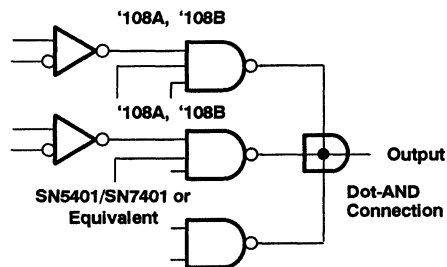


**Figure 12. Single-Ended Operation**

**APPLICATION INFORMATION**

**'108A, '108B dot-AND output connections**

The '108A, '108B line receivers feature an open-collector-output circuit that can be connected in the dot-AND logic configuration with other similar open-collector outputs. This allows a level of logic to be implemented without additional logic delay.



**Figure 13. Dot-AND Connection**

**increasing common-mode input voltage range of receiver**

The common-mode voltage range or CMVR is defined as the range of voltage applied simultaneously to both input terminals that if exceeded does not allow normal operation of the receiver.

The recommended operating CMVR is  $\pm 3$  V, making it useful in all but the noisiest environments. In extremely noisy environments, common-mode voltage can easily reach  $\pm 10$  V to  $\pm 15$  V if some precautions are not taken to reduce ground and power supply noise, as well as crosstalk problems. When the receiver must operate in such conditions, input attenuators should be used to decrease the system common-mode noise to a tolerable level at the receiver inputs. Differential noise is also reduced by the same ratio. These attenuators have been intentionally omitted from the receiver input terminals so the designer may select resistors that will be compatible with his particular application or environment. Furthermore, the use of attenuators adversely affects the input sensitivity, the propagation delay time, the power dissipation, and in some cases (depending on the selected resistor values) the input impedance, therefore reducing the versatility of the receiver.

The ability of the receiver to operate with approximately  $\pm 15$  V common-mode voltage at the inputs has been checked using the circuit shown in Figure 14. The resistors R1 and R2 provide a voltage divider network. Dividers with three different values presenting a 5-to-1 attenuation were used so as to operate the differential inputs at approximately  $\pm 3$  V common-mode voltage. Careful matching of the two attenuators is needed so as to balance the overdrive at the input stage. The resistors used are shown in Table 1.

**Table 1**

Attenuator 1:	R1 = 2 k $\Omega$ ,	R2 = 0.5 k $\Omega$
Attenuator 2:	R1 = 6 k $\Omega$ ,	R2 = 1.5 k $\Omega$
Attenuator 3:	R1 = 12 k $\Omega$ ,	R2 = 3 k $\Omega$

**Table 2. Typical Propagation Delays for Receiver With Attenuator Test Circuit Shown in Figure 14**

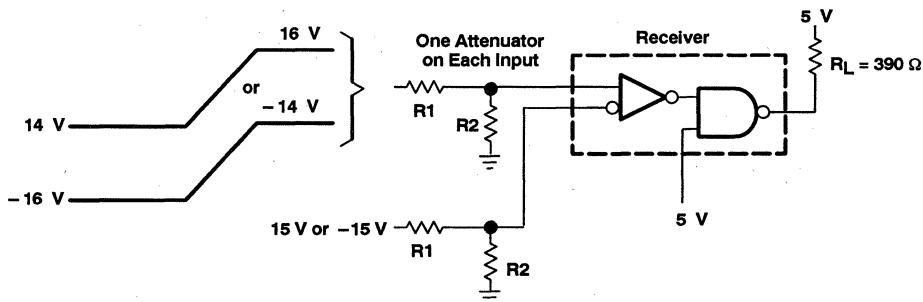
DEVICE	PARAMETERS	INPUT ATTENUATOR	TYPICAL (ns)
'107A, '107B	t <sub>PLH</sub>	1	20
		2	32
		3	42
	t <sub>PHL</sub>	1	22
		2	31
		3	33
'108A, '108B	t <sub>PLH</sub>	1	36
		2	47
		3	57
	t <sub>PHL</sub>	1	29
		2	38
		3	41

Table 2 shows some of the typical switching results obtained under such conditions.

**SN55107A, SN55107B, SN55108A, SN55108B  
SN75107A, SN75107B, SN75108A, SN75108B  
DUAL LINE RECEIVERS**

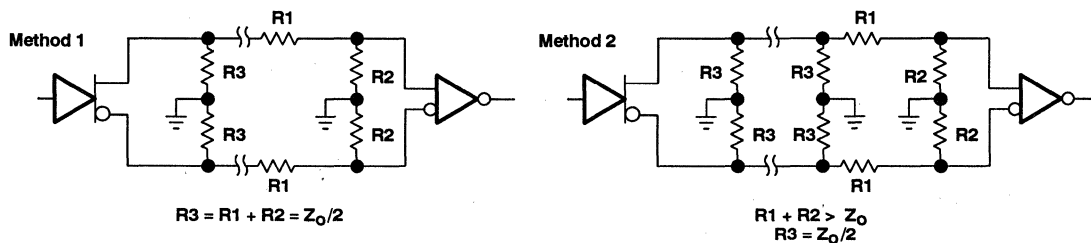
SLLS069A - D2304, JANUARY 1977 - REVISED JANUARY 1993

**APPLICATION INFORMATION**



**Figure 14. Common-Mode Circuit for Testing Input Attenuators With Results Shown In Table 2**

Two methods of terminating a transmission line to reduce reflections are:



**Figure 15. Termination Techniques**

The first method uses the resistors as the attenuation network and line termination. The second method uses two additional resistors for the line terminations.

### APPLICATION INFORMATION

For party-line operation, method 2 should be used as follows:

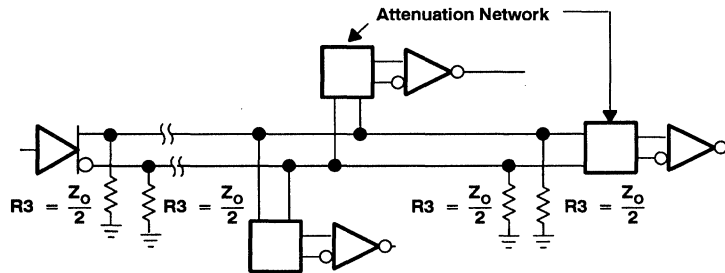


Figure 16. Party-Line Termination Technique

To minimize the loading, the values of R1 and R2 should be fairly large. Examples of possible values are shown in Table 1.

### furnace control using the SN75108A

The furnace control circuit in Figure 17 is an example of the possible use of the SN55107A Series in areas other than what would normally be considered electronic systems. Basically, the operation of this control is as follows. When the room temperature is below the desired level, the resistance of the room temperature sensor is high and channel 1 noninverting input is below (less positive than) the reference level set on the input differential amplifier. This situation causes a low output, operating the heat on relay and turning on the heat. The channel 2 noninverting input is below the reference level when the bonnet temperature of the furnace reaches the desired level. This causes a low output thus operating the blower relay. Normally the furnace is shut down when the room temperature reaches the desired level and the channel 1 output goes high, turning the heat off. The blower remains on as long as the bonnet temperature is high, even after the heat on relay is off. There is also a safety switch in the bonnet that shuts the furnace down if the temperature there exceeds desired limitations. The types of temperature-sensing devices and bias-resistor values used are determined by the particular operating conditions encountered.

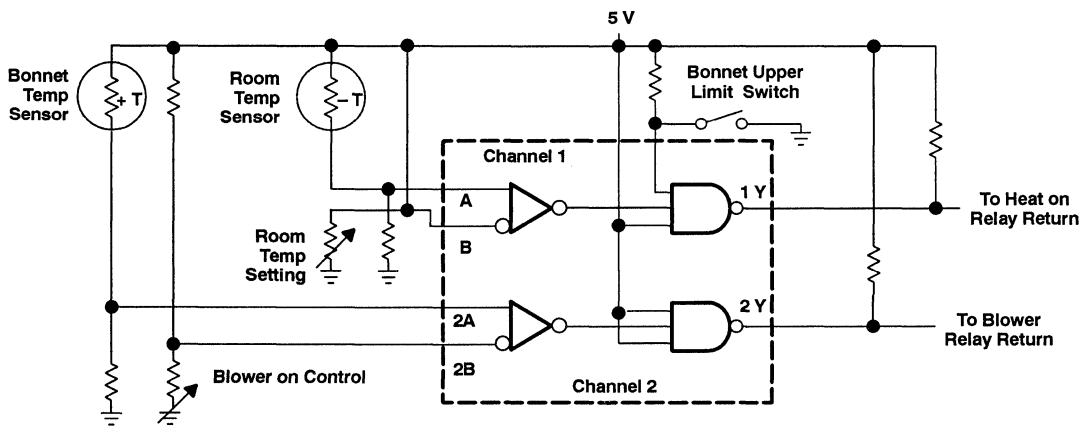


Figure 17. Furnace Control Using SN75108A

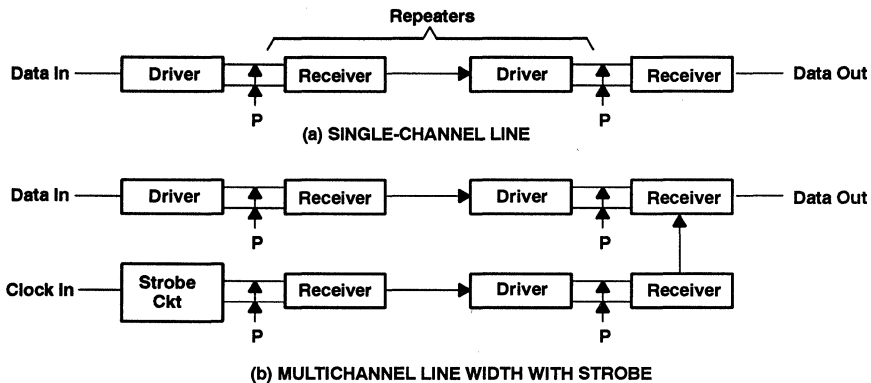
**SN55107A, SN55107B, SN55108A, SN55108B**  
**SN75107A, SN75107B, SN75108A, SN75108B**  
**DUAL LINE RECEIVERS**

SLLS069A - D2304, JANUARY 1977 - REVISED JANUARY 1993

**APPLICATION INFORMATION**

**repeaters for long lines**

In some cases, the driven line may be so long that the noise level on the line reaches the common-mode limits or the attenuation becomes too large and results in poor reception. In such a case, a simple application of a receiver and a driver as repeaters [shown in Figure 18(a)] restores the signal level and allows an adequate signal level at the receiving end. If multichannel operation is desired, then proper gating for each channel must be sent through the repeater station using another repeater set as in Figure 18(b).

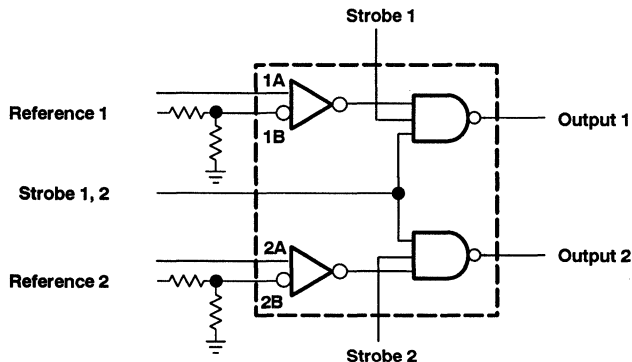


**Figure 18. Receiver-Driver Repeaters**

**receiver as dual differential comparator**

There are many applications for differential comparators, such as voltage comparison, threshold detection, controlled Schmitt triggering, and pulse-width control.

As a differential comparator, a '107A or '108A may be connected so as to compare the noninverting input terminal with the inverting input as shown in Figure 19. Thus the output will be high or low resulting from the A input being greater or less than the reference. The strobe inputs allow additional control over the circuit so that either output or both may be inhibited.

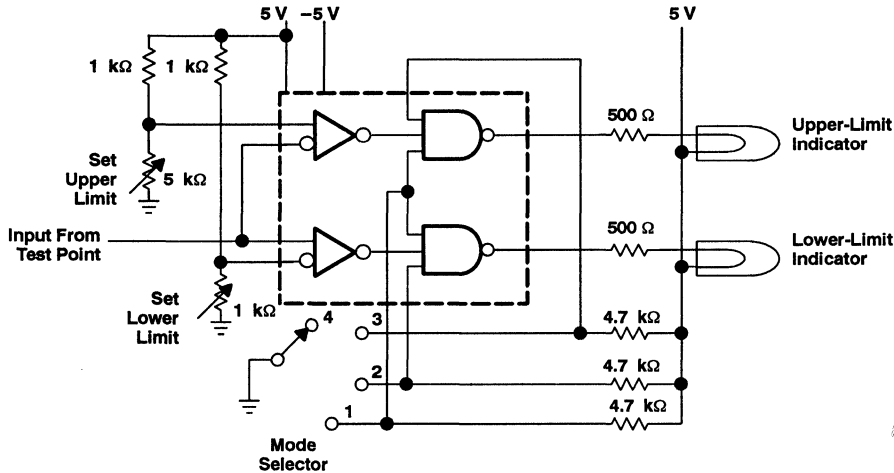


**Figure 19. SN55107A Series Receiver as a Dual Differential Comparator**

APPLICATION INFORMATION

window detector

The window detector circuit in Figure 20 has a large number of applications in test equipment and in determining upper limits, lower limits, or both at the same time - such as detecting whether a voltage or signal has exceeded its limits or window. Illumination of the upper-limit (lower-limit) indicator shows that the input voltage is above (below) the selected upper (lower) limit. A mode selector is provided for selecting the desired test. For window detecting, the upper and lower limits test position is used.



MODE SELECTOR LEGEND

POSITION	CONDITION
1	Off
2	Test for Upper Limit
3	Test for Lower Limit
4	Test for Upper and Lower Limits

Figure 20. Window Detector Using SN75108A

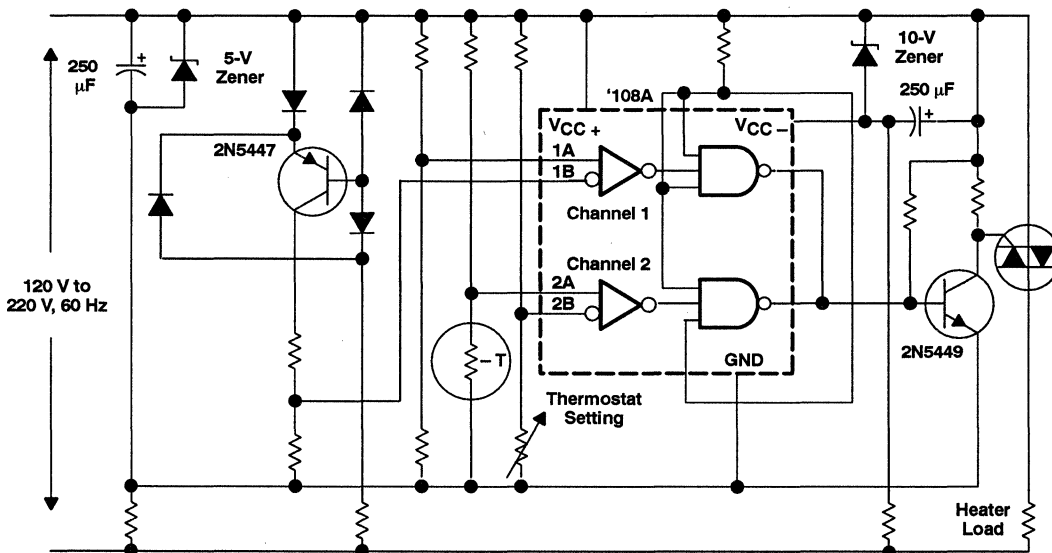
**SN55107A, SN55107B, SN55108A, SN55108B  
SN75107A, SN75107B, SN75108A, SN75108B  
DUAL LINE RECEIVERS**

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**APPLICATION INFORMATION**

**temperature controller with zero-voltage switching**

The circuit in Figure 21 switches an electric-resistive heater on or off by providing negative-going pulses to the gate of a triac during the time interval when the line voltage is passing through zero. The pulse generator is the 2N5447 and four diodes. This portion of the circuit provides negative-going pulses during the short time (approximately 100  $\mu$ s) when the line voltage is near zero. These pulses are fed to the inverting input of one channel of the '108A. If the room temperature is below the desired level, the resistance of the thermistor is high and the noninverting input of channel 2 is above the reference level determined by the thermostat setting. This provides a high-level output from channel 2. This output is ANDed with the positive-going pulses from the output of channel 1, which are reinverted in the 2N5449. This provides a high-level output from channel 2. This output is ANDed with the positive-going pulses from the output of channel 1, which are reinverted in the 2N5449.



**Figure 21. Zero-Voltage Switching Temperature Controller**

# SN55109A, SN55110A SN75109A, SN75110A, SN75112 DUAL LINE DRIVERS

SLLS106A – D2106, DECEMBER 1975 – REVISED FEBRUARY 1993

- Improved Stability Over Supply Voltage and Temperature Ranges
- Constant-Current Outputs
- High Speed
- Standard Supply Voltages
- High Output Impedance
- High Common-Mode Output Voltage Range  
–3 V to 10 V
- TTL-Input Compatibility
- Inhibitor Available for Driver Selection
- Glitch-Free During Power Up/Power Down
- SN75112 Complies With Requirements of CCITT Recommendation V.35

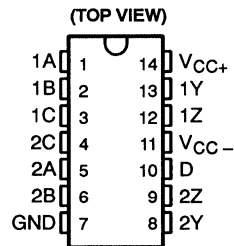
## description

The SN55109A, SN55110A, SN75109A, SN75110A, and SN75112 dual line drivers have improved output current regulation with supply voltage and temperature variations. In addition, the higher current of the SN75112 (27 mA) allows data to be transmitted over longer lines. These drivers offer optimum performance when used with the SN55107A, SN55108A, SN75107A, and SN75108A line receivers.

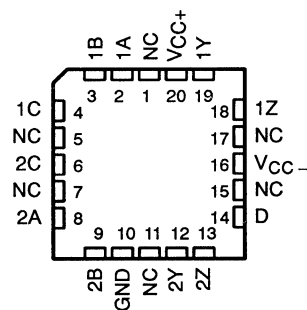
These drivers feature independent channels with common voltage supply and ground terminals. The significant difference between the three drivers is in the output current specification. The driver circuits feature a constant output current that is switched to either of two output terminals by the appropriate logic levels at the input terminals. The output current can be switched off (inhibited) by low logic levels on the enable inputs. The output current is nominally 6 mA for the '109A, 12 mA for the '110A, and 27 mA for the SN75112.

The enable/inhibit feature is provided so the circuits can be used in party-line or data-bus applications. A strobe or inhibitor (enable D), common to both drivers, is included for increased driver-logic versatility. The output current in the inhibited mode,  $I_{O(off)}$ , is specified so that minimum line loading is induced when the driver is used in a party-line system with other drivers. The output impedance of the driver in the inhibited mode is very high. The output impedance of a transistor is biased to cutoff.

SN55109A, SN55110A . . . J OR W PACKAGE  
SN75109A, SN75110A, SN75112 . . . D OR N PACKAGE



SN55109A, SN55110A . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

**THE SN75109A IS NOT  
RECOMMENDED FOR NEW DESIGN**

## AVAILABLE OPTIONS

T <sub>A</sub>	PACKAGE				CERAMIC FLATPACK (W)
	SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	
0°C to 70°C	SN75109AD SN75110AD SN75112D		SN75109AJ SN75110AJ SN75112J	SN75109AN SN75110AN SN75112N	
–55°C to 125°C		SN55109AFK SN55110AFK	SN55109AJ SN55110AJ	SN55109AJ SN55110AJ	SN55109AW SN55110AW

The D package is available taped and reeled. Add the suffix R to the device type, (e.g., SN75109ADR).

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

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**SN55109A, SN55110A  
SN75109A, SN75110A, SN75112  
DUAL LINE DRIVERS**

SLLS106A - D2106, DECEMBER 1975 - REVISED FEBRUARY 1993

**description (continued)**

The driver outputs have a common-mode voltage range of  $-3\text{ V}$  to  $10\text{ V}$ , allowing common-mode voltage on the line without affecting driver performance.

All inputs are diode clamped and are designed to satisfy TTL-system requirements. The inputs are tested at  $2\text{ V}$  for high-logic-level input conditions and  $0.8\text{ V}$  for low-logic-level input conditions. These tests ensure  $400\text{-mV}$  noise margin when interfaced with TTL Series 54/74.

The SN55109A and SN55110A are characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75109A, SN75110A, and SN75112 are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

**FUNCTION TABLE**  
(each driver)

LOGIC INPUTS		ENABLE INPUTS		OUTPUTS†	
A	B	C	D	Y	Z
X	X	L	X	Off	Off
X	X	X	L	Off	Off
L	X	H	H	On	Off
X	L	H	H	On	Off
H	H	H	H	Off	On

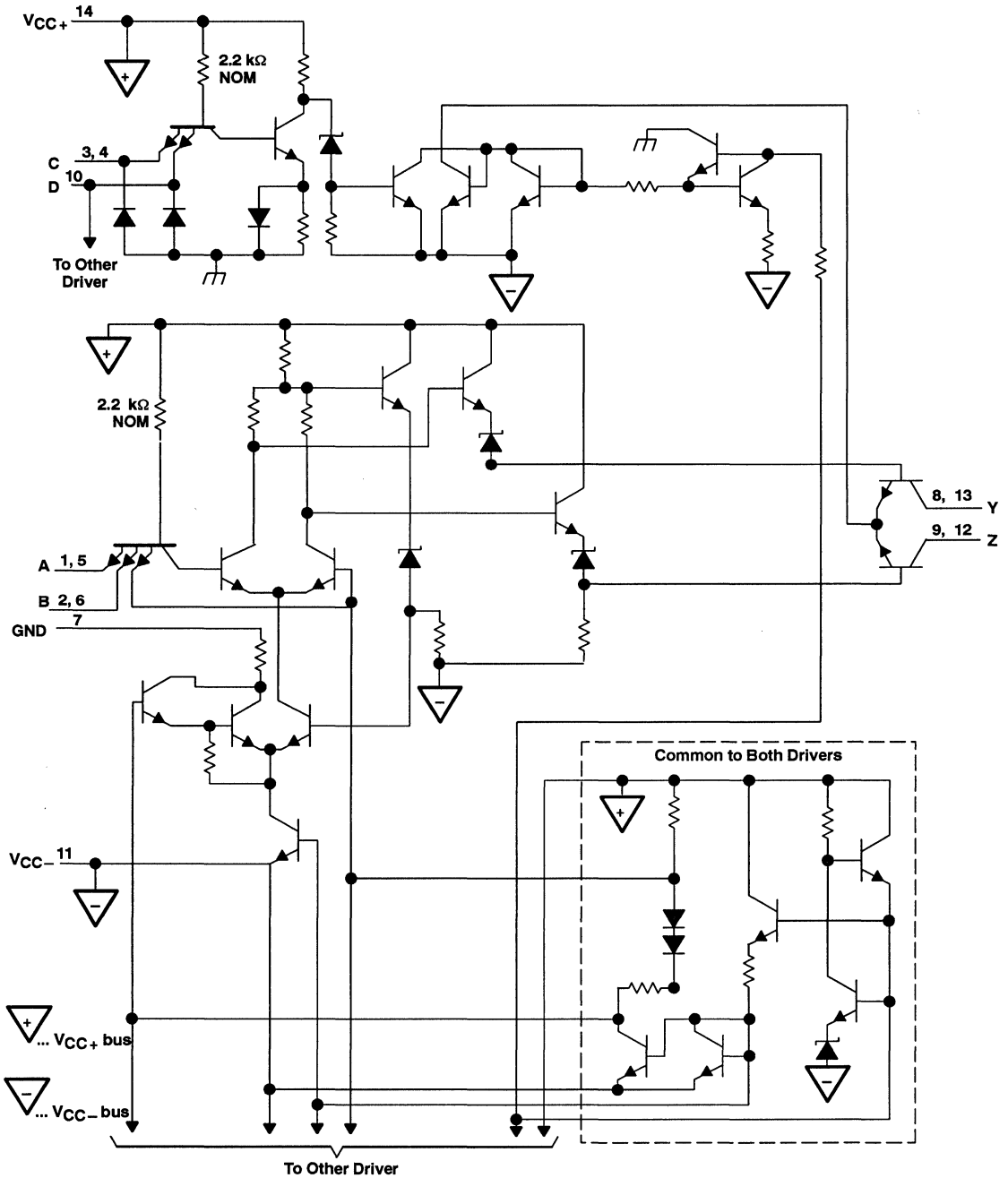
H = high level, L = low level, X = irrelevant

† When using only one channel of the line drivers, the other channel should be inhibited and/or have its outputs grounded.

**SN55109A, SN55110A  
SN75109A, SN75110A, SN75112  
DUAL LINE DRIVERS**

SLLS106A - D2106, DECEMBER 1975 - REVISED FEBRUARY 1993

**schematic (each driver)**



Pin numbers shown are for D, J, N, and W packages.



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**SN55109A, SN55110A  
SN75109A, SN75110A, SN75112  
DUAL LINE DRIVERS**

SLLS106A - D2106, DECEMBER 1975 - REVISED FEBRUARY 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

	SN55109A SN55110A	SN75109A SN75110A	SN75112	UNIT
Supply voltage, $V_{CC+}$ (see Note 1)	7	7	7	V
Supply voltage, $V_{CC-}$	-7	-7	-7	V
Input voltage, $V_I$	5.5	5.5	5.5	V
Output voltage range	-5 to 12	-5 to 12	-5 to 12	V
Continuous total power dissipation (see Note 2)	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	0 to 70	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds: FK package	260			°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or W package	300		°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package		260	°C

NOTES: 1. Voltage values are with respect to network ground terminal.

2. In the FK, J, or W package, SN55109A and SN55110A chips are either silver glass or alloy mounted, and SN75109A, SN75110A, and SN75112 chips are glass mounted.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—
W	1000 mW	8.0 mW/°C	640 mW	200 mW

**recommended operating conditions (see Note 3)**

	SN55109A, SN55110A			SN75109A, SN75110A, SN75112			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC+}$	4.5	5	5.5	4.75	5	5.25	V
Supply voltage, $V_{CC-}$	-4.5	-5	-5.5	-4.75	-5	-5.25	V
Positive common-mode output voltage	0		10	0		10	V
Negative common-mode output voltage	0		-3	0		-3	V
High-level input voltage, $V_{IH}$	2			2			V
Low-level output current, $V_{IL}$			0.8			0.8	V
Operating free-air temperature, $T_A$	-55		125	0		70	°C

NOTE 3: When using only one channel of the line drivers, the other channel should be inhibited and/or have its outputs grounded.



electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	SN55109A, SN75109A		SN55110A, SN75110A		SN75112		UNIT	
		MIN	TYP‡	MAX	MIN	TYP‡	MAX		MIN
$V_{IK}$ Input clamp voltage	$V_{CC\pm} = \text{MIN}, I_L = -12 \text{ mA}$	-0.9	-1.5	-0.9	-1.5	-0.9	-1.5	V	
$I_{O(\text{on})}$ On-state output current	$V_{CC\pm} = \text{MAX}, V_O = 10 \text{ V}$	6	7	12	15	27	36	mA	
	$V_{CC} = \text{MIN to MAX}, V_O = -1 \text{ V to } 1 \text{ V}$					24	28		32
	$V_{CC\pm} = \text{MIN}, V_O = -3 \text{ V}$	3.5	6	6.5	12	18	27		
$I_{O(\text{off})}$ Off-state output current	$V_{CC\pm} = \text{MIN}, V_O = 10 \text{ V}$			100	100	100	100	$\mu\text{A}$	
$I_I$ Input current at maximum input voltage	A, B, or C inputs			1	1	1	1	mA	
	D input			2	2	2	2		
$I_{IH}$ High-level input current	A, B, or C inputs			40	40	40	40	$\mu\text{A}$	
	D input			80	80	80	80		
$I_{IL}$ Low-level input current	A, B, or C inputs			-3	-3	-3	-3	mA	
	D input			-6	-6	-6	-6		
$I_{CC+ (\text{on})}$ Supply current from $V_{CC+}$ with driver enabled	$V_{CC\pm} = \text{MAX},$ A and B inputs at 0.4 V, C and D inputs at 2 V	18	30	23	35	25	40	mA	
$I_{CC- (\text{on})}$ Supply current from $V_{CC-}$ with driver enabled		-18	-30	-34	-50	-65	-100		
$I_{CC+ (\text{off})}$ Supply current from $V_{CC+}$ with driver inhibited	$V_{CC\pm} = \text{MAX},$ A, B, C, and D inputs at 0.4 V	18		21		30		mA	
$I_{CC- (\text{off})}$ Supply current from $V_{CC-}$ with driver inhibited		-10		-17		-32			

† For conditions shown as MIN or MAX, use appropriate value specified under recommended operating conditions.

‡ All typical values are at  $V_{CC+} = 5 \text{ V}, V_{CC-} = -5 \text{ V}, T_A = 25^\circ\text{C}.$

**SN55109A, SN55110A**  
**SN75109A, SN75110A, SN75112**  
**DUAL LINE DRIVERS**

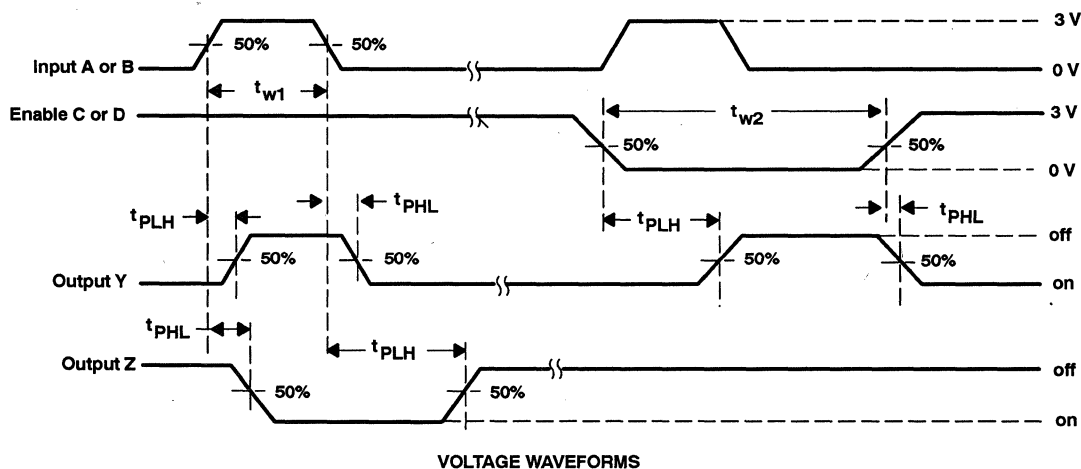
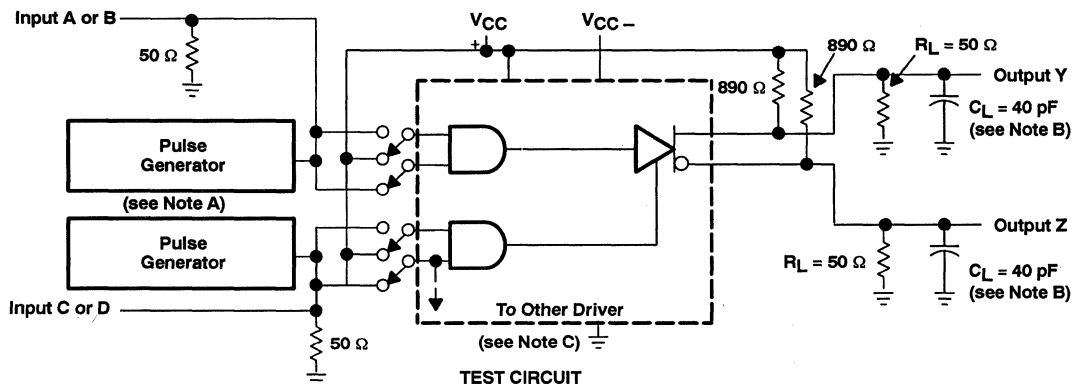
SLLS106A—D2106, DECEMBER 1975—REVISED FEBRUARY 1993

switching characteristics,  $V_{CC\pm} = \pm 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER †	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	A or B	Y or Z	$C_L = 40\text{ pF}$ , $R_L = 50\ \Omega$ , See Figure 1	9	15		ns
$t_{PHL}$				9	15		ns
$t_{PLH}$	C or D	Y or Z		16	25		ns
$t_{PHL}$				13	25		ns

†  $t_{PLH}$  = Propagation delay time, low-to-high-level output  
 $t_{PHL}$  = Propagation delay time, high-to-low-level output

**PARAMETER MEASUREMENT INFORMATION**



- NOTES: A. The pulse generators have the following characteristics:  $Z_O = 50\ \Omega$ ,  $t_r = t_f = 10 \pm 5\text{ ns}$ ,  $t_{w1} = 500\text{ ns}$ ,  $\text{PRR} \leq 1\text{ MHz}$ ,  $t_{w2} = 1\ \mu\text{s}$ ,  $\text{PRR} \leq 500\text{ kHz}$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. For simplicity, only one channel and the enable connections are shown.

**Figure 1. Test Circuit and Voltage Waveforms**



TYPICAL CHARACTERISTICS

ON-STATE OUTPUT CURRENT  
 vs  
 NEGATIVE SUPPLY VOLTAGE

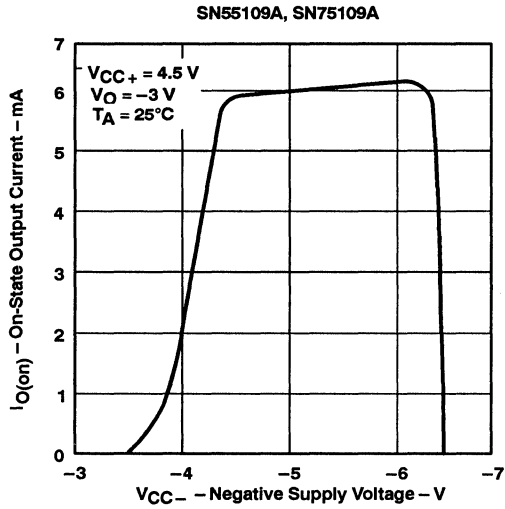


Figure 2

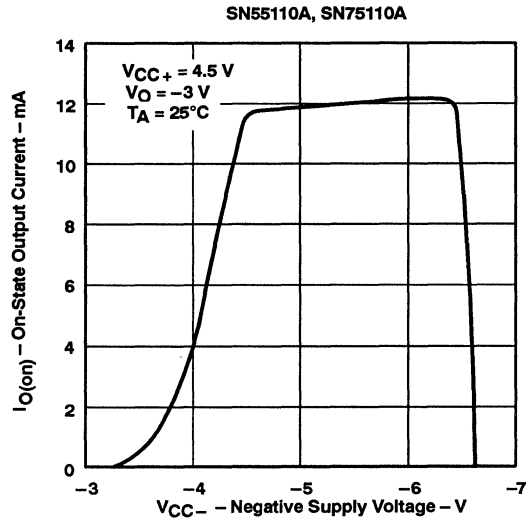


Figure 3

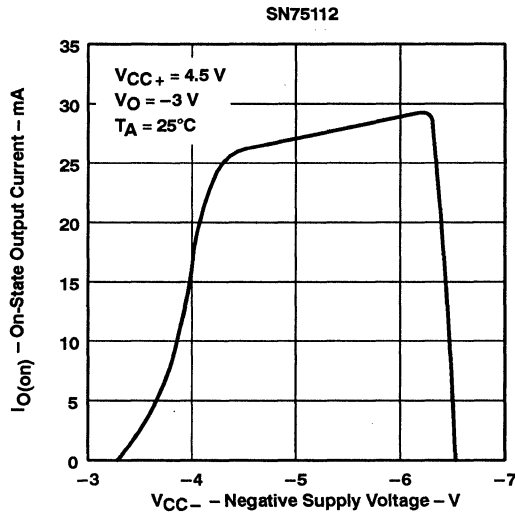


Figure 4

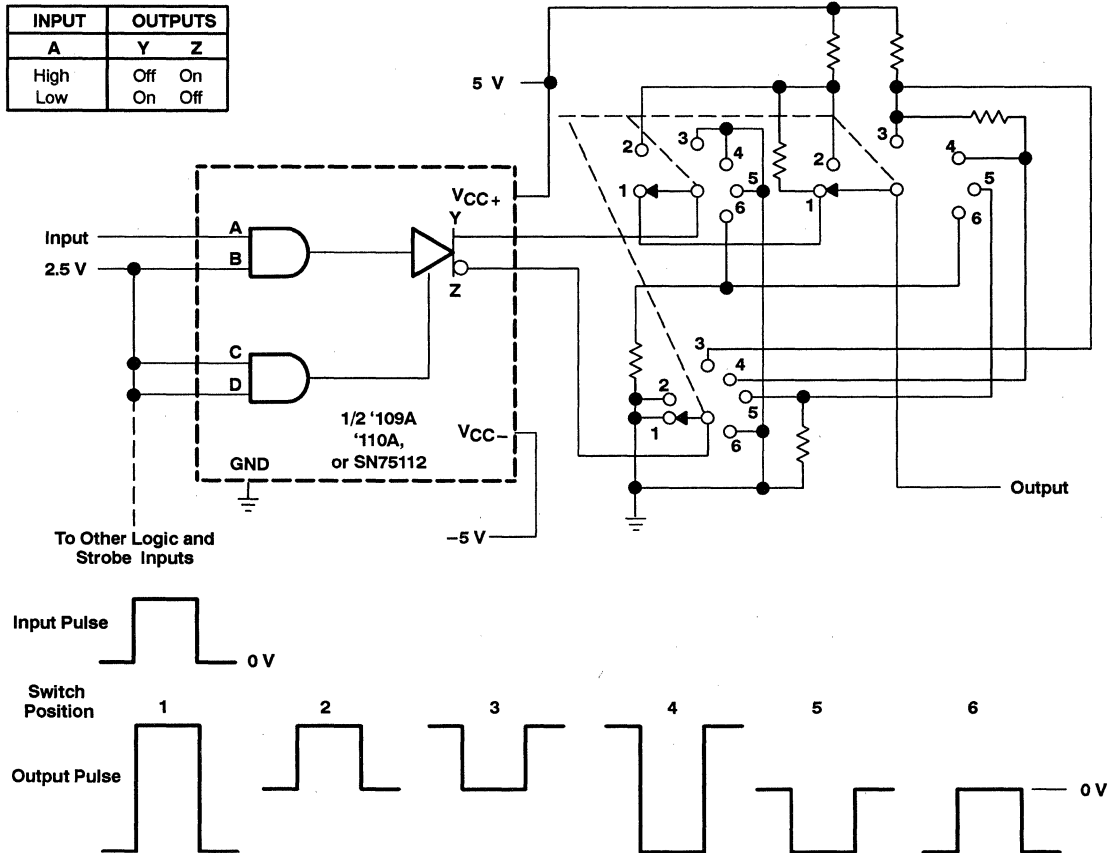
**SN55109A, SN55110A  
SN75109A, SN75110A, SN75112  
DUAL LINE DRIVERS**

SLLS106A - D2106, DECEMBER 1975 - REVISED FEBRUARY 1993

**APPLICATION INFORMATION**

**special pulse-control circuit**

Figure 5 shows a circuit that may be used as a pulse generator output or in many other testing applications.



**Figure 5. Pulse-Control Circuit**

APPLICATION INFORMATION

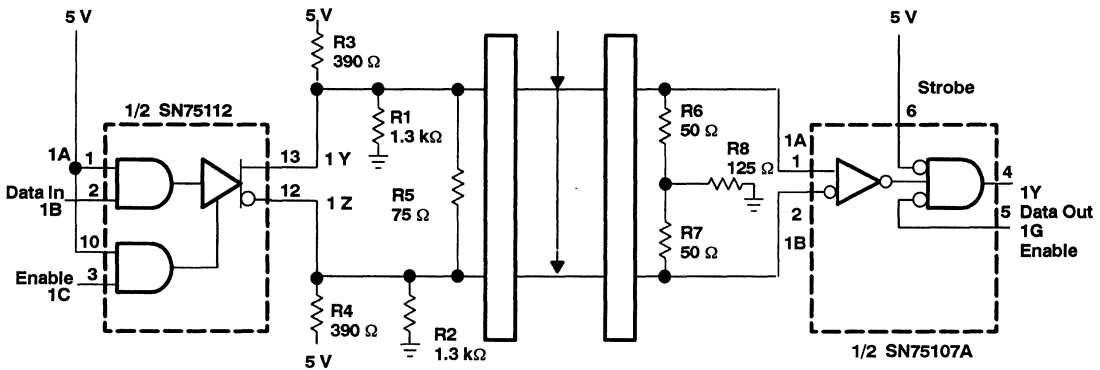
using the SN75112 as a CCITT recommendation V.35 line driver

The SN75112 dual line driver, the SN75107A dual line receiver, and some external resistors can be used to implement the data interchange circuit of CCITT recommendation V.35 (1976) modem specification. The circuit of one channel is shown in Figure 1 and meets the requirement of the interface as specified by Appendix 11 of CCITT V.35 and summarized in Table 1.

Table 1. CCITT V.35 Electrical Requirements

	MIN	MAX	UNIT
<b>GENERATOR</b>			
Source impedance, $Z_{source}$	50	150	$\Omega$
Resistance to ground, R	135	165	$\Omega$
Differential output voltage, $V_{OD}$	440	660	MV
10% to 90% rise time, $t_r$		40	ns
or		$0.01 \times u_i \uparrow$	
Common-mode output voltage, $V_{CC}$	-0.6	0.6	V
<b>LOAD (RECEIVER)</b>			
Input impedance, $Z_I$	90	110	$\Omega$
Resistance to ground, R	135	165	$\Omega$

$\uparrow u_i$  = unit interval or minimum signal element pulse width



All resistors are 5%, 1/4 W.

Figure 6. CCITT Recommendation V.35 Interface Using the SN75112 and SN75107A





# SN55113, SN75113 DUAL DIFFERENTIAL LINE DRIVERS

SLLS070A – D1315, SEPTEMBER 1973 – REVISED FEBRUARY 1993

- Choice of Open-Collector, Open-Emitter, or 3-State Outputs
- High-Impedance Output State for Party-Line Applications
- Single-Ended or Differential AND/NAND Outputs
- Single 5-V Supply
- Dual Channel Operation
- Compatible With TTL
- Short-Circuit Protection
- High-Current Outputs
- Common and individual Output Controls
- Clamp Diodes at Inputs and Outputs
- Easily Adaptable to SN55114 and SN75114 Applications
- Designed for Use With SN55115 and SN75115

## description

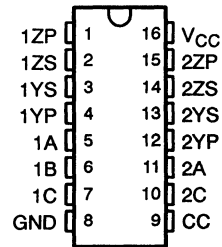
The SN55113 and SN75113 dual differential line drivers with 3-state outputs are designed to provide all the features of the SN55114 and SN75114 line drivers with the added feature of driver output controls. Individual controls are provided for each output pair, as well as a common control for both output pairs. If any output is low, the associated output is in a high-impedance state and the output can neither drive nor load the bus.

This permits many devices to be connected together on the same transmission line for party-line applications.

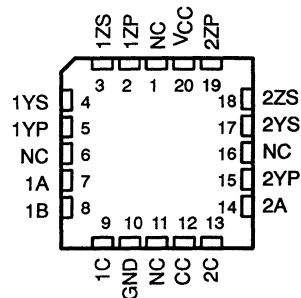
The output stages are similar to TTL totem-pole outputs, but with the sink outputs, YS and ZS, and the corresponding active pullup terminals, YP and ZP, available on adjacent package pins.

The SN55113 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75113 is characterized for operation over the temperature range of  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

SN55113 . . . J OR W PACKAGE  
SN75113 . . . D OR N PACKAGE  
(TOP VIEW)



SN55113 . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

FUNCTION TABLE

INPUTS			OUTPUTS		
OUTPUT	CONTROL	DATA		AND	NAND
C	CC	A	B <sup>†</sup>	Y	Z
L	X	X	X	Z	Z
X	L	X	X	Z	Z
H	H	L	X	L	H
H	H	X	L	L	H
H	H	H	H	H	L

H = high level, L = low level, X = irrelevant,

Z = high impedance (off)

<sup>†</sup>B input and 4th line of function table are applicable only to driver number 1.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

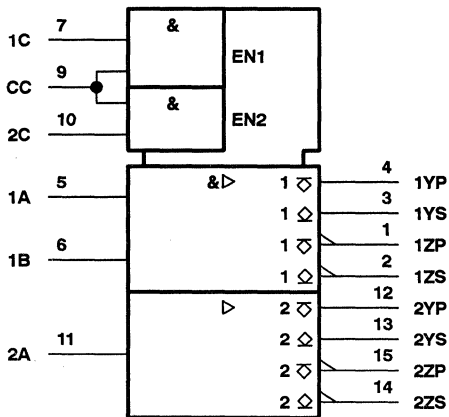
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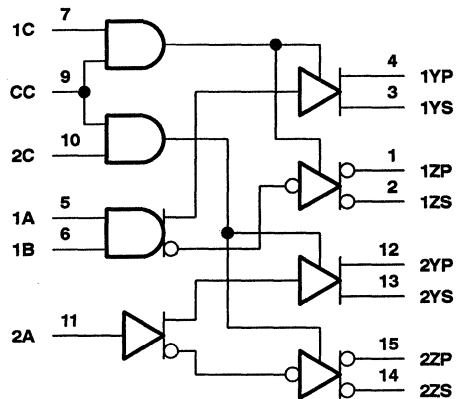
# SN55113, SN75113 DUAL DIFFERENTIAL LINE DRIVERS

SLLS070A-D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## logic symbol†



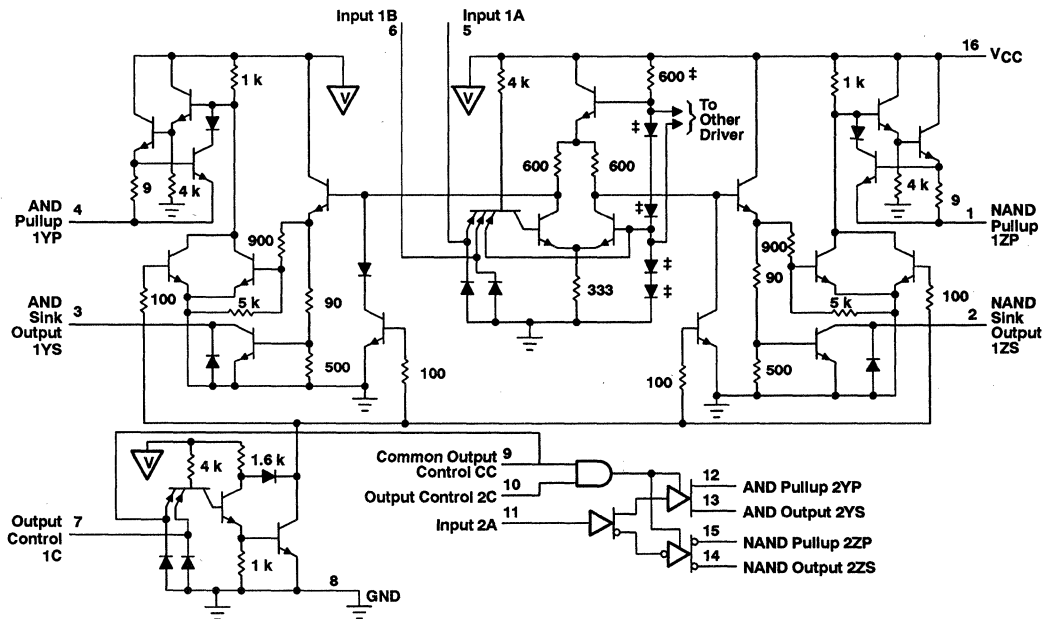
## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, N, and W packages.

## schematic



▽... VCC bus

‡ These components are common to both drivers.  
Resistor values shown are nominal and in ohms.

# SN55113, SN75113 DUAL DIFFERENTIAL LINE DRIVERS

SLLS070A – D1315, SEPTEMBER 1973 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Off-state voltage applied to open-collector outputs	12 V
Continuous total power dissipation (see Note 2)	See Dissipation Rating Table
Operating free-air temperature range: SN55113	–55°C to 125°C
SN75113	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	300°C

- NOTES: 1. All voltage values are with respect to network ground terminal.  
 2. In the J, FK, and W packages, SN55113 chips are alloy mounted; SN75113 chips are glass mounted.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	N/A
W	1000 mW	8.0 mW/°C	640 mW	200 mW

## recommended operating conditions

	SN55113			SN75113			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			2			V
Low-level input voltage, $V_{IL}$	0.8			0.8			V
High-level output current, $I_{OH}$	–40			–40			mA
Low-level output current, $I_{OL}$	40			40			mA
Operating free-air temperature, $T_A$	–55      125			0      70			°C



# SN55113, SN75113 DUAL DIFFERENTIAL LINE DRIVERS

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electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS†			SN55113			SN75113			UNIT			
					MIN	TYP‡	MAX	MIN	TYP‡	MAX				
V <sub>IK</sub>	Input clamp voltage	V <sub>CC</sub> = MIN, I <sub>I</sub> = -12 mA			-0.9	-1.5		-0.9	-1.5		V			
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN, V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V			I <sub>OH</sub> = -10 mA			2.4 3.4			V			
					I <sub>OH</sub> = -40 mA			2 3.0						
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN, V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V, I <sub>OL</sub> = 40 mA			0.23 0.4			0.23 0.4			V			
V <sub>OK</sub>	Output clamp voltage	V <sub>CC</sub> = MAX, I <sub>O</sub> = -40 mA			-1.1 -1.5			-1.1 -1.5			V			
I <sub>O(off)</sub>	Off-state open-collector output current	V <sub>CC</sub> = MAX			V <sub>OH</sub> = 12 V, T <sub>A</sub> = 25°C			1 10			μA			
					T <sub>A</sub> = 125°C			200						
					V <sub>OH</sub> = 5.25 V, T <sub>A</sub> = 25°C			1 10						
I <sub>OZ</sub>	Off-state (high-impedance-state) output current	V <sub>CC</sub> = MAX, Output controls at 0.8 V			T <sub>A</sub> = 25°C, V <sub>O</sub> = 0 to V <sub>CC</sub>			±10			μA			
					T <sub>A</sub> = MAX, V <sub>O</sub> = 0			-150						
					V <sub>O</sub> = 0.4 V			±80						
					V <sub>O</sub> = 2.4 V			±80						
I <sub>i</sub>	Input current at maximum input voltage	V <sub>CC</sub> = MAX, V <sub>I</sub> = 5.5 V			1			1			mA			
					2			2						
	High-level input current				V <sub>CC</sub> = MAX, V <sub>I</sub> = 2.4 V			40			40			μA
					80			80						
I <sub>iL</sub>	Low-level input current	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0.4 V			-1.6			-1.6			mA			
					-3.2			-3.2						
I <sub>OS</sub>	Short-circuit output current§	V <sub>CC</sub> = MAX, V <sub>O</sub> = 0, T <sub>A</sub> = 25°C			-40 -90 -120			-40 -90 -120			mA			
I <sub>CC</sub>	Supply current (both drivers)	All inputs at 0 V, No load, T <sub>A</sub> = 25°C			V <sub>CC</sub> = MAX			47 65			mA			
					V <sub>CC</sub> = 7 V			65 85						

† All parameters with the exception of off-state open-collector output current are measured with the active pullup connected to the sink output. For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at T<sub>A</sub> = 25°C and V<sub>CC</sub> = 5 V, with the exception of I<sub>CC</sub> at 7 V.

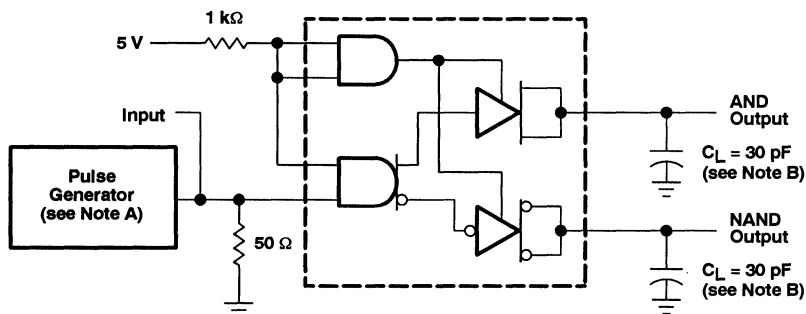
§ Only one output should be shorted at a time, and duration of the short-circuit should not exceed one second.

## switching characteristics, V<sub>CC</sub> = 5 V, C<sub>L</sub> = 30 pF, T<sub>A</sub> = 25°C

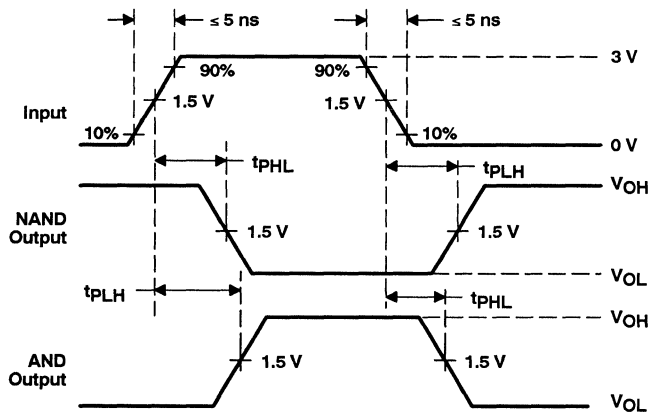
PARAMETER		TEST CONDITIONS		SN55113			SN75113			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	See Figure 1		13 20			13 30			ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output			12 20			12 30			
t <sub>PZH</sub>	Output enable time to high level	R <sub>L</sub> = 180 Ω, See Figure 2		7 15			7 20			ns
t <sub>PZL</sub>	Output enable time to low level	R <sub>L</sub> = 250 Ω, See Figure 3		14 30			14 40			ns
t <sub>PHZ</sub>	Output disable time from high level	R <sub>L</sub> = 180 Ω, See Figure 2		10 20			10 30			ns
t <sub>PLZ</sub>	Output disable time from low level	R <sub>L</sub> = 250 Ω, See Figure 3		17 35			17 35			ns



PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

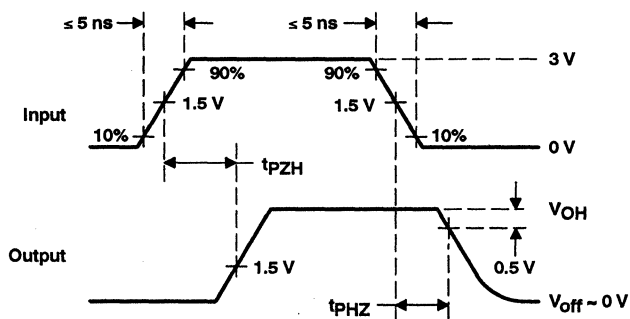
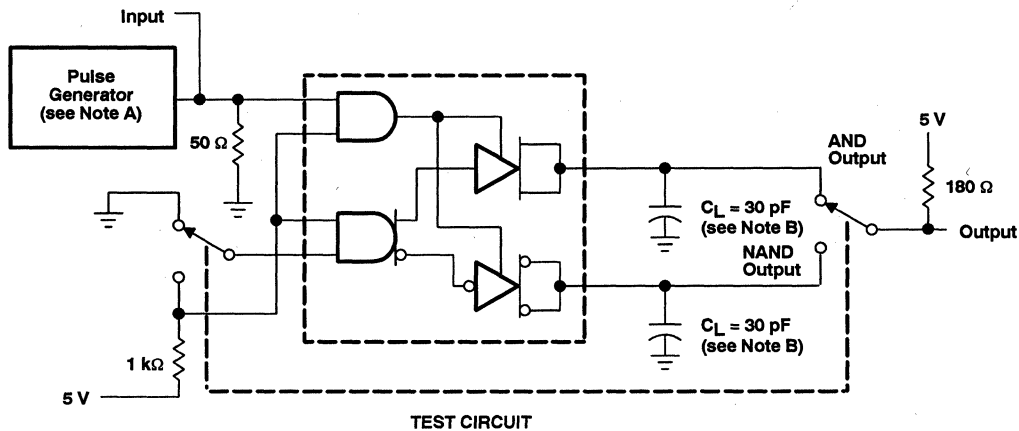
NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ ,  $t_w = 100 \text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms  $t_{PLH}$  and  $t_{PHL}$

# SN55113, SN75113 DUAL DIFFERENTIAL LINE DRIVERS

SLLS070A - D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

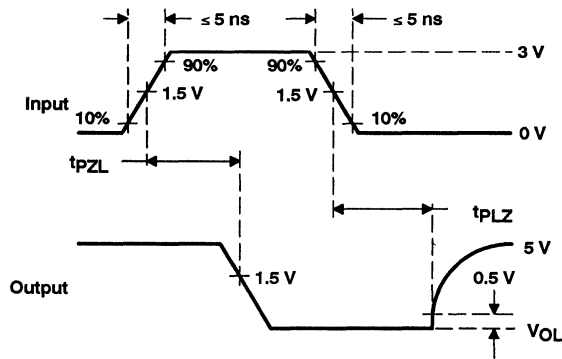
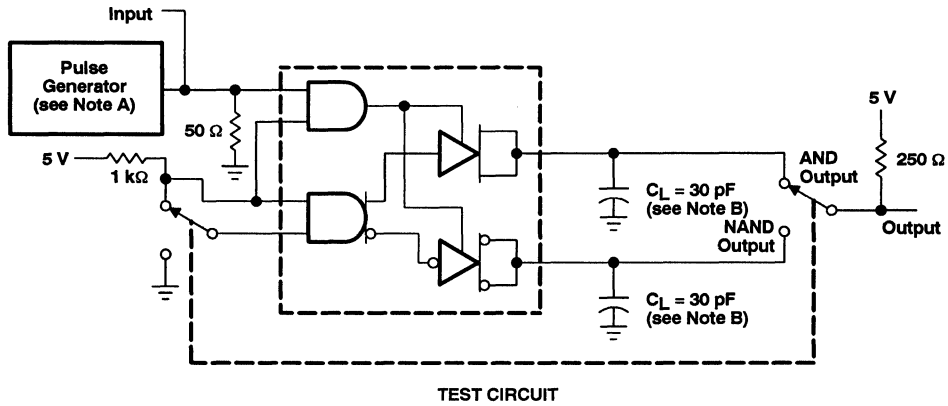


VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ ,  $t_w = 100 \text{ ns}$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 2. Test Circuit and Voltage Waveforms  $t_{pZH}$  and  $t_{pHZ}$

PARAMETER MEASUREMENT INFORMATION



NOTES: C. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ ,  $t_w = 100 \text{ ns}$ .  
 D.  $C_L$  includes probe and jig capacitance.

Figure 3. Test Circuit and Voltage Waveforms,  $t_{pZL}$  and  $t_{PLZ}$



# SN55113, SN75113 DUAL DIFFERENTIAL LINE DRIVERS

SLLS070A - D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS†

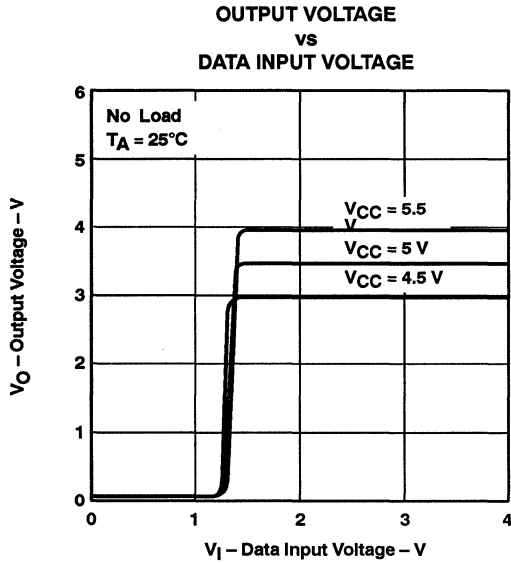


Figure 4

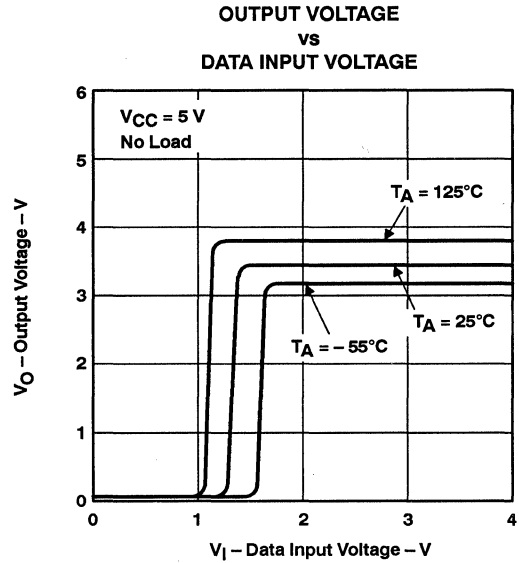


Figure 5

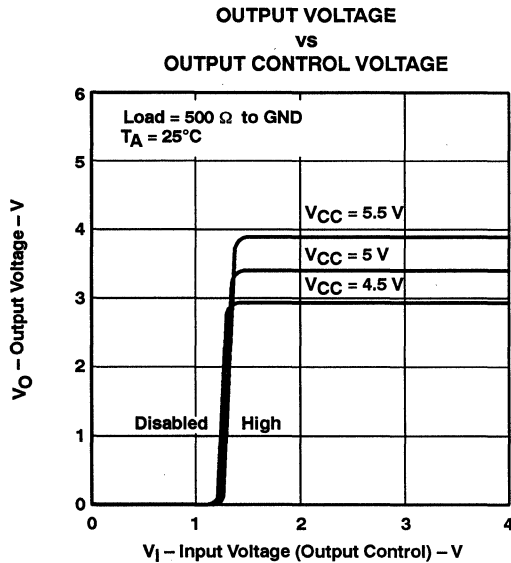


Figure 6

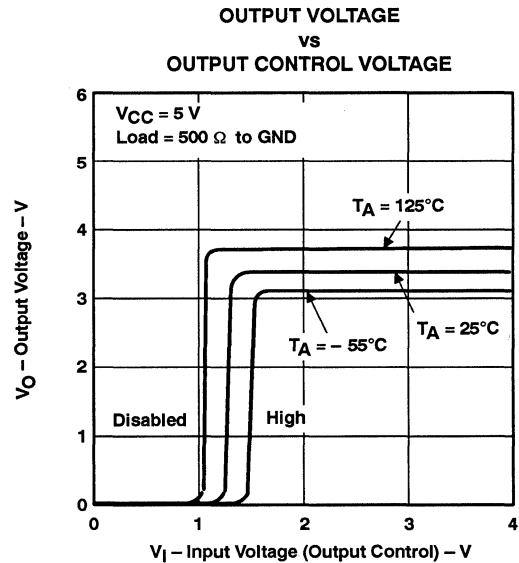


Figure 7

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  and for supply voltages below 4.75 V and above 5.25 V are applicable to SN55113 circuits only. These parameters were measured with the active pullup connected to the sink output.

TYPICAL CHARACTERISTICS†

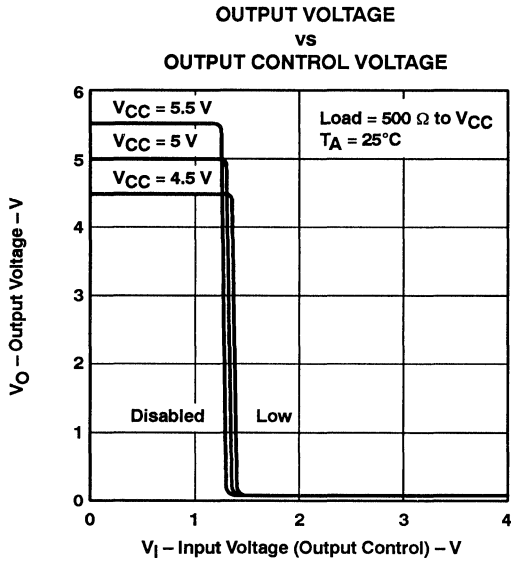


Figure 8

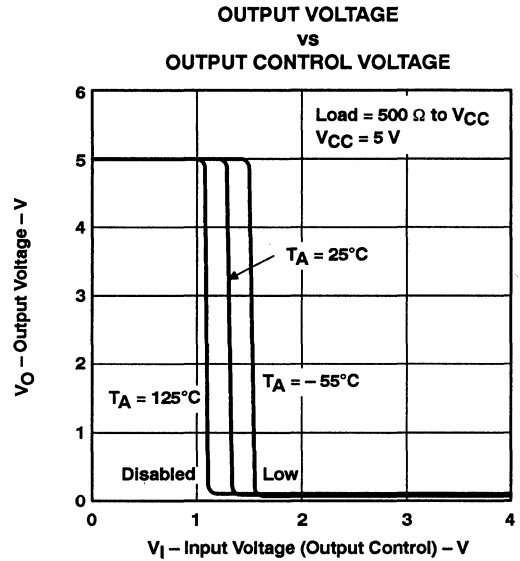


Figure 9

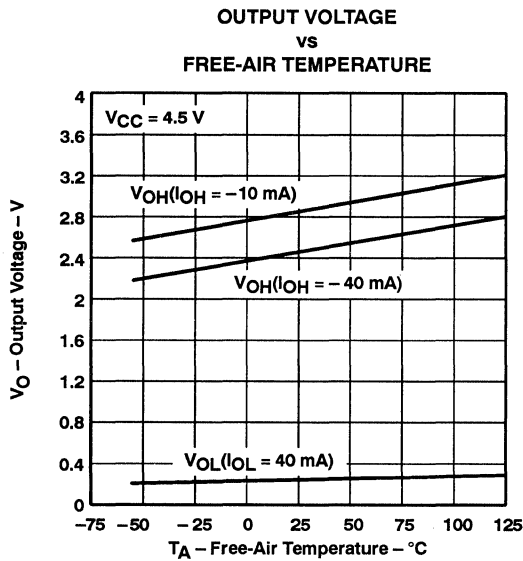


Figure 10

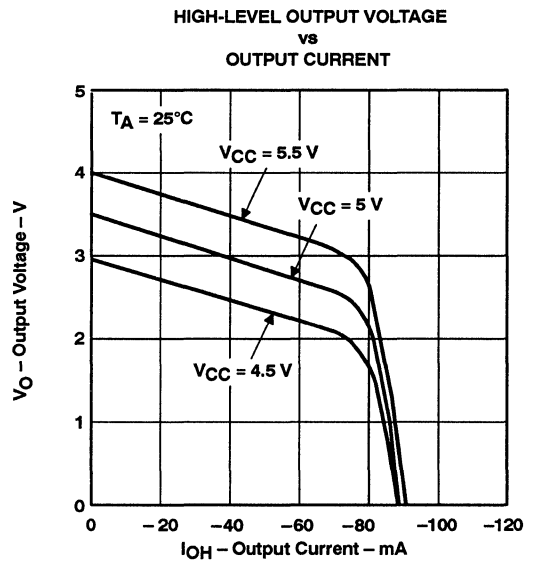


Figure 11

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  and for supply voltages below  $4.75\text{ V}$  and above  $5.25\text{ V}$  are applicable to SN55113 circuits only. These parameters were measured with the active pullup connected to the sink output.

# SN55113, SN75113 DUAL DIFFERENTIAL LINE DRIVERS

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## TYPICAL CHARACTERISTICS†

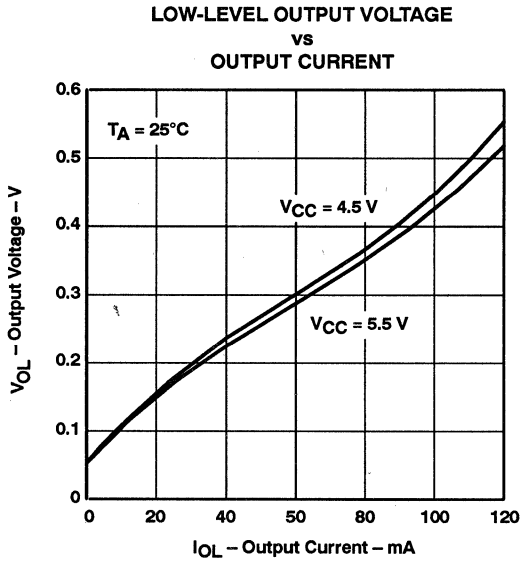


Figure 12

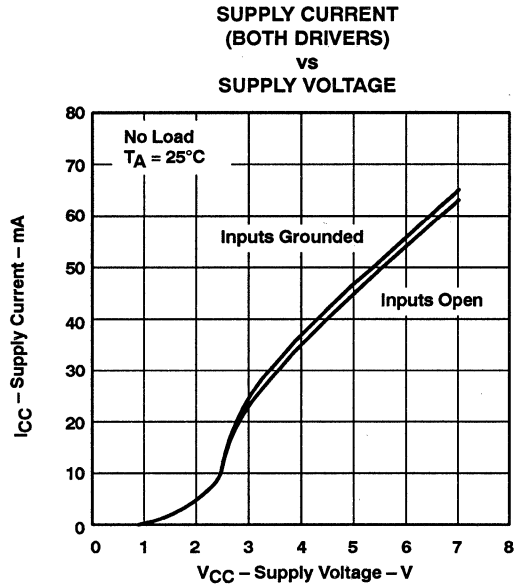


Figure 13

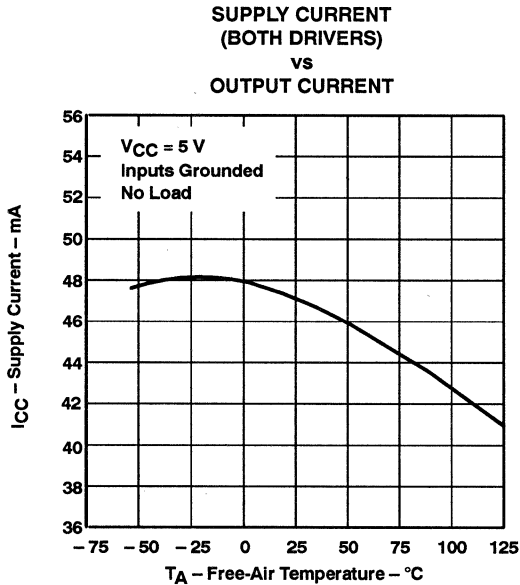


Figure 14

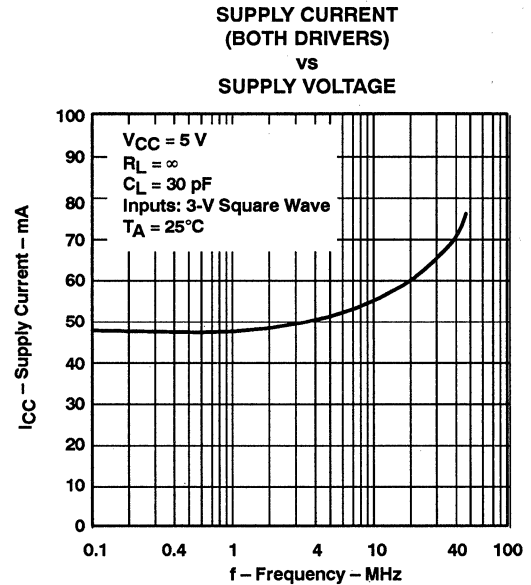
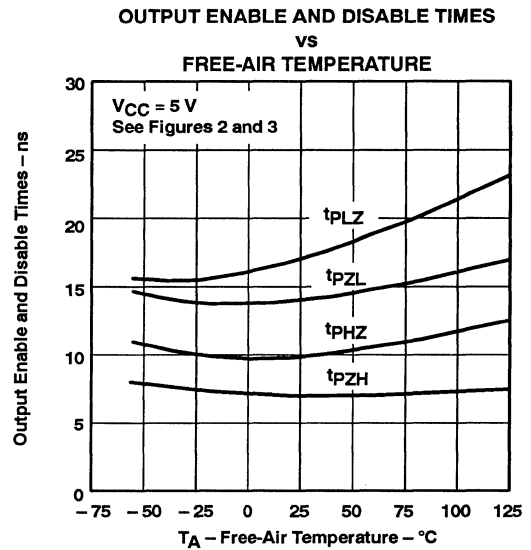
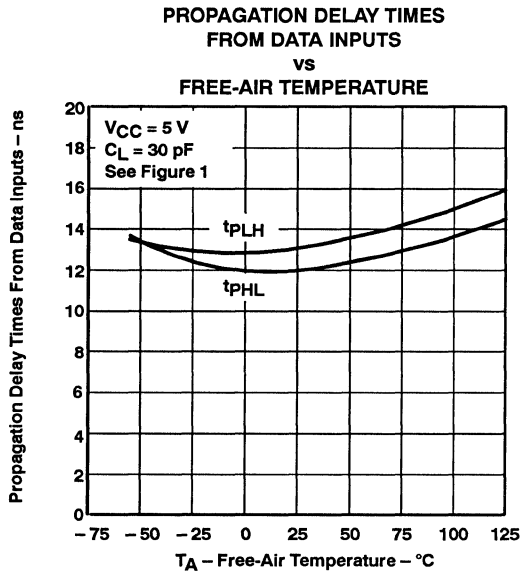


Figure 15

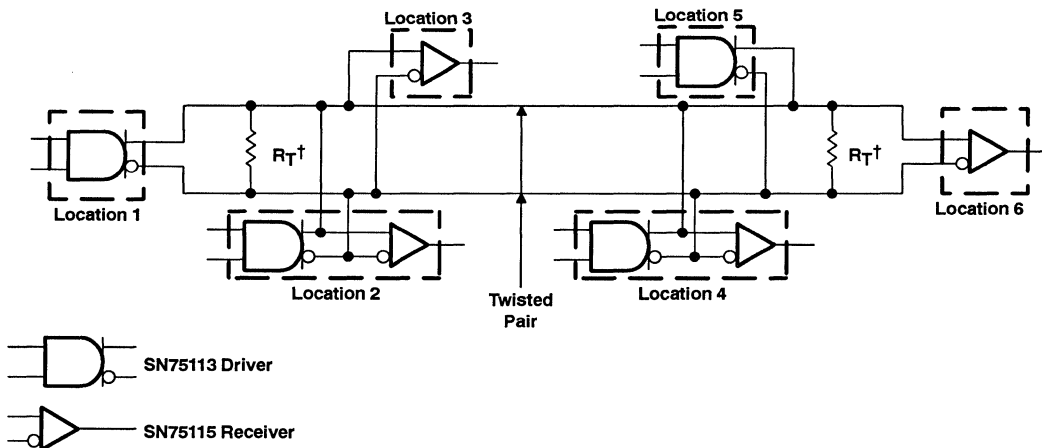
† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  and for supply voltages below 4.75 V and above 5.25 V are applicable to SN55113 circuits only. These parameters were measured with the active pullup connected to the sink output.

TYPICAL CHARACTERISTICS†



† Data for temperatures below 0°C and above 70°C and for supply voltages below 4.75 V and above 5.25 V are applicable to SN55113 circuits only. These parameters were measured with the active pullup connected to the sink output.

APPLICATION INFORMATION



†  $R_T = Z_0$ . A capacitor may be connected in series with  $R_T$  to reduce power dissipation.

Figure 18. Basic Party-Line or Data-Bus Differential Data Transmission



# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

SLLS071A – D1315, SEPTEMBER 1973 – REVISED FEBRUARY 1993

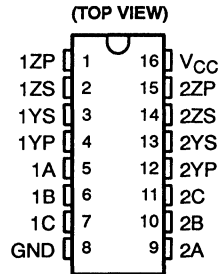
- Choice of Open-Collector, Open-Emitter, or Totem-Pole Outputs
- Single-Ended or Differential AND/NAND Outputs
- Single 5-V Supply
- Dual-Channel Operation
- TTL Compatible
- Short-Circuit Protection
- High-Current Outputs
- Triple inputs
- Clamp Diodes at Inputs and Outputs
- Designed for Use With SN55115 and SN75115 Differential Line Receivers
- Designed to Be Interchangeable With Fairchild 9614 Line Driver

## description

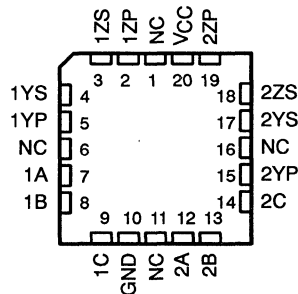
The SN55114 and SN75114 dual differential line drivers are designed to provide differential output signals with the high-current capability for driving balanced lines, such as twisted pair, at normal line impedances without high power dissipation. The output stages are similar to TTL totem-pole outputs, but with the sink outputs, YS and ZS, and the corresponding active pullup terminals, YP and ZP, available on adjacent package pins. Since the output stages provide TTL-compatible output levels, these devices may also be used as TTL expanders or phase splitters.

The SN55114 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75114 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

SN55114 . . . J OR W PACKAGE  
SN75114 . . . D OR N PACKAGE



SN55114 . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

FUNCTION TABLE

INPUTS			OUTPUTS	
A	B	C	Y	Z
H	L	H	H	L
All other inputs combinations			L	H

H = high level, L = low level

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

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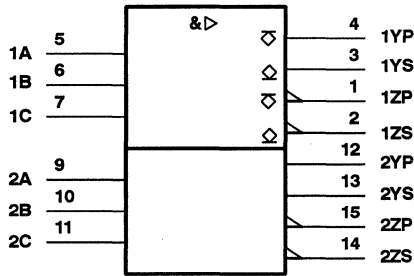
Copyright © 1993, Texas Instruments Incorporated

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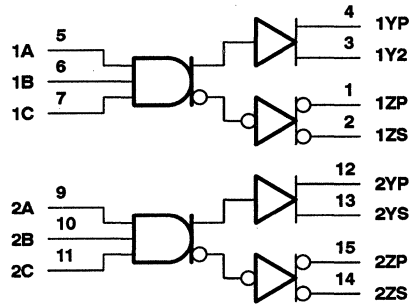
# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

SLLS071A-D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## logic symbol†



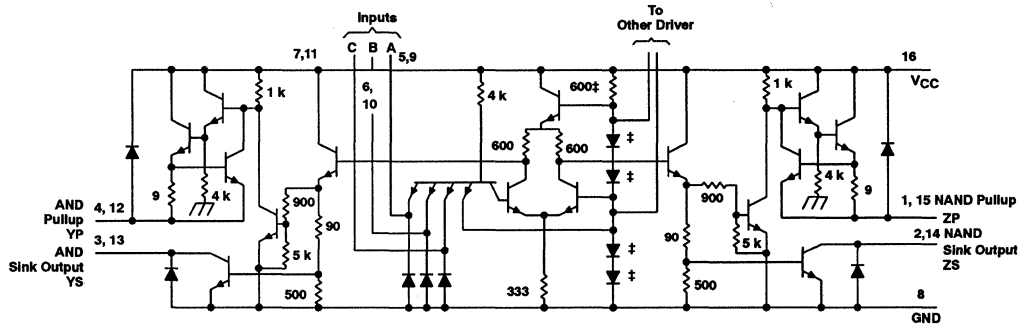
## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, N, and W packages.

## schematic (each driver)



‡ These components are common to both drivers. Resistor values shown are nominal and in ohms. Pin numbers shown are for the D, J, N, and W packages.

# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55114	SN75114	UNIT
Supply voltage, $V_{CC}$ (see Note 1)	7	7	V
Input voltage	5.5	5.5	V
Off-state voltage applied to open-collector outputs	12	12	V
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds: FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	300		°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package		260	°C

NOTE 1: Voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	—
FK†	1375 mW	11.0 mW/°C	880 mW	275 mW
J†	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—
W†	1000 mW	8.0 mW/°C	640 mW	200 MW

† In the FK, J, and W packages, SN55114 chips are either silver glass or alloy mounted.

## recommended operating conditions

	SN55114			SN75114			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			2			V
Low-level input voltage, $V_{IL}$			0.8			0.8	V
High-level output current, $I_{OH}$			-40			-40	mA
Low-level output current, $I_{OL}$			40			40	mA
Operating free-air temperature, $T_A$	-55		125	0		70	°C



# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

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electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	SN55114		SN75114		UNIT		
		MIN	TYP‡	MAX	MIN		TYP‡	MAX
$V_{IK}$ Input clamp voltage	$V_{CC} = \text{MIN}$ , $I_I = -12 \text{ mA}$	-0.9	-1.5	-0.9	-1.5	V		
$V_{OH}$ High-level output voltage	$V_{CC} = \text{MIN}$ , $V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$	$I_{OH} = -10 \text{ mA}$	2.4	3.4	2.4	3.4	V	
		$I_{OH} = -40 \text{ mA}$	2	3	2	3		
$V_{OL}$ Low-level output voltage	$V_{CC} = \text{MIN}$ , $V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ , $I_{OL} = 40 \text{ mA}$	0.2	0.4	0.2	0.45	V		
$V_{OK}$ Output clamp voltage	$V_{CC} = 5 \text{ V}$ , $I_O = 40 \text{ mA}$ , $T_A = 25^\circ\text{C}$	6.1	6.5	6.1	6.5	V		
	$V_{CC} = \text{MAX}$ , $I_O = -40 \text{ mA}$ , $T_A = 25^\circ\text{C}$	-1.1	-1.5	-1.1	-1.5			
$I_{O(\text{off})}$ Off-state open collector output current	$V_{CC} = \text{MAX}$	$V_{OH} = 12 \text{ V}$	$T_A = 25^\circ\text{C}$	1	100	$\mu\text{A}$		
			$T_A = 125^\circ\text{C}$	200				
		$V_{OH} = 5.25 \text{ V}$	$T_A = 25^\circ\text{C}$	1			100	
			$T_A = 70^\circ\text{C}$	200				
$I_I$ Input current at maximum input voltage	$V_{CC} = \text{MAX}$ , $V_I = 5.5 \text{ V}$	1		1		mA		
$I_{IH}$ High-level input current	$V_{CC} = \text{MAX}$ , $V_I = 2.4 \text{ V}$	40		40		$\mu\text{A}$		
$I_{IL}$ Low-level input current	$V_{CC} = \text{MAX}$ , $V_I = 0.4 \text{ V}$	-1.1	-1.6	-1.1	-1.6	mA		
$I_{OS}$ Short-circuit output current‡	$V_{CC} = \text{MAX}$ , $V_O = 0$ , $T_A = 25^\circ\text{C}$	-40	-90	-120	-40	-90	-120	mA
$I_{CC}$ Supply current (both drivers)	All inputs at 0 V, No load, $T_A = 25^\circ\text{C}$	$V_{CC} = \text{MAX}$	37	50	37	50	mA	
		$V_{CC} = 7 \text{ V}$	47	65	47	70		

† All parameters with the exception of off-state open-collector output current are measured with the active pullup connected to the sink output. For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at  $T_A = 25^\circ\text{C}$  and  $V_{CC} = 5 \text{ V}$ , with the exception of  $I_{CC}$  at 7 V.

§ Only one output should be shorted at a time, and duration of the short circuit should not exceed one second.

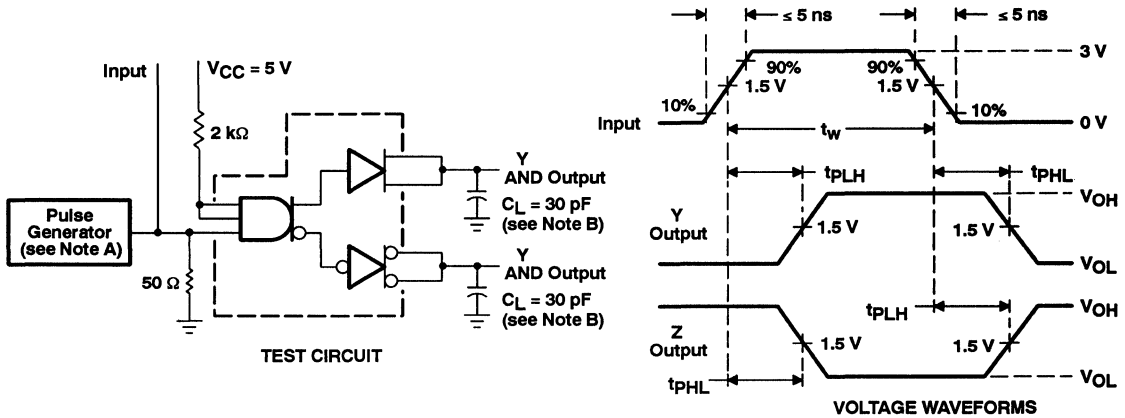
## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	SN55114		SN75114		UNIT
		MIN	TYP	MAX	MIN	
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 30 \text{ pF}$ , See Figure 1	15		20		ns
$t_{PHL}$ Propagation delay time, high-to-low-level output		11		20		

# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

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## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics:  $Z_O = 500 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ ,  $t_w \leq 100 \text{ ns}$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

## TYPICAL CHARACTERISTICS†

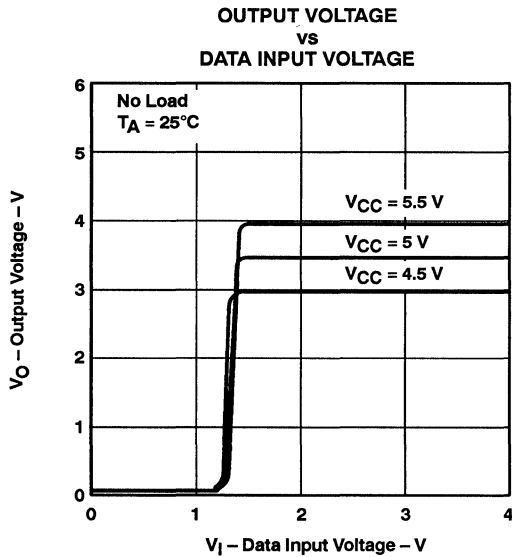


Figure 2

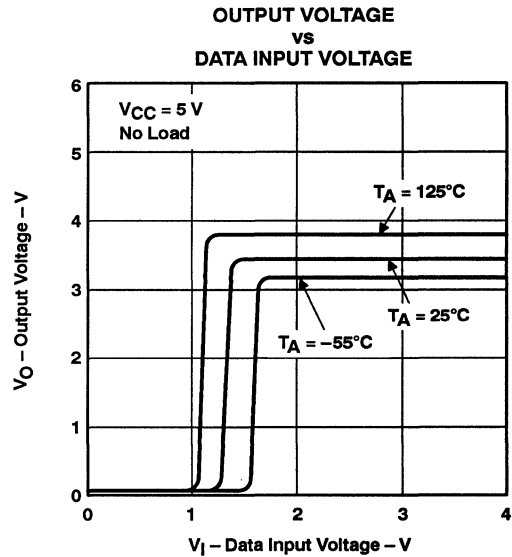


Figure 3

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  and for supply voltages below 4.75 V and above 5.25 V are applicable to SN55114 circuits only. These parameters were measured with the active pullup connected to the sink output.

# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

SLLS071A - D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT

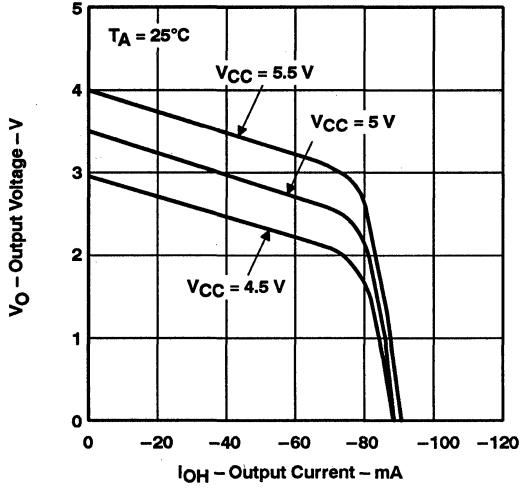


Figure 4

LOW-LEVEL OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT

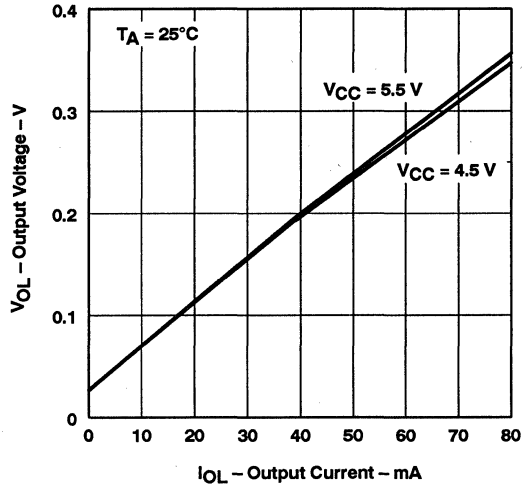


Figure 5

OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

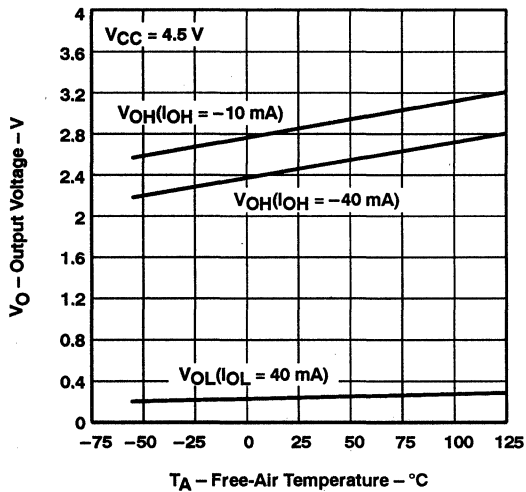


Figure 6

PROPAGATION DELAY TIMES  
vs  
FREE-AIR TEMPERATURE

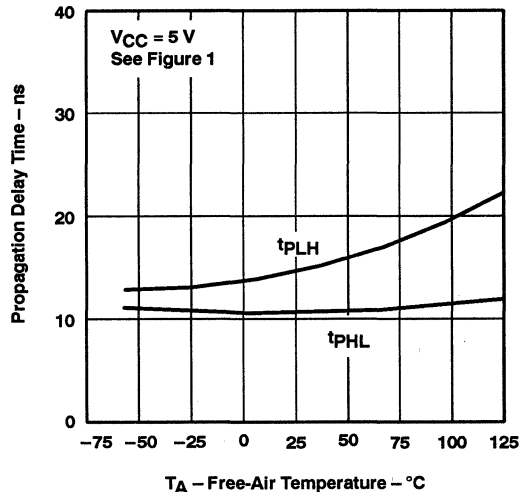


Figure 7

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  are applicable to SN55114 circuits only. These parameters were measured with the active pullup connected to the sink output.

# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

SLLS071A - D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS†

**SUPPLY CURRENT  
(BOTH DRIVERS)  
vs  
SUPPLY VOLTAGE**

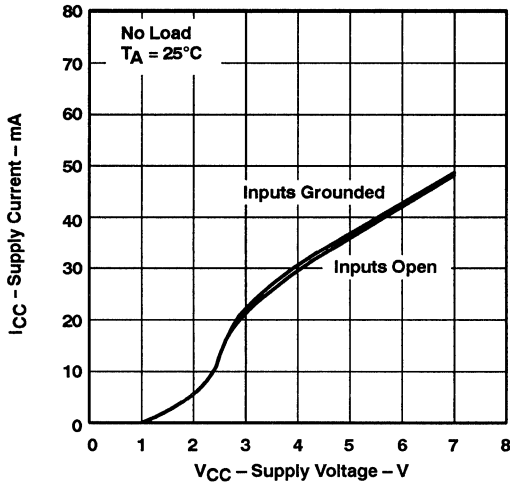


Figure 8

**SUPPLY CURRENT  
(BOTH DRIVERS)  
vs  
FREE-AIR TEMPERATURE**

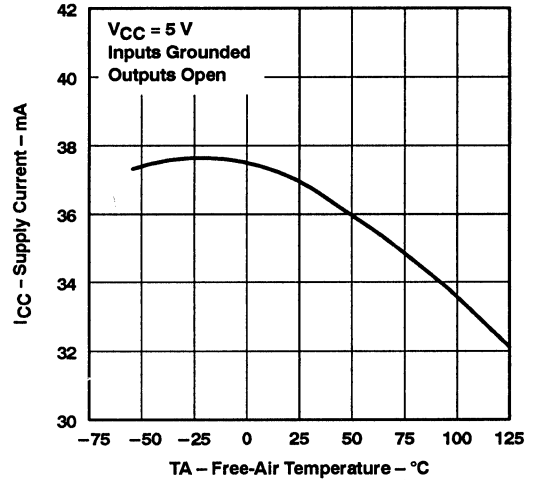


Figure 9

**SUPPLY CURRENT  
(BOTH DRIVERS)  
vs  
FREQUENCY**

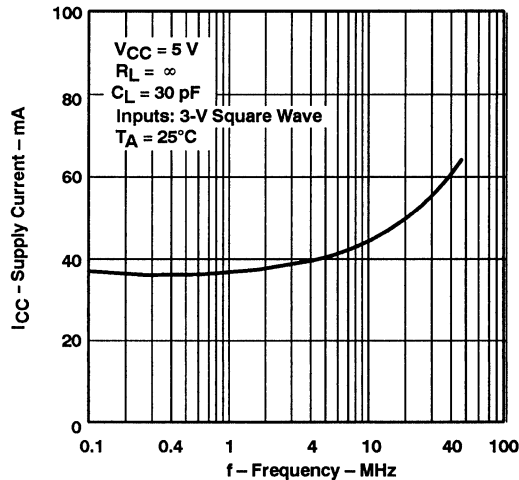


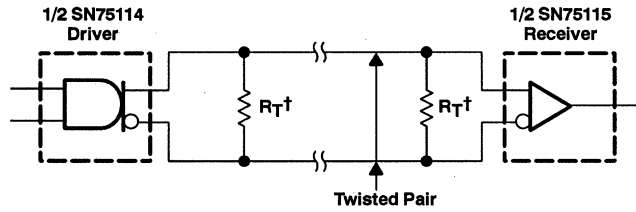
Figure 10

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  are applicable to SN55114 circuits only. These parameters were measured with the active pullup connected to the sink output.

# SN55114, SN75114 DUAL DIFFERENTIAL LINE DRIVERS

SLLS071A - D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## APPLICATION INFORMATION†



†  $R_T = Z_0$ . A capacitor may be connected in series with  $R_T$  to reduce power dissipation.

**Figure 11. Basic Party-Line or Data-Bus Differential Data Transmission**

# SN55115, SN75115 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS072A – D1315, SEPTEMBER 1973 – REVISED FEBRUARY 1993

- Choice of Open-Collector or Active Pullup (Totem-Pole) Outputs
- Single 5-V Supply
- Differential Line Operation
- Dual-Channel Operation
- TTL Compatible
- $\pm 15$ -V Common-Mode Input Voltage Range
- Optional-Use Built-In 130- $\Omega$  Line-Terminating Resistor
- Individual Frequency Response Controls
- Individual Channel Strobes
- Designed for Use With SN55113, SN75113, SN55114, and SN75114 Drivers
- Designed to Be interchangeable With Fairchild 9615 Line Receivers

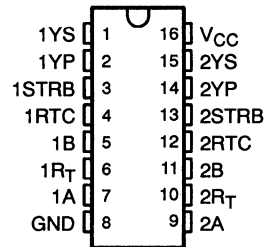
## description

The SN55115 and SN75115 dual differential line receivers are designed to sense small differential signals in the presence of large common-mode noise. These devices give TTL-compatible output signals as a function of the differential input voltage. The open-collector output configuration permits the wire-ANDing of similar TTL outputs (such as SN5401/SN7401) or other SN55115/SN75115 line receivers. This permits a level of logic to be implemented without extra delay. The output stages are similar to TTL totem-pole outputs, but with sink outputs, 1YS and 2YS, and the corresponding active pullup terminals, 1YP and 2YP, available on adjacent package pins. The frequency response and noise immunity may be provided by a single external capacitor. A strobe input is provided for each channel. With the strobe in the low level, the receiver is disabled and the outputs are forced to a high level.

The SN55115 is characterized for operation over the full military range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75115 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

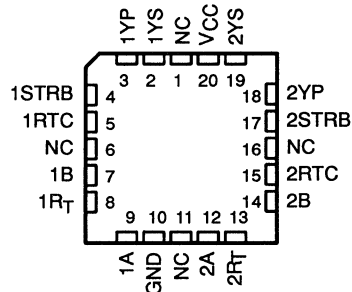
SN55115 . . . J OR W PACKAGE  
SN75115 . . . D OR N PACKAGE

(TOP VIEW)



SN55114 . . . FK PACKAGE

(TOP VIEW)



NC – No internal connection

FUNCTION TABLE

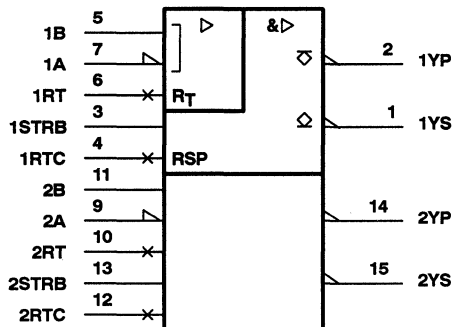
STRB	DIFF INPUT (A AND B)	OUTPUT (YP AND YS TIED TOGETHER)
L	X	H
H	L	H
H	H	L

H =  $V_I \geq V_{IH}$  min or  $V_{ID}$  more positive than  $V_{T+}$  max  
L =  $V_I \leq V_{IL}$  max or  $V_{ID}$  more negative than  $V_{T-}$  max  
X = irrelevant

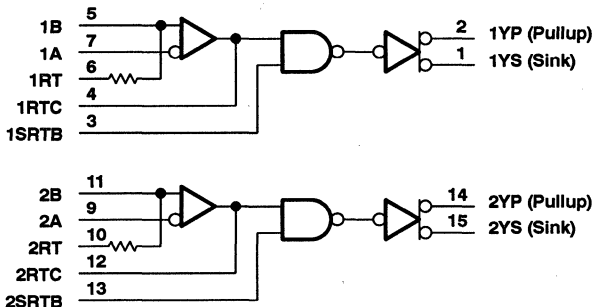
# SN55115, SN75115 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS072A - D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## logic symbol†

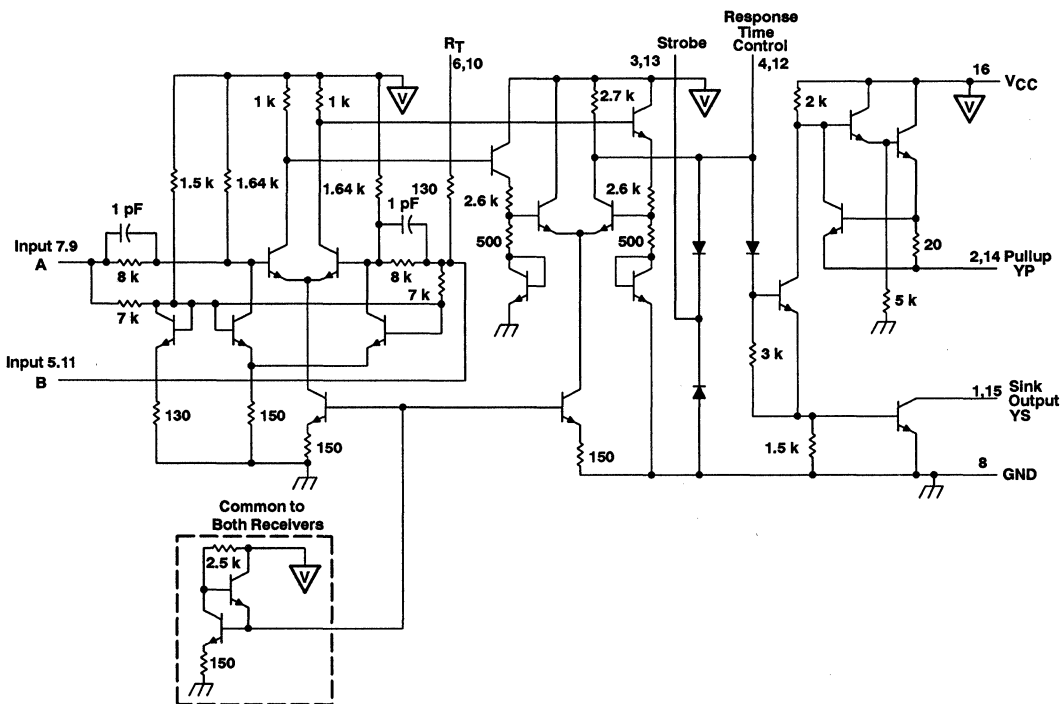


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematic (each receiver)



Resistor values are nominal and in ohms.  
Pin numbers shown are for D, J, N, and W packages.

# SN55115, SN75115 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS072A – D1315, SEPTEMBER 1973 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55115	SN75115	UNIT
Supply voltage, $V_{CC}$ (see Note 1)	7	7	V
Input voltage at A, B, and $R_T$	$\pm 25$	$\pm 25$	V
Input voltage at STRB	5.5	5.5	V
Off-state voltage applied to open-collector outputs	14	14	V
Continuous total power dissipation (see Note 2)	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds: FK package	260		$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	300		$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package		260	$^{\circ}\text{C}$

NOTE 1: All voltage values, except differential input voltage, are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW	—
FK†	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
J†	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW	—
W†	1000 mW	8.0 mW/ $^{\circ}\text{C}$	640 mW	200 mW

† In the FK, J, and W packages, SN55115 chips are either silver glass or alloy mounted. SN75115 chips are glass mounted.

## recommended operating conditions

	SN55115			SN75115			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.75	5	5.25	V
High-level input voltage at STRB, $V_{IH}$	2.4			2.4			V
Low-level input voltage at STRB, $V_{IL}$	0.4			0.4			V
High-level output current, $I_{OH}$	-5			-5			mA
Low-level output current, $I_{OL}$	15			15			mA
Operating free-air temperature, $T_A$	-55		125	0		70	$^{\circ}\text{C}$

  
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# SN55115, SN75115 DUAL DIFFERENTIAL LINE RECEIVERS

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electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITION†	SN55115			SN75115			UNIT	
		MIN	TYP‡	MAX	MIN	TYP‡	MAX		
$V_{T+}$ §	Positive-going threshold voltage	$V_O = 0.4$ V,	$I_{OL} = 15$ mA,	$V_{IC} = 0$	500			mV	
$V_{T-}$ §	Negative-going threshold voltage	$V_O = 2.4$ V,	$I_{OH} = -5$ mA,	$V_{IC} = 0$	-500¶			mV	
$V_{ICR}$	Common-mode input voltage range	$V_{ID} = \pm 1$ V			+15 to -15	+24 to -19	+15 to -15	+24 to -19	V
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ , $I_{OH} = -5$ mA	$V_{ID} = -0.5$ V,	$T_A = \text{MIN}$	2.2		2.4		V
				$T_A = 25^\circ\text{C}$	2.4	3.4	2.4	3.4	
				$T_A = \text{MAX}$	2.4		2.4		
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ , $I_{OL} = 15$ mA	$V_{ID} = -0.5$ V,	0.22 0.4			0.22 0.45	V	
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$ , Other input at 5.5 V	$V_I = 0.4$ V,	$T_A = \text{MIN}$	-0.9		-0.9		mA
				$T_A = 25^\circ\text{C}$	-0.5	-0.7	-0.5	-0.7	
				$T_A = \text{MAX}$	-0.7		-0.7		
$I_{SH}$	High-level strobe current	$V_{CC} = \text{MIN}$ , $V_{\text{strobe}} = 4.5$ V	$V_{ID} = -0.5$ V,	$T_A = 25^\circ\text{C}$	2		5		$\mu\text{A}$
				$T_A = \text{MAX}$	5		10		
$I_{SL}$	Low-level strobe current	$V_{CC} = \text{MAX}$ , $V_{\text{strobe}} = 0.4$ V	$V_{ID} = 0.5$ V,	$T_A = 25^\circ\text{C}$	-1.15	-2.4	-1.15	-2.4	mA
$I(\text{RTC})$	Response-time-control current	$V_{CC} = \text{MAX}$ , $V_{RC} = 0$	$V_{ID} = 0.5$ V,	$T_A = 25^\circ\text{C}$	-1.2	-3.4	-1.2	-3.4	mA
$I_{O(\text{off})}$	Off-state open-collector output current	$V_{CC} = \text{MIN}$ , $V_{ID} = -4.5$ V	$V_{OH} = 12$ V,	$T_A = 25^\circ\text{C}$	100				$\mu\text{A}$
				$T_A = \text{MAX}$	200				
				$T_A = 25^\circ\text{C}$			100		
				$T_A = \text{MAX}$			200		
$R_T$	Line-terminating resistance	$V_{CC} = 5$ V		$T_A = 25^\circ\text{C}$	77 130 167	74 130 179		$\Omega$	
$I_{OS}$	Supply-circuit output current#	$V_{CC} = \text{MAX}$ , $V_O = 0$	$V_{ID} = -0.5$ V,	$T_A = 25^\circ\text{C}$	-15 -40 -80	-14 -40 -100		mA	
$I_{CC}$	Supply current (both receivers)	$V_{CC} = \text{MAX}$ , $V_{IC} = 0$	$V_{ID} = 0.5$ V,	$T_A = 25^\circ\text{C}$	32 50	32 50		mA	

† Unless otherwise noted,  $V_{\text{strobe}} = 2.4$  V. All parameters with the exception of off-state open-collector output current are measured with the active pull-up connected to the sink output.

‡ All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ , and  $V_{IC} = 0$ .

§ Differential voltages are at the B input terminal with respect to the A input terminal.

¶ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold voltages only.

# Only one output should be shorted to ground at a time, and duration of the short circuit should not exceed one second.



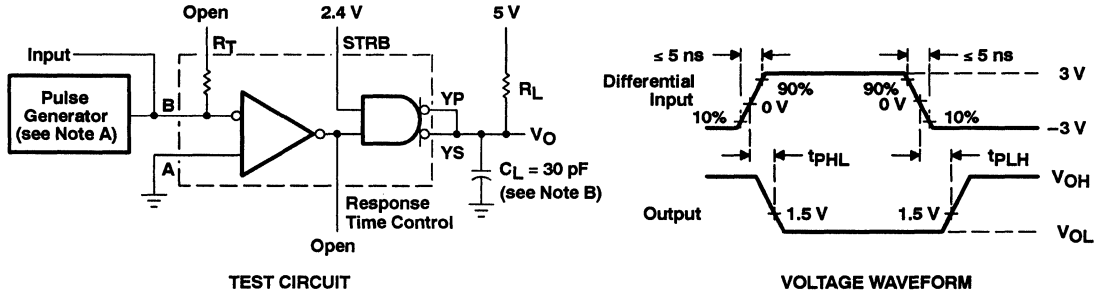
# SN55115, SN75115 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS072A - D1315, SEPTEMBER 1973 - REVISED FEBRUARY 1993

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 30\text{ pF}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	SN55115			SN75115			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$t_{PLH}$ Propagation delay time, low-to-high level output	$R_L = 3.9\text{ k}\Omega$ , See Figure 1		18	50	18	75		ns
$t_{PHL}$ Propagation delay time, high-to-low level output	$R_L = 390\ \Omega$ , See Figure 1		20	50	20	75		ns

## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50\ \Omega$ ,  $PRR \leq 500\text{ kHz}$ ,  $t_w \leq 100\text{ ns}$ , duty cycle = 50%.  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

## TYPICAL CHARACTERISTICS

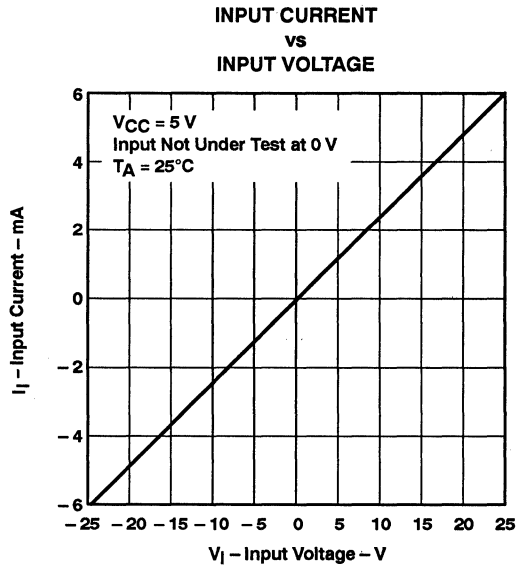


Figure 2

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# SN55115, SN75115 DUAL DIFFERENTIAL LINE RECEIVERS

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## TYPICAL CHARACTERISTICS†

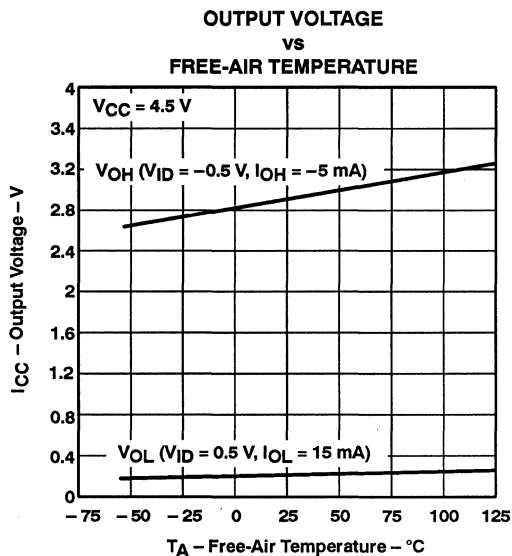


Figure 3

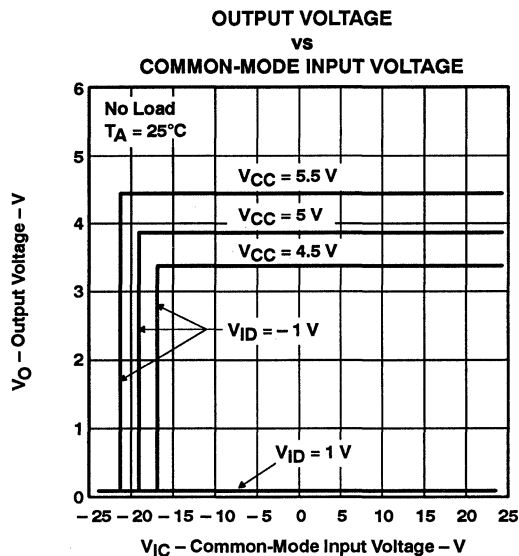


Figure 4

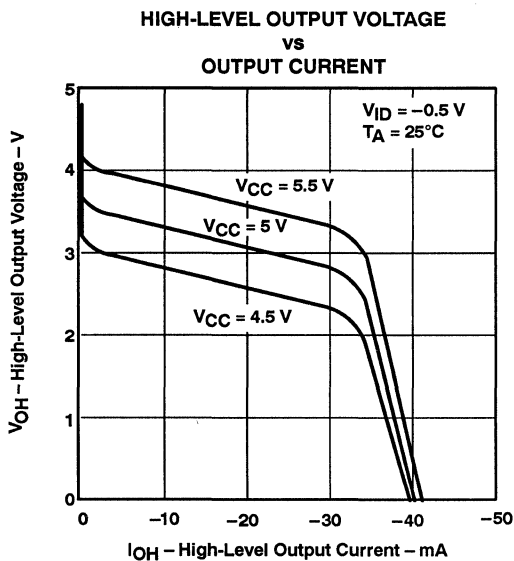


Figure 5

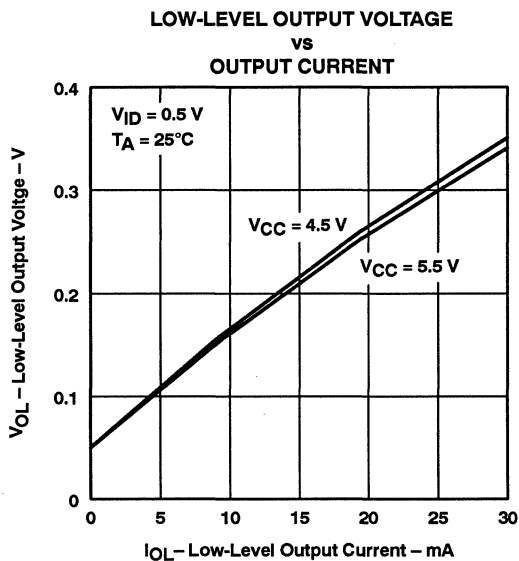


Figure 6

† Data for temperatures below 0°C and above 70°C and for supply voltages below 4.75 V and above 5.25 V are applicable to SN55115 circuits only. These parameters were measured with the active pullup connected to the sink output.

TYPICAL CHARACTERISTICS†

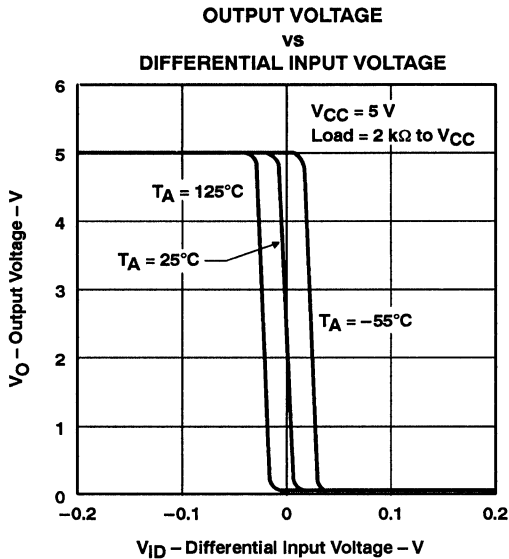


Figure 7

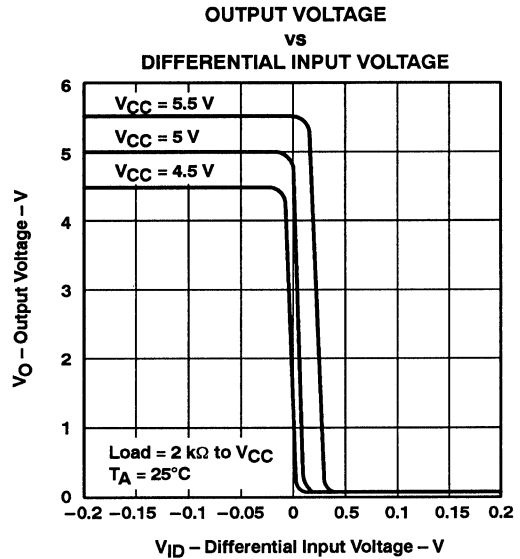


Figure 8

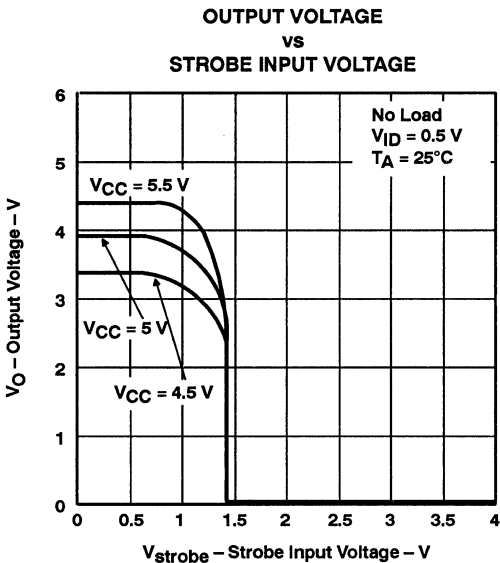


Figure 9

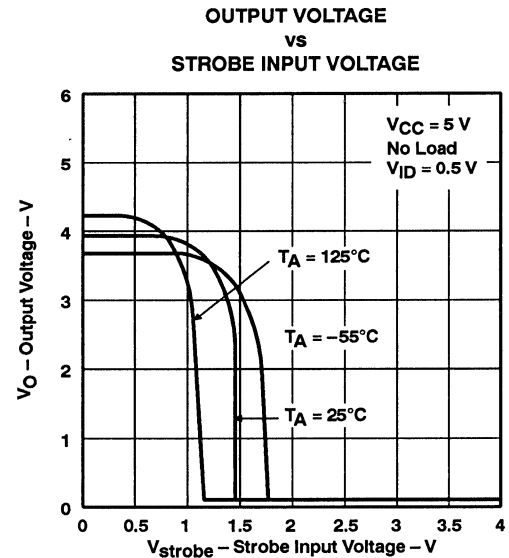


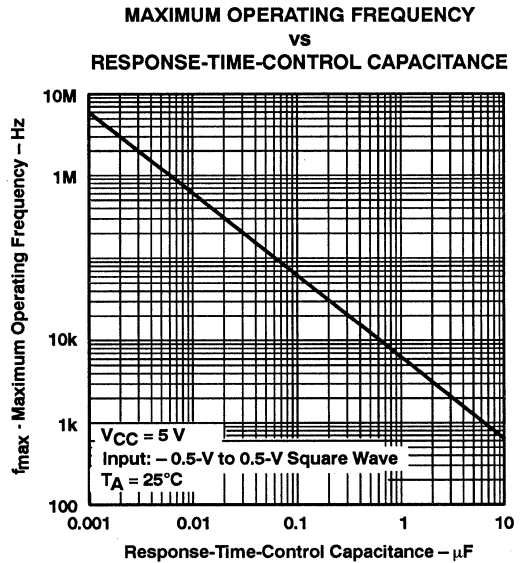
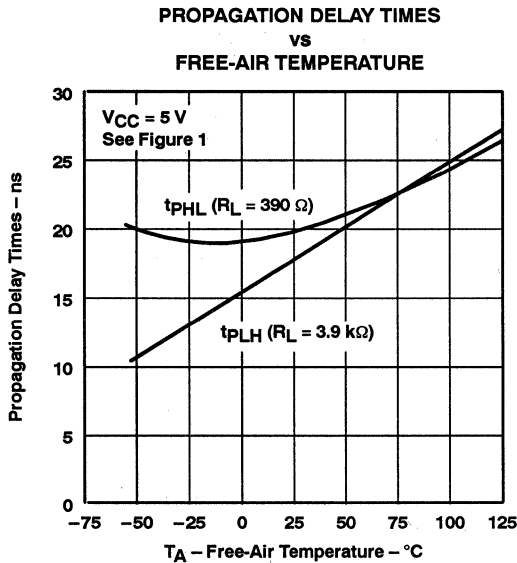
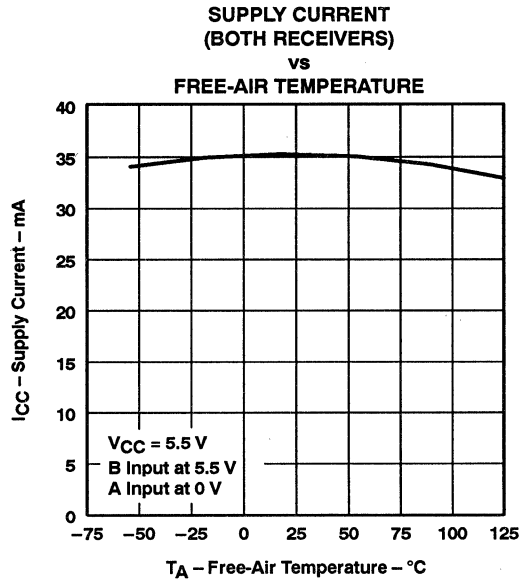
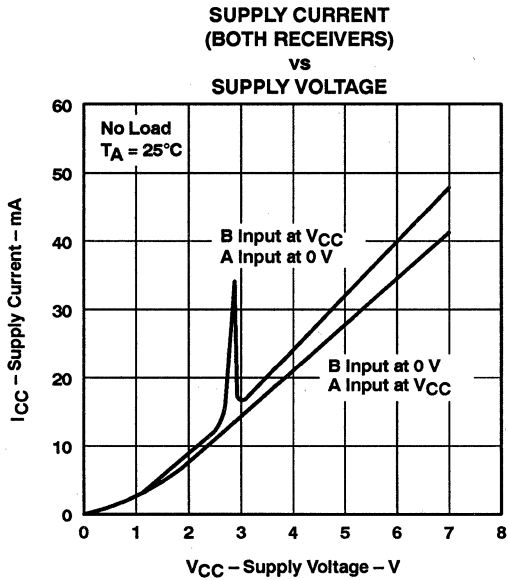
Figure 10

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  and for supply voltages below 4.75 V and above 5.25 V are applicable to SN55115 circuits only. These parameters were measured with the active pullup connected to the sink output.

# SN55115, SN75115 DUAL DIFFERENTIAL LINE RECEIVERS

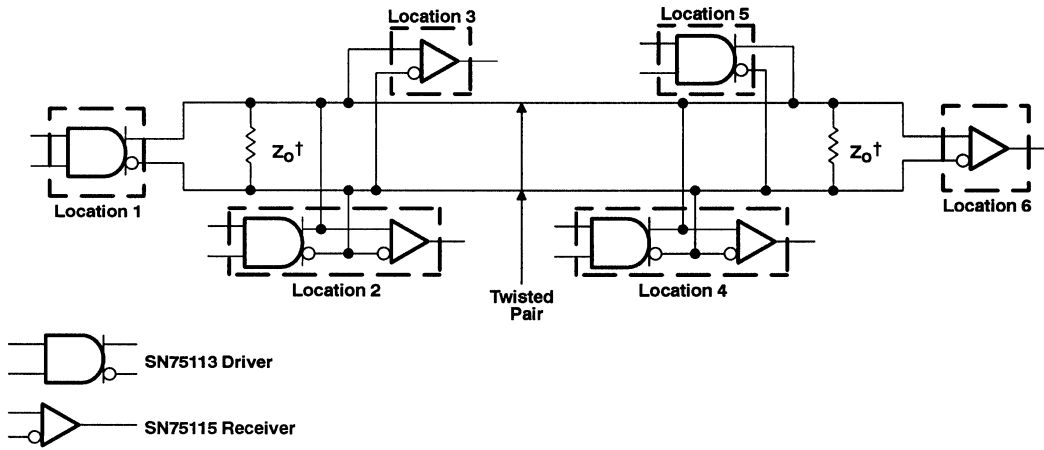
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## TYPICAL CHARACTERISTICS†



† Data for temperatures below 0°C and above 70°C and for supply voltages below 4.75 V and above 5.25 V are applicable to SN55115 circuits only. These parameters were measured with the active pullup connected to the sink output.

APPLICATION INFORMATION



† A capacitor may be connected in series with  $Z_o$  to reduce power dissipation.

Figure 15. Basic Party-Line or Data-Bus Differential Data Transmission



# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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## features common to all types

- Single 5-V Supply
- 3-State Driver Output Circuitry
- TTL-Compatible Driver Inputs
- TTL-Compatible Receiver Output
- Differential Line Operation
- Receiver Output Strobe ('116, SN75117) or Enable (SN75118, SN75119)
- Designed for Party-Line (Data-Bus) Applications
- Choice of Ceramic or Plastic Packages

## additional features of the SN55116/SN75116

- Independent Driver and Receiver
- Choice of Open-Collector or Totem-Pole Outputs on Both Driver and Receiver
- Dual Data Inputs on Driver
- Optional Line-Termination Resistor in Receiver
- $\pm 15$ -V Receiver Common-Mode Capability
- Receiver Frequency Response Control

## additional features of the SN75117

- Driver Output Internally Connected to Receiver Input

**The SN75118 is an SN75116 With 3-State Receiver Output Circuitry**

**The SN75119 is an SN75117 With 3-State Receiver Output Circuitry**

## description

These integrated circuits are designed for use in interfacing between TTL-type digital systems and differential data transmission lines. They are especially useful for party-line (data-bus) applications. Each of these circuit types combine in one package a 3-state differential line driver and a differential-input line receiver, both of which operate from a single 5-V power supply. The driver inputs and receiver outputs are TTL compatible. The driver employed is similar to the SN55113/SN75113 3-state line driver, and the receiver is similar to the SN55115/SN75115 line receiver.

The '116 and SN75118 circuits offer all the features of the SN55113/SN75113 driver and the SN55115/SN75115 receiver combined. The driver performs the dual input AND and NAND functions when enabled or presents a high impedance to the load when in the disabled state. The driver output stages are similar to TTL totem-pole outputs, but have the current-sink portion separated from the current-sourcing portion and both are brought out to adjacent package pins. This feature allows the user the option of using the driver in the open-collector output configuration, or, by connecting the adjacent source and sink pins together, of using the driver in the normal totem-pole output configuration.

The receiver portion of the '116 and SN75118 features a differential-input circuit having a common-mode voltage range of  $\pm 15$  V. An internal 130- $\Omega$  resistor is also provided, which may optionally be used for terminating the transmission line. A frequency response control pin allows the user to reduce the speed of the receiver or to improve differential noise immunity. The receiver of the '116 also has an output strobe and a split totem-pole output. The receiver of the SN75118 has an output-enable for the 3-state split totem-pole output. The receiver section of either circuit is independent of the driver section except for the  $V_{CC}$  and ground pins.

The SN75117 and SN75119 circuits provide the basic driver and receiver functions of the '116 and SN75118, but use a package that is only half as large. The SN75117 and SN75119 are intended primarily for party-line or bus-organized systems as the driver outputs are internally connected to the receiver inputs. The driver has a single data input and a single enable input, and the SN75117 receiver has an output strobe while the SN75119 receiver has a 3-state-output enable. These devices do not, however, provide output connection options, line termination resistors, or receiver frequency-response controls.

The SN55116 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75116, SN75117, SN75118, and SN75119 are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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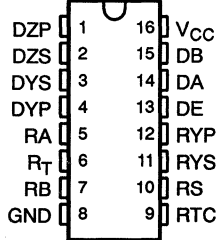
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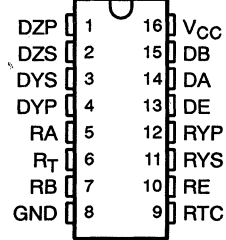
# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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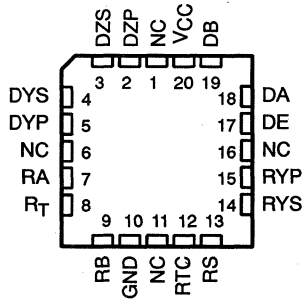
SN55116 . . . J PACKAGE  
SN75116 . . . D OR N PACKAGE  
(TOP VIEW)



SN75118 . . . D OR N PACKAGE  
(TOP VIEW)

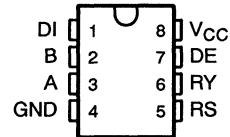


SN55116 . . . FK PACKAGE  
(TOP VIEW)

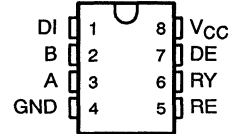


NC - No internal connection

SN75117 . . . D OR P PACKAGE  
(TOP VIEW)



SN75119 . . . D OR P PACKAGE  
(TOP VIEW)



## Function Tables

'116, SN75118  
DRIVER

INPUTS			OUTPUTS	
DE	DA	DB	DY	DZ
L	X	X	Z	Z
H	L	X	L	H
H	X	L	L	H
H	H	H	H	L

'116, SN75118  
RECEIVER

RS/RE	DIFF INPUT	OUTPUTS RY	
		'116	SN75118
L	X	H	Z
H	L	H	H
H	H	L	L

SN75117, SN75119  
DRIVER

INPUTS		OUTPUTS	
DI	DE	A	B
H	H	H	L
L	H	L	H
X	L	Z	Z

SN75117, SN75119  
RECEIVER

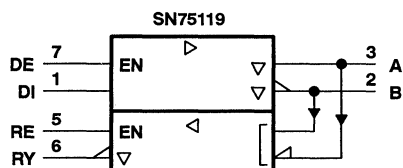
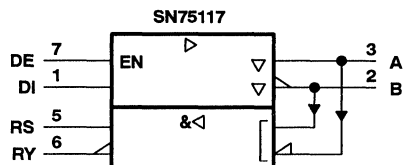
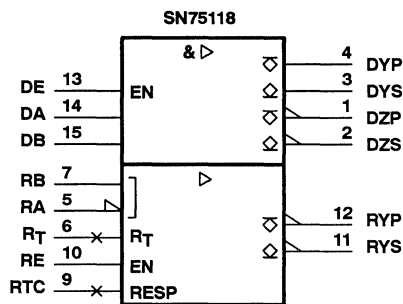
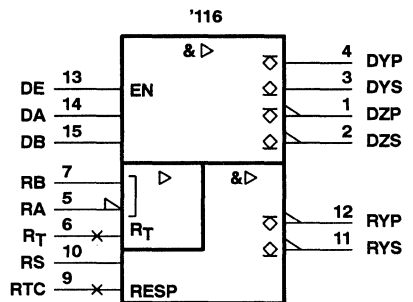
INPUTS			OUTPUT RY	
A	B	RS/RE	SN75117	SN75119
H	L	H	H	H
L	H	H	L	L
X	X	L	H	Z

H = high level ( $V_I \geq V_{IH}$  min or  $V_{ID}$  more positive than  $V_{TH}$  max)  
L = low level ( $V_I \leq V_{IL}$  max or  $V_{ID}$  more negative than  $V_{TL}$  max)  
X = irrelevant  
Z = high impedance (off)

# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

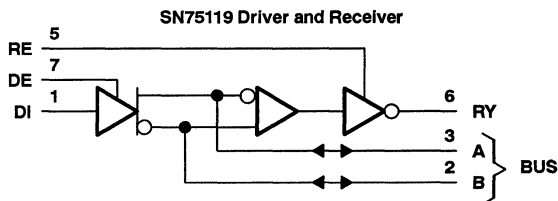
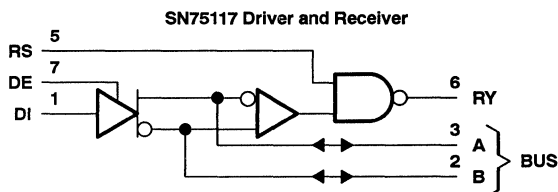
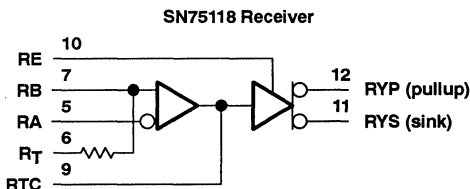
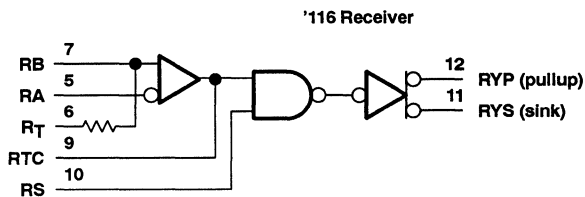
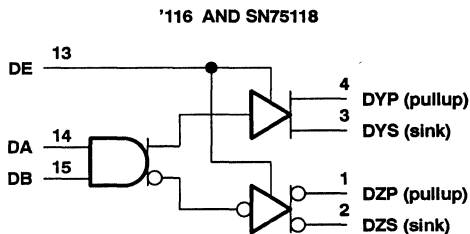
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## logic symbols†



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagrams (positive logic)

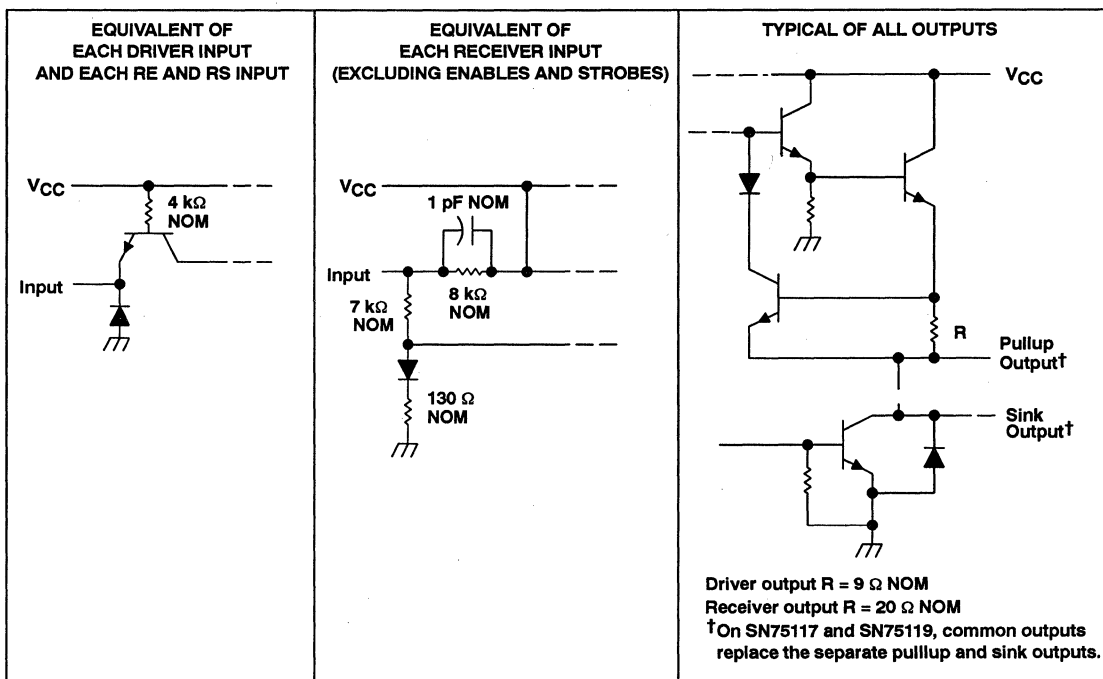


Pin numbers shown for the SN55116 are for the J package, those shown for the SN75118 are for the N package, those shown for SN75117 and SN75119 are for the P package.

# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

		'116, SN75118	SN75117, SN75119	UNIT
Supply voltage, $V_{CC}$ (see Note 1)		7	7	V
Input voltage, $V_I$	DA, DB, DE, DI, RE, RS	5.5	5.5	V
	RA, RB, RT	±25		
	A and B		0 to 6	
Off-state voltage applied to open-collector outputs		12		V

		SN55116	SN75116 THRU SN75119	UNIT
Continuous total power dissipation (see Note 2)		See Dissipation Rating Table		
Operating free-air temperature range		-55 to 125	0 to 70	°C
Storage temperature range		-65 to 50	-65 to 50	°C
Case temperature for 60 seconds: FK package		260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package		300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or P package			260	°C

NOTES: 1. All voltage values are with respect to network ground terminal.

2. In the FK and J packages, SN55116 chip is alloy mounted and SN75116 through SN75119 chips are glass mounted.

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# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D (8 pin)	725 mW	5.8 mW/ $^\circ\text{C}$	464 mW	—
D (16 pin)	950 mW	7.6 mW/ $^\circ\text{C}$	608 mW	—
FK	1375 mW	11.0 mW/ $^\circ\text{C}$	880 mW	275 mW
J	1375 mW	11.0 mW/ $^\circ\text{C}$	880 mW	275 mW
N	1150 mW	9.2 mW/ $^\circ\text{C}$	736 mW	—
P	1000 mW	8.0 mW/ $^\circ\text{C}$	640 mW	—

## recommended operating conditions

PARAMETER		SN55116			SN75'			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
Supply voltage, $V_{CC}$		4.5	5	5.5	4.5	5	5.5	V
High-level input voltage, $V_{IH}$	All inputs except differential inputs	2			2			V
Low-level input voltage, $V_{IL}$		0.8			0.8			V
High-level output current, $I_{OH}$	Drivers	-40			-40			mA
	Receivers	-5			-5			
Low-level output current, $I_{OL}$	Drivers	40			40			mA
	Receivers	15			15			
Receiver input voltage, $V_I$	'116, '118	$\pm 15$			$\pm 15$			V
	'117, '119	0	6	6	0	6	6	
Common-mode receiver input voltage, $V_{ICR}$	'116, '118	$\pm 15$			$\pm 15$			V
	'117, '119	0	6	6	0	6	6	
Operating free-air temperature, $T_A$		-55		125	0		70	$^\circ\text{C}$

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

driver section

PARAMETER		TEST CONDITIONS†			'116, SN75118		SN75117, SN75119		UNIT
					MIN	TYP‡	MAX	MIN	
V <sub>IK</sub>	Input clamp voltage	V <sub>CC</sub> = MIN, I <sub>I</sub> = -12 mA			-0.9	-1.5	-0.9	-1.5	V
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN, V <sub>IL</sub> = 0.8 V, I <sub>IH</sub> = 2 V	T <sub>A</sub> = 25°C (SN55116)	I <sub>OH</sub> = -10 mA	2.4	3.4	2.4	3.4	V
			T <sub>A</sub> = 0°C to 70°C (SN75')	I <sub>OH</sub> = -40 mA	2	3	2	3	
			T <sub>A</sub> = -55°C to 125°C (SN55')	I <sub>OH</sub> = -10 mA	2		2		
				I <sub>OH</sub> = -40 mA	1.8		1.8		
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN, V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V, I <sub>OL</sub> = 40 mA			0.4		0.4		V
V <sub>OK</sub>	Output clamp voltage	V <sub>CC</sub> = MAX, I <sub>O</sub> = -40 mA, DE at 0.8 V			-1.5		-1.5		V
I <sub>O(off)</sub>	Off-state open-collector output current	V <sub>CC</sub> = MAX, V <sub>O</sub> = 12 V	T <sub>A</sub> = 25°C		1	10			μA
			T <sub>A</sub> = MAX	SN55116	200				
				SN75'	20				
I <sub>OZ</sub>	Off-state (high-impedance-state) output current	V <sub>CC</sub> = MAX, DE at 0.8 V, T <sub>A</sub> = MAX	V <sub>O</sub> = 0 to V <sub>CC</sub> , DE at 0.8 V, T <sub>A</sub> = 25°C		±10				μA
			V <sub>O</sub> = 0	SN55116	-300				
			V <sub>O</sub> = 0.4 V to V <sub>CC</sub>	SN55116	±150				
			V <sub>O</sub> = 0 to V <sub>CC</sub>	SN75'	±20				
I <sub>I</sub>	Input current at maximum input voltage	Driver or enable input	V <sub>CC</sub> = MAX, V <sub>I</sub> = 5.5 V		1		1		mA
I <sub>IH</sub>	High-level input current		V <sub>CC</sub> = MAX, V <sub>I</sub> = 2.4 V		40		40		μA
I <sub>IL</sub>	Low-level input current		V <sub>CC</sub> = MAX, V <sub>I</sub> = 0.4 V		-1.6		-1.6		mA
I <sub>OS</sub>	Short-circuit output current§	V <sub>CC</sub> = MAX, V <sub>O</sub> = 0, T <sub>A</sub> = 25°C			-40	-120	-40	-120	mA
I <sub>CC</sub>	Supply current (driver and receiver combined)	V <sub>CC</sub> = MAX, T <sub>A</sub> = 25°C			42	60	42	60	mA

† All parameters with the exception of off-state open-collector output current are measured with the active pullup connected to the sink output. For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

§ Not more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

SLLS073A – D2143, MAY 1976 – REVISED FEBRUARY 1993

**switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 30\text{ pF}$ ,  $T_A = 25^\circ\text{C}$**

**driver section**

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
tPLH	Propagation delay time, low-to-high level output	See Figure 13			14	30	ns
tPHL	Propagation delay time, high-to-low level output				12	30	
tpZH	Output enable time to high level	$R_L = 180\ \Omega$ ,	See Figure 14		8	20	ns
tpZL	Output enable time to low level	$R_L = 250\ \Omega$ ,	See Figure 15		17	40	ns
tPHZ	Output disable time from high level	$R_L = 180\ \Omega$ ,	See Figure 14		16	30	ns
tPLZ	Output disable time from low level	$R_L = 250\ \Omega$ ,	See Figure 15		20	35	ns

**electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)**
**receiver section**

PARAMETER		TEST CONDITIONS†		'116, SN75118		SN75117, SN75119		UNIT		
				MIN	TYP‡	MAX	MIN		TYP‡	MAX
V <sub>T+</sub>	Positive-going threshold voltage§	V <sub>O</sub> = 0.4 V, See Note 3	I <sub>OL</sub> = 15 mA,	V <sub>CC</sub> = MIN, V <sub>ICR</sub> = 0, See Note 4	0.5		0.5		V	
				V <sub>CC</sub> = 5 V, V <sub>ICR</sub> = MAX, See Note 5	1		1			
V <sub>T-</sub>	Negative-going threshold voltage§	V <sub>O</sub> = 2.4 V, See Note 3	I <sub>OL</sub> = -5 mA,	V <sub>CC</sub> = MIN, V <sub>ICR</sub> = 0, See Note 4	-0.5¶		-0.5¶		V	
				V <sub>CC</sub> = 5 V, V <sub>ICR</sub> = MAX, See Note 5	-1¶		-1¶			
V <sub>I</sub>	Input voltage range#	V <sub>CC</sub> = 5 V,	V <sub>ID</sub> = -1 V or 1 V,	See Note 3		15 to -15	6 to 0	V		
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -5 mA, See Note 3	V <sub>CC</sub> = MIN, V <sub>ICR</sub> = 0,	V <sub>ID</sub> = -0.5 V, See Notes 4 and 6	2.4		2.4		V	
			V <sub>CC</sub> = 5 V, V <sub>ICR</sub> = MAX,	V <sub>ID</sub> = -1 V, See Note 5	2.4		2.4			
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 15 mA, See Note 3	V <sub>CC</sub> = MIN, V <sub>ICR</sub> = 0,	V <sub>ID</sub> = 0.5 V, See Notes 4 and 7	0.4		0.4		V	
			V <sub>CC</sub> = 5 V, V <sub>ICR</sub> = MAX,	V <sub>ID</sub> = 1 V, See Note 5	0.4		0.4			
I <sub>I(rec)</sub>	Receiver input current	V <sub>CC</sub> = MAX, See Note 3	V <sub>I</sub> = 0,	Other input at 0 V	-0.5	-0.9	-0.5	-1	mA	
			V <sub>I</sub> = 0.4 V,	Other input at 2.4 V	-0.4	-0.7	-0.4	-0.8		
			V <sub>I</sub> = 2.4 V,	Other input at 0.4 V	0.1	0.3	0.1	0.4		
I <sub>I</sub>	Input current at maximum input voltage	Strobe	V <sub>CC</sub> = MIN, V <sub>strobe</sub> = 4.5 V	V <sub>ID</sub> = -0.5 V	'116, SN75117		5		5	μA
			Enable	V <sub>CC</sub> = MAX,	V <sub>I</sub> = 5.5 V	SN75118, SN75119		1		1

† Unless otherwise noted, V<sub>strobe</sub> = 2.4 V. All parameters with the exception of off-state open-collector output current are measured with the active pullup connected to the sink output.

For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C, and V<sub>IC</sub> = 0.

§ Differential voltages are at the B input terminal with respect to the A input terminal. Neither receiver input of the SN75117 or SN75119 should be taken negative with respect to GND.

¶ The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold voltages only.

# Input voltage range is the voltage range that, if exceeded at either input, will cause the receiver to cease functioning properly.

NOTES: 3. Measurement of these characteristics on the SN75117 and SN75119 requires the driver to be disabled with the driver enable at 0.8 V.

4. This applies with the less positive receiver input grounded.

5. For '116 and SN75118, this applies with the more positive receiver input at 15 V or the more negative receiver input at -15 V. For SN75117 and SN75119, this applies with the more positive receiver input at 6 V.

6. For SN55116, V<sub>ID</sub> = -1 V.

7. For SN55116, V<sub>ID</sub> = 1 V.

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

receiver section

PARAMETER			TEST CONDITIONS†			'116, SN75118		SN75117, SN75119		UNIT
						MIN	TYP‡	MAX	MIN	
I <sub>IH</sub>	High-level input current	Enable	V <sub>CC</sub> = MAX, V <sub>I</sub> = 2.4 V	SN75118, SN75119		40		40	μA	
I <sub>I</sub>	Low-level input current	Strobe	V <sub>CC</sub> = MAX, V <sub>strobe</sub> = 0.4 V, V <sub>ID</sub> = 0.5 V, See Notes 4 and 7	'116, SN75117		-2.4		-2.4	mA	
		Enable	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0.4 V	SN75118, SN75119		-1.6		-1.6		
I <sub>(RC)</sub>	Response-time-control current (RTC)		V <sub>CC</sub> = MAX, RC at 0 V, V <sub>ID</sub> = 0.5 V, See Notes 4 and 7	T <sub>A</sub> = 25°C		-1.2			mA	
I <sub>O(off)</sub>	Off-state open-collector output current		V <sub>CC</sub> = MAX, V <sub>O</sub> = 12 V, V <sub>ID</sub> = -1 V	T <sub>A</sub> = 25°C		1	10		μA	
				T <sub>A</sub> = MAX	SN55116, SN75_		200	20		
I <sub>OZ</sub>	Off-state (high-impedance state) output current		V <sub>CC</sub> = MAX, V <sub>O</sub> = 0 to V <sub>CC</sub> , RE at 0.4 V	T <sub>A</sub> = 25°C	SN75118, SN75119		±10		±10	
				T <sub>A</sub> = MAX	SN75118		±20			
					SN75119			±20		
R <sub>T</sub>	Line-terminating resistance		V <sub>CC</sub> = 5 V	T <sub>A</sub> = 25°C		77	167		Ω	
I <sub>OS</sub>	Short-circuit output current§		V <sub>CC</sub> = MAX, V <sub>ID</sub> = -0.5 V, V <sub>O</sub> = 0, See Notes 4 and 6	T <sub>A</sub> = 25°C		-15	-80	-15	-80	mA
I <sub>CC</sub>	Short current (driver and receiver combined)		V <sub>CC</sub> = MAX, V <sub>ID</sub> = 0.5 V, See Notes 4 and 7	T <sub>A</sub> = 25°C		42	60	42	60	mA

† Unless otherwise noted V<sub>strobe</sub> = 2.4 V. All parameters with the exception of off-state open-collector output current are measured with the active pullup connected to the sink output.

For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C, and V<sub>IC</sub> = 0.

§ Not more than one output should be shorted at a time.

NOTES: 4. This applies with the less positive receiver input grounded.

6. For SN55116, V<sub>ID</sub> = -1 V.

7. For SN55116, V<sub>ID</sub> = 1 V.



# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 30\text{ pF}$ ,  $T_A = 25^\circ\text{C}$

receiver section

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$R_L = 400\ \Omega$ , See Figure 16		20	75	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output			17	75	ns
$t_{PZH}$	Output enable time to high level	$R_L = 480\ \Omega$ , See Figure 14		9	20	ns
$t_{PZL}$	Output enable time to low level	$R_L = 250\ \Omega$ , See Figure 15		16	35	ns
$t_{PHZ}$	Output disable time from high level	$R_L = 480\ \Omega$ , See Figure 14		12	30	ns
$t_{PLZ}$	Output disable time from low level	$R_L = 250\ \Omega$ , See Figure 15		17	35	ns

## TYPICAL CHARACTERISTICS

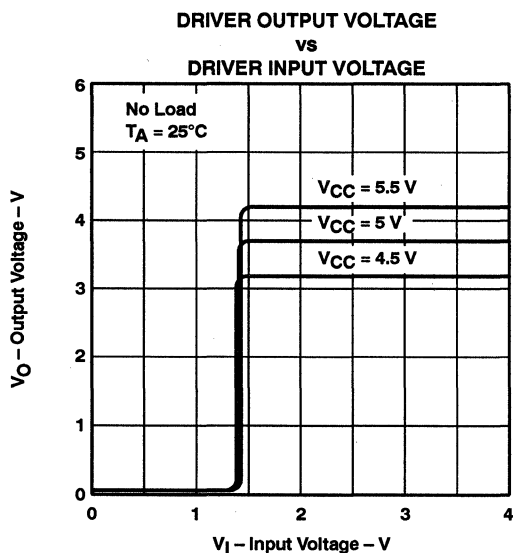


Figure 1

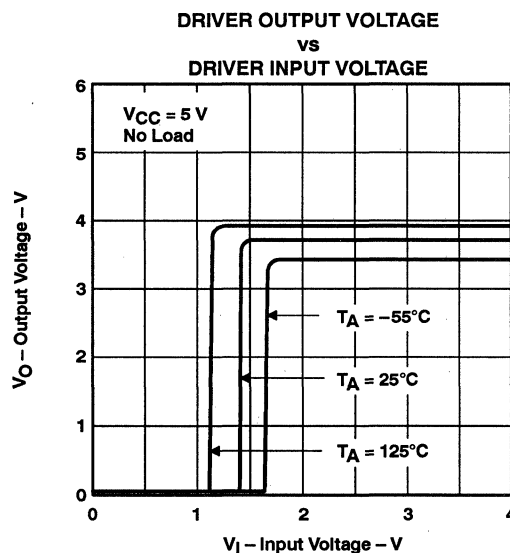
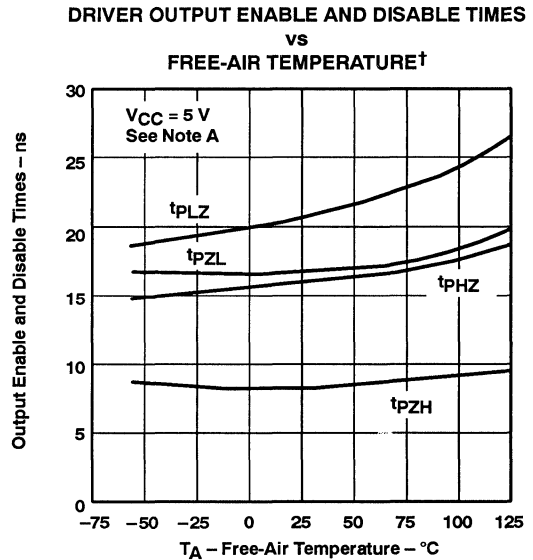
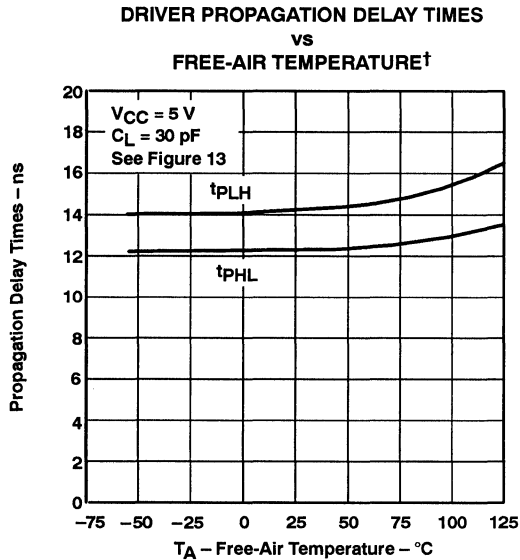
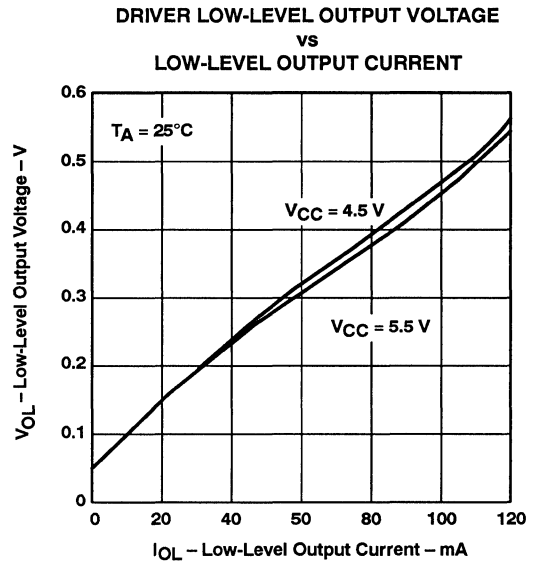
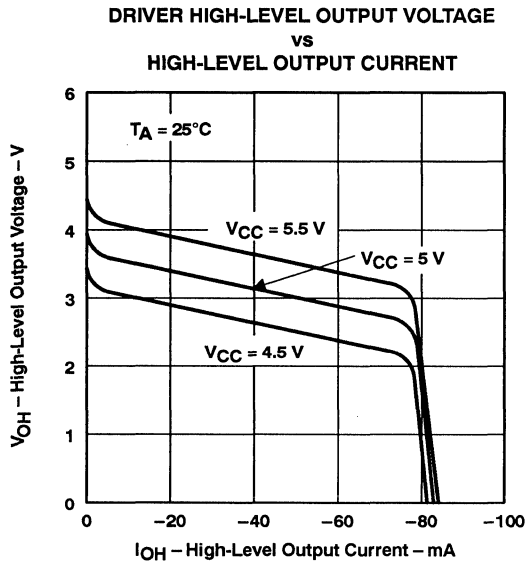


Figure 2

# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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## TYPICAL CHARACTERISTICS



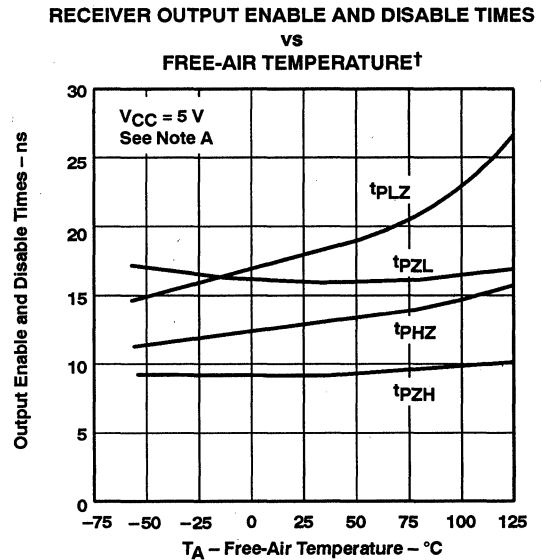
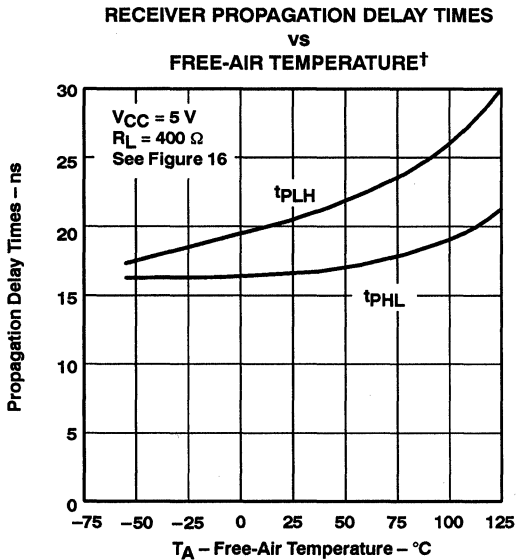
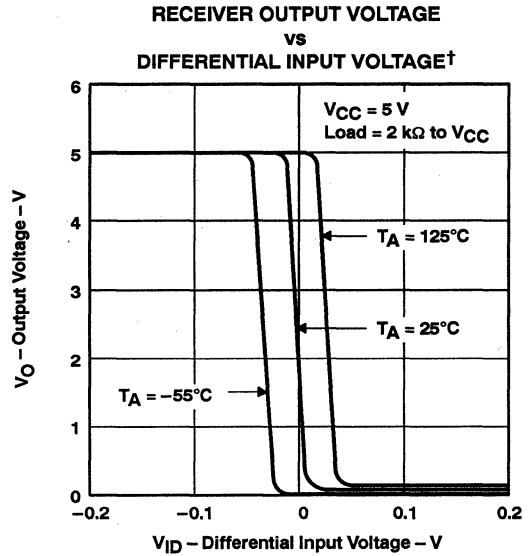
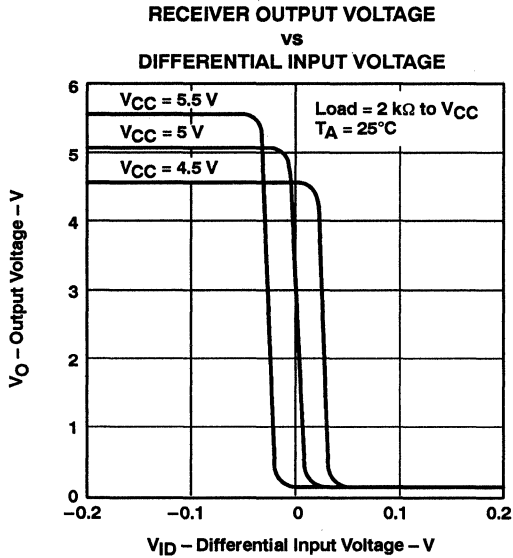
† Data for temperatures below 0°C and above 70°C are applicable to SN55116.

NOTE A: For  $t_{pZH}$  and  $t_{pHZ}$ :  $R_L = 180 \Omega$ , see Figure 14. For  $t_{pZL}$  and  $t_{pLZ}$ :  $R_L = 250 \Omega$ , see Figure 15.

# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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## TYPICAL CHARACTERISTICS



† Data for temperatures below 0°C and above 70°C are applicable to SN55116.

NOTE A: For  $t_{pZH}$  and  $t_{pHZ}$ :  $R_L = 480 \Omega$ , see Figure 14. For  $t_{pZL}$  and  $t_{pLZ}$ :  $R_L = 250 \Omega$ , see Figure 15.

**TEXAS  
INSTRUMENTS**

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# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

SUPPLY CURRENT (DRIVER AND RECEIVER)  
vs  
SUPPLY VOLTAGE

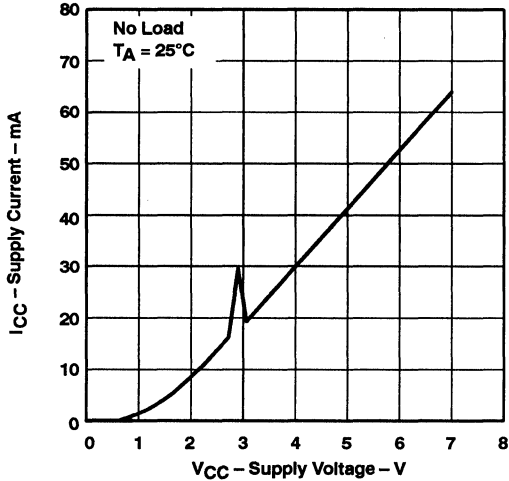


Figure 11

SUPPLY CURRENT (DRIVER AND RECEIVER)  
vs  
FREE-AIR TEMPERATURE†

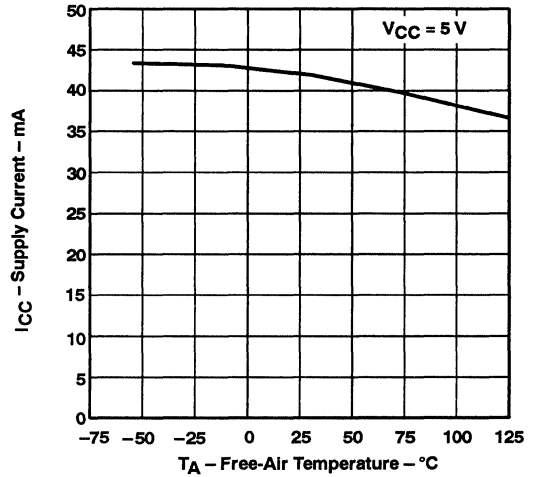


Figure 12

† Data for temperatures below 0°C and above 70°C are applicable to SN55116.

# SN55116, SN75116 THRU SN75119 DIFFERENTIAL LINE TRANSCIEVERS

SLLS073A - D2143, MAY 1976 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

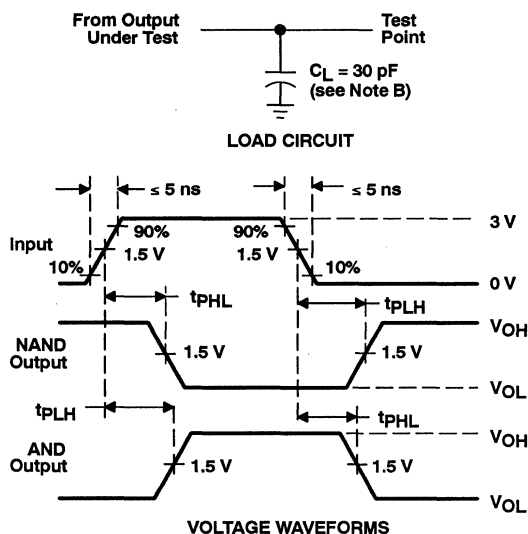


Figure 13.  $t_{pLH}$  and  $t_{pHL}$  (drivers only)

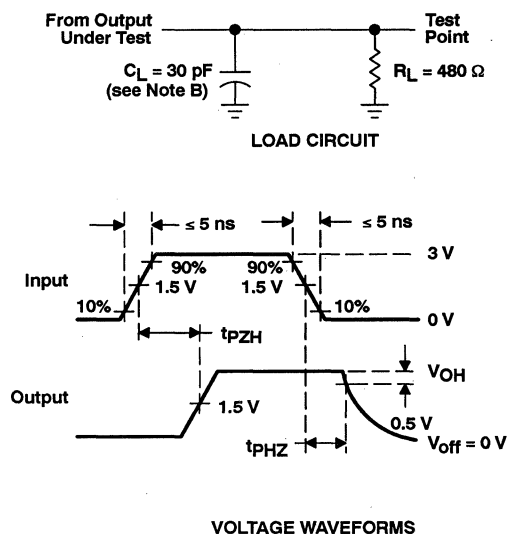


Figure 14.  $t_{pZH}$  and  $t_{pHZ}$

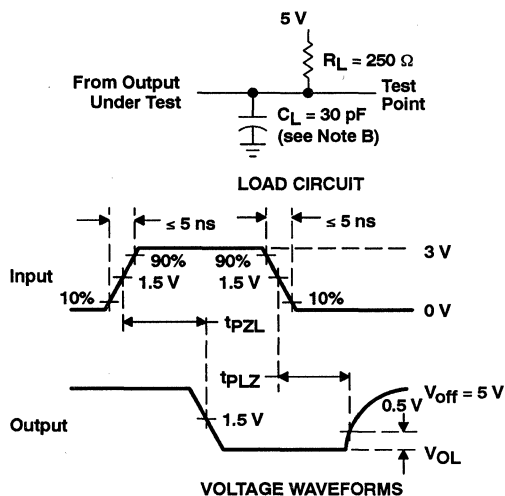


Figure 15.  $t_{pZL}$  and  $t_{pLZ}$

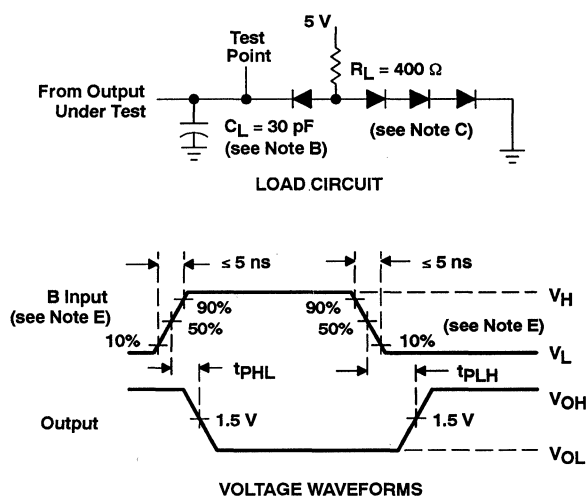


Figure 16.  $t_{pLH}$  and  $t_{pHL}$  (receivers only)

- NOTES: A. Input pulses are supplied by generators having the following characteristics  $Z_O = 50 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ ,  $t_w = 100 \text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.  
 D. When testing the '116 and SN75118 receiver sections, the response-time control and the termination resistor pins are left open.  
 E. For '116 and SN75118,  $V_H = 3 \text{ V}$ ,  $V_L = -3 \text{ V}$ , the A input is at 0 V.  
 For SN75117 and SN75119,  $V_H = 3 \text{ V}$ ,  $V_L = 0$ , the A input is at 1.5 V.

# SN55121, SN75121 DUAL LINE DRIVERS

SLLS074A – D1334, SEPTEMBER 1973 – REVISED FEBRUARY 1993

- Designed for Digital Data Transmission Over 50-Ω to 500-Ω Coaxial Cable, Strip Line, or Twisted Pair
- High Speed  
 $t_{pd} = 20 \text{ ns Max at } C_L = 15 \text{ pF}$
- TTL Compatible With Single 5-V Supply
- 2.4-V Output at  $I_{OH} = -75 \text{ mA}$
- Uncommitted Emitter-Follower Output Structure for Party-Line Operation
- Short-Circuit Protection
- AND-OR Logic Configuration
- Designed for Use With Triple Line Receivers SN55122, SN75122
- Designed to Be Interchangeable With Signetics N8T13

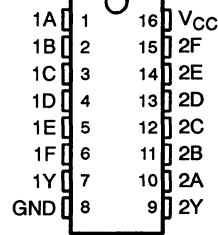
## description

The SN55121 and SN75121 dual line drivers are designed for digital data transmission over lines having impedances from 50 to 500 Ω. They are also compatible with standard TTL logic and supply voltage levels.

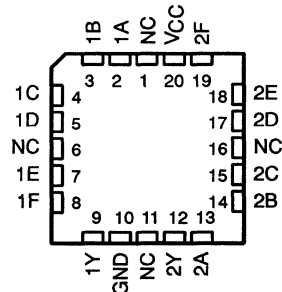
The low-impedance emitter-follower outputs of the SN55121 and SN75121 will drive terminated lines such as coaxial cable or twisted pairs. Having the outputs uncommitted allows wired-OR logic to be performed in party-line applications. Output short-circuit protection is provided by an internal clamping network that turns on when the output voltage drops below approximately 1.5 V. All of the inputs are in conventional TTL configuration and the gating can be used during power-up and power-down sequences to ensure that no noise is introduced to the line.

The SN55121 is characterized for operation over the full military temperature range of -55°C to 125°C. The SN75121 is characterized for operation from 0°C to 70°C.

SN55121 ... J PACKAGE  
SN75121 ... D OR N PACKAGE  
(TOP VIEW)



SN55121 ... FK PACKAGE  
(TOP VIEW)



NC—No internal connection

**THE SN75121 IS NOT  
RECOMMENDED FOR NEW DESIGN**

FUNCTION TABLE

INPUTS						OUTPUT
A	B	C	D	E	F	Y
H	H	H	H	X	X	H
X	X	X	X	H	H	H
All other input combinations						L

H = high level L = low level X = irrelevant

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

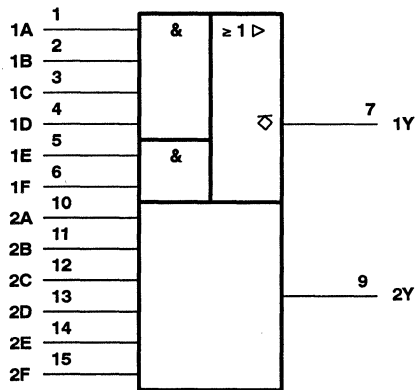
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# SN55121, SN75121 DUAL LINE DRIVERS

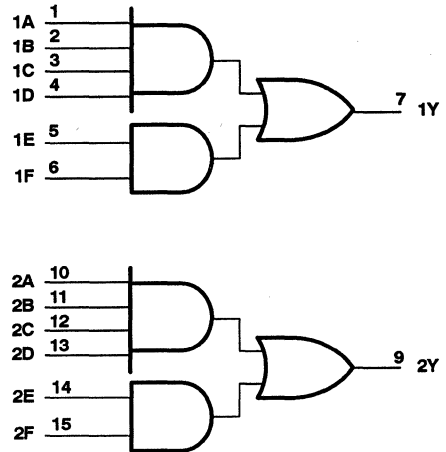
SLLS074A - D1334, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## logic symbol†

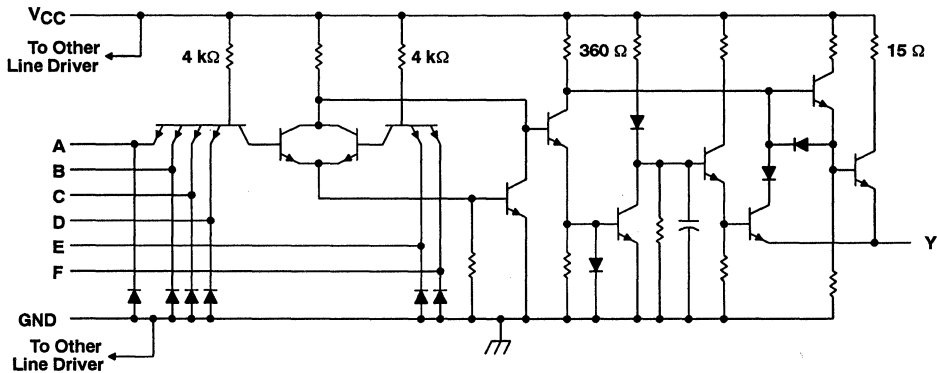


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12. Pin numbers shown are for the D, J, and N packages.

## logic diagram (positive logic)



## schematic (each driver)



All resistor values shown are nominal.

# SN55121, SN75121 DUAL LINE DRIVERS

SLLS074A – D1334, SEPTEMBER 1973 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55121	SN75121	UNIT
Supply voltage, $V_{CC}$ (see Note 1)	6	6	V
Input voltage	6	6	V
Output voltage	6	6	V
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds: FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package		260	°C

NOTE 1: All voltage values are with respect to both ground terminals connected together.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	—
FK†	1375 mW	11.0 mW/°C	880 mW	275 mW
J†	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—

† In the FK and J packages, SN55121 chips are either silver glass or alloy mounted.

## recommended operating conditions

	SN55121			SN75121			UNIT	
	MIN	NOM	MAX	MIN	NOM	MAX		
Supply voltage, $V_{CC}$	4.75	5	5.25	4.75	5	5.25	V	
High-level input voltage, $V_{IH}$	2			2			V	
Low-level input voltage, $V_{IL}$	0.8			0.8			V	
High-level output current, $I_{OH}$	-75			-75			mA	
Operating free-air temperature, $T_A$	-55			0			70	°C



# SN55121, SN75121 DUAL LINE DRIVERS

SLLS074A - D1334, SEPTEMBER 1973 - REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT	
$V_{IK}$	Input clamp voltage	$V_{CC} = 5\text{ V}$ ,	$I_I = -12\text{ mA}$		-1.5	V	
$V_{(BR)}$	Breakdown voltage	$V_{CC} = 5\text{ V}$ ,	$I_I = 10\text{ mA}$	5.5		V	
$V_{OH}$	High-level output voltage	$V_{IH} = 2\text{ V}$ ,	$I_{OH} = -75\text{ mA}$ , See Note 2	2.4		V	
$I_{OH}$	High-level output current	$V_{CC} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$ ,	$V_{IH} = 4.5\text{ V}$ , See Note 2	$V_{OH} = 2\text{ V}$ ,	-100	-250	mA
$I_{OL}$	Low-level output current	$V_{IL} = 0.8\text{ V}$ ,	$V_{OL} = 0.4\text{ V}$ ,	See Note 2	-800		$\mu\text{A}$
$I_{O(off)}$	Off-state output current	$V_{CC} = 3\text{ V}$ ,	$V_O = 3\text{ V}$		500		$\mu\text{A}$
$I_{IH}$	High-level output current	$V_I = 4.5\text{ V}$			40		$\mu\text{A}$
$I_{IL}$	Low-level output current	$V_I = 0.4\text{ V}$			-0.1	-1.6	mA
$I_{OS}$	Short-circuit output current†	$V_{CC} = 5\text{ V}$ ,	$T_A = 25^\circ\text{C}$		-30		mA
$I_{CCH}$	Supply current, outputs high	$V_{CC} = 5.25\text{ V}$ ,	All inputs at 2 V, Outputs open		28		mA
$I_{CCL}$	Supply current, outputs low	$V_{CC} = 5.25\text{ V}$ ,	All inputs at 0.8 V, Outputs open		60		mA

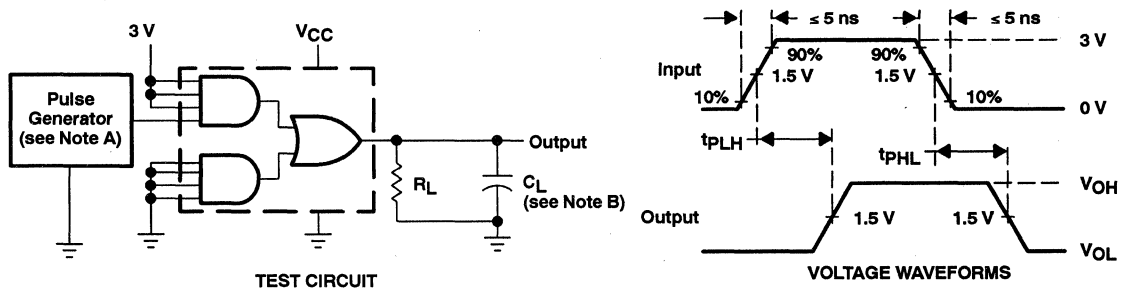
† Not more than one output should be shorted at a time.

NOTE 2: The output voltage and current limits are valid for any appropriate combination of high and low inputs specified by the function table for the desired output.

**switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high level output	$R_L = 37\ \Omega$ , $C_L = 15\text{ pF}$ ,	See Figure 1		11	20	ns
$t_{PHL}$	Propagation delay time, high-to-low level output				8	20	
$t_{PLH}$	Propagation delay time, low-to-high level output	$R_L = 37\ \Omega$ , $C_L = 1000\text{ pF}$ ,	See Figure 1		22	50	ns
$t_{PHL}$	Propagation delay time, high-to-low level output				20	50	

## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generators have the following characteristics:  $Z_O = 50\ \Omega$ ,  $t_W = 200\text{ ns}$ , duty cycle  $\leq 50\%$ , PRR  $\leq 500\text{ kHz}$ .

B.  $C_L$  includes probe and jig capacitance.

**Figure 1. Test Circuit and Voltage Waveforms**

TYPICAL CHARACTERISTICS

OUTPUT CURRENT vs OUTPUT VOLTAGE

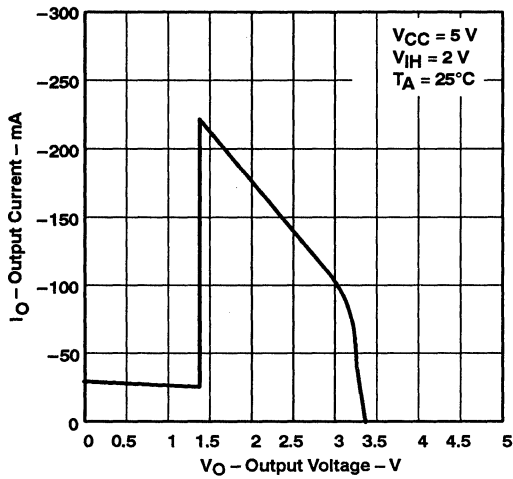


Figure 2

APPLICATION INFORMATION

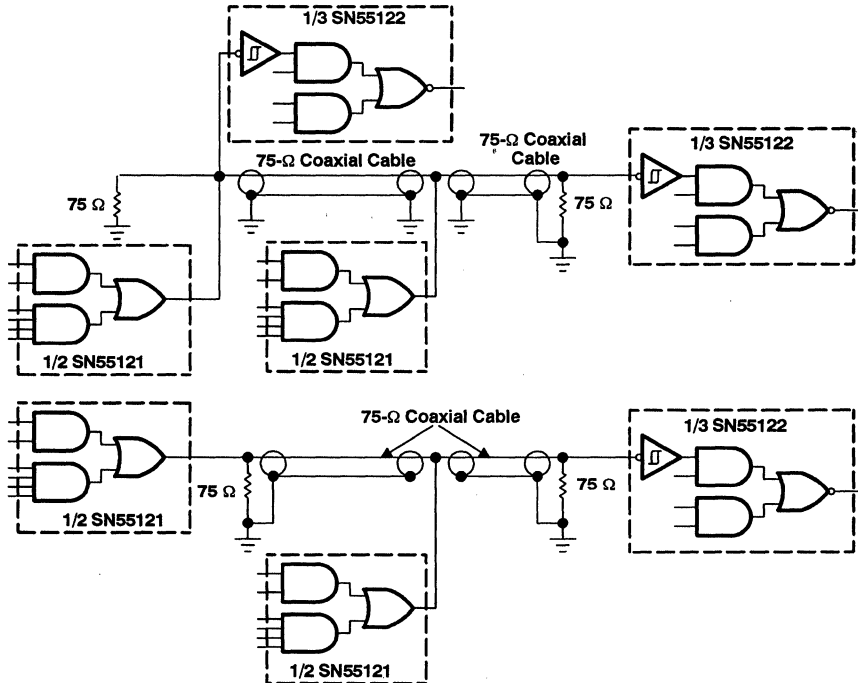


Figure 3. Single-Ended Party Line Circuits



# SN75ALS121 DUAL LINE DRIVER

SLLS030A - D1334, SEPTEMBER 1987 - REVISED AUGUST 1989

- Permits Digital Data Transmission Over Coaxial Cable, Strip Line, or Twisted Pair
- Operates With 50-Ω to 500-Ω Transmission Lines
- TTL Compatible With 5-V Supply
- 2.4-V Output at  $I_{OH} = -75 \text{ mA}$
- Uncommitted Emitter-Follower Output Structure for Party-Line Operation
- IMPACT™ Low-Power Schottky Technology
- Improved Replacement for the SN75121 and Signetics 8T13
- Glitchless Power Up/Power Down
- Short-Circuit Protection
- AND-OR Logic Configuration
- High Speed . . . Maximum Propagation Delay Time of 14 ns at  $C_L = 15 \text{ pF}$

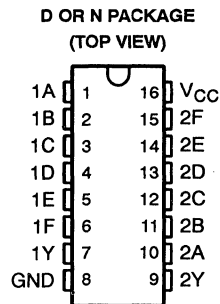
## description

The SN75ALS121 dual line driver is designed for digital data transmission over lines having impedances from 50 to 500 Ω. It is compatible with standard TTL logic and supply voltage levels.

The low-impedance emitter-follower outputs drive terminated lines such as coaxial cable, strip line, or twisted pair. Having the outputs uncommitted allows wired-OR logic to be performed in party-line applications. Output short-circuit protection is provided by an internal clamping network that turns on when the output voltage drops below approximately 1.5 V. All inputs are in conventional TTL configuration. Gating can be used during power-up and power-down sequences to ensure that no noise is introduced on the line.

The SN75ALS121 employs the IMPACT™ process to achieve fast switching speeds, low power dissipation, and reduced input current requirements.

The SN75ALS121 is characterized for operation from 0°C to 70°C.



NOT RECOMMENDED FOR NEW DESIGN

FUNCTION TABLE

INPUTS						OUTPUT
A	B	C	D	E	F	Y
H	H	H	H	X	X	H
X	X	X	X	H	H	H
All other input combinations						L

H = high level, L = low level, X = irrelevant

IMPACT is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

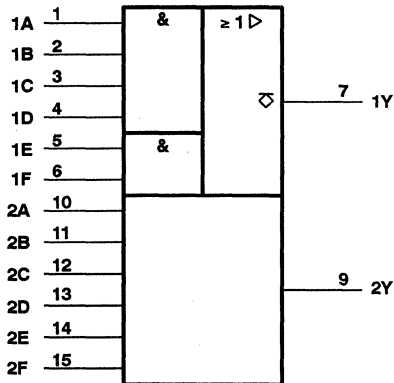
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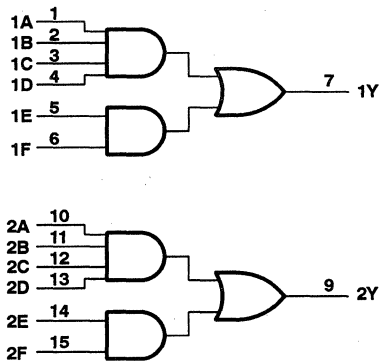
# SN75ALS121 DUAL LINE DRIVER

SLLS030A - D1334, SEPTEMBER 1987 - REVISED AUGUST 1989

## logic symbol†

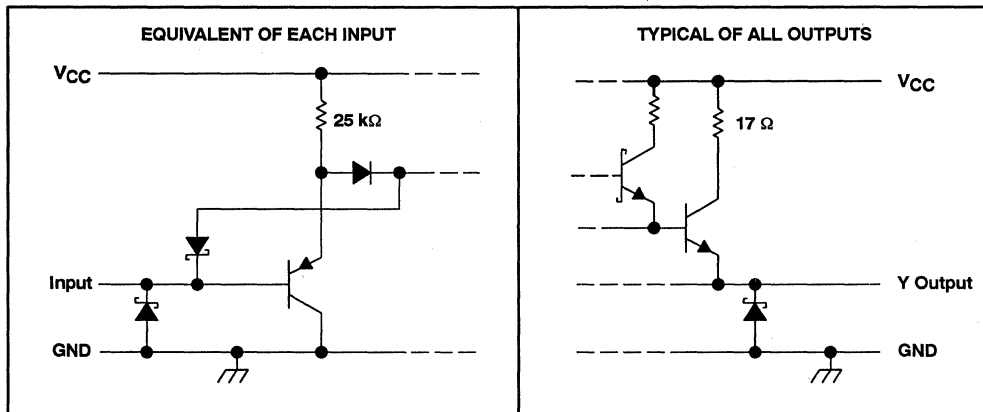


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematics of inputs and outputs



# SN75ALS121 DUAL LINE DRIVER

SLLS030A – D1334, SEPTEMBER 1987 – REVISED AUGUST 1989

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	6 V
Input voltage	6 V
Output voltage	6 V
Continuous total dissipation at (or below) 25°C free air temperature	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	– 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			– 75	mA
Operating free-air temperature range, $T_A$	0		70	°C

## electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$V_{CC} = 5\text{ V}$ , $I_I = -12\text{ mA}$			– 1.5	V
$V_{(BR)}$ Input breakdown voltage	$V_{CC} = 5\text{ V}$ , $I_I = 10\text{ mA}$	5.5			V
$V_{OH}$ High-level output voltage	$V_{IH} = 2\text{ V}$ , $I_{OH} = -75\text{ mA}$ , See Note 2	2.4	3.2		V
$I_{OH}$ High-level output current	$V_{CC} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$ , $V_{IH} = 4.5\text{ V}$ , See Note 2	– 100	– 200	– 250	mA
$I_{OL}$ Low-level output current	$V_{IL} = 0.8\text{ V}$ , $V_{OL} = 0.4\text{ V}$ , See Note 2			– 800	μA
$I_{O(off)}$ Off-state output current	$V_{CC} = 3\text{ V}$ , $V_O = 3\text{ V}$			500	μA
$I_{IH}$ High-level input current	$V_I = 4.5\text{ V}$			40	μA
$I_{IL}$ Low-level input current	$V_I = 0.4\text{ V}$			– 250	μA
$I_{OS}$ Short-circuit output current	$V_{CC} = 5\text{ V}$	– 5	– 30		mA
$I_{CCH}$ Supply current, outputs high	$V_{CC} = 5.25\text{ V}$ , All inputs at 2 V, No load		9	14	mA
$I_{CCL}$ Supply current, outputs low	$V_{CC} = 5.25\text{ V}$ , All inputs at 0.8 V, No load		13	30	mA

† All typical values are at  $V_{CC} = 5\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

NOTE 2: The output voltage and current limits are ensured for any appropriate combination of high and low inputs specified by the function table for the desired output.



# SN75ALS121 DUAL LINE DRIVER

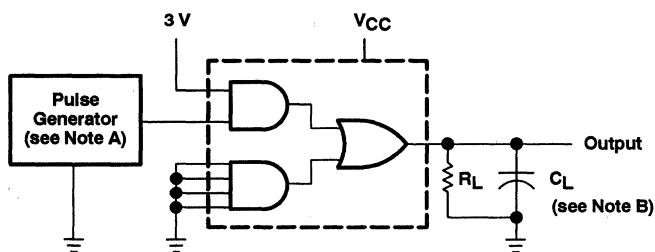
SLLS030A - D1334, SEPTEMBER 1987 - REVISED AUGUST 1989

switching characteristics over recommended ranges of supply voltage and operating free-air temperature

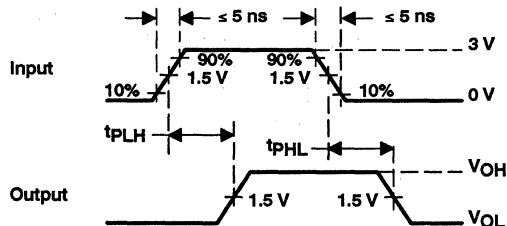
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 37 \Omega$ , $C_L = 15 \text{ pF}$ , See Figure 1		6	14	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			4	14	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 37 \Omega$ , $C_L = 1000 \text{ pF}$ , See Figure 1		18	30	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			29	50	ns

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $t_w = 200 \text{ ns}$ , duty cycle = 50%.  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

OUTPUT CURRENT  
vs  
OUTPUT VOLTAGE

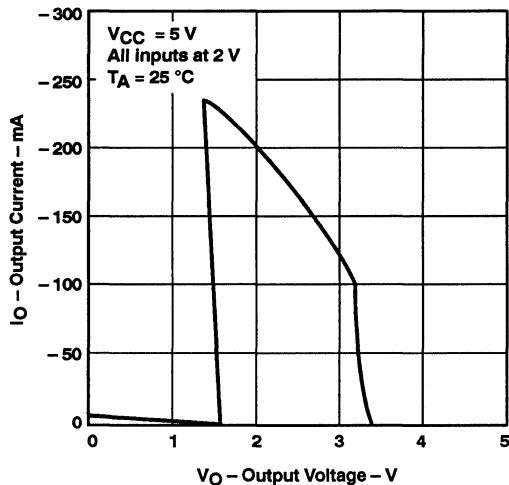


Figure 2





# N8T14, SN55122, SN75122 TRIPLE LINE RECEIVERS

SLLS075A - D1334, SEPTEMBER 1973 - REVISED FEBRUARY 1993

- Designed for Digital Data Transmission Over Coaxial Cable, Strip Line, or Twisted Pair
- Designed for Operation With 50-Ω to 500-Ω Transmission Lines
- TTL Compatible
- Single 5-V Supply
- Built-Input Threshold Hysteresis
- High-Speed . . . Typical Propagation Delay Time = 20 ns
- Independent Channel Strobes
- Input Gating Increases Application Flexibility
- Fanout to 10 Series 54/74 Standard Loads
- Can Be Used With Dual Line Drivers SN55121 and SN75121
- Interchangeable With Signetics N8T14

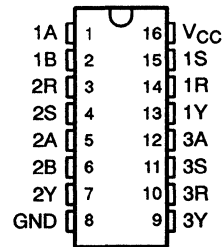
## description

The N8T14, SN55122, and SN75122 are triple line receivers that are designed for digital data transmission over lines having impedances from 50 Ω to 500 Ω. They are also compatible with standard TTL-logic and supply voltage levels.

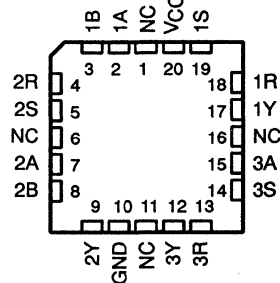
The N8T14, SN55122, and SN75122 have receiver inputs with built-in hysteresis to provide increased noise margin for single-ended systems. The high impedance of this input presents a minimum load to the driver and allows termination of the transmission line in its characteristic impedance to minimize line reflection. An open line will affect the receiver input as would a low-level voltage. The receiver can withstand a level of -0.15 V with power on or off. The other inputs are in TTL configuration. The S input must be high to enable the receiver input. Two of the line receivers have A and B inputs that, if both are high, will hold the output low. The third receiver has only an A input that, if high, will hold the output low.

The SN55122 is characterized for operation over the full military temperature range of -55°C to 125°C. The N8T14 and SN75122 are characterized for operation from 0°C to 70°C.

SN55122 . . . J PACKAGE  
N8T14, SN75122 . . . D OR N PACKAGE  
(TOP VIEW)



SN55122 . . . FK PACKAGE  
(TOP VIEW)



NC—No internal connection

**THE N8T14 AND SN75122 ARE NOT  
RECOMMENDED FOR NEW DESIGN**

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

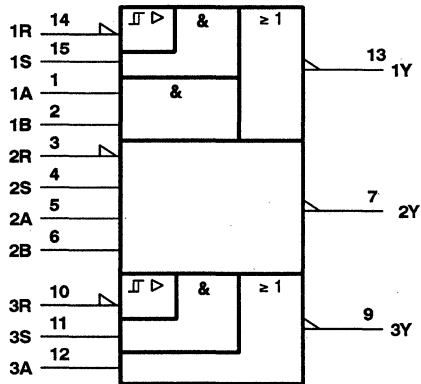
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# N8T14, SN55122, SN75122 TRIPLE LINE RECEIVERS

SLLS075A-D1334, SEPTEMBER 1973 - REVISED FEBRUARY 1993

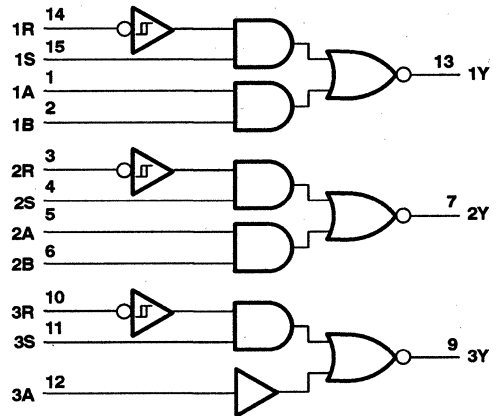
## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, and N packages.

## logic diagram



FUNCTION TABLE

INPUTS				OUTPUT
A	B‡	R	S	Y
H	H	X	X	L
X	X	L	H	L
L	X	H	X	H
L	X	X	L	H
X	L	H	X	H
X	L	X	L	H

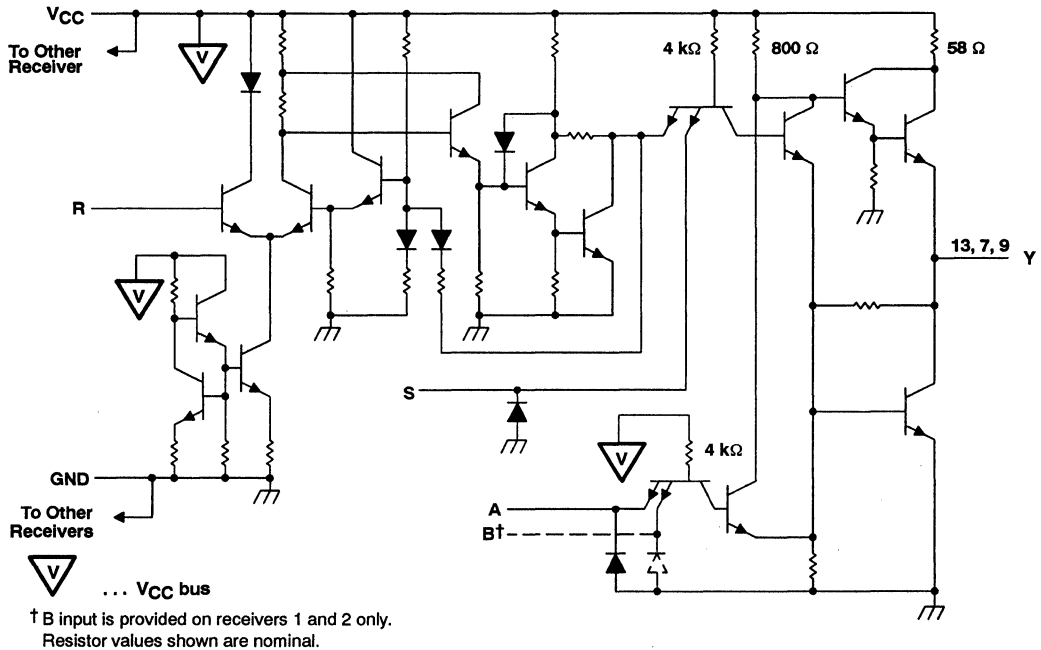
‡ B input and last two lines of the function table are applicable to receivers 1 and 2 only.

H = high level, L = low level, X = irrelevant

# N8T14, SN55122, SN75122 TRIPLE LINE RECEIVERS

SLLS075A - D1334, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## schematic diagram (each receiver)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	6 V
Input voltage: R input	6 V
A, B, or S input	5.5 V
Output voltage	6 V
Output current	$\pm 100$ mA
Continuous total power dissipation (see Note 2)	See Dissipation Rating Table
Operating free-air temperature range: SN55122	$-55^{\circ}\text{C}$ to $125^{\circ}\text{C}$
N8T14, SN75122	$0^{\circ}\text{C}$ to $70^{\circ}\text{C}$
Storage temperature range	$-65^{\circ}\text{C}$ to $150^{\circ}\text{C}$
Case temperature for 60 seconds: FK package	$260^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	$300^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	$260^{\circ}\text{C}$

- NOTES: 1. Voltage values are with respect to network ground terminal.  
2. The SN55122 chips are alloy mounted, and the SN75122 chips are glass mounted.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	$T_A = 70^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING		POWER RATING	POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW	—
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
J	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW	—

TEXAS  
INSTRUMENTS

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# N8T14, SN55122, SN75122 TRIPLE LINE RECEIVERS

SLLS075A - D1334, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	A, B, R, or S	2			V
Low-level input voltage, $V_{IL}$	A, B, R, or S			0.8	V
High-level output current, $I_{OH}$				-500	$\mu$ A
Low-level output current, $I_{OL}$				16	mA
Operating free-air temperature, $T_A$	SN55122	-55		125	$^{\circ}$ C
	SN75122	0		70	

## electrical characteristics over recommended operating free-air temperature, $V_{CC} = 4.75$ V to 5.25 V (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	R	$V_{CC} = 5$ V, $T_A = 25^{\circ}$ C, See Figures 2 and 4	0.3	0.6		V
$V_{IK}$	Input clamp voltage	A, B, or S	$V_{CC} = 5$ V, $I_I = -12$ mA			-1.5	V
$V_{I(BR)}$	Input breakdown voltage	A, B, or S	$V_{CC} = 5$ V, $I_I = 10$ mA	5.5			V
$V_{OH}$	High-level output voltage		$V_{IH} = 2$ V, $V_{IL} = 0.8$ V, $I_{OH} = -500$ $\mu$ A	2.6			V
			$V_I(A) = 0$ , $V_I(B) = 0$ , $V_I(S) = 2$ V, $V_I(R) = 1.45$ V, $I_{OH} = -500$ $\mu$ A, See Note 3	2.6			
$V_{OL}$	Low-level output voltage		$V_{IH} = 2$ V, $V_{IL} = 0.8$ V, $I_{OL} = 16$ mA			0.4	V
			$V_I(A) = 0$ , $V_I(B) = 0$ , $V_I(S) = 2$ V, $V_I(R) = 1.45$ V, $I_{OL} = 16$ mA, See Note 4			0.4	
$I_{IH}$	High-level input current	A, B, or S	$V_I = 4.5$ V			40	$\mu$ A
		R	$V_I = 3.8$ V			170	
$I_{IL}$	Low-level input current	A, B, or S	$V_I = 0.4$ V, $V_{IR} = 0.8$ V	-0.1		-1.6	mA
$I_{OS}^{\ddagger}$	Short-circuit output current		$V_{CC} = 5$ V, $T_A = 25^{\circ}$ C	-50		-100	mA
$I_{CCH}$	High-level supply current		$V_{CC} = \text{MAX}$ , All inputs at 0.8 V, Outputs open			72	mA
$I_{CCL}$	Low-level supply current		$V_{CC} = \text{MAX}$ , All inputs at 2 V, Outputs open			100	mA

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^{\circ}$ C.

‡ Not more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

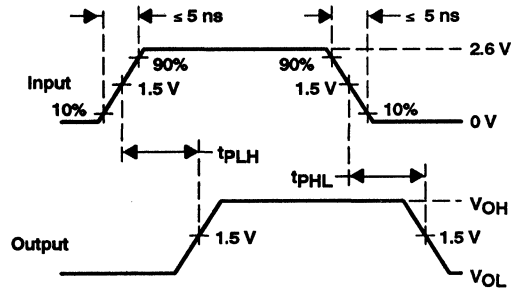
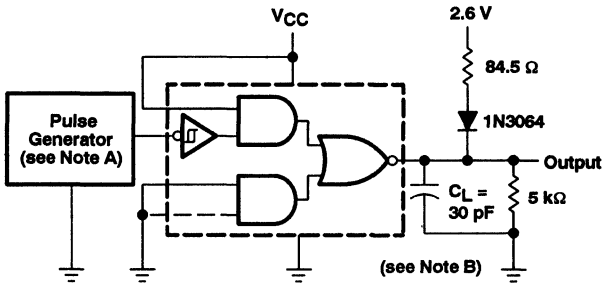
NOTES: 3. The receiver input is high immediately before being reduced to 1.45 V.

4. The receiver input is low immediately before being increased to 1.45 V.

## switching characteristics, $V_{CC} = 5$ V, $T_A = 25^{\circ}$ C

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output from R input	See Figure 1		20	30	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output from R input			20	30	

PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT

VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $t_w = 200 \text{ ns}$ , duty cycle = 50%,  $\text{PRR} \leq 500 \text{ kHz}$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

OUTPUT VOLTAGE  
vs  
INPUT VOLTAGE

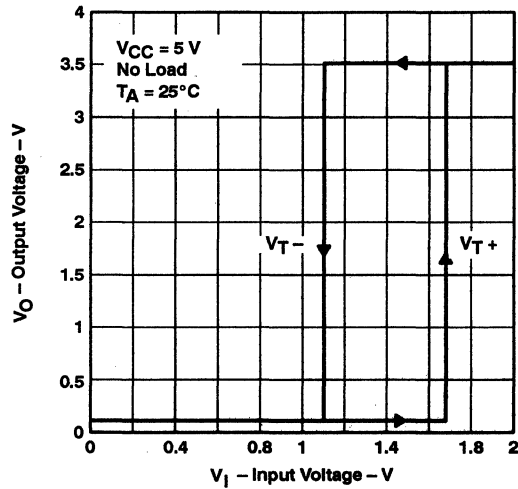


Figure 2

# N8T14, SN55122, SN75122 TRIPLE LINE RECEIVERS

SLLS075A - D1334, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## APPLICATION INFORMATION

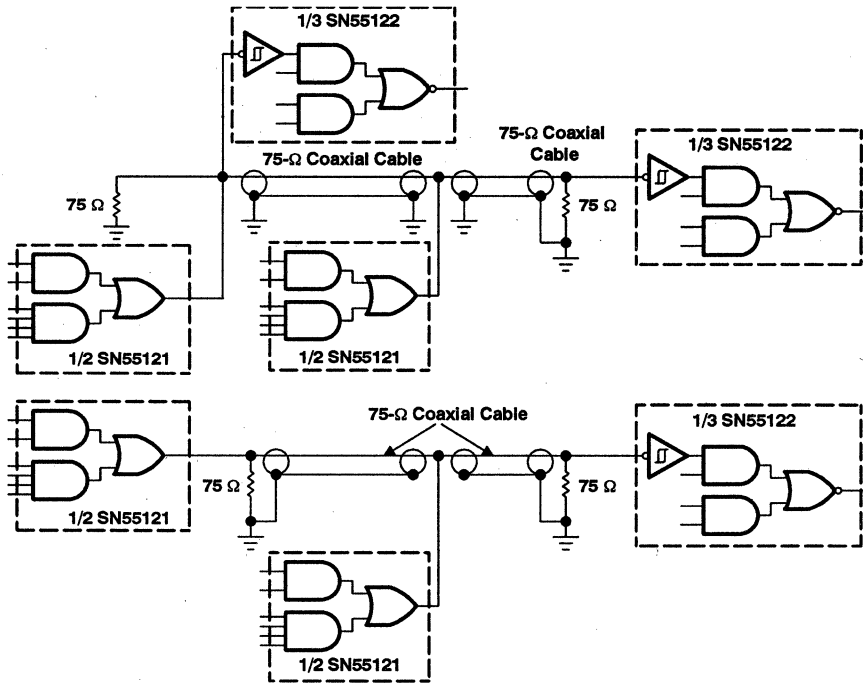
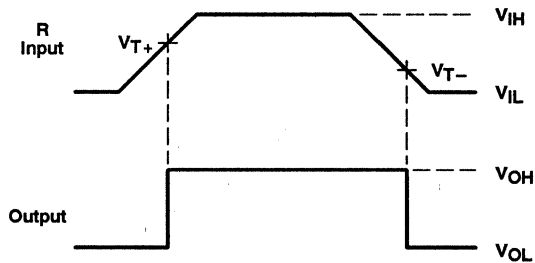


Figure 3. Single-Ended Party Line Circuits



NOTE: The high gain and built-in hysteresis of the SN55122 and SN75122 line receivers enable them to be used as Schmitt triggers in squaring pulses.

Figure 4. Pulse Squaring

# N8T13, N8T23, SN75123 DUAL LINE DRIVERS

SLLS086A - D1322, SEPTEMBER 1973 - REVISED JANUARY 1993

- Meets IBM System 360 input/Output interface Specifications
- Operates From Single 5-V Supply
- TTL Compatible
- 3.11-V Output at  $I_{OH} = -59.3$  mA
- Uncommitted Emitter-Follower Output Structure for Party-Line Operation
- Short-Circuit Protection
- AND-OR Logic Configuration
- Designed for Use With Triple Line Receiver SN75124
- Designed to Be Interchangeable With Signetics N8T13 and N8T23

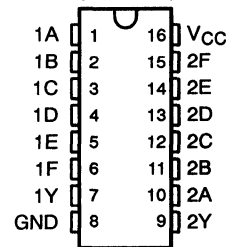
## description

The N8T13, N8T23, and SN75123 are dual line drivers specifically designed to meet the input/output interface specifications for IBM System 360. It is also compatible with standard TTL logic and supply voltage levels.

The low-impedance emitter-follower outputs of the N8T13, N8T23, and SN75123 will drive terminated lines such as coaxial cable or twisted pair. Having the outputs uncommitted allows wired-OR logic to be performed in party-line applications. Output short-circuit protection is provided by an internal clamping network that turns on when the output voltage drops below approximately 1.5 V. All the inputs are in conventional TTL configuration and the gating can be used during power-up and power-down sequences to ensure that no noise is introduced to the line.

The N8T13, N8T23, and SN75123 are characterized for operation from 0°C to 70°C.

D OR N PACKAGE  
(TOP VIEW)

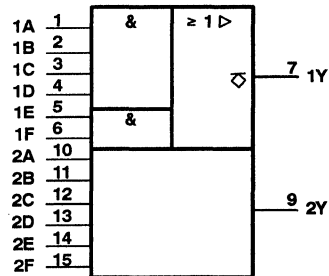


FUNCTION TABLE

INPUTS						OUTPUT
A	B	C	D	E	F	Y
H	H	H	H	X	X	H
X	X	X	X	H	H	L
All other input combinations						L

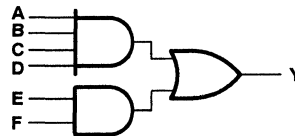
H = high level, L = low level, X = irrelevant

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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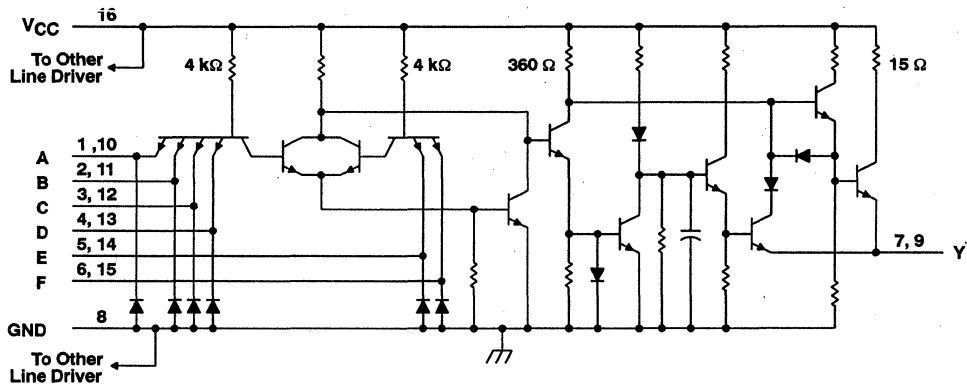
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# N8T13, N8T23, SN75123 DUAL LINE DRIVERS

SLLS086A – D1322, SEPTEMBER 1973 – REVISED JANUARY 1993

## schematic (each driver)



Resistor values shown are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Output voltage, $V_O$	7 V
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 2):	
D package	950 mW
N package	1150 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.

2. For operation above 25°C free-air temperature, derate the D package to 608 mW at 70°C at the rate of 7.6 mW/°C and the N package to 736 mW at 70°C at the rate of 9.2 mW/°C.

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-100	mA
Operating free-air temperature, $T_A$	0		70	°C

TEXAS  
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# N8T13, N8T23, SN75123 DUAL LINE DRIVERS

SLLS086A - D1322, SEPTEMBER 1973 - REVISED JANUARY 1993

## electrical characteristics, $V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$ , $T_A = 0^\circ\text{C to } 70^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS		MIN	MAX	UNIT	
$V_{IK}$ Input clamp voltage	$V_{CC} = 5 \text{ V}$ ,	$I_I = -12 \text{ mA}$		-1.5	V	
$V_{(BR)I}$ Input breakdown voltage	$V_{CC} = 5 \text{ V}$ ,	$I_I = 10 \text{ mA}$	5.5		V	
$V_{OH}$ High-level output voltage	$V_{CC} = 5 \text{ V}$ , $I_{OH} = -59.3 \text{ mA}$ ,	$V_{IH} = 2 \text{ V}$ , See Note 3	$T_A = 25^\circ\text{C}$ 3.11		V	
$V_{OL}$ Low-level output voltage	$V_{IL} = 0.8 \text{ V}$ ,	$I_{OL} = -240 \mu\text{A}$ ,	$T_A = 0^\circ\text{C to } 70^\circ\text{C}$ 2.9		V	
$I_{OH}$ High-level output current	$V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$ ,	$V_{IH} = 4.5 \text{ V}$ , See Note 3	$V_{OH} = 2 \text{ V}$ ,	-100	-250	mA
$I_{O(off)}$ Off-state output current	$V_{CC} = 0$ ,	$V_O = 3 \text{ V}$		40	$\mu\text{A}$	
$I_{IH}$ High-level input current	$V_I = 4.5 \text{ V}$			40	$\mu\text{A}$	
$I_{IL}$ Low-level input current	$V_I = 0.4 \text{ V}$			-0.1	-1.6	mA
$I_{OS}$ Short-circuit output current†	$V_{CC} = 5 \text{ V}$ ,	$T_A = 25^\circ\text{C}$		-30	mA	
$I_{CCH}$ Supply current, outputs high	$V_{CC} = 5.25 \text{ V}$ ,	All inputs at 2 V, Outputs open		28	mA	
$I_{CCL}$ Supply current, outputs low	$V_{CC} = 5.25 \text{ V}$ ,	All inputs at 0.8 V, Outputs open		60	mA	

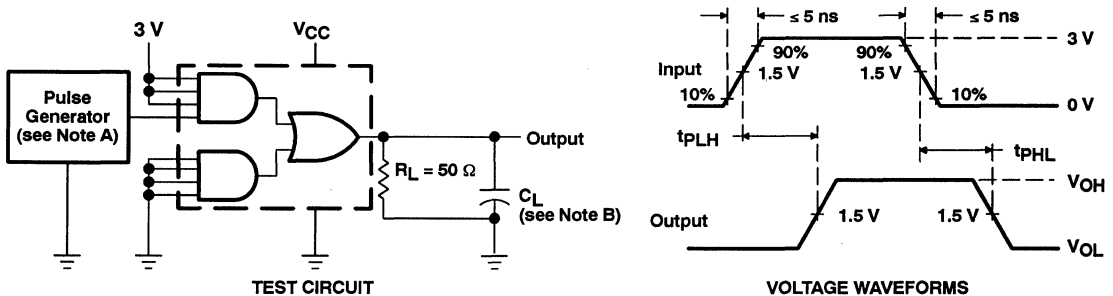
† Not more than one output should be shorted at a time.

NOTE 3: The output voltage and current limits are valid for any appropriate combination of high and low inputs specified by the function table for the desired output.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 50 \Omega$ ,	$C_L = 15 \text{ pF}$ , See Figure 1		12	20	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output				12	20	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 50 \Omega$ ,	$C_L = 100 \text{ pF}$ , See Figure 1		20	35	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output				15	25	ns

## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ;  $t_w = 200 \text{ ns}$ , duty cycle = 50%.

B.  $C_L$  Includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

TEXAS  
INSTRUMENTS

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2-261

# N8T13, N8T23, SN75123 DUAL LINE DRIVERS

SLLS086A - D1322, SEPTEMBER 1973 - REVISED JANUARY 1993

## TYPICAL CHARACTERISTICS

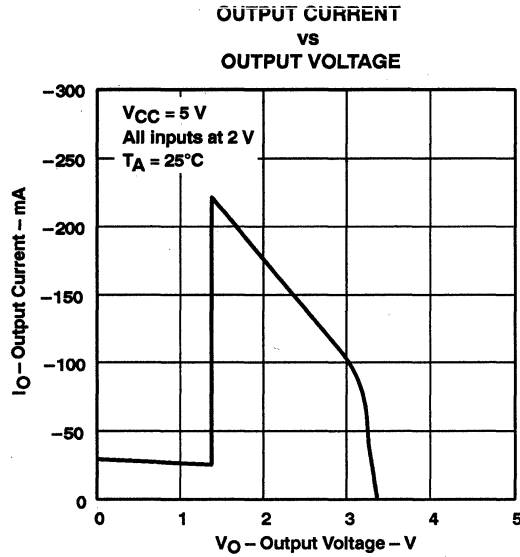


Figure 2

## APPLICATION INFORMATION

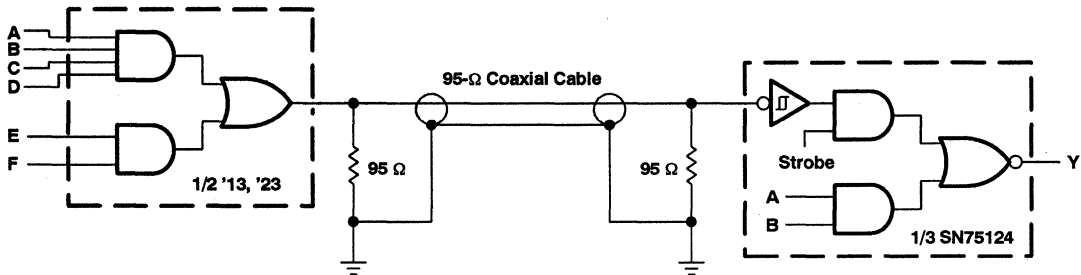


Figure 3. Unbalanced Line Communication Using '13, '23, and '124

  
**TEXAS  
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# SN75ALS123 DUAL LINE DRIVER

SLLS031B - D1332, SEPTEMBER 1987 - REVISED FEBRUARY 1993

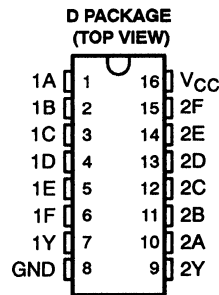
- Meets IBM 360 Input Interface Specifications
- Permits Digital Data Transmission Over Coaxial Cable, Strip Line, or Twisted Pair
- TTL Compatible With 5-V Supply
- 3.11-V Output at  $I_{OH} = -59.3$  mA
- Uncommitted Emitter-Follower Output Structure for Party-Line Operation
- IMPACT™ Low-Power Schottky Technology
- Improved Replacement for the SN75123 and Signetics 8T13
- Glitchless Power-Up/Power-Down Protection
- Short-Circuit Protection
- AND-OR Logic Configuration
- High Speed . . . Maximum Propagation Delay Time of 14 ns at  $C_L = 15$  pF

## description

The SN75ALS123 dual line driver is specifically designed to meet the input interface specifications for the IBM System 360. It is compatible with standard TTL logic and supply voltage levels. The low-impedance, emitter-follower outputs drive terminated lines such as coaxial cable, strip line, or twisted pair. The uncommitted output allows wired-OR logic to be performed in party-line applications. Output short-circuit protection is provided by an internal clamping network that turns on when the output voltage drops below approximately 1.5 V. All inputs are in conventional TTL configuration. Gating can be used during power-up and power-down sequences to ensure that no noise is introduced on the line.

The SN75ALS123 employs the IMPACT™ process to achieve fast switching speeds, low power dissipation, and reduced input current requirements.

The SN75ALS123 is characterized for operation from 0°C to 70°C.



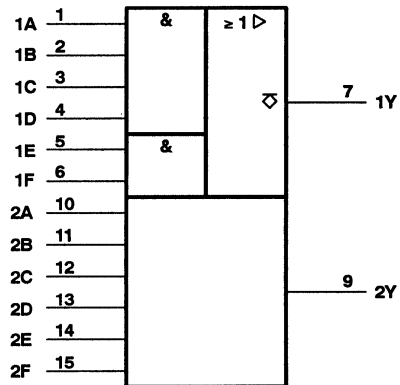
NOT RECOMMENDED FOR NEW DESIGN

FUNCTION TABLE

INPUTS						OUTPUT
A	B	C	D	E	F	Y
H	H	H	H	X	X	H
X	X	X	X	H	H	H
All other input combinations						L

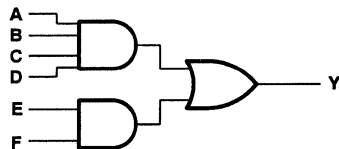
H = high level L = low level X = irrelevant

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram, each driver (positive logic)



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**TEXAS  
INSTRUMENTS**

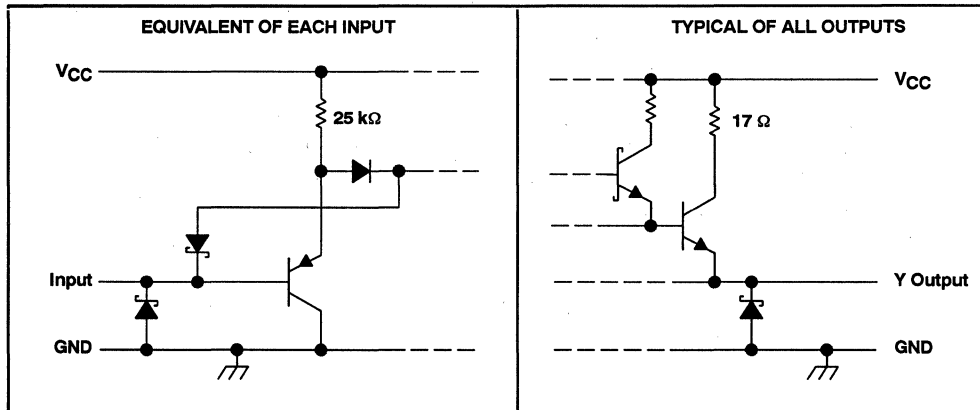
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# SN75ALS123 DUAL LINE DRIVER

SLLS031B – D1332, SEPTEMBER 1987 – REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Output voltage	6 V
Continuous total dissipation at (or below) 25°C free air temperature (see Note 2)	950 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

2. For operation above 25°C free-air temperature, derate to 608 mW at 70°C at the rate of 7.6 mW/°C.

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-100	mA
Operating free-air temperature range, $T_A$	0		70	°C

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$V_{CC} = 5\text{ V}$ , $I_I = -12\text{ mA}$			-1.5	V
$V_{(BR)}$ Input breakdown voltage	$V_{CC} = 5\text{ V}$ , $I_I = 10\text{ mA}$	5.5			V
$V_{OH}$ High-level output voltage	$V_{CC} = 5\text{ V}$ , See Note 3 $V_{IH} = 2\text{ V}$ , $I_{OH} = -59.3\text{ mA}$ ,	2.9			V
	$V_{CC} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$ , $V_{IH} = 2\text{ V}$ , See Note 3 $I_{OH} = -59.3\text{ mA}$ ,	3.11	3.3		
$V_{OL}$ Low-level output voltage	$V_{IL} = 0.8\text{ V}$ , $I_{OL} = -240\text{ }\mu\text{A}$ , See Note 3			0.15	V
$I_{OH}$ High-level output current	$V_{CC} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$ , $V_{IH} = 4.5\text{ V}$ , See Note 3 $V_{OH} = 2\text{ V}$ ,	-100	-200	-250	mA
$I_{O(off)}$ Off-state output current	$V_{CC} = 0$ , $V_O = 3\text{ V}$			40	$\mu\text{A}$
$I_{IH}$ High-level input current	$V_I = 4.5\text{ V}$			40	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_I = 0.4\text{ V}$			-250	$\mu\text{A}$
$I_{OS}$ Short-circuit output current	$V_{CC} = 5\text{ V}$			-5 -30	mA
$I_{CCH}$ Supply current, outputs high	$V_{CC} = 5.25\text{ V}$ , All inputs at 2 V, No load		9	14	mA
$I_{CCL}$ Supply current, outputs low	$V_{CC} = 5.25\text{ V}$ , All inputs at 0.8 V, No load		13	30	mA

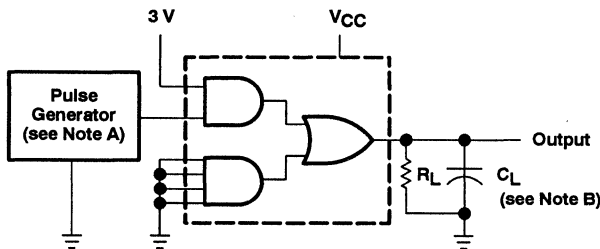
NOTE 3: The output voltage and current limits are ensured for any appropriate combination of high and low inputs specified by the function table for the desired output.

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

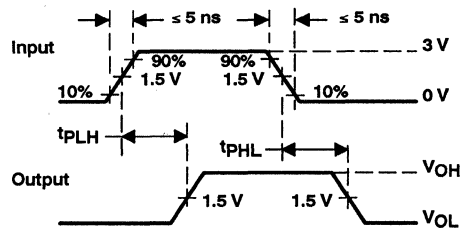
PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 50\text{ }\Omega$ , $C_L = 15\text{ pF}$ , See Figure 1		4	14	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			5	14	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 50\text{ }\Omega$ , $C_L = 100\text{ pF}$ , See Figure 1		8	20	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			8	20	ns

† All typical values are at  $V_{CC} = 5\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

### PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

**Figure 1. Test Circuit and Voltage Waveforms**

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50\text{ }\Omega$ ,  $t_w = 200\text{ ns}$ , duty cycle = 50%.  
B.  $C_L$  includes probe and jig capacitance.

# SN75ALS123 DUAL LINE DRIVER

SLLS031B - D1332, SEPTEMBER 1987 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

OUTPUT CURRENT  
vs  
OUTPUT VOLTAGE

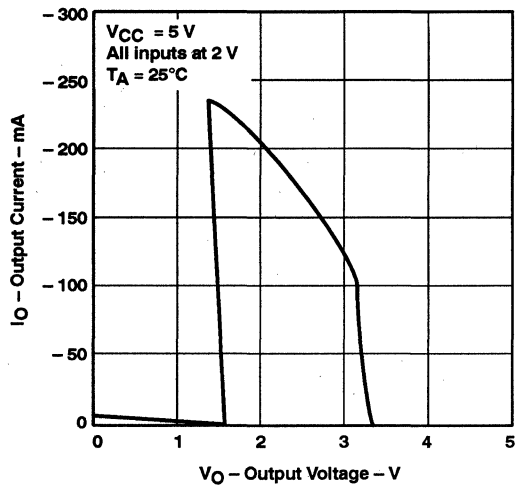


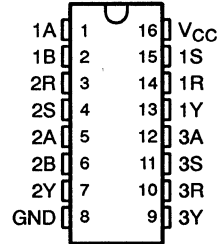
Figure 2

# SN75124 TRIPLE LINE RECEIVER

SLLS058A – D1322, SEPTEMBER 1973 – REVISED JANUARY 1993

- Meets IBM System 360 Input/Output Interface Specifications
- Operates From Single 5-V Supply
- TTL Compatible
- Built-in Input Threshold Hysteresis
- High Speed . . . Typical Propagation Delay Time = 20 ns
- Independent Channel Strobes
- Input Gating Increases Application Flexibility
- Designed for Use With Dual Line Driver SN75123
- Designed to Be Interchangeable With Signetics N8T24

D OR N PACKAGE  
(TOP VIEW)



## description

The SN75124 triple line receiver is specifically designed to meet the input/output interface specifications for IBM System 360. It is also compatible with standard TTL logic and supply voltage levels.

The SN75124 has receiver inputs with built-in hysteresis to provide increased noise margin for single-ended systems. An open line will affect the receiver input as would a low-level input voltage and the receiver input can withstand a level of  $-0.15$  V with power on or off. The other inputs are in TTL configuration. The S input must be high to enable the receiver input. Two of the line receivers have A and B inputs that, if both are high, will hold the output low. The third receiver has only an A input that, if high, will hold the output low.

The SN75124 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

FUNCTION TABLE

INPUTS				OUTPUT
A	B†	R	S	Y
H	H	X	X	L
X	X	L	H	L
L	X	H	X	H
L	X	X	L	H
X	L	H	X	H
X	L	X	L	H

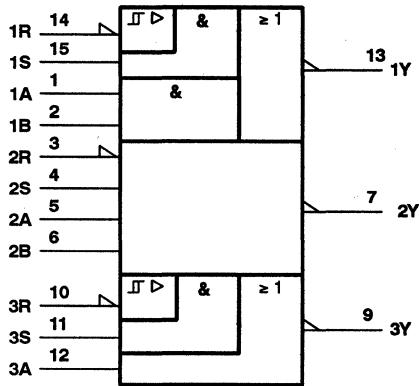
† B input and last two lines of the function table are applicable to receivers 1 and 2 only.



# SN75124 TRIPLE LINE RECEIVER

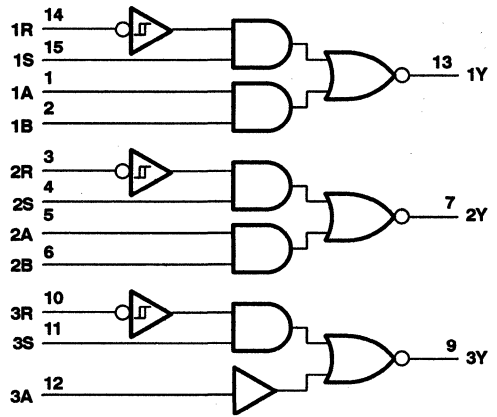
SLLS058A - D1322, SEPTEMBER 1973 - REVISED JANUARY 1993

## logic symbol†

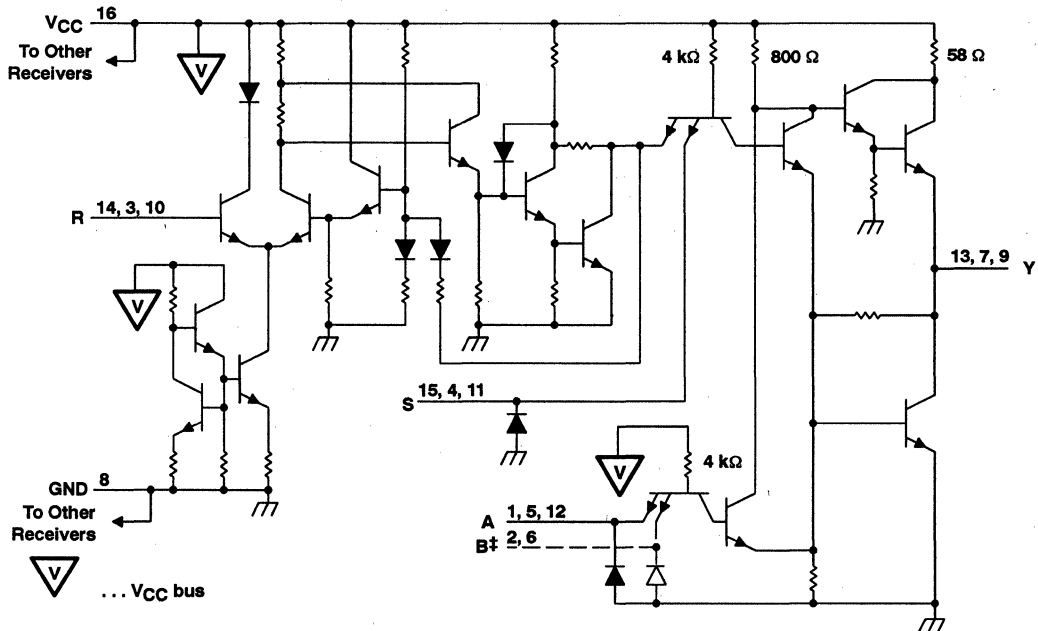


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram



## schematic (each receiver)



‡ B input is provided on receivers 1 and 2 only.  
Resistor values shown are nominal.

# SN75124 TRIPLE LINE RECEIVER

SLLS058A - D1322, SEPTEMBER 1973 - REVISED JANUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$ : R input with $V_{CC}$ applied	7 V
R input with $V_{CC}$ not applied	6 V
A, B, or S input	5.5 V
Output voltage, $V_O$	7 V
Output current, $I_O$	±100 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	A, B, or S	2			V
	R	1.7			
Low-level input voltage, $V_{IL}$	A, B, or S	0.8			V
	R	0.7			
High-level output current, $I_{OH}$		-800			μA
Low-level output current, $I_{OL}$		16			mA
Operating free-air temperature, $T_A$		0		70	°C



# SN75124 TRIPLE LINE RECEIVER

SLLS058A – D1322, SEPTEMBER 1973 – REVISED JANUARY 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	R	$V_{CC} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$	0.2	0.5		V
$V_{IK}$	Input clamp voltage	A, B, or S	$V_{CC} = 5\text{ V}$ , $I_I = 12\text{ mA}$			-1.5	V
$V(BR)_I$	Input breakdown voltage	A, B, or S	$V_{CC} = 5\text{ V}$ , $I_I = 10\text{ mA}$	5.5			V
$V_{OH}$	High-level output voltage		$V_{IH} = V_{IHmin}$ , $I_{OH} = -800\ \mu\text{A}$ , $V_{IL} = V_{ILmax}$ , See Note 2	2.6			V
$V_{OL}$	Low-level output voltage		$V_{IH} = V_{IHmin}$ , $I_{OL} = 16\text{ mA}$ , $V_{IL} = V_{ILmax}$ , See Note 2			0.4	V
$I_I$	Input current at maximum input voltage	R	$V_I = 7\text{ V}$			5	mA
			$V_I = 6\text{ V}$ , $V_{CC} = 0$			5	
$I_{IH}$	High-level input current	A, B, or S	$V_I = 4.5\text{ V}$			40	$\mu\text{A}$
		R	$V_I = 3.11\text{ V}$			170	
$I_{IL}$	Low-level input current	A, B, or S	$V_I = 0.4\text{ V}$ , $V_{IR} = 0.8\text{ V}$	-0.1		-1.6	mA
$I_{OS}$	Short-circuit output current†			-50		-100	mA
$I_{CC}$	Supply current		All inputs = 0.8 V			72	mA
			All inputs = 2 V			100	

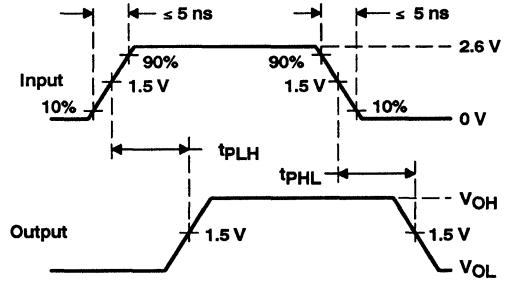
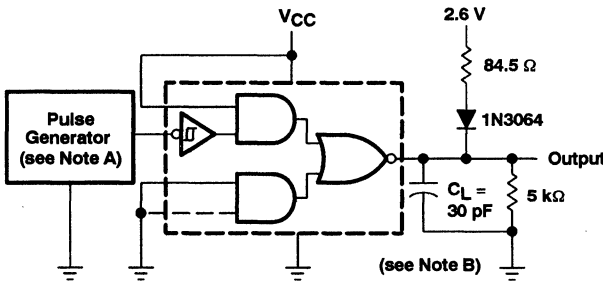
† Not more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

NOTE 2: The output voltage and current limits are characterized for any appropriate combination of high and low inputs specified by the function table for the desired output.

**switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output from R input		See Figure 1		20	30	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output from R input				20	30	

PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT

VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 5 \text{ MHz}$ , duty cycle = 50%.  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

OUTPUT VOLTAGE  
vs  
RECEIVER INPUT VOLTAGE

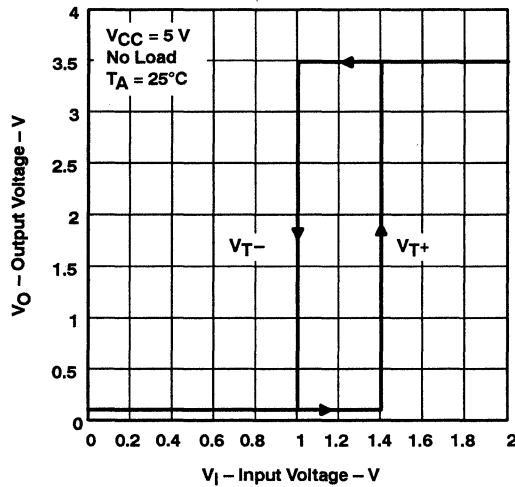


Figure 2

# SN75124 TRIPLE LINE RECEIVER

SLLS058A - D1322, SEPTEMBER 1973 - REVISED JANUARY 1993

## APPLICATION INFORMATION

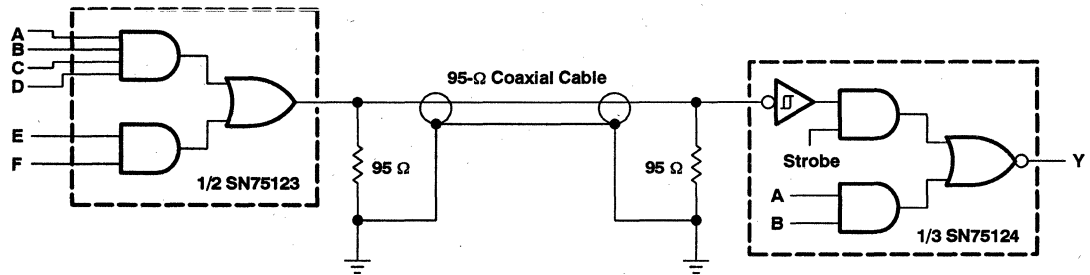


Figure 3. Unbalanced Line Communication Using SN75123 and SN75124

# SN75125, SN75127 SEVEN-CHANNEL LINE RECEIVERS

SLLS108A – D239, JANUARY 1977 – REVISED FEBRUARY 1993

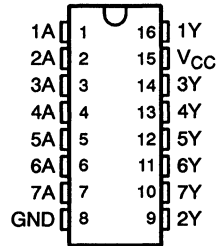
- Meets IBM 360/370 I/O Specification
- Input Resistance . . . 7 k $\Omega$  to 20 k $\Omega$
- Output Compatible With TTL
- Schottky-Clamped Transistors
- Operates From Single 5-V Supply
- High Speed . . . Low Propagation Delay
- Ratio Specification for Propagation Delay Time, Low-to-High/High-to-Low
- Seven Channels in One 16-Pin Package
- Standard V<sub>CC</sub> and Ground Positioning on SN75127

## description

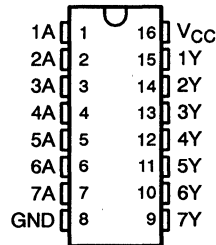
The SN75125 and SN75127 are monolithic seven-channel line receivers designed to satisfy the requirements of the IBM System 360/370 input/output interface specifications. Special low-power design and Schottky-clamped transistors allow for low supply-current requirements while maintaining fast switching speeds and high-current TTL outputs.

The SN75125 and SN75127 are characterized for operation from 0°C to 70°C.

SN75125 . . . D OR N PACKAGE  
(TOP VIEW)

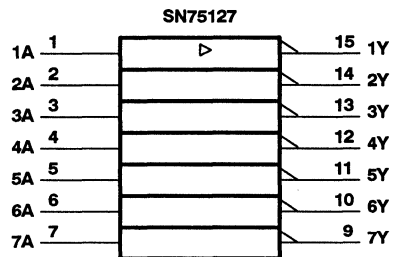
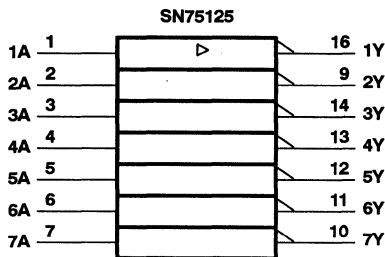


SN75127 . . . D OR N PACKAGE  
(TOP VIEW)



**THE SN75125 IS NOT  
RECOMMENDED FOR NEW DESIGN**

## logic symbols†



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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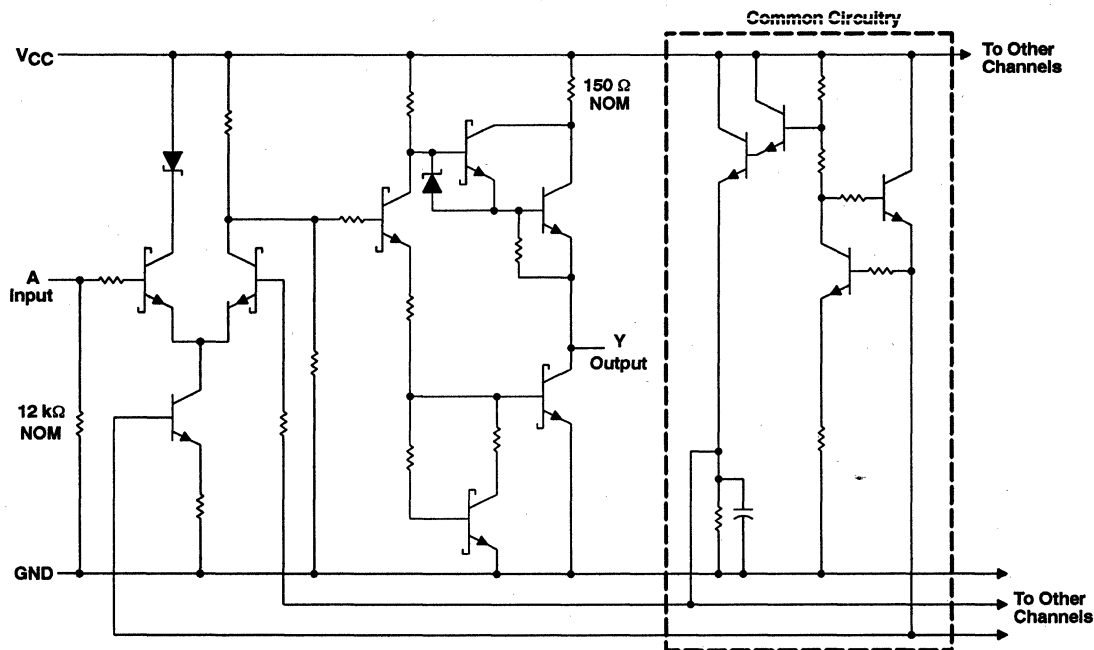
**TEXAS  
INSTRUMENTS**

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# SN75125, SN75127 SEVEN-CHANNEL LINE RECEIVERS

SLLS108A - D239, JANUARY 1977 - REVISED FEBRUARY 1993

## schematic (each receiver)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage range: SN75125	-0.15 V to 7 V
SN75127	-2 V to 7 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1050 mW	9.2 mW/°C	736 mW

# SN75125, SN75127 SEVEN-CHANNEL LINE RECEIVERS

SLLS108A - D239, JANUARY 1977 - REVISED FEBRUARY 1993

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.5	V
High-level input voltage, $V_{IH}$	1.7			V
Low-level input voltage, $V_{IL}$			0.7	V
High-level output current, $I_{OH}$			-0.4	mA
Low-level output current, $I_{OL}$			16	mA
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}$ High-level output voltage	$V_{CC} = 4.5$ V, $V_{IL} = 0.7$ V, $I_{OH} = -0.4$ mA	2.4	3.1		V
$V_{OL}$ Low-level output voltage	$V_{CC} = 4.5$ V, $V_{IH} = 1.7$ V, $I_{OL} = 16$ mA		0.4	0.5	V
$I_{IH}$ High-level input current	$V_{CC} = 5.5$ V, $V_I = 3.11$ V		0.3	0.42	mA
$I_{IL}$ Low-level input current	$V_{CC} = 5.5$ V, $V_I = 0.15$ V			30	μA
$I_{OS}$ Short-circuit output current‡	$V_{CC} = 5.5$ V, $V_O = 0$	-18		-60	mA
$r_i$ Input resistance	$V_{CC} = 4.5$ V, 0 V, or open, $\Delta V_I = 0.15$ V to 4.15 V	7		20	kΩ
$I_{CC}$ Supply current	$V_{CC} = 5.5$ V, $I_{OH} = -0.4$ mA, All inputs at 0.7 V		15	25	mA
	$V_{CC} = 5.5$ V, $I_{OL} = 16$ mA, All inputs at 4 V		28	47	mA

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

‡ Not more than one output should be shorted at a time.

## switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 400$ Ω, $C_L = 50$ pF, See Figure 1	7	14	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output		10	18	30	ns
$\frac{t_{PLH}}{t_{PHL}}$ Ratio of propagation delay times		0.5	0.8	1.3	
$t_{TLH}$ Transition time, low-to-high-level output		1	7	12	ns
$t_{THL}$ Transition time, high-to-low-level output		1	3	12	ns

  
**TEXAS**  
**INSTRUMENTS**

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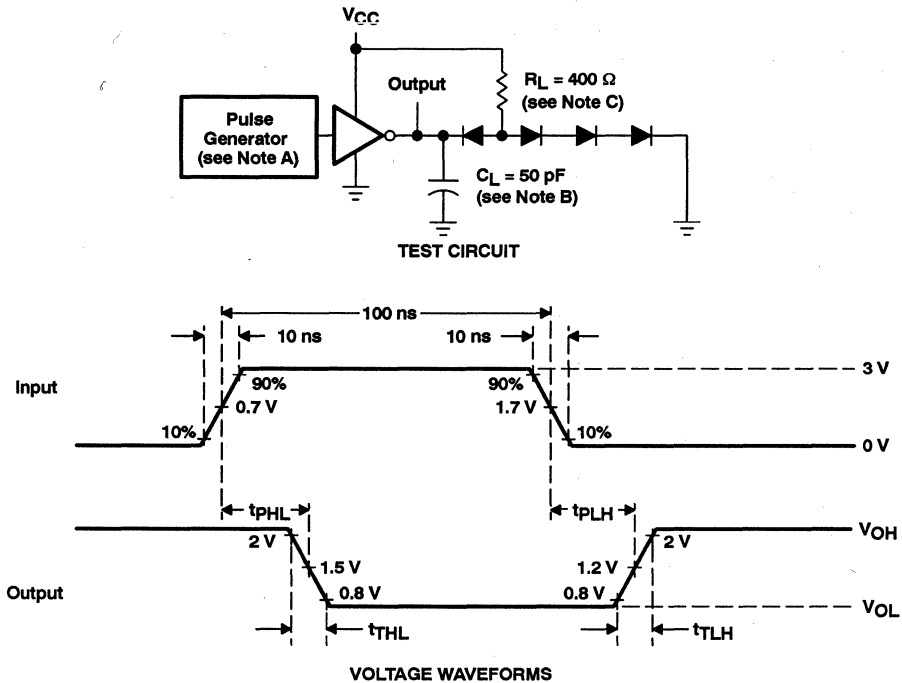
2-275



# SN75125, SN75127 SEVEN-CHANNEL LINE RECEIVERS

SLLS108A – D239, JANUARY 1977 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The pulse generator has the following characteristics:  $Z_O \sim 50 \Omega$ ,  $PRR \leq 5 \text{ MHz}$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.

Figure 1. Tests Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

VOLTAGE TRANSFER CHARACTERISTICS

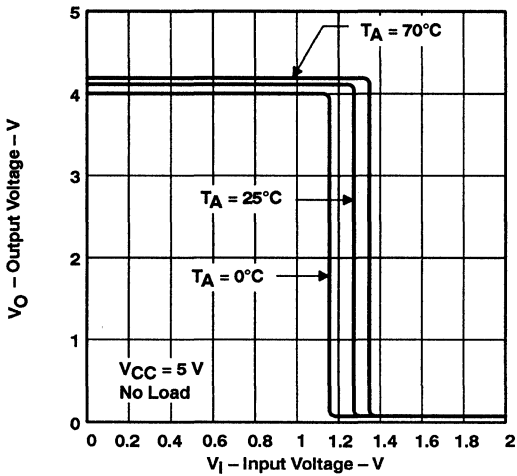


Figure 2

VOLTAGE TRANSFER CHARACTERISTICS

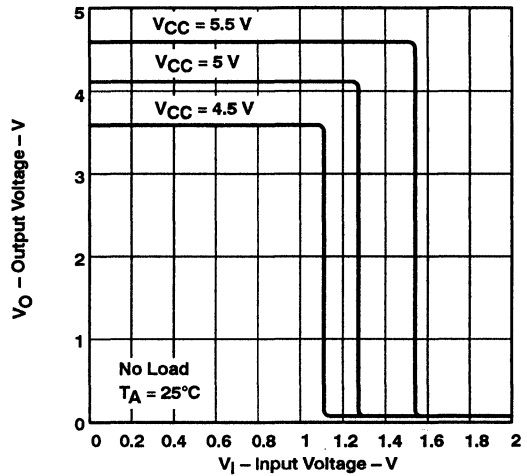


Figure 3

INPUT CURRENT  
vs  
INPUT VOLTAGE

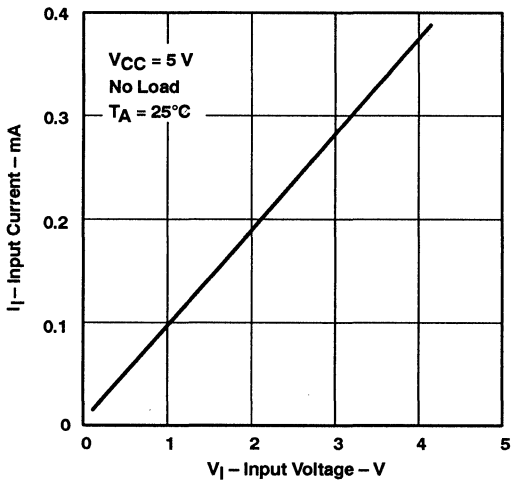


Figure 4

LOW-LEVEL OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT

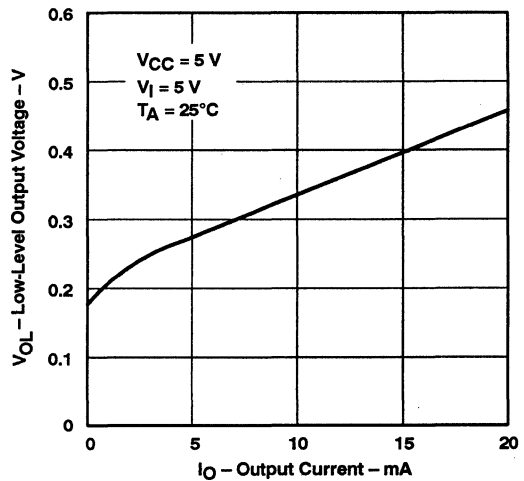


Figure 5

# SN75125, SN75127 SEVEN-CHANNEL LINE RECEIVERS

SLLS108A - D239, JANUARY 1977 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

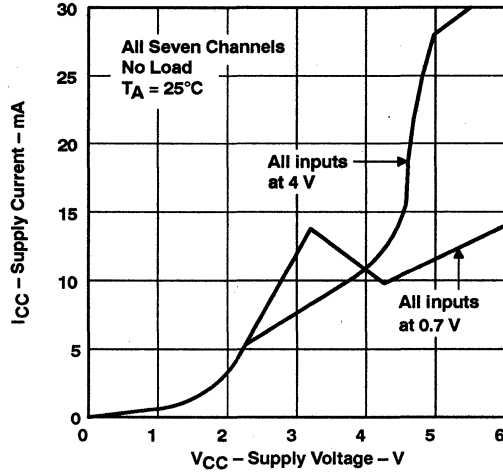


Figure 6

TEXAS  
INSTRUMENTS

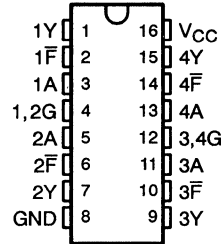
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# SN75126 QUADRUPLE LINE DRIVER

SLLS060A – D3405, FEBRUARY 1990 – REVISED FEBRUARY 1993

- **Meets IBM 360/370 I/O Interface Specification GA22-6974-3 (Also See SN55ALS126 and SN75ALS126)**
- **Minimum Output Voltage of 3.11 V at  $I_{OH} = -59.3$  mA**
- **Fault-Flag Circuit Output Signals Driver Output Fault**
- **Fault-Detection Current-Limit Circuit Minimizes Power Dissipation During a Fault Condition**
- **Dual Common Enable**
- **Individual Fault Flags**
- **Designed to Replace the MC3481**

**D OR N PACKAGE  
(TOP VIEW)**



**FUNCTION TABLE**

INPUTS		OUTPUTS	
G	A	Y	F
L	X	L	H
H	H	H	H
H	H	S	L

H = high level,  
L = low level,  
X = irrelevant,  
S = shorted to ground

## description

The SN75126 quadruple line driver is designed to meet the IBM 360/370 I/O specification A22-6974-3. The output voltage is 3.11 V minimum (at  $I_{OH} = -59.3$  mA) over the recommended ranges of supply voltage (4.5 V to 5.95 V) and temperature. Driver outputs use a fault-detection current-limit circuit to allow high drive current but still minimize power dissipation when the output is shorted to ground. The SN75126 is compatible with standard TTL logic and supply voltages.

Fault-flag circuitry is designed to sense and signal a line short on any Y line. Upon detecting an output fault condition, the fault-flag circuit forces the driver output into a low state and signals a fault condition by causing the fault-flag output to go low.

The SN75126 can drive a 50- $\Omega$  load as required in the IBM GA22-6974-3 specification or a 90- $\Omega$  load as used in many I/O systems. Optimum performance can be achieved when the device is used with either the SN75125, SN75127, SN75128, or SN75129 line receivers.

The SN75126 is characterized for operation from 0°C to 70°C.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



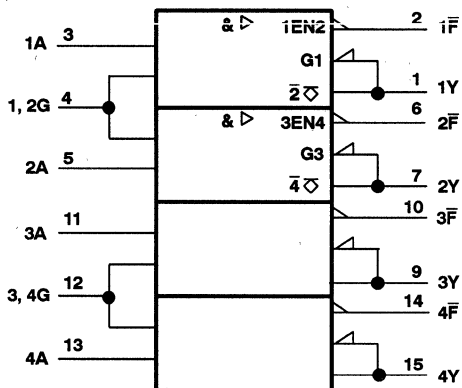
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# SN75126 QUADRUPLE LINE DRIVER

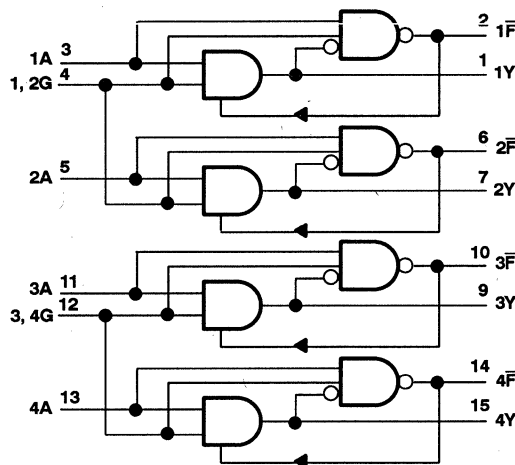
SLLS060A – D3405, FEBRUARY 1990 – REVISED FEBRUARY 1993

## logic symbol†

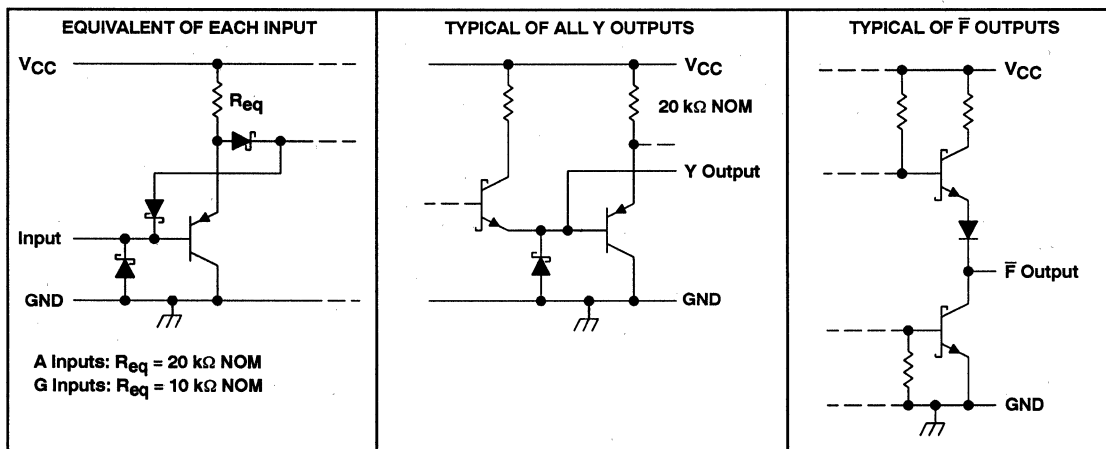


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, VCC	7 V
Input voltage, Vi	7 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, TA	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

DISSIPATION RATING TABLE

PACKAGE	TA = 25°C POWER RATING	DERATING FACTOR ABOVE TA = 25°C	TA = 70°C POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

# SN75126 QUADRUPLE LINE DRIVER

SLLS060A – D3405, FEBRUARY 1990 – REVISED FEBRUARY 1993

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.95	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-59.3	mA
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended operating free-air temperature range

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT
$V_{IK}$	Input clamp voltage	A, G	$V_{CC} = 4.5\text{ V}$ , $I_I = -18\text{ mA}$		-1.5	V
$V_{OH}$	High-level output voltage	Y	$V_{CC} = 4.5\text{ V}$ , $I_{OH} = -59.3\text{ mA}$ , $V_{IH} = 2\text{ V}$	3.11		V
		Y	$V_{CC} = 5.25\text{ V}$ , $I_{OH} = -41\text{ mA}$ , $V_{IH} = 2\text{ V}$	3.9		
		$\bar{F}$	$V_{CC} = 4.5\text{ V}$ , $I_{OH} = -400\text{ }\mu\text{A}$ , $V_{IH} = 2\text{ V}$	2.5		
$V_{OL}$	Low-level output voltage	Y	$V_{CC} = 5.5\text{ V}$ , $I_{OL} = -240\text{ }\mu\text{A}$ , $V_{IL} = 0.8\text{ V}$		0.15	V
		Y	$V_{CC} = 5.95\text{ V}$ , $I_{OL} = -1\text{ mA}$ , $V_{IL} = 0.8\text{ V}$		0.15	
		$\bar{F}$	$V_{CC} = 4.5\text{ V}$ , $I_{OL} = 8\text{ mA}$ , Y at 0 V, $V_{IH} = 2\text{ V}$		0.5	
$I_{O(off)}$	Off-state output current	Y	$V_{CC} = 4.5\text{ V}$ , $V_I = 0$ , $V_O = 3.11\text{ V}$		100	$\mu\text{A}$
		Y	$V_{CC} = 0$ , $V_I = 0$ , $V_O = 3.11\text{ V}$		200	
$I_I$	Input current	A	$V_{CC} = 4.5\text{ V}$ , $V_I = 5.5\text{ V}$		100	$\mu\text{A}$
		G			200	
$I_{IH}$	High-level input current	A	$V_{CC} = 4.5\text{ V}$ , $V_I = 2.7\text{ V}$		20	$\mu\text{A}$
		G			40	
$I_{IL}$	Low-level input current	A	$V_{CC} = 5.95\text{ V}$ , $V_I = 0.4\text{ V}$		-250	$\mu\text{A}$
		G			-500	
$I_{OS}$	Short-circuit output current	Y	$V_{CC} = 5.5\text{ V}$ , $V_O = 0$ , $V_{IH} = 2.7\text{ V}$		-5	mA
		$\bar{F}$	$V_{CC} = 5.5\text{ V}$ , $V_O = 0$	-15	-100	
		Y	$V_{CC} = 5.95\text{ V}$ , $V_O = 0$ , $V_{IH} = 2.7\text{ V}$		-5	
		$\bar{F}$	$V_{CC} = 5.95\text{ V}$ , $V_O = 0$	-15	-110	
$I_{CC(H)}$	Supply current, all outputs high		$V_{CC} = 5.5\text{ V}$ , No load, $V_{IH} = 2\text{ V}$		70	mA
			$V_{CC} = 5.95\text{ V}$ , No load, $V_{IH} = 2\text{ V}$		80	
$I_{CC(L)}$	Supply current, Y outputs low		$V_{CC} = 5.5\text{ V}$ , No load, $V_{IL} = 0.8\text{ V}$		55	mA
			$V_{CC} = 5.95\text{ V}$ , No load, $V_{IL} = 0.8\text{ V}$		70	



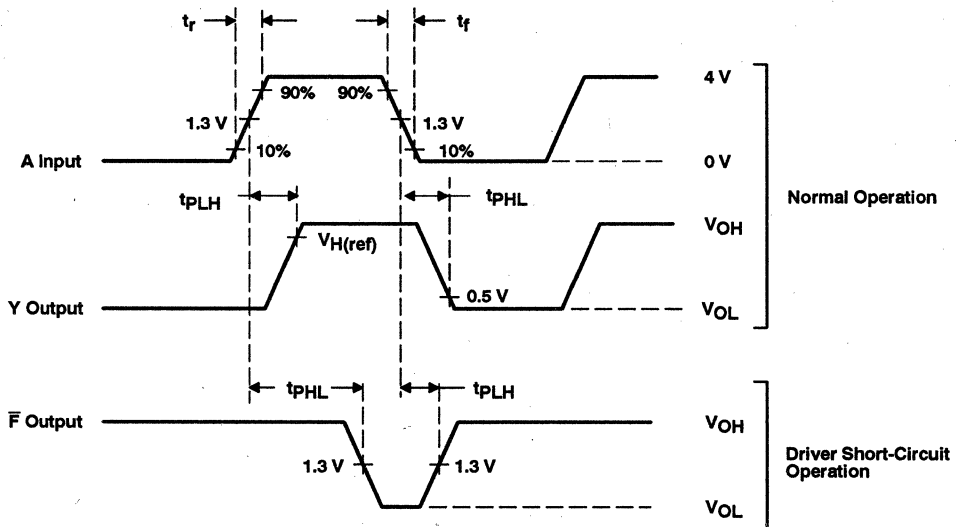
# SN75126 QUADRUPLE LINE DRIVER

SLLS060A – D3405, FEBRUARY 1990 – REVISED FEBRUARY 1993

## switching characteristics at $T_A = 25^\circ\text{C}$

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	A	Y	V <sub>CC</sub> = 4.5 V to 5.5 V, R <sub>L</sub> = 50 Ω, C <sub>L</sub> = 50 pF, V <sub>H(ref)</sub> = 3.11 V, See Figures 1 and 2		40	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					37	ns
$\frac{t_{PLH}}{t_{PHL}}$	Ratio of propagation delay times				0.3	3	
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	A	Y	V <sub>CC</sub> = 5.25 V to 5.95 V, R <sub>L</sub> = 90 Ω, C <sub>L</sub> = 50 pF, V <sub>H(ref)</sub> = 3.9 V, See Figures 1 and 2		45	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					45	ns
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	A	F	V <sub>CC</sub> = 5 V, C <sub>L</sub> = 15 pF, R <sub>L</sub> = 2 kΩ, See Figures 1 and 2		60	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					100	ns

### PARAMETER MEASUREMENT INFORMATION

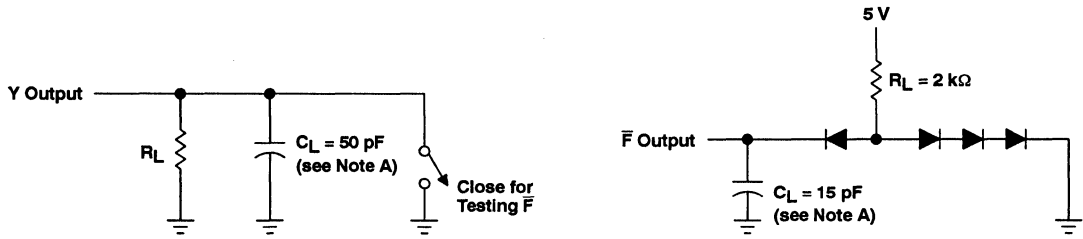


NOTE: The input pulse is supplied by a generator having the following characteristics: PRR ≤ 1 MHz, duty cycle ≤ 50%, t<sub>r</sub> ≤ 6 ns, t<sub>f</sub> ≤ 6 ns, Z<sub>O</sub> ~ 50 Ω.

Figure 1. Input and Output Voltage Waveforms



PARAMETER MEASUREMENT INFORMATION



NOTE A:  $C_L$  includes probe and stray capacitance.

Figure 2. Switching Characteristics Load Circuits





# SN75ALS126 QUADRUPLE LINE DRIVER

SGLS017B – D2299, FEBRUARY 1986 – REVISED FEBRUARY 1993

- Meets IBM 360/370 I/O Interface Specification GA22-6974-3 (Also See SN75ALS130)
- Minimum Output Voltage of 3.11 V at  $I_{OH} = -60$  mA
- Fault-Flag Circuit Output Signals Driver Output Fault
- Fault-Detection Current Limit Circuit Minimizes Power Dissipation During a Fault Condition
- Advanced Low-Power Schottky Circuitry
- Dual Common Enable
- Individual Fault Flags
- Designed to Be an Improved Replacement for the MC3481

## description

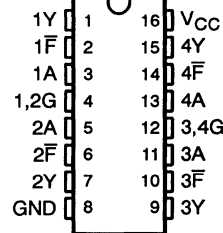
The SN75ALS126 quadruple line driver is designed to meet the IBM 360/370 I/O specification GA22-6974-3. The output voltage is 3.11 V minimum (at  $I_{OH} = -59.3$  mA) over the recommended ranges of supply voltage (4.5 V to 5.95 V) and temperature. Driver outputs use a fault-detection current-limit circuit to allow high drive current but still minimize power dissipation when the output is shorted to ground. The SN75ALS126 is compatible with standard TTL logic and supply voltages

The SN75ALS126 employs the IMPACT™ process to achieve fast switching speeds and low power dissipation. Fault-flag circuitry is designed to sense and signal a line short on any Y line. Upon detecting an output fault condition, the fault-flag circuit forces the driver output into a low state and signals a fault condition by causing the fault-flag output to go low.

The SN75ALS126 can drive a 50-Ω load as required in the IBM GA22-6974-3 specification or a 90-Ω load as used in many I/O systems. Optimum performance can be achieved when the device is used with either the SN75127, SN75128, or SN75129 line receivers.

The SN75ALS126 is characterized for operation from 0°C to 70°C.

D OR N PACKAGE  
(TOP VIEW)



FUNCTION TABLE

INPUTS		OUTPUTS	
G	A	Y	F
L	X	L	H
H	H	H	H
H	H	S	L

H = high level, L = low level,  
X = irrelevant,  
S = shorted to GND

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PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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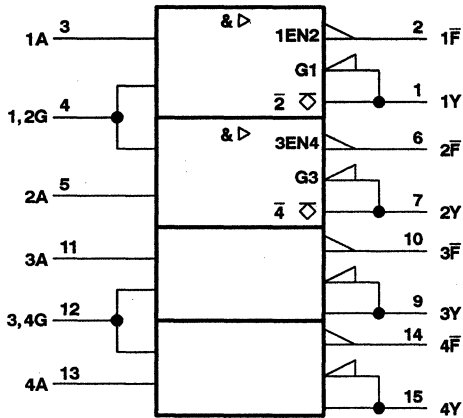
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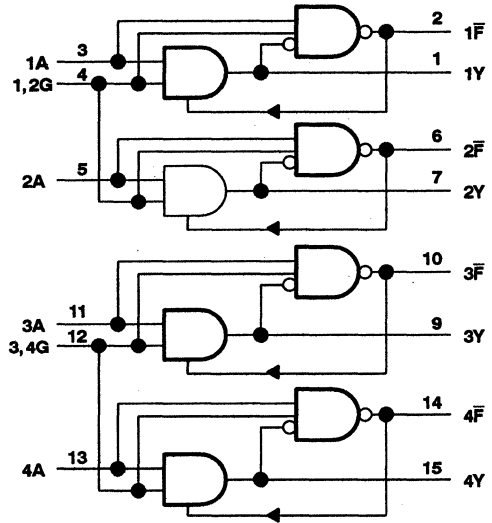
# SN75ALS126 QUADRUPLE LINE DRIVER

SGLS017B - D2299, FEBRUARY 1986 - REVISED FEBRUARY 1993

## logic symbol†

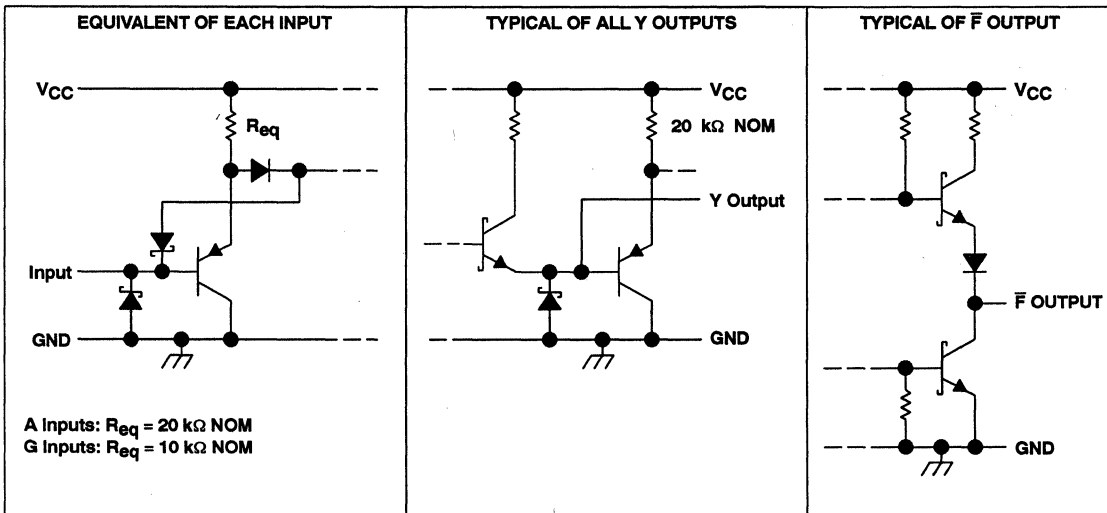


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematics of inputs and outputs



# SN75ALS126 QUADRUPLE LINE DRIVER

SGLS017B – D2299, FEBRUARY 1986 – REVISED FEBRUARY 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ .....	7 V
Input voltage .....	7 V
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	– 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.95	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			– 59.3	mA
Operating free-air temperature, $T_A$	0		70	°C



# SN75ALS126 QUADRUPLE LINE DRIVER

SGLS017B - D2299, FEBRUARY 1986 - REVISED FEBRUARY 1993

## electrical characteristics over recommended operating free-air temperature range

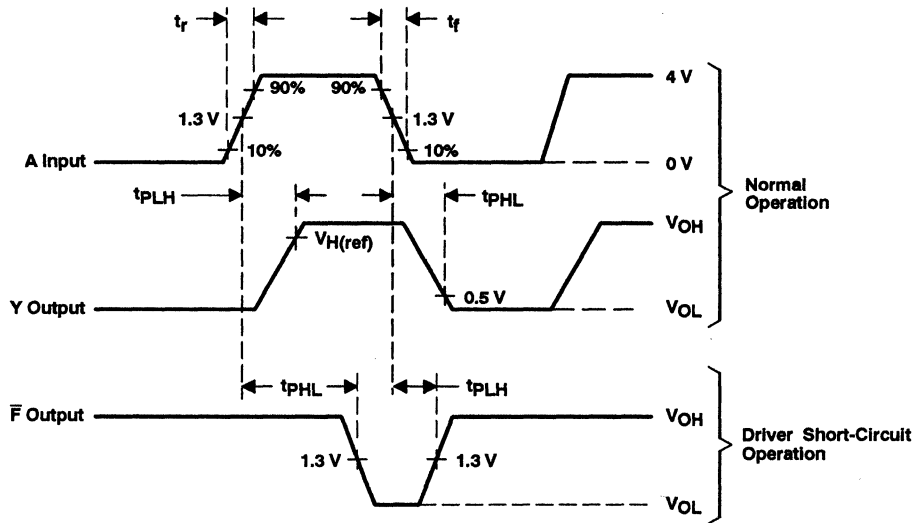
PARAMETER			TEST CONDITIONS		MIN	MAX	UNIT	
V <sub>IK</sub>	Input clamp voltage	A,G	V <sub>CC</sub> = 4.5 V,	I <sub>I</sub> = -18 mA		-1.5	V	
		Y	V <sub>CC</sub> = 4.5 V,	I <sub>OH</sub> = -59.3 mA, V <sub>IH</sub> = 2 V	3.11			
V <sub>OH</sub>	High-level output voltage	Y	V <sub>CC</sub> = 5.25 V,	I <sub>OH</sub> = -41 mA, V <sub>IH</sub> = 2 V	3.9		V	
		F	V <sub>CC</sub> = 4.5 V,	I <sub>OH</sub> = -400 μA, V <sub>IH</sub> = 2 V	2.5			
V <sub>OL</sub>	Low-level output voltage	Y	V <sub>CC</sub> = 5.5 V,	I <sub>OL</sub> = -240 μA, V <sub>IL</sub> = 0.8 V		0.15		
		Y	V <sub>CC</sub> = 5.95 V,	I <sub>OL</sub> = -1 mA, V <sub>IL</sub> = 0.8 V		0.15	V	
		F	V <sub>CC</sub> = 4.5 V,	I <sub>OL</sub> = 8 mA, Y at 0 V		0.5		
I <sub>O(off)</sub>	Off-state output current	Y	V <sub>CC</sub> = 4.5 V,	V <sub>IL</sub> = 0, V <sub>O</sub> = 3.11 V		100	μA	
		Y	V <sub>CC</sub> = 0 V,	V <sub>IL</sub> = 0, V <sub>O</sub> = 3.11 V		200		
I <sub>I</sub>	Input current	A	V <sub>CC</sub> = 4.5 V,	V <sub>I</sub> = 5.5 V		100	μA	
		G			400			
I <sub>IH</sub>	High-level input current	A	V <sub>CC</sub> = 4.5 V,	V <sub>I</sub> = 2.7 V		20	μA	
		G			80			
I <sub>IL</sub>	Low-level input current	A	V <sub>CC</sub> = 5.95 V,	V <sub>I</sub> = 0.4 V		-250	μA	
		G			-1000			
I <sub>OS</sub>	Short-circuit output	Y	V <sub>CC</sub> = 5.5 V,	V <sub>O</sub> = 0, V <sub>IH</sub> = 2.7 V		-5	mA	
		F	V <sub>CC</sub> = 5.5 V,	V <sub>O</sub> = 0		-15		-100
		Y	V <sub>CC</sub> = 5.95 V,	V <sub>O</sub> = 0, V <sub>IH</sub> = 2.7 V		-5		
		F	V <sub>CC</sub> = 5.95 V,	V <sub>O</sub> = 0		-15		-110
I <sub>CCH</sub>	Supply current, all outputs high		V <sub>CC</sub> = 5.5 V,	No load, V <sub>IH</sub> = 2.7 V		25	mA	
			V <sub>CC</sub> = 5.95 V,	No load, V <sub>IH</sub> = 2.7 V		27		
I <sub>CCL</sub>	Supply current, Y outputs low		V <sub>CC</sub> = 5.5 V,	No load, V <sub>IL</sub> = 0.4 V		45	mA	
			V <sub>CC</sub> = 5.95 V,	No load, V <sub>IL</sub> = 0.4 V		47		

## switching characteristics over recommended operating free-air temperature range

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS		MIN	MAX	UNIT
t <sub>PLH</sub>	A	Y	V <sub>CC</sub> = 4.5 V to 5.5 V, R <sub>L</sub> = 50 Ω, See Figures 1 and 2	C <sub>L</sub> = 50 pF, V <sub>H(ref)</sub> = 3.11 V,		30	ns
t <sub>PHL</sub>						28	ns
t <sub>PLH</sub> / t <sub>PHL</sub>						0.3	3
t <sub>PLH</sub>	A	Y	V <sub>CC</sub> = 5.25 V to 5.95 V, R <sub>L</sub> = 90 Ω, See Figures 1 and 2	C <sub>L</sub> = 50 pF, V <sub>H(ref)</sub> = 3.9 V,		34	ns
t <sub>PHL</sub>						34	ns
t <sub>PLH</sub>	A	F	V <sub>CC</sub> = 5 V, C <sub>L</sub> = 15 pF,	R <sub>L</sub> = 2 kΩ, See Figures 1 and 2		45	ns
t <sub>PHL</sub>						75	ns

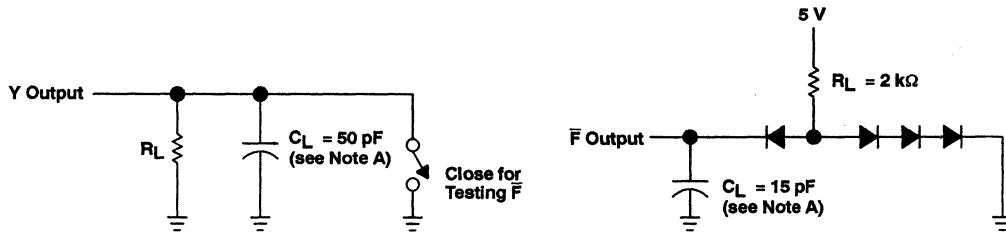


PARAMETER MEASUREMENT INFORMATION



NOTE: The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

Figure 1. Input and Output Voltage Waveforms



NOTE A:  $C_L$  includes probe and stray capacitance.

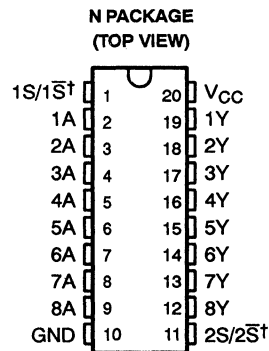
Figure 2. Switching Characteristics Load Circuits



# SN75128, SN75129 EIGHT-CHANNEL LINE RECEIVERS

SLLS076A – D2305, JANUARY 1977 – REVISED MARCH 1993

- Meets IBM 360/370 I/O Specification
- Input Resistance . . . 7 k $\Omega$  to 20 k $\Omega$
- Output Compatible With TTL
- Schottky-Clamped Transistors
- Operates From a Single 5-V Supply
- High Speed . . . Low Propagation Delay
- Ratio Specification . . . t<sub>PLH</sub> / t<sub>PHL</sub>
- Common Strobe for Each Group of Four Receivers
- SN75128 . . . Active-High Strobes  
SN75129 . . . Active-Low Strobes



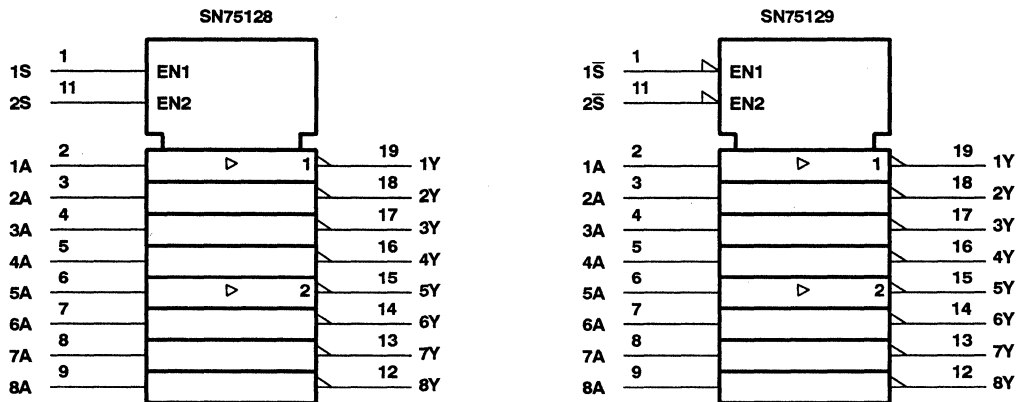
† S and  $\bar{S}$  for SN75128 and SN75129, respectively

## description

The SN75128 and SN75129 are eight-channel line receivers designed to satisfy the requirements of the input-output interface specification for IBM 360/370. Both devices feature common strobes for each group of four devices. The SN75128 has active-high strobes; the SN75129 has active-low strobes. Special low-power design and Schottky-diode-clamped transistors allow low supply-current requirements while maintaining fast switching speeds and high-current TTL outputs.

The SN75128 and SN75129 are characterized for operation from 0°C to 70°C.

## logic symbols†



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

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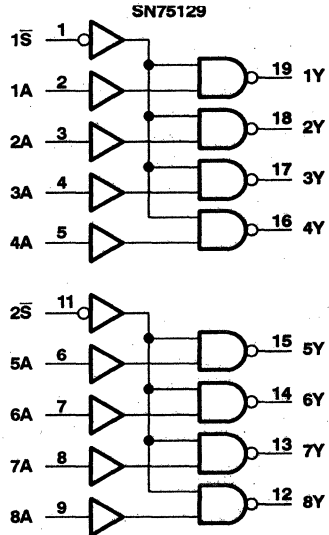
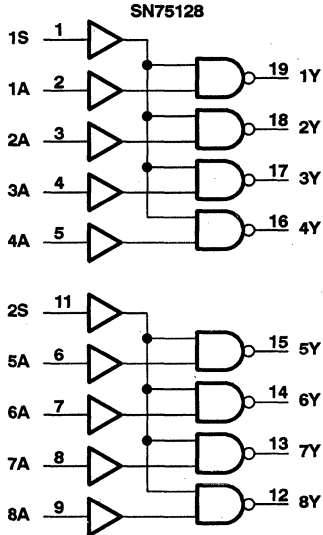
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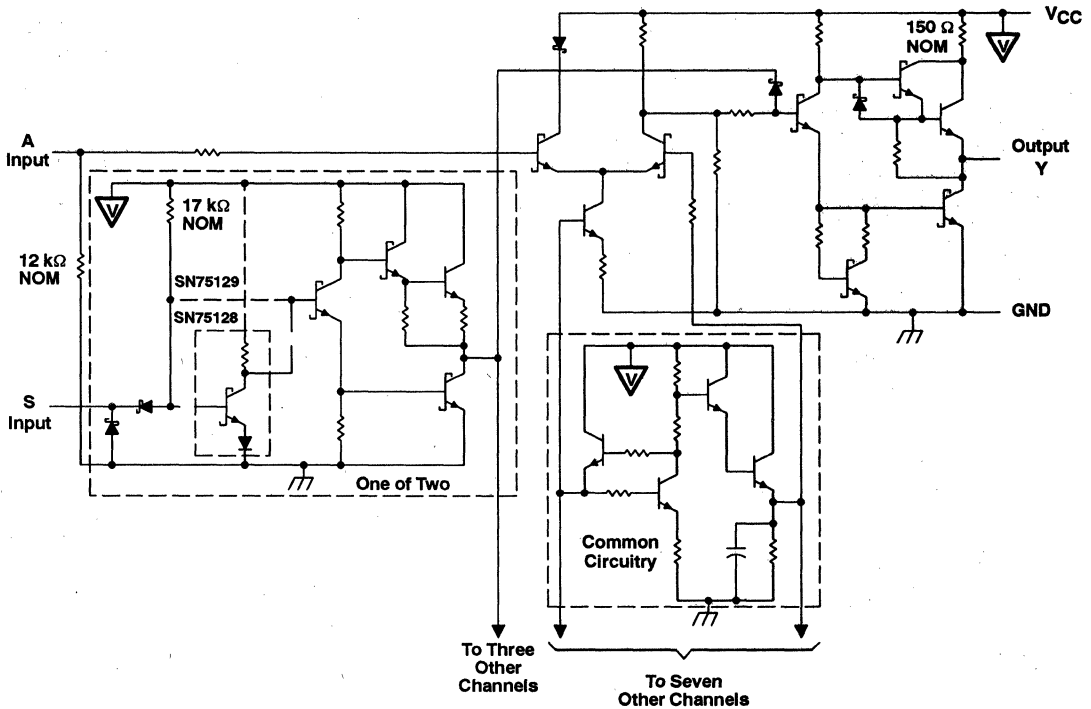
# SN75128, SN75129 EIGHT-CHANNEL LINE RECEIVERS

SLLS076A - D2305, JANUARY 1977 - REVISED MARCH 1993

## logic diagrams (positive logic)



## schematic (each driver)



# SN75128, SN75129 EIGHT-CHANNEL LINE RECEIVERS

SLLS076A – D2305, JANUARY 1977 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage range, A, $V_I$	-0.15 V to 7 V
Input voltage, S, $V_I$	7 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	T <sub>A</sub> ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.5	5	5.5	V
High-level input voltage, $V_{IH}$	A	1.7			V
	S	2			
Low-level input voltage, $V_{IL}$	A	0.7			V
	S	0.7			
High-level output current, $I_{OH}$		-0.4			mA
Low-level output current, $I_{OL}$		16			mA
Operating free-air temperature, T <sub>A</sub>		0	70		°C

## electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS			MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage	$V_{CC} = 4.5$ V, $V_{IL} = 0.7$ V, $I_{OH} = -0.4$ mA			2.4	3.1		V
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.5$ V, $V_{IH} = 1.7$ V, $I_{OL} = 16$ mA				0.4	0.5	V
$V_{IK}$	Input clamp voltage	S	$V_{CC} = 4.5$ V, $I_I = -18$ mA				-1.5	V
$I_{IH}$	High-level input current	A	$V_{CC} = 5.5$ V, $V_I = 3.11$ V			0.3	0.42	mA
		S	$V_{CC} = 5.5$ V, $V_I = 2.7$ V				20	μA
$I_{IL}$	Low-level input current	A	$V_{CC} = 5.5$ V, $V_I = 0.15$ V				30	μA
		S	$V_{CC} = 5.5$ V, $V_I = 0.4$ V				-0.4	mA
$I_{OS}$	Short-circuit output current‡	$V_{CC} = 5.5$ V, $V_O = 0$			-18		-60	mA
$r_i$	Input resistance	$V_{CC} = 4.5$ V, 0 V or open, $\Delta V_I = 0.15$ V to 4.15 V			7		20	kΩ
$I_{CC}$	Supply current	SN75128	$V_{CC} = 5.5$ V, Strobe at 2.4 V, All A inputs at 0.7 V			19	31	mA
		SN75129	$V_{CC} = 5.5$ V, Strobe at 0.4 V, All A inputs at 0.7 V			19	31	
		SN75128	$V_{CC} = 5.5$ V, Strobe at 2.4 V, All A inputs at 4 V			32	53	
		SN75129	$V_{CC} = 5.5$ V, Strobe at 0.4 V, All A inputs at 4 V			32	53	

† All typical values are at  $V_{CC} = 5$  V, T<sub>A</sub> = 25°C.

‡ Not more than one output should be shorted at a time.



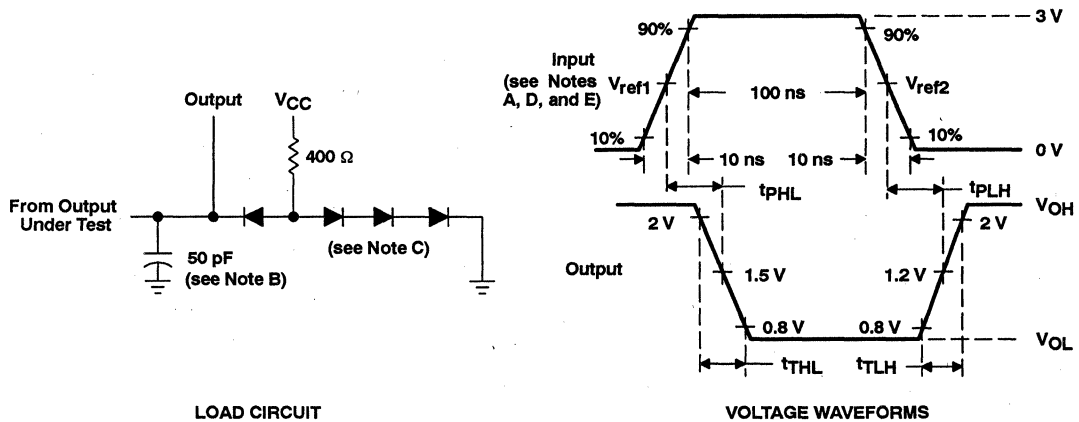
# SN75128, SN75129 EIGHT-CHANNEL LINE RECEIVERS

SLLS076A - D2305, JANUARY 1977 - REVISED MARCH 1993

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	FROM	TEST CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	A	$R_L = 400\ \Omega$ , $C_L = 50\ \text{pF}$ , See Figure 1	7	14	25	7	14	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output	A		10	18	30	10	18	30	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	S		26	40		20	35	ns	
$t_{PHL}$ Propagation delay time, high-to-low-level output	S		22	35		16	30	ns	
$\frac{t_{PLH}}{t_{PHL}}$ Ratio of propagation delay times	A		0.5	0.8	1.3	0.5	0.8	1.3	
$t_{TLH}$ Transition time, low-to-high-level output			1	7	12	1	7	12	ns
$t_{THL}$ Transition time, high-to-low-level output			1	3	12	1	3	12	ns

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. Input pulses are supplied by a generator having the following characteristics:  $Z_O = 50\ \Omega$ ,  $\text{PRR} \leq 5\ \text{MHz}$ .  
 B. Includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.  
 D. The strobe inputs of SN75129 are in phase with the output.  
 E.  $V_{ref1} = 0.7\ \text{V}$  and  $V_{ref2} = 1.7\ \text{V}$  for testing data (A) inputs,  $V_{ref1} = V_{ref2} = 1.3\ \text{V}$  for strobe inputs.

Figure 1. Load Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

VOLTAGE TRANSFER CHARACTERISTICS

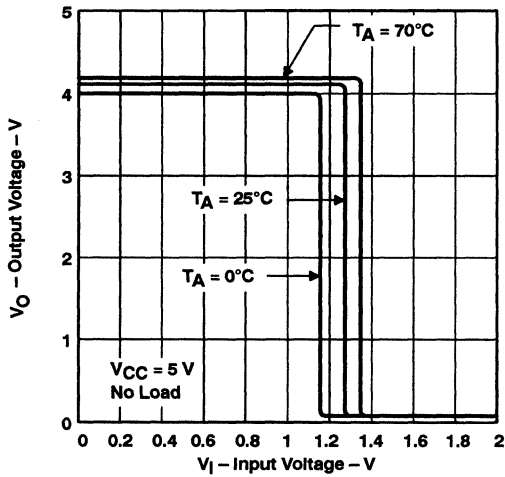


Figure 2

VOLTAGE TRANSFER CHARACTERISTICS FROM A INPUTS

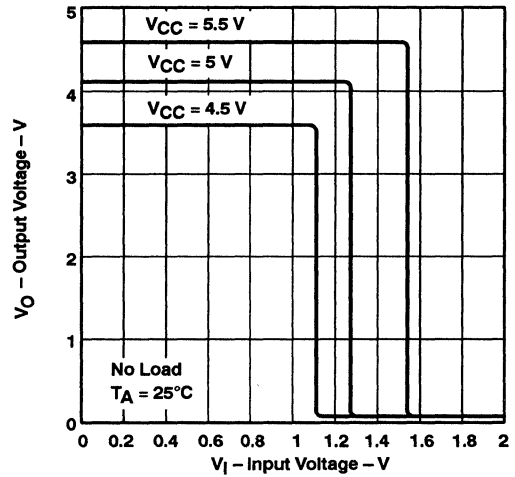


Figure 3

INPUT CURRENT  
vs  
INPUT VOLTAGE

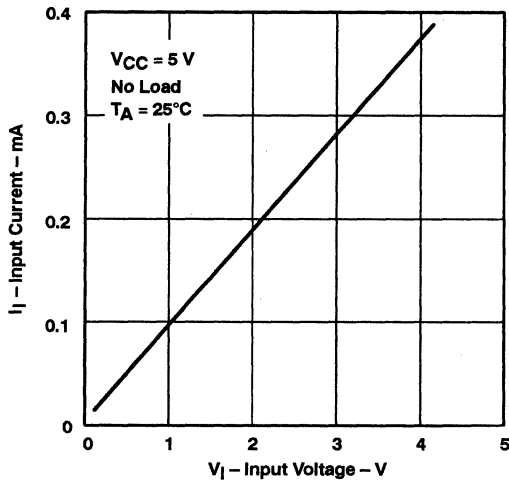


Figure 4

LOW-LEVEL OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT

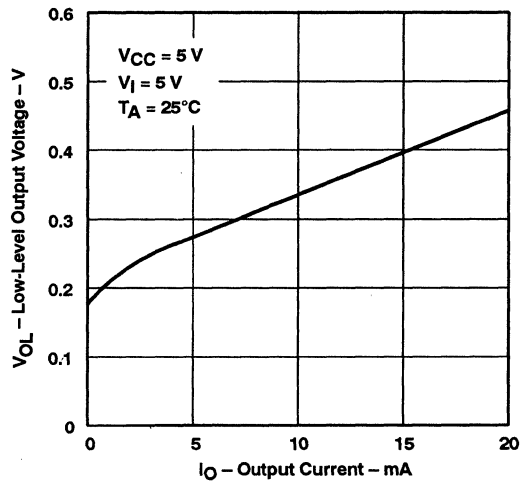


Figure 5

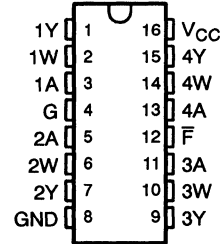


# SN75130 QUADRUPLE LINE DRIVER

SLLS077A – D3406, FEBRUARY 1990 – REVISED FEBRUARY 1993

- Meets IBM 360/370 I/O Interface Specification GA22-6974-3 (Also See SN75ALS130)
- Minimum Output Voltage of 3.11 V at  $I_{OH} = -59.3$  mA
- Fault-Flag Circuit Output Signals Driver Output Fault
- Fault-Detection Current-Limit Circuit Minimizes Power Dissipation During a Fault Condition
- Common Enable and Common Fault Flag
- Designed to Be an Improved Replacement for the MC3485

D OR N PACKAGE  
(TOP VIEW)



FUNCTION TABLE

INPUTS		OUTPUTS		
G†	A	Y	F	W
L	X	L	H	H
X	L	L	H	H
H	H	H	H	L
H	H	S	L	H

H = high level, L = low level, X = irrelevant,  
S = shorted to ground

† G and F are common to the four drivers. If any of the four Y outputs is shorted, the fault flag will respond.

## description

The SN75130 quadruple line driver is designed to meet the IBM 360/370 I/O specification GA22-6974-3. The output voltage is 3.11 V minimum (at  $I_{OH} = -59.3$  mA) over the recommended ranges of supply voltage (4.5 V to 5.5 V) and temperature (0°C to 70°C). Driver outputs use a fault-detection current-limit circuit to allow high drive current but still minimize power dissipation when the output is shorted to ground. The SN75130 is compatible with standard TTL logic and supply voltages.

Fault-flag circuitry is designed to sense and signal a line short on any Y line. Upon detecting an output fault condition, the fault-flag circuit forces the driver output into the off (low) state and signals a fault condition by causing the fault-flag output to go low.

The SN75130 can drive a 50-Ω load as required in the IBM GA22-6974-3 specification or a 90-Ω load as used in many I/O systems. Optimum performance can be achieved when the device is used with either the SN75125, SN75127, SN75128, or SN75129 line receivers.

The SN75130 is characterized for operation from 0°C to 70°C.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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**INSTRUMENTS**

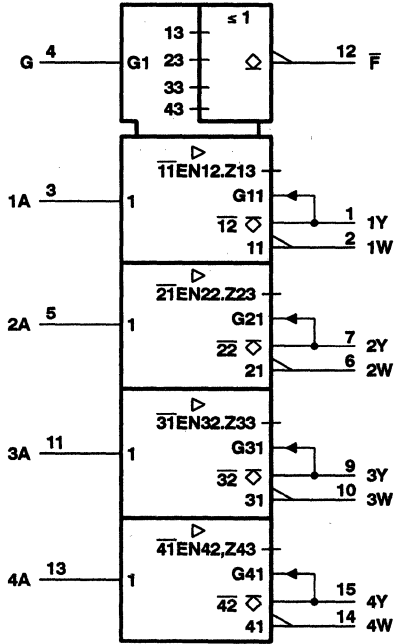
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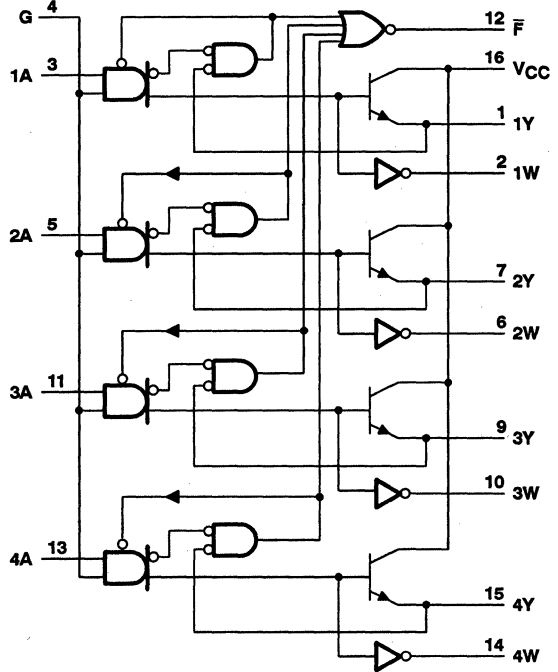
# SN75130 QUADRUPLE LINE DRIVER

SLLS077A - D3406, FEBRUARY 1990 - REVISED FEBRUARY 1993

logic symbol†



logic diagram (positive logic)

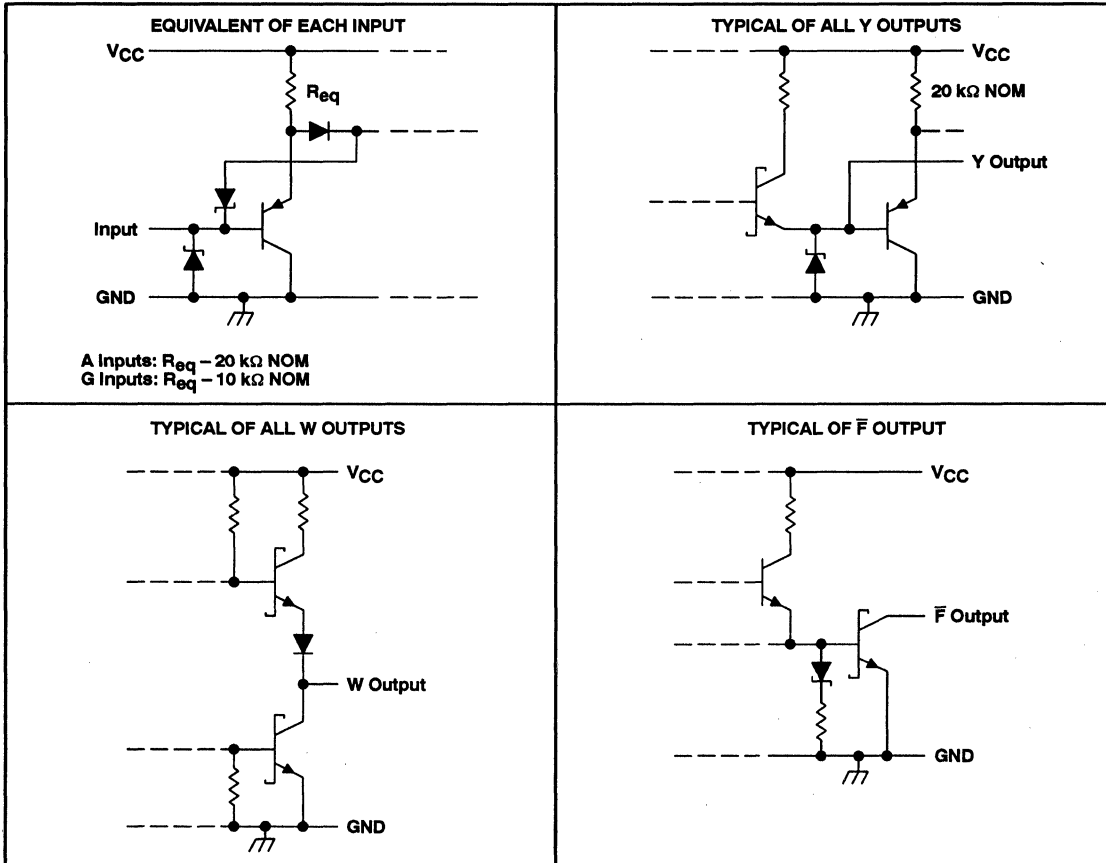


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

# SN75130 QUADRUPLE LINE DRIVER

SLLS077A - D3406, FEBRUARY 1990 - REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC</sub>	7 V
Input voltage	7 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, T <sub>A</sub>	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

DISSIPATION RATING TABLE

PACKAGE	T <sub>A</sub> ≤ 25°C POWER RATING	OPERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

**TEXAS**  
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# SN75130 QUADRUPLE LINE DRIVER

SLLS077A - D3406, FEBRUARY 1990 - REVISED FEBRUARY 1993

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.95	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-59.3	mA
Operating free-air temperature, $T_A$	0		70	°C

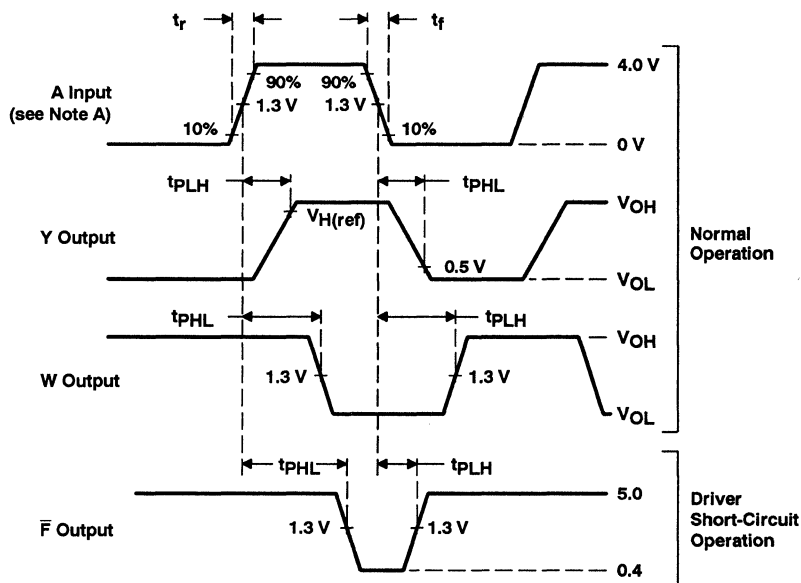
## electrical characteristics over recommended operating free-air temperature range

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT
$V_{IK}$	Input clamp voltage	A, G	$I_I = -18$ mA		-1.5	V
$V_{OH}$	High-level output voltage	Y	$V_{CC} = 4.5$ V, $I_{OH} = -59.3$ mA, $V_{IH} = 2$ V	3.11		V
		Y	$V_{CC} = 5.25$ V, $I_{OH} = -41$ mA, $V_{IH} = 2$ V	3.9		
		W	$V_{CC} = 4.5$ V, $I_{OH} = -400$ $\mu$ A, $V_{IH} = 2$ V	2.5		
$V_{OL}$	Low-level output voltage	Y	$V_{CC} = 5.5$ V, $I_{OL} = -240$ $\mu$ A, $V_{IL} = 0.8$ V		0.15	V
		Y	$V_{CC} = 5.95$ V, $I_{OL} = -1$ mA, $V_{IL} = 0.8$ V		0.15	
		$\bar{F}$	$V_{CC} = 4.5$ V, $I_{OL} = 8$ mA, Y at 0 V		0.5	
		W	$V_{CC} = 4.5$ V, $I_{OL} = 8$ mA		0.5	
$I_{O(off)}$	Off-state output current	Y	$V_{CC} = 4.5$ V, $V_{IL} = 0$ , $V_O = 3.11$ V		100	$\mu$ A
		Y	$V_{CC} = 0$ , $V_{IL} = 0$ , $V_O = 3.11$ V		200	
$I_{OH}$	High-level output current	$\bar{F}$	$V_{CC} = 5.95$ V, $V_{OH} = 5.95$ V		100	$\mu$ A
$I_I$	Input current	A	$V_{CC} = 4.5$ V, $V_{IH} = 5.5$ V		100	$\mu$ A
		G		400		
$I_{IH}$	High-level input current	A	$V_{CC} = 4.5$ V, $V_{IH} = 2.7$ V		20	$\mu$ A
		G		80		
$I_{IL}$	Low-level input current	A	$V_{CC} = 5.95$ V, $V_{IL} = 0.4$ V		250	$\mu$ A
		G		-1000		
$I_{OS}$	Short-circuit output	Y	$V_{CC} = 5.5$ V, $V_O = 0$		-5	mA
		W		-15	-100	
		Y	$V_{CC} = 5.95$ V, $V_O = 0$		-5	
		W		-15	-110	
$I_{CCH}$	Supply current, all outputs high		$V_{CC} = 5.5$ V, $V_I = 2$ V		75	mA
			$V_{CC} = 5.95$ V, $V_I = 2$ V		85	
$I_{CCL}$	Supply current, Y outputs low		$V_{CC} = 5.5$ V, $V_I = 0.8$ V		55	mA
			$V_{CC} = 5.95$ V, $V_I = 0.8$ V		70	

switching characteristics over recommended operating free-air temperature range

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
$t_{PLH}$	A	Y	$V_{CC} = 4.5\text{ V to }5.5\text{ V}$ , $C_L = 50\text{ pF}$ , Input $f = 1\text{ MHz}$ , $V_{H(ref)} = 3.11\text{ V}$ , $R_L = 50\ \Omega$ , See Figures 1 and 2		40	ns
$t_{PHL}$					37	ns
$\frac{t_{PLH}}{t_{PHL}}$					0.3	3
$t_{PLH}$	A	Y	$V_{CC} = 5.25\text{ V to }5.59\text{ V}$ , $C_L = 50\text{ pF}$ , Input $f = 5\text{ MHz}$ , $V_{H(ref)} = 3.9\text{ V}$ , $R_L = 90\ \Omega$ , See Figures 1 and 2		45	ns
$t_{PHL}$					45	ns
$t_{PLH}$	A	W	$V_{CC} = 5\text{ V}$ , $C_L = 15\text{ pF}$ , $R_L = 2\text{ k}\Omega$ , See Figures 1 and 2		45	ns
$t_{PHL}$					28	ns
$t_{PLH}$	A	$\bar{F}$	$V_{CC} = 5\text{ V}$ , $C_L = 15\text{ pF}$ , $R_L = 2\text{ k}\Omega$ , See Figures 1 and 2		60	ns
$t_{PHL}$					100	ns

PARAMETER MEASUREMENT INFORMATION



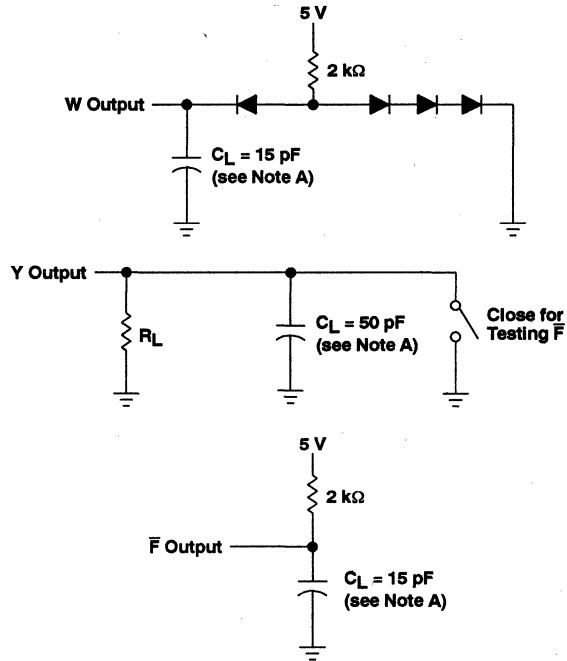
NOTE A: The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r \leq 6\text{ ns}$ ,  $t_f \leq 6\text{ ns}$ ,  $Z_O = 50\ \Omega$ .

Figure 1. Input and Output Voltage Waveforms

# SN75130 QUADRUPLE LINE DRIVER

SLLS077A - D3406, FEBRUARY 1990 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



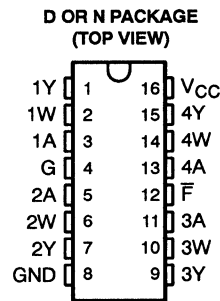
NOTE A:  $C_L$  includes probe and stray capacitance.

Figure 2. Switching Characteristics Load Circuits

# SN75ALS130 QUADRUPLE LINE DRIVER

SLLS024B – D2299, FEBRUARY 1986 – REVISED FEBRUARY 1993

- Meets IBM 360/370 I/O Interface Specification GA22-6974-3 (Also See SN75ALS126)
- Minimum Output Voltage of 3.11 V at  $I_{OH} = -60$  mA
- Fault-Flag Circuit Output Signals Driver Output Fault
- Fault-Detection Current-Limit Circuit Minimizes Power Dissipation During a Fault Condition
- Advanced Low-Power Schottky Circuitry
- Common Enable and Common Fault Flag
- Designed to Be an Improved Replacement for the MC3485



**NOT RECOMMENDED FOR NEW DESIGN**

## description

The SN75ALS130 quadruple line driver is designed to meet the IBM 360/370 I/O specification GA22-6974-3. The output voltage is 3.11 V minimum (at  $I_{OH} = -59.3$  mA) over the recommended ranges of supply voltage (4.5 V to 5.95 V) and temperature. Driver outputs use a fault-detection current-limit circuit to allow high drive current but still minimize power dissipation when the output is shorted to ground. The SN75ALS130 is compatible with standard TTL logic and supply voltages.

The SN75ALS130 employs the IMPACT™ process to achieve fast switching speeds and low power dissipation. Fault-flag circuitry is designed to sense and signal a line short on any Y line. Upon detecting an output fault condition, the fault-flag circuit forces the driver output into a low state and signals a fault condition by causing the fault-flag output to go low.

The SN75ALS130 can drive a 50-Ω load as required in the IBM GA22-6974-3 specification or a 90-Ω load as used in many I/O systems. Optimum performance can be achieved when the devices are used with either the SN75125, SN75127, SN75128, or SN75129 line receivers.

The SN75ALS130 is characterized for operation from 0°C to 70°C.

FUNCTION TABLE

INPUTS		OUTPUTS		
G†	A	Y	F̄†	W
L	X	L	H	H
X	L	L	H	H
H	H	H	H	L
H	H	S	L	H

H = high level, L = low level,

X = irrelevant, S = shorted to ground

† G and F̄ are common to the four drivers. If any of the four Y outputs is shorted, the fault flag will respond.

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PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

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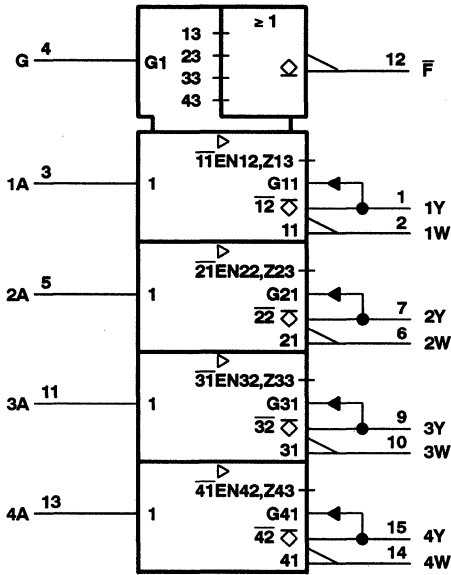
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2-303

# SN75ALS130 QUADRUPLE LINE DRIVER

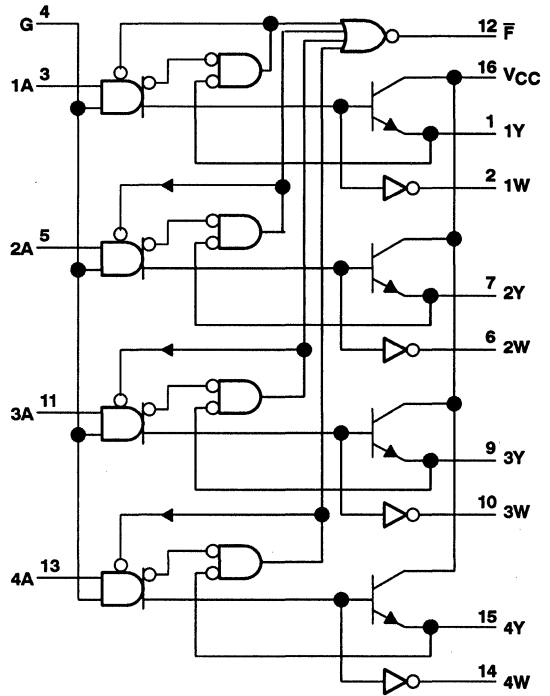
SLLS024B - D2299, FEBRUARY 1986 - REVISED FEBRUARY 1993

logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

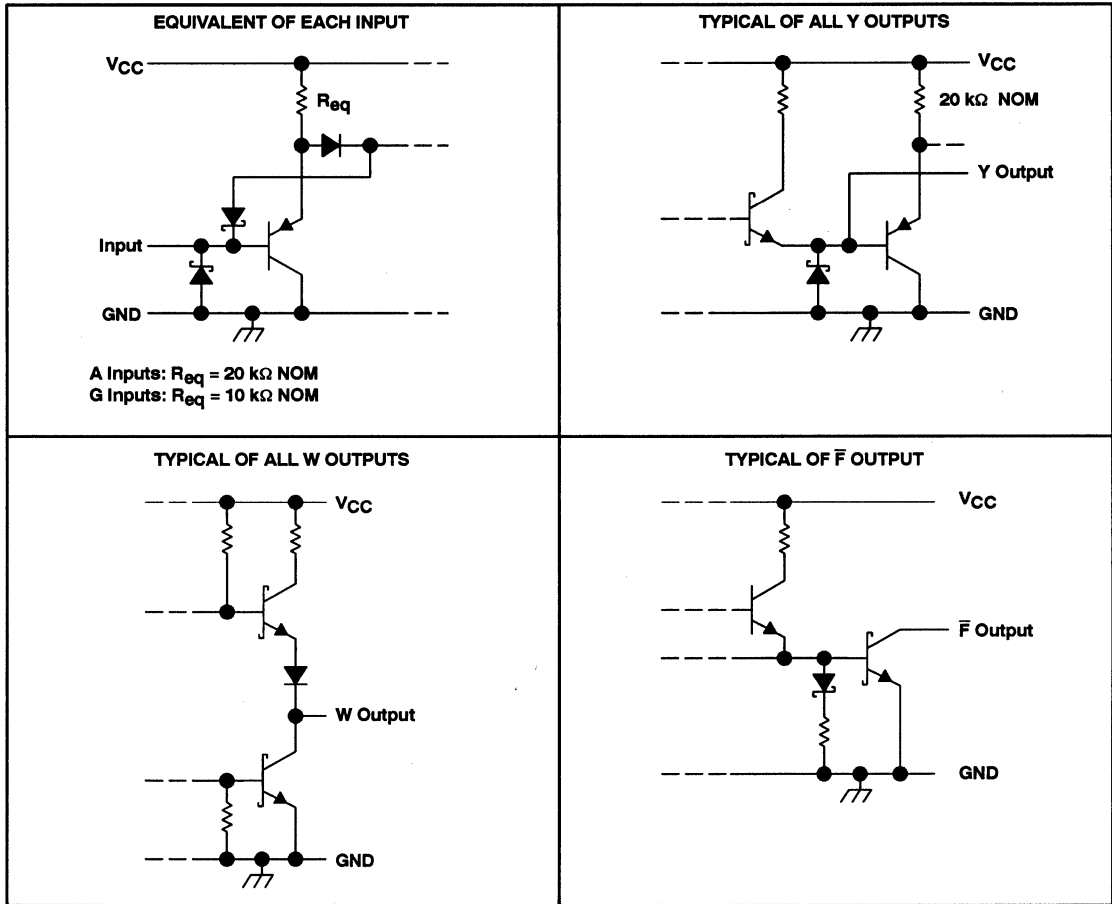
logic diagram (positive logic)



# SN75ALS130 QUADRUPLE LINE DRIVER

SLLS024B – D2299, FEBRUARY 1986 – REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$	7 V
Input voltage	7 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

TEXAS  
INSTRUMENTS

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# SN75ALS130 QUADRUPLE LINE DRIVER

SLLS024B - D2299, FEBRUARY 1986 - REVISED FEBRUARY 1993

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.95	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$	0.8			V
High-level output current, $I_{OH}$	-59.3			mA
Operating free-air temperature, $T_A$	0	70		°C

## electrical characteristics over recommended operating free-air temperature range

PARAMETER		TEST CONDITIONS			MIN	MAX	UNIT	
$V_{IK}$	Input clamp voltage	A, G	$V_{CC} = 4.5\text{ V}$ ,	$I_I = -18\text{ mA}$		-1.5	V	
$V_{OH}$	High-level output voltage	Y	$V_{CC} = 4.5\text{ V}$ ,	$I_{OH} = -59.3\text{ mA}$ , $V_{IH} = 2\text{ V}$	3.11		V	
		Y	$V_{CC} = 5.25\text{ V}$ ,	$I_{OH} = -41\text{ mA}$ , $V_{IH} = 2\text{ V}$	3.9			
		W	$V_{CC} = 4.5\text{ V}$ ,	$I_{OH} = -400\text{ }\mu\text{A}$ , $V_{IH} = 2\text{ V}$	2.5			
$V_{OL}$	Low-level output voltage	Y	$V_{CC} = 5.5\text{ V}$ ,	$I_{OL} = -240\text{ }\mu\text{A}$ , $V_{IL} = 0.8\text{ V}$		0.15	V	
		Y	$V_{CC} = 5.95\text{ V}$ ,	$I_{OL} = -1\text{ mA}$ , $V_{IL} = 0.8\text{ V}$		0.15		
		$\bar{F}$	$V_{CC} = 4.5\text{ V}$ ,	$I_{OL} = 8\text{ mA}$ , Y at 0 V		0.5		
		W	$V_{CC} = 4.5\text{ V}$ ,	$I_{OL} = 8\text{ mA}$		0.5		
$I_{O(off)}$	Off-state output current	Y	$V_{CC} = 4.5\text{ V}$ ,	$V_{IL} = 0$ , $V_O = 3.11\text{ V}$		100	$\mu\text{A}$	
		Y	$V_{CC} = 0$ ,	$V_{IL} = 0$ , $V_O = 3.11\text{ V}$		200		
$I_{OH}$	High-level output current	$\bar{F}$	$V_{CC} = 5.95\text{ V}$ ,	$V_{OH} = 5.95\text{ V}$		100	$\mu\text{A}$	
$I_I$	Input current	A	$V_{CC} = 4.5\text{ V}$ ,	$V_{IH} = 5.5\text{ V}$		100	$\mu\text{A}$	
		G			400			
$I_{IH}$	High-level input current	A	$V_{CC} = 4.5\text{ V}$ ,	$V_{IH} = 2.7\text{ V}$		20	$\mu\text{A}$	
		G			80			
$I_{IL}$	Low-level input current	A	$V_{CC} = 5.95\text{ V}$ ,	$V_{IL} = 0.4\text{ V}$		250	$\mu\text{A}$	
		G			-1000			
$I_{OS}$	Short-circuit output current	Y	$V_{CC} = 5.5\text{ V}$ ,	$V_O = 0$ , $V_{IH} = 2.7\text{ V}$		-5	mA	
		W	$V_{CC} = 5.5\text{ V}$ ,	$V_O = 0$		-15		-100
		Y	$V_{CC} = 5.95\text{ V}$ ,	$V_O = 0$ , $V_{IH} = 2.7\text{ V}$		-5		
		W	$V_{CC} = 5.95\text{ V}$ ,	$V_O = 0$		-15		-110
$I_{CCH}$	Supply current, all outputs high		$V_{CC} = 5.5\text{ V}$ ,	No load, $V_{IH} = 2.7\text{ V}$		30	mA	
			$V_{CC} = 5.95\text{ V}$ ,	No load, $V_{IH} = 2.7\text{ V}$		32		
$I_{CCL}$	Supply current, Y outputs low		$V_{CC} = 5.5\text{ V}$ ,	No load, $V_{IL} = 0.4\text{ V}$		45	mA	
			$V_{CC} = 5.95\text{ V}$ ,	No load, $V_{IL} = 0.4\text{ V}$		47		



# SN75ALS130 QUADRUPLE LINE DRIVER

SLLS024B – D2299, FEBRUARY 1986 – REVISED FEBRUARY 1993

## switching characteristics over recommended operating free-air temperature range

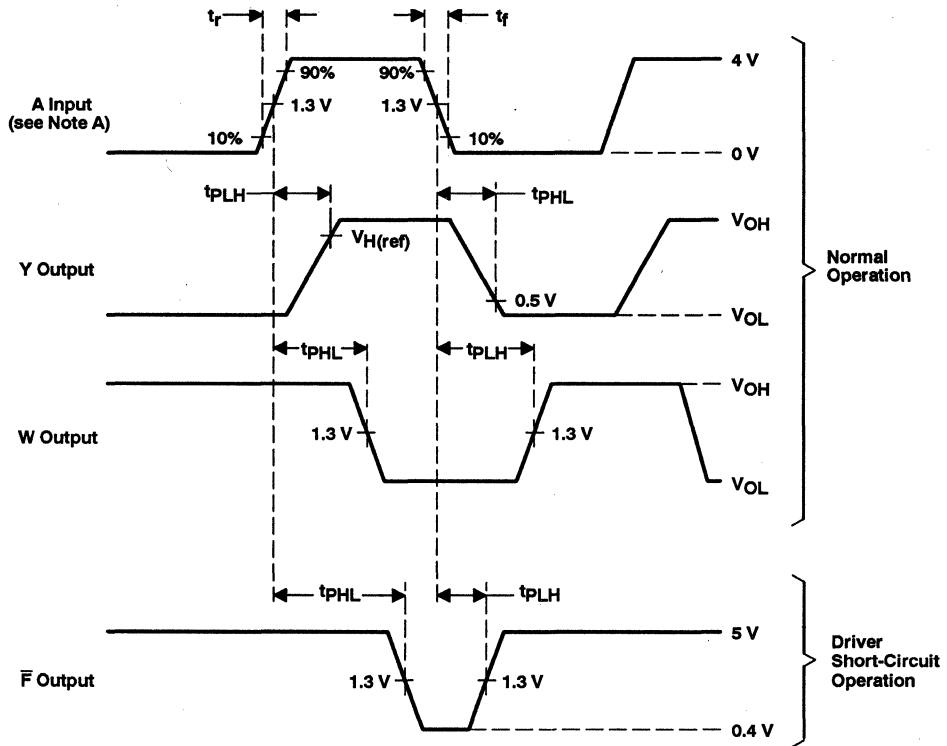
PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	A	Y	V <sub>CC</sub> = 4.5 V to 5.5 V, C <sub>L</sub> = 50 pF, Input f = 1 MHz, R <sub>L</sub> = 50 Ω, V <sub>H(ref)</sub> = 3.11 V, See Figures 1 and 2		30	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					28	ns
$\frac{t_{PLH}}{t_{PHL}}$	Ratio of propagation delay times				0.3	3	
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	A	Y	V <sub>CC</sub> = 5.25 V to 5.95 V, C <sub>L</sub> = 50 pF, Input f = 5 MHz, R <sub>L</sub> = 90 Ω, V <sub>H(ref)</sub> = 3.9 V, See Figures 1 and 2		34	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					34	ns
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	A	W	V <sub>CC</sub> = 5 V, C <sub>L</sub> = 15 pF, R <sub>L</sub> = 2 kΩ, See Figures 1 and 2		34	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					21	ns
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	A	$\bar{F}$	V <sub>CC</sub> = 5 V, C <sub>L</sub> = 15 pF, R <sub>L</sub> = 2 kΩ, See Figures 1 and 2		45	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					75	ns



# SN75ALS130 QUADRUPLE LINE DRIVER

SLLS024B – D2299, FEBRUARY 1986 – REVISED FEBRUARY 1993

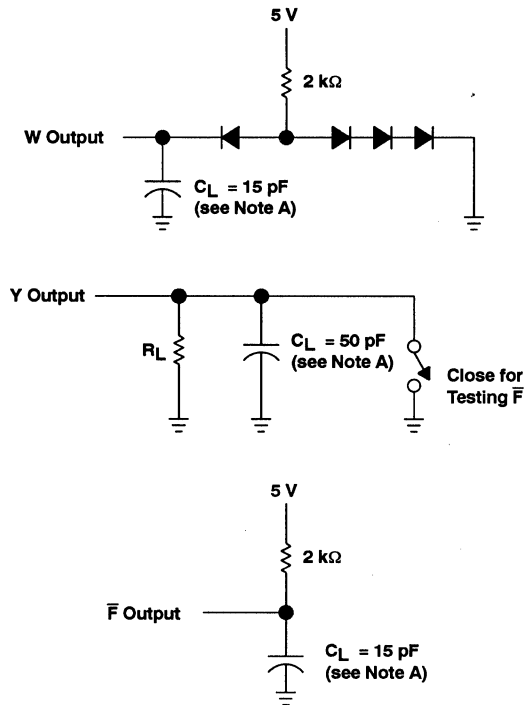
## PARAMETER MEASUREMENT INFORMATION



NOTE A: The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle  $\leq$  50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O \sim 50 \Omega$ .

Figure 1. Input and Output Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION



NOTE A:  $C_L$  includes probe and stray capacitance.

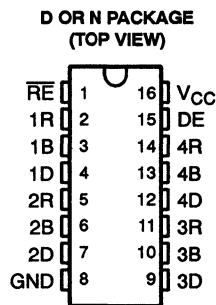
Figure 2. Switching Characteristics Load Circuits



# SN75136 QUAD BUS TRANSCEIVER WITH 3-STATE OUTPUT

SLLS078A - D2291, JANUARY 1977 - REVISED FEBRUARY 1993

- PNP Inputs for Minimal Input Loading (200  $\mu$ A Maximum)
- High-Speed Schottky Circuitry
- 3-State Outputs for Driver and Receiver
- Party-Line (Data-Bus) Operation
- Single 5-V Supply
- Driver Has 40-mA Current Sink Capability
- Designed to Be Functionally Interchangeable With Signetics N8T26, Also Called 8T26



## description

The SN75136 is a quad transceiver utilizing Schottky-diode-clamped transistors. Both the driver and receiver have 3-state outputs. With pnp inputs, the input loading is reduced to a maximum input current of 200  $\mu$ A.

The SN75136 is characterized for operation from 0°C to 70°C.

### Function Tables

#### DRIVER

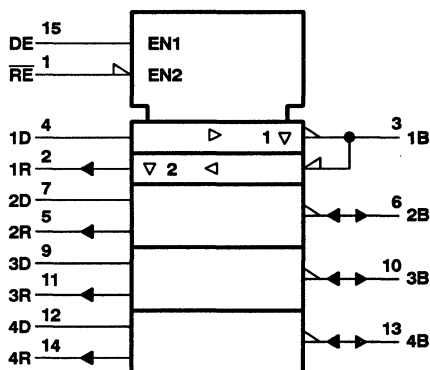
INPUTS		OUTPUT
D	DE	B
L	H	H
H	H	L
X	L	Z

#### RECEIVER

INPUTS		OUTPUT
B	RE	R
L	L	H
H	L	L
X	H	Z

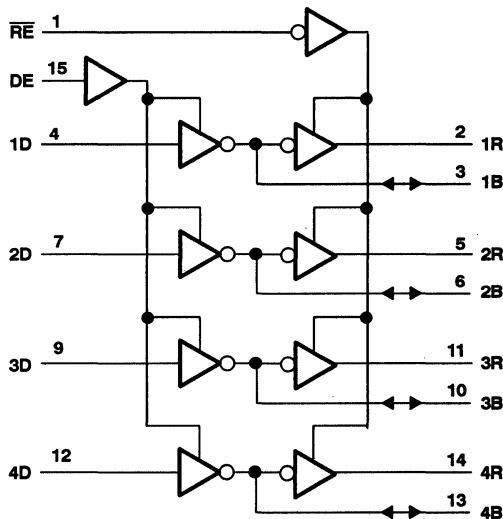
H = high level  
L = low level  
X = irrelevant  
Z = high impedance

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



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**TEXAS  
INSTRUMENTS**

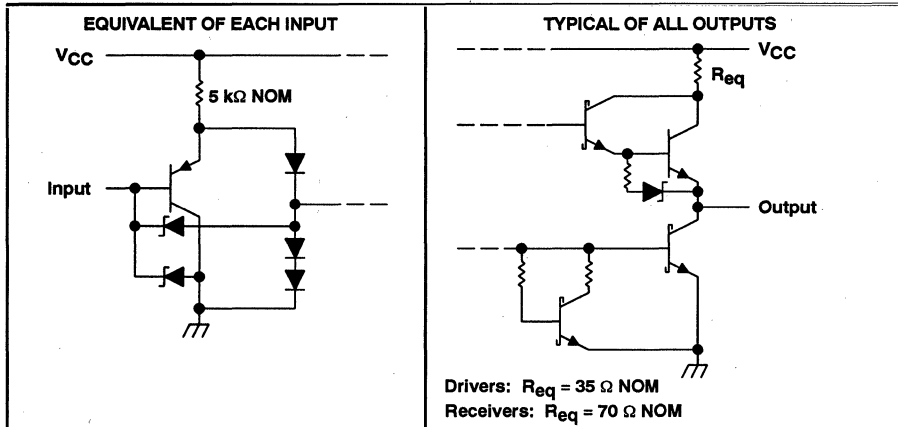
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**SN75136**  
**QUAD BUS TRANSCEIVER**  
**WITH 3-STATE OUTPUT**

SLLS078A - D2291, JANUARY 1977 - REVISED FEBRUARY 1993

**schematics of inputs and outputs**



**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	B, D, DE, $\overline{RE}$	2			V
Low-level input voltage, $V_{IL}$	B, D, DE, $\overline{RE}$			0.85	V
High-level output current, $I_{OH}$	Driver, B			-10	mA
	Receiver, R			-2	
Low-level output current, $I_{OL}$	Driver, B			40	mA
	Receiver, R			16	
Operating free-air temperature, $T_A$		0		70	°C



# SN75136 QUAD BUS TRANSCEIVER WITH 3-STATE OUTPUT

SLLS078A - D2291, JANUARY 1977 - REVISED FEBRUARY 1993

**electrical characteristics over recommended operating free-air temperature and supply voltage range (unless otherwise noted)**

PARAMETER		TEST CONDITIONS			MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	B, D, DE, $\overline{RE}$	$I_I = -5 \text{ mA}$				-1	V
$V_{OH}$	High-level output voltage	B	$V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.85 \text{ V}$ , $I_{OH} = -10 \text{ mA}$	2.6	3.1		V	
		R	$V_{IL} = 0.85 \text{ V}$ , $I_{OH} = -2 \text{ mA}$	2.6	3.1			
$V_{OL}$	Low-level output voltage	B	$V_{IH} = 2 \text{ V}$ , $I_{OL} = 40 \text{ mA}$			0.5	V	
		R	$V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.85 \text{ V}$ , $I_{OL} = 16 \text{ mA}$			0.5		
$I_{OZ}$	Off-state (high-impedance state) output current	B, R	DE at 0.85 V, $\overline{RE}$ at 2 V, $V_O = 2.6 \text{ V}$			100	$\mu\text{A}$	
		R	$\overline{RE}$ at 2 V, $V_O = 0.5 \text{ V}$			-100		
$I_{IH}$	High-level input current	D, DE, $\overline{RE}$	$V_I = 5.25 \text{ V}$				25	$\mu\text{A}$
$I_{IL}$	Low-level input current	B, D, DE, $\overline{RE}$	$V_I = 0.4 \text{ V}$				-200	$\mu\text{A}$
$I_{OS}$	Short-circuit output current‡	B	$V_{CC} = 5.25 \text{ V}$	-50		-150	mA	
		R		-30		-75		
$I_{CC}$	Supply current		$V_{CC} = 5.25 \text{ V}$ , No load				87	mA

† All typical values are at  $T_A = 25^\circ\text{C}$  and  $V_{CC} = 5 \text{ V}$ .

‡ Only one output should be shorted to ground at a time, and duration of the short circuit should not exceed one second.

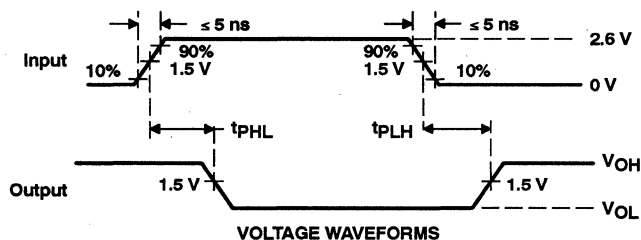
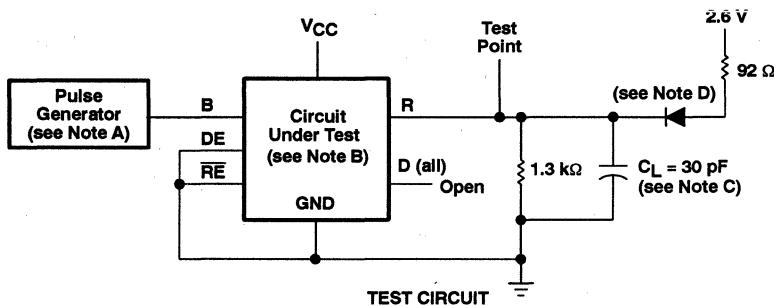
**switching characteristics,  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	B	R	$C_L = 30 \text{ pF}$ , See Figure 1		8	18	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output					7	14	
$t_{PLH}$	Propagation delay time, low-to-high-level output	D	B	$C_L = 300 \text{ pF}$ , See Figure 2		11	20	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output					16	24	
$t_{PLZ}$	Output disable time from low level	$\overline{RE}$	R	$C_L = 30 \text{ pF}$ , See Figure 3		16	24	ns
$t_{PZL}$	Output enable time to low level					15	30	
$t_{PLZ}$	Output disable time from low level	DE	B	$C_L = 300 \text{ pF}$ , See Figure 4		9	24	ns
$t_{PZL}$	Output enable time to low level					31	38	

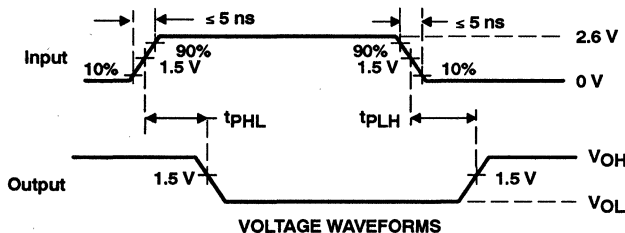
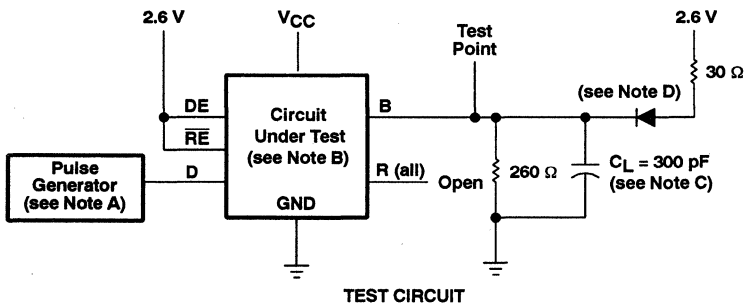
**SN75136**  
**QUAD BUS TRANSCEIVER**  
**WITH 3-STATE OUTPUT**

SLLS078A - D2291, JANUARY 1977 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



**Figure 1. Propagation Delay Times From Bus to Receiver Output**



**Figure 2. Propagation Delay Times From Driver Input to Bus**

- NOTES: A. The pulse generator in Figures 1 and 2 has the following characteristics: PRR  $\leq$  10 MHz, duty cycle = 50%,  $Z_0 \sim 50 \Omega$ .  
 B. All inputs and outputs not shown are open.  
 C.  $C_L$  includes probe and jig capacitance.  
 D. All diodes are 1N916 or 1N3064.



PARAMETER MEASUREMENT INFORMATION

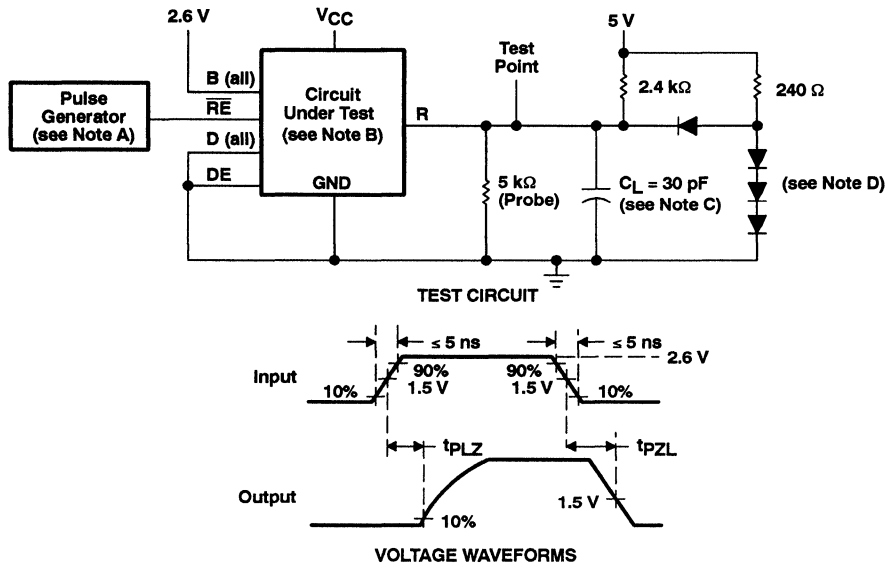


Figure 3. Receiver Test Circuit and Voltage Waveforms

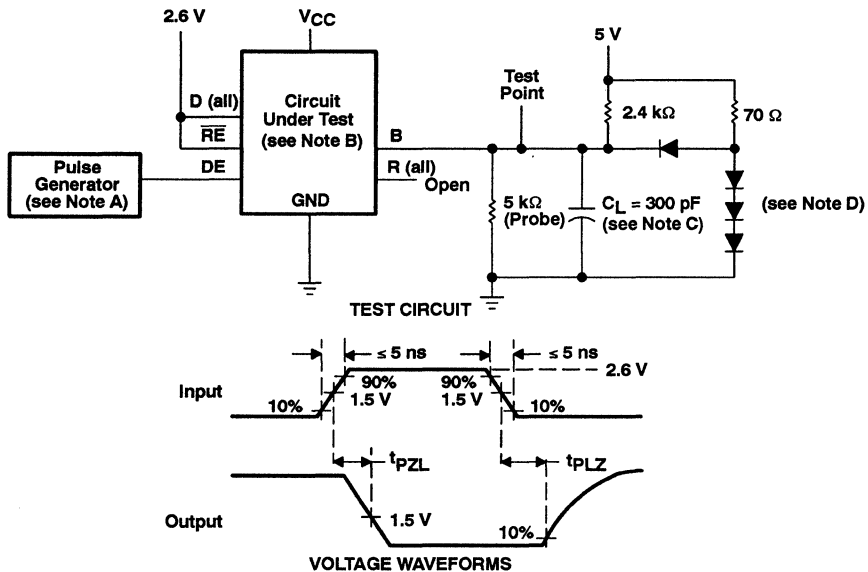


Figure 4. Driver Test Circuit and Voltage Waveforms

- NOTES: A. The pulse generator in Figures 3 and 4 has the following characteristics: PRR  $\leq$  5 MHz, duty cycle = 50%,  $Z_0 \sim 50 \Omega$ .  
 B. All inputs and outputs now shown are open.  
 C.  $C_L$  includes probe and jig capacitance.  
 D. All diodes are 1N916 or 1N3064.





# SN55138, SN75138 QUAD BUS TRANSCEIVERS

SLLS079A - D1663, SEPTEMBER 1973 - REVISED FEBRUARY 1993

- Single 5-V Supply
- High-Input-Impedance, High-Threshold Receivers
- Common Driver Strobe
- TTL-Compatible Driver and Strobe Inputs With Clamp Diodes
- High-Speed Operation
- 100-mA Open-Collector Driver Outputs
- Four Independent Channels
- TTL-Compatible Receiver Output

## description

The SN55138 and SN75138 quad bus transceivers are designed for two-way data communication over single-ended transmission lines. Each of the four identical channels consists of a driver with TTL inputs and a receiver with a TTL output. The driver open-collector output is designed to handle loads of up to 100-mA open collector. The receiver input is internally connected to the driver output, and has a high impedance to minimize loading of the transmission line. Because of the high driver-output current and the high receiver-input impedance, a very large number (typically hundreds) of transceivers may be connected to a single data bus.

The receiver design also features a threshold of 2.3 V (typical), providing a wider noise margin than would be possible with a receiver having the usual TTL threshold. A strobe turns off all drivers (high impedance) but does not affect receiver operation. These circuits are designed for operation from a single 5-V supply and include a provision to minimize loading of the data bus when the power-supply voltage is zero.

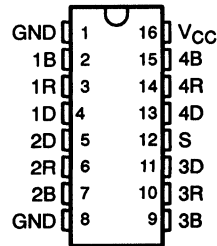
The SN55138 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75138 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## Function Tables

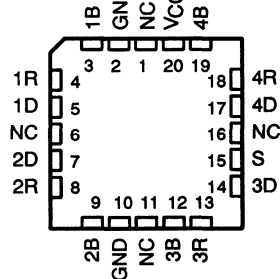
TRANSMITTING				RECEIVING			
INPUTS		OUTPUTS		INPUTS			OUTPUT
S	D	B	R	S	B	D	R
L	H	L	H	H	H	X	L
L	L	H	L	H	L	X	H

H = high level, L = low level, X = irrelevant

SN55138 ... J OR W PACKAGE  
SN75138 ... D OR N PACKAGE  
(TOP VIEW)



SN55138 ... FK PACKAGE  
(TOP VIEW)

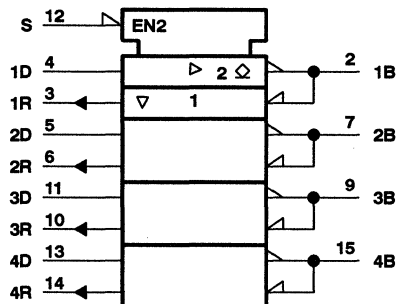


NC - No internal connection

# SN55138, SN75138 QUAD BUS TRANSCEIVERS

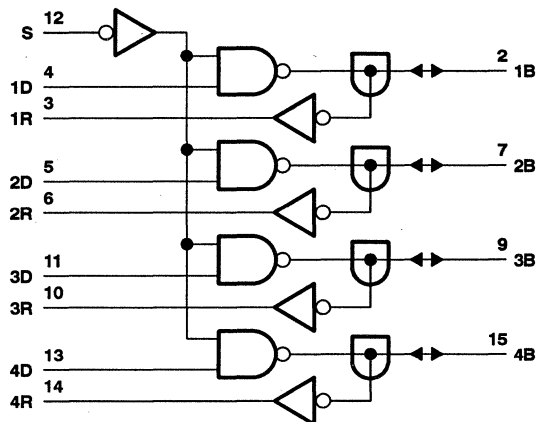
SLLS079A - D1663, SEPTEMBER 1973 - REVISED FEBRUARY 1983

## logic symbol†

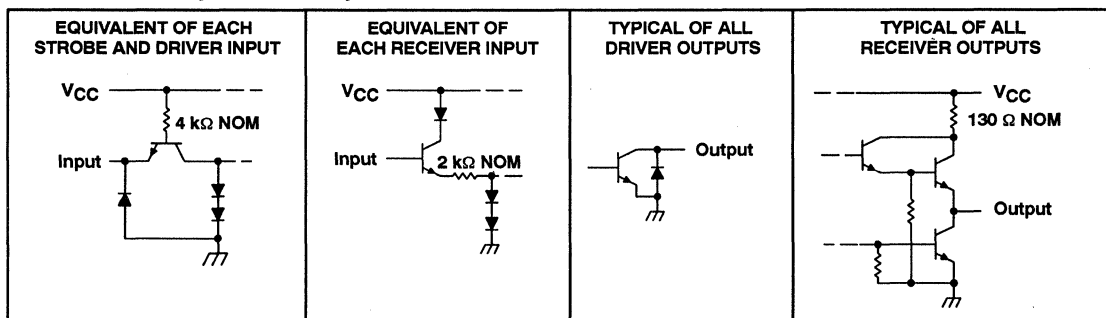


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12. Pin numbers shown are for D, J, N, and W packages.

## logic diagram (positive logic)



## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55138	SN75138	UNIT
Supply voltage, $V_{CC}$ (see Note 1)	7	7	V
Input voltage	5.5	5.5	V
Driver off-state output voltage	7	7	V
Low-level output current into the driver output	150	150	mA
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 125	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or W package	260		
Case temperature for 60 seconds: FK package	260	°C	
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300	°C	

NOTE 1: All voltage values are with respect to both ground terminals connected together.

# SN55138, SN75138 QUAD BUS TRANSCEIVERS

SLLS079A - D1663, SEPTEMBER 1973 - REVISED FEBRUARY 1993

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	—
FK†	1375 mW	11.0 mW/°C	880 mW	275 mW
J†	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—
W	1000 mW	8.0 mW/°C	640 mW	200 mW

† In the FK and J packages, the SN55138 chip is alloy mounted.

## recommended operating conditions

		SN55138			SN75138			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$		4.5		5.5	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	Driver or strobe	2			2			V
	Receiver	3.2			2.9			
Low-level input voltage, $V_{IL}$	Driver or strobe				0.8			V
	Receiver				1.5			
High-level output current, $I_{OH}$	Receiver output				-400			$\mu\text{A}$
Low-level output current, $I_{OL}$	Driver output				100			mA
	Receiver output				16			
Operating free-air temperature, $T_A$		-55		125	0		70	°C



# SN55138, SN75138 QUAD BUS TRANSCEIVERS

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electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER			TEST CONDITIONS†		SN55138			SN75138			UNIT
					MIN	TYP‡	MAX	MIN	TYP‡	MAX	
V <sub>IK</sub>	Input clamp voltage	Driver or strobe	V <sub>CC</sub> = MIN,	I <sub>I</sub> = -12 mA			-1.5			-1.5	V
V <sub>OH</sub>	High-level output voltage	Receiver	V <sub>CC</sub> = MIN, V <sub>IL(R)</sub> = V <sub>IL</sub> max,	V <sub>IH(S)</sub> = 2 V, I <sub>OH</sub> = -400 µA	2.4	3.5		2.4	3.5		V
V <sub>OL</sub>	Low-level output voltage	Driver	V <sub>CC</sub> = MIN, V <sub>IL(S)</sub> = 0.8 V,	V <sub>IH(D)</sub> = 2 V, I <sub>OL</sub> = 100 mA			0.45			0.45	V
		Receiver	V <sub>CC</sub> = MIN, V <sub>IH(S)</sub> = 2 V,	V <sub>IH(R)</sub> = V <sub>IH</sub> min, I <sub>OL</sub> = 16 mA			0.4			0.4	
I <sub>I</sub>	Input current at maximum input voltage	Driver or strobe	V <sub>CC</sub> = MAX,	V <sub>I</sub> = V <sub>CC</sub>			1			1	mA
I <sub>IH</sub>	High-level input current	Driver or strobe	V <sub>CC</sub> = MAX,	V <sub>I</sub> = 2.4 V			40			40	µA
		Receiver	V <sub>CC</sub> = 5 V, V <sub>I(S)</sub> = 2 V	V <sub>I(R)</sub> = 4.5 V,		25	300		25	300	
I <sub>IL</sub>	Low-level input current	Driver or strobe	V <sub>CC</sub> = MAX,	V <sub>I</sub> = 0.4 V		-1	-1.6		-1	-1.6	mA
		Receiver	V <sub>CC</sub> = MAX, V <sub>I(S)</sub> = 2 V	V <sub>I(R)</sub> = 0.45 V,			-50			-50	µA
	Input current with power off	Receiver	V <sub>CC</sub> = 0,	V <sub>I</sub> = 4.5 V		1.1	1.5		1.1	1.5	mA
I <sub>OS</sub>	Short-circuit output current§	Receiver	V <sub>CC</sub> = MAX			-20	-55		-18	-55	mA
I <sub>CC</sub>	Supply current	All driver outputs low	V <sub>CC</sub> = MAX, V <sub>I(S)</sub> = 0.8 V	V <sub>I(D)</sub> = 2 V,		50	65		50	65	mA
		All driver outputs high	V <sub>CC</sub> = MAX, V <sub>I(S)</sub> = 2 V, Receiver outputs open	V <sub>I(R)</sub> = 3.5 V,		42	55		42	55	

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions. Parenthetical letters D, R, and S used with V<sub>I</sub> refer to the driver input, receiver input, and strobe input, respectively.

‡ All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

§ Not more than one output should be shorted at a time.

## switching characteristics, V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C

PARAMETER <sup>¶</sup>	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
t <sub>PLH</sub>	Driver	Driver	C <sub>L</sub> = 50 pF, See Figure 1	R <sub>L</sub> = 50 Ω,		15	24	ns	
t <sub>PHL</sub>						14	24		
t <sub>PLH</sub>	Strobe	Driver					18	28	ns
t <sub>PHL</sub>							22	32	
t <sub>PLH</sub>	Receiver	Receiver	C <sub>L</sub> = 15 pF, See Figure 2	R <sub>L</sub> = 400 Ω,		7	15	ns	
t <sub>PHL</sub>							8		15

¶ t<sub>PLH</sub> = propagation delay time, low-to-high-level output

t<sub>PHL</sub> = propagation delay time, high-to-low-level output

TEXAS  
INSTRUMENTS

PARAMETER MEASUREMENT INFORMATION

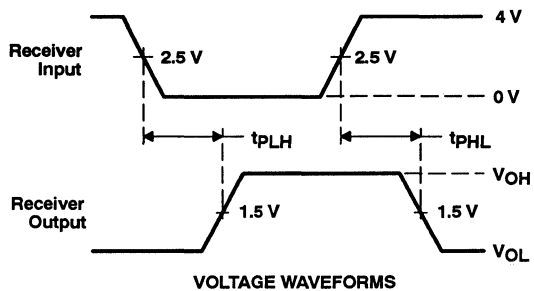
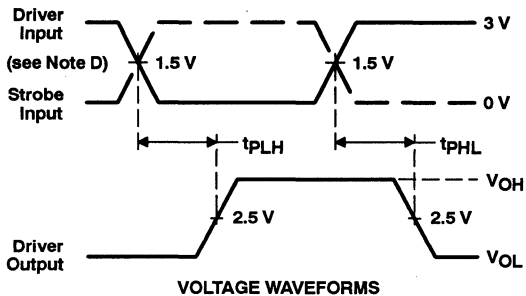
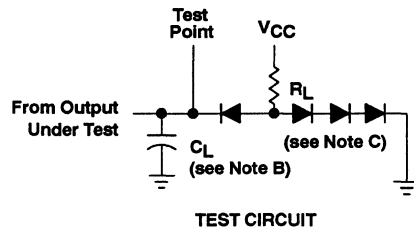
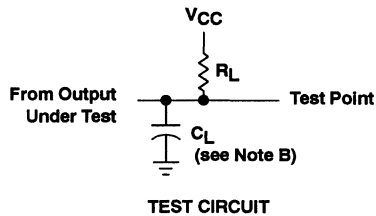


Figure 1. Propagation Delay Times From Data and Strobe Inputs

Figure 2. Propagation Delay Times From Receiver Input

- NOTES: A. Input pulses are supplied by generators having the following characteristics:  $t_w = 100$  ns,  $PRR \leq 1$  MHz,  $t_r \leq 10$  ns,  $t_f \leq 10$  ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N916 or 1N3064.  
 D. When testing driver input (solid line) strobe must be low; when testing strobe input (dashed line) driver input must be high.

# SN55138, SN75138 QUAD BUS TRANSCEIVERS

SLLS079A - D1663, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS†

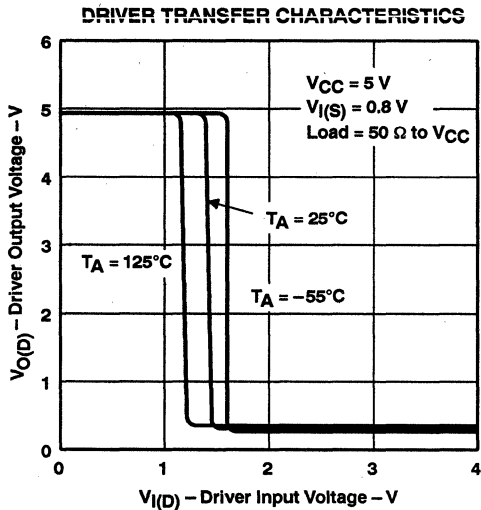


Figure 3

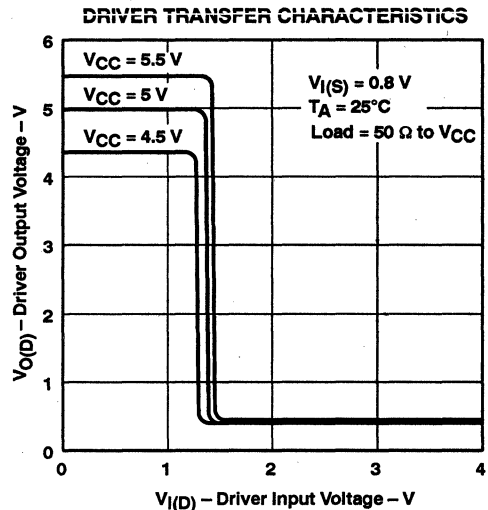


Figure 4

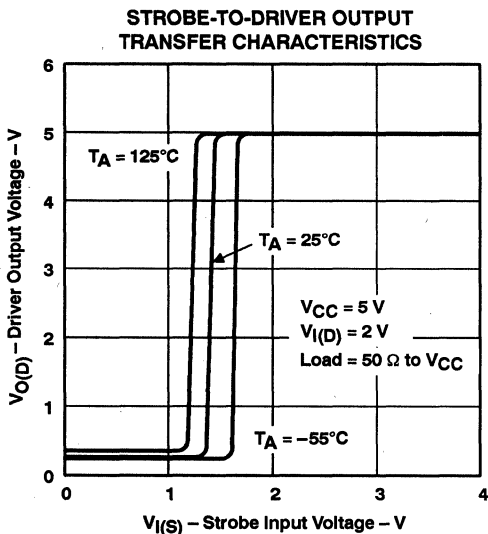


Figure 5

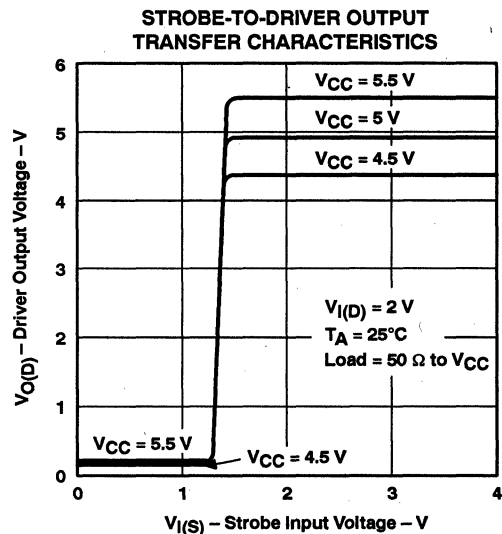


Figure 6

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  is applicable to SN55138 circuits only.

TYPICAL CHARACTERISTICS†

RECEIVER TRANSFER CHARACTERISTICS

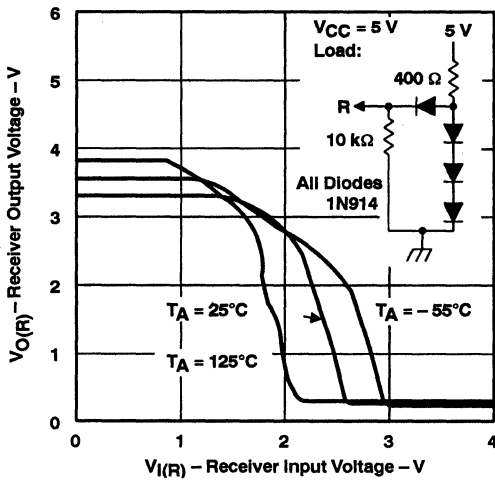


Figure 7

RECEIVER TRANSFER CHARACTERISTICS

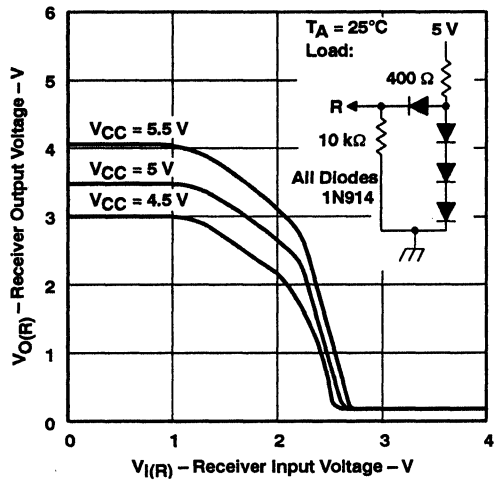


Figure 8

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT (RECEIVER)

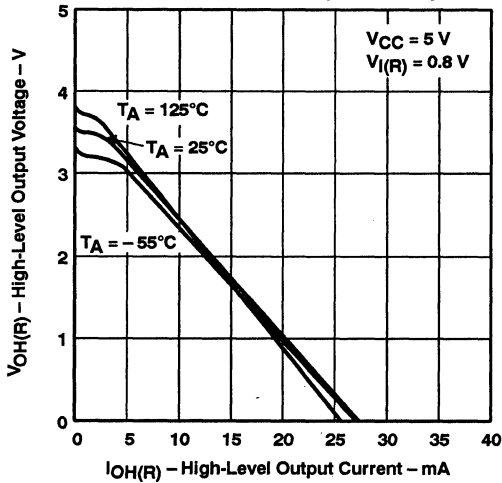


Figure 9

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT (RECEIVER)

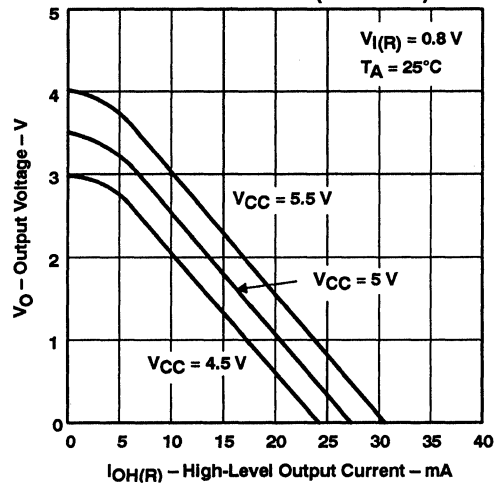


Figure 10

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  is applicable to SN55138 circuits only.



# SN55138, SN75138 QUAD BUS TRANSCEIVERS

SLLS079A - D1663, SEPTEMBER 1973 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS†

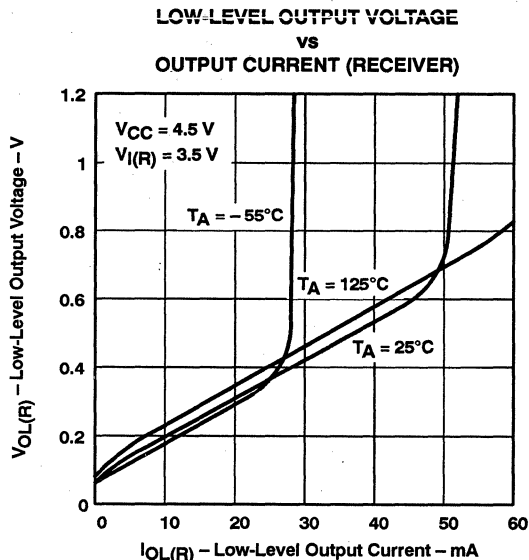


Figure 11

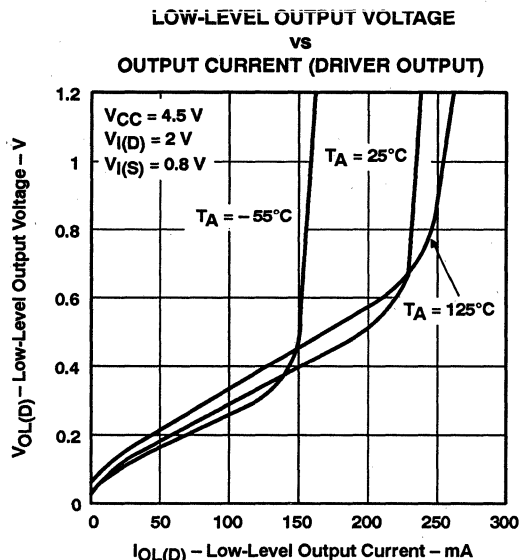


Figure 12

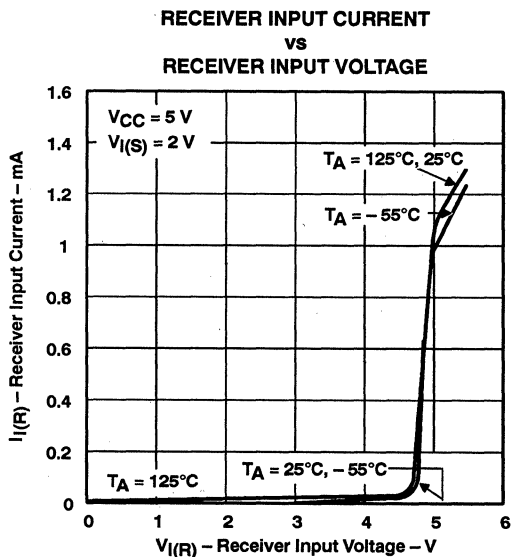


Figure 13

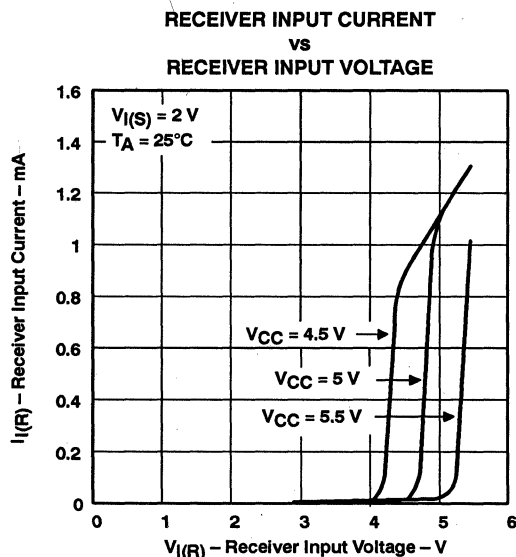


Figure 14

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  is applicable to SN55138 circuits only.

TYPICAL CHARACTERISTICS†

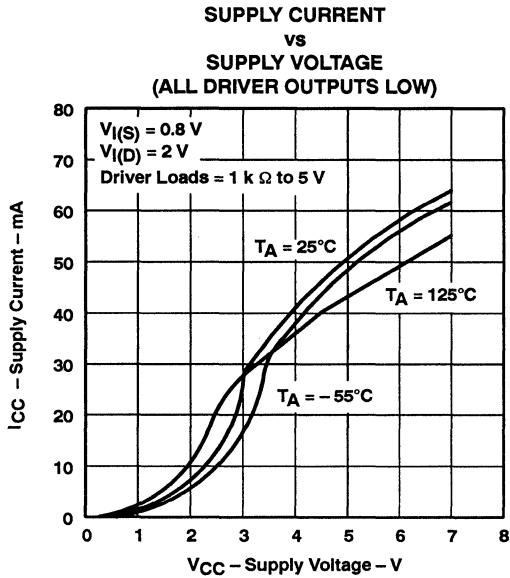


Figure 15

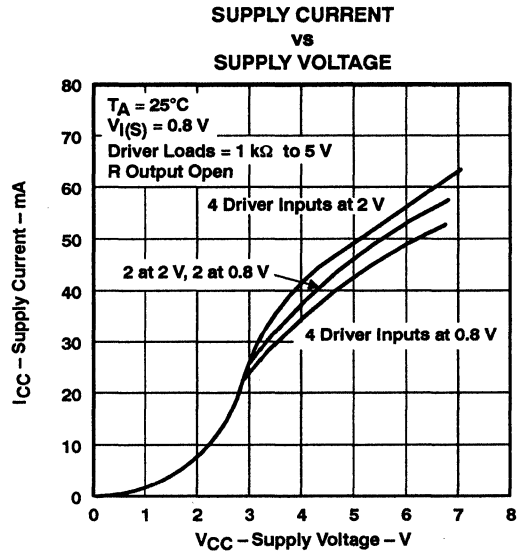


Figure 16

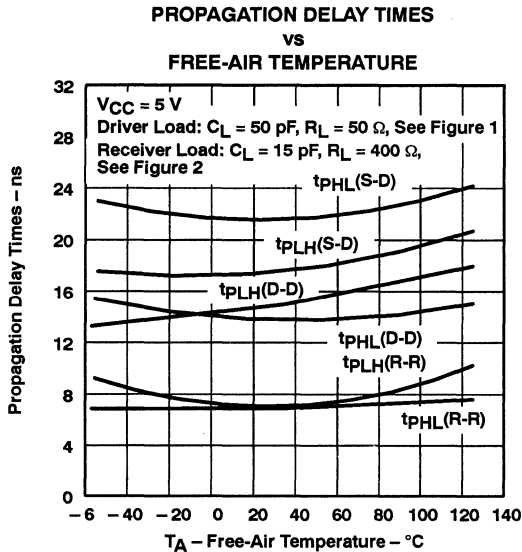


Figure 17

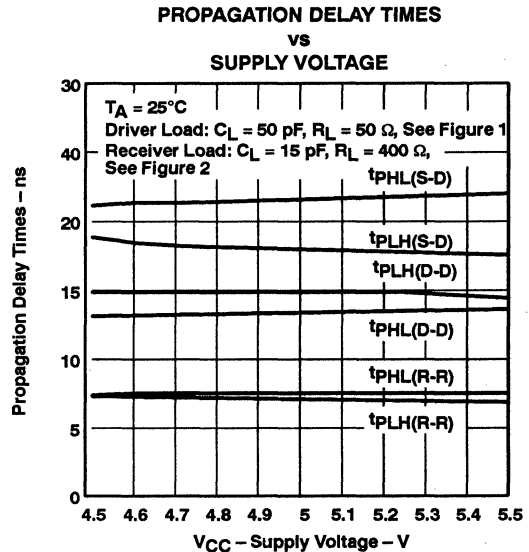


Figure 18

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  is applicable to SN55138 circuits only.

# SN55138, SN75138 QUADRUPLE BUS TRANSCEIVERS

SLLS079 – D1663, SEPTEMBER 1973 – REVISED SEPTEMBER 1986

## TYPICAL CHARACTERISTICS

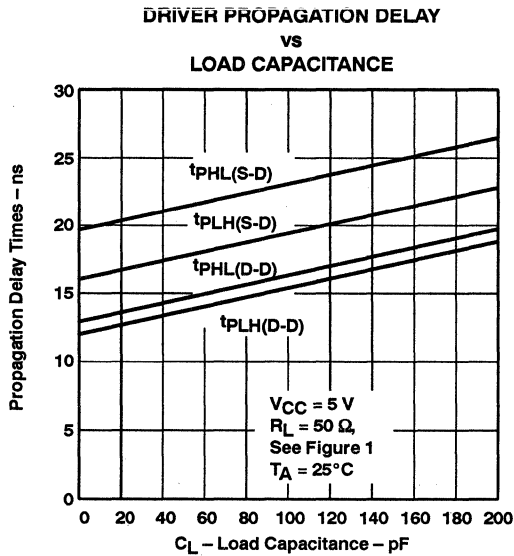


Figure 19

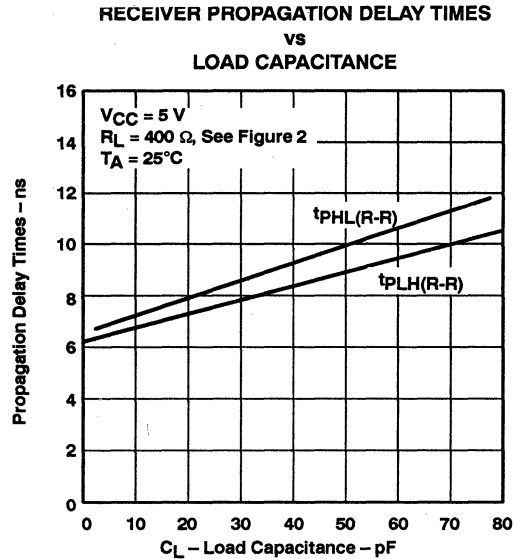


Figure 20

## APPLICATION INFORMATION

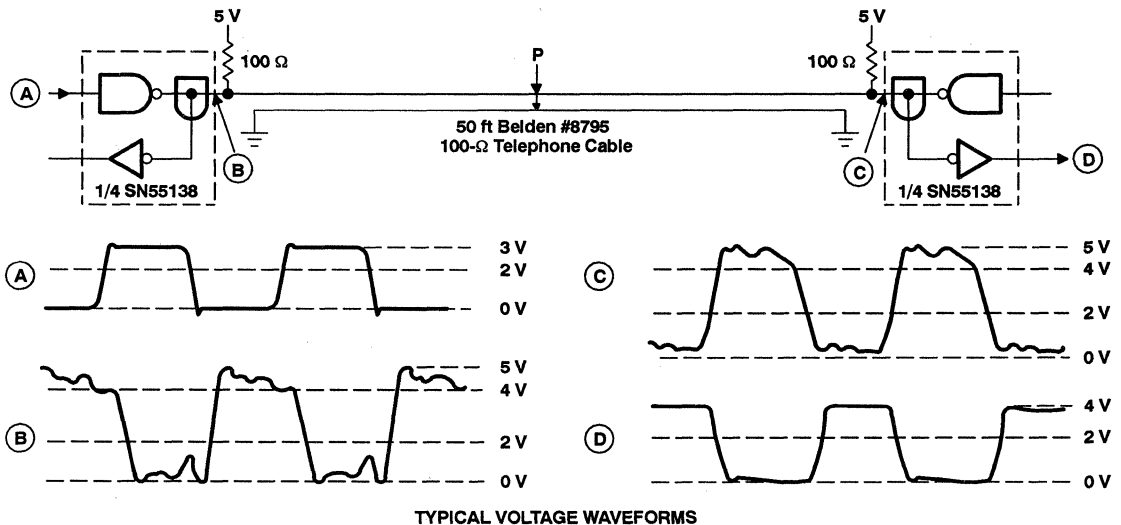
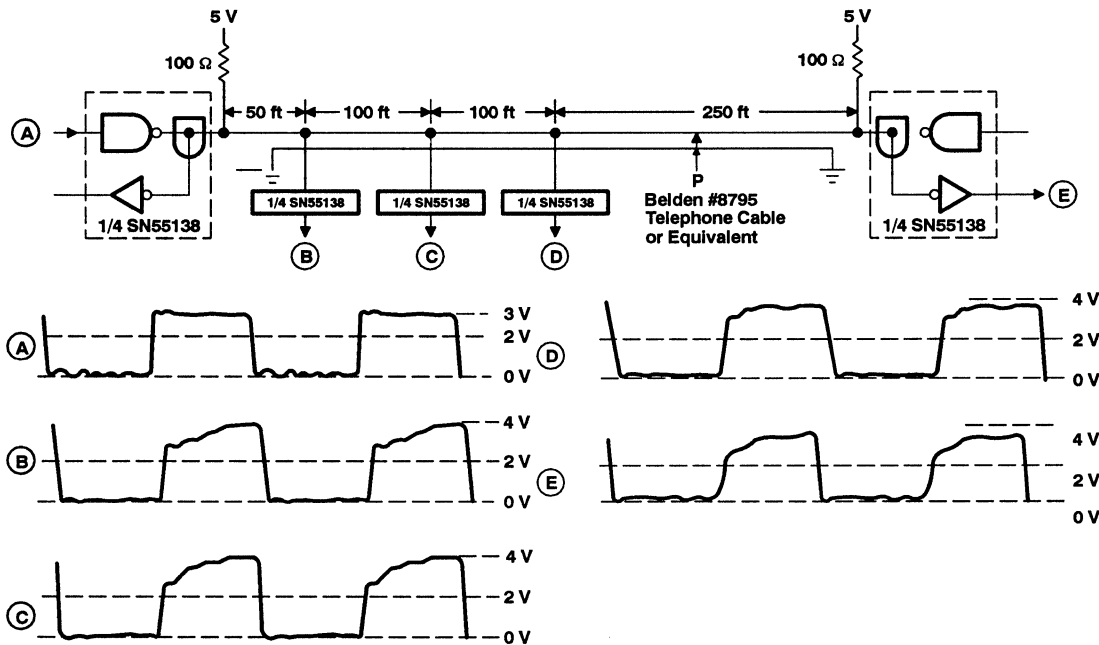


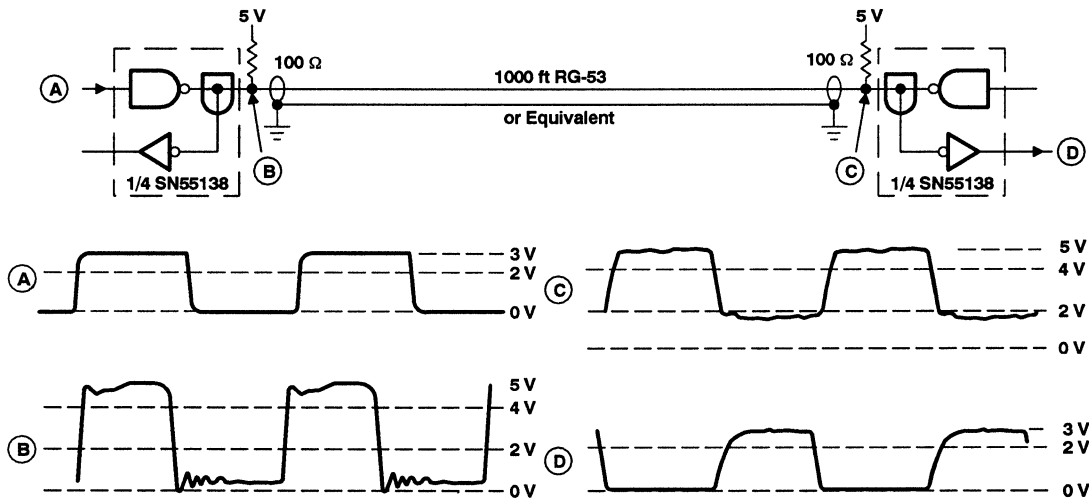
Figure 21. Point-to-Point Communication Over 50 Feet of Twisted Pair at 5 MHz

APPLICATION INFORMATION



TYPICAL VOLTAGE WAVEFORMS

Figure 22. Party-Line Communication on 500 Feet of Twisted Pair at 1 MHz



TYPICAL VOLTAGE WAVEFORMS

Figure 23. Point-to-Point Communication Over 1000 Feet of Coax at 1 MHz



# SN75140, SN75141 DUAL LINE RECEIVERS

SLLS080A – D2155, JANUARY 1977 – REVISED FEBRUARY 1993

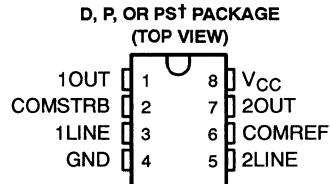
- **Single 5-V Supply**
- **±100-mV Sensitivity**
- **For Application as:**  
**Single-Ended Line Receiver**  
**Gated Oscillator**  
**Level Comparator**
- **Adjustable Reference Voltage**
- **TTL Outputs**
- **TTL-Compatible Strobe**
- **Designed for Party-Line (Data-Bus) Applications**
- **Common Reference Pin**
- **Common Strobe**
- **SN75141 Has Diode-Protected Input Stage for Power-Off Condition**

## description

Each of these devices consists of a dual single-ended line receiver with TTL-compatible strobes and outputs. The reference voltage (switching threshold) is applied externally and can be adjusted from 1.5 V to 3.5 V, making it possible to optimize noise immunity for a given system design. Due to their low input current (less than 100  $\mu$ A), they are ideally suited for party-line (bus-organized) systems.

The SN75140 has a common reference voltage pin and a common strobe. The SN75141 is the same as the SN75140 except that the input stage is diode protected.

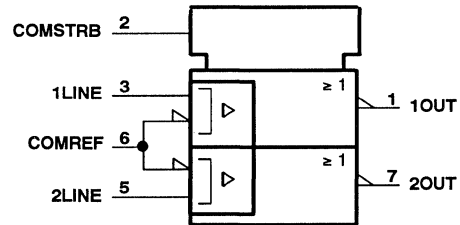
The SN75140 and SN75141 are characterized for operation from 0°C to 70°C.



† The PS package is only available left-ended taped and reeled (order SN75140 PSLE).

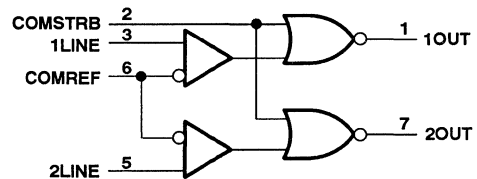
## SN75141 NOT RECOMMENDED FOR NEW DESIGN

### logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

### logic diagram (positive logic)



FUNCTION TABLE  
(each receiver)

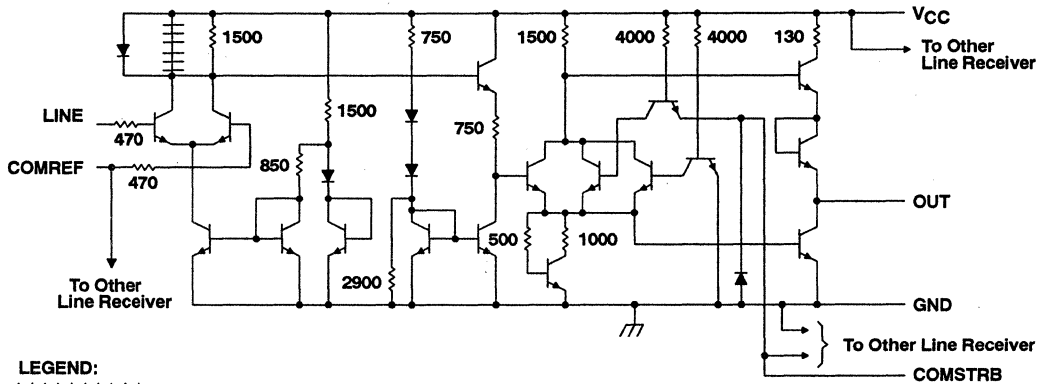
LINE INPUT	STROBE	OUTPUT
$\leq V_{ref} - 100 \text{ mV}$	L	H
$\geq V_{ref} + 100 \text{ mV}$	X	L
X	H	L

H = high level, L = low level, X = irrelevant

# SN75140, SN75141 DUAL LINE RECEIVERS

SLLS080A – D2155, JANUARY 1977 – REVISED FEBRUARY 1993

## schematic (each receiver)



**LEGEND:**

SN75140 device only

Resistor values shown are nominal and in ohms.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Reference input voltage, $V_{ref}$	5.5 V
Line input voltage range with respect to GND	-2 V to 5.5 V
Line input voltage with respect to $V_{ref}$	$\pm 5$ V
Strobe input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Unless otherwise specified, voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW
PS	450 mW	3.6 mW/°C	288 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	5.5	V
Reference input voltage, $V_{ref}$	1.5		3.5	V
High-level line input voltage, $V_{IH(L)}$	$V_{ref} + 0.1$		$V_{CC} - 1$	V
Low-level line input voltage, $V_{IL(L)}$	0		$V_{ref} - 0.1$	V
High-level strobe input voltage, $V_{IH(S)}$	2		5.5	V
Low-level strobe input voltage, $V_{IL(S)}$	0		0.8	V



# SN75140, SN75141 DUAL LINE RECEIVERS

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**electrical characteristics over recommended operating free-air temperature range,  $V_{CC} = 5\text{ V} \pm 10\%$ ,  $V_{ref} = 1.5\text{ V}$  to  $3.5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$V_{IK}$	Strobe input clamp voltage	$I_{I(S)} = -12\text{ mA}$			-1.5	V	
$V_{OH}$	High-level output voltage	$V_{IL(L)} = V_{ref} - 100\text{ mV}$ , $V_{IL(S)} = 0.8\text{ V}$ , $I_{OH} = -400\text{ }\mu\text{A}$	2.4			V	
$V_{OL}$	Low-level output voltage	$V_{IH(L)} = V_{ref} + 100\text{ mV}$ , $V_{IL(S)} = 0.8\text{ V}$ , $I_{OL} = 16\text{ mA}$			0.4	V	
		$V_{IL(L)} = V_{ref} - 100\text{ mV}$ , $V_{IH(S)} = 2\text{ V}$ , $I_{OL} = 16\text{ mA}$			0.4		
$I_{I(S)}$	Strobe input current at maximum input voltage	Strobe			1	mA	
		COMSTRB	$V_I(S) = 5.5\text{ V}$		2		
$I_{IH}$	High-level input current	Strobe			40	$\mu\text{A}$	
		COMSTRB	$V_I(S) = 2.4\text{ V}$		80		
		LINE	$V_{I(L)} = 3.5\text{ V}$ , $V_{ref} = 1.5\text{ V}$	35	100		
		Reference		35	100		
		COMREF	$V_{I(L)} = 0$ , $V_{ref} = 3.5\text{ V}$	70	200		
$I_{IL}$	Low-level input current	Strobe			-1.6	mA	
		COMSTRB	$V_I(S) = 0.4\text{ V}$		-3.2		
		LINE	$V_{I(L)} = 0$ , $V_{ref} = 1.5\text{ V}$			-10	$\mu\text{A}$
		Reference				-10	
		COMREF	$V_{I(L)} = 1.5\text{ V}$ , $V_{ref} = 0$			-20	
$I_{OS}$	Short-circuit output current‡	$V_{CC} = 5.5\text{ V}$	-18		-55	mA	
$I_{CCH}$	Supply current, output high	$V_I(S) = 0$ , $V_{I(L)} = V_{ref} - 100\text{ mV}$		18	30	mA	
$I_{CCL}$	Supply current, output low	$V_I(S) = 0$ , $V_{I(L)} = V_{ref} + 100\text{ mV}$		20	35	mA	

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ Only one output should be shorted at a time.

**switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $V_{ref} = 2.5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

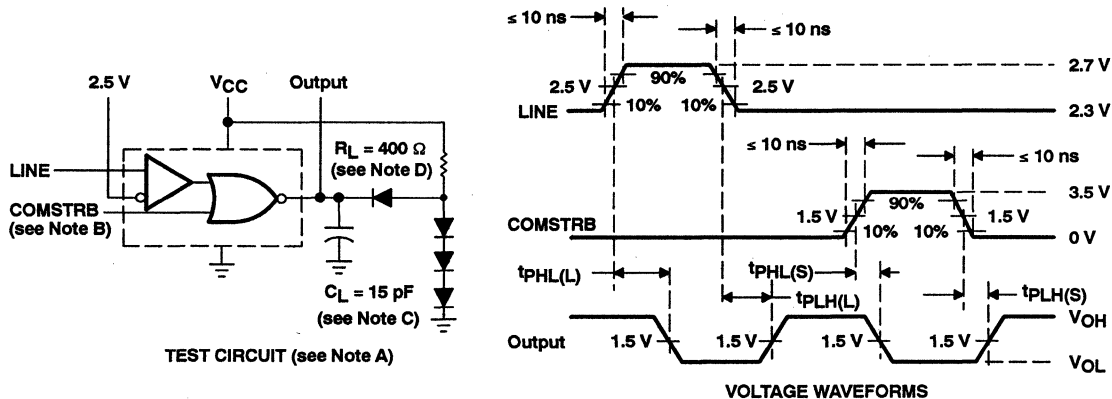
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH(L)}$	Propagation delay time, low-to-high-level output from LINE	$C_L = 15\text{ pF}$ , $R_L = 400\text{ k}\Omega$ , See Figure 1		22	35	ns
$t_{PHL(L)}$	Propagation delay time, high-to-low-level output from LINE			22	30	
$t_{PLH(S)}$	Propagation delay time, low-to-high-level output from COMSTRB			12	22	ns
$t_{PHL(S)}$	Propagation delay time, high-to-low-level output from COMSTRB			8	15	



# SN75140, SN75141 DUAL LINE RECEIVERS

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## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. Input pulses are supplied by generators having the following characteristics: PRR  $\leq$  1 MHz, duty cycle  $\leq$  50%,  $Z_O = 50 \Omega$ .  
 B. Unused strobes are to be grounded.  
 C.  $C_L$  includes probe and jig capacitance.  
 D. All diodes are 1N3064.

Figure 1. Test Circuit and Voltage Waveforms

## TYPICAL CHARACTERISTICS

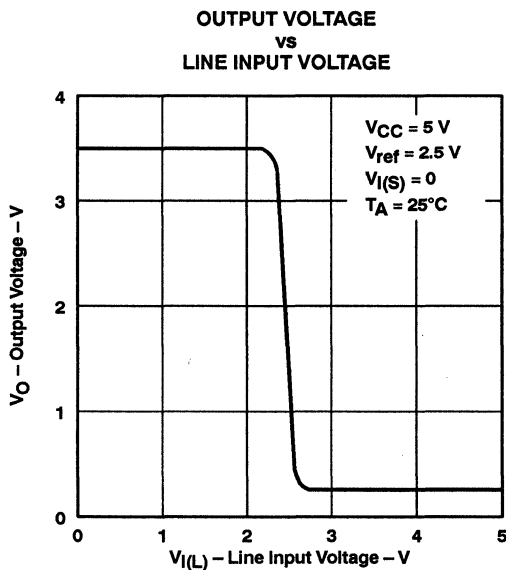


Figure 2

APPLICATION INFORMATION

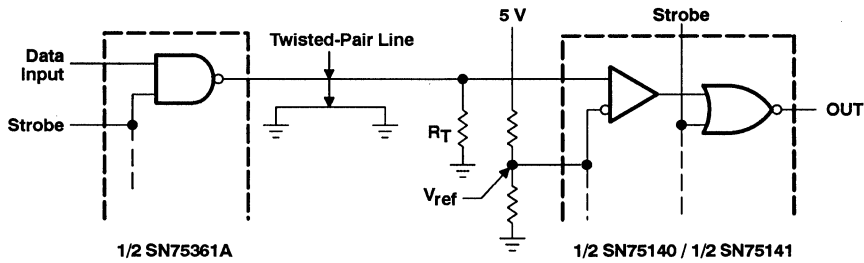
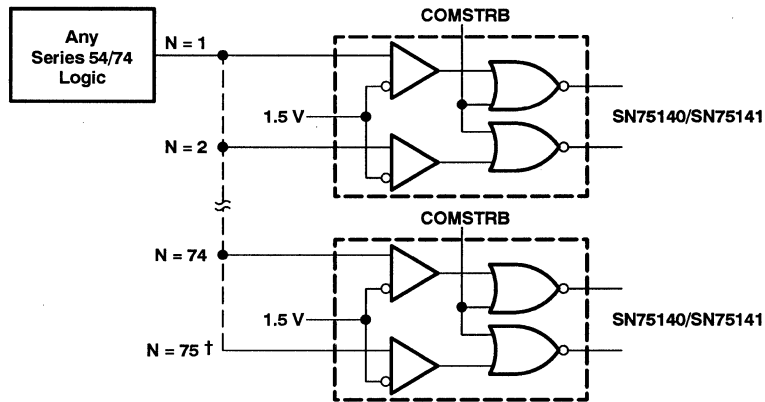


Figure 3. Line Receiver



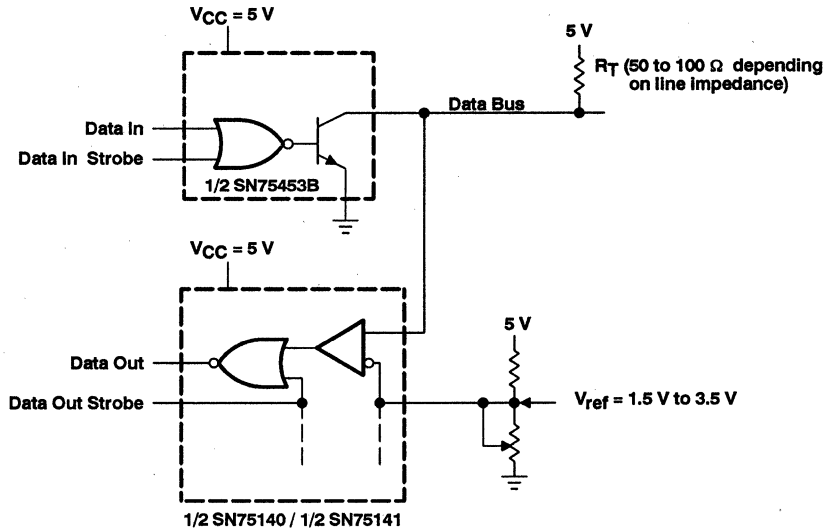
† Although most Series 54/74 circuits have a 2.4-V output at 400  $\mu$ A, they are typically capable of maintaining a 2.4-V output level under a load of 7.5 mA.

Figure 4. High Fanout From Standard TTL Gate

# SN75140, SN75141 DUAL LINE RECEIVERS

SLLS080A—D2155, JANUARY 1977—REVISED FEBRUARY 1993

## APPLICATION INFORMATION



NOTE: Using this arrangement, as many as 100 transceivers can be connected to a single data bus. The adjustable reference voltage feature allows the noise margin to be optimized for a given system. The complete dual bus transceiver (SN75453B driver and SN75140 receiver) can be assembled in approximately the same space required by a single 16-pin package and only one power supply is required (5 V). Data In and Data Out terminals are TTL compatible.

Figure 5. Dual Bus Transceiver

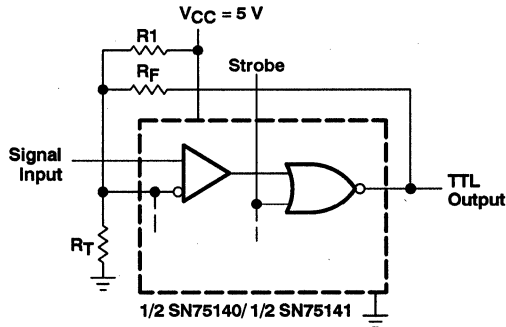
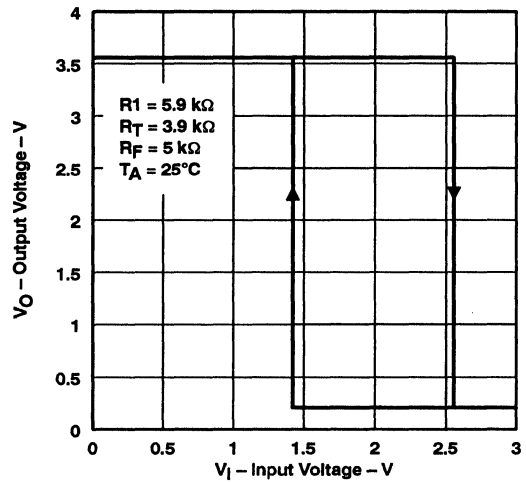
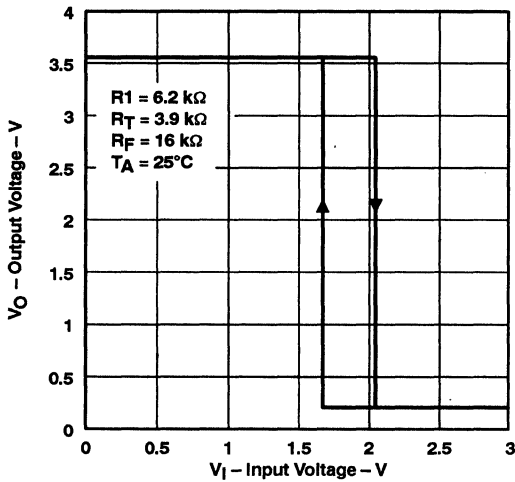


Figure 6. Schmitt Trigger

APPLICATION INFORMATION



NOTE: Slowly changing input levels from data lines, optical detectors, and other types of transducers may be converted to standard TTL signals with this Schmitt trigger circuit.  $R_1$ ,  $R_F$ , and  $R_T$  may be adjusted for the desired hysteresis and trigger levels.

Figure 7. Examples of Transfer Characteristics

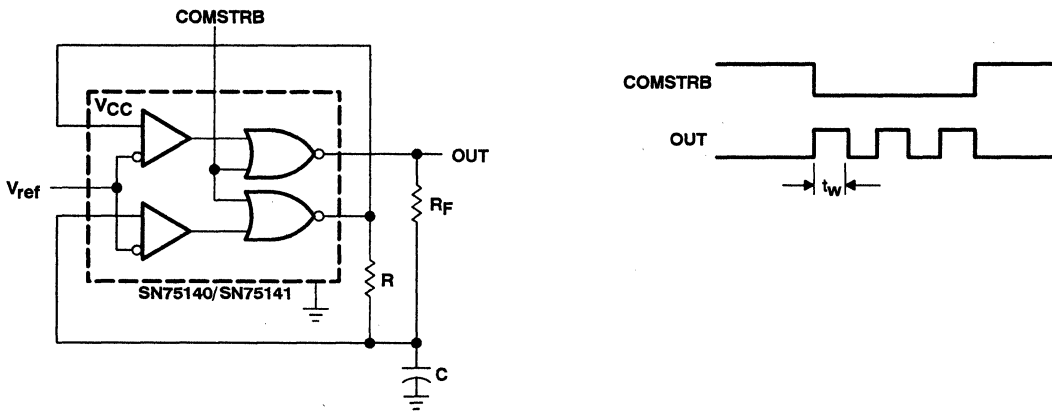


Figure 8. Gated Oscillator

# SN75140, SN75141 DUAL LINE RECEIVERS

SLLS080A - D2155, JANUARY 1977 - REVISED FEBRUARY 1993

## APPLICATION INFORMATION

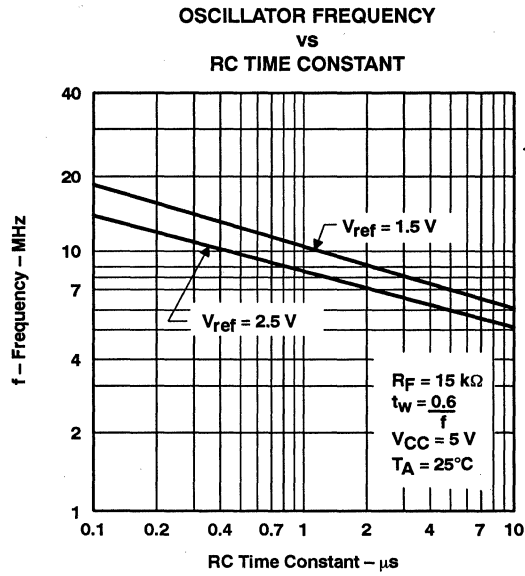


Figure 9

# SN75146 DUAL DIFFERENTIAL LINE RECEIVER

SLLS015A – D2609, FEBRUARY 1986 – REVISED FEBRUARY 1993

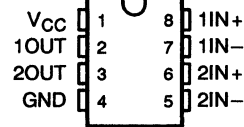
- Meets EIA Standards RS-422-A and RS-423-A
- Meets EIA Standards RS-232 and CCITT V.28 With External Components
- Meets Federal Standards 1020 and 1030
- Built-in 5-MHz Low-Pass Filter
- Operates From Single 5-V Power Supply
- Wide Common-Mode Voltage Range
- High Input Impedance
- TTL-Compatible Outputs
- 8-Pin Dual-in-Line Package
- Pinout Compatible With the  $\mu$ A9637 and  $\mu$ A9639

## description

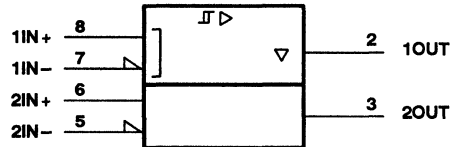
The SN75146 is a dual differential line receiver designed to meet EIA Standards RS-422-A and RS-423-A. The receiver is designed to have a constant impedance with input voltages of  $\pm 3$  V to  $\pm 25$  V allowing it to meet the requirements of EIA Standard RS-232-C and CCITT recommendation V.28 with the addition of an external bias resistor. This receiver is designed for low-speed operation below 355 kHz and has a built-in 5-MHz low-pass filter to attenuate high-frequency noise. The inputs are compatible with either a single-ended or a differential line system and the outputs are TTL compatible. This device operates from a single 5-V power supply and is supplied in both the 8-pin dual-in-line and small-outline packages.

The SN75146 is characterized for operation from 0°C to 70°C.

D OR P PACKAGE  
(TOP VIEW)

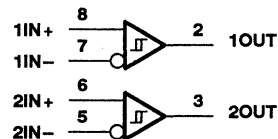


## logic symbol



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

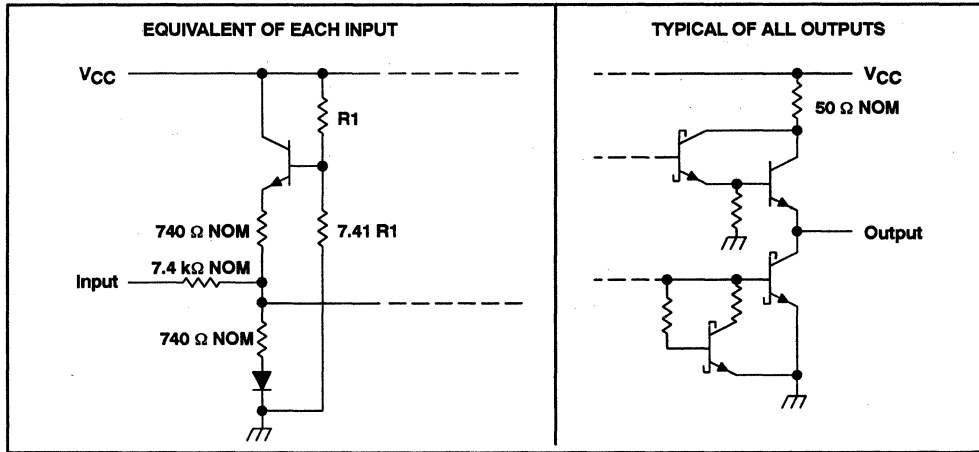
## logic diagram



# SN75146 DUAL DIFFERENTIAL LINE RECEIVER

SLLS015A - D2609, FEBRUARY 1986 - REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	-0.5 V to 7 V
Input voltage	$\pm 25$ V
Differential input voltage (see Note 2)	$\pm 25$ V
Output voltage range (see Note 1)	-0.5 V to 5.5 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to the network ground terminal.  
2. Differential input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Operating free-air temperature, $T_A$	0	25	70	°C

# SN75146 DUAL DIFFERENTIAL LINE RECEIVER

SLLS015A - D2609, FEBRUARY 1986 - REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of supply voltage, common-mode input voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_T$	Threshold voltage ( $V_{T+}$ and $V_{T-}$ )		-0.2‡		0.2	V
		See Note 3	-0.4‡		0.4	
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )		70			mV
$V_{IB}$	Input bias voltage	$I_I = 0$	2		2.4	V
$V_{OH}$	High-level output voltage	$V_{ID} = 0.2$ V, $I_O = -1$ mA	2.5	3.5		V
$V_{OL}$	Low-level output voltage	$V_{ID} = -0.2$ V, $I_O = 20$ mA		0.35	0.5	V
$r_i$	Input resistance	$V_I = 3$ V to 25 V or $V_I = -3$ V to -25 V, See Note 4	6	7.8	10.5	k $\Omega$
$I_I$	Input current	$V_{CC} = 0$ to 5.5 V, See Note 5	$V_I = 10$ V	1.1	3.25	mA
			$V_I = -10$ V	-1.6	-3.25	
$I_{OS}$	Short-circuit output current§	$V_O = 0$ , $V_{ID} = 0.2$ V	-40	-75	-100	mA
$I_{CC}$	Supply current	$V_{ID} = -0.5$ V, No load		35	50	mA

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold levels only.

§ Only one output should be shorted at a time, and duration of the short circuit should not exceed one second.

NOTES: 3. The expanded threshold parameter is tested with a 500- $\Omega$  resistor in series with each input.

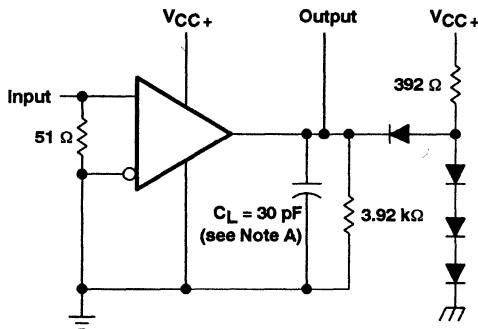
4.  $r_i$  is defined by  $\Delta V_I / \Delta I_I$ .

5. The input not under test is grounded.

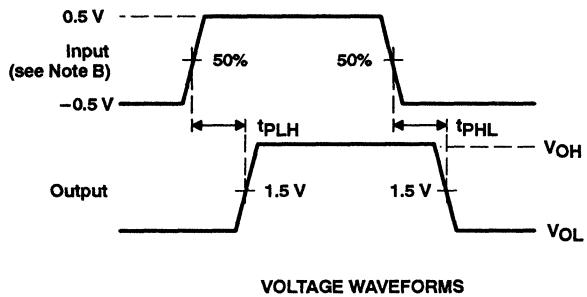
## switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	$C_L = 30$ pF, See Figure 1	100	150	300	ns
$t_{PHL}$					

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5$  ns,  $t_f \leq 5$  ns, PRR  $\leq 300$  kHz, duty cycle = 50%.

Figure 1. Test Circuit and Voltage Waveforms

**TEXAS**  
**INSTRUMENTS**

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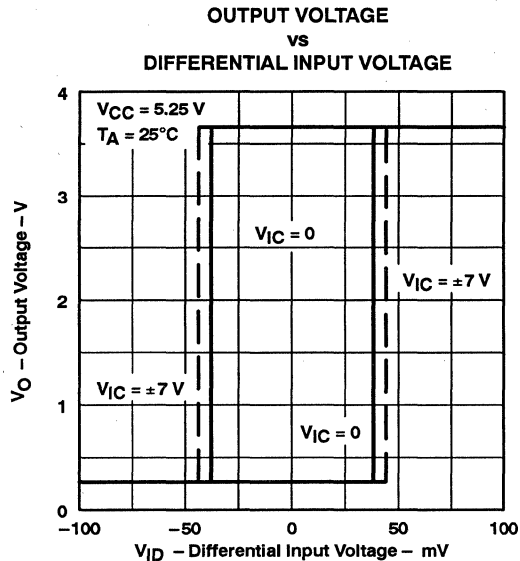
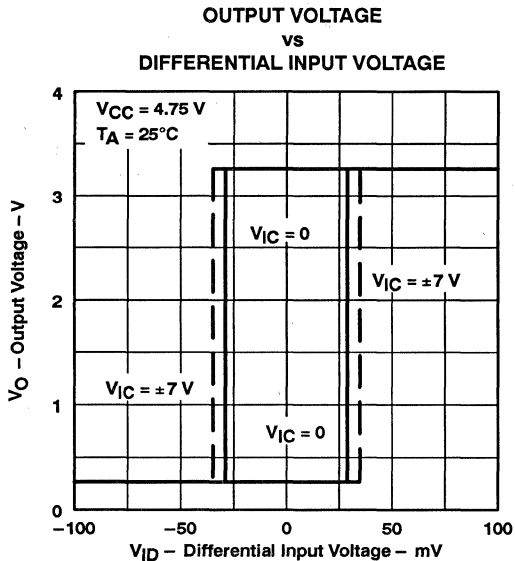
2-339



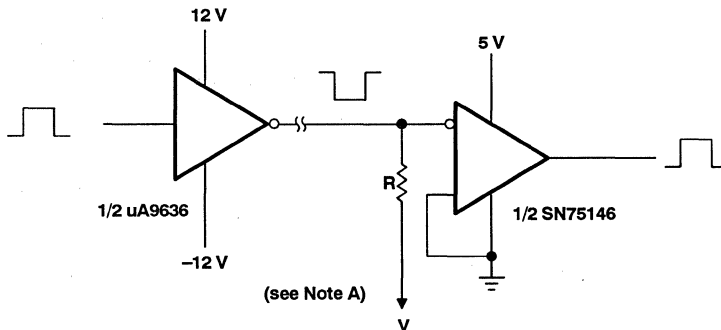
# SN75146 DUAL DIFFERENTIAL LINE RECEIVER

SLLS015A – D2609, FEBRUARY 1986 – REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS



## APPLICATION INFORMATION



NOTE A: In order to meet the input-impedance and open-circuit-input voltage requirements of RS-232-C and CCITT V.28 and ensure open-circuit-input fail-safe operation, R and V are selected to satisfy the following equations:

$$V = -1.1 - 3.3 \frac{R}{r_i} \text{ volts}$$

$$3 \text{ k}\Omega \leq \frac{R(r_i)}{R + r_i} \leq 7 \text{ k}\Omega$$

Figure 4. RS-232-C System Applications

APPLICATION INFORMATION

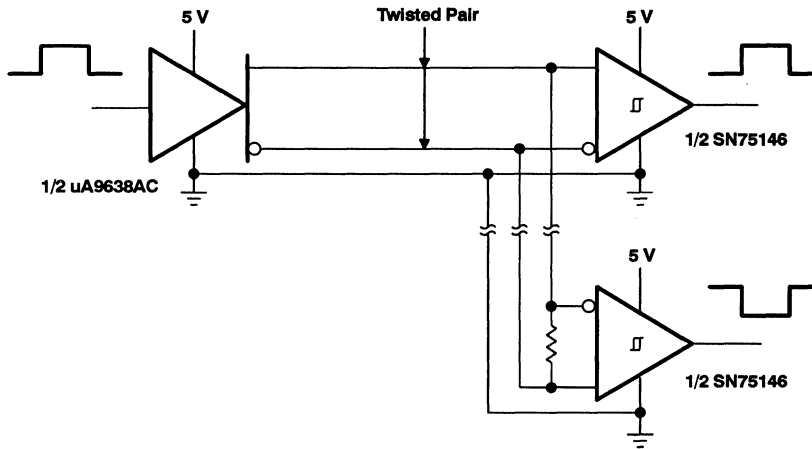


Figure 5. RS-422-A System Applications



# SN75150 DUAL LINE DRIVER

SLLS081A – D951, JANUARY 1971 – REVISED MARCH 1993

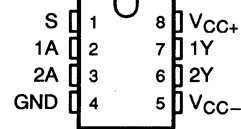
- Satisfies Requirement of EIA Standard RS-232-C
- Withstands Sustained Output Short Circuit to Any Low-Impedance Voltage Between  $-25\text{ V}$  and  $25\text{ V}$
- $2\text{-}\mu\text{s}$  Max Transition Time Through the 3-V to  $-3\text{-V}$  Transition Region Under Full 2500-pF Load
- Inputs Compatible With Most TTL Families
- Common Strobe Input
- Inverting Output
- Slew Rate Can Be Controlled With an External Capacitor at the Output
- Standard Supply Voltages . . .  $\pm 12\text{ V}$

## description

The SN75150 is a monolithic dual line driver designed to satisfy the requirements of the standard interface between data terminal equipment and data communication equipment as defined by EIA Standard RS-232-C. A rate of 20000 bits per second can be transmitted with a full 2500-pF load. Other applications are in data-transmission systems using relatively short single lines, in level translators, and for driving MOS devices. The logic input is compatible with most TTL families. Operation is from 12-V and  $-12\text{-V}$  power supplies.

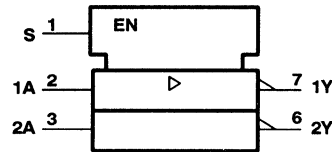
The SN75150 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

D, P, OR PS† PACKAGE  
(TOP VIEW)



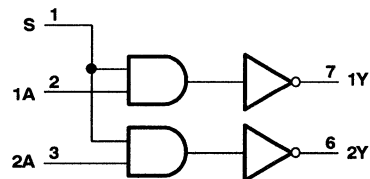
† The PS package is only available left-end taped and reeled, i.e., order device SN75150PSLE.

## logic symbol‡



‡ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

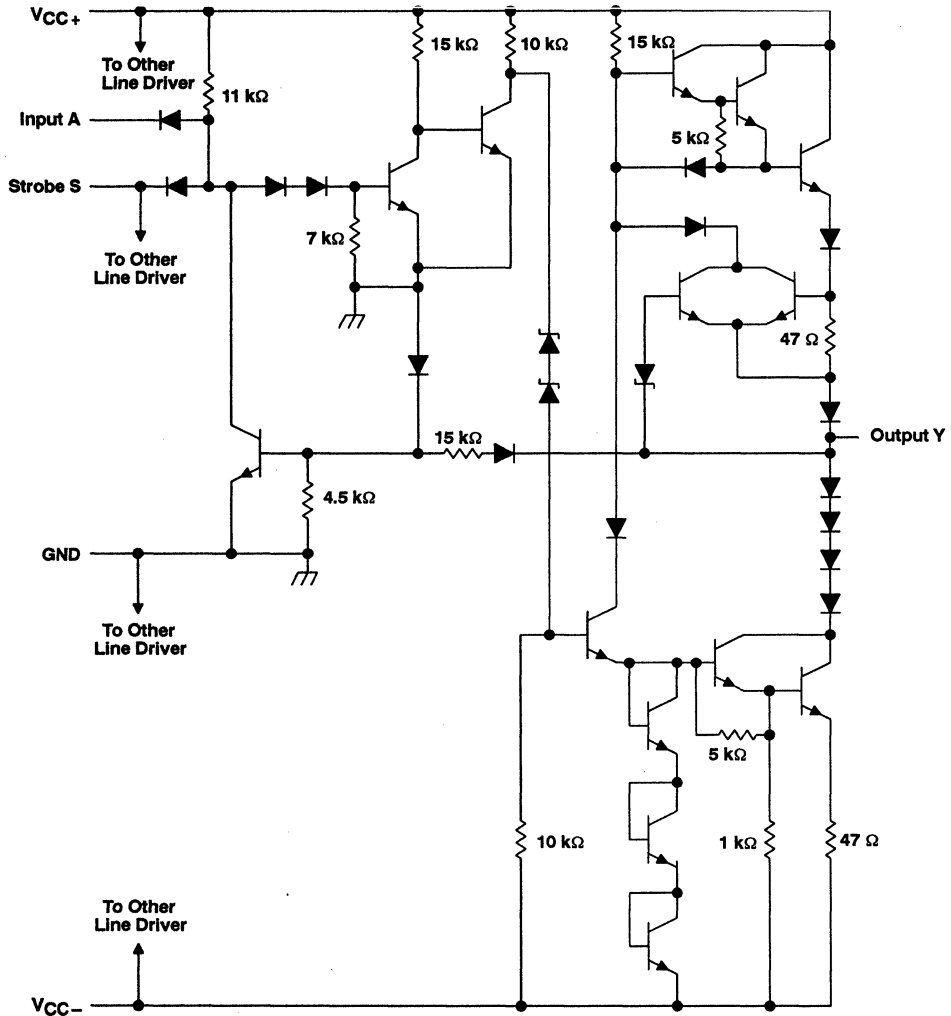
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# SN75150 DUAL LINE DRIVER

SLLS081A-D951, JANUARY 1971 - REVISED MARCH 1993

## schematic (each line driver)



Resistor values shown are nominal.

# SN75150 DUAL LINE DRIVER

SLLS081A - D951, JANUARY 1971 - REVISED MARCH 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC+}$ (see Note 1) .....	15 V
Supply voltage, $V_{CC-}$ .....	-15 V
Input voltage, $V_I$ .....	15 V
Applied output voltage .....	$\pm 25$ V
Continuous total power dissipation .....	See Dissipation Rating Table
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

NOTE 1: Voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW
PS	450 mW	3.6 mW/°C	288 mW

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$	10.8	12	13.2	V
Supply voltage, $V_{CC-}$	-10.8	-12	-13.2	V
High-level input voltage, $V_{IH}$	2		5.5	V
Low-level input voltage, $V_{IL}$	0		0.8	V
Applied output voltage, $V_O$			$\pm 15$	V
Operating free-air temperature, $T_A$	0		70	°C



# SN75150 DUAL LINE DRIVER

SLLS081A - D951, JANUARY 1971 - REVISED MARCH 1993

**electrical characteristics over recommended operating free-air temperature range,  $V_{CC\pm} = \pm 13.2 V$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage	$V_{CC+} = 10.8 V$ , $V_{CC-} = -13.2 V$ , $V_{IL} = 0.8 V$ , $R_L = 3 k\Omega$ to $7 k\Omega$	5	8		V
$V_{OL}$	Low-level output voltage (see Note 2)	$V_{CC+} = 10.8 V$ , $V_{CC-} = -10.8 V$ , $V_{IH} = 2 V$ , $R_L = 3 k\Omega$ to $7 k\Omega$		-8	-5	V
$I_{IH}$	High-level input current	$V_I = 2.4 V$		1	10	$\mu A$
	Data input			2	20	
$I_{IL}$	Low-level input current	$V_I = 0.4 V$		-1	-1.8	mA
	Strobe input			-2	-3.2	
$I_{OS}$	Short-circuit output current‡	$V_O = 25 V$		2	8	mA
		$V_O = -25 V$		-3	-8	
		$V_O = 0$ , $V_I = 3 V$	10	15	30	
		$V_O = 0$ , $V_I = 0$	-10	-15	-30	
$I_{CCH+}$	Supply current from $V_{CC+}$ , high-level output	$V_I = 0$ , $R_L = 3 k\Omega$ , $T_A = 25^\circ C$		10	22	mA
$I_{CCH-}$	Supply current from $V_{CC-}$ , high-level output			-1	-10	
$I_{CCL+}$	Supply current from $V_{CC+}$ , low-level output	$V_I = 3 V$ , $R_L = 3 k\Omega$ , $T_A = 25^\circ C$		8	17	mA
$I_{CCL-}$	Supply current from $V_{CC-}$ , low-level output			-9	-20	

† All typical values are at  $V_{CC+} = 12 V$ ,  $V_{CC-} = -12 V$ ,  $T_A = 25^\circ C$ .

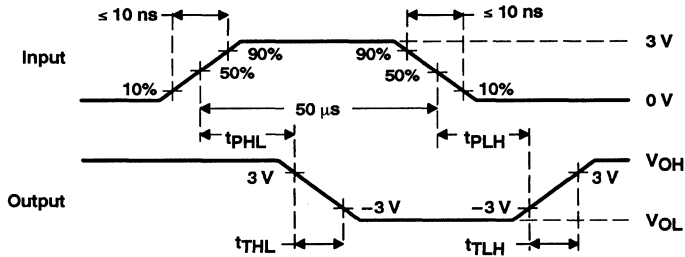
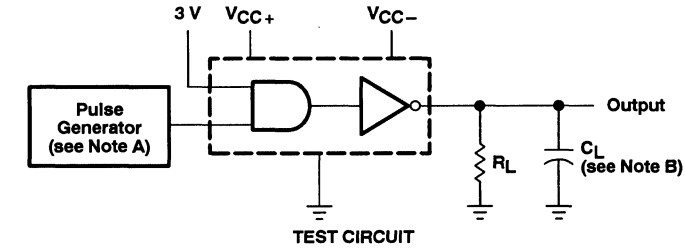
‡ Not more than one output should be shorted at a time.

NOTE 2: The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for logic levels only, e.g., when  $-5 V$  is the maximum, the typical value is a more negative voltage.

**switching characteristics,  $V_{CC+} = 12 V$ ,  $V_{CC-} = -12 V$ ,  $T_A = 25^\circ C$  (see Figure 1)**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{TLH}$	Transition time, low-to-high-level output	$C_L = 2500 pF$ , $R_L = 3 k\Omega$ to $7 k\Omega$	0.2	1.4	2	$\mu s$
$t_{THL}$	Transition time, high-to-low-level output		0.2	1.5	2	$\mu s$
$t_{TLH}$	Transition time, low-to-high-level output	$C_L = 15 pF$ , $R_L = 7 k\Omega$		40		ns
$t_{THL}$	Transition time, high-to-low-level output			20		ns
$t_{PLH}$	Propagation delay time, low-to-high-level output	$C_L = 15 pF$ , $R_L = 7 k\Omega$		60		ns
$t_{PHL}$	Propagation delay time, high-to-low-level output			45		ns

PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics: duty cycle  $\leq 50\%$ ,  $Z_O \sim 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms



# SN75150 DUAL LINE DRIVER

SLLS081A - D951, JANUARY 1971 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

OUTPUT CURRENT  
vs  
APPLIED OUTPUT VOLTAGE

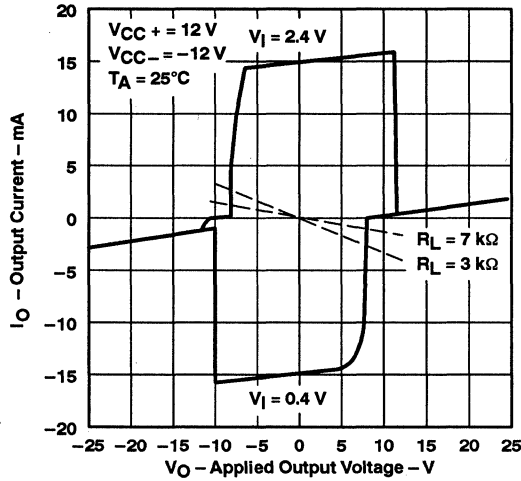


Figure 2

## APPLICATION INFORMATION

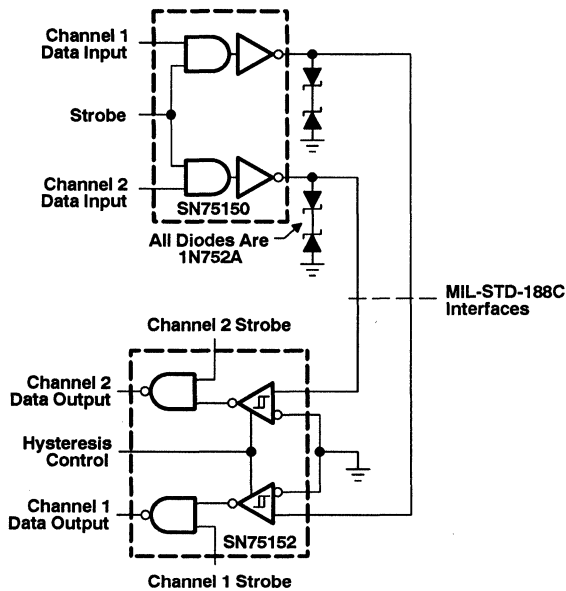
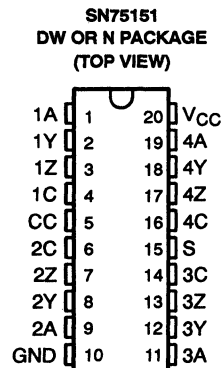


Figure 3. Dual-Channel Single-Ended Interface Circuit Meeting MIL-STD-188C, Paragraph 7.2.

# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS082A - D2453, DECEMBER 1978 - REVISED FEBRUARY 1993

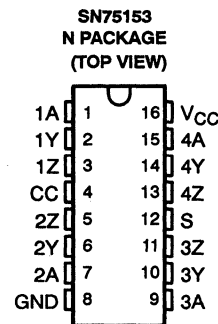
- Meets EIA Standard RS-422-A
- High-Impedance Output State for Party-Line Operation
- High Output Impedance in Power-Off Condition
- Low Input Current to Minimize Loading
- Single 5-V Supply
- 40-mA Sink- and Source-Current Capability
- High-Speed Schottky Circuitry
- Low Power Requirements



## description

These line drivers are designed to provide differential signals with high current capability on balanced lines. These circuits provide strobe and enable inputs to control all four drivers, and the SN75151 provides an additional enable input for each driver. The output circuits have active pullup and pulldown and are capable of sinking or sourcing 40 mA.

The SN75151 and SN75153 meet all requirements of EIA Standard RS-422-A and Federal Standard 1020. They are characterized for operation from 0°C to 70°C.



**SN75153**  
NOT RECOMMENDED FOR NEW DESIGN

## Function Tables

SN75151				OUTPUTS	
INPUTS				Y	Z
ENABLE CC	ENABLE C	STROBE S	DATA A		
L	X	X	X	Z	Z
X	L	X	X	Z	Z
H	H	L	X	L	H
H	H	X	L	L	H
H	H	H	H	H	L

SN75153			OUTPUTS	
ENABLE CC	STROBE S	DATA A	Y	Z
L	X	X	Z	Z
H	L	X	L	H
H	X	L	L	H
H	H	H	H	L

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

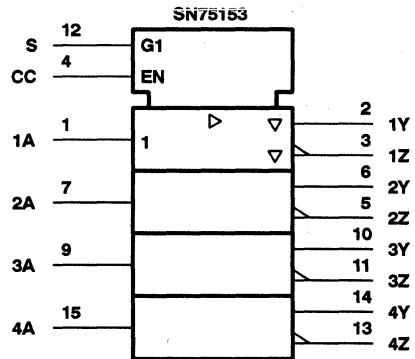
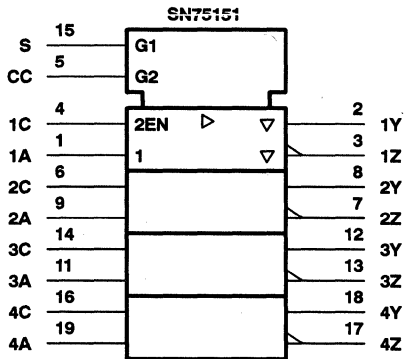
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# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

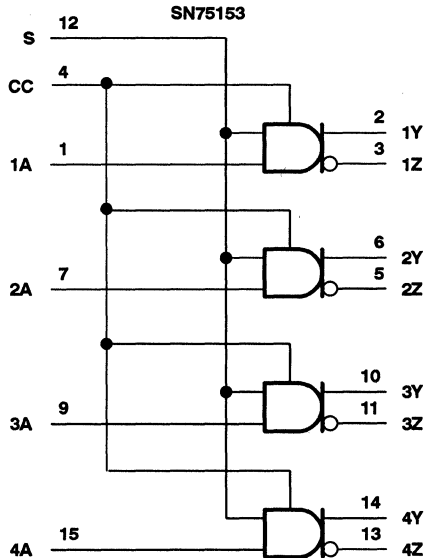
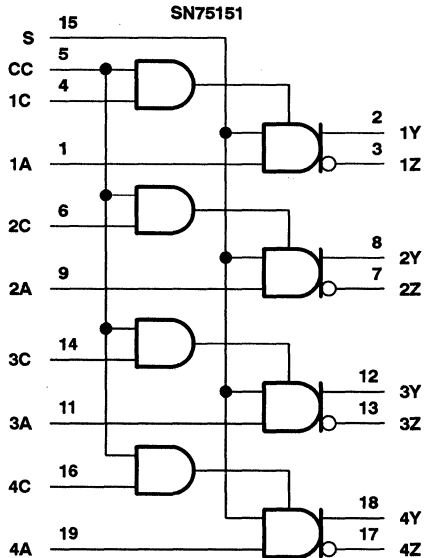
SLLS082A - D2453, DECEMBER 1978 - REVISED FEBRUARY 1993

## logic symbols†



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

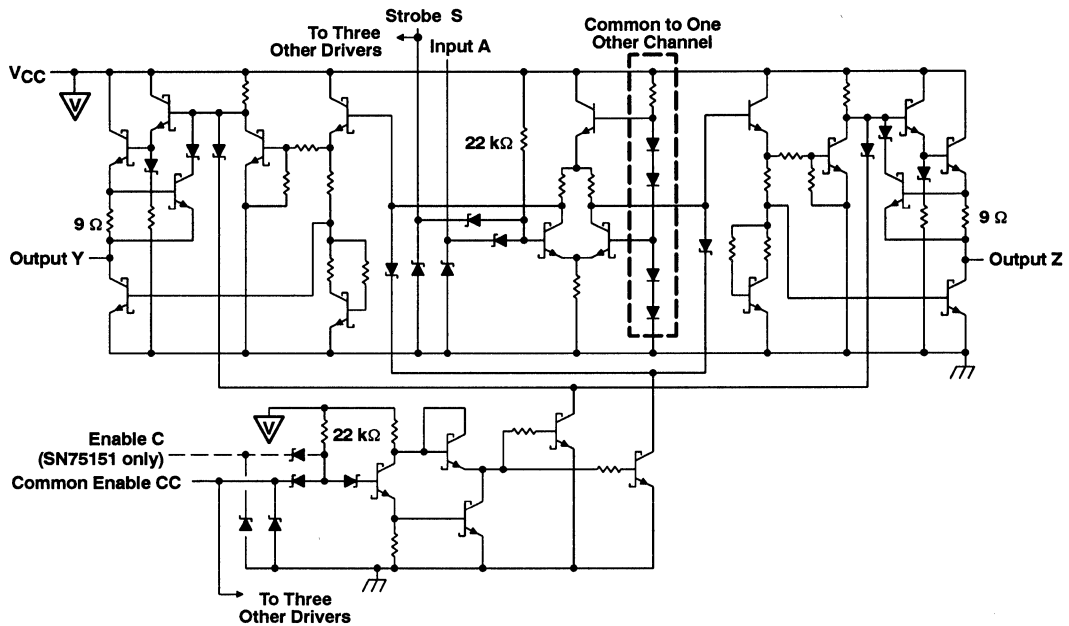
## logic diagrams (positive logic)



# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS082A – D2453, DECEMBER 1978 – REVISED FEBRUARY 1993

## schematic



Resistor values shown are nominal.

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values, except differential output voltage  $V_{OD}$ , are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	OPERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
N	1150 mW	9.2 mW/°C	736 mW

  
**TEXAS  
INSTRUMENTS**

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# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS082A - D2453, DECEMBER 1978 - REVISED FEBRUARY 1993

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
Common-mode output voltage, $V_{OC}$	-0.25		6	V
High-level output current, $I_{OH}$			-40	mA
Low-level output current, $I_{OL}$			40	mA
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{IK}$	Input clamp voltage	$V_{CC} = \text{MIN}$ , $I_I = -12 \text{ mA}$	CC, S			-2	V
			All others		-0.9	-1.5	
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ , $V_{IH} = 2 \text{ V}$ , $V_{IL} = \text{MAX}$	$I_{OH} = -20 \text{ mA}$	2.5			V
			$I_{OH} = -40 \text{ mA}$	2.4			
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ , $I_{OL} = 40 \text{ mA}$ , $V_{IH} = 2 \text{ V}$ , $V_{IL} = \text{MAX}$				0.5	V
$ V_{OD1} $	Differential output voltage	$V_{CC} = \text{MAX}$ , $I_O = 0$		3.4	$2V_{OD2}$		V
$ V_{OD2} $	Differential output voltage	$V_{CC} = \text{MIN}$		2	2.8		V
$\Delta V_{OD} $	Change in magnitude of differential output voltage§	$V_{CC} = \text{MIN}$	$R_L = 100 \Omega$ , See Figure 1		$\pm 0.01$	$\pm 0.4$	V
$V_{OC}$	Common-mode output voltage¶	$V_{CC} = \text{MAX}$		1.8	3		V
		$V_{CC} = \text{MIN}$		1.6	3		
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage§	$V_{CC} = \text{MIN or MAX}$			$\pm 0.02$	$\pm 0.4$	
$I_{OZ}$	Off-state (high-impedance state) output current	$V_{CC} = \text{MAX}$ , Enable at 0.8 V	$V_O = 0.5 \text{ V}$			-20	$\mu\text{A}$
			$V_O = 2.5 \text{ V}$			20	
			$V_O = V_{CC}$			20	
$I_O$	Output current with power off	$V_{CC} = 0$	$V_O = 6 \text{ V}$		0.1	100	$\mu\text{A}$
			$V_O = -0.25 \text{ V}$		-0.1	-100	
			$V_O = -0.25 \text{ V to } 6 \text{ V}$			$\pm 100$	
$I_I$	Input current at maximum input voltage	$V_{CC} = \text{MAX}$ , $V_I = 5.5 \text{ V}$				0.1	mA
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$ , $V_I = 2.4 \text{ V}$	C(SN75151), A			20	$\mu\text{A}$
			CC, S			80	
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$ , $V_I = 0.4 \text{ V}$	C(SN75151), A			-0.36	mA
			CC, S			-1.6	
$I_{OS}$	Short-circuit output current#	$V_{CC} = \text{MAX}$		-50	-90	-150	mA
$I_{CC}$	Supply current (both drivers)	$V_{CC} = \text{MAX}$ , No load	Outputs disabled		30	60	mA
			Outputs enabled		60	80	

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at  $T_A = 25^\circ\text{C}$  and  $V_{CC} = 5 \text{ V}$  except for  $V_{OC}$ , for which  $V_{CC}$  is as stated under test conditions.

§  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitudes of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

¶ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

# Only one output should be shorted at a time, and duration of the short-circuit should not exceed one second.

TEXAS  
INSTRUMENTS

# SN75151, SN75153

## QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS082A – D2453, DECEMBER 1978 – REVISED FEBRUARY 1993

switching characteristics over recommended operating free-air temperature range,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 30\text{ pF}$ , See Figure 2, $R_L = 100\ \Omega$ , Termination A		15	30	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			15	30	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 30\text{ pF}$ , See Figure 2, Termination B		13	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			13	25	ns
$t_{TLH}$ Transition time, low-to-high-level output	$C_L = 30\text{ pF}$ , See Figure 2, $R_L = 100\ \Omega$ , Termination A		12	20	ns
$t_{THL}$ Transition time, high-to-low-level output			12	20	ns
$t_{pZH}$ Output enable time to high level	$C_L = 30\text{ pF}$ , See Figure 3, $R_L = 60\ \Omega$ ,		18	35	ns
$t_{pZL}$ Output enable time to low level	$C_L = 30\text{ pF}$ , See Figure 4, $R_L = 111\ \Omega$ ,		20	35	ns
$t_{pHZ}$ Output disable time from high level	$C_L = 30\text{ pF}$ , See Figure 3, $R_L = 60\ \Omega$ ,		19	30	ns
$t_{pLZ}$ Output disable time from low level	$C_L = 30\text{ pF}$ , See Figure 4, $R_L = 111\ \Omega$ ,		13	30	ns
Overshoot factor	$R_L = 100\ \Omega$ , See Figure 2, Termination C			10	%

† All typical values are at  $T_A = 25^\circ\text{C}$ .

### PARAMETER MEASUREMENT INFORMATION

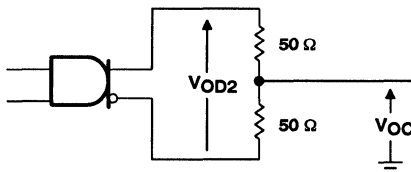
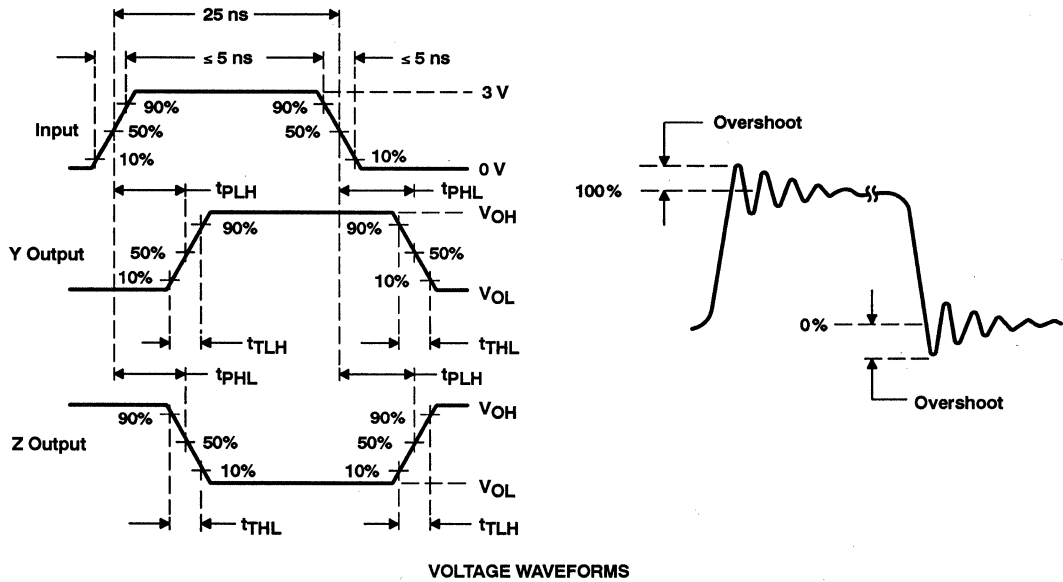
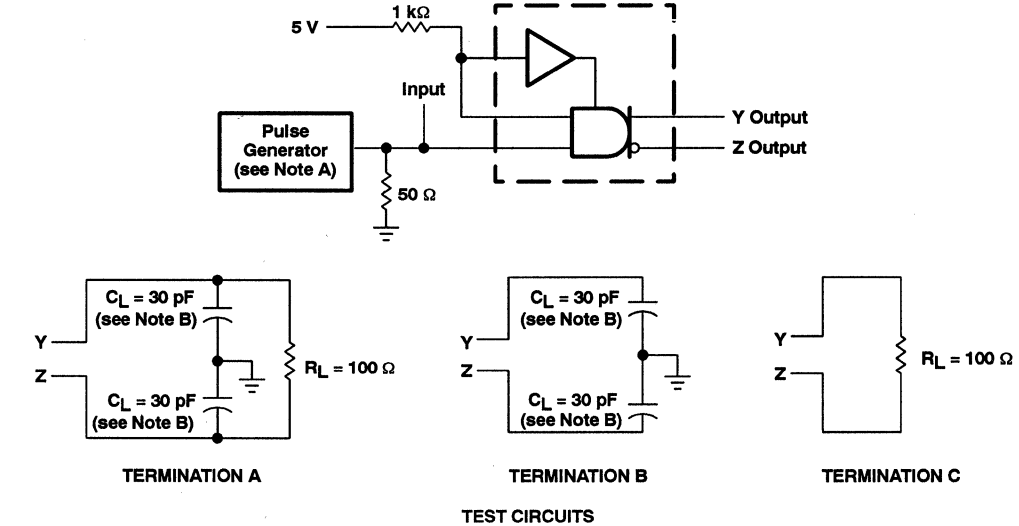


Figure 1. Differential and Common-Mode Output Voltages

# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS082A – D2453, DECEMBER 1978 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



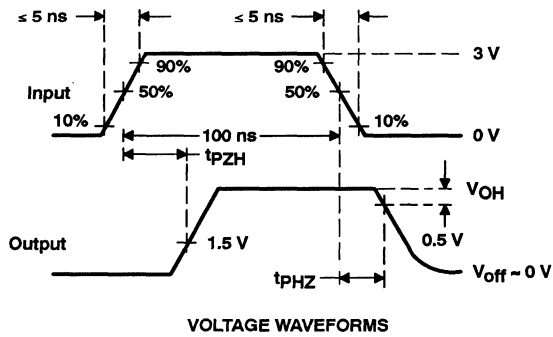
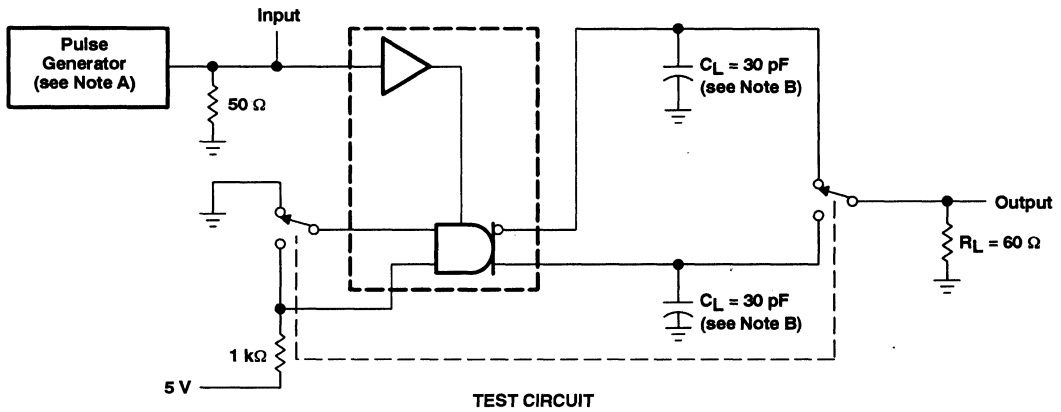
NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50\ \Omega$ ,  $PRR \leq 10\text{ MHz}$ .  
B.  $C_L$  includes probe and jig capacitance.

**Figure 2. Test Circuits, Voltage Waveforms, and Overshoot Factor**

# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

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## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ .  
B.  $C_L$  includes probe and jig capacitance.

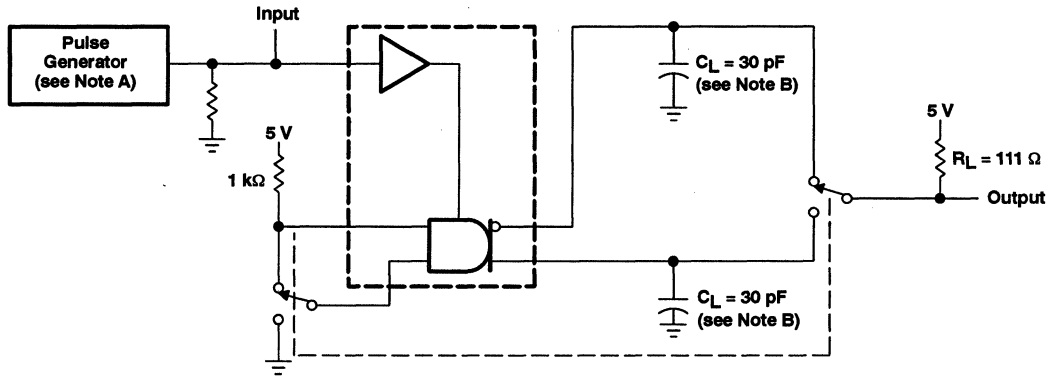
**Figure 3. Test Circuit and Voltage Waveforms**



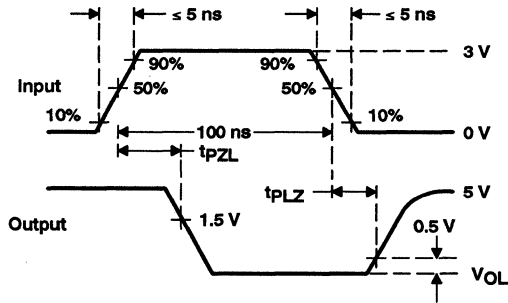
# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS082A – D2453, DECEMBER 1978 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

- NOTES: A. The pulse generators have the following characteristics:  $Z_O = 50 \Omega$ ,  $\text{PRR} \leq 500 \text{ kHz}$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 4. Test Circuit and Voltage Waveforms

# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

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## TYPICAL CHARACTERISTICS

**Y OUTPUT VOLTAGE  
vs  
DATA INPUT VOLTAGE**

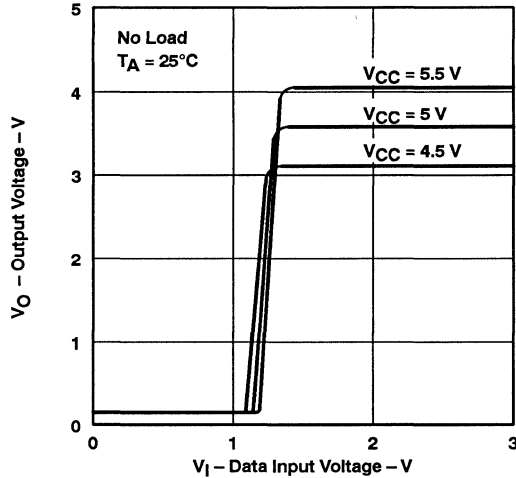
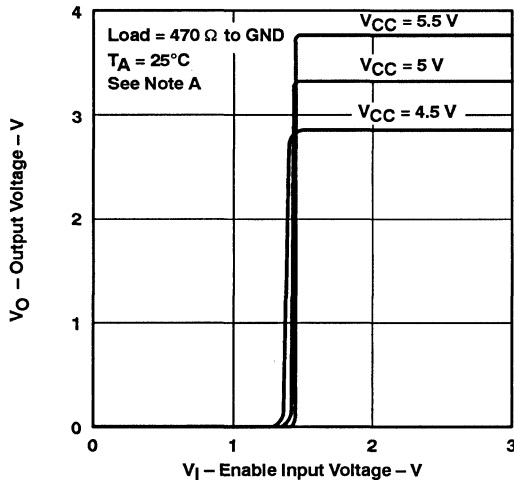


Figure 5

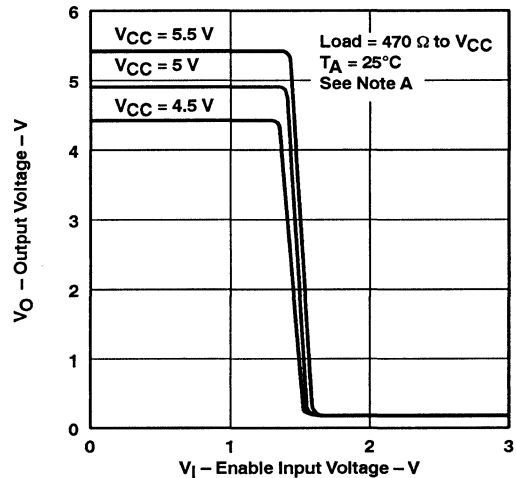
**Y OR Z OUTPUT VOLTAGE  
vs  
ENABLE INPUT VOLTAGE**



NOTE A: The A input is connected to  $V_{CC}$  during the testing of the Y outputs and to ground during testing of the Z outputs.

Figure 6

**Y OR Z OUTPUT VOLTAGE  
vs  
ENABLE INPUT VOLTAGE**



NOTE A: The A input is connected to GND during the testing of the Y outputs and to  $V_{CC}$  during the testing of the Z outputs.

Figure 7

# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

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## TYPICAL CHARACTERISTICS

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**

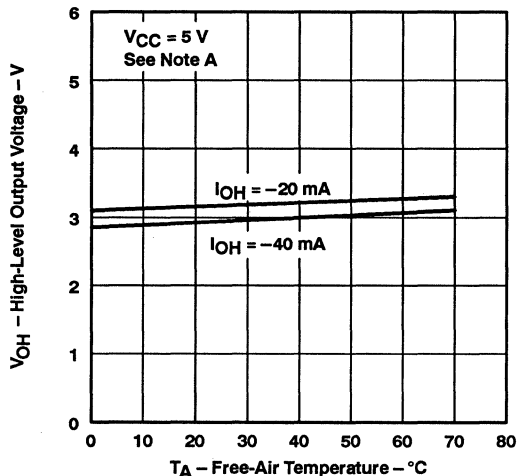


Figure 8

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT**

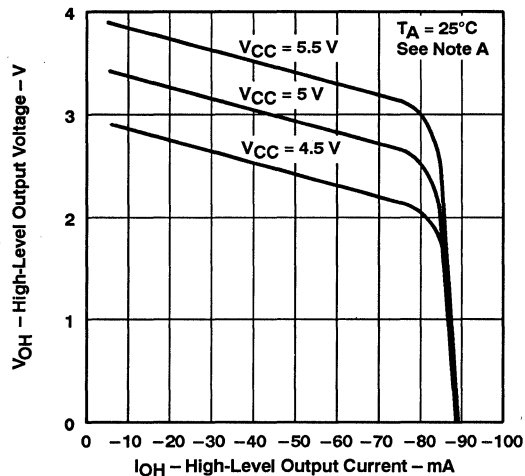


Figure 9

NOTE A: The A input is connected to V<sub>CC</sub> during the testing of the Y outputs and to ground during testing of the Z outputs.

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**

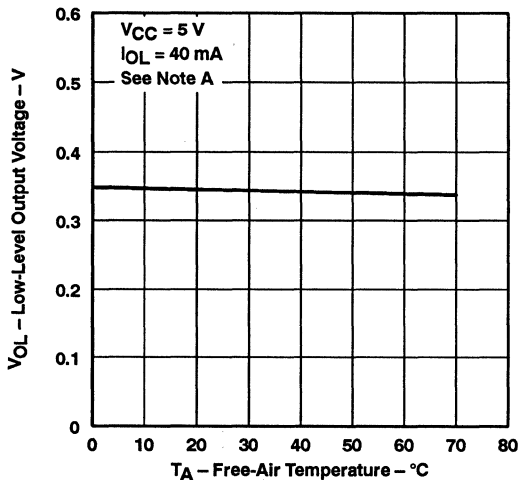


Figure 10

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT**

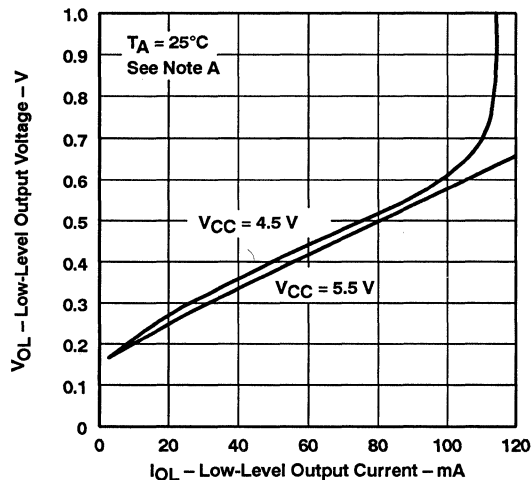


Figure 11

NOTE A: The A input is connected to GND during the testing of the Y outputs and to V<sub>CC</sub> during the testing of the Z outputs.



# SN75151, SN75153 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

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## TYPICAL CHARACTERISTICS

SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

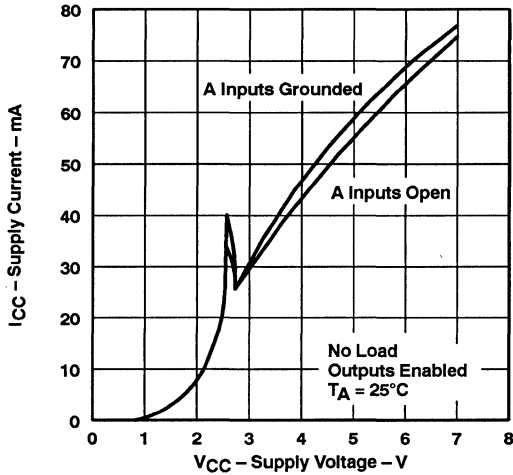


Figure 12

SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

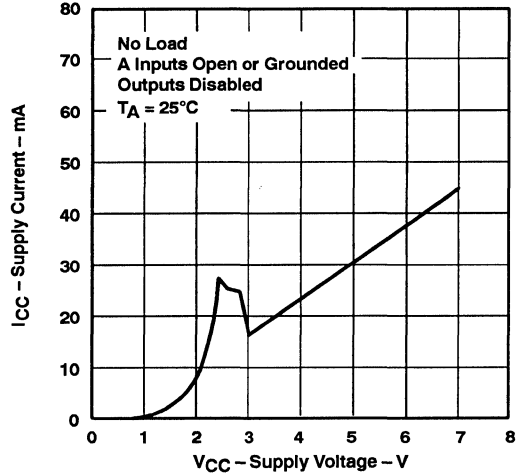


Figure 13

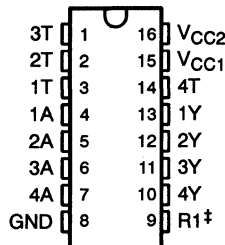


# SN75154 QUAD LINE RECEIVER

SLLS083A – D899, NOVEMBER 1970 – REVISED MARCH 1993

- Satisfies Requirements of EIA Standard RS-232-C
- Input Resistance . . . 3 k $\Omega$  to 7 k $\Omega$  Over Full RS-232-C Voltage Range
- Input Threshold Adjustable to Meet Fail-Safe Requirements Without Using External Components
- Built-In Hysteresis for Increased Noise Immunity
- Inverting Output Compatible With TTL
- Output With Active Pullup for Symmetrical Switching Speeds
- Standard Supply Voltages . . . 5 V or 12 V

D, N, OR NS† PACKAGE  
(TOP VIEW)



† The NS package is only available left-end taped and reeled, i.e., order device SN75154NSLE.

‡ For function of R1, see schematic

## description

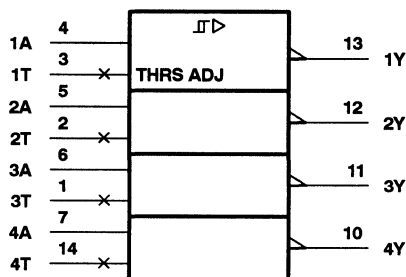
The SN75154 is a monolithic low-power Schottky line receiver designed to satisfy the requirements of the standard interface between data terminal equipment and data communication equipment as defined by EIA Standard RS-232-C. Other applications are for relatively short, single-line, point-to-point data transmission and for level translators. Operation is normally from a single 5-V supply; however, a built-in option allows operation from a 12-V supply without the use of additional components. The output is compatible with most TTL circuits when either supply voltage is used.

In normal operation, the threshold-control terminals are connected to the  $V_{CC1}$  terminal, even if power is being supplied via the alternate  $V_{CC2}$  terminal. This provides a wide hysteresis loop, which is the difference between the positive-going and negative-going threshold voltages. See typical characteristics. In this mode of operation, if the input voltage goes to zero, the output voltage will remain at the low or high level as determined by the previous input.

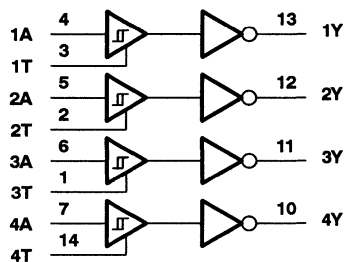
For fail-safe operation, the threshold-control terminals are open. This reduces the hysteresis loop by causing the negative-going threshold voltage to be above zero. The positive-going threshold voltage remains above zero as it is unaffected by the disposition of the threshold terminals. In the fail-safe mode, if the input voltage goes to zero or an open-circuit condition, the output will go to the high level regardless of the previous input condition.

The SN75154 is characterized for operation from 0°C to 70°C.

## logic symbol†



## logic diagram



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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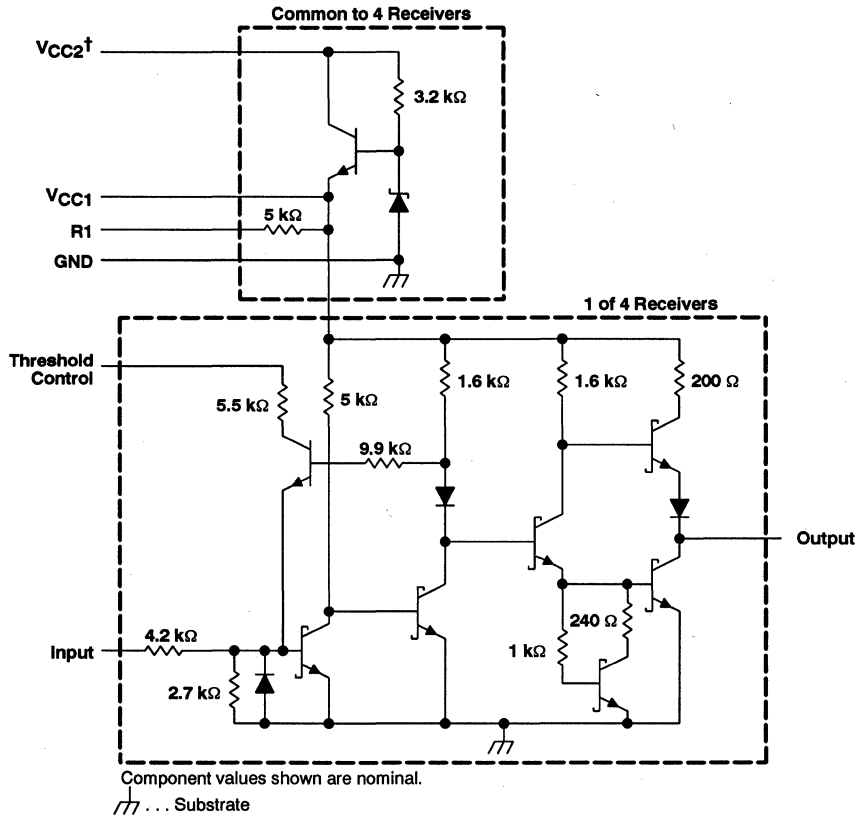
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# SN75154 QUAD LINE RECEIVER

SLLS083A - D899, NOVEMBER 1970 - REVISED MARCH 1993

## schematic



† When VCC1 is used, VCC2 may be left open or shorted to VCC1. When VCC2 is used, VCC1 must be left open or connected to the threshold control pins.

# SN75154 QUAD LINE RECEIVER

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Normal supply voltage, $V_{CC1}$ (see Note 1)	7 V
Alternate supply voltage, $V_{CC2}$	14 V
Input voltage, $V_I$	$\pm 25$ V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	$0^{\circ}\text{C}$ to $70^{\circ}\text{C}$
Storage temperature range	$-65^{\circ}\text{C}$ to $150^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	$260^{\circ}\text{C}$

NOTE 1: Voltage values are with respect to network GND terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A = 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	$T_A = 70^{\circ}\text{C}$ POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW
NS	625 mW	5.0 mW/ $^{\circ}\text{C}$	400 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Normal supply voltage, $V_{CC1}$	4.5	5	5.5	V
Alternate supply voltage, $V_{CC2}$	10.8	12	13.2	V
High-level input voltage, $V_{IH}$ (see Note 2)	3		15	V
Low-level input voltage, $V_{IL}$ (see Note 2)	-15		-3	V
High-level output current, $I_{OH}$			-400	$\mu\text{A}$
Low-level output current, $I_{OL}$			16	mA
Operating free-air temperature, $T_A$	0		70	$^{\circ}\text{C}$

NOTE 2: The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for logic and threshold levels only, e.g., when 0 V is the maximum, the minimum limit is a more negative voltage.



# SN75154 QUAD LINE RECEIVER

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**electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)**

PARAMETER		TEST FIGURE	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
V <sub>T+</sub>	Positive-going threshold voltage	Normal operation		0.8	2.2	3	V
		Fail-safe operation		0.8	2.2	3	
V <sub>T-</sub>	Negative-going threshold voltage	Normal operation		-3	-1.1	0	V
		Fail-safe operation		0.8	1.4	3	
V <sub>hys</sub>	Hysteresis (V <sub>T+</sub> - V <sub>T-</sub> )	Normal operation		0.8	3.3	6	V
		Fail-safe operation		0	0.8	2.2	
V <sub>OH</sub>	High-level output voltage	1	I <sub>OH</sub> = -400 μA	2.4	3.5		V
V <sub>OL</sub>	Low-level output voltage	1	I <sub>OL</sub> = 16 mA		0.29	0.4	V
r <sub>i</sub>	Input resistance	2	ΔV <sub>I</sub> = -25 V to -14 V	3	5	7	kΩ
			ΔV <sub>I</sub> = -14 V to -3 V	3	5	7	
			ΔV <sub>I</sub> = -3 V to 3 V	3	6	8	
			ΔV <sub>I</sub> = 3 V to 14 V	3	5	7	
			ΔV <sub>I</sub> = 14 V to 25 V	3	5	7	
V <sub>I(open)</sub>	Open-circuit input voltage	3	I <sub>I</sub> = 0	0	0.2	2	V
I <sub>OS</sub>	Short-circuit output current‡	4	V <sub>CC1</sub> = 5.5 V, V <sub>I</sub> = -5 V	-10	-20	-40	mA
I <sub>CC1</sub>	Supply current from V <sub>CC1</sub>	5	V <sub>CC1</sub> = 5.5 V, T <sub>A</sub> = 25°C		20	35	mA
I <sub>CC2</sub>	Supply current from V <sub>CC2</sub>		V <sub>CC2</sub> = 13.2 V, T <sub>A</sub> = 25°C		23	40	

† All typical values are at V<sub>CC1</sub> = 5 V, T<sub>A</sub> = 25°C.

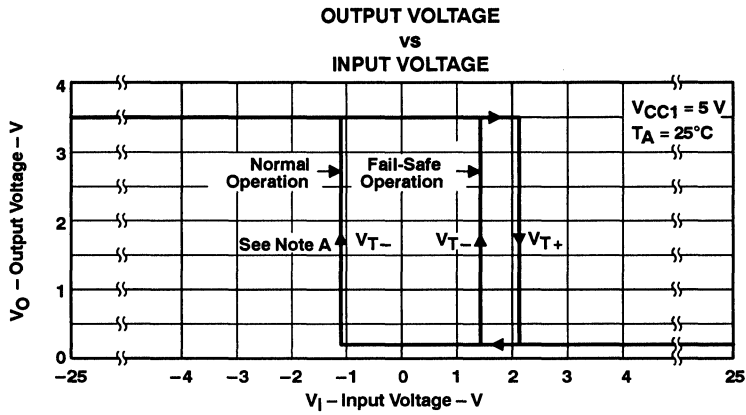
‡ Not more than one output should be shorted at a time.

**switching characteristics, V<sub>CC1</sub> = 5 V, T<sub>A</sub> = 25°C, N = 10**

PARAMETER		TEST FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	6	C <sub>L</sub> = 50 pF, R <sub>L</sub> = 390 Ω		11		ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output				8		ns
t <sub>TLH</sub>	Transition time, low-to-high-level output				7		ns
t <sub>THL</sub>	Transition time, high-to-low-level output				2.2		ns

TEXAS  
INSTRUMENTS

TYPICAL CHARACTERISTICS



NOTE A: For normal operation, the threshold controls are connected to  $V_{CC1}$ . For fail-safe operation, the threshold controls are open.

Figure 1

# SN75154 QUAD LINE RECEIVER

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## PARAMETER MEASUREMENT INFORMATION

### dc test circuits†

TEST TABLE

TEST	MEASURE	A	T	Y	VCC1	VCC2
Open-circuit input (fail safe)	V <sub>OH</sub>	Open	Open	I <sub>OH</sub>	4.5 V	Open
	V <sub>OH</sub>	Open	Open	I <sub>OH</sub>	Open	10.8 V
V <sub>T+</sub> min, V <sub>T-</sub> min (fail safe)	V <sub>OH</sub>	0.8 V	Open	I <sub>OH</sub>	5.5 V	Open
	V <sub>OH</sub>	0.8 V	Open	I <sub>OH</sub>	Open	13.2 V
V <sub>T+</sub> min (normal)	V <sub>OH</sub>	Note A	VCC1	I <sub>OH</sub>	5.5 V and T	Open
	V <sub>OH</sub>	Note A	VCC1	I <sub>OH</sub>	T	13.2 V
V <sub>IL</sub> max, V <sub>T+</sub> min (normal)	V <sub>OH</sub>	-3 V	VCC1	I <sub>OH</sub>	5.5 V and T	Open
	V <sub>OH</sub>	-3 V	VCC1	I <sub>OH</sub>	T	13.2 V
V <sub>IH</sub> min, V <sub>T+</sub> max, V <sub>T-</sub> max (fail safe)	V <sub>OL</sub>	3 V	Open	I <sub>OL</sub>	4.5 V	Open
	V <sub>OL</sub>	3 V	Open	I <sub>OL</sub>	Open	10.8 V
V <sub>IH</sub> min, V <sub>T+</sub> max (normal)	V <sub>OL</sub>	3 V	VCC1	I <sub>OL</sub>	4.5 V and T	Open
	V <sub>OL</sub>	3 V	VCC1	I <sub>OL</sub>	T	10.8 V
V <sub>T-</sub> max (normal)	V <sub>OL</sub>	Note B	VCC1	I <sub>OL</sub>	5.5 V and T	Open
	V <sub>OL</sub>	Note B	VCC1	I <sub>OL</sub>	T	13.2 V

NOTES: A. Momentarily apply -5 V, then 0.8 V.  
B. Momentarily apply 5 V, then GND.

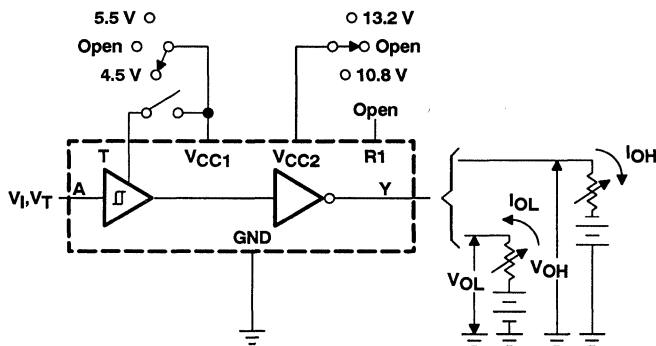


Figure 2. V<sub>IH</sub>, V<sub>IL</sub>, V<sub>T+</sub>, V<sub>T-</sub>, V<sub>OH</sub>, V<sub>OL</sub>

† Arrows indicate actual direction of current flow. Current into a terminal is a positive value.

PARAMETER MEASUREMENT INFORMATION

dc test circuits† (continued)

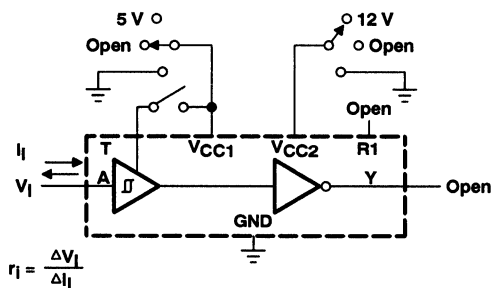


Figure 3. Input Resistance

TEST TABLE

T	VCC1	VCC2
Open	5 V	Open
Open	GND	Open
Open	Open	Open
VCC1	T and 5 V	Open
GND	GND	Open
Open	Open	12 V
Open	Open	GND
VCC1	T	12 V
VCC1	T	GND
VCC1	T	Open

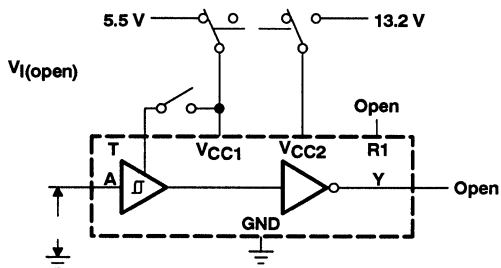
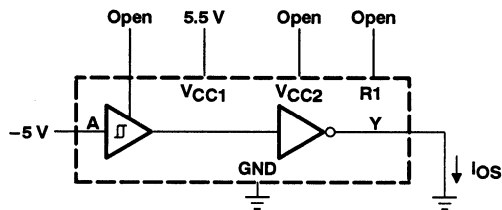


Figure 4. Input Voltage (open)

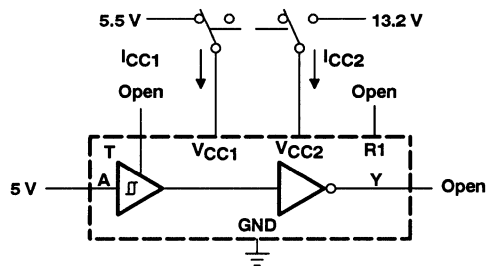
TEST TABLE

T	VCC1	VCC2
Open	5.5 V	Open
VCC1	5.5 V	Open
Open	Open	13.2 V
VCC1	T	13.2 V



Each output is tested separately.

Figure 5. Output Short-Circuit Current



All four line receivers are tested simultaneously.

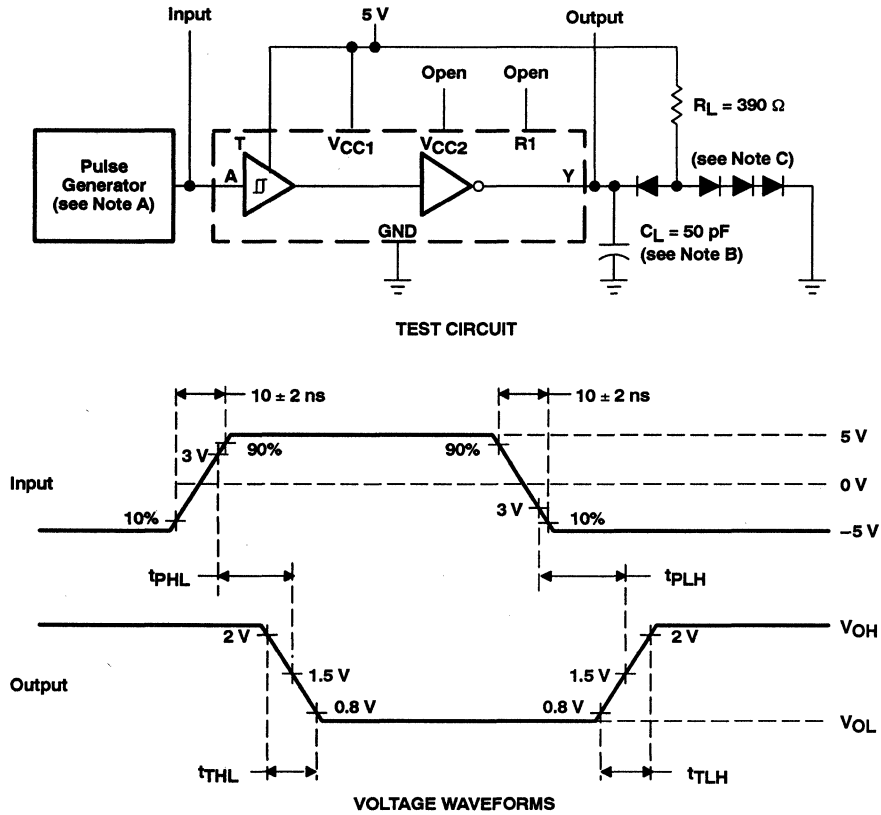
Figure 6. Supply Current

† Arrows indicate actual direction of current flow. Current into a terminal is a positive value.

# SN75154 QUAD LINE RECEIVER

SLLS083A - D899, NOVEMBER 1970 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $t_W \leq 200 \text{ ns}$ , duty cycle  $\leq 20\%$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064.

Figure 6. Test Circuit and Voltage Waveforms

# SN75155 LINE DRIVER AND RECEIVER

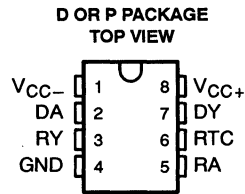
SLLS017B - D2951, JULY 1986 - REVISED MARCH 1993

- Meets EIA-Standard RS-232-C
- 10-mA Current Limited Output
- Wide Range of Supply Voltage  
V<sub>CC</sub> = 4.5 V to 15 V
- Low Power . . . 130 mW
- Built-In 5-V Regulator
- Response Control Provides:  
Input Threshold Shifting  
Input Noise Filtering
- Power-Off Output Resistance . . . 300 Ω Typ
- Driver Input TTL Compatible

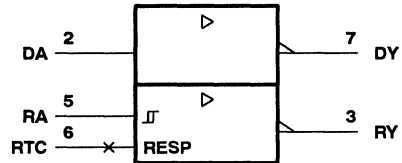
## description

The SN75155 is a monolithic line driver and receiver that is designed to satisfy the requirements of the standard interface between data terminal equipment and data communication equipment as defined by EIA standard RS-232-C. A response control input is provided for the receiver. A resistor or a resistor and a bias voltage can be connected between the response control input and ground to provide noise filtering. The driver used is similar to the SN75188. The receiver used is similar to the SN75189A.

The SN75155 is characterized for operation from 0°C to 70°C.

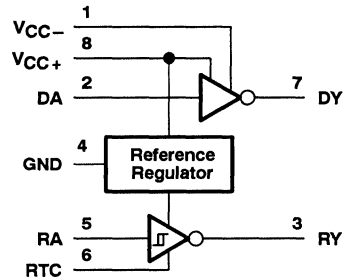


## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12

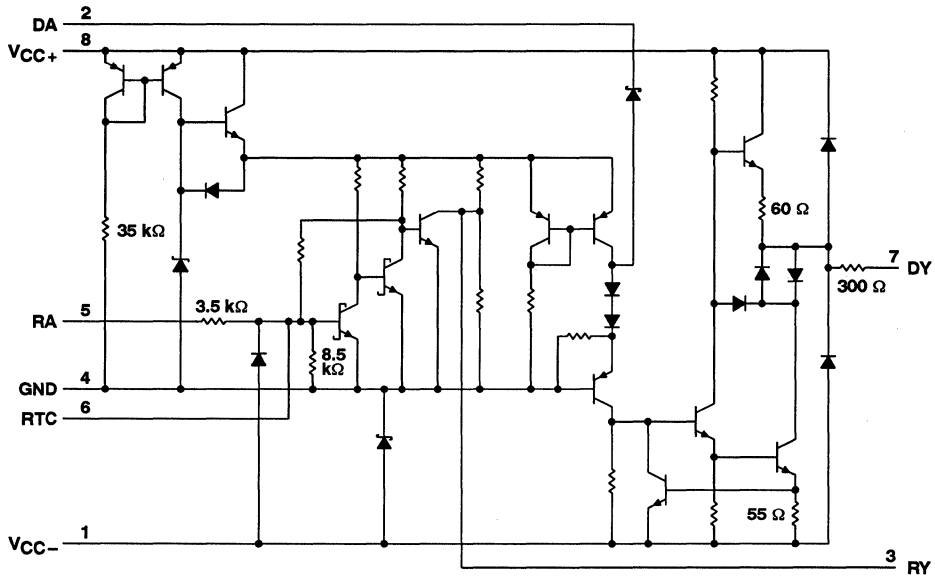
## logic diagram



# SN75155 LINE DRIVER AND RECEIVER

SLLS017B - D2951, JULY 1986 - REVISED MARCH 1993

## schematic



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	15 V
Supply voltage, $V_{CC-}$ (see Note 1)	-15 V
Input voltage range:	
Driver	-15 V to 15 V
Receiver	-30 V to 30 V
Output voltage range (driver)	-15 V to 15 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

# SN75155 LINE DRIVER AND RECEIVER

SLLS017B - D2951, JULY 1986 - REVISED MARCH 1993

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$	4.5	12	15	V
Supply voltage, $V_{CC-}$	-4.5	-12	-15	V
Input voltage, driver, $V_{I(D)}$			±15	V
Input voltage, receiver, $V_{I(R)}$	-25		25	V
High-level input voltage, driver, $V_{IH}$	2			V
Low-level voltage, driver, $V_{IL}$			0.8	V
Response control current			±5.5	mA
Output current, receiver, $I_{O(R)}$			24	mA
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

### total device

PARAMETER	TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$I_{CCH+}$ High-level supply current	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	$V_{I(D)} = 2\text{ V}$ , $V_{I(R)} = 2.3\text{ V}$ , Output open		6.3	8.1	mA
	$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$		9.1	11.9		
	$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$		10.4	14		
$I_{CCL+}$ Low-level supply current	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	$V_{I(D)} = 0.8\text{ V}$ , $V_{I(R)} = 0.6\text{ V}$ , Output open		2.5	3.4	mA
	$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$		3.7	5.1		
	$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$		4.1	5.6		
$I_{CC+}$ Supply current	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = 0$	$V_{I(R)} = 2.3\text{ V}$ , $V_{I(D)} = 0$		4.8	6.4	mA
	$V_{CC+} = 9\text{ V}$ , $V_{CC-} = 0$		6.7	9.1		
$I_{CCH-}$ High-level supply current	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	$V_{I(D)} = 2\text{ V}$ , $V_{I(R)} = 2.3\text{ V}$ , Output open		-2.4	-3.1	mA
	$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$		-3.9	-4.9		
	$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$		-4.8	-6.1		
$I_{CCL-}$ Low-level supply current	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	$V_{I(D)} = 0.8\text{ V}$ , $V_{I(R)} = 0.6\text{ V}$ , Output open		-0.2	-0.35	mA
	$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$		-0.25	-0.4		
	$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$		-0.27	-0.45		

† All typical values are at  $T_A = 25^\circ\text{C}$ .





# SN75155

## LINE DRIVER AND RECEIVER

SLLS017B – D2951, JULY 1986 – REVISED MARCH 1993

electrical characteristics over recommended operating free-air temperature range,  $V_{CC+} = 12\text{ V}$ ,  $V_{CC-} = -12\text{ V}$  (unless otherwise noted)

### driver section

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT		
$V_{OH}$	High-level output voltage	$V_{IL} = 0.8\text{ V}$ , $R_L = 3\text{ k}\Omega$	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	3.2	3.7		V		
			$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$	6.5	7.2				
			$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$	8.9	9.8				
$V_{OL}$	Low-level output voltage (see Note 2)	$V_{IH} = 2\text{ V}$ , $R_L = 3\text{ k}\Omega$	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$		-3.6	-3.2	V		
			$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$		-7.1	-6.4			
			$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$		-9.7	-8.8			
$I_{IH}$	High-level input current	$V_I = 7\text{ V}$				5	$\mu\text{A}$		
$I_{IL}$	Low-level input current	$V_I = 0$				-0.73	-1.2	mA	
$I_{OS(H)}$	High-level short-circuit output current	$V_I = 0.8\text{ V}$ , $V_O = 0$				-7	-12	-14.5	mA
$I_{OS(L)}$	Low-level short-circuit output current	$V_I = 2\text{ V}$ , $V_O = 0$				6.5	11.5	15	mA
$r_o$	Output resistance with power off	$V_O = -2\text{ V}$ to $2\text{ V}$				300			$\Omega$

### receiver section (see Figure 1)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT	
$V_{T+}$	Positive-going threshold voltage			1.2	1.9	2.3	V	
$V_{T-}$	Negative-going threshold voltage			0.6	0.95	1.2	V	
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )			0.6			V	
$V_{O(H)}$	High-level output voltage	$V_I = 0.6\text{ V}$ , $I_{OH} = 10\text{ }\mu\text{A}$	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	3.7	4.1	4.5	V	
			$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$	4.4	4.7	5.2		
		$V_I = 0.6\text{ V}$ , $I_{OH} = 0.4\text{ mA}$	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	3.1	3.4	3.8		
			$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$	3.6	4	4.5		
$V_{O(L)}$	Low-level output voltage	$V_I = 2.3\text{ V}$ , $I_{OL} = 24\text{ mA}$		0.2	0.3	V		
$I_{IH}$	High-level input current	$V_I = 2.5\text{ V}$		3.6	6.7	10	mA	
		$V_I = 3\text{ V}$		0.43	0.67	1	mA	
$I_{IL}$	Low-level input current	$V_I = -25\text{ V}$		-3.6	-6.7	-10	mA	
		$V_I = -3\text{ V}$		-0.43	-0.67	-1	mA	
$I_{OS}$	Short-circuit output current	$V_I = 0.6\text{ V}$				-2.8	-3.7	mA

† All typical values are at  $T_A = 25^\circ\text{C}$ .

NOTE 2: The algebraic limit system, in which the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic voltage levels only, e.g., if  $-8.8\text{ V}$  is the maximum, the typical value is a more negative value.



switching characteristics over recommended operating free-air temperature range,  $V_{CC+} = 5\text{ V}$ ,  $V_{CC-} = -5\text{ V}$ ,  $C_L = 50\text{ pF}$  (unless otherwise noted)

driver section (see Figure 2)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high level output	$R_L = 3\text{ k}\Omega$		250	480	ns
$t_{PHL}$ Propagation delay time, high-to-low level output			80	150	
$t_r$ Output rise time	$R_L = 3\text{ k}\Omega$		67	180	ns
	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 2500\text{ pF}$		2.4	3	$\mu\text{s}$
$t_f$ Output fall time	$R_L = 3\text{ k}\Omega$		48	160	ns
	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 2500\text{ pF}$		1.9	3	$\mu\text{s}$

receiver section (see Figure 3)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high level output	$R_L = 400\ \Omega$		175	245	ns
$t_{PHL}$ Propagation delay time, high-to-low level output			37	100	
$t_r$ Output rise time	$R_L = 400\ \Omega$		255	360	ns
$t_f$ Output fall time	$R_L = 400\ \Omega$		23	50	ns

† All typical values are at  $T_A = 25^\circ\text{C}$ .

**PARAMETER MEASUREMENT INFORMATION**

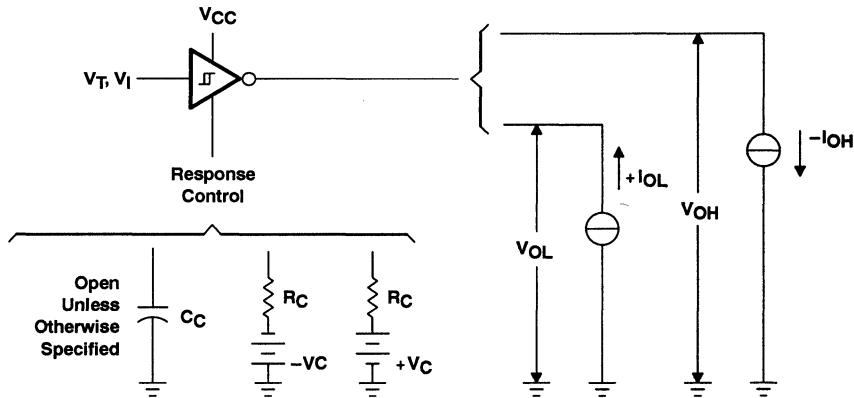


Figure 1. Receiver Section Test Circuit ( $V_{T+}$ ,  $V_{T-}$ ,  $V_{OH}$ ,  $V_{OL}$ )

# SN75155 LINE DRIVER AND RECEIVER

SLLS017B - D2951, JULY 1986 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

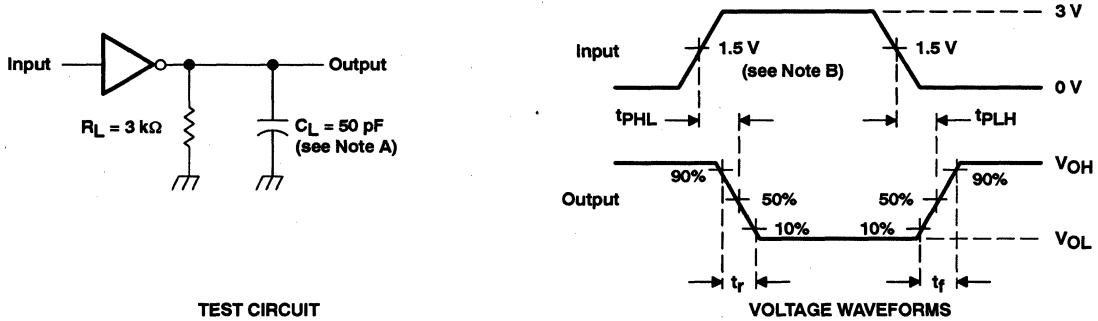


Figure 2. Driver Section Switching Test Circuit and Voltage Waveforms

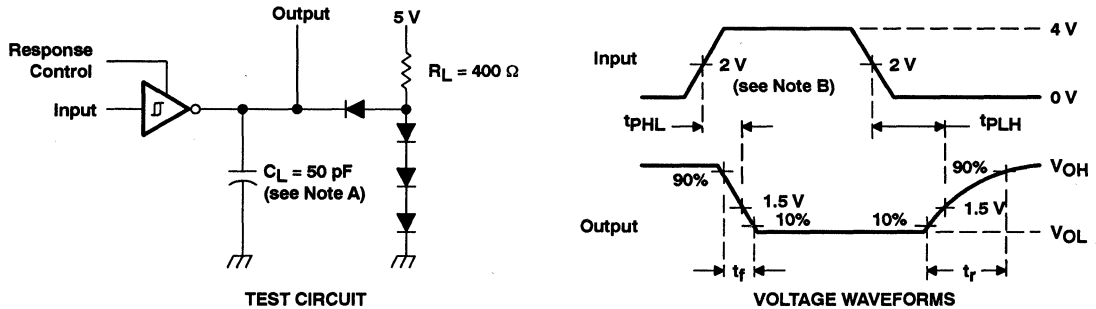


Figure 3. Receiver Section Switching Test Circuit and Voltage Waveforms

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input waveform is supplied by a generator with the following characteristics:  $Z_O = 50 \Omega$ ,  $t_W = 1 \mu s$ ,  $t_r \leq 10 ns$ ,  $t_f \leq 10 ns$ .

TYPICAL CHARACTERISTICS  
(DRIVER)

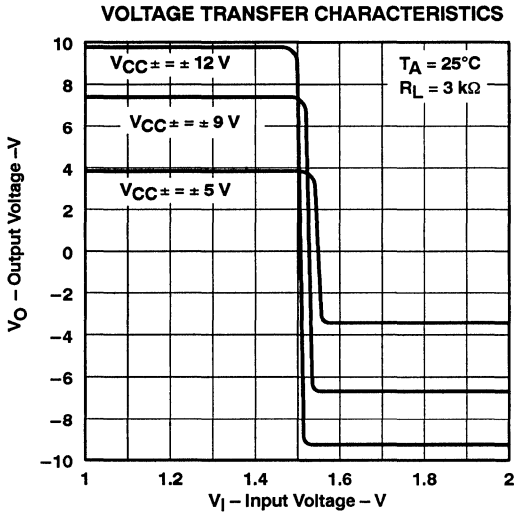


Figure 4

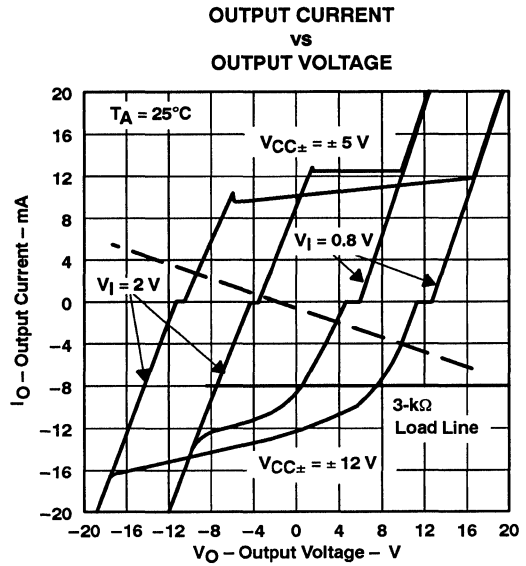


Figure 5

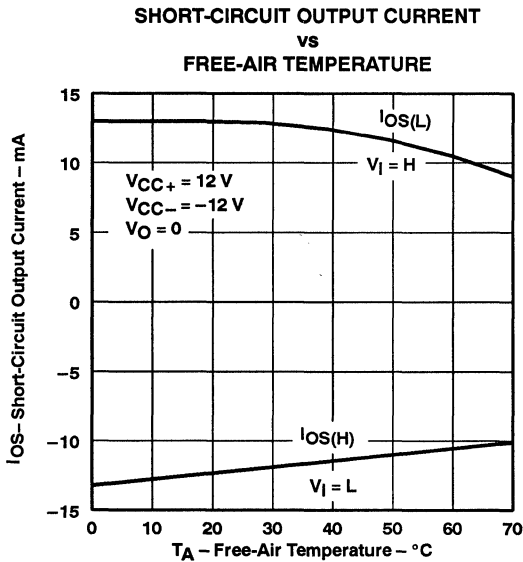


Figure 6

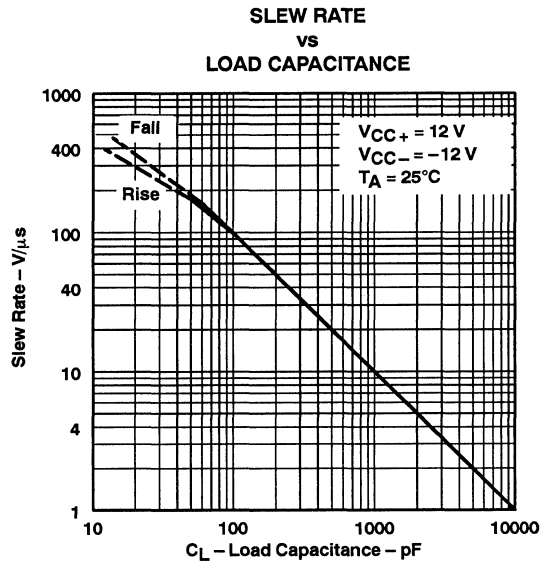


Figure 7

# SN75155 LINE DRIVER AND RECEIVER

SLLS017B - D2951, JULY 1986 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS (RECEIVER)

OUTPUT VOLTAGE  
vs  
INPUT VOLTAGE

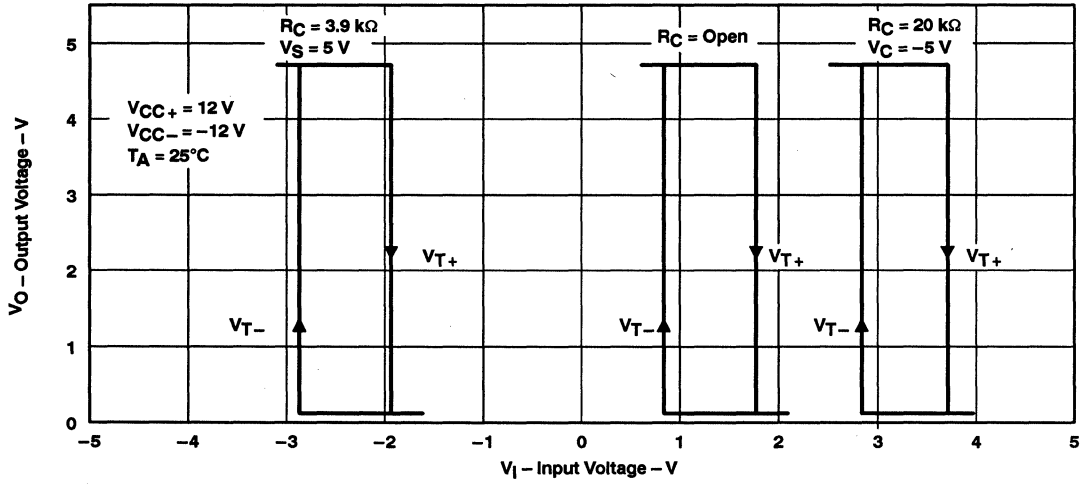


Figure 8

OUTPUT VOLTAGE  
vs  
INPUT VOLTAGE

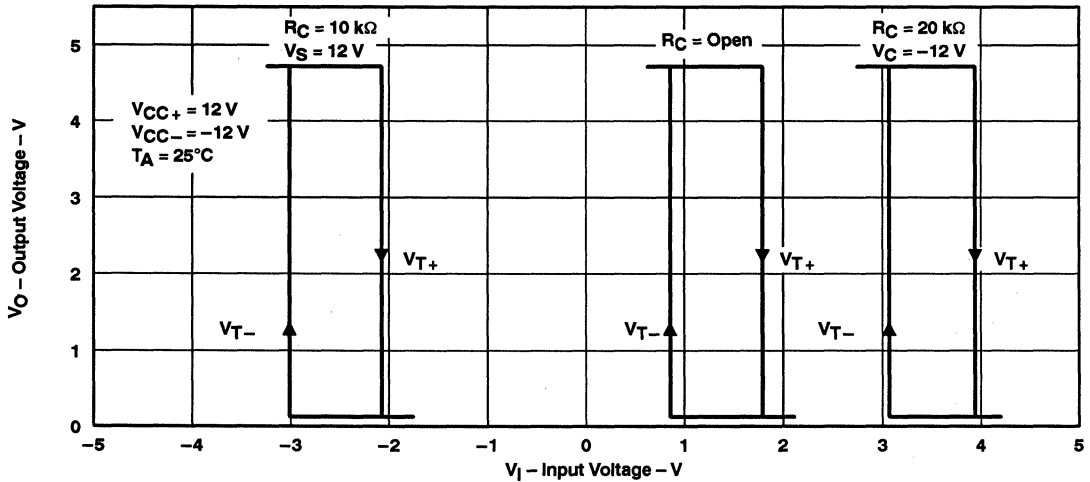


Figure 9

TYPICAL CHARACTERISTICS  
(RECEIVER)

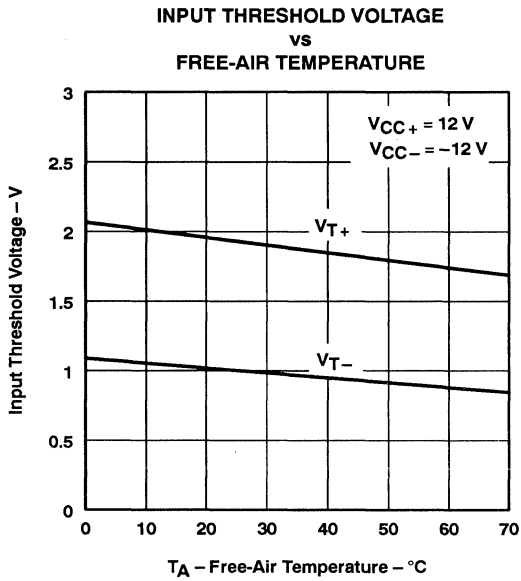


Figure 10

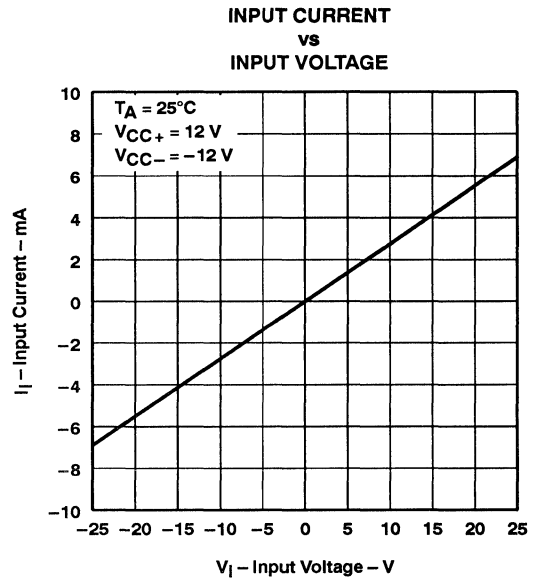


Figure 11

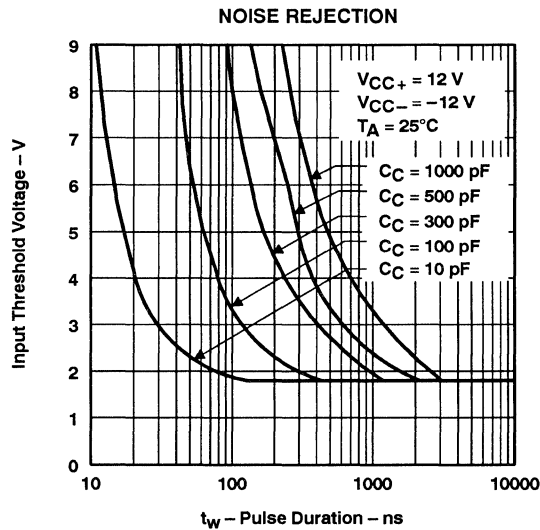


Figure 12



# SN75157 DUAL DIFFERENTIAL LINE RECEIVER

SLLS084A – D2300, SEPTEMBER 1980 – REVISED FEBRUARY 1993

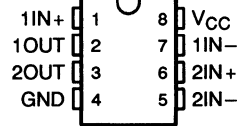
- Meets EIA Standards RS-422-A and RS-423-A
- Meets Federal Standards 1020 and 1030
- Operates From Single 5-V Power Supply
- Wide Common-Mode Voltage Range
- High Input Impedance
- TTL-Compatible Outputs
- High-Speed Schottky Circuitry
- 8-Pin Dual-In-Line Package
- Similar to  $\mu$ A9637AC Except for Corner  $V_{CC}$  and GND Positions

## description

The SN75157 is a dual differential line receiver designed to meet EIA Standards RS-422-A and RS-423-A and Federal Standards 1020 and 1030. It utilizes Schottky circuitry and has TTL-compatible outputs. The inputs are compatible with either a single-ended or a differential-line system. The device operates from a single 5-V power supply and is supplied in an 8-pin dual-in-line package and small outline package.

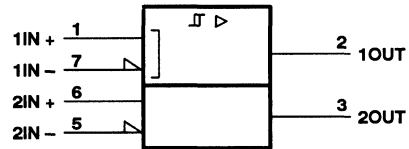
The SN75157 is characterized for operation from 0°C to 70°C.

D, P OR PS† PACKAGE  
(TOP VIEW)



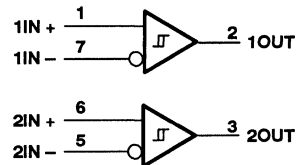
† The PS package is only available left-ended taped and reeled (order SN75157PSLE).

## logic symbol†

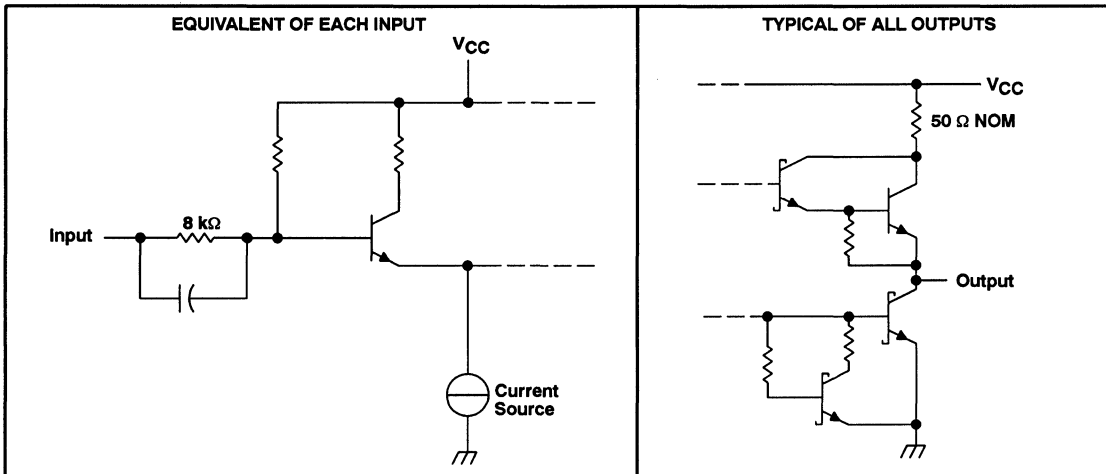


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## schematics of inputs and outputs



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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# SN75157 DUAL DIFFERENTIAL LINE RECEIVER

SLLS084A - D2300, SEPTEMBER 1980 - REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	-0.5 V to 7 V
Input voltage, $V_I$	$\pm 15$ V
Differential input voltage (see Note 2)	$\pm 15$ V
Output voltage range (see Note 1)	-0.5 V to 5.5 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to the network ground terminal.  
2. Differential input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW
PS	450 mW	3.6 mW/°C	288 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Operating free-air temperature, $T_A$	0	25	70	°C

## electrical characteristics over recommended ranges of supply voltage, common-mode input voltage, and operating free-air temperature (unless otherwise noted)<sup>†</sup>

PARAMETER	TEST CONDITIONS	MIN	TYP <sup>‡</sup>	MAX	UNIT
$V_T$ Threshold voltage ( $V_{T+}$ and $V_{T-}$ )		-0.2		0.2	V
	See Note 3	-0.4		0.4	
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )			70		mV
$V_{OH}$ High-level output voltage	$V_{ID} = 0.2$ V, $I_O = -1$ mA	2.5	3.5		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -0.2$ V, $I_O = 20$ mA		0.35	0.5	V
$I_I$ Input current	$V_{CC} = 0$ to 5.5 V, See Note 4	$V_I = 10$ V	1.1	3.25	mA
		$V_I = -10$ V	-1.6	-3.25	
$I_{OS}$ Short-circuit output current <sup>§</sup>	$V_O = 0$ , $V_{ID} = 0.2$ V	-40	-75	-100	mA
$I_{CC}$ Supply current	$V_{ID} = -0.5$ V, No load		35	50	mA

<sup>†</sup> The algebraic convention, where the less-positive (more-negative) limit is designated as minimum, is used in this data sheet for threshold levels only.

<sup>‡</sup> All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

<sup>§</sup> Only one output should be shorted at a time and duration of the short circuit should not exceed one second.

NOTES: 3. The expanded threshold parameter is tested with a 500- $\Omega$  resistor in series with each input.

4. The input not under test is grounded.



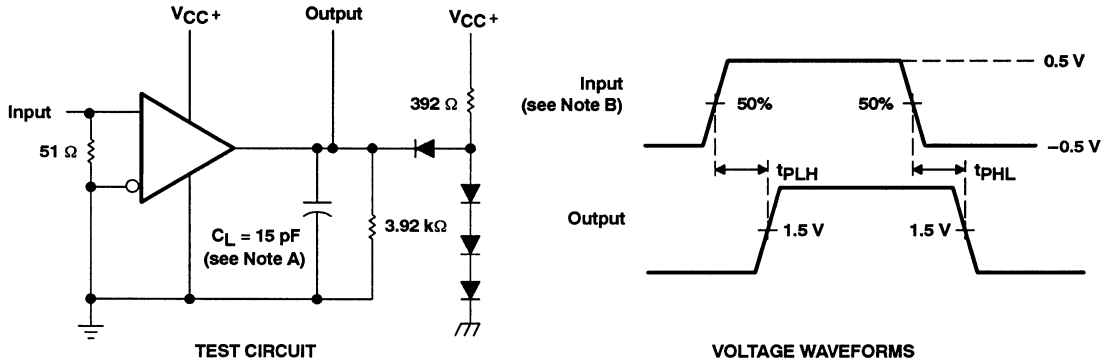
# SN75157 DUAL DIFFERENTIAL LINE RECEIVER

SLLS084A – D2300, SEPTEMBER 1980 – REVISED FEBRUARY 1993

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 15\text{ pF}$ , See Figure 1		15	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			13	25	ns

## PARAMETER MEASUREMENT INFORMATION



NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5\text{ ns}$ ,  $t_f \leq 5\text{ ns}$ ,  $PRR \leq 5\text{ MHz}$ , duty cycle = 50%.

**Figure 1. Test Circuit and Voltage Waveforms**

# SN75157 DUAL DIFFERENTIAL LINE RECEIVER

SLLS084A – D2300, SEPTEMBER 1980 – REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

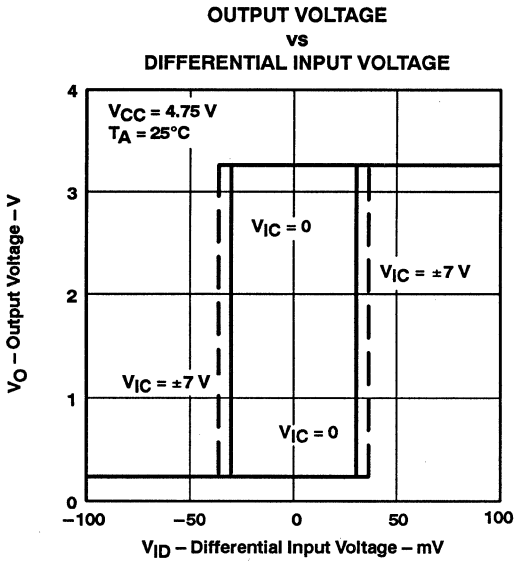


Figure 2

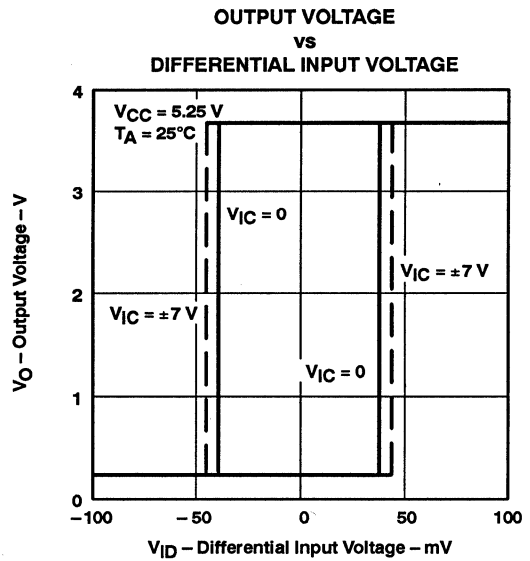


Figure 3

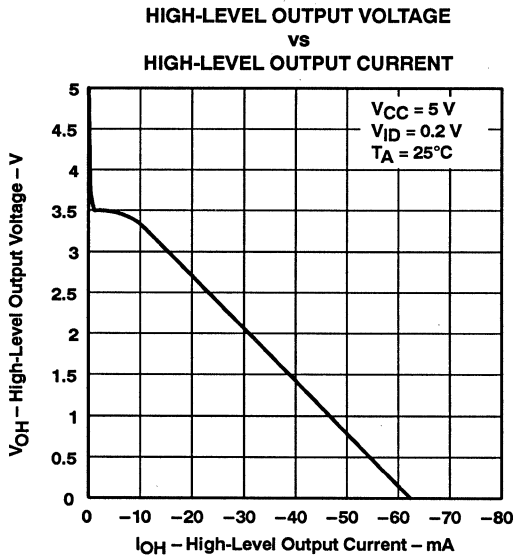


Figure 4

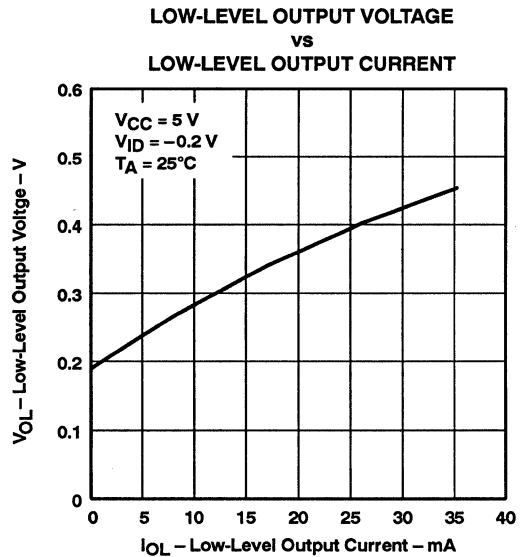


Figure 5

TYPICAL CHARACTERISTICS

SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE

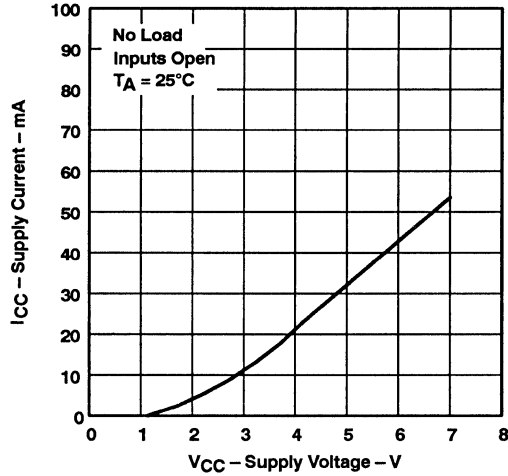


Figure 6

APPLICATION INFORMATION

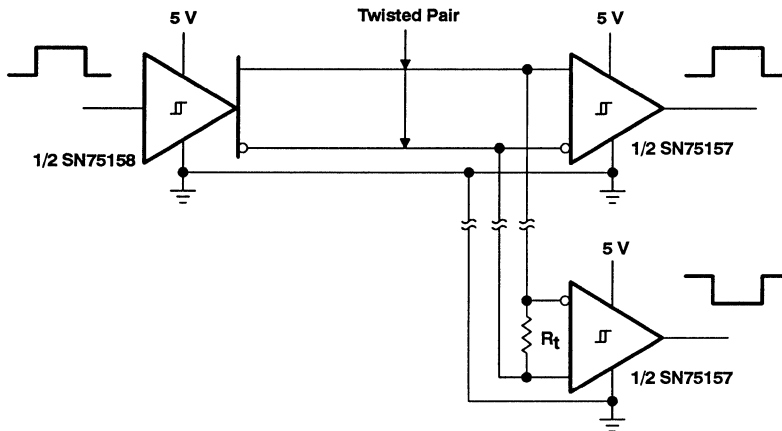


Figure 7. RS-422-A System Applications

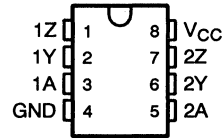


# SN75158 DUAL DIFFERENTIAL LINE DRIVER

SLLS085A - D2292, JANUARY 1977 - REVISED MARCH 1993

- Meets EIA Standard RS-422-A
- Single 5-V Supply
- Balanced-Line Operation
- TTL Compatible
- High Output Impedance in Power-Off Condition
- High-Current Active-Pullup Outputs
- Short-Circuit Protection
- Dual Channels
- Input Clamp Diodes

D, PST, OR P PACKAGE  
(TOP VIEW)



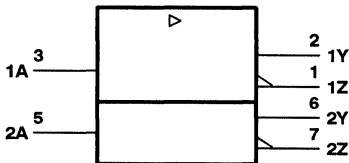
† The PS package is only available left-end taped and reeled, i.e., order SN75158PSLE.

## description

The SN75158 is a dual differential line driver designed to satisfy the requirements set by the EIA Standard RS-422-A interface specifications. The outputs provide complementary signals with high-current capability for driving balanced lines, such as twisted pair, at normal line impedance without high power dissipation. The output stages are TTL totem-pole outputs providing a high-impedance state in the power-off condition.

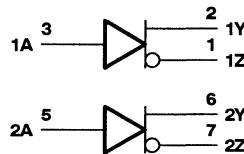
The SN75158 is characterized for operation from 0°C to 70°C.

## logic symbol†

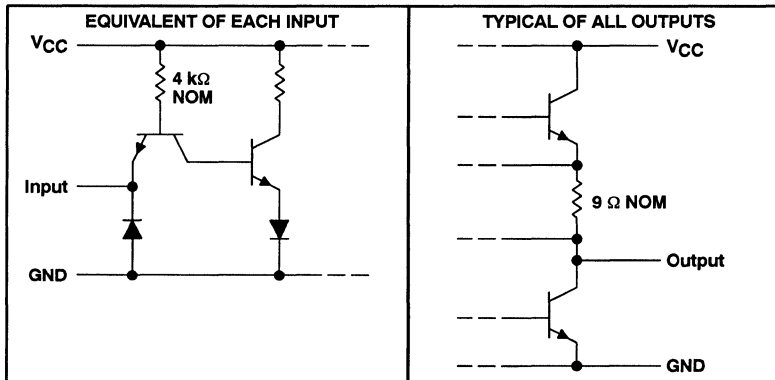


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## schematics of inputs and outputs



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

# SN75158 DUAL DIFFERENTIAL LINE DRIVER

SLLS085A - D2292, JANUARY 1977 - REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values except differential output voltage  $V_{OD}$  are with respect to network ground terminal.  $V_{OD}$  is at the Y output with respect to the Z output.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW
PS	450 mW	3.6 mW/°C	288 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$	0.8			V
High-level output current, $I_{OH}$	-40			mA
Low-level output current, $I_{OL}$	40			mA
Operating free-air temperature, $T_A$	0	70		°C

# SN75158 DUAL DIFFERENTIAL LINE DRIVER

SLLS085A – D2292, JANUARY 1977 – REVISED MARCH 1993

## electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP‡	MAX	UNIT
$V_{IK}$ Input clamp voltage	$V_{CC} = \text{MIN}$ , $I_I = -12 \text{ mA}$	-0.9	-1.5		V
$V_{OH}$ High-level output voltage	$V_{CC} = \text{MIN}$ , $V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ , $I_{OH} = -40 \text{ mA}$	2.4	3		V
$V_{OL}$ Low-level output voltage	$V_{CC} = \text{MIN}$ , $V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ , $I_{OL} = 40 \text{ mA}$		0.2	0.4	V
$ V_{OD1} $ Differential output voltage	$V_{CC} = \text{MAX}$ , $I_O = 0$		3.5	$2 V_{OD2}$	V
$ V_{OD2} $ Differential output voltage	$V_{CC} = \text{MIN}$	2	3		V
$\Delta V_{OD}$ Change in magnitude of differential output voltage§	$V_{CC} = \text{MIN}$		$\pm 0.02$	$\pm 0.4$	V
$V_{OC}$ Common-mode output voltage¶	$V_{CC} = \text{MAX}$		1.8	3	V
	$V_{CC} = \text{MIN}$		1.5	3	V
$\Delta V_{OC}$ Change in magnitude of common-mode output voltage§	$V_{CC} = \text{MIN}$ or $\text{MAX}$		$\pm 0.02$	$\pm 0.4$	V
$I_O$ Output current with power off	$V_{CC} = 0$	$V_O = 6 \text{ V}$	0.1	100	$\mu\text{A}$
		$V_O = -0.25 \text{ V}$	-0.1	-100	
		$V_O = -0.25 \text{ to } 6 \text{ V}$		$\pm 100$	
$I_I$ Input current at maximum input voltage	$V_{CC} = \text{MAX}$ , $V_I = 5.5 \text{ V}$			1	mA
$I_{IH}$ High-level input current	$V_{CC} = \text{MAX}$ , $V_I = 2.4 \text{ V}$			40	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_{CC} = \text{MAX}$ , $V_I = 0.4 \text{ V}$		-1	-1.6	mA
$I_{OS}$ Short-circuit output current#	$V_{CC} = \text{MAX}$	-40	-90	-150	mA
$I_{CC}$ Supply current (both drivers)	$V_{CC} = \text{MAX}$ , No load, $T_A = 25^\circ\text{C}$		37	50	mA

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$  except for  $V_{OC}$ , for which  $V_{CC}$  is as stated under test conditions.

§  $\Delta V_{OD}$  and  $\Delta|V_{OC}|$  are the changes in magnitudes of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

¶ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

# Only one output should be shorted at a time, and duration of the short circuit should not exceed one second.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	See Figure 2, Termination A		16	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			10	20	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	See Figure 2, Termination B		13	20	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			9	15	ns
$t_{TLH}$ Transition time, low-to-high-level output	See Figure 2, Termination A		4	20	ns
$t_{TLH}$ Transition time, high-to-low-level output			4	20	ns
Overshoot factor	See Figure 2, Termination C			10%	



# SN75158 DUAL DIFFERENTIAL LINE DRIVER

SLLS085A - D2292, JANUARY 1977 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

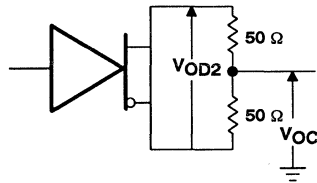
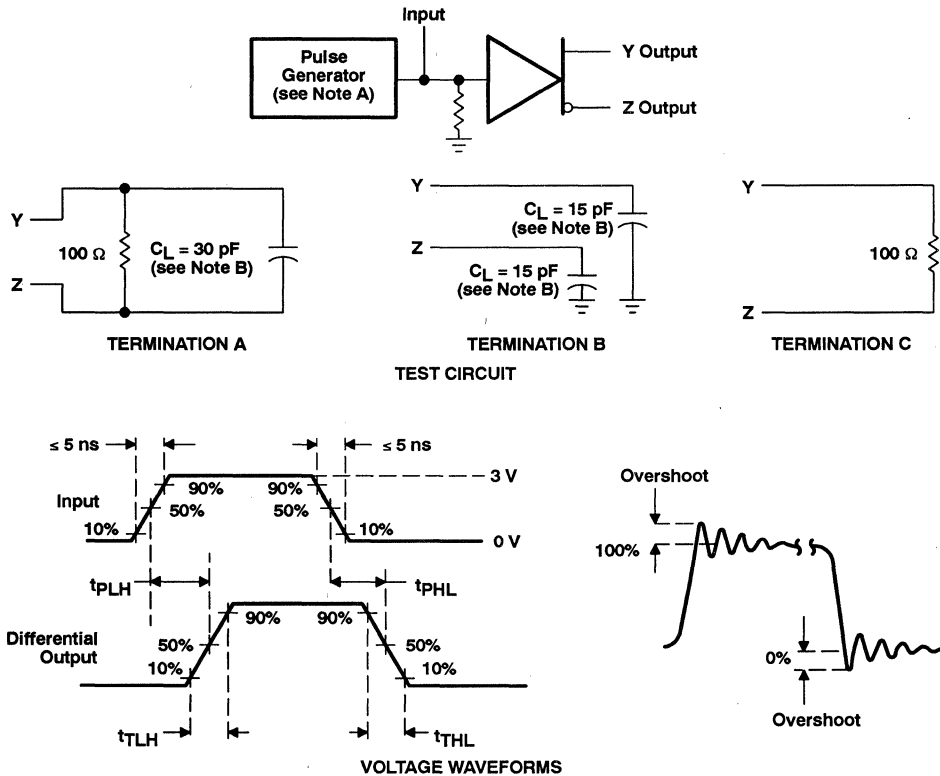


Figure 1. Differential and Common-Mode Output Voltages



NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $t_w = 25 \text{ ns}$ ,  $\text{PRR} \leq 10 \text{ MHz}$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 2. Test Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

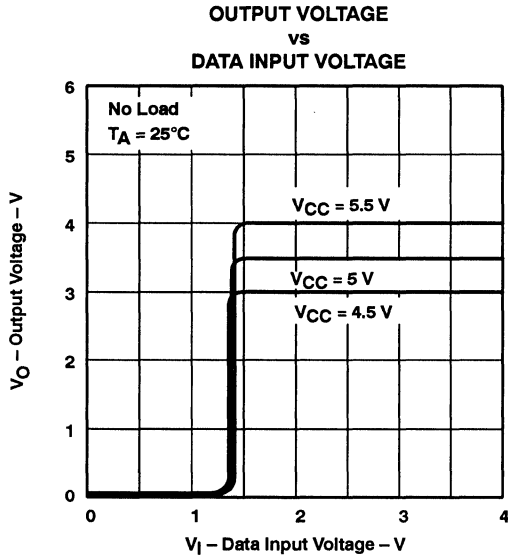


Figure 3

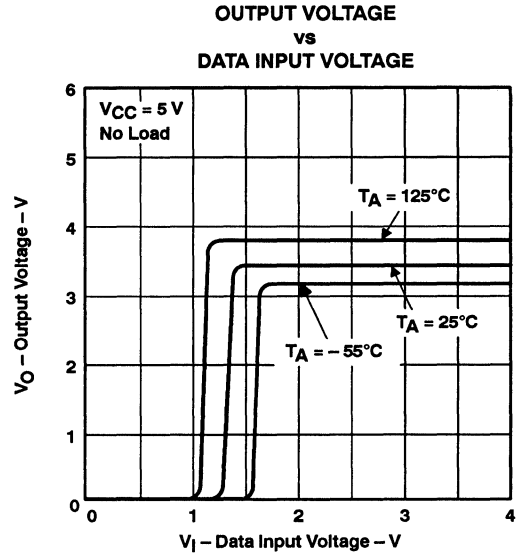


Figure 4

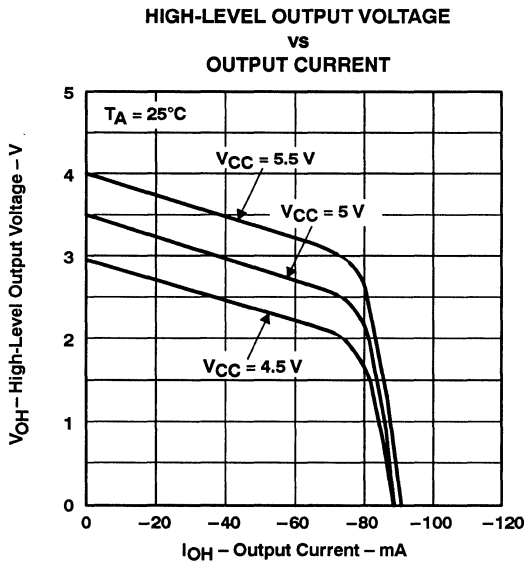


Figure 5

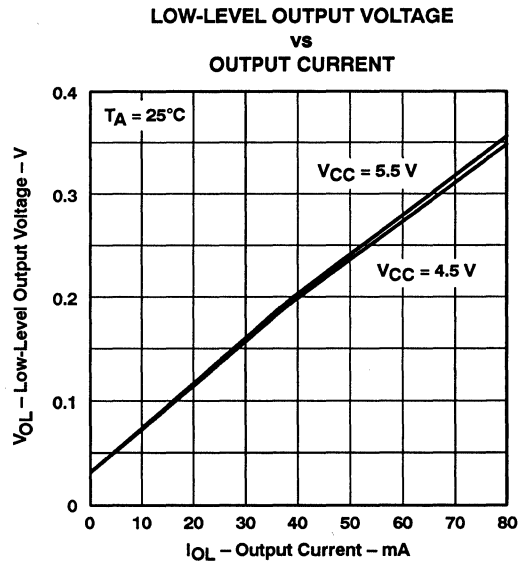


Figure 6

# SN75158 DUAL DIFFERENTIAL LINE DRIVER

SLLS085A - D2292, JANUARY 1977 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

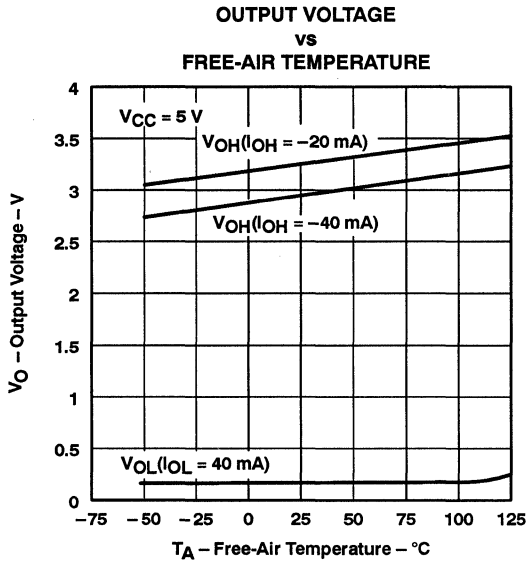


Figure 7

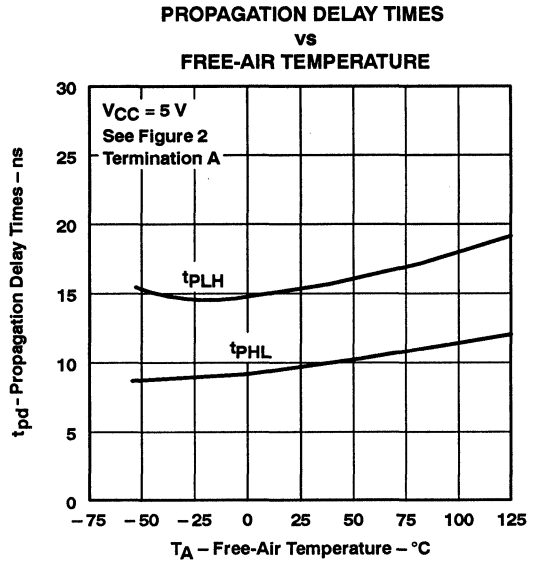


Figure 8

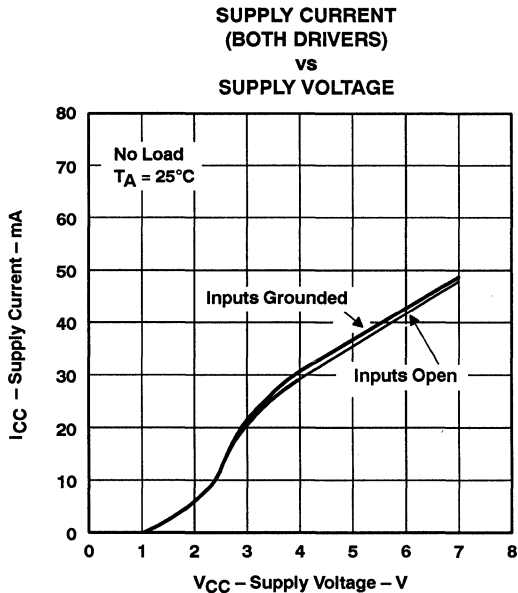


Figure 9

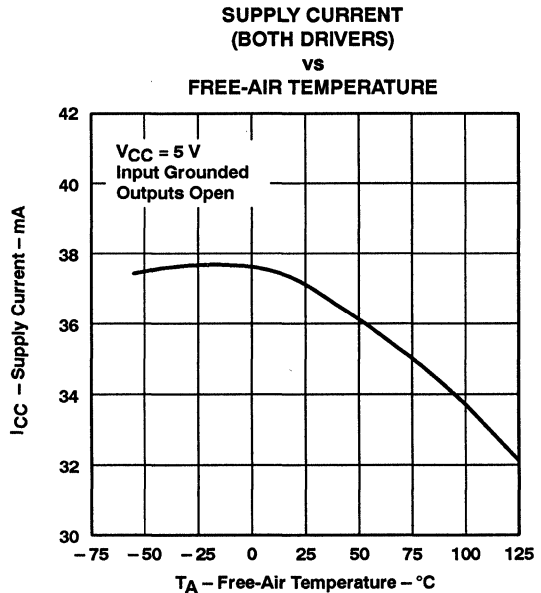


Figure 10

TYPICAL CHARACTERISTICS

SUPPLY CURRENT  
(BOTH DRIVERS)  
vs  
FREQUENCY

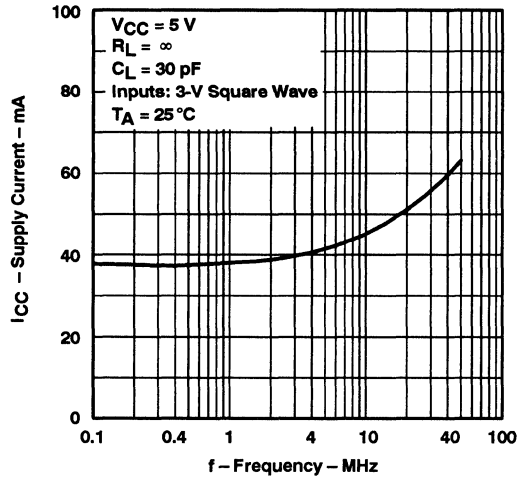


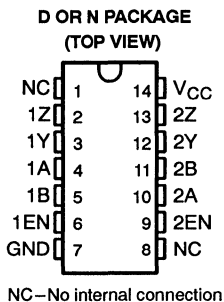
Figure 11



# SN75159 DUAL DIFFERENTIAL LINE DRIVER WITH 3-STATE OUTPUTS

SLLS088A – D2325, JANUARY 1977 – REVISED FEBRUARY 1993

- Meets EIA Standard RS-422-A
- Single 5-V Supply
- Balanced Line Operation
- TTL Compatible
- High-Impedance Output State for Party-Line Applications
- High-Current Active-Pullup Outputs
- Short-Circuit Protection
- Dual Channels
- Clamp Diodes at Inputs

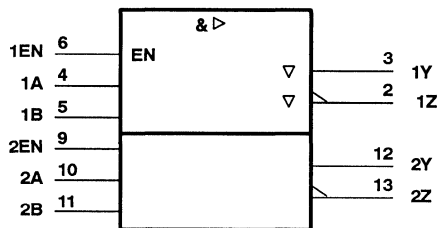


## description

The SN75159 dual differential line driver with 3-state outputs is designed to provide all the features of the SN75158 line driver with the added feature of driver output controls. There is an individual control for each driver. When the output control is low, the associated outputs are in a high-impedance state and the outputs can neither drive nor load the bus. This permits many devices to be connected together on the same transmission line for party-line applications.

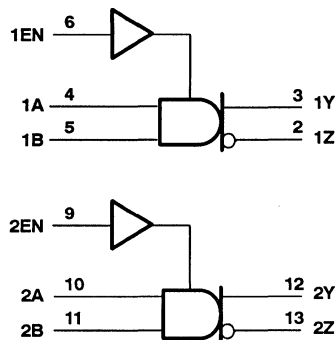
The SN75159 is characterized for operation from 0°C to 70°C.

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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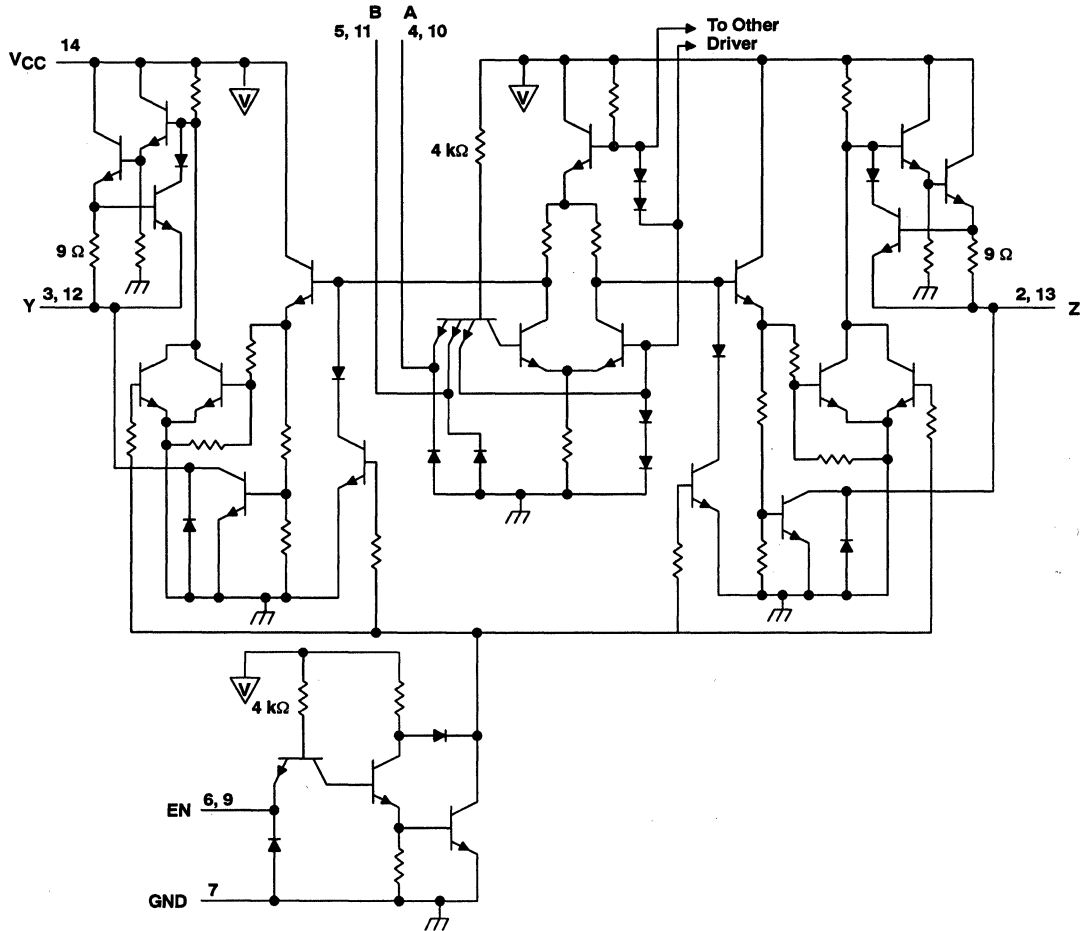
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**SN75159**  
**DUAL DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

**schematic (each driver)**



△ ... VCC bus  
 Resistor values shown are nominal.

**SN75159**  
**DUAL DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS088A – D2325, JANUARY 1977 – REVISED FEBRUARY 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Off-state voltage applied to open-collector outputs	12 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values except differential output voltage  $V_{OD}$  are with respect to the network ground terminal.  $V_{OD}$  is at the Y output with respect to the Z output.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output voltage, $I_{OH}$			-40	mA
Low-level output current, $I_{OL}$			40	mA
Operating free-air temperature, $T_A$	0		70	°C





**SN75159**  
**DUAL DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

**electrical characteristics over operating free-air temperature range (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$V_{CC} = 4.75 \text{ V}$ ,	$I_I = -12 \text{ mA}$	-0.9	-1.5		V
$V_{OH}$	High-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IH} = 2 \text{ V}$ ,	$V_{IL} = 0.8 \text{ V}$ , $I_{OH} = -40 \text{ mA}$	2.4	3		V
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IH} = 2 \text{ V}$ ,	$V_{IL} = 0.8 \text{ V}$ , $I_{OL} = 40 \text{ mA}$		0.25	0.4	V
$V_{OK}$	Output clamp voltage	$V_{CC} = 5.25 \text{ V}$ ,	$I_O = -40 \text{ mA}$	-1.1	-1.5		V
$V_O$	Output voltage	$V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$ ,	$I_O = 0$	0		6	V
$ V_{OD1} $	Differential output voltage	$V_{CC} = 5.25 \text{ V}$ ,	$I_O = 0$		3.5	$\frac{2V_{OD}}{2}$	V
$ V_{OD2} $	Differential output voltage	$V_{CC} = 4.75 \text{ V}$		2	3		V
$\Delta V_{OD} $	Change in magnitude of differential output voltage‡	$V_{CC} = 4.75 \text{ V}$	$R_L = 100 \Omega$ , See Figure 1	$\pm 0.02$		$\pm 0.4$	V
$V_{OC}$	Common-mode output voltage§	$V_{CC} = 5.25 \text{ V}$		1.8		3	V
		$V_{CC} = 4.75 \text{ V}$		1.5		3	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage‡	$V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$		$\pm 0.01$		$\pm 0.4$	V
$I_O$	Output current with power off	$V_{CC} = 0$	$V_O = 6 \text{ V}$		0.1	100	$\mu\text{A}$
			$V_O = -0.25 \text{ V}$		-0.1	-100	
			$V_O = -0.25 \text{ V to } 6 \text{ V}$			$\pm 100$	
$I_{OZ}$	Off-state (high-impedance state) output current	$V_{CC} = 5.25 \text{ V}$ , Output controls at 0.8 V	$T_A = 25^\circ\text{C}$	$V_O = 0 \text{ to } V_{CC}$		$\pm 10$	$\mu\text{A}$
			$T_A = 70^\circ\text{C}$	$V_O = 0$		-20	
				$V_O = 0.4 \text{ V}$		$\pm 20$	
				$V_O = 2.4 \text{ V}$		$\pm 20$	
				$V_O = V_{CC}$		20	
$I_I$	Input current at maximum input voltage	$V_{CC} = 5.25 \text{ V}$ ,	$V_I = 5.5 \text{ V}$			1	$\text{mA}$
$I_{IH}$	High-level input current	$V_{CC} = 5.25 \text{ V}$ ,	$V_I = 2.4 \text{ V}$			40	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_{CC} = 5.25 \text{ V}$ ,	$V_I = 0.4 \text{ V}$		-1	-1.6	$\text{mA}$
$I_{OS}$	Short-circuit output current¶	$V_{CC} = 5.25 \text{ V}$		-40	-90	-150	$\text{mA}$
$I_{CC}$	Supply current (both drivers)	$V_{CC} = 5.25 \text{ V}$ , No load,	Inputs grounded, $T_A = 25^\circ\text{C}$		47	65	$\text{mA}$

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$  except for  $V_{OC}$ , for which  $V_{CC}$  is as stated under test conditions.

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitudes of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

§ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to GND, is called output offset voltage,  $V_{OS}$ .

¶ Only one output should be shorted at a time, and duration of the short circuit should not exceed one second.



**SN75159**  
**DUAL DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS088A – D2325, JANUARY 1977 – REVISED FEBRUARY 1993

**switching characteristics over operating free-air temperature range,  $V_{CC} = 5\text{ V}$**

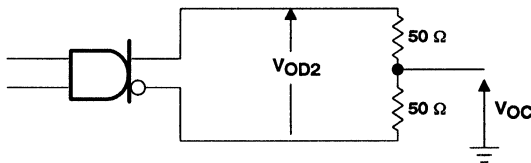
PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 30\text{ pF}$ , $R_L = 100\ \Omega$ , See Figure 2, Termination A		16	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			11	20	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 15\text{ pF}$ , See Figure 2, Termination B		13	20	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			9	15	ns
$t_{TLH}$ Transition time, low-to-high-level output	$C_L = 30\text{ pF}$ , $R_L = 100\ \Omega$ , See Figure 2, Termination A		4	20	ns
$t_{THL}$ Transition time, high-to-low-level output			4	20	ns
$t_{PZH}$ Output enable time to high level	$C_L = 30\text{ pF}$ , $R_L = 180\ \Omega$ , See Figure 3		7	20	ns
$t_{PZL}$ Output enable time to low level	$C_L = 30\text{ pF}$ , $R_L = 250\ \Omega$ , See Figure 4		14	40	ns
$t_{PHZ}$ Output disable time from high level	$C_L = 30\text{ pF}$ , $R_L = 180\ \Omega$ , See Figure 3		10	30	ns
$t_{PLZ}$ Output disable time from low level	$C_L = 30\text{ pF}$ , $R_L = 250\ \Omega$ , See Figure 4		17	35	ns
Overshoot factor	$R_L = 100\ \Omega$ , See Figure 2, Termination C			10%	

† All typical values are at  $T_A = 25^\circ\text{C}$ .

**SYMBOL EQUIVALENTS**

DATA SHEET PARAMETER	RS-422-A
$V_O$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_o$
$ V_{OD2} $	$V_t$
$\Delta V_{OD} $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $
$\Delta V_{OC} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $
$I_O$	$ I_{xa} ,  I_{xb} $

**PARAMETER MEASUREMENT INFORMATION**

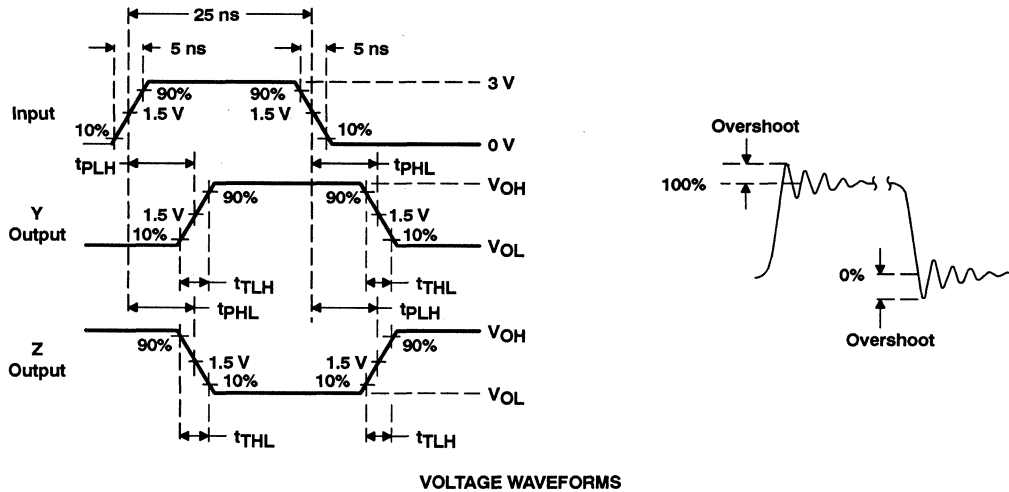
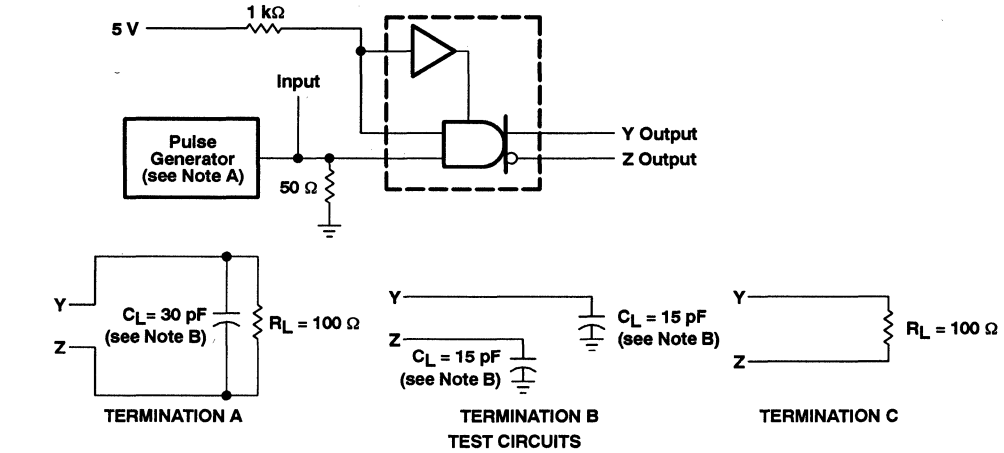


**Figure 1. Differential and Common-Mode Output Voltages**

**SN75159**  
**DUAL DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 10 \text{ MHz}$ .  
 B.  $C_L$  includes probe and jig capacitance.

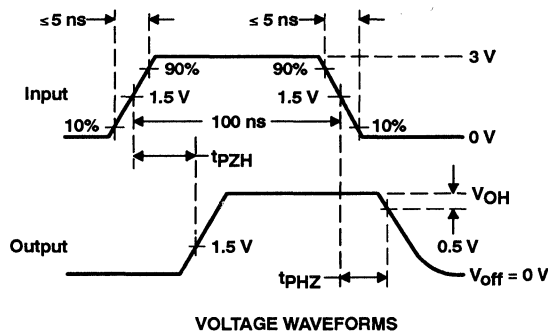
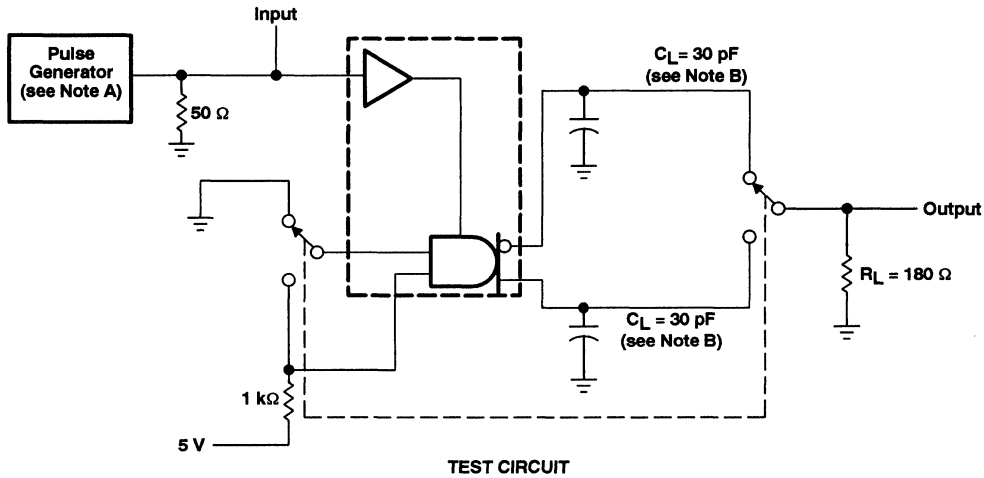
**Figure 2. Test Circuits, Voltage Waveforms, and Overshoot Factor**



**SN75159**  
**DUAL DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



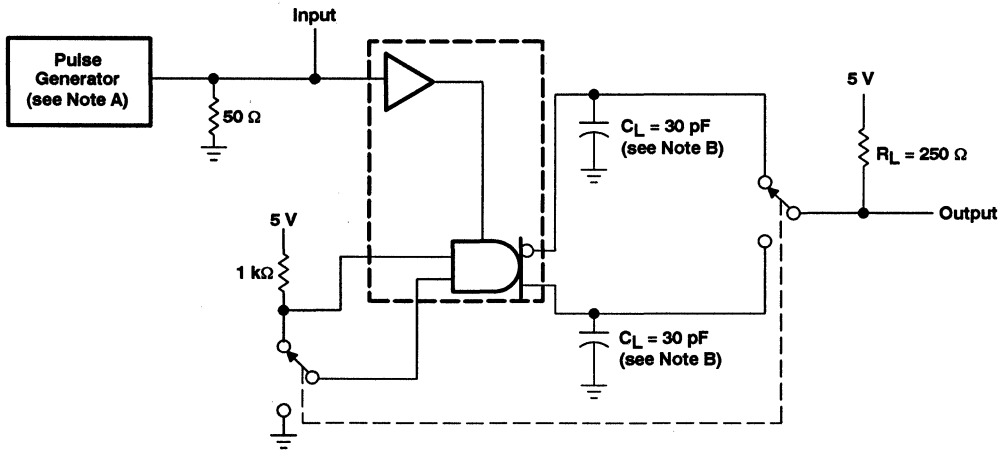
- NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50\ \Omega$ ,  $PRR \leq 500\text{ kHz}$ .  
 B.  $C_L$  includes probe and jig capacitance.

**Figure 3. Test Circuit and Voltage Waveforms**

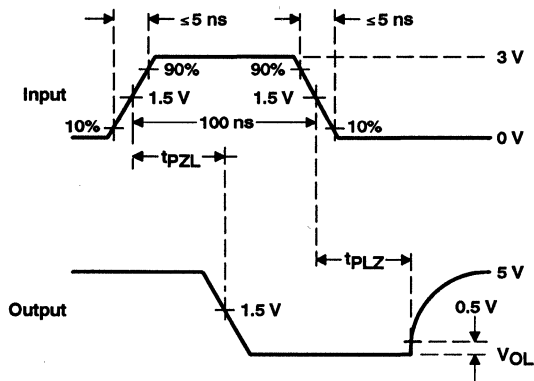
**SN75159  
DUAL DIFFERENTIAL LINE DRIVER  
WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



**TEST CIRCUIT**



**VOLTAGE WAVEFORMS**

NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ .  
B.  $C_L$  includes probe and jig capacitance.

**Figure 4. Test Circuit and Voltage Waveform**

SN75159  
**DUAL DIFFERENTIAL LINE DRIVER  
 WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

**TYPICAL CHARACTERISTICS**

**OUTPUT VOLTAGE  
 vs  
 DATA INPUT VOLTAGE**

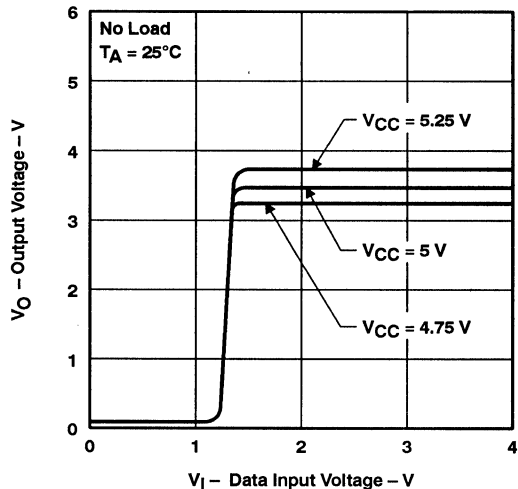


Figure 5

**OUTPUT VOLTAGE  
 vs  
 DATA INPUT VOLTAGE**

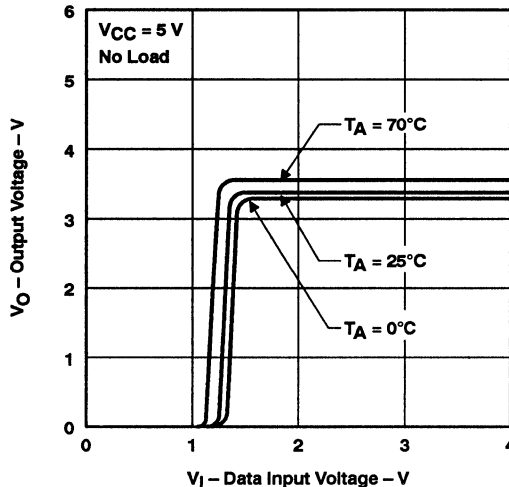


Figure 6

**OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE**

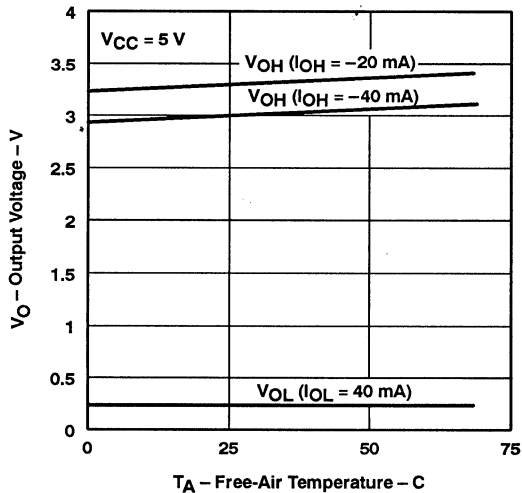


Figure 7

**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT**

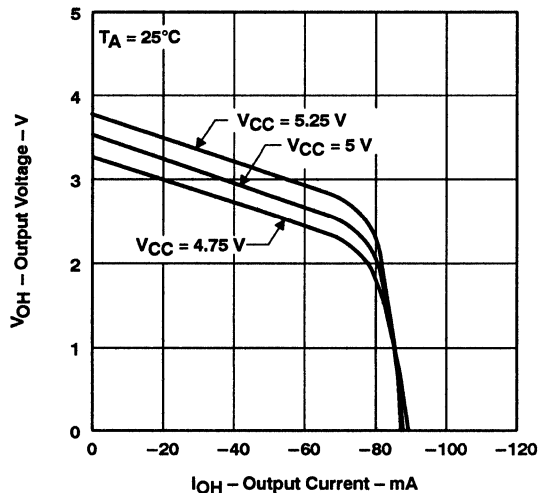


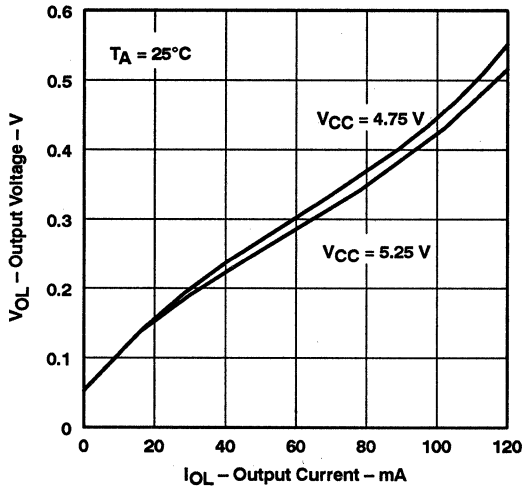
Figure 8

**SN75159  
DUAL DIFFERENTIAL LINE DRIVER  
WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

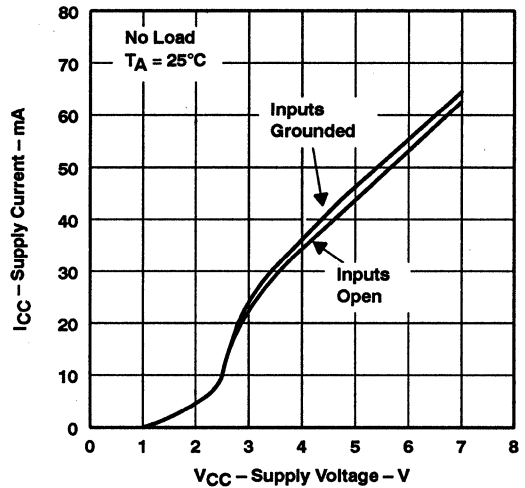
**TYPICAL CHARACTERISTICS**

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**



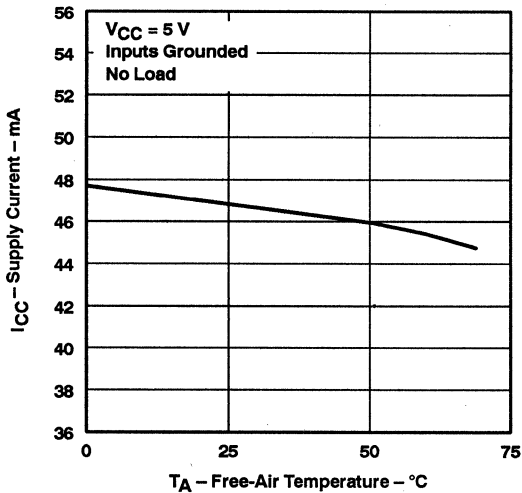
**Figure 9**

**SUPPLY CURRENT  
(BOTH DRIVERS)  
vs  
SUPPLY VOLTAGE**



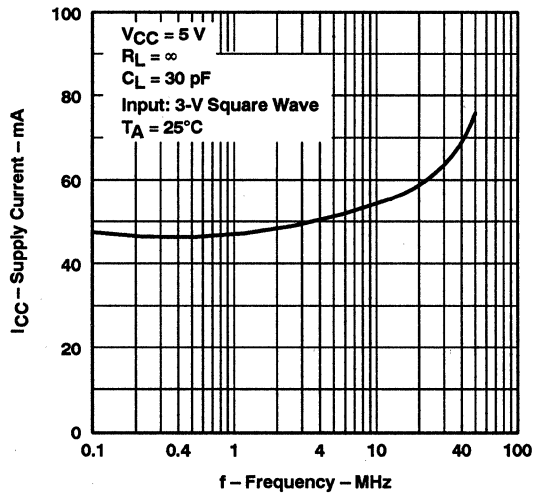
**Figure 10**

**SUPPLY CURRENT  
(BOTH DRIVERS)  
vs  
FREE-AIR TEMPERATURE**



**Figure 11**

**SUPPLY CURRENT  
(BOTH DRIVERS)  
vs  
FREQUENCY**



**Figure 12**



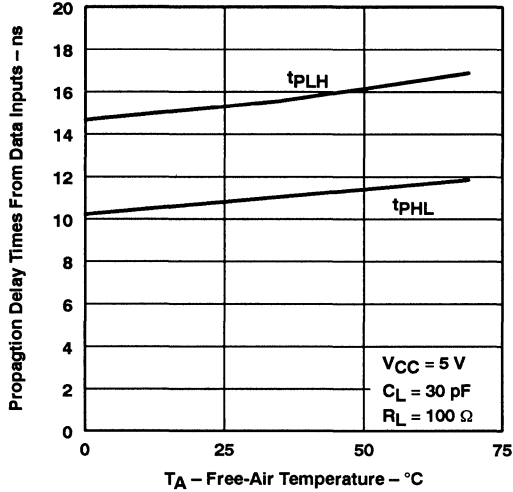
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**SN75159**  
**DUAL DIFFERENTIAL LINE DRIVER**  
**WITH 3-STATE OUTPUTS**

SLLS088A - D2325, JANUARY 1977 - REVISED FEBRUARY 1993

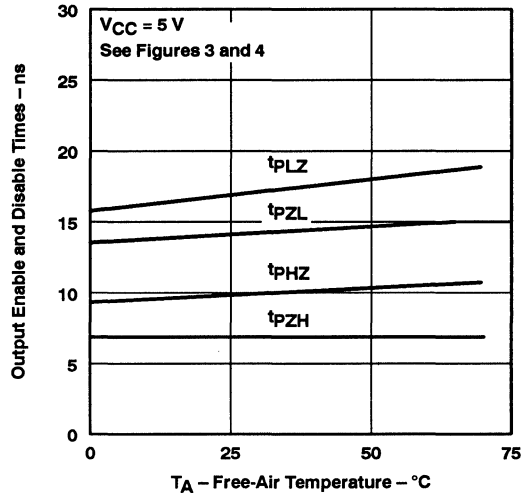
**TYPICAL CHARACTERISTICS**

**PROPAGATION DELAY TIMES  
 FROM DATA INPUTS  
 vs  
 FREE-AIR TEMPERATURE**



**Figure 13**

**OUTPUT ENABLE AND DISABLE TIMES  
 vs  
 FREE-AIR TEMPERATURE**



**Figure 14**



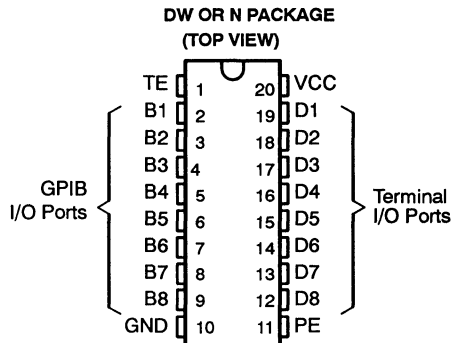


# SN75160B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS004A - D2525, OCTOBER 1985 - REVISED FEBRUARY 1993

## MEETS IEEE STANDARD 488-1978 (GPIB)

- 8-Channel Bidirectional Transceiver
- Power-Up/Power-Down Protection (Glitch Free)
- High-Speed, Low-Power Schottky Circuitry
- Low-Power Dissipation . . . 72 mW Max Per Channel
- Fast Propagation Times . . . 22 ns Max
- High-Impedance PNP Inputs
- Receiver Hysteresis . . . 650 mV Typ
- Open-Collector Driver Output Option
- No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )



### description

The SN75160B 8-channel general-purpose interface bus transceiver is a monolithic, high-speed, low-power Schottky device designed for two-way data communications over single-ended transmission lines. It is designed to meet the requirements of IEEE Standard 488-1978. The transceiver features driver outputs that can be operated in either the passive-pullup or 3-state mode. If talk enable (TE) is high, these ports have the characteristics of passive-pullup outputs when pullup enable (PE) is low, and of 3-state outputs when PE is high. Taking TE low places these ports in the high-impedance state. The driver outputs are designed to handle loads up to 48 mA of sink current.

Output glitches during power up and power down are eliminated by an internal circuit that disables both the bus and receiver outputs. The outputs do not load the bus when  $V_{CC} = 0$ . When combined with the SN75161B or SN75162B management bus transceivers, the pair provides the complete 16-wire interface for the IEEE-488 bus.

The SN75160B is characterized for operation from 0°C to 70°C.

### Function Tables

EACH DRIVER			
INPUTS			OUTPUT
D	TE	PE	B
H	H	H	H
L	H	X	L
H	X	L	Z†
X	L	X	Z†

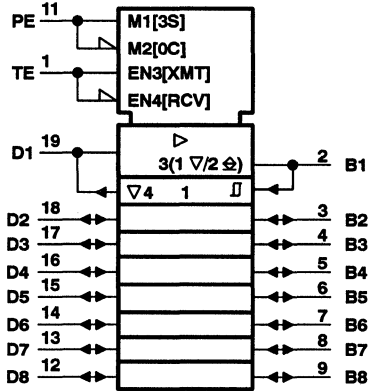
EACH RECEIVER			
INPUTS			OUTPUT
B	TE	PE	D
L	L	X	L
H	L	X	H
X	H	X	Z

H = high level, L = low level, X = irrelevant, Z = high-impedance state

† This is the high-impedance state of a normal 3-state output modified by the internal resistors to  $V_{CC}$  and GND.

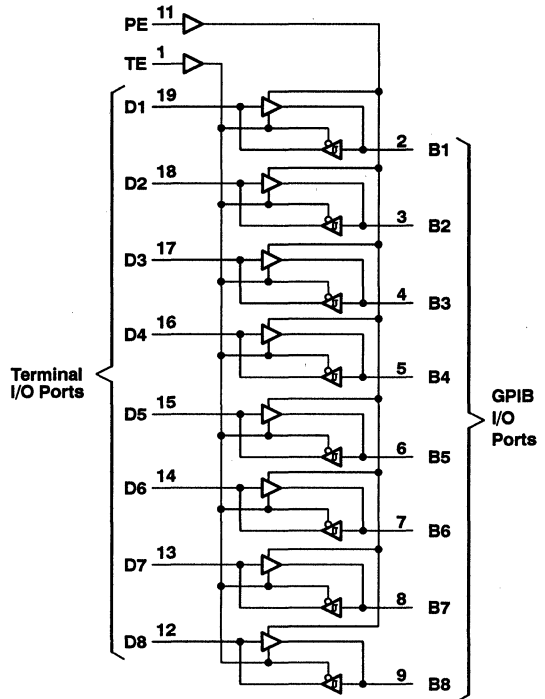
**SN75160B**  
**OCTAL GENERAL-PURPOSE**  
**INTERFACE BUS TRANSCEIVER**  
 SLLS004A - D2525, OCTOBER 1985 - REVISED FEBRUARY 1993

**logic symbol†**

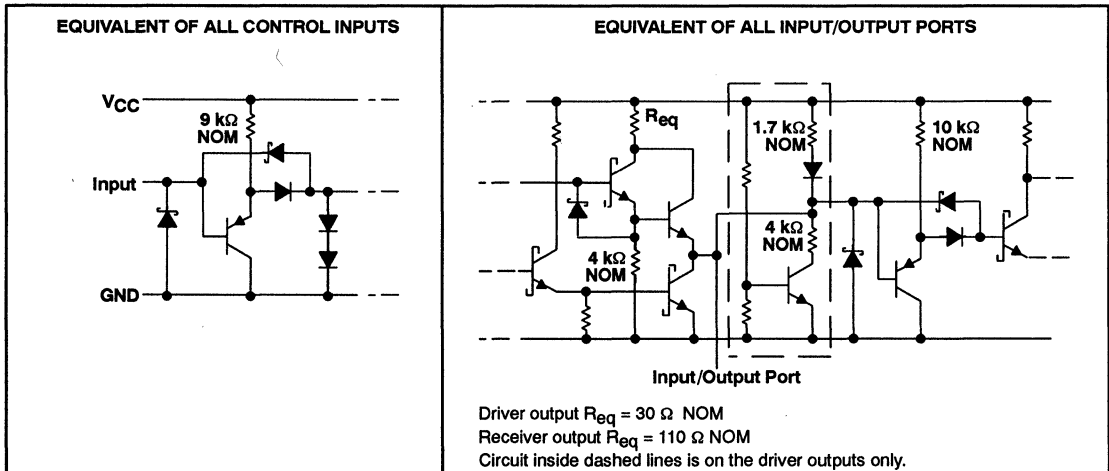


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.  
 ∇ Designates 3-state outputs  
 ⊕ Designates passive-pullup outputs

**logic diagram (positive logic)**



**schematics of inputs and outputs**



**SN75160B**  
**OCTAL GENERAL-PURPOSE**  
**INTERFACE BUS TRANSCEIVER**

SLLS004A – D2525, OCTOBER 1985 – REVISED FEBRUARY 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Low-level driver output current	100 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$	0.8			V
High-level output current, $I_{OH}$	Bus ports with pullups active		-5.2	mA
	Terminal ports		-800	$\mu\text{A}$
High-level output current, $I_{OL}$	Bus ports		48	mA
	Terminal ports		16	
Operating free-air temperature, $T_A$	0		70	°C

# SN75160B

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS004A - D2525, OCTOBER 1985 - REVISED FEBRUARY 1993

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V <sub>IK</sub>	Input clamp voltage		I <sub>I</sub> = -18 mA	-0.8	-1.5		V
V <sub>hys</sub>	Hysteresis (V <sub>T+</sub> - V <sub>T-</sub> )	Bus	See Figure 8	0.4	0.65		V
V <sub>OH</sub>	High-level output voltage	Terminal	I <sub>OH</sub> = -800 μA, TE at 0.8 V	2.7	3.5		V
		Bus	I <sub>OH</sub> = -5.2 mA, PE and TE at 2 V	2.5	3.3		
V <sub>OL</sub>	Low-level output voltage	Terminal	I <sub>OL</sub> = 16 mA, TE at 0.8 V		0.3	0.5	V
		Bus	I <sub>OL</sub> = 48 mA, TE at 2 V		0.35	0.5	
I <sub>I</sub>	Input current at maximum input voltage	Terminal	V <sub>I</sub> = 5.5 V		0.2	100	μA
I <sub>IH</sub>	High-level input current	Terminal	V <sub>I</sub> = 2.7 V		0.1	20	μA
I <sub>IL</sub>	Low-level input current	Terminal	V <sub>I</sub> = 0.5 V		-10	-100	μA
V <sub>I/O(bus)</sub>	Voltage at bus port	Driver disabled	I <sub>I(bus)</sub> = 0	2.5	3.0	3.7	V
			I <sub>I(bus)</sub> = -12 mA			-1.5	
I <sub>I/O(bus)</sub>	Current into bus port	Power on	Driver disabled	V <sub>I(bus)</sub> = -1.5 V to 0.4 V	-1.3		mA
				V <sub>I(bus)</sub> = 0.4 V to 2.5 V	0	-3.2	
				V <sub>I(bus)</sub> = 2.5 V to 3.7 V		2.5	
				V <sub>I(bus)</sub> = 3.7 V to 5 V	0	2.5	
				V <sub>I(bus)</sub> = 5 V to 5.5 V	0.7	2.5	
		Power off	V <sub>CC</sub> = 0, V <sub>I(bus)</sub> = 0 to 2.5 V		-40		
I <sub>OS</sub>	Short-circuit output current	Terminal		-15	-35	-75	mA
		Bus		-25	-50	-125	
I <sub>CC</sub>	Supply current	No load	Receivers low and enabled		70	90	mA
			Drivers low and enabled		85	110	
C <sub>I/O(bus)</sub>	Bus-port capacitance		V <sub>CC</sub> = 0 to 5 V, V <sub>I/O</sub> = 0 to 2 V, f = 1 MHz		30		pF

† All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

switching characteristics, V<sub>CC</sub> = 5 V, C<sub>L</sub> = 15 pF, T<sub>A</sub> = 25°C (unless otherwise noted)

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Terminal	Bus	C <sub>L</sub> = 30 pF, See Figure 1		14	20	ns
t <sub>PHL</sub>					14	20	
t <sub>PLH</sub>	Bus	Terminal	C <sub>L</sub> = 30 pF, See Figure 2		10	20	ns
t <sub>PHL</sub>					15	22	
t <sub>PZH</sub>	TE	BUS	See Figure 3		25	35	ns
t <sub>PHZ</sub>					13	22	
t <sub>PZL</sub>					22	35	
t <sub>PLZ</sub>					22	32	
t <sub>PZH</sub>	TE	Terminal	See Figure 4		20	30	ns
t <sub>PHZ</sub>					12	20	
t <sub>PZL</sub>					23	32	
t <sub>PLZ</sub>					19	30	
t <sub>en</sub>	PE	Bus	See Figure 5		15	22	ns
t <sub>dis</sub>					13	20	



SN75160B  
 OCTAL GENERAL-PURPOSE  
 INTERFACE BUS TRANSCEIVER

SLLS004A - D2525, OCTOBER 1985 - REVISED FEBRUARY 1993

PARAMETER MEASUREMENT INFORMATION

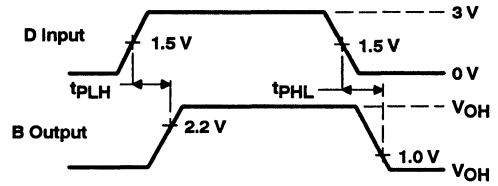
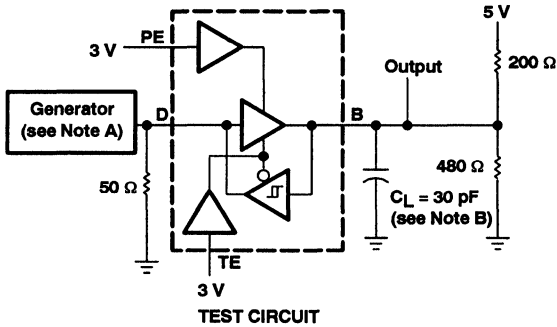


Figure 1. Terminal-to-Bus Test Circuit and Voltage Waveforms

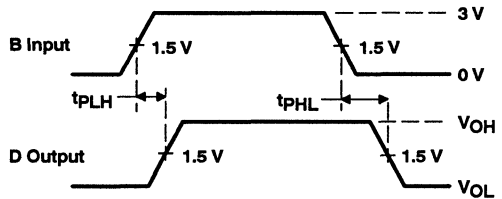
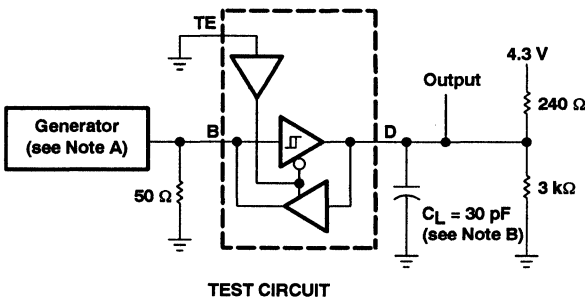


Figure 2. Bus-to-Terminal Test Circuit and Voltage Waveforms

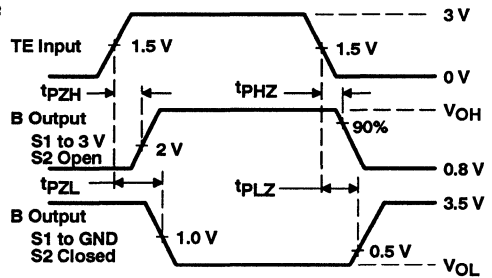
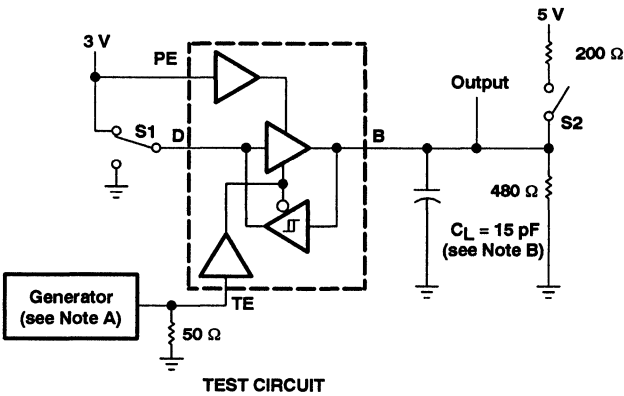
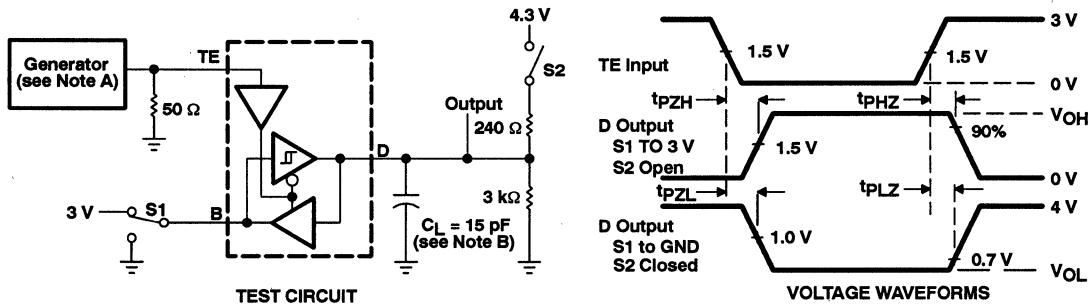


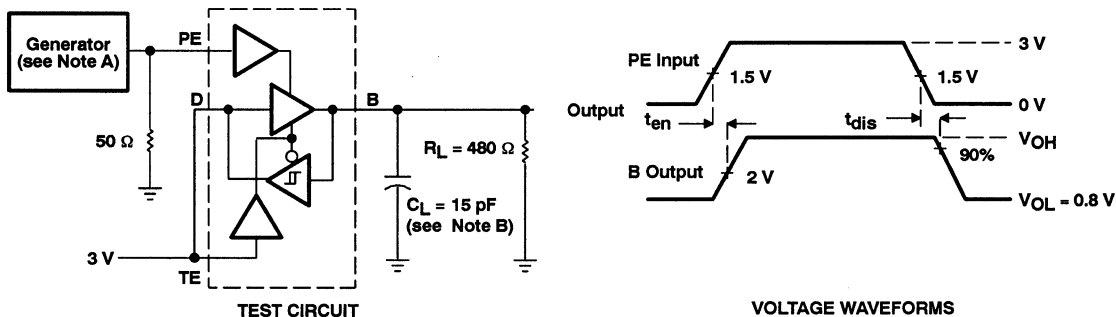
Figure 3. TE-to-Bus Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

**PARAMETER MEASUREMENT INFORMATION**



**Figure 4. TE-to-Terminal Test Circuit and Voltage Waveforms**



**Figure 5. PE-to-Bus Pullup Test Circuit and Voltage Waveforms**

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

TYPICAL CHARACTERISTICS

TERMINAL HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT

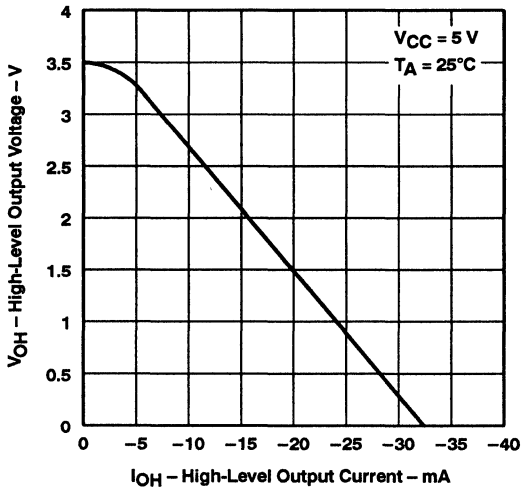


Figure 6

TERMINAL LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

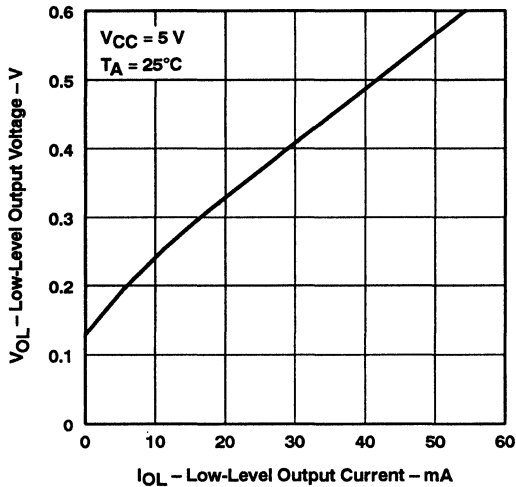


Figure 7

TERMINAL OUTPUT VOLTAGE  
 vs  
 BUS INPUT VOLTAGE

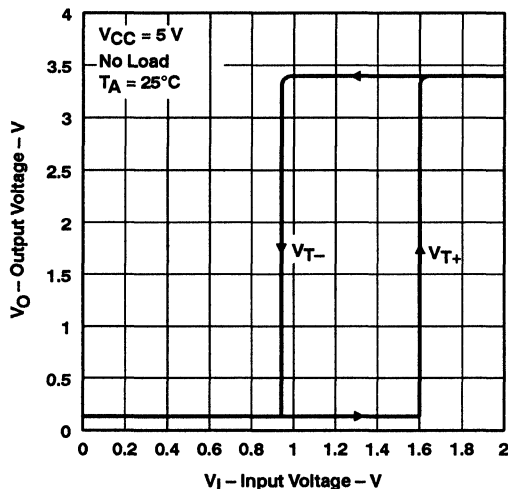


Figure 8



TYPICAL CHARACTERISTICS

BUS HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT

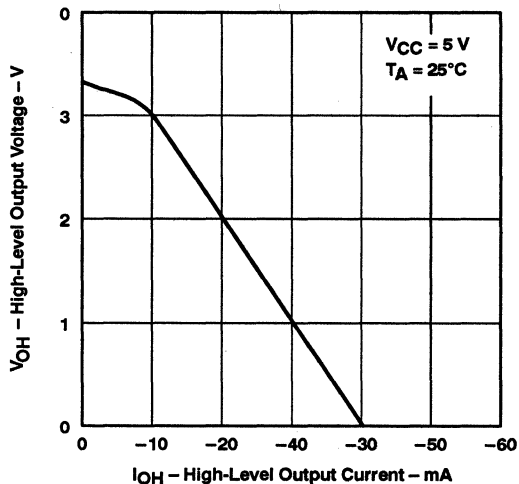


Figure 9

BUS LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

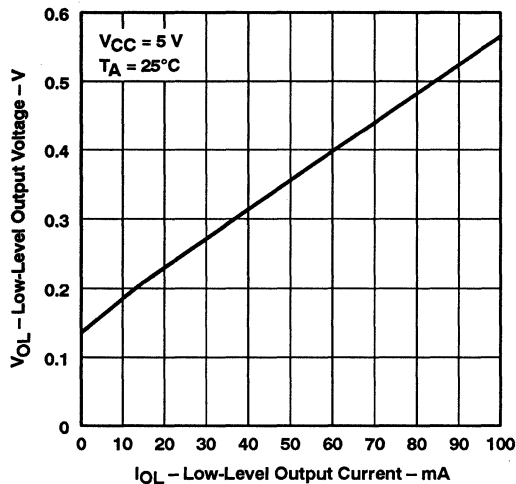


Figure 10

BUS OUTPUT VOLTAGE  
 vs  
 THERMAL INPUT VOLTAGE

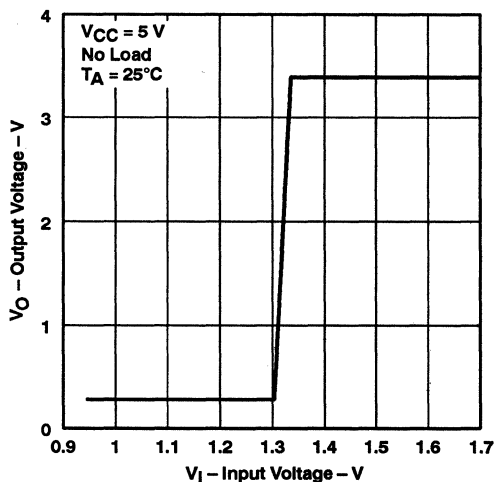


Figure 11

BUS CURRENT  
 vs  
 BUS VOLTAGE

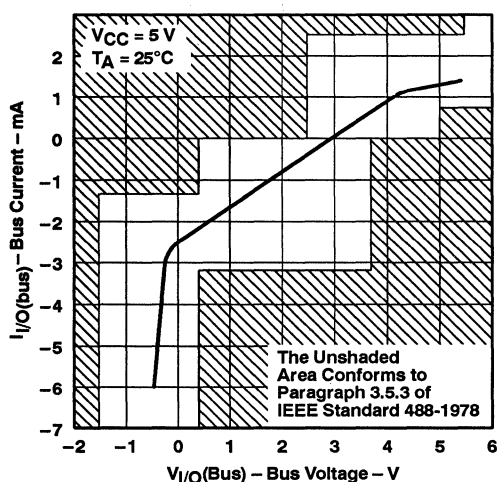


Figure 12

# SN55ALS160, SN75ALS160 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

SLLS018C - D2525, JUNE 1986 - REVISED FEBRUARY 1993

## SUITABLE FOR IEEE STANDARD 488-1978 (GPIB)<sup>†</sup>

- **8-Channel Bidirectional Transceiver**
- **High-Speed Advanced Low-Power Schottky Circuitry**
- **Low Power Dissipation:**  
     SN55ALS160 . . . 56 mW Max  
     Per Channel  
     SN75ALS160 . . . 46 mW Max  
     Per Channel
- **Fast Propagation Times . . . 20 ns Max**
- **High-Impedance PNP Inputs**
- **Receiver Hysteresis:**  
     SN55ALS160 . . . 550 mV Typ  
     SN75ALS160 . . . 650 mV Typ
- **Open-Collector Driver Output Option**
- **No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )**
- **Power-Up/Power-Down Protection (Glitch Free)**

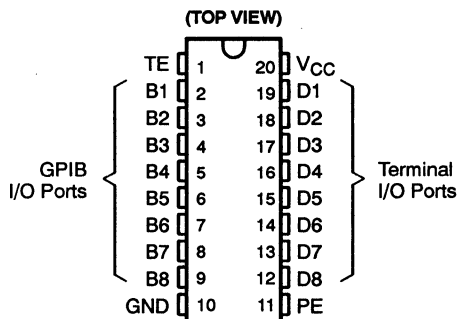
### description

The SN55ALS160 and SN75ALS160 eight-channel general-purpose interface bus transceivers are monolithic, high-speed, advanced low-power Schottky devices designed for two-way data communications over single-ended transmission lines. They are designed to meet the requirements of IEEE Standard 488-1978. The transceivers feature driver outputs that can be operated in either the passive-pullup or 3-state mode. If talk enable (TE) is high, these ports have the characteristics of passive-pullup outputs when pullup enable (PE) is low, and of 3-state outputs when PE is high. Taking TE low places these ports in the high-impedance state. The driver outputs are designed to handle loads up to 48 mA of sink current.

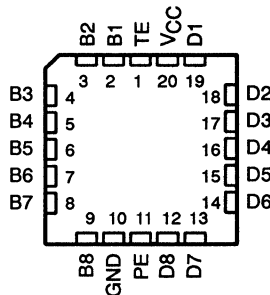
An active turn-off feature has been incorporated into the bus-terminating resistors so that the device exhibits a high impedance to the bus when  $V_{CC} = 0$ . When combined with the SN55ALS161, SN75ALS161, or SN75ALS162 management bus transceiver, the pair provides the complete 16-wire interface for the IEEE-488 bus.

The SN55ALS160 is characterized for operation from  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75ALS160 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

SN55ALS160 . . . J OR W PACKAGE  
SN75ALS160 . . . DW OR N PACKAGE



SN55ALS160 . . . FK PACKAGE  
(TOP VIEW)



### Function Tables

EACH DRIVER				EACH RECEIVER			
INPUTS			OUTPUT	INPUTS			OUTPUT
D	TE	PE	B	B	TE	PE	D
H	H	H	H	L	L	X	L
L	H	X	L	H	L	X	H
H	X	L	Z <sup>‡</sup>	X	H	X	Z
X	L	X	Z <sup>‡</sup>				

H = high level, L = low level, X = irrelevant,  
Z = high-impedance state

<sup>‡</sup> This is the high-impedance state of a normal 3-state output modified by the internal resistors to  $V_{CC}$  and GND.

<sup>†</sup> The transceivers are suitable for IEEE Standard 896 applications to the extent of the operating conditions and characteristics specified in this data sheet. Certain limits contained in the IEEE specification are not met or cannot be tested over the entire military temperature range.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



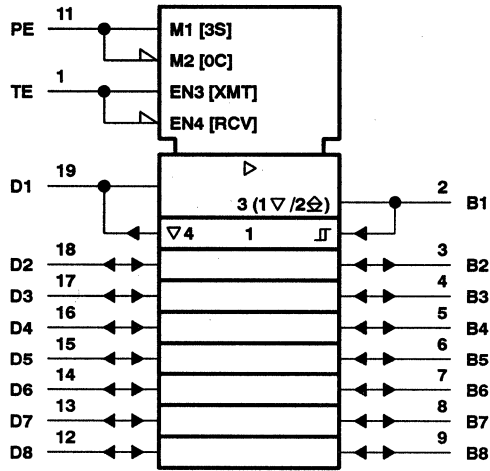
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# SN55ALS160, SN75ALS160 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## logic symbol†

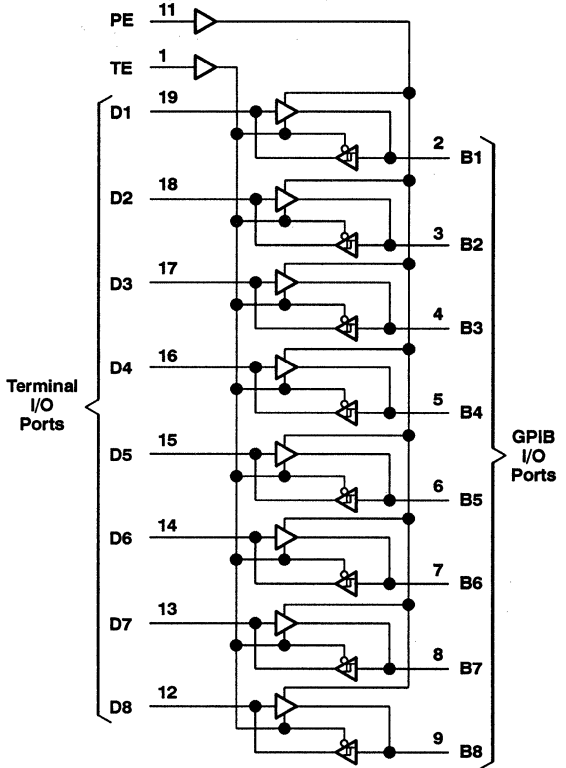


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

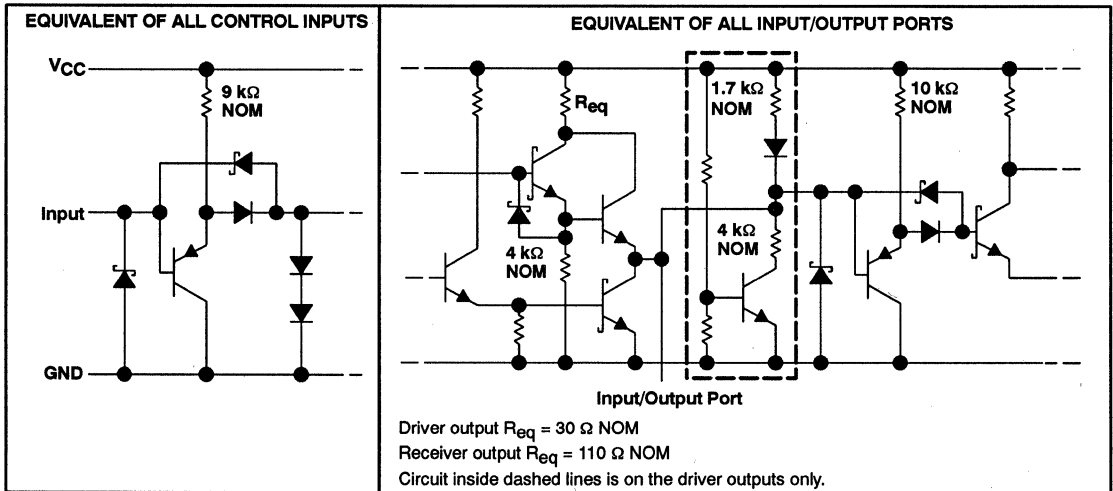
▽ Designates 3-state outputs

⊕ Designates open-collector outputs

## logic diagram (positive logic)



## schematics of inputs and outputs



TEXAS  
INSTRUMENTS

# SN55ALS160, SN75ALS160 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Low-level driver output current	100 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN55ALS160	–55°C to 125°C
SN75ALS160	0°C to 70°C
Storage temperature range	–65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds: DW or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from the case for 60 seconds: J or W package	300°C

NOTES: 1. All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—
W	1000 mW	8.0 mW/°C	640 mW	200 mW

## SN55ALS160 recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	TE and PE at $T_A = -55^\circ\text{C}$ to $125^\circ\text{C}$	2			V
	Bus and terminal at $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	2			
	Bus and terminal at $T_A = -55^\circ\text{C}$	2.1			
Low-level input voltage, $V_{IL}$	TE and PE at $T_A = -55^\circ\text{C}$ to $125^\circ\text{C}$				V
	Bus and terminal at $T_A = 25^\circ\text{C}$ to $-55^\circ\text{C}$	0.8			
	Bus and terminal at $T_A = 125^\circ\text{C}$	0.7			
High-level output current, $I_{OH}$	Bus ports with pullups active ( $V_{CC} = 5\text{ V}$ )	–5.2			mA
	Terminal ports	–800			$\mu\text{A}$
Low-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			
Operating free-air temperature, $T_A$		–55		125	°C

## SN75ALS160 recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$		2			V
Low-level input voltage, $V_{IL}$		0.8			V
High-level output current, $I_{OH}$	Bus ports with pullups active	–5.2			mA
	Terminal ports	–800			$\mu\text{A}$
Low-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			
Operating free-air temperature, $T_A$		0		70	°C



SN55ALS160, SN75ALS160  
OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		SN55ALS160			SN75ALS160			UNIT	
				MIN	TYP‡	MAX	MIN	TYP‡	MAX		
V <sub>IK</sub>	Input clamp voltage	I <sub>I</sub> = -18 mA, V <sub>CC</sub> = MIN		-0.8	-1.5		-0.8	-1.5	V		
V <sub>hys</sub>	Hysteresis (V <sub>T+</sub> - V <sub>T-</sub> )	Bus					0.4	0.65	V		
		Bus	V <sub>CC</sub> = 5 V, T <sub>A</sub> = -55°C and 25°C V <sub>CC</sub> = 5 V, T <sub>A</sub> = 125°C	0.4	0.55						
V <sub>OH</sub> §	High-level output voltage	Terminal	I <sub>OH</sub> = -800 µA, TE at 0.8 V, V <sub>CC</sub> = MIN	2.7	3.5		2.7	3.5	V		
		Bus	I <sub>OH</sub> = -5.2 mA, PE and TE at 2 V, V <sub>CC</sub> = MIN	2.5	3.3		2.5	3.3			
V <sub>OL</sub>	Low-level output voltage	Terminal	I <sub>OL</sub> = 16 mA, TE at 0.8 V, V <sub>CC</sub> = MIN		0.3	0.5		0.3	0.5	V	
		Bus	I <sub>OL</sub> = 48 mA, TE at 2 V, V <sub>CC</sub> = MIN		0.35	0.5		0.35	0.5		
I <sub>I</sub>	Input current at maximum input voltage	Terminal	V <sub>I</sub> = 5.5 V, V <sub>CC</sub> = MAX		0.2	100		0.2	100	µA	
I <sub>IH</sub>	High-level input current	Terminal,	V <sub>I</sub> = 2.7 V, V <sub>CC</sub> = MAX		0.1	20		0.1	20	µA	
I <sub>IL</sub>	Low-level input current	PE, or TE	V <sub>I</sub> = 0.5 V, V <sub>CC</sub> = MAX		-30	-100		-10	-100	µA	
V <sub>I/O</sub> (bus)	Voltage at bus port	Driver disabled, V <sub>CC</sub> = 5 V (SN55)		I <sub>I</sub> (bus) = 0	2.5	3	3.7	2.5	3	3.7	V
				I <sub>I</sub> (bus) = -12 mA			-1.5			-1.5	
I <sub>I/O</sub> (bus)	Current into bus port	Power on	Driver disabled, V <sub>CC</sub> = 5 V (SN55)	V <sub>I</sub> (bus) = -1.5 V to 0.4 V	-1.3			-1.3			mA
				V <sub>I</sub> (bus) = 0.4 V to 2.5 V		0	-3.2		0	-3.2	
				V <sub>I</sub> (bus) = 2.5 V to 3.7 V			2.5			2.5	
				V <sub>I</sub> (bus) = 3.7 V to 5 V		0	2.5		0	2.5	
				V <sub>I</sub> (bus) = 5 V to 5.5 V		0.7	2.5		0.7	2.5	
		Power off	V <sub>CC</sub> = 0	V <sub>I</sub> (bus) = 0 to 2.5 V			40			40	µA
I <sub>OS</sub>	Short-circuit output current	Terminal	V <sub>CC</sub> = MAX	-15	-35	-75	-15	-35	-75	mA	
		Bus	V <sub>CC</sub> = MAX	-25	-50	-125	-25	-50	-125		
I <sub>CC</sub>	Supply current	No load, V <sub>CC</sub> = MAX		Terminal outputs low and enabled			42			56	mA
				Bus outputs low and enabled			52			85	
C <sub>I/O</sub> (bus)	Bus-port capacitance	V <sub>CC</sub> = 0 to 5 V, V <sub>I/O</sub> = 0 to 2 V, f = 1 MHz		30			30			pF	

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

§ V<sub>OH</sub> applies to 3-state outputs only.

# SN55ALS160

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANCEIVER

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**switching characteristics at  $V_{CC} = 4.75\text{ V}$ ,  $5\text{ V}$ , and  $5.25\text{ V}$ ,  $C_L = 50\text{ pF}$  (unless otherwise noted)**

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	$T_A$ †	MIN	TYP ‡	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	Terminal	Bus	See Figure 1	25°C		10	17	ns
				Full range			20	
$t_{PHL}$ Propagation delay time, high-to-low-level output				25°C		10	14	
				Full range			16	
$t_{PLH}$ Propagation delay time, low-to-high-level output	Bus	Terminal	See Figure 2	25°C		8	15	ns
				Full range			18	
$t_{PHL}$ Propagation delay time, high-to-low-level output				25°C		8	15	
				Full range			18	
$t_{PZH}$ Output enable time to high level	TE	Bus	See Figure 3	25°C		24	30	ns
				Full range			41	
$t_{PHZ}$ Output disable time from high level				25°C		9	14	
				Full range			16	
$t_{PZL}$ Output enable time to low level				25°C		16	28	
				Full range			34	
$t_{PLZ}$ Output disable time from low level				25°C		12	19	
				Full range			24	
$t_{PZH}$ Output enable time to high level	TE	Terminal	See Figure 4	25°C		24	36	ns
				Full range			50	
$t_{PHZ}$ Output disable time from high level				25°C		10	18	
				Full range			23	
$t_{PZL}$ Output enable time to low level				25°C		15	26	
				Full range			30	
$t_{PLZ}$ Output disable time from low level				25°C		15	24	
				Full range			31	
$t_{en}$ Output pullup enable time	PE	Bus	See Figure 5	25°C		16	24	ns
				Full range			25	
$t_{dis}$ Output pullup disable time				25°C		9	16	
				Full range			20	

† Full range is  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

‡ All typical values are at  $V_{CC} = 5\text{ V}$ .



# SN75ALS160

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANCEIVER

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switching characteristics over recommended range of operating free-air temperature,  $V_{CC} = 5\text{ V}$

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Terminal	Bus	C <sub>L</sub> = 30 pF, See Figure 1	7	20	ns	
t <sub>PHL</sub> Propagation delay time, high-to-low-level output				8	20		
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Bus	Terminal	C <sub>L</sub> = 30 pF, See Figure 2	7	14	ns	
t <sub>PHL</sub> Propagation delay time, high-to-low-level output				9	14		
t <sub>PZH</sub> Output enable time to high level	TE	Bus	C <sub>L</sub> = 15 pF, See Figure 3	19	30	ns	
t <sub>PHZ</sub> Output disable time from high level				5	12		
t <sub>PZL</sub> Output enable time to low level				16	35		
t <sub>PLZ</sub> Output disable time from low level				9	20		
t <sub>PZH</sub> Output enable time to high level	TE	Terminal	C <sub>L</sub> = 15 pF, See Figure 4	13	30	ns	
t <sub>PHZ</sub> Output disable time from high level				12	20		
t <sub>PZL</sub> Output enable time to low level				12	20		
t <sub>PLZ</sub> Output disable time from low level				11	20		
t <sub>en</sub> Output pullup enable time	PE	Bus	C <sub>L</sub> = 15 pF, See Figure 5	11	22	ns	
t <sub>dis</sub> Output pullup disable time				6	12		

† Typical values are at T<sub>A</sub> = 25°C.



# SN55ALS160, SN75ALS160 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION

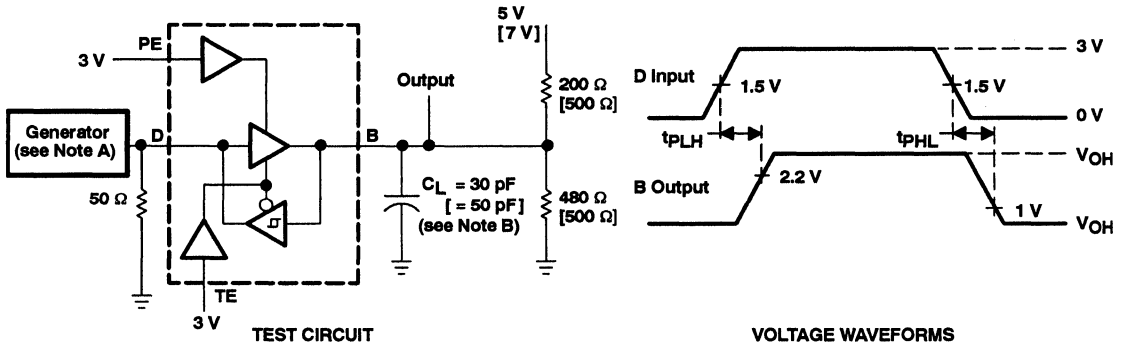


Figure 1. Terminal-to-Bus Test Circuit and Voltage Waveforms

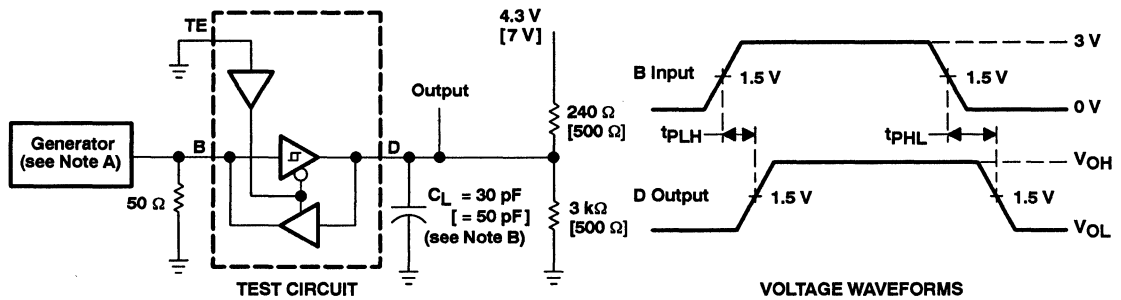


Figure 2. Bus-to-Terminal Test Circuit and Voltage Waveforms

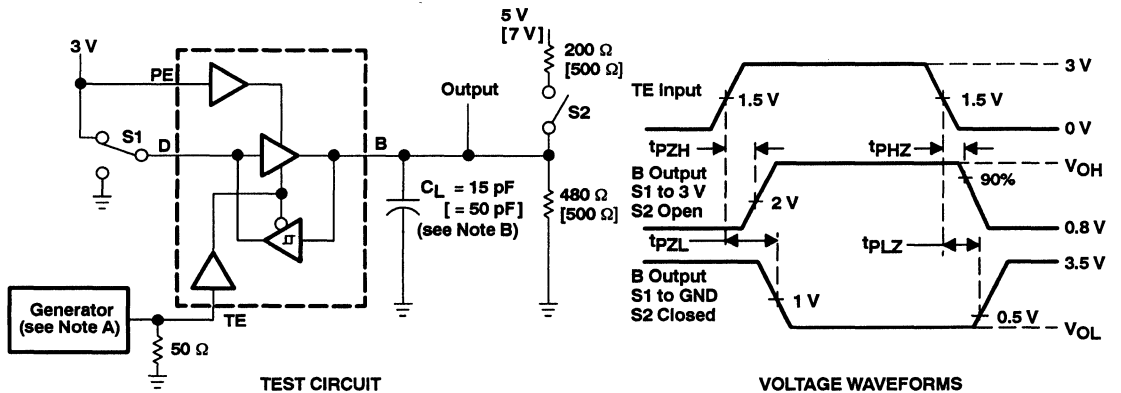


Figure 3. TE-to-Bus Test Circuit and Voltage Waveforms

[ ] denotes the SN55ALS160 military test conditions.

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.





# SN55ALS160, SN75ALS160 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION

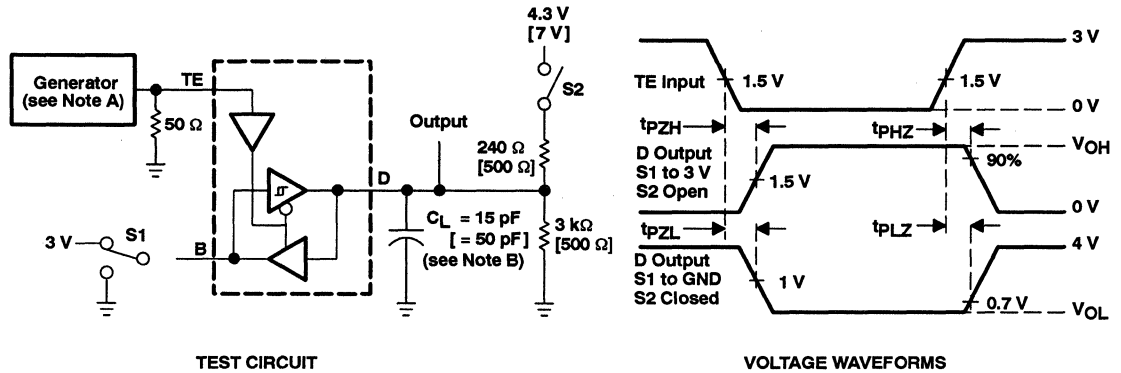


Figure 4. TE-to-Terminal Test Circuit and Voltage Waveforms

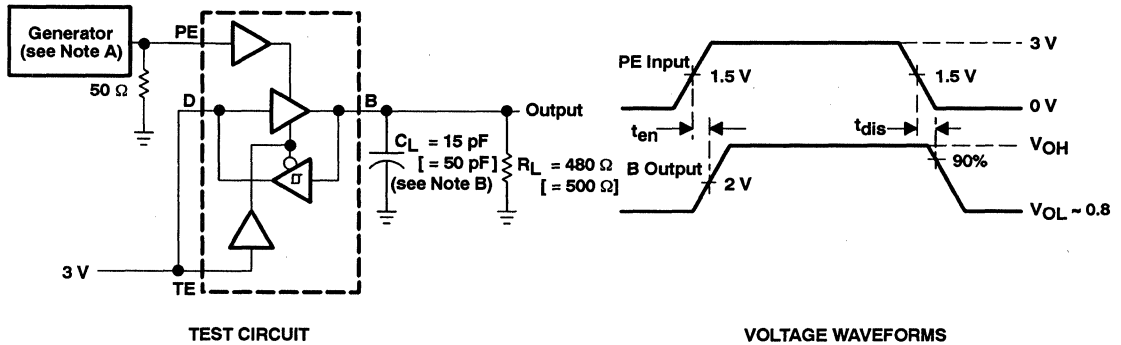


Figure 5. PE-to-Bus Test Circuit and Voltage Waveforms

[ ] denotes the SN55ALS160 military test conditions.

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

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# SN55ALS160, SN75ALS160 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

**TERMINAL HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

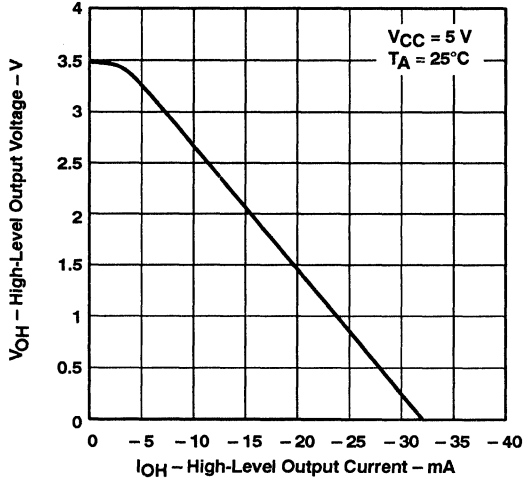


Figure 6

**TERMINAL LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

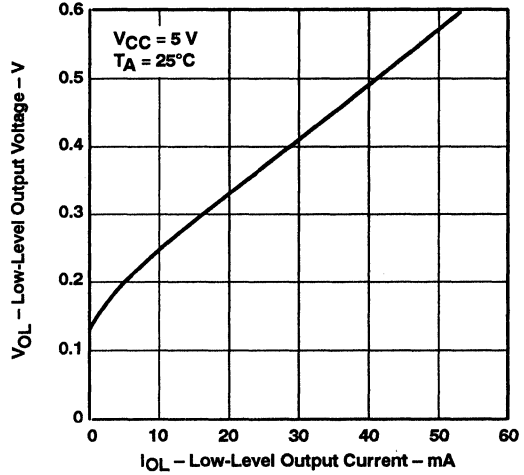


Figure 7

**TERMINAL OUTPUT VOLTAGE  
vs  
BUS INPUT VOLTAGE**

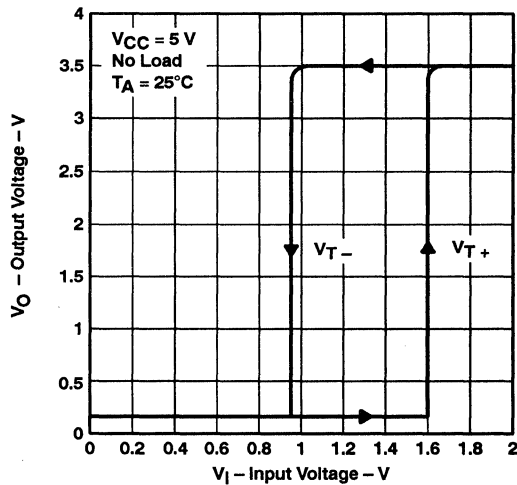


Figure 8

**TEXAS  
INSTRUMENTS**

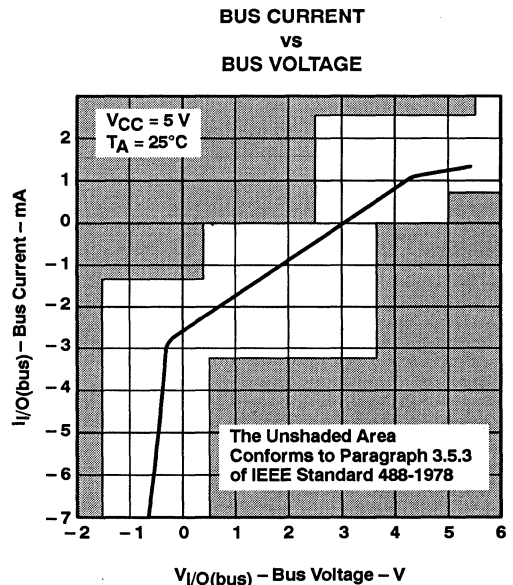
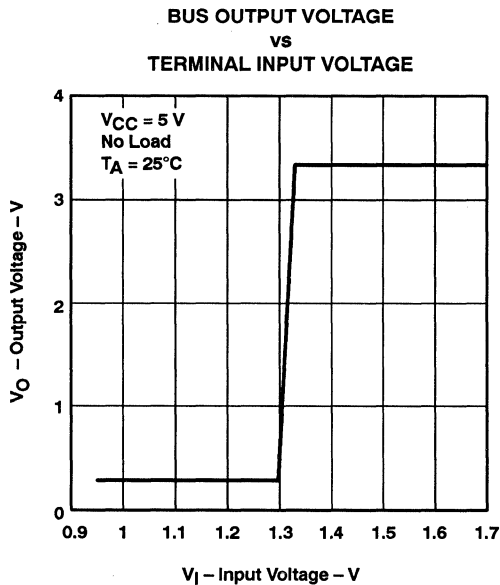
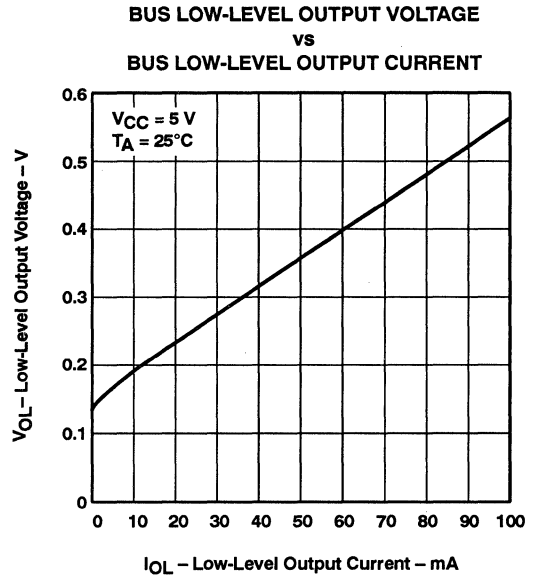
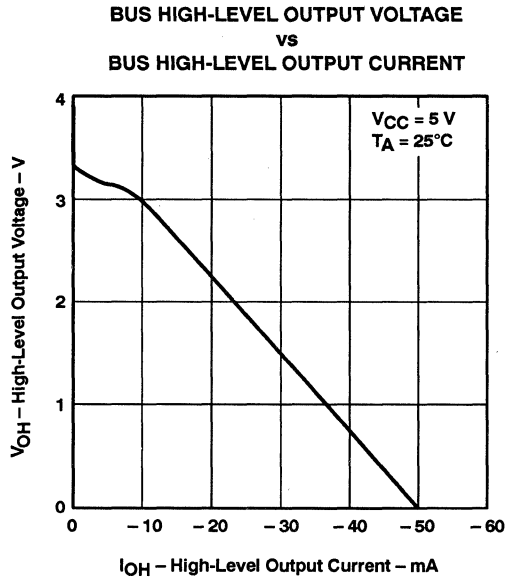
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# SN55ALS160, SN75ALS160 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS



# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

SLLS005A – D2618, OCTOBER 1980 – REVISED FEBRUARY 1993

## MEETS IEEE STANDARD 488-1978 (GPIB)

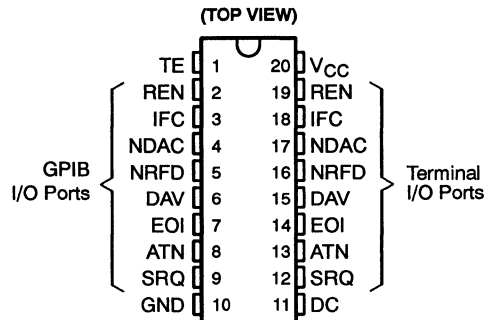
- **8-Channel Bidirectional Transceiver**
- **Power-Up/Power-Down Protection (Glitch Free)**
- **Designed to Implement Control Bus Interface**
- **SN75161B Designed for Single Controller**
- **SN75162B Designed for Multiple Controllers**
- **High-Speed, Low-Power Schottky Circuitry**
- **Low-Power Dissipation . . . 72 mW Max Per Channel**
- **Fast Propagation Times . . . 22 ns Max**
- **High-impedance PNP Inputs**
- **Receiver Hysteresis . . . 650 mV Typ**
- **Bus-Terminating Resistors Provided on Driver Outputs**
- **No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )**

### description

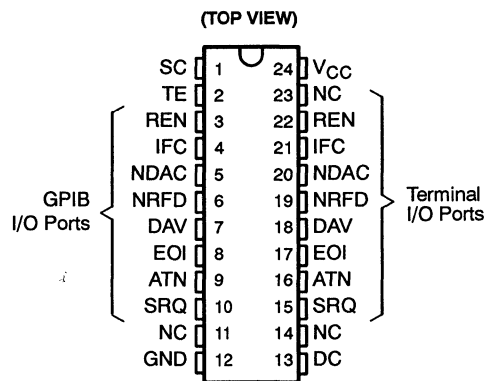
The SN75161B and SN75162B eight-channel, general-purpose interface bus transceivers are monolithic, high-speed, low-power Schottky devices designed to meet the requirements of IEEE Standard 488-1978. Each transceiver is designed to provide the bus-management and data-transfer signals between operating units of a single- or multiple-controller instrumentation system. When combined with the SN75160B octal bus transceiver, the SN75161B or SN75162B provides the complete 16-wire interface for the IEEE-488 bus.

The SN75161B and SN75162B each features eight driver-receiver pairs connected in a front-to-back configuration to form input/output (I/O) ports at both the bus and terminal sides. A power-up/down disable circuit is included on all bus and receiver outputs. This provides glitch-free operation during  $V_{CC}$  power up and power down. The direction of data through these driver-receiver pairs is determined by the DC, TE, and SC (on SN75162B) enable signals. The SC input on the SN75162B allows the REN and IFC transceivers to be controlled independently.

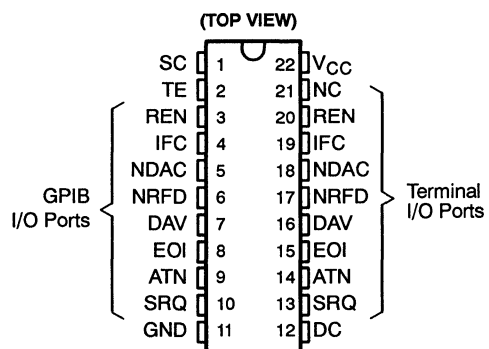
SN75161B . . . DW OR N PACKAGE



SN75162B . . . DW PACKAGE



SN75162B . . . N PACKAGE



NC—No internal connection

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## description (continued)

The driver outputs (GPIB I/O ports) feature active bus-terminating resistor circuits designed to provide a high impedance to the bus when supply voltage  $V_{CC}$  is 0. The drivers are designed to handle loads up to 48 mA of sink current. Each receiver features pnp transistor inputs for high input impedance and hysteresis of 400 mV for increased noise immunity. All receivers have 3-state outputs to present a high impedance to the terminal when disabled.

The SN75161B and SN75162B are characterized for operation from 0°C to 70°C.

CHANNEL IDENTIFICATION TABLE

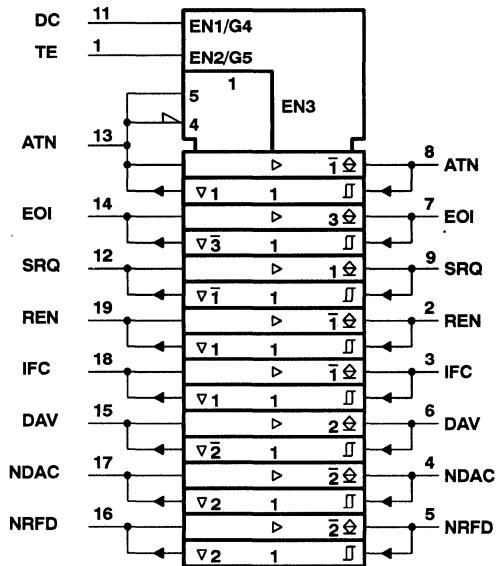
NAME	IDENTITY	CLASS
DC	Direction Control	
TE	Talk Enable	Control
SC	System Control (SN75162B only)	
ATN	Attention	
SRQ	Service Request	
REN	Remote Enable	Bus
IFC	Interface Clear	Management
EOI	End of Identity	
DAV	Data Valid	
NDAC	Not Data Accepted	Data
NRFD	Not Ready for Data	Transfer

# SN75161B

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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**logic symbol†**

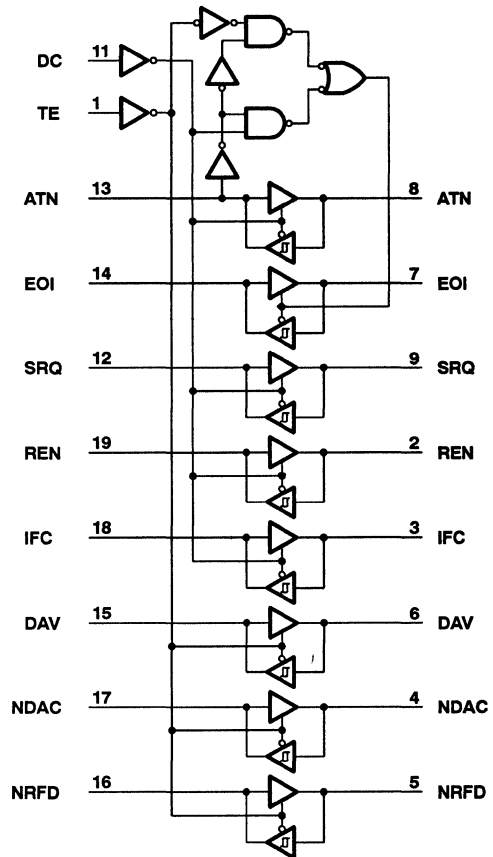


† This symbol is in accordance with IEEE Std 91-1984 and IEC Publication 617-12.

▽ Designates 3-state outputs

⊕ Designates passive-pullup outputs

**logic diagram (positive logic)**



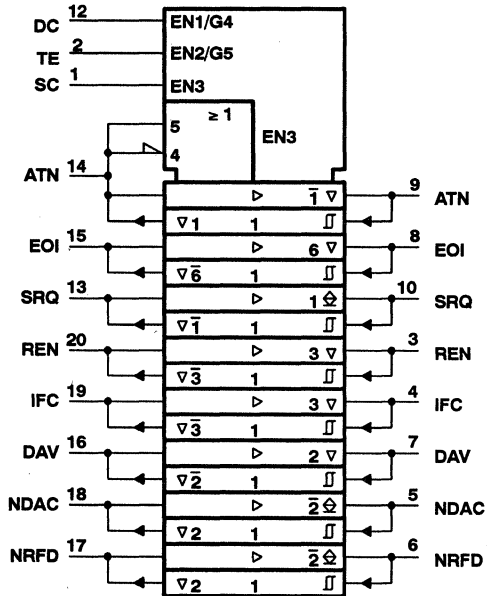
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# SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

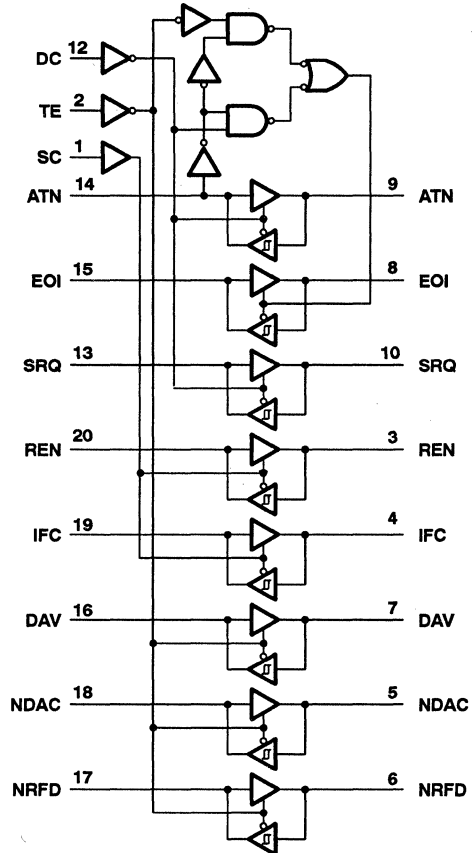
SLLS005A—D2618, OCTOBER 1980—REVISED FEBRUARY 1993

logic symbol†



†This symbol is in accordance with IEEE Std 91-1984 and IEC Publication 617-12.  
 ▽ Designates 3-state outputs  
 ⊕ Designates passive-pullup outputs

logic diagram (positive logic)



Pin numbers shown are for the N package.

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# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## Function Tables

### SN75161B RECEIVE/TRANSMIT

CONTROLS			BUS-MANAGEMENT CHANNELS				DATA-TRANSFER CHANNELS			
DC	TE	ATN†	ATN†	SRQ	REN	IFC	EOI	DAV	NDAC	NRFD
			(Controlled by DC)				(Controlled by TE)			
H	H	H	R	T	R	R	T	T	R	R
H	H	L					R			
L	L	H	T	R	T	T	R	R	T	T
L	L	L					T			
H	L	X	R	T	R	R	R	R	T	T
L	H	X	T	R	T	T	T	T	R	R

### SN75162B RECEIVE/TRANSMIT

CONTROLS				BUS-MANAGEMENT CHANNELS				DATA-TRANSFER CHANNELS				
SC	DC	TE	ATN†	ATN†	SRQ	REN	IFC	EOI	DAV	NDAC	NRFD	
				(Controlled by DC)		(Controlled by SC)		(Controlled by TE)				
	H	H	H	R	T			T	T	R	R	
	H	H	L					R				
	L	L	H	T	R			R	R	T	T	T
	L	L	L					T				
	H	L	X	R	T			R	R	R	T	T
	L	H	X	T	R			R	R	T	T	R
H						T	T					
L						R	R					

H = high level, L = low level, R = receive, T = transmit, X = irrelevant

Direction of data transmission is from the terminal side to the bus side, and the direction of data receiving is from the bus side to the terminal side.

Data transfer is noninverting in both directions.

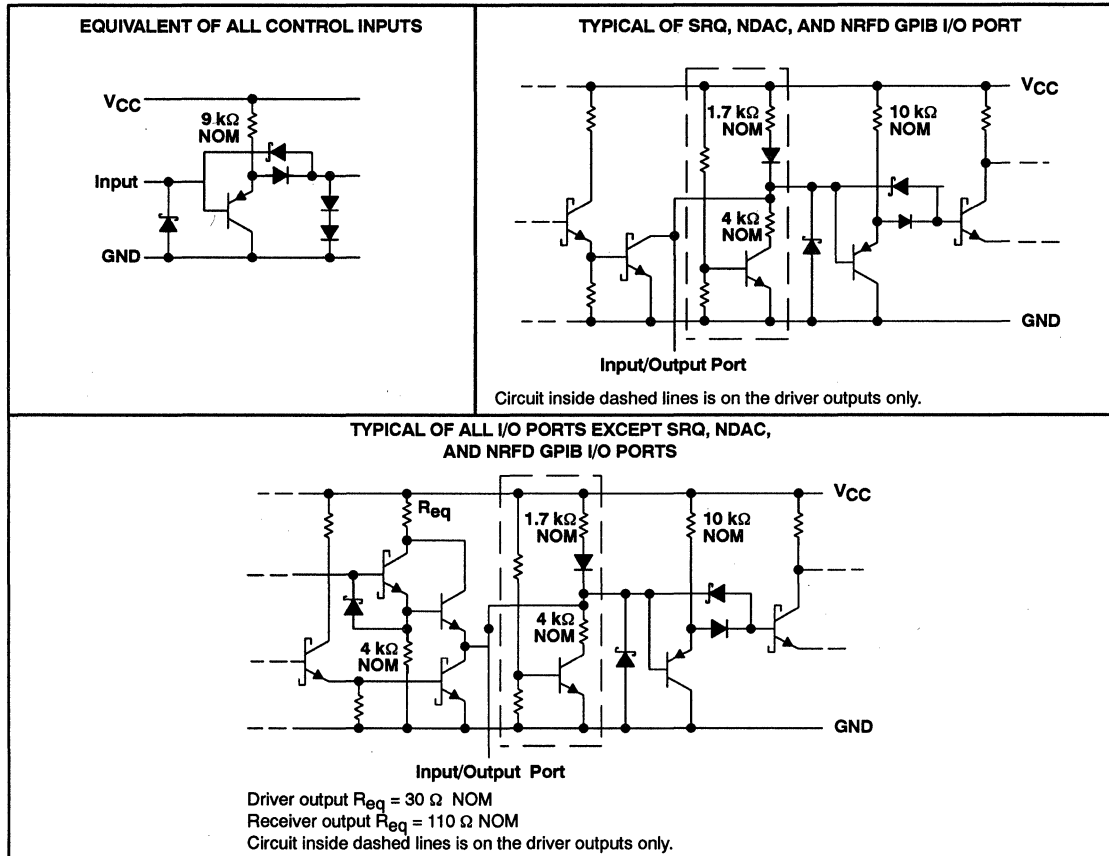
† ATN is a normal transceiver channel that functions additionally as an internal direction control or talk enable for EOI whenever the DC and TE inputs are in the same state. When DC and TE are in opposite states, the ATN channel functions as an independent transceiver only.



# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

SLLS005A - D2618, OCTOBER 1980 - REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Low-level driver output current	100 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16) inch from the case for 10 seconds	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

TEXAS  
INSTRUMENTS

# SN75161B, SN75162B

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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**DISSIPATION RATING TABLE**

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW (20 pin)	1125 mW	9.0 mW/°C	720 mW
DW (24 pin)	1350 mW	10.8 mW/°C	864 mW
N (20 pin)	1150 mW	9.2 mW/°C	736 mW
N (22 pin)	1700 mW	13.6 mW/°C	1088 mW

### recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level Input voltage, $V_{IH}$		2			V
Low-level Input voltage, $V_{IL}$		0.8			V
High-level output current, $I_{OH}$	Bus ports with 3-state outputs	-5.2			mA
	Terminal ports	-800			$\mu\text{A}$
Low-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			mA
Operating free-air temperature, $T_A$		0			70 °C



# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage		$I_I = -18 \text{ mA}$	-0.8	-1.5		V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	Bus	See Figure 2	0.4	0.65		V
$V_{OH}^\ddagger$	High-level output voltage	Terminal	$I_{OH} = -800 \mu\text{A}$	2.7	3.5		V
		Bus	$I_{OH} = -5.2 \text{ mA}$	2.5	3.3		
$V_{OL}$	Low-level output voltage	Terminal	$I_{OL} = 16 \text{ mA}$	0.3	0.5		V
		Bus	$I_{OL} = 48 \text{ mA}$	0.35	0.5		
$I_I$	Input current at maximum input voltage	Terminal	$V_I = 5.5 \text{ V}$	0.2	100		$\mu\text{A}$
$I_{IH}$	High-level input current	Terminal and control inputs	$V_I = 2.7 \text{ V}$	0.1	20		$\mu\text{A}$
$I_{IL}$	Low-level input current	Terminal and control inputs	$V_I = 0.5 \text{ V}$	-10	-100		$\mu\text{A}$
$V_{I/O(\text{bus})}$	Voltage at bus port	Driver disabled	$I_{I(\text{bus})} = 0$	2.5	3.0	3.7	V
			$I_{I(\text{bus})} = -12 \text{ mA}$			-1.5	
$I_{I/O(\text{bus})}$	Current into bus port	Power on	Driver disabled	$V_{I(\text{bus})} = -1.5 \text{ V to } 0.4 \text{ V}$	-1.3		mA
				$V_{I(\text{bus})} = 0.4 \text{ V to } 2.5 \text{ V}$	0	-3.2	
				$V_{I(\text{bus})} = 2.5 \text{ V to } 3.7 \text{ V}$		2.5	
				$V_{I(\text{bus})} = 3.7 \text{ V to } 5 \text{ V}$	0	2.5	
				$V_{I(\text{bus})} = 5 \text{ V to } 5.5 \text{ V}$	0.7	2.5	
				Power off	$V_{CC} = 0, V_{I(\text{bus})} = 0 \text{ V to } 2.5 \text{ V}$		
$I_{OS}$	Short-circuit output current	Terminal		-15	-35	-75	mA
		Bus		-25	-50	-125	
$I_{CC}$	Supply current		No load, TE, DE, and SC low			110	mA
$C_{I/O(\text{bus})}$	Bus-port capacitance		$V_{CC} = 5 \text{ V to } 0, V_{I/O} = 0 \text{ to } 2 \text{ V}, f = 1 \text{ MHz}$		30		pF

† All typical values are at  $V_{CC} = 5 \text{ V}, T_A = 25^\circ\text{C}$ .

‡  $V_{OH}$  applies for 3-state outputs only.



# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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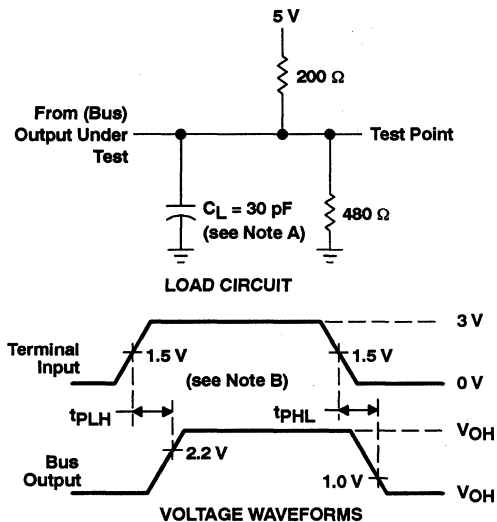
**switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 15\text{ pF}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	Terminal	Bus	C <sub>L</sub> = 30 pF, See Figure 1		14	20	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low level output					14	20	
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	Terminal	Bus (SRQ, NDAC NRFD)	C <sub>L</sub> = 30 pF, See Figure 1		29	35	ns
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	Bus	Terminal	C <sub>L</sub> = 30 pF, See Figure 2		10	20	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low level output					15	22	
t <sub>PZH</sub>	Output enable time to high level	TE,DC, or SC	Bus (ATN, EOI, REN, IFC, and DAV)	See Figure 3			60	ns
t <sub>PHZ</sub>	Output disable time from high level						45	
t <sub>PZL</sub>	Output enable time to low level						60	
t <sub>PLZ</sub>	Output disable time from low level						55	
t <sub>PZH</sub>	Output enable time to high level	TE,DC, or SC	Terminal	See Figure 4			55	ns
t <sub>PHZ</sub>	Output disable time from high level						50	
t <sub>PZL</sub>	Output enable time to low level						45	
t <sub>PLZ</sub>	Output disable time from low level						55	

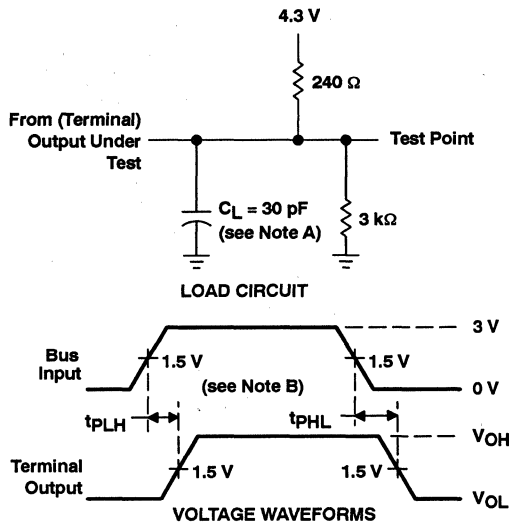
# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION



**Figure 1. Terminal-to-Bus  
Load Circuit and Voltage Waveforms**



**Figure 2. Bus-to-Terminal  
Load Circuit and Voltage Waveforms**

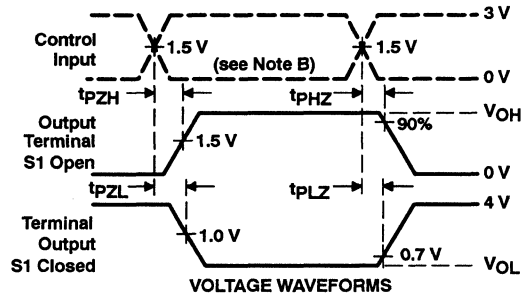
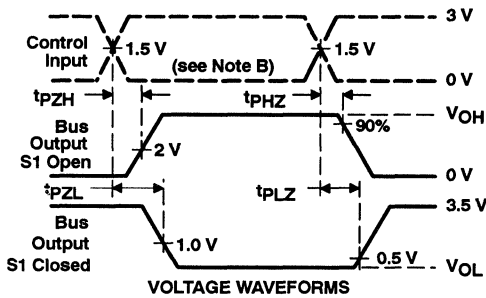
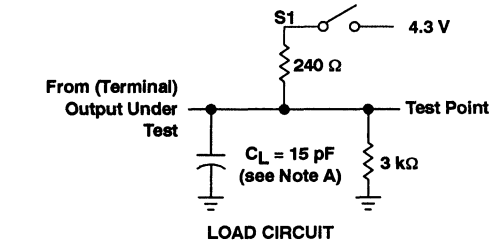
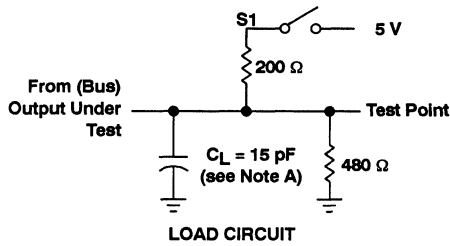
NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .

# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION



**Figure 3. Bus Enable and Disable Times  
Load Circuit and Voltage Waveforms**

**Figure 4. Terminal Enable and Disable Times  
Load Circuit and Voltage Waveforms**

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The Input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_O = 50 \Omega$ .

# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

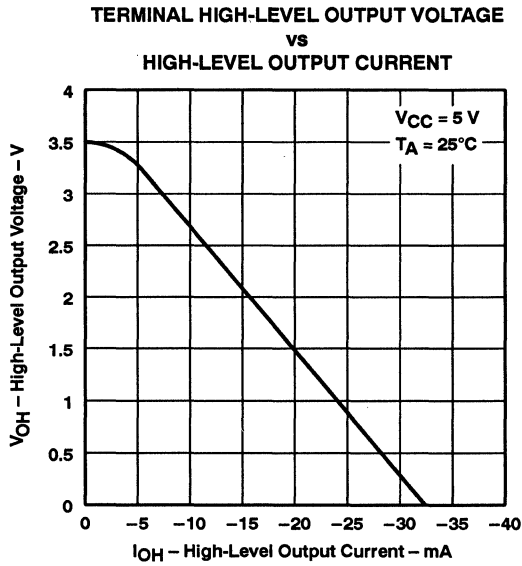


Figure 5

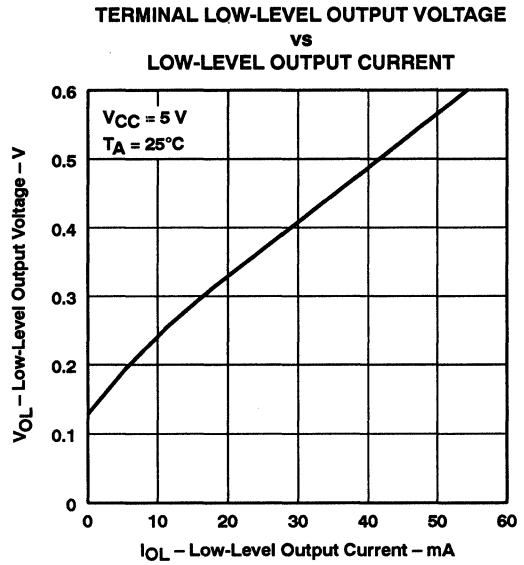


Figure 6

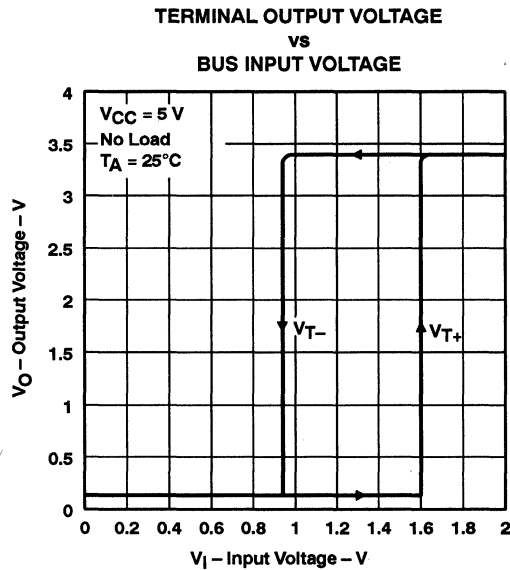


Figure 7

# SN75161B, SN75162B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

**BUS HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

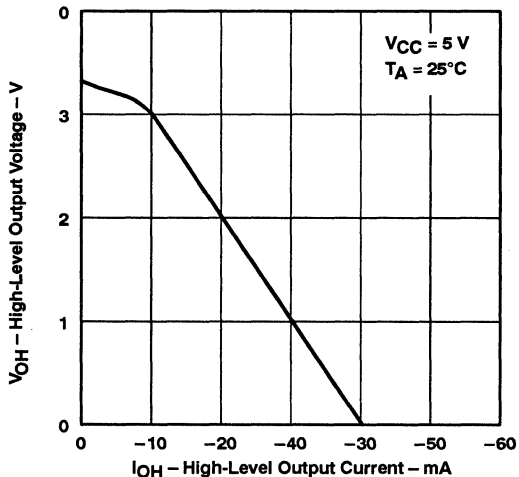


Figure 8

**BUS LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

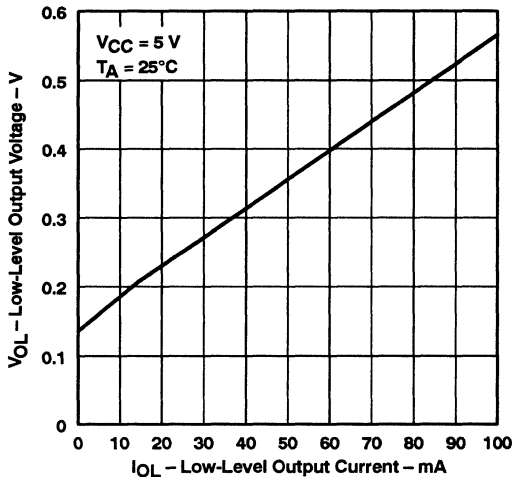


Figure 9

**BUS OUTPUT VOLTAGE  
vs  
THERMAL INPUT VOLTAGE**

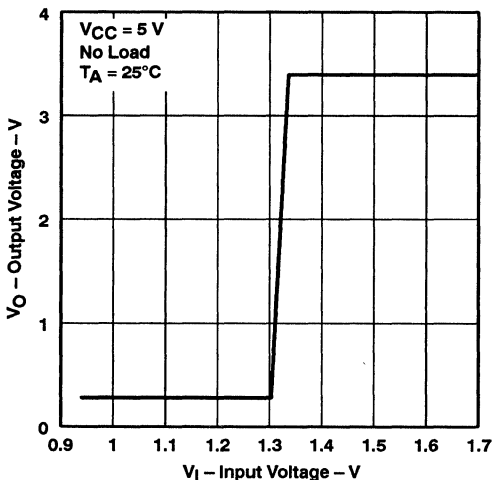


Figure 10

**BUS CURRENT  
vs  
BUS VOLTAGE**

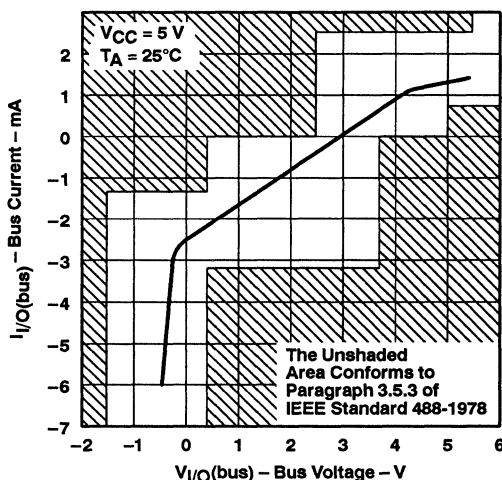


Figure 11





# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

SLLS019C - D2618, JUNE 1986 - REVISED FEBRUARY 1993

## SUITABLE FOR IEEE STANDARD 488-1978 (GPIB)<sup>†</sup>

- 8-Channel Bidirectional Transceiver
- Designed to Implement Control Bus Interface
- Designed for Single Controller
- High-Speed Advanced Low-Power Schottky Circuitry
- Low-Power Dissipation:  
SN55ALS161 . . . 59 mW Max Per Channel  
SN75ALS161 . . . 46 mW Max Per Channel
- Fast Propagation Times:  
SN55ALS161 . . . 25 ns Max  
SN75ALS161 . . . 20 ns Max
- High-impedance PNP Inputs
- Receiver Hysteresis:  
SN55ALS161 . . . 550 mV Typ  
SN75ALS161 . . . 650 mV Typ
- Bus-Terminating Resistors Provided on Driver Outputs
- No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )
- Power-Up/Power-Down Protection (Glitch Free)

### description

The SN55ALS161 and SN75ALS161 eight-channel general-purpose interface bus transceivers are monolithic, high-speed, advanced low-power Schottky process devices designed to provide the bus-management and data-transfer signals between operating units of a single controller instrumentation system. When combined with the SN55ALS160 and SN75ALS160 octal bus transceivers, the 'ALS161 provides the complete 16-wire interface for the IEEE 488 bus.

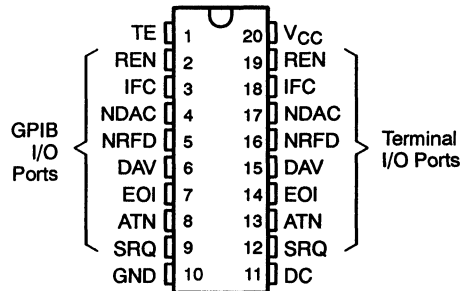
The SN55ALS161 and SN75ALS161 feature eight driver-receiver pairs connected in a front-to-back configuration to form input/output (I/O) ports at both the bus and terminal sides. The direction of data through these driver-receiver pairs is determined by the DC and TE enable signals.

The driver outputs (GPIB I/O ports) feature active bus-terminating resistor circuits designed to provide a high impedance to the bus when  $V_{CC} = 0$ . The drivers are designed to handle loads up to 48 mA of sink current. Each receiver features pnp transistor inputs for high input impedance and hysteresis of 400 mV on the commercial part, 250 mV on the military part minimum for increased noise immunity. All receivers have 3-state outputs to present a high impedance to the terminal when disabled.

<sup>†</sup> The transceivers are suitable for IEEE Standard 488 applications to the extent of the operating conditions and characteristics specified in this data sheet. Certain limits contained in the IEEE specification are not met or cannot be tested over the entire military temperature range.

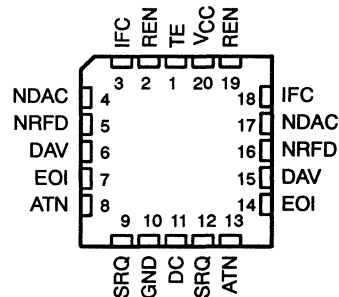
SN55ALS161 . . . J OR W PACKAGE  
SN75ALS161 . . . DW OR N PACKAGE

(TOP VIEW)



SN55ALS161 . . . FK PACKAGE

(TOP VIEW)



CHANNEL IDENTIFICATION TABLE

NAME	IDENTITY	CLASS
DC	Direction Control	Control
TE	Talk Enable	
ATN	Attention	Bus Management
SRQ	Service Request	
REN	Remote Enable	
IFC	Interface Clear	
EOI	End or Identify	Data Transfer
DAV	Data Valid	
NRFD	Not Ready for Data	

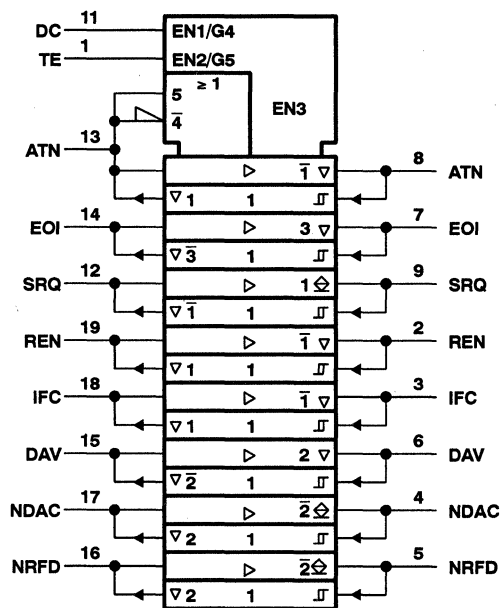
# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

SLLS019C - D2618, JUNE 1986 - REVISED FEBRUARY 1993

## description (continued)

The SN5ALS161 is characterized for operation from  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75ALS161 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## logic symbol†

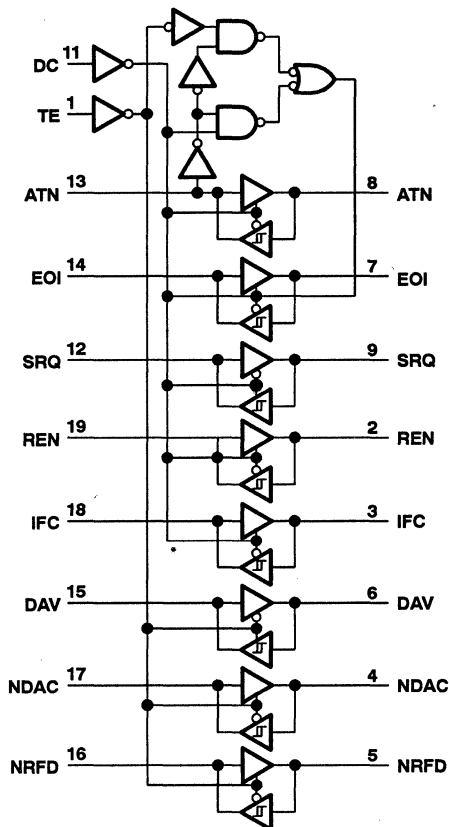


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

▽ Designates 3-state outputs

⊕ Designates passive-pullup outputs

## logic diagram (positive logic)



# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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**RECEIVE/TRANSMIT FUNCTION TABLE**

CONTROLS			BUS-MANAGEMENT CHANNELS					DATA-TRANSFER CHANNELS			
DC	TE	ATN†	ATN†	SRQ	REN	IFC	EOI	DAV	NDAC	NRFD	
			(controlled by DC)						(controlled by TE)		
H	H	H	R	T	R	R	T	T	R	R	
H	H	L					R				
L	L	H	T	R	T	T	R	R	T	T	
L	L	L					T				
H	L	X	R	T	R	R	R	R	T	T	
L	H	X	T	R	T	T	T	T	R	R	

H = high level, L = low level, R = receive, T = transmit, X = irrelevant

Direction of data transmission is from the terminal side to the bus side, and the direction of data receiving is from the bus side to the terminal side.

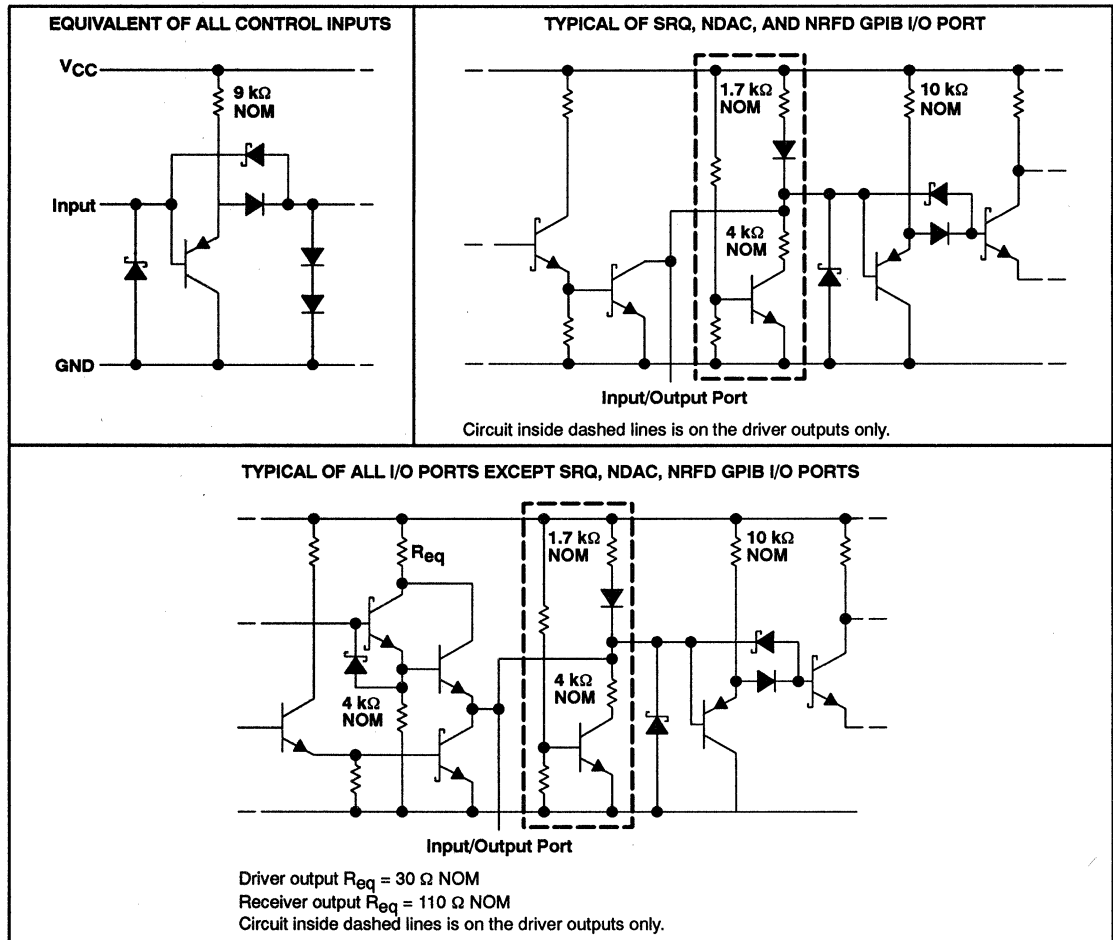
Data transfer is noninverting in both directions.

† ATN is a normal transceiver channel that functions additionally as an internal direction control or talk enable for EOI whenever the DC and TE inputs are in the same state. When DC and TE are in opposite states, the ATN channel functions as an independent transceiver only.

# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

SLLS019C - D2618, JUNE 1986 - REVISED FEBRUARY 1993

## schematics of inputs and outputs



# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

SLLS019C – D2618, JUNE 1986 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Low-level driver output current	100 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN55ALS161	–55°C to 125°C
SN75ALS161	0°C to 70°C
Storage temperature range	–65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from the case for 60 seconds: J or W package	300°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds: DW or N package	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	—
W	1000 mW	8.0 mW/°C	640 mW	200 mW

## SN55ALS161 recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	TE and DC at $T_A = -55^\circ\text{C}$ to $125^\circ\text{C}$	2			V
	Bus and terminal at $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	2			
	Bus and terminal at $T_A = -55^\circ\text{C}$	2.1			
Low-level input voltage, $V_{IL}$	TE and DC at $T_A = -55^\circ\text{C}$ to $125^\circ\text{C}$				V
	Bus and terminal at $T_A = 25^\circ\text{C}$ to $-55^\circ\text{C}$	0.8			
	Bus and terminal at $T_A = 125^\circ\text{C}$	0.7			
High-level output current, $I_{OH}$	Bus ports with pullups active ( $V_{CC} = 5\text{ V}$ )	–5.2			mA
	Terminal ports	–800			$\mu\text{A}$
Low-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			
Operating free-air temperature, $T_A$		–55		125	°C

## SN75ALS171 recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$		2			V
Low-level input voltage, $V_{IL}$					0.8 V
High-level output current, $I_{OH}$	Bus ports with pullups active	–5.2			mA
	Terminal ports	–800			$\mu\text{A}$
Low-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			
Operating free-air temperature, $T_A$		0		70	°C



electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		SN55ALS161			SN75ALS161			UNIT		
				MIN	TYP‡	MAX	MIN	TYP‡	MAX			
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$		-0.8 -1.5			-0.8 -1.5			V		
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	Bus						0.4 0.65			V	
		Bus	$V_{CC} = 5 \text{ V}$ , $T_A = -55^\circ\text{C}$ and $25^\circ\text{C}$	0.4 0.55								
$V_{OH}^{\S}$	High-level output voltage	Terminal	$I_{OH} = -800 \mu\text{A}$		2.7 3.5			2.7 3.5			V	
		Bus	$I_{OH} = -5.2 \text{ mA}$ , $V_{CC} = 5 \text{ V}$ (SN55 <sup>†</sup> )	2.2		2.2						
$V_{OL}$	Low-level output voltage	Terminal	$I_{OL} = 16 \text{ mA}$ , $V_{CC} = \text{MIN}$		0.3 0.5			0.3 0.5			V	
		Bus	$I_{OL} = 48 \text{ mA}$ , $V_{CC} = \text{MIN}$	$T_A = -55^\circ\text{C}$ and $25^\circ\text{C}$ (SN55 <sup>†</sup> )		0.35 0.5			0.35 0.5			
				$T_A = 25^\circ\text{C}$ (SN55 <sup>†</sup> )		0.35 0.5			0.35 0.5			
$I_I$	Input current at maximum input voltage	Terminal	$V_I = 5.5 \text{ V}$ , $V_{CC} = \text{MAX}$		0.2 100			0.2 100			$\mu\text{A}$	
$I_{IH}$	High-level input current	Terminal	$V_I = 2.7 \text{ V}$ , $V_{CC} = \text{MAX}$		0.1 20			0.1 20			$\mu\text{A}$	
$I_{IL}$	Low-level input current	Terminal and control inputs	$V_I = 0.5 \text{ V}$ , $V_{CC} = \text{MAX}$		-30 -100			-10 -100			$\mu\text{A}$	
$V_{I/O(\text{bus})}$	Voltage at bus port	Driver disabled, $V_{CC} = 5 \text{ V}$ (SN55 <sup>†</sup> )		$I_I(\text{bus}) = 0$	2.5 3 3.7			2.5 3 3.7			V	
				$I_I(\text{bus}) = -12 \text{ mA}$	-1.5			-1.5				
$I_{I/O(\text{bus})}$	Current into bus port	Power on	Driver disabled, $V_{CC} = 5 \text{ V}$ (SN55 <sup>†</sup> )	$V_I(\text{bus}) = -1.5 \text{ V to } 0.4 \text{ V}$	-1.3			-1.3			mA	
				$V_I(\text{bus}) = 0.4 \text{ V to } 2.5 \text{ V}$	0 -3.2			0 -3.2				
				$V_I(\text{bus}) = 2.5 \text{ V to } 3.7 \text{ V}$	2.5 -3.2			2.5 -3.2				
				$V_I(\text{bus}) = 3.7 \text{ V to } 5 \text{ V}$	0 2.5			0 2.5				
				$V_I(\text{bus}) = 5 \text{ V to } 5.5 \text{ V}$	0.7 2.5			0.7 2.5				
				$V_I(\text{bus}) = 0 \text{ to } 2.5 \text{ V}$	40			40				
$I_{OS}$	Short-circuit output current	Terminal	$V_{CC} = \text{MAX}$		-15 -35 -75			-15 -35 -75			mA	
		Bus			-25 -50 -125			-25 -50 -125				
$I_{CC}$	Supply current	No load, TE and DC low, $V_{CC} = \text{MAX}$		55 90			55 75			mA		
$C_{I/O(\text{bus})}$	Bus-port capacitance	$V_{CC} = 0 \text{ to } 5 \text{ V}$ , $V_{I/O} = 0 \text{ to } 2 \text{ V}$ , $f = 1 \text{ MHz}$		30			30			pF		

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

§  $V_{OH}$  and  $I_{OS}$  apply to 3-state outputs only.

# SN55ALS161

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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**SN55ALS161 switching characteristics  $V_{CC} = 4.75\text{ V}$ ,  $V_{CC} = 5\text{ V}$ , and  $V_{CC} = 5.25\text{ V}$  and  $C_L = 50\text{ pF}$  (unless otherwise noted)**

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	$T_A$ †	MIN	TYP‡	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Terminal	Bus (Except SRQ, NDAC, and NRFD)	See Figure 1	25°C		10	17	ns
				Full range			20	
t <sub>PHL</sub> Propagation delay time, high-to-low-level output	Terminal	Bus (Except SRQ, NDAC, and NRFD)	See Figure 1	25°C		10	14	ns
				Full range			16	
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Terminal	Bus (NRFD, SRQ, NDAC)	See Figure 2	25°C			25	ns
				Full range			30	
t <sub>PHL</sub> Propagation delay time, high-to-low-level output	Terminal	Bus (NRFD, SRQ, NDAC)	See Figure 2	25°C		10	14	ns
				Full range			16	
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	Bus	Terminal	See Figure 2	25°C		10	15	ns
				Full range			18	
t <sub>PHL</sub> Propagation delay time, high-to-low-level output	Bus	Terminal	See Figure 2	25°C		10	15	ns
				Full range			18	
t <sub>PZH</sub> Output enable time to high level	TE or DC	Bus (ATN, REN, IFC, and DAV)	See Figure 3	25°C		20	30	ns
				Full range			41	
t <sub>PHZ</sub> Output disable time from high level	TE or DC	Bus (ATN, REN, IFC, and DAV)	See Figure 3	25°C		8	14	ns
				Full range			16	
t <sub>PZL</sub> Output enable time to low level	TE or DC	Bus (ATN, REN, IFC, and DAV)	See Figure 3	25°C		16	28	ns
				Full range			34	
t <sub>PLZ</sub> Output disable time from low level	TE or DC	Bus (ATN, REN, IFC, and DAV)	See Figure 3	25°C		10	19	ns
				Full range			24	
t <sub>PZH</sub> Output enable time to high level	TE or DC	Bus (EOI)	See Figure 3	25°C		24	30	ns
				Full range			48	
t <sub>PHZ</sub> Output disable time from high level	TE or DC	Bus (EOI)	See Figure 3	25°C		13	19	ns
				Full range			25	
t <sub>PZL</sub> Output enable time to low level,	TE or DC	Bus (EOI)	See Figure 3	25°C		21	35	ns
				Full range			43	
t <sub>PLZ</sub> Output disable time from low level	TE or DC	Bus (EOI)	See Figure 3	25°C		13	20	ns
				Full range			27	
t <sub>PZH</sub> Output enable time to high level	TE or DC	Terminal	See Figure 4	25°C		24	36	ns
				Full range			50	
t <sub>PHZ</sub> Output disable time from high level	TE or DC	Terminal	See Figure 4	25°C		12	20	ns
				Full range			33	
t <sub>PZL</sub> Output enable time to low level	TE or DC	Terminal	See Figure 4	25°C		20	34	ns
				Full range			41	
t <sub>PLZ</sub> Output disable time from low level	TE or DC	Terminal	See Figure 4	25°C		13	24	ns
				Full range			35	

† Full range is  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

‡ All typical values are at  $V_{CC} = 5\text{ V}$ .





# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

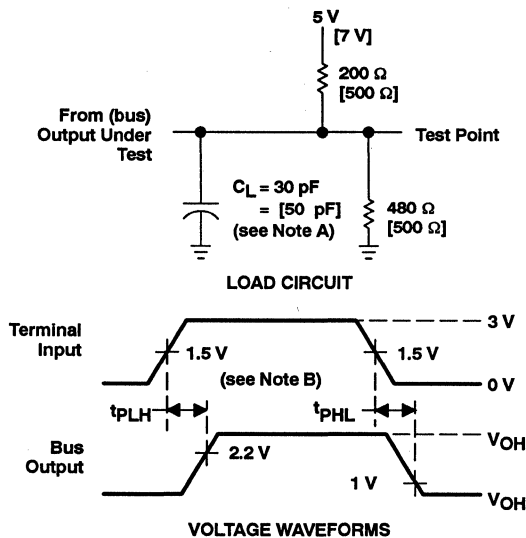
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**SN75ALS161 switching characteristics over recommended range of operating free-air temperature,  $V_{CC} = 5\text{ V}$**

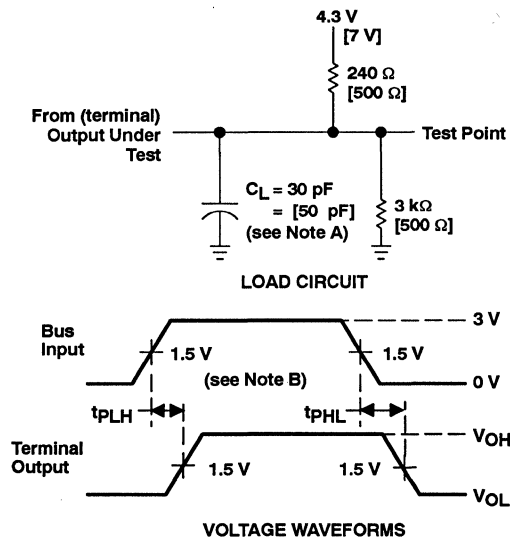
PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	Terminal	Bus	$C_L = 30\text{ pF}$ , See Figure 1		10	20	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output					12	20	
$t_{PLH}$ Propagation delay time, low-to-high-level output	Bus	Terminal	$C_L = 30\text{ pF}$ , See Figure 2		5	10	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output					7	14	
$t_{PZH}$ Output enable time to high level	TE or DC	Bus (ATN, EOI, REN, IFC, and DAV)	$C_L = 15\text{ pF}$ , See Figure 3			30	ns
$t_{PHZ}$ Output disable time from high level						20	
$t_{PZL}$ Output enable time to low level						45	
$t_{PLZ}$ Output disable time from low level						20	
$t_{PZH}$ Output enable time to high level	TE or DC	Terminal	$C_L = 15\text{ pF}$ , See Figure 4			30	ns
$t_{PHZ}$ Output disable time from high level						25	
$t_{PZL}$ Output enable time to low level						30	
$t_{PLZ}$ Output disable time from low level						25	

† All typical values are at  $T_A = 25^\circ\text{C}$ .

## PARAMETER MEASUREMENT INFORMATION



**Figure 1. Terminal-to-Bus Load Circuit and Voltage Waveforms**



**Figure 2. Bus-to-Terminal Load Circuit and Voltage Waveforms**

[ ] denotes the SN55ALS161 military test conditions.

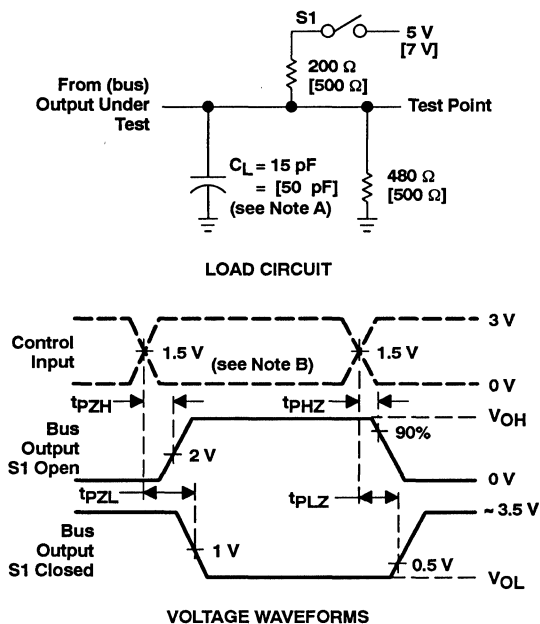
NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq 1\text{ MHz}$ , 50% duty cycle,  $t_r \leq 6\text{ ns}$ ,  $t_f \leq 6\text{ ns}$ ,  $Z_0 = 50\ \Omega$ .

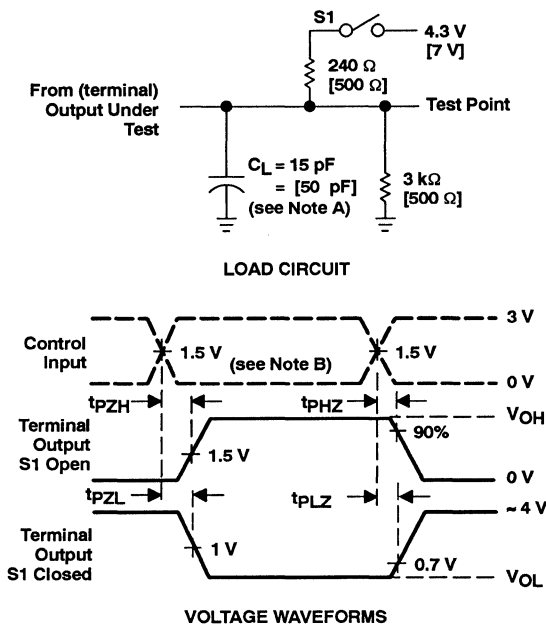
# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION



**Figure 3. Bus Load Circuit and Voltage Waveforms**



**Figure 4. Terminal Load Circuit and Voltage Waveforms**

[ ] denotes the SN55ALS161 military test conditions.

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

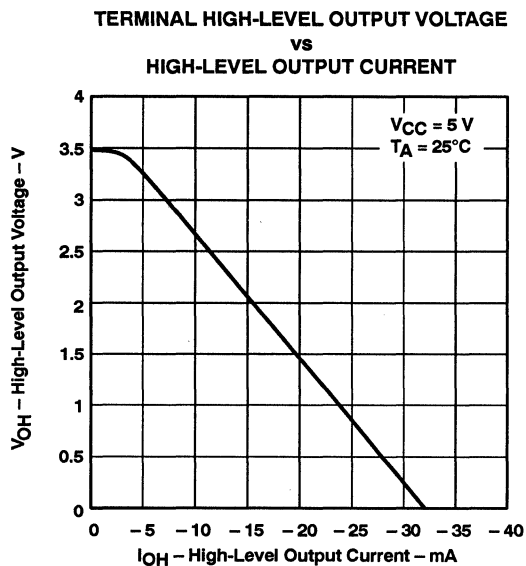


Figure 5

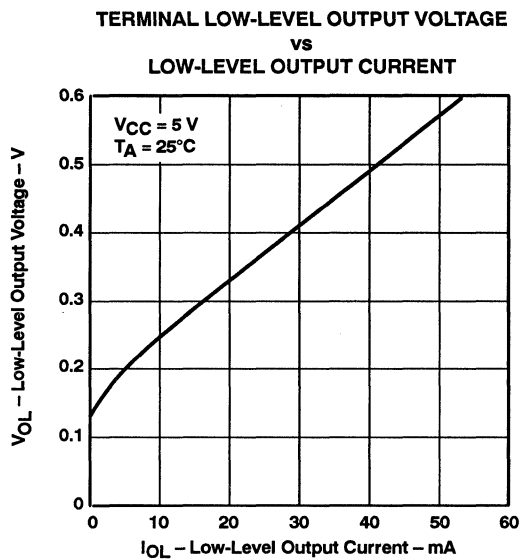


Figure 6

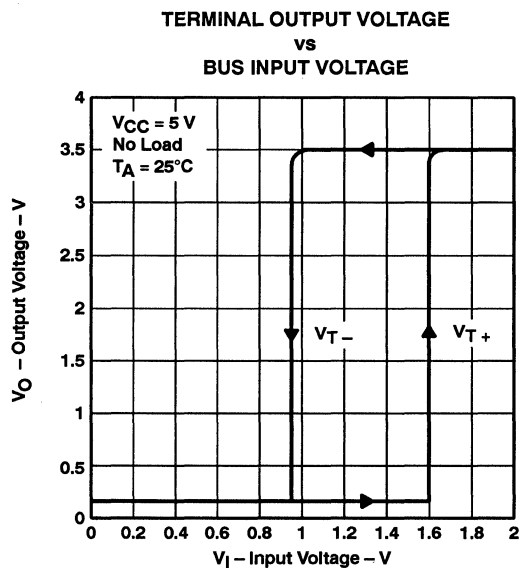


Figure 7

# SN55ALS161, SN75ALS161 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

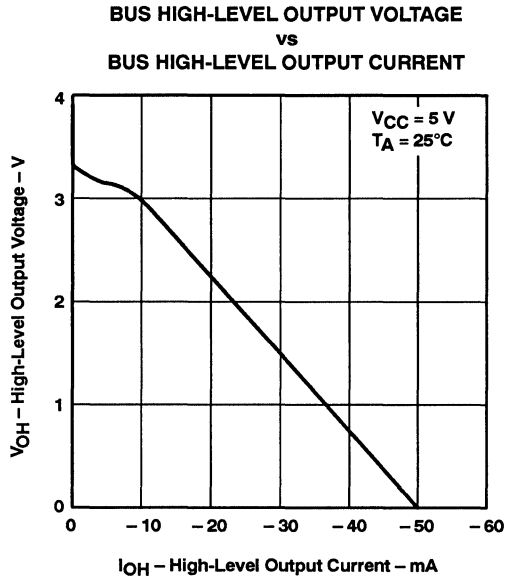


Figure 8

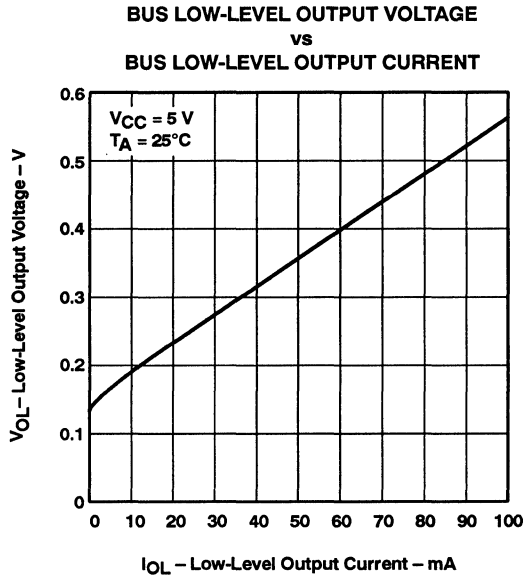


Figure 9

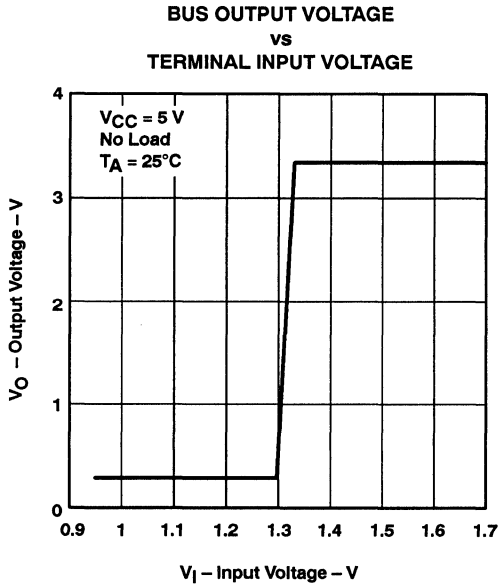


Figure 10

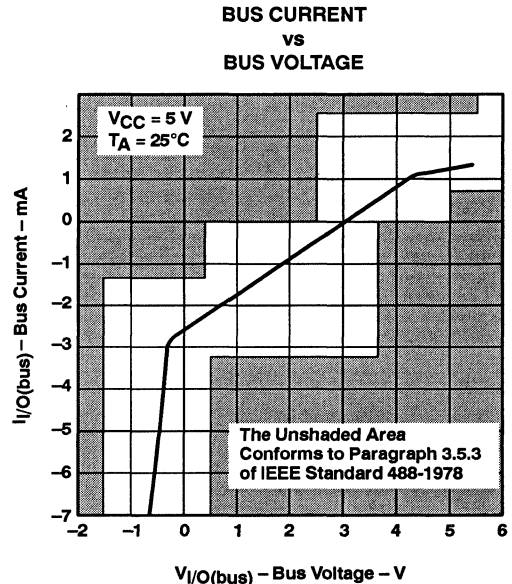


Figure 11



# SN75ALS162

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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### MEETS IEEE STANDARD 488-1978 ( GPIB )

- 8-Channel Bidirectional Transceiver
- Designed to Implement Control Bus Interface
- Designed for Multicontrollers
- High-Speed Advanced Low-Power Schottky Circuitry
- Low-Power Dissipation . . . 46 mW Max per Channel
- Fast Propagation Times . . . 20 ns Max
- High-impedance PNP Inputs
- Receiver Hysteresis . . . 650 mV Typ
- Bus-Terminating Resistors Provided on Driver Outputs
- No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )
- Power-Up/Power-Down Protection (Glitch Free)

#### description

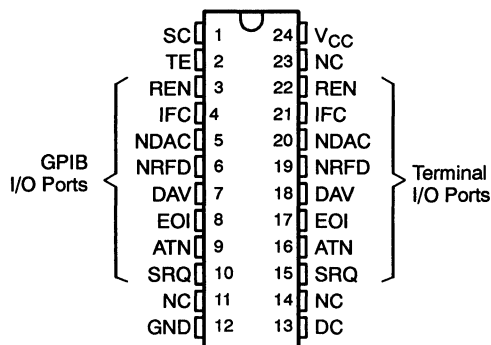
The SN75ALS162 eight-channel general-purpose interface bus transceiver is a monolithic, high-speed, advanced low-power Schottky process device designed to provide the bus-management and data-transfer signals between operating units of a multiple-controller instrumentation system. When combined with the SN75ALS160 octal bus transceiver, the SN75ALS162 provides the complete 16-wire interface for the IEEE 488 bus.

The SN75ALS162 features eight driver-receiver pairs connected in a front-to-back configuration to form input/output (I/O) ports at both the bus and terminal sides. The direction of data through these driver-receiver pairs is determined by the DC, TE, and SC enable signals. The SC input allows the REN and IFC transceivers to be controlled independently.

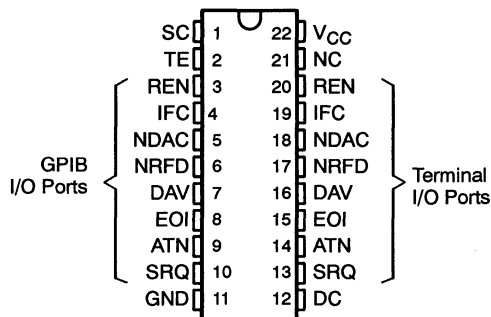
The driver outputs (GPIB I/O ports) feature active bus-terminating resistor circuits designed to provide a high impedance to the bus when  $V_{CC} = 0$ . The drivers are designed to handle loads up to 48 mA of sink current. Each receiver features pnp transistor inputs for high input impedance and hysteresis of 400 mV minimum for increased noise immunity. All receivers have 3-state outputs to present a high impedance to the terminal when disabled.

The SN75ALS162 is characterized for operation from 0°C to 70°C.

DW PACKAGE  
(TOP VIEW)



N PACKAGE  
(TOP VIEW)



NC--No internal connection

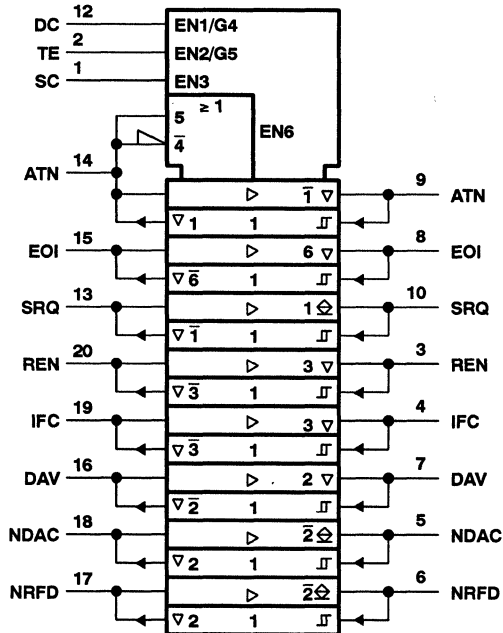
# SN75ALS162 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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CHANNEL IDENTIFICATION TABLE

NAME	IDENTITY	CLASS
DC	Direction Control	Control
TE	Talk Enable	
SC	System Control	
ATN	Attention	Bus Management
SRQ	Service Request	
REN	Remote Enable	
IFC	Interface Clear	
EOI	End or Identify	
DAV	Data Valid	Data Transfer
NDAC	No Data Accepted	
NRFD	Not Ready for Data	

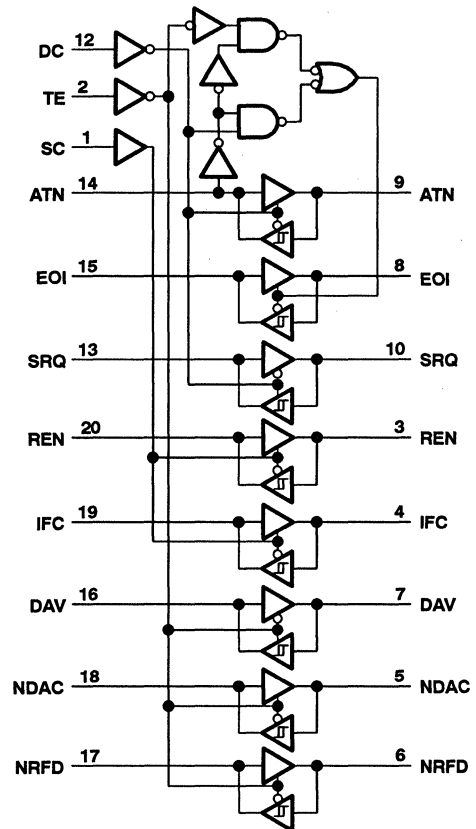
logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

- ▽ Designates 3-state outputs
- ⊗ Designates passive-pullup outputs

logic diagram (positive logic)



Pin numbers shown are for the N package.

# SN75ALS162

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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**RECEIVE/TRANSMIT FUNCTION TABLE**

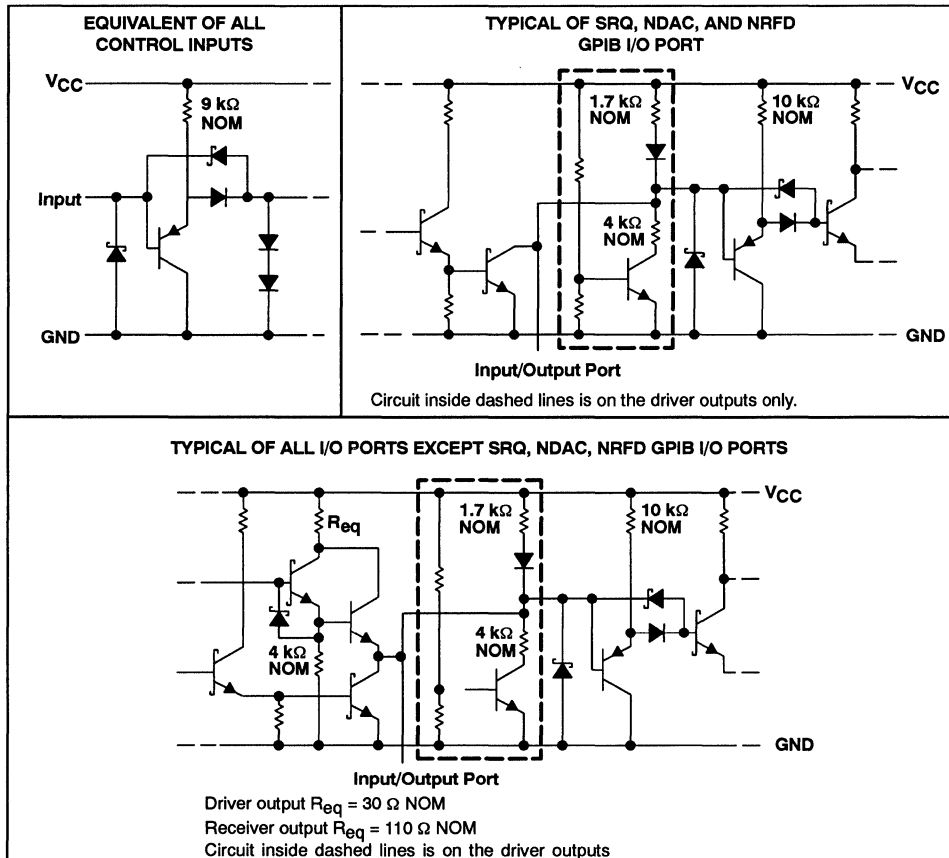
CONTROLS				BUS-MANAGEMENT CHANNELS				DATA-TRANSFER CHANNELS			
SC	DC	TE	ATN†	ATN† (controlled by DC)	SRQ	REN (controlled by SC)	IFC	EOI	DAV	NDAC	NRFD
	H	H	H		R	T		T		T	R
	H	H	L		T	R		R			
	L	L	H		R	T		T			
	L	L	L		R	T		T			
	H	L	X		R	T		T			
L	H	X	T	R	R	R					
H						T	T				
L						R	R				

H = high level, L = low level, R = receive, T = transmit, X = irrelevant

Direction of data transmission is from the terminal side to the bus side, and the direction of data receiving is from the bus side to the terminal side. Data transfer is noninverting in both directions.

† ATN is a normal transceiver channel that functions additionally as an internal direction control or talk enable for EOI whenever the DC and TE inputs are in the same state. When DC and TE are in opposite states, the ATN channel functions as an independent transceiver only.

### schematics of inputs and outputs





# SN75ALS162 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Low-level driver output current	100 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	– 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1350 mW	10.8 mW/°C	864 mW
N	1700 mW	13.6 mW/°C	1088 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$	Bus ports with 3-state outputs		– 5.2	mA
	Terminal ports		– 800	$\mu\text{A}$
Low-level output current, $I_{OL}$	Bus ports		48	mA
	Terminal ports		16	
Operating free-air temperature, $T_A$	0		70	°C



# SN75ALS162

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$		-0.8	-1.5		V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	Bus		0.4	0.65		V
$V_{OH}\ddagger$	High-level output voltage	Terminal	$I_{OH} = -800 \mu\text{A}$	2.7	3.5		V
		Bus	$I_{OH} = -5.2 \text{ mA}$	2.5	3.3		
$V_{OL}$	Low-level output voltage	Terminal	$I_{OL} = 16 \text{ mA}$	0.3	0.5		V
		Bus	$I_{OL} = 48 \text{ mA}$	0.35	0.5		
$I_I$	Input current at maximum input voltage	Terminal	$V_I = 5.5 \text{ V}$	0.2	100		$\mu\text{A}$
$I_{IH}$	High-level input current	Terminal and control inputs		0.1	20		$\mu\text{A}$
$I_{IL}$	Low-level input current			-10	-100		$\mu\text{A}$
$V_{I/O(\text{bus})}$	Voltage at bus port	Driver disabled	$I_{I(\text{bus})} = 0$	2.5	3.0	3.7	V
			$I_{I(\text{bus})} = -12 \text{ mA}$			-1.5	
$I_{I/O(\text{bus})}$	Current into bus port	Power on	Driver disabled	$V_{I(\text{bus})} = -1.5 \text{ V to } 0.4 \text{ V}$	-1.3		mA
				$V_{I(\text{bus})} = 0.4 \text{ V to } 2.5 \text{ V}$	0	-3.2	
				$V_{I(\text{bus})} = 2.5 \text{ V to } 3.7 \text{ V}$		+2.5 -3.2	
				$V_{I(\text{bus})} = 3.7 \text{ V to } 5 \text{ V}$	0	2.5	
				$V_{I(\text{bus})} = 5 \text{ V to } 5.5 \text{ V}$	0.7	2.5	
	Power off	$V_{CC} = 0,$	$V_{I(\text{bus})} = 0 \text{ to } 2.5 \text{ V}$		-40		$\mu\text{A}$
$I_{OS}$	Short-circuit output current	Terminal		-15	-35	-75	mA
		Bus		-25	-50	-125	
$I_{CC}$	Supply current	No load, TE, DC, and SC low		55	75		mA
$C_{I/O(\text{bus})}$	Bus-port capacitance	$V_{CC} = 0 \text{ to } 5 \text{ V}, V_{I/O} = 0 \text{ to } 2 \text{ V}, f = 1 \text{ MHz}$		30			pF

† All typical values are at  $V_{CC} = 5 \text{ V}, T_A = 25^\circ\text{C}$ .

‡  $V_{OH}$  applies to 3-state outputs only.

# SN75ALS162 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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switching characteristics over recommended range of operating free-air temperature,  $V_{CC} = 5\text{ V}$

PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	Terminal	Bus	$C_L = 30\text{ pF}$ , See Figure 1	10	20	ns	
$t_{PHL}$	Propagation delay time, high-to-low-level output				12	20		
$t_{PLH}$	Propagation delay time, low-to-high-level output	Bus	Terminal	$C_L = 30\text{ pF}$ , See Figure 2	5	10	ns	
$t_{PHL}$	Propagation delay time, high-to-low-level output				7	14		
$t_{pZH}$	Output enable time to high level	TE, DC, or SC	Bus (ATN, EOI, REN, IFC, and DAV)	$C_L = 15\text{ pF}$ , See Figure 3	30	ns		
$t_{pHZ}$	Output disable time from high level				20			
$t_{pZL}$	Output enable time to low level				45			
$t_{pLZ}$	Output disable time from low level				20			
$t_{pZH}$	Output enable time to high level	TE, DC, or SC	Terminal	$C_L = 15\text{ pF}$ , See Figure 4	30	ns		
$t_{pHZ}$	Output disable time from high level				25			
$t_{pZL}$	Output enable time to low level				30			
$t_{pLZ}$	Output disable time from low level				25			

† All typical values are at  $T_A = 25^\circ\text{C}$ .

## PARAMETER MEASUREMENT INFORMATION

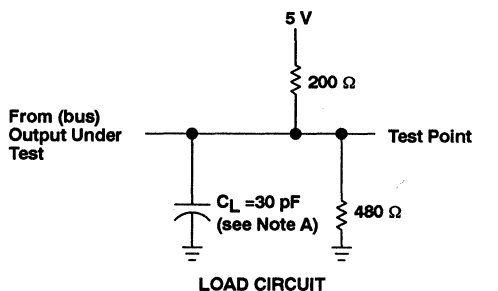


Figure 1. Terminal-to-Bus Load Circuit and Voltage Waveforms

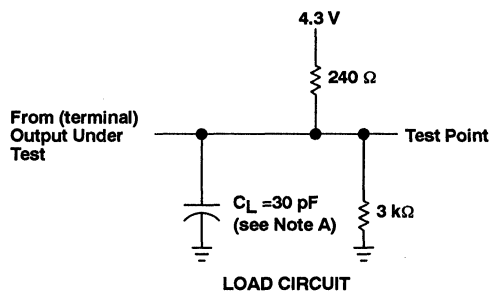


Figure 2. Bus-to-Terminal Load Circuit and Voltage Waveforms

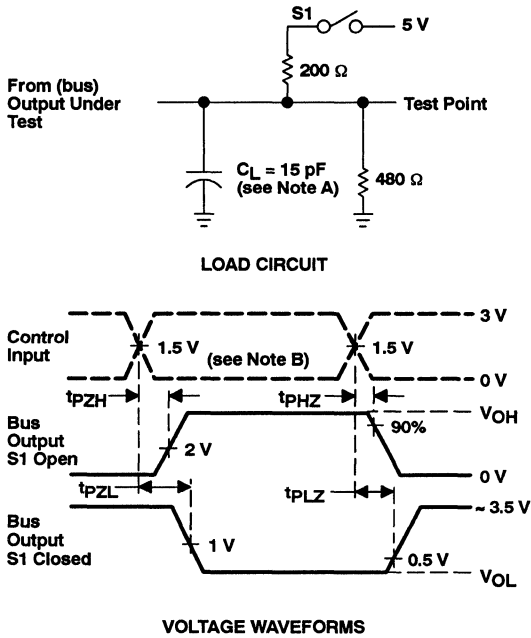
NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1\text{ MHz}$ , 50% duty cycle,  $t_r \leq 6\text{ ns}$ ,  $t_f \leq 6\text{ ns}$ ,  $Z_0 = 50\ \Omega$ .

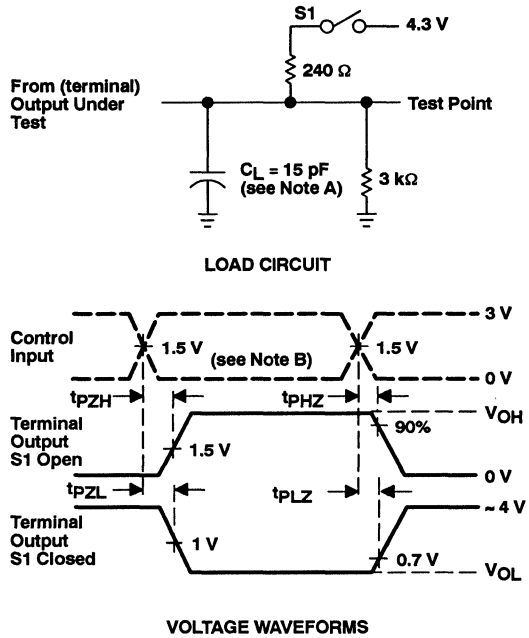
# SN75ALS162 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## PARAMETER MEASUREMENT INFORMATION



**Figure 3. Bus Load Circuit and Voltage Waveforms**



**Figure 4. Terminal Load Circuit and Voltage Waveforms**

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

# SN75ALS162 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## TYPICAL CHARACTERISTICS

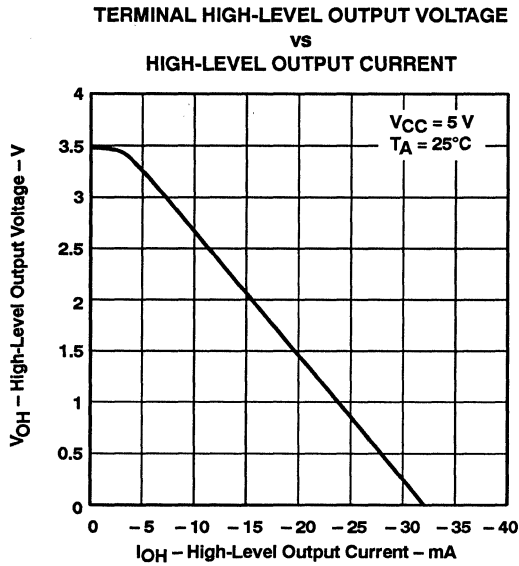


Figure 5

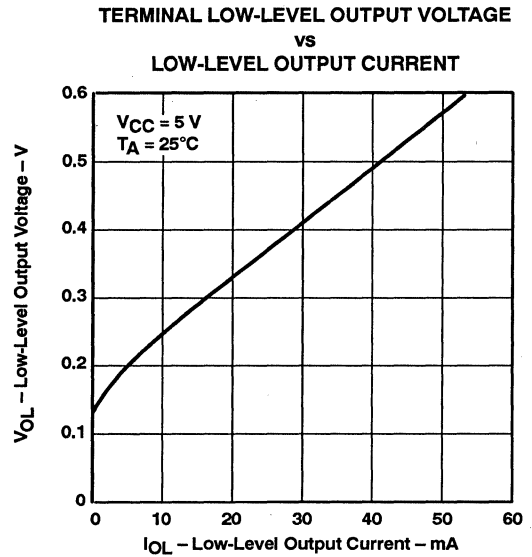


Figure 6

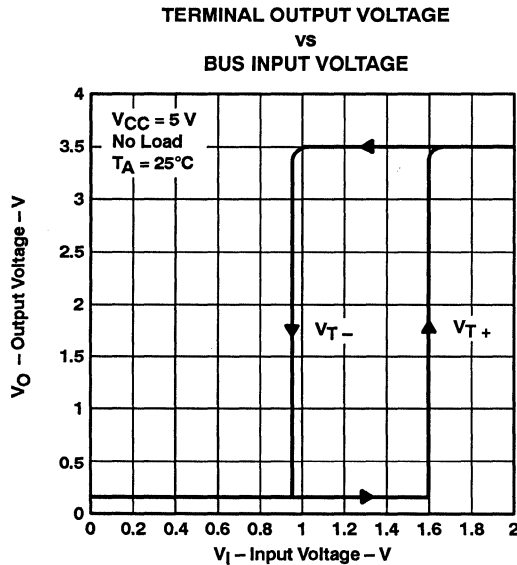


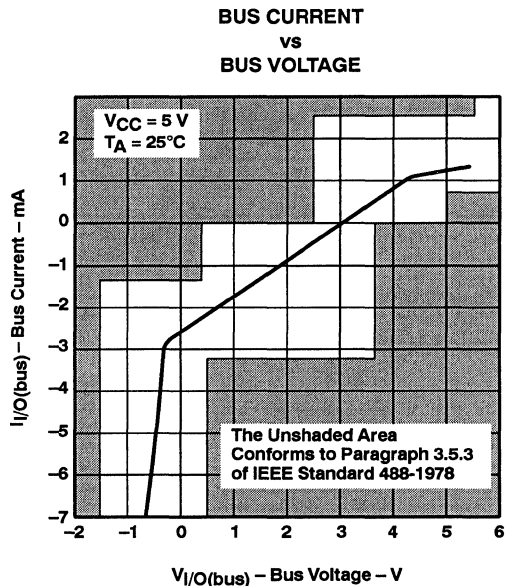
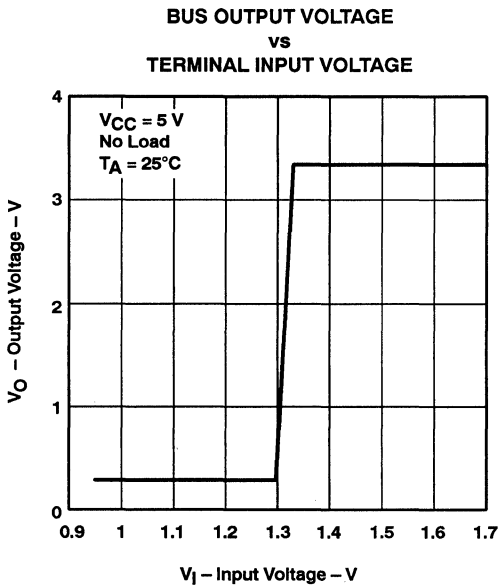
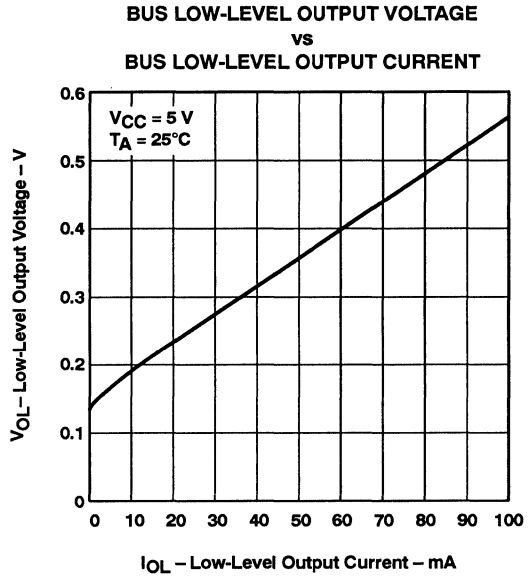
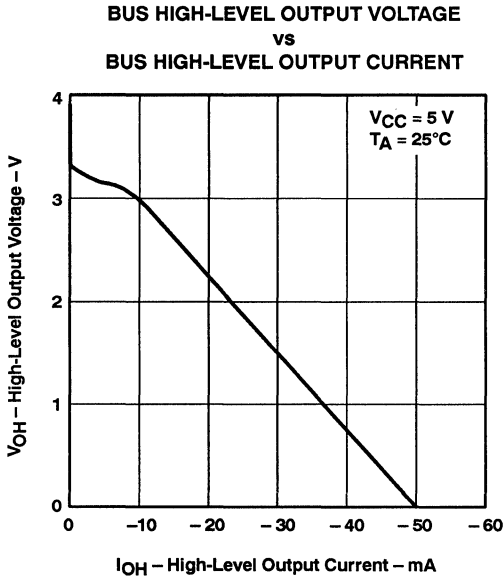
Figure 7

# SN75ALS162

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS020B – D2618, JUNE 1986 – REVISED AUGUST 1989

### TYPICAL CHARACTERISTICS

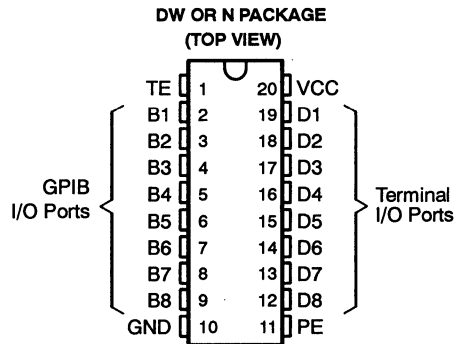




# SN75163B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A – D2611, OCTOBER 1985 – REVISED FEBRUARY 1993

- 8-Channel Bidirectional Transceivers
- Power-Up/Power-Down Protection (Glitch Free)
- High-Speed Low-Power Schottky Circuitry
- Low Power Dissipation . . . 66 mW Max Per Channel
- High-impedance PNP Inputs
- Receiver Hysteresis . . . 650 mV Typ
- Open-Collector Driver Output Option
- No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )



## description

The SN75163B octal general-purpose interface bus transceiver is a monolithic, high-speed, low-power Schottky device. It is designed for two-way data communications over single-ended transmission lines. The transceiver features driver outputs that can be operated in either the open-collector or 3-state modes. If talk enable (TE) is high, these outputs have the characteristics of open-collector outputs when pullup enable (PE) is low and of 3-state outputs when PE is high. Taking TE low places the outputs in the high-impedance state. The driver outputs are designed to handle loads of up to 48 mA of sink current. Each receiver features pnp transistor inputs for high input impedance and 400 mV of hysteresis for increased noise immunity.

**NOT RECOMMENDED FOR NEW DESIGN**

Output glitches during power up and power down are eliminated by an internal circuit that disables both the bus and receiver outputs. The outputs do not load the bus when  $V_{CC} = 0$ .

The SN75163B is characterized for operation from 0°C to 70°C.

### Function Tables

EACH DRIVER			OUTPUT B
INPUTS			
D	TE	PE	
H	H	H	H
L	H	H	L
H	X	L	Z
L	H	L	L
X	L	X	Z

EACH RECEIVER			OUTPUT D
INPUTS			
B	TE	PE	
L	L	X	L
H	L	X	H
X	H	X	Z

H = high level, L = low level, X = irrelevant, Z = high-impedance state

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

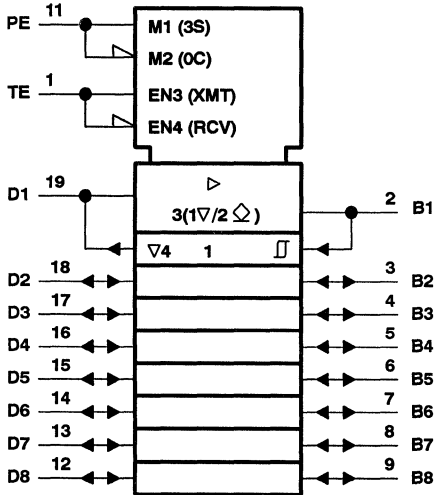




# SN75163B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A - D2611, OCTOBER 1985 - REVISED FEBRUARY 1993

## logic symbol†

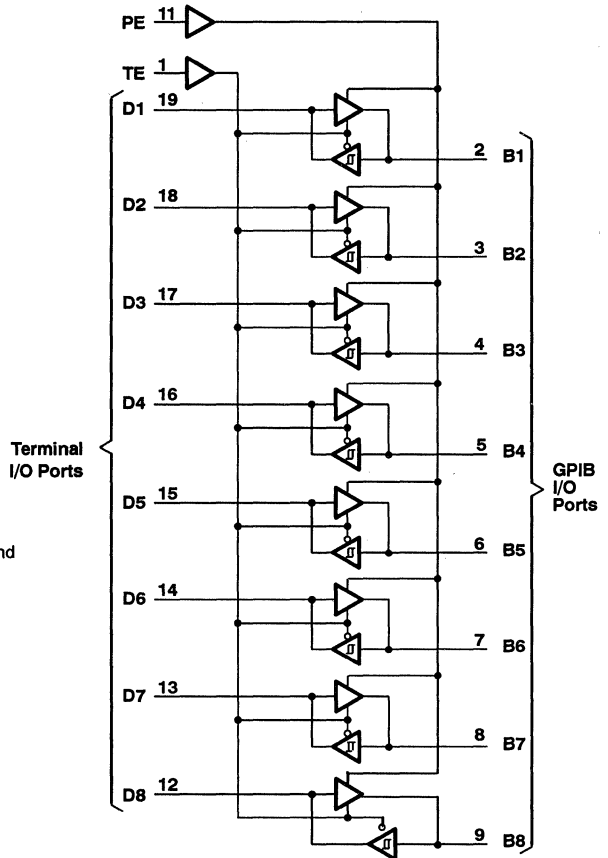


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

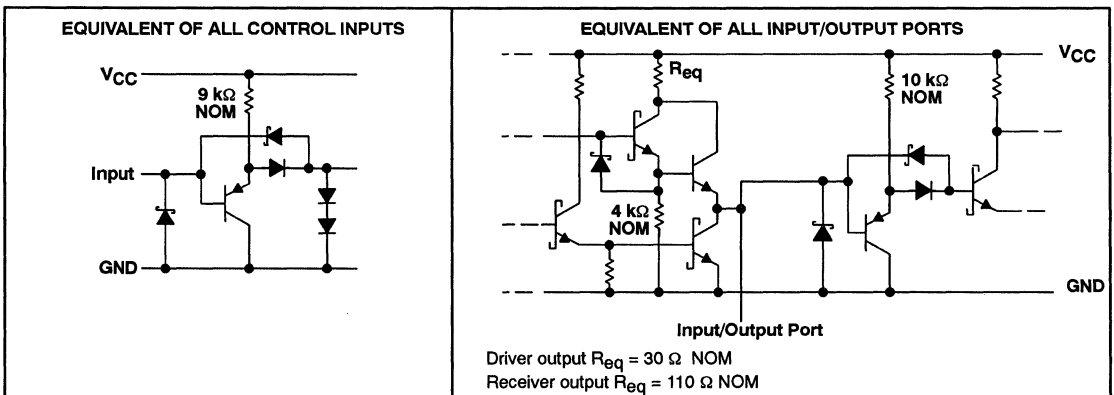
∇ Designates 3-state outputs

◇ Designates open-collector outputs

## logic diagram (positive logic)



## schematics of inputs and outputs



# SN75163B

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A – D2611, OCTOBER 1985 – REVISED FEBRUARY 1993

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Input voltage .....	5.5 V
Low-level driver output current .....	100 mA
Continuous total power dissipation (see Note 2) .....	See Dissipation Rating Table
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	– 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds .....	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
N	1150 mW	9.2 mW/°C	736 mW

### recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$		2			V
Low-level input voltage, $V_{IL}$		0.8			V
High-level output current, $I_{OH}$	Bus ports with pullups active	–10			mA
	Terminal ports	–800			μA
High-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			
Operating free-air temperature, $T_A$		0			70 °C



# SN75163B

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A - D2811, OCTOBER 1985 - REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V <sub>IK</sub>	Input clamp voltage		I <sub>I</sub> = -18 mA	-0.8	-1.5		V
V <sub>hys</sub>	Hysteresis (V <sub>T+</sub> - V <sub>T-</sub> )	Bus	See Figure 8	0.4	0.65		V
V <sub>OH</sub>	High-level output voltage	Terminal	I <sub>OH</sub> = -800 μA, TE at 0.8 V	2.7	3.5		V
		Bus	I <sub>OH</sub> = -10 mA, PE and TE at 2 V	2.5	3.3		
V <sub>OL</sub>	Low-level output voltage	Terminal	I <sub>OL</sub> = 16 mA, TE at 0.8 V		0.3	0.5	V
		Bus	I <sub>OL</sub> = 48 mA, PE and TE at 2 V		0.4	0.5	
I <sub>OH</sub>	High-level output current (open-collector mode)	Bus	V <sub>O</sub> = 5.5 V, PE at 0.8 V, D and TE at 2 V			100	μA
I <sub>OZ</sub>	Off-state output current (3-state mode)	Bus	PE at 2 V, TE at 0.8 V	V <sub>O</sub> = 2.7 V		20	μA
				V <sub>O</sub> = 0.4 V		-20	
I <sub>I</sub>	Input current at maximum input voltage	Terminal	V <sub>I</sub> = 5.5 V		0.2	100	μA
I <sub>IH</sub>	High-level input current	Terminal	V <sub>I</sub> = 2.7 V		0.1	20	μA
I <sub>IL</sub>	Low-level input current	Terminal	V <sub>I</sub> = 0.5 V		-10	-100	μA
I <sub>OS</sub>	Short-circuit output current	Terminal		-15	-35	-75	mA
		Bus		-25	-50	-125	
I <sub>IL</sub>	Supply current	No load	Receivers low and enabled			80	mA
			Drivers low and enabled			100	
C <sub>I/O (bus)</sub>	Bus-port capacitance		V <sub>CC</sub> = 5 V to 0, V <sub>I/O</sub> = 0 to 2 V, f = 1 MHz		30		pF

† All typical values are at V<sub>CC</sub> = 5, T<sub>A</sub> = 25°C.

**switching characteristics, V<sub>CC</sub> = 5 V, C<sub>L</sub> = 15 pF, T<sub>A</sub> = 25°C (unless otherwise noted)**

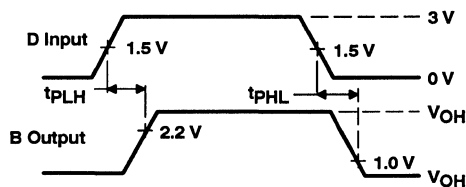
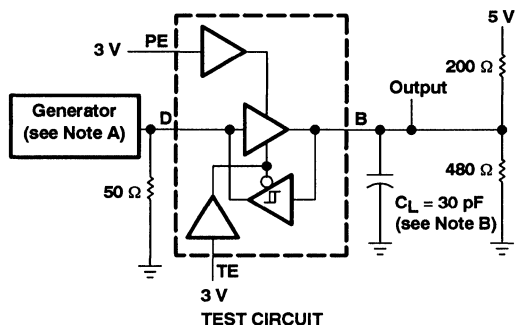
PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Terminal	Bus	C <sub>L</sub> = 30 pF, See Figure 1		14	20	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					14	20	
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Bus	Terminal	C <sub>L</sub> = 30 pF, See Figure 2		10	20	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					15	22	
t <sub>PZH</sub>	Output enable time to high level	TE	Bus	See Figure 3		25	35	ns
t <sub>PHZ</sub>	Output disable time from high level					13	22	
t <sub>PZL</sub>	Output enable time to low level					22	35	
t <sub>PLZ</sub>	Output disable time from low level					22	32	
t <sub>PZH</sub>	Output enable time to high level					20	30	
t <sub>PHZ</sub>	Output disable time from high level	TE	Terminal	See Figure 4		12	20	ns
t <sub>PZL</sub>	Output enable time to low level					23	32	
t <sub>PLZ</sub>	Output disable time from low level					19	30	
t <sub>en</sub>	Output pullup enable time					15	22	
t <sub>dis</sub>	Output pullup disable time	PE	Terminal	See Figure 5		13	20	ns

**TEXAS**  
**INSTRUMENTS**

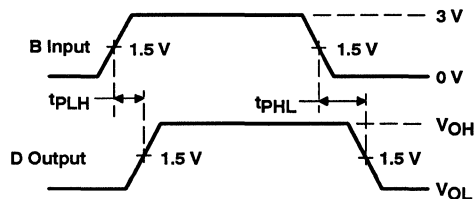
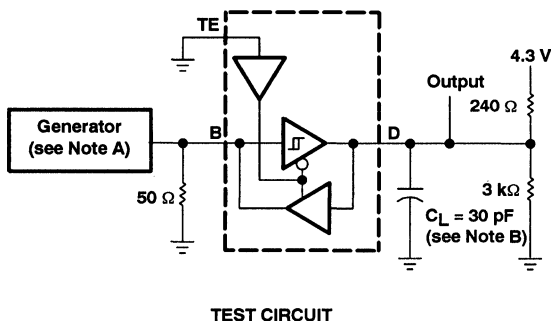
# SN75163B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A - D2611, OCTOBER 1985 - REVISED FEBRUARY 1993

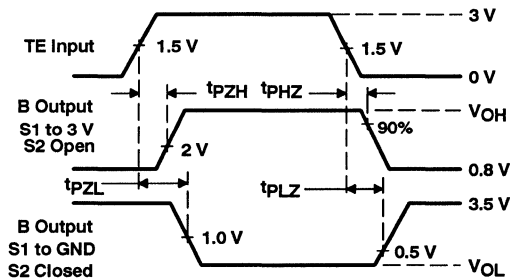
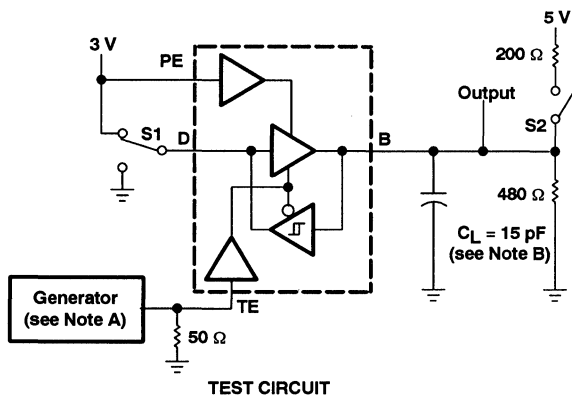
## PARAMETER MEASUREMENT INFORMATION



**Figure 1. Terminal-to-Bus Test Circuit and Voltage Waveforms**



**Figure 2. Bus-to-Terminal Test Circuit and Voltage Waveforms**



**Figure 3. TE-to-Bus Test Circuit and Voltage Waveforms**

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

# SN75163B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A – D2611, OCTOBER 1985 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

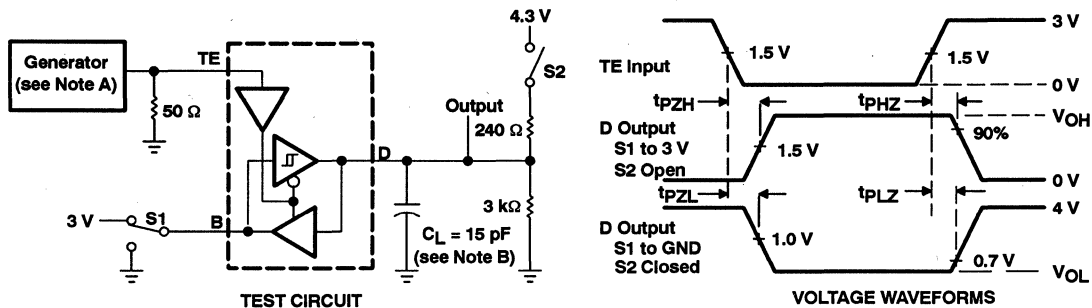


Figure 4. TE-to-Terminal Test Circuit and Voltage Waveforms

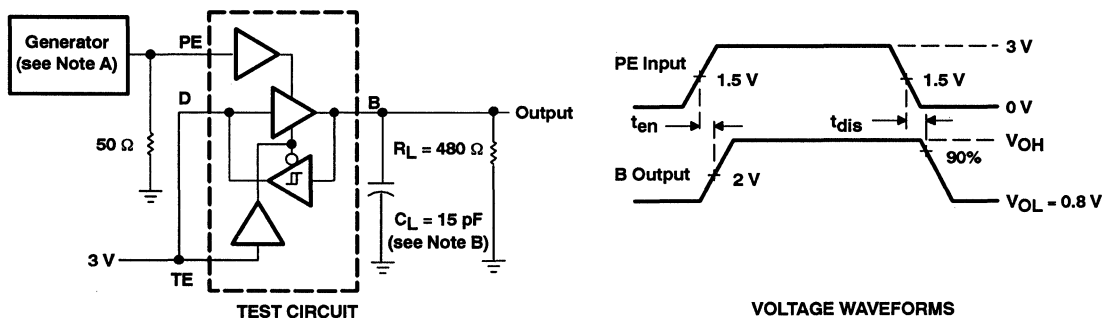


Figure 5. PE-to-Bus Pullup Test Circuit and Voltage Waveforms

- NOTES: C. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .  
 D.  $C_L$  includes probe and jig capacitance.

# SN75163B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A - D2611, OCTOBER 1985 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

**TERMINAL HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

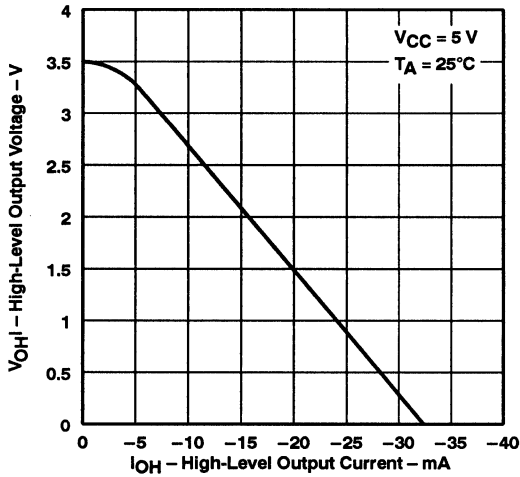


Figure 6

**TERMINAL LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

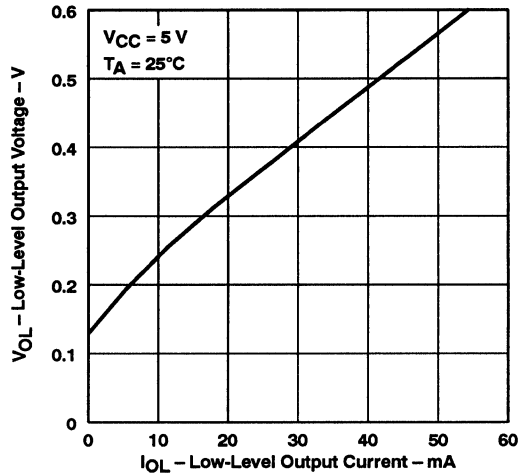


Figure 7

**TERMINAL OUTPUT VOLTAGE  
vs  
BUS INPUT VOLTAGE**

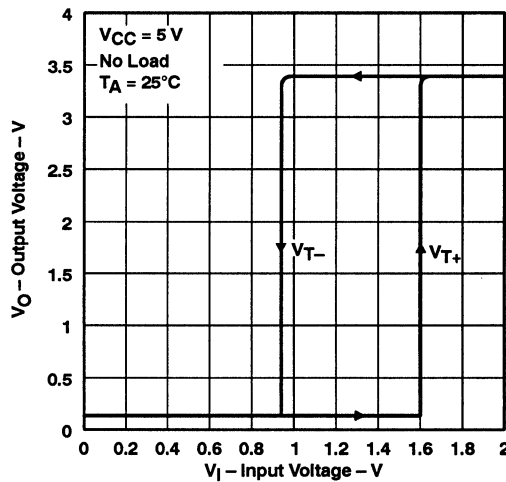


Figure 8

**TEXAS  
INSTRUMENTS**

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2-465

# SN75163B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS006A – D2611, OCTOBER 1985 – REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

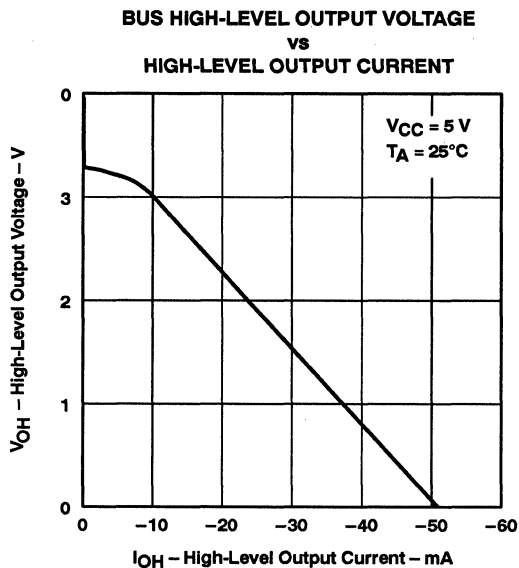


Figure 9

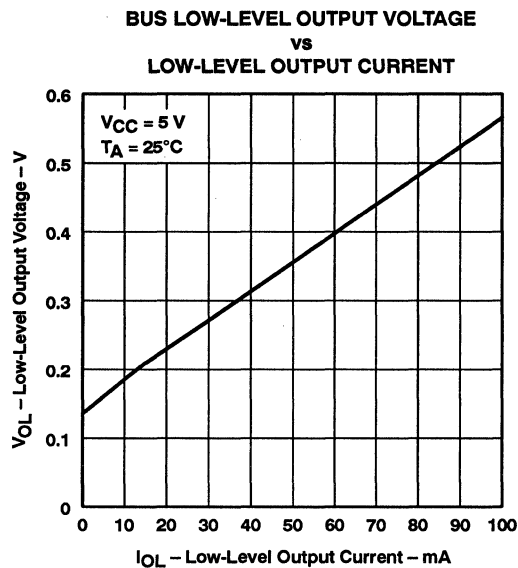


Figure 10

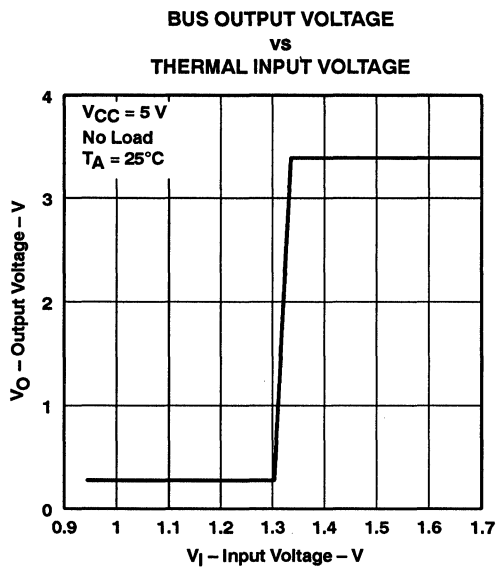


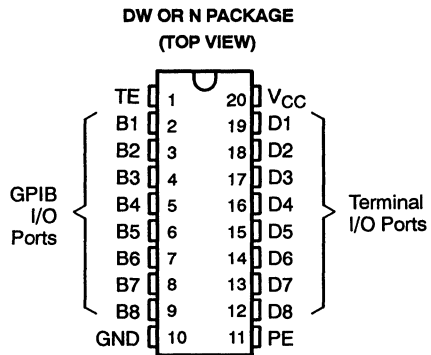
Figure 11

# SN75ALS163

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D – D2611, JUNE 1986 – REVISED FEBRUARY 1993

- **8-Channel Bidirectional Transceiver**
- **High-Speed Advanced Low-Power Schottky Circuitry**
- **Low Power Dissipation . . . 46 mW Max per Channel**
- **Fast Propagation Times . . . 20 ns Max**
- **High-Impedance PNP Inputs**
- **Receiver Hysteresis . . . 650 mV Typ**
- **Open-Collector Driver Output Option**
- **No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )**
- **Power-Up/Power-Down Protection (Glitch Free)**



### description

The SN75ALS163 octal general-purpose interface bus transceiver is a monolithic, high-speed, advanced low-power Schottky device. It is designed for two-way data communications over single-ended transmission lines. The transceiver features driver outputs that can be operated in either the open-collector or 3-state mode. If talk enable (TE) is high, these outputs have the characteristics of open-collector outputs when pullup enable (PE) is low and of 3-state outputs when PE is high. Taking TE low places the outputs in the high-impedance state. The driver outputs are designed to handle loads of up to 48 mA of sink current. Each receiver features pnp transistor inputs for high input impedance and 400 mV minimum of hysteresis for increased noise immunity.

Output glitches during power up and power down are eliminated by an internal circuit that disables both the bus and receiver outputs. The outputs do not load the bus when  $V_{CC} = 0$ .

The SN75ALS163 is characterized for operation from 0°C to 70°C.

### Function Tables

**EACH DRIVER**

INPUTS			OUTPUT
D	TE	PE	B
H	H	H	H
L	H	X	L
H	X	L	Z
X	L	X	Z

**EACH RECEIVER**

INPUTS			OUTPUT
B	TE	PE	D
L	L	X	L
H	L	X	H
X	H	X	Z

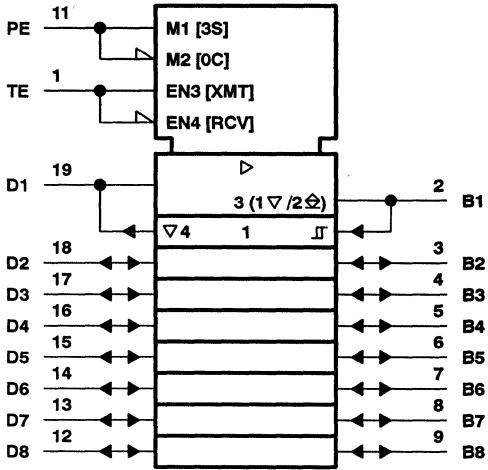
H = high level, L = low level, X = irrelevant,  
Z = high-impedance state



# SN75ALS163 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D - D2611, JUNE 1986 - REVISED FEBRUARY 1993

## logic symbol†

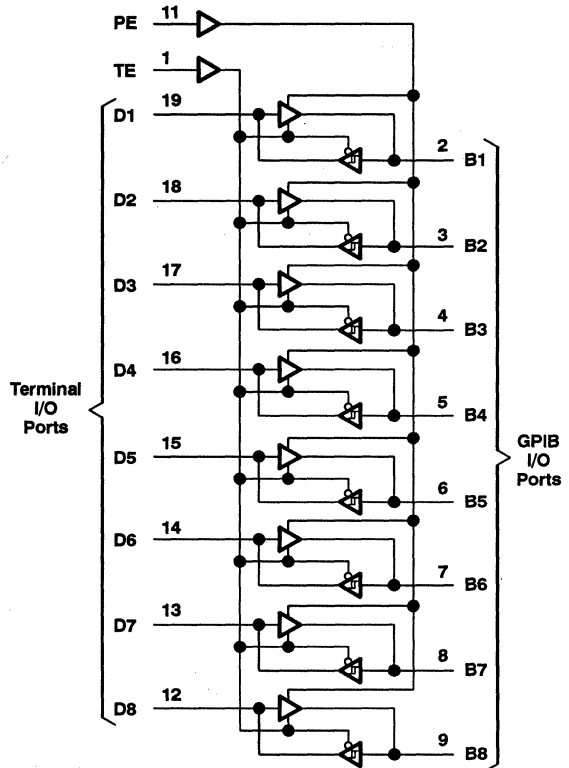


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

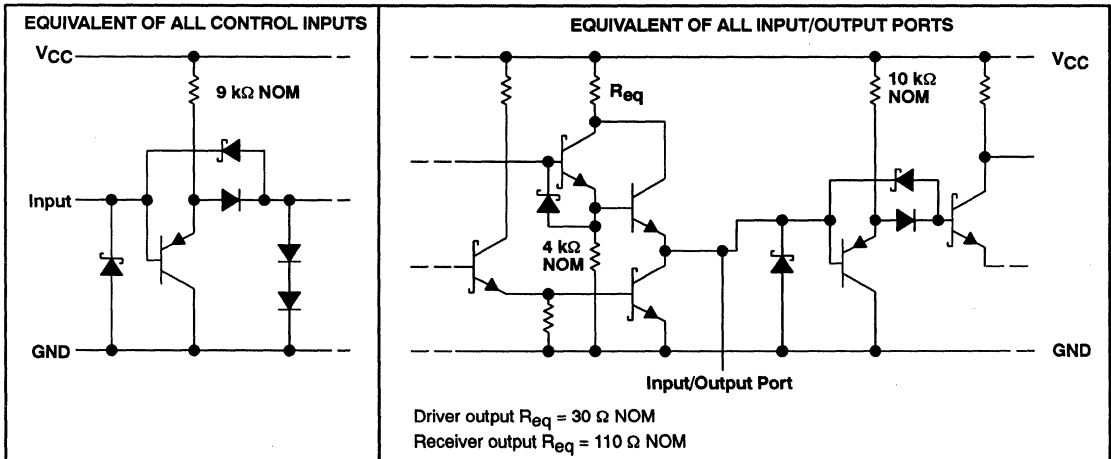
∇ Designates 3-state outputs

⊕ Designates open-collector outputs

## logic diagram (positive logic)



## schematics of inputs and outputs



# SN75ALS163

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D – D2611, JUNE 1986 – REVISED FEBRUARY 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Input voltage, $V_I$ .....	5.5 V
Low-level driver output current .....	100 mA
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	– 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds .....	260°C

NOTE: 1. All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$		2			V
Low-level input voltage, $V_{IL}$		0.8			V
High-level output current, $I_{OH}$	Bus ports with pullups active	– 5.2			mA
	Terminal ports	– 800			μA
Low-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			mA
Operating free-air temperature, $T_A$		0			70 °C



# SN75ALS163 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D - D2611, JUNE 1986 - REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA		-0.8	-1.5		V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	Bus		0.4	0.65		V
$V_{OH}$	High-level output voltage	Terminal	$I_{OH} = -800$ $\mu$ A, TE at 0.8 V	2.7	3.5		V
		Bus	$I_{OH} = -5.2$ mA, PE and TE at 2 V	2.5	3.3		
$V_{OL}$	Low-level output voltage	Terminal	$I_{OL} = 16$ mA, TE at 0.8 V	0.3	0.5		V
		Bus	$I_{OL} = 48$ mA, TE at 2 V	0.35	0.5		
$I_{OH}$	High-level output current (open-collector mode)	Bus	$V_O = 5.5$ V, PE at 0.8 V, D and TE at 2 V			100	$\mu$ A
$I_{OZ}$	Off-state output current (3-state mode)	Bus	PE at 2 V, $V_O = 2.7$ V			20	$\mu$ A
			TE at 0.8 V, $V_O = 0.5$ V			-100	
$I_I$	Input current at maximum input voltage	Terminal	$V_I = 5.5$ V	0.2	100		$\mu$ A
$I_{IH}$	High-level input current	Terminal, PE, or TE	$V_I = 2.7$ V	0.1	20		$\mu$ A
$I_{IL}$	Low-level input current	Terminal, PE, or TE	$V_I = 0.5$ V	-10	-100		$\mu$ A
$I_{OS}$	Short-circuit output current	Terminal		-15	-35	-75	mA
		Bus		-25	-50	-125	
$I_{CC}$	Supply current	No load	Terminal outputs low and enabled	42	65		mA
			Bus outputs low and enabled	52	80		
$C_{I/O}(\text{bus})$	Bus-port capacitance	$V_{CC} = 0$ to 5 V, $V_{I/O} = 0$ to 2 V, $f = 1$ MHz		30			pF

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

**switching characteristics over recommended range of operating free-air temperature (unless otherwise noted),  $V_{CC} = 5$  V**

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP‡	MAX	UNIT
$t_{PLH}$	Terminal	Bus	$C_L = 30$ pF, See Figure 1	7	20		ns
$t_{PHL}$				8	20		
$t_{PLH}$	Bus	Terminal	$C_L = 30$ pF, See Figure 2	7	14		ns
$t_{PHL}$				9	14		
$t_{PZH}$	TE	Bus	$C_L = 15$ pF, See Figure 3	19	30		ns
$t_{PHZ}$				5	12		
$t_{PZL}$				16	35		
$t_{PLZ}$				9	20		
$t_{PZH}$	TE	Terminal	$C_L = 15$ pF, See Figure 4	13	30		ns
$t_{PHZ}$				12	20		
$t_{PZL}$				12	20		
$t_{PLZ}$				11	20		
$t_{en}$	PE	Bus	$C_L = 15$ pF, See Figure 5	11	22		ns
$t_{dis}$				6	12		

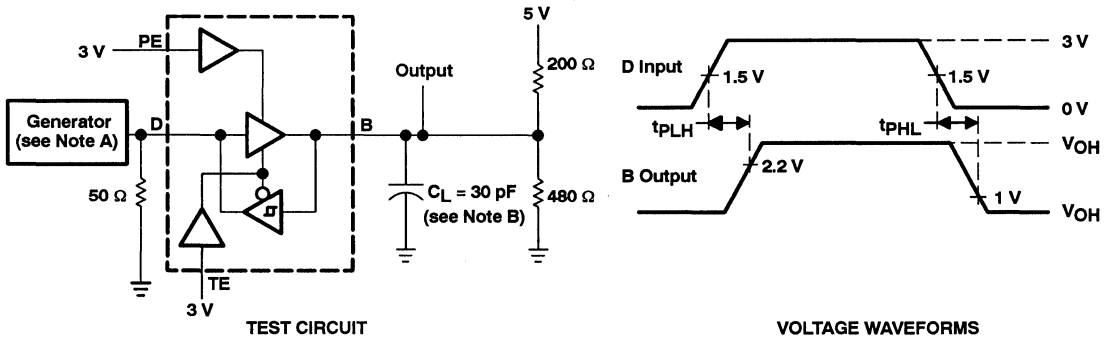
‡ Typical values are at  $T_A = 25^\circ\text{C}$ .



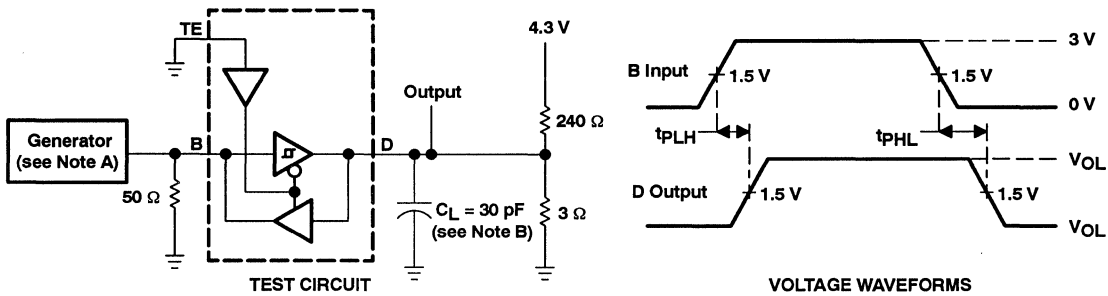
# SN75ALS163 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D - D2611, JUNE 1986 - REVISED FEBRUARY 1993

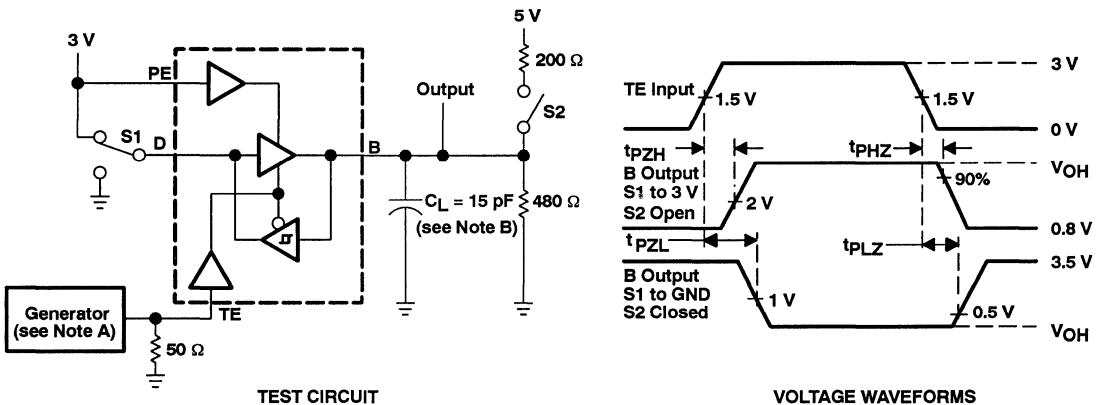
## PARAMETER MEASUREMENT INFORMATION



**Figure 1. Terminal-to-Bus Test Circuit and Voltage Waveforms**



**Figure 2. Bus-to-Terminal Test Circuit and Voltage Waveforms**



**Figure 3. TE-to-Bus Test Circuit and Voltage Waveforms**

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

# SN75ALS163 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D - D2611, JUNE 1986 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

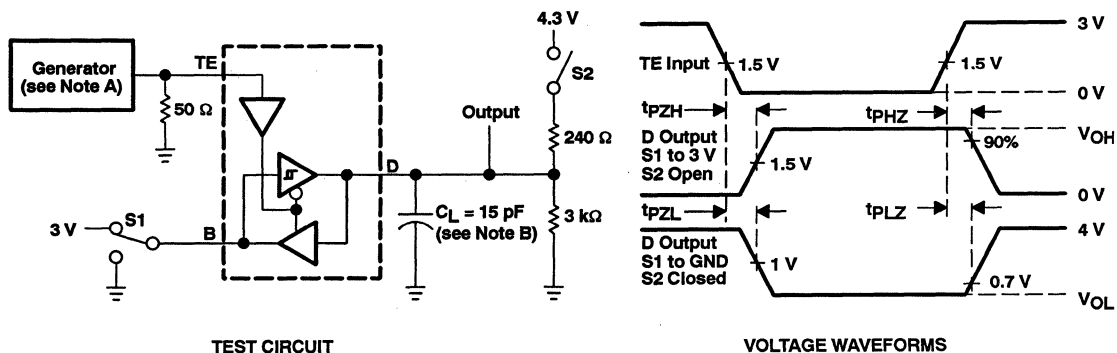


Figure 4. TE-to-Terminal Test Circuit and Voltage Waveforms

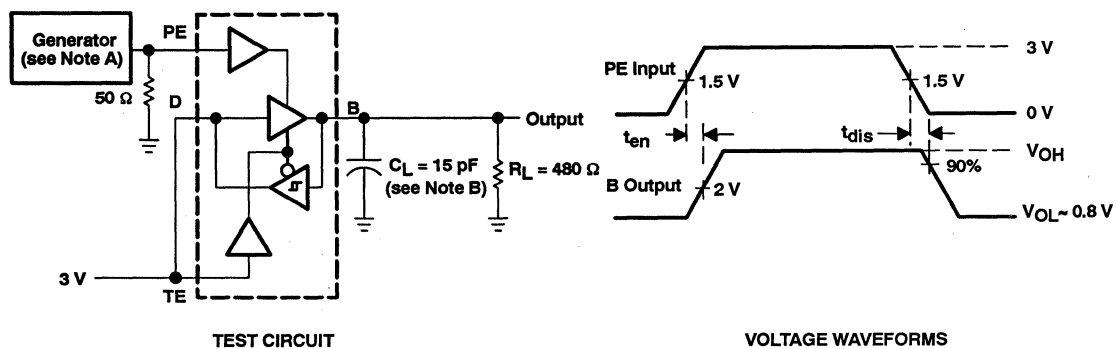


Figure 5. PE-to-Bus Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .
- B.  $C_L$  includes probe and jig capacitance.

# SN75ALS163 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D - D2611, JUNE 1986 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

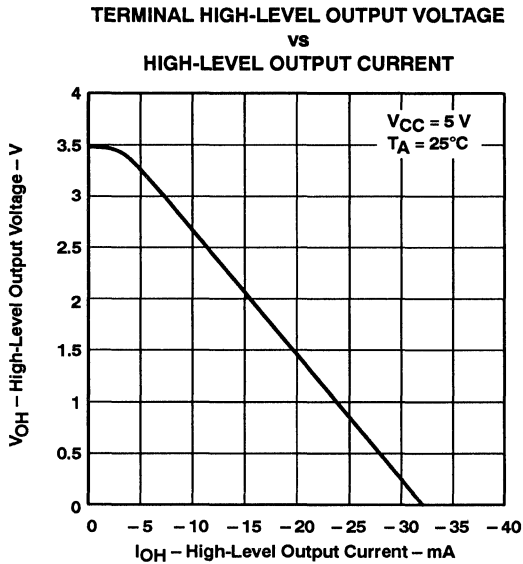


Figure 6

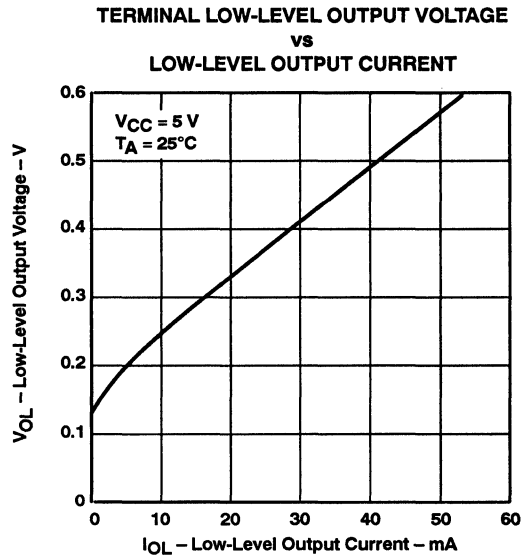


Figure 7

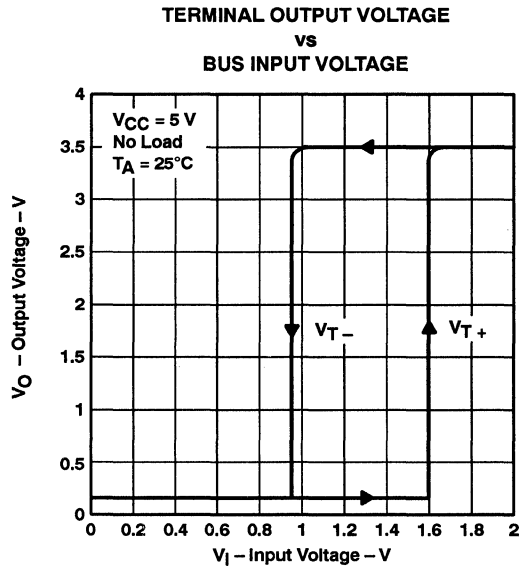


Figure 8

# SN75ALS163 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS021D - D2611, JUNE 1986 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

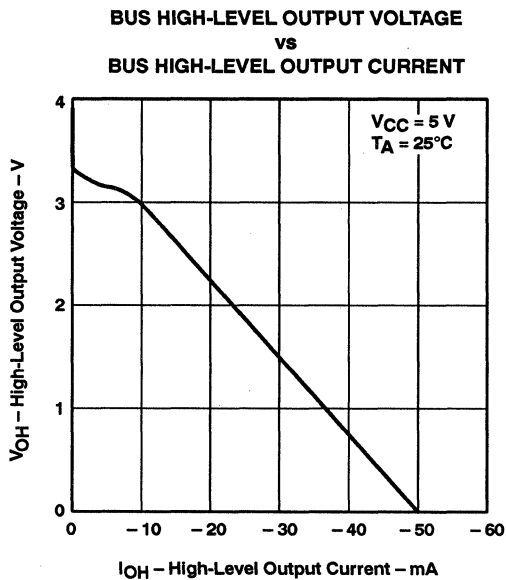


Figure 9

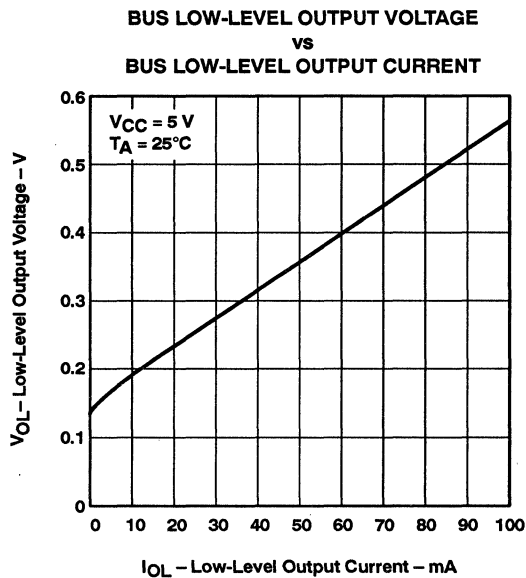


Figure 10

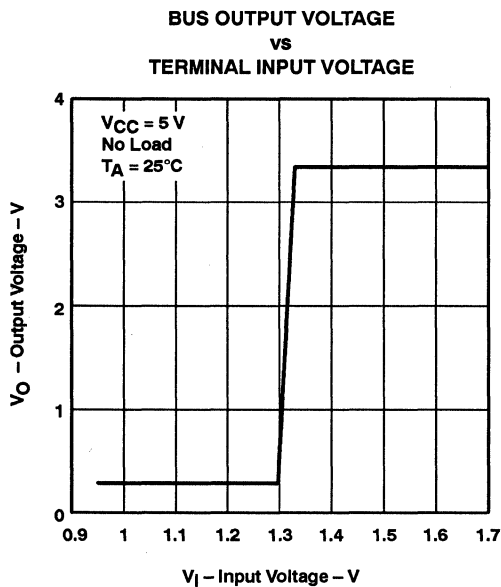
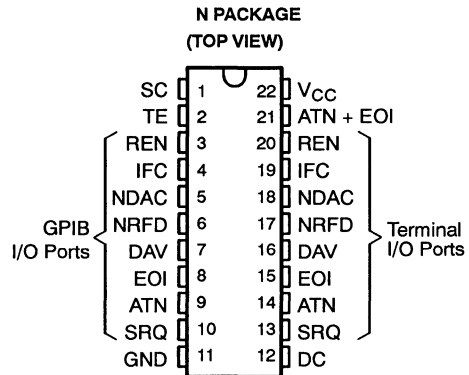


Figure 11

# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS011A – D2908, OCTOBER 1985 – REVISED FEBRUARY 1993

- 8-Channel Bidirectional Transceiver
- Power-Up/Power-Down Protection (Glitch Free)
- ATN+EOI (OR Function) Output to Simplify Board Layout
- Designed to Implement Control Bus Interface for Multiple Controllers
- Low-Power Dissipation . . . 72 mW Max Per Channel
- Fast Propagation Times . . . 22 ns Max
- High-impedance PNP Inputs
- Receiver Hysteresis . . . 650 mV Typ
- Bus-Terminating Resistors Provided on Driver Outputs
- No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )



NC – No internal connection

**NOT RECOMMENDED FOR NEW DESIGN**

## description

The SN75164B eight-channel general-purpose interface bus transceiver is a monolithic, high-speed, low-power Schottky device designed to meet the requirements of IEEE Standard 488-1978. Each transceiver is designed to provide the bus-management and data-transfer signals between operating units of a multiple-controller instrumentation system. When combined with the SN75160B octal bus transceiver, the SN75164B provides the complete 16-wire interface for the IEEE-488 bus.

The SN75164B features eight driver-receiver pairs connected in a front-to-back configuration to form input/output (I/O) ports at both the bus and terminal sides. All outputs are disabled (at a high-impedance state) during  $V_{CC}$  power-up and power-down transitions for glitch-free operation. The direction of data flow through these driver-receiver pairs is determined by the DC, TE, and SC enable signals. The SN75164B is identical to the SN75162B with the addition of an OR gate to help simplify board layouts in several popular applications. The ATN and EOI signals are ORed to pin 21, which is a standard totem-pole output.

CHANNEL IDENTIFICATION TABLE

NAME	IDENTITY	CLASS
DC TE SC	Direction Control Talk Enable System Control	Control
ATN SRQ REN IFC EOI	Attention Service Request Remote Enable Interface Clear End or Identify	Bus Management
ATN + EOI	ATN Logical or EOI	Logic
DAV NDAC NRFD	Data Valid Not Data Accepted Not Ready for Data	Data Transfer

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# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

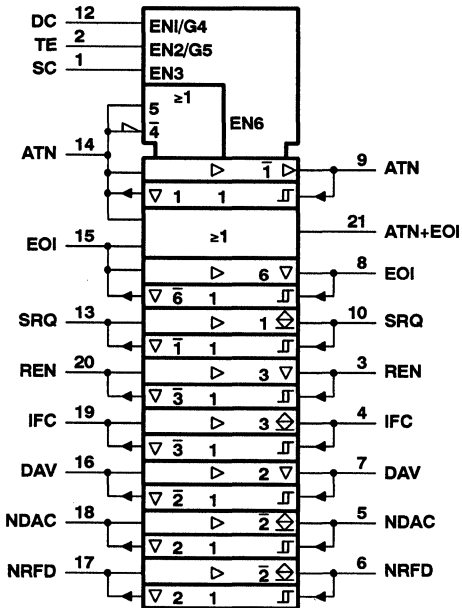
SLLS011A - D2908, OCTOBER 1985 - REVISED FEBRUARY 1993

## description (continued)

The driver outputs (GPIO I/O ports) feature active bus-terminating resistor circuits designed to provide a high impedance to the bus when supply voltage  $V_{CC}$  is 0. The drivers are designed to handle loads up to 48 mA of sink current. Each receiver features pnp transistor inputs for high input impedance and an ensured hysteresis of 400 mV for increased noise immunity. All receivers have 3-state outputs to present a high impedance to the terminal when disabled.

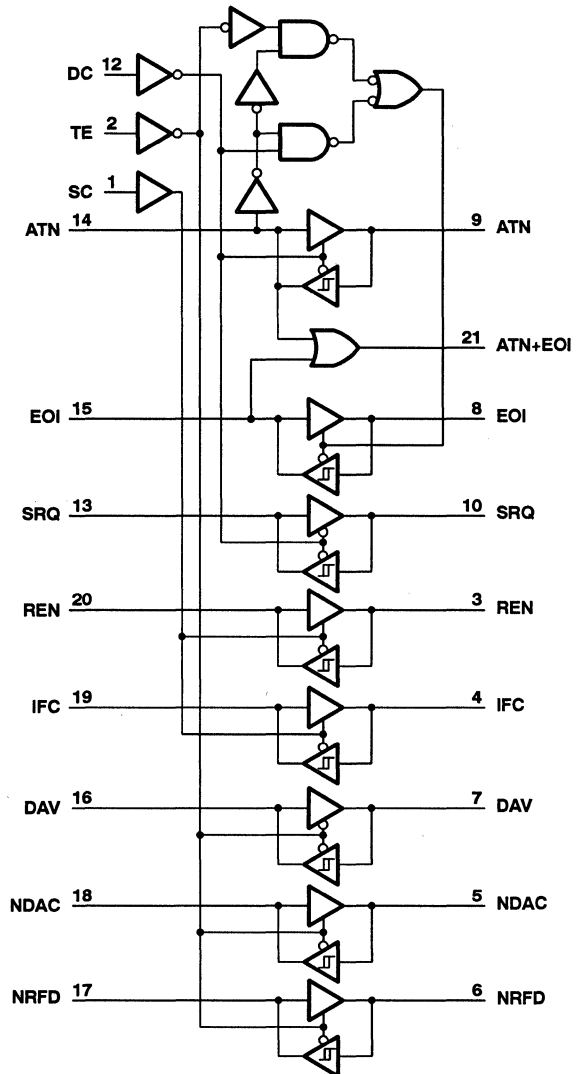
The SN75164B is characterized for operation from 0°C to 70°C.

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12

## logic diagram (positive logic)



# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS011A – D2908, OCTOBER 1985 – REVISED FEBRUARY 1993

**RECEIVE/TRANSMIT FUNCTION TABLE**

CONTROLS				BUS-MANAGEMENT CHANNELS				DATA-TRANSFER CHANNELS			
SC	DC	TE	ATN†	ATN†	SRQ	REN	IFC	EOI	DAV	NDAC	NRFD
				(controlled by DC)		(controlled by SC)			(controlled by TE)		
	H	H	H	R	T			T	T	R	R
	H	H	L	T	R			R	R	T	T
	L	L	H					R	T	T	
	L	L	L	R	T			R	T	T	
	H	L	X	T	R			T	R	T	T
	L	H	X	T	R			T	T	R	R
H						T	T				
L						R	R				

H = high level, L = low level, R = receive, T = transmit, X = irrelevant

Direction of data transmission is from the terminal side to the bus side, and the direction of data receiving is from the bus side to the terminal side. Data transfer is noninverting in both directions.

† ATN is a normal transceiver channel that functions additionally as an internal direction control or talk enable for EOI whenever the DC and TE inputs are in the same state. When DC and TE are in opposite states, the ATN channel functions as an independent transceiver only.

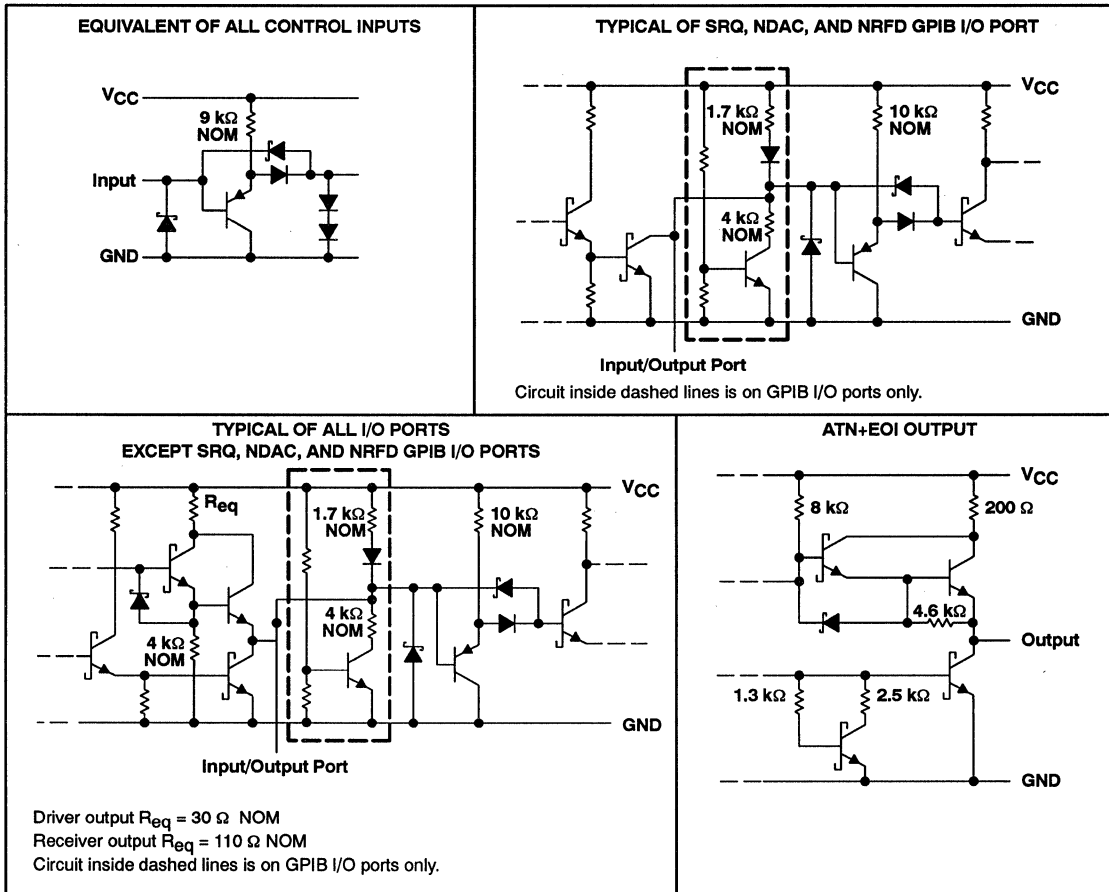
**ATN + EOI FUNCTION TABLE**

INPUTS		OUTPUT ATN+EOI
ATN	EOI	
H	X	H
X	H	H
L	L	L

# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Low-level driver output current	100 mA
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 2)	1700 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: DW or N package	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

2. For operation above 25°C free-air temperature, derate the N package at the rate of 13.6 mW/°C.

TEXAS  
INSTRUMENTS

# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level Input voltage, $V_{IH}$		2			V
Low-level Input voltage, $V_{IL}$				0.8	V
High-level output current, $I_{OH}$	Bus ports with 3-state outputs			-5.2	mA
	Terminal ports			-800	μA
	ATN+EOI			-400	
Low-level output current, $I_{OL}$	Bus ports			48	mA
	Terminal ports			16	
	ATN+EOI			4	
Operating free-air temperature, $T_A$		0		70	°C

## electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT	
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA				-1.5	V	
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	Bus	See Figure 8	0.4			V	
$V_{OH}‡$	High-level output voltage	Terminal	$I_{OH} = -800$ μA			2.7	V	
		Bus	$I_{OH} = -5.2$ mA			2.5		
		ATN+EOI	$I_{OH} = -400$ μA			2.7		
$V_{OL}$	Low-level output voltage	Terminal	$I_{OL} = 16$ mA			0.5	V	
		Bus	$I_{OL} = 48$ mA			0.5		
		ATN+EOI	$I_{OL} = 4$ mA			0.4		
$I_I$	Input current at maximum input voltage	Terminal§	$V_I = 5.5$ V			100	μA	
		ATN+EOI	$V_I = 5.5$ V			200		
$I_{IH}$	High-level input current	Terminal, control	$V_I = 2.7$ V			20	μA	
		ATN, EOI	$V_I = 2.7$ V			40		
$I_{IL}$	Low-level input current	Terminal, control	$V_I = 0.5$ V			-100	μA	
		ATN, EOI	$V_I = 0.5$ V			-500		
$V_{I/O}(\text{bus})$	Voltage at bus port	Driver disabled	$I_{I(\text{bus})} = 0$			2.5	3.7	V
			$I_{I(\text{bus})} = -12$ mA				-1.5	
$I_{I/O}(\text{bus})$	Current into bus port	Power on	Driver disabled	$V_{I(\text{bus})} = -1.5$ V to 0.4 V		-1.3		mA
				$V_{I(\text{bus})} = 0.4$ V to 2.5 V		0	-3.2	
				$V_{I(\text{bus})} = 2.5$ V to 3.7 V			+2.5	
				$V_{I(\text{bus})} = 3.7$ V to 5 V		0	2.5	
				$V_{I(\text{bus})} = 5$ V to 5.5 V		0.7	2.5	
		Power off	$V_{CC} = 0$ , $V_{I(\text{bus})} = 0$ V to 2.5 V			-40	μA	
$I_{OS}$	Short-circuit output current	Terminal				-15	-75	mA
		Bus				-25	-125	
		ATN+EOI				-10	-100	
$I_{CC}$	Supply current	No load, TE, DE, and SC low					120	mA
$C_{I/O}(\text{bus})$	Bus-port capacitance	$V_{CC} = 5$ V to 0 V, $V_{I/O} = 0$ to 2 V, $f = 1$ MHz				30		pF

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

‡  $V_{OH}$  applies for 3-state outputs only.

§ Except ATN and EOI terminal pins



# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 15\text{ pF}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

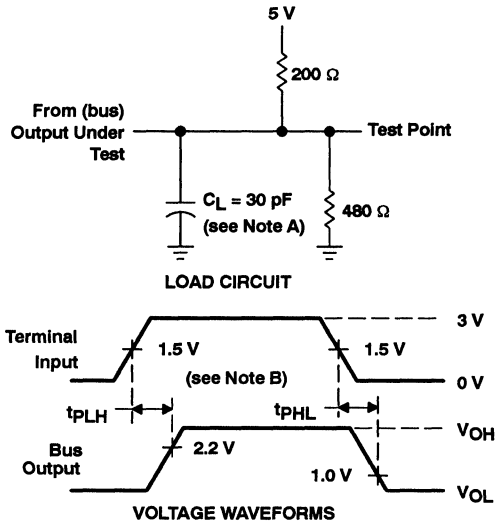
PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high level output	Terminal	Bus	$C_L = 30\text{ pF}$ , See Figure 1		14	20	ns
$t_{PHL}$ Propagation delay time, high-to-low level output					14	20	
$t_{PLH}$ Propagation delay time, low-to-high level output	Terminal	Bus (SRQ, NDAC, NRFD)	$C_L = 30\text{ pF}$ , See Figure 1		29	35	ns
$t_{PLH}$ Propagation delay time, low-to-high level output	Bus	Terminal	$C_L = 30\text{ pF}$ , See Figure 2		10	20	ns
$t_{PHL}$ Propagation delay time, high-to-low level output					15	22	
$t_{PLH}$ Propagation delay time, low-to-high level output	Terminal ATN or Terminal EOI	ATN+EOI	See Figure 3		14		ns
$t_{PHL}$ Propagation delay time, high-to-low level output	Terminal ATN or Terminal EOI	ATN+EOI	See Figure 3		14		ns
$t_{PZH}$ Output enable time to high level	TE, DC, or SC	Bus (ATN, EOI, REN, IFC, and DAV)	See Figure 4			60	ns
$t_{PHZ}$ Output disable time from high level					45		
$t_{PZL}$ Output enable time to low level					60		
$t_{PLZ}$ Output disable time from low level					55		
$t_{PZH}$ Output enable time to high level	TE, DC, or SC	Terminal	See Figure 5			55	ns
$t_{PHZ}$ Output disable time from high level					50		
$t_{PZL}$ Output enable time to low level					45		
$t_{PLZ}$ Output disable time from low level					55		



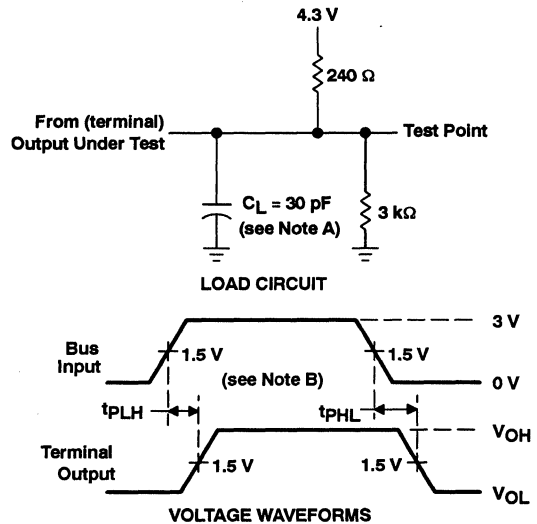
# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS011A - D2908, OCTOBER 1985 - REVISED FEBRUARY 1993

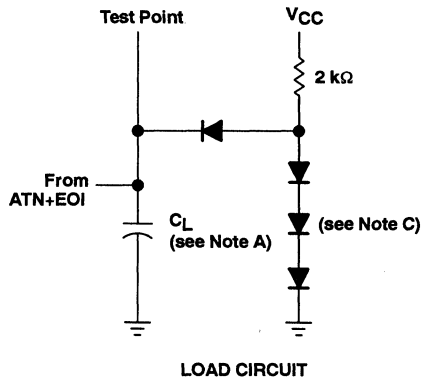
## PARAMETER MEASUREMENT INFORMATION



**Figure 1. Terminal-to-Bus Load Circuit and Voltage Waveforms**



**Figure 2. Bus-to-Terminal Load Circuit and Voltage Waveforms**



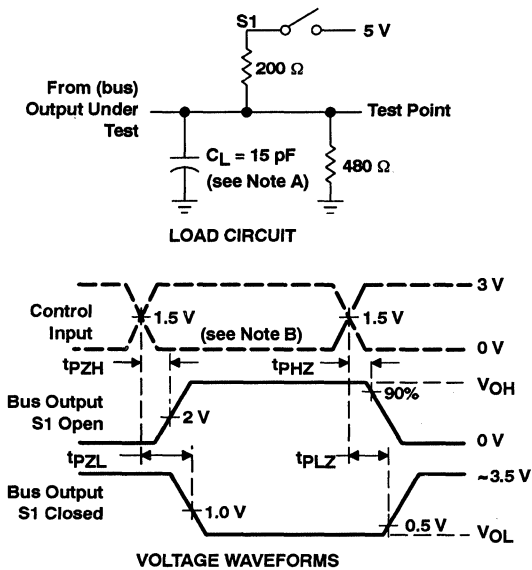
**Figure 3. ATN+EOI Load Circuit and Voltage Waveforms**

- NOTES: A.  $C_L$  includes probe and jig capacitance.  
 B. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  ns,  $Z_0 = 50 \Omega$ .  
 C. All diodes are 1N916 or 1N3064.

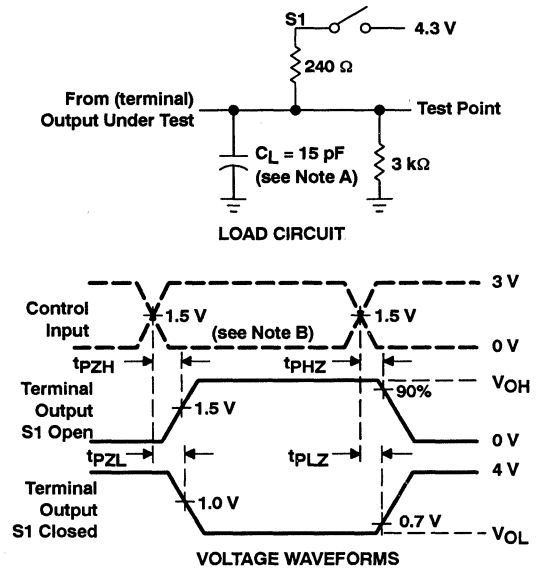
# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS011A – D2908, OCTOBER 1985 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



**Figure 4. Bus Enable and Disable Times  
Load Circuit and Voltage Waveforms**



**Figure 5. Terminal Enable and Disable Times  
Load Circuit and Voltage Waveforms**

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .

# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS011A - D2908, OCTOBER 1985 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

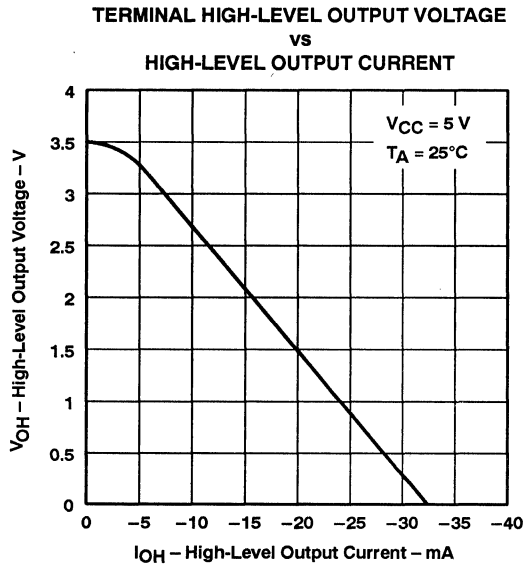


Figure 6

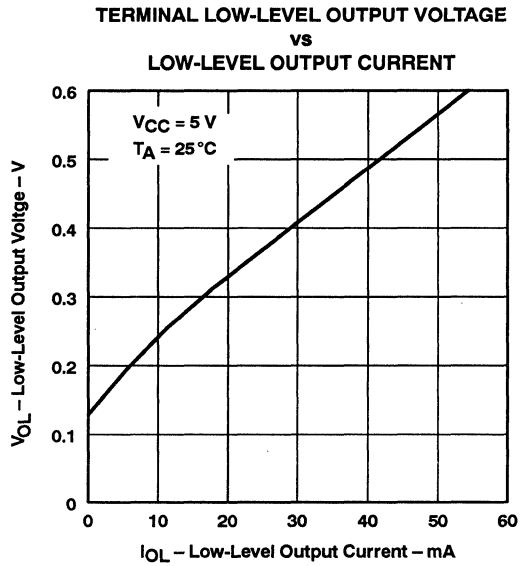


Figure 7

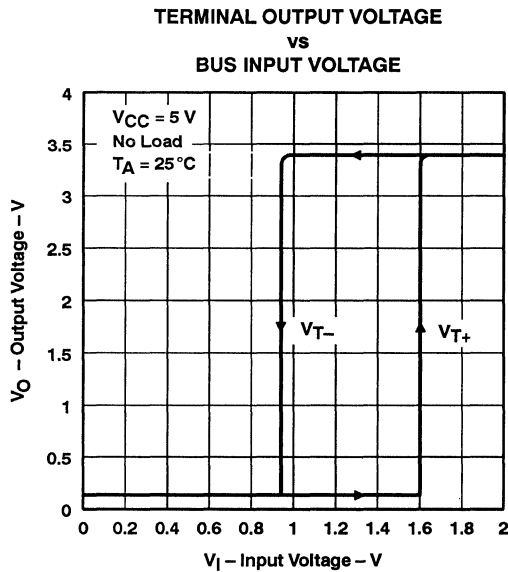


Figure 8



# SN75164B OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS011A - D2908, OCTOBER 1985 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

**BUS HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

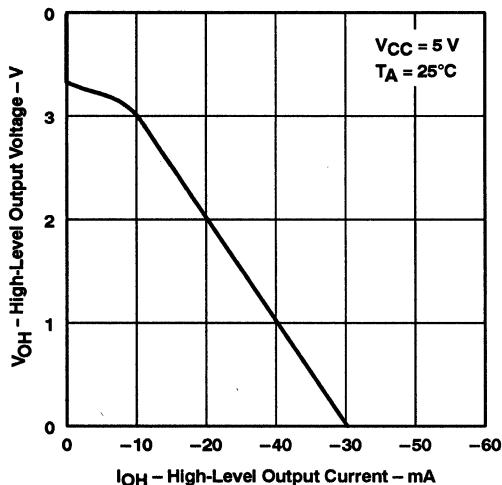


Figure 9

**BUS LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

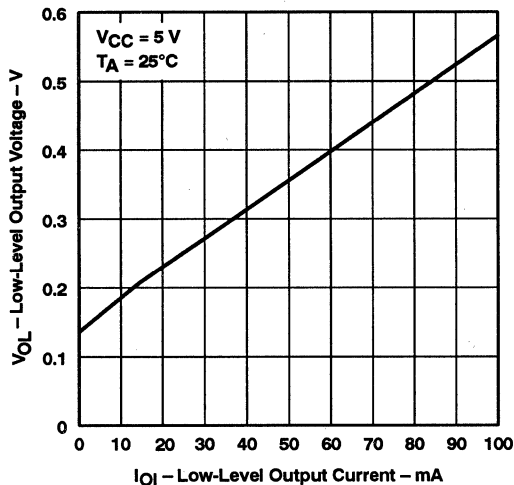


Figure 10

**BUS OUTPUT VOLTAGE  
vs  
THERMAL INPUT VOLTAGE**

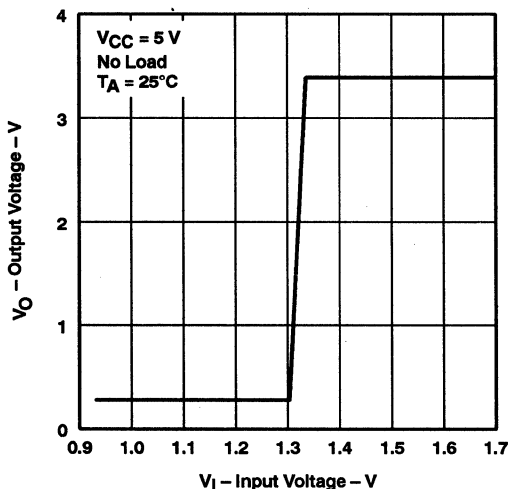


Figure 11

**BUS CURRENT  
vs  
BUS VOLTAGE**

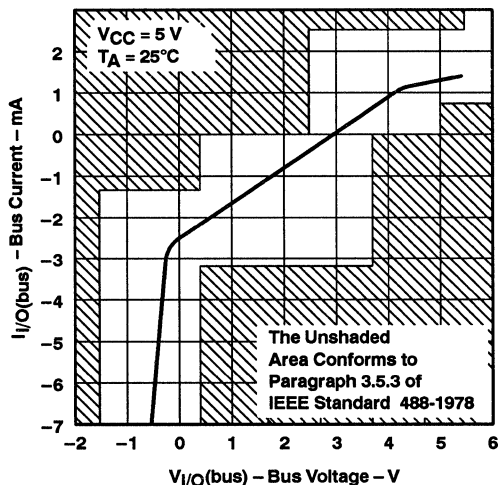


Figure 12

# SN75ALS164

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

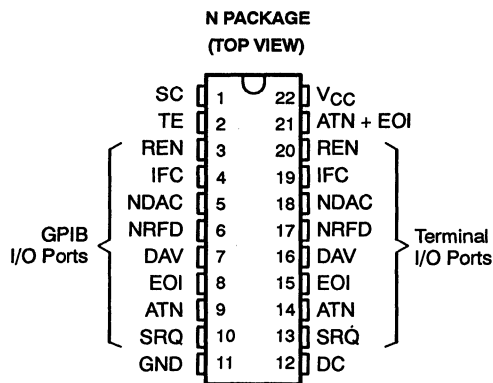
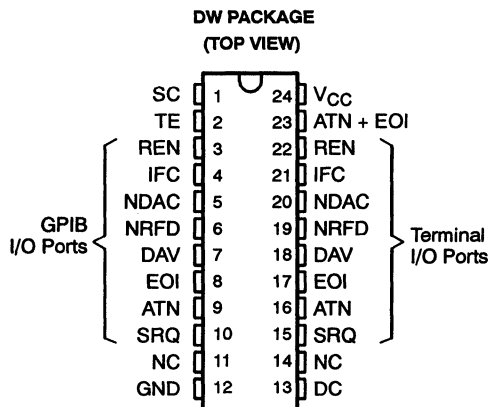
SLLS022B - D2908, JUNE 1986 - REVISED AUGUST 1989

- **8-Channel Bidirectional Transceiver**
- **Designed to Implement Control Bus Interface**
- **Designed for Multicontrollers**
- **High-Speed Advanced Low-Power Schottky Circuitry**
- **Low-Power Dissipation . . . 46 mW Max Per Channel**
- **Fast Propagation Times . . . 20 ns Max**
- **High-Impedance PNP Inputs**
- **Receiver Hysteresis . . . 650 mV Typ**
- **Bus-Terminating Resistors Provided on Driver Outputs**
- **No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )**
- **Power-Up/Power-Down Protection (Glitch Free)**

### description

The SN75ALS164 eight-channel general-purpose interface bus transceiver is a monolithic, high-speed, advanced low-power Schottky device designed to meet the requirements of IEEE Standard 488-1978. Each transceiver is designed to provide the bus-management and data-transfer signals between operating units of a multiple-controller instrumentation system. When combined with the SN75ALS160 octal bus transceiver, the SN75ALS164 provides the complete 16-wire interface for the IEEE 488 bus.

The SN75ALS164 features eight driver-receiver pairs connected in a front-to-back configuration to form input/output (I/O) ports at both the bus and terminal sides. All outputs are disabled (at a high-impedance state) during  $V_{CC}$  power-up and power-down transitions for glitch-free operation. The direction of data flow through these driver-receiver pairs is determined by the DC, TE, and SC enable signals. The SN75ALS164 is identical to the SN75ALS162 with the addition of an OR gate to help simplify board layouts in several popular applications. The ATN and EOI signals are ORed to provide the ATN + EOI output, which is a standard totem-pole output.



NC - No internal connection

**NOT RECOMMENDED FOR NEW DESIGN**

**CHANNEL IDENTIFICATION TABLE**

NAME	IDENTITY	CLASS
DC	Direction Control	Control
TE	Talk Enable	
SC	System Control	
ATN	Attention	Bus Management
SRQ	Service Request	
REN	Remote Enable	
IFC	Interface Clear	
EOI	End or Identify	
ATN+EOI	ATN Logical or EOI	Logic
DAV	Data Valid	Data Transfer
NDAC	No Data Accepted	
NRFD	Not Ready for Data	

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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# SN75ALS164 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

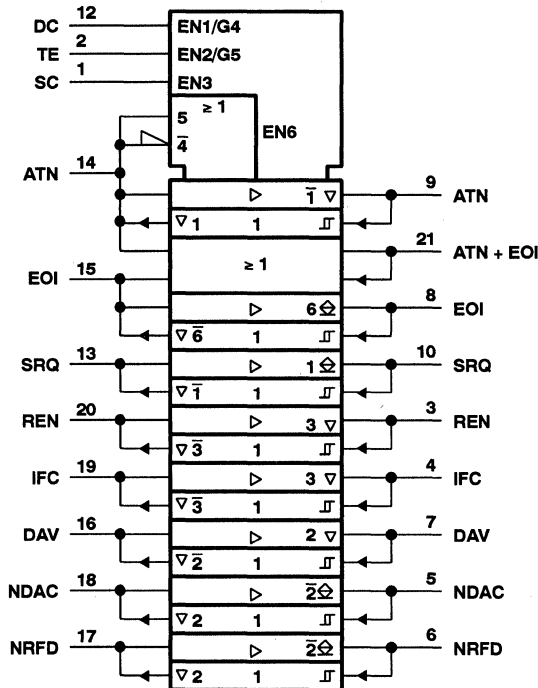
SLLS022B - D2908, JUNE 1986 - REVISED AUGUST 1989

## description (continued)

The driver outputs (GPIO I/O ports) feature active bus-terminating resistor circuits designed to provide a high impedance to the bus when supply voltage  $V_{CC}$  is 0. The drivers are designed to handle loads up to 48 mA of sink current. Each receiver features pnp transistor inputs for high input impedance and hysteresis of 400 mV minimum for increased noise immunity. All receivers have 3-state outputs to present a high impedance to the terminal when disabled.

The SN75ALS164 is characterized for operation from 0°C to 70°C.

## logic symbol†

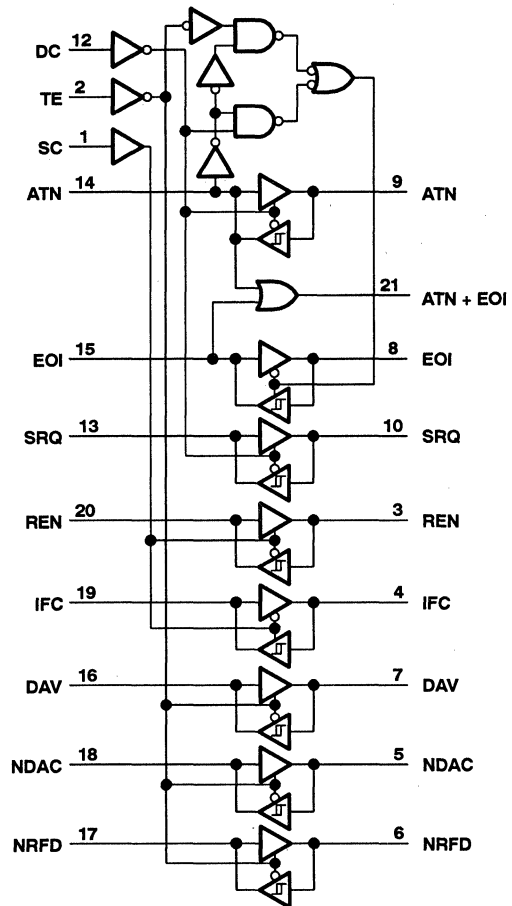


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

▽ Designates 3-state outputs

⊕ Designates passive-pullup outputs

## logic diagram (positive logic)



Pin numbers shown are for the N package.

# SN75ALS164 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS022B - D2908, JUNE 1986 - REVISED AUGUST 1989

**RECEIVE/TRANSMIT FUNCTION TABLE**

CONTROLS				BUS-MANAGEMENT CHANNELS				DATA-TRANSFER CHANNELS			
SC	DC	TE	ATN†	ATN† (controlled by DC)	SRQ	REN (controlled by SC)	IFC	EOI	DAV	NDAC	NRFD
									(controlled by TE)		
	H	H	H	R	T			T	T	R	R
	H	H	L					R			
	L	L	H	T	R			R	R	T	T
	L	L	L					T			
	H	L	X	R	T			R	R	T	T
	L	H	X	T	R			T	T	R	R
H						T	T				
L						R	R				

H = high level, L = low level, R = receive, T = transmit, X = irrelevant

Direction of data transmission is from the terminal side to the bus side, and the direction of data receiving is from the bus side to the terminal side. Data transfer is noninverting in both directions.

† ATN is a normal transceiver channel that functions additionally as an internal direction control or talk enable for EOI whenever the DC and TE inputs are in the same state. When DC and TE are in opposite states, the ATN channel functions as an independent transceiver only.

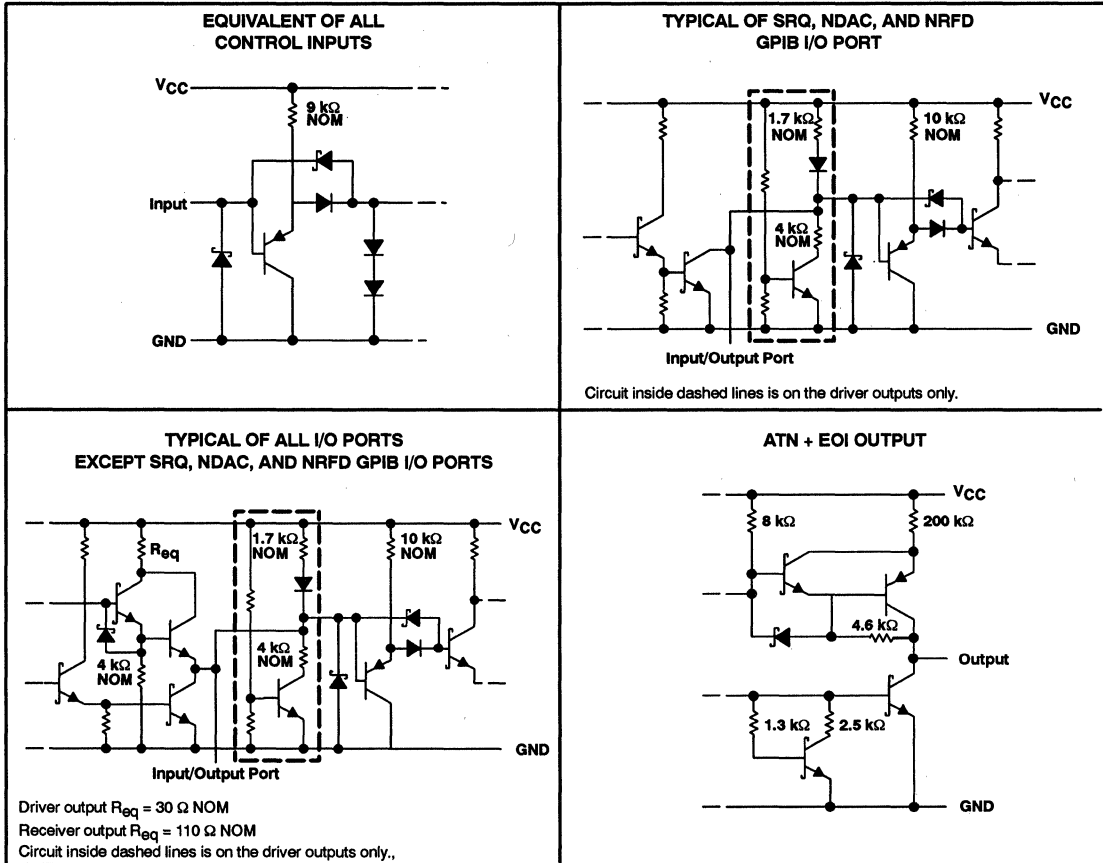
**ATN + EOI FUNCTION TABLE**

INPUTS		OUTPUT ATN + EOI
ATN	EOI	
H	X	H
X	H	H
L	L	L

# SN75ALS164 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS022B - D2908, JUNE 1986 - REVISED AUGUST 1989

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Low-level driver output current	100 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1350 mW	10.8 mW/°C	864 mW
N	1700 mW	13.6 mW/°C	1088 mW

# SN75ALS164

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS022B – D2908, JUNE 1986 – REVISED AUGUST 1989

### recommended operating conditions

		MIN	NOM	MAX	UNIT	
Supply voltage, $V_{CC}$		4.75	5	5.25	V	
High-level input voltage, $V_{IH}$		2			V	
Low-level input voltage, $V_{IL}$		0.8			V	
High-level output current, $I_{OH}$	Bus ports with 3-state outputs	-5.2			mA	
	Terminal ports	-800			$\mu$ A	
	ATN + EOI	-400				
Low-level output current, $I_{OL}$	Bus ports	48			mA	
	Terminal ports	16				
	ATN + EOI	4				
Operating free-air temperature, $T_A$		0			70	$^{\circ}$ C

### electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA		-0.8	-1.5		V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	Bus		0.4	0.65		V
$V_{OH}^{\ddagger}$	High-level output voltage	Terminal	$I_{OH} = -800$ $\mu$ A	2.7	3.5		V
		Bus	$I_{OH} = -5.2$ mA	2.5	3.3		
		ATN+EOI	$I_{OH} = -400$ $\mu$ A	2.7			
$V_{OL}$	Low-level output voltage	Terminal	$I_{OL} = 16$ mA		0.3	0.5	V
		Bus	$I_{OL} = 48$ mA		0.35	0.5	
		ATN+EOI	$I_{OL} = 4$ mA			0.4	
$I_I$	Input current at maximum input voltage	Terminal <sup>§</sup>	$V_I = 5.5$ V		0.2	100	$\mu$ A
		ATN, EOI	$V_I = 5.5$ V			200	
$I_{IH}$	High-level input current	Terminal control	$V_I = 2.7$ V		0.1	20	$\mu$ A
		ATN, EOI	$V_I = 2.7$ V			40	
$I_{IL}$	Low-level input current	Terminal control	$V_I = 0.5$ V		-10	-100	$\mu$ A
		ATN, EOI	$V_I = 0.5$ V			-500	
$V_{I/O}(\text{bus})$	Voltage at bus port	Driver disabled	$I_I(\text{bus}) = 0$	2.5	3.0	3.7	V
			$I_I(\text{bus}) = -12$ mA			-1.5	
$I_{I/O}(\text{bus})$	Current into bus port	Power on	Driver disabled	$V_I(\text{bus}) = -1.5$ V to 0.4 V	-1.3		mA
			$V_I(\text{bus}) = 0.4$ V to 2.5 V	0		-3.2	
			$V_I(\text{bus}) = 2.5$ V to 3.7 V			+2.5	
		$V_I(\text{bus}) = 3.7$ V to 5 V	0		2.5		
		$V_I(\text{bus}) = 5$ V to 5.5 V	0.7		2.5		
		Power off	$V_{CC} = 0$ , $V_I(\text{bus}) = 0$ to 2.5 V			-40	
$I_{OS}$	Short-circuit output current	Terminal		-15	-35	-75	mA
		Bus		-25	-50	-125	
		ATN + EOI		-10		-100	
$I_{CC}$	Supply current	No load, TE, DC, and SC low		55	75		mA
$C_{I/O}(\text{bus})$	Bus-port capacitance	$V_{CC} = 0$ to 5 V, $V_{I/O} = 0$ to 2 V, $f = 1$ MHz		30			pF

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^{\circ}$ C.

‡  $V_{OH}$  applies for 3-state outputs only.

§ Except ATN and EOI terminals.



# SN75ALS164 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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switching characteristics over recommended range of operating free-air temperature,  $V_{CC} = 5\text{ V}$

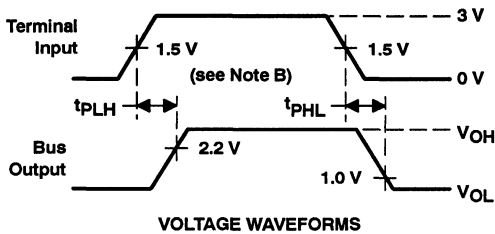
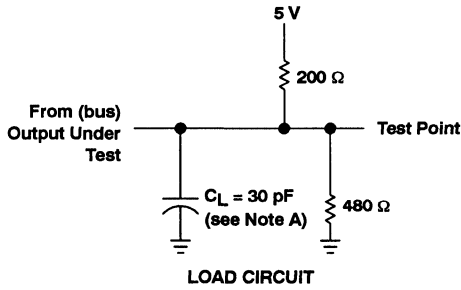
PARAMETER		FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Terminal	Bus	C <sub>L</sub> = 30 pF, See Figure 1		10	20	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					12	20	
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Bus	Terminal	C <sub>L</sub> = 30 pF, See Figure 2		5	10	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output					7	14	
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Terminal ATN or Terminal EOI	ATN+EOI	C <sub>L</sub> = 15 pF, See Figure 3		3.5	10	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output	Terminal ATN or Terminal EOI	ATN+EOI	C <sub>L</sub> = 15 pF, See Figure 3		7	15	ns
t <sub>PZH</sub>	Output enable time to high level	TE, DC, or SC	Bus (ATN, EOI, REN, IFC, and DAV)	C <sub>L</sub> = 15 pF, See Figure 4			30	ns
t <sub>PHZ</sub>	Output disable time from high level					20		
t <sub>PZL</sub>	Output enable time to low level					45		
t <sub>PLZ</sub>	Output disable time from low level					20		
t <sub>PZH</sub>	Output enable time to high level	TE, DC, or SC	Terminal	C <sub>L</sub> = 15 pF, See Figure 5			30	ns
t <sub>PHZ</sub>	Output disable time from high level					25		
t <sub>PZL</sub>	Output enable time to low level					30		
t <sub>PLZ</sub>	Output disable time from low level					25		



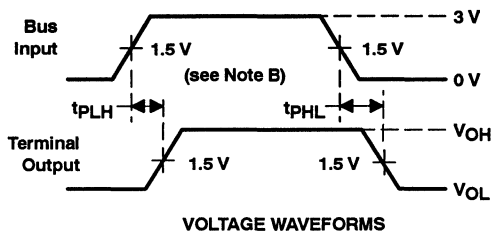
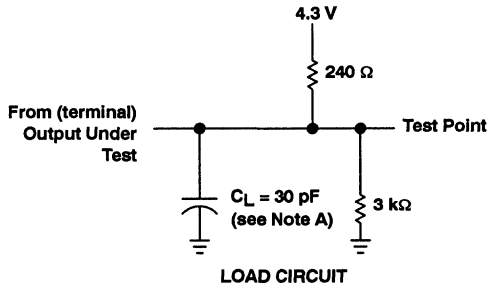
# SN75ALS164 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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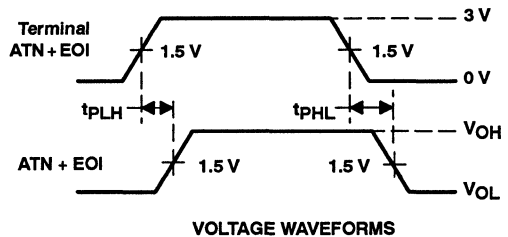
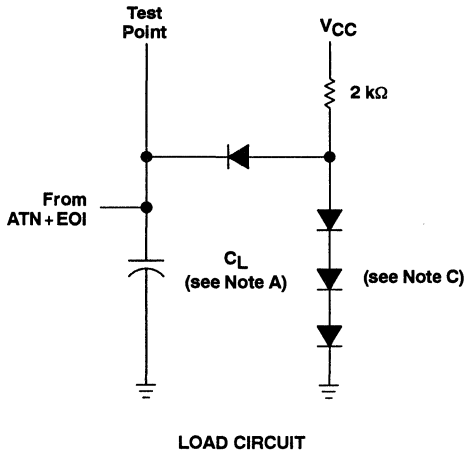
## PARAMETER MEASUREMENT INFORMATION



**Figure 1. Terminal-to-Bus Load Circuit and Voltage Waveforms**



**Figure 2. Bus-to-Terminal Load Circuit and Voltage Waveforms**



**Figure 3. ATN + EOI Load Circuit and Voltage Waveforms**

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

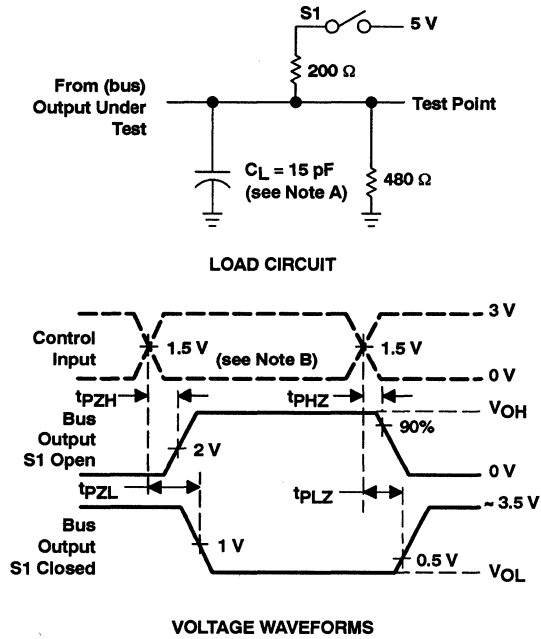
C. All diodes are 1N916 or 1N3064



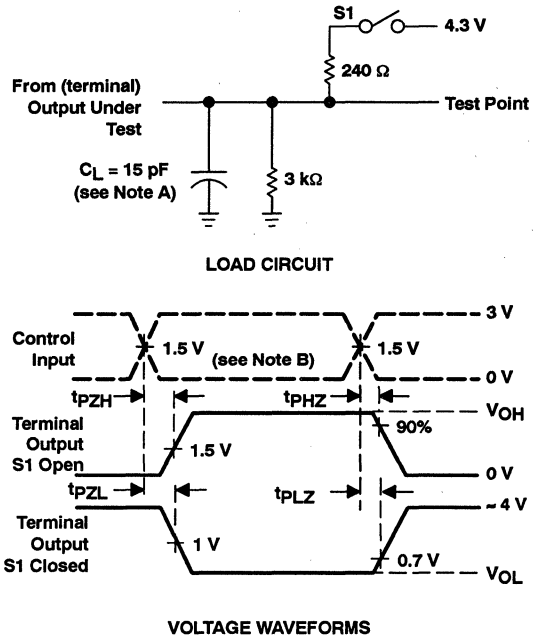
# SN75ALS164 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## PARAMETER MEASUREMENT INFORMATION



**Figure 4. Bus Load Circuit and Voltage Waveforms**



**Figure 5. Terminal Load Circuit and Voltage Waveforms**

NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .

# SN75ALS164

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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### TYPICAL CHARACTERISTICS

**TERMINAL HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

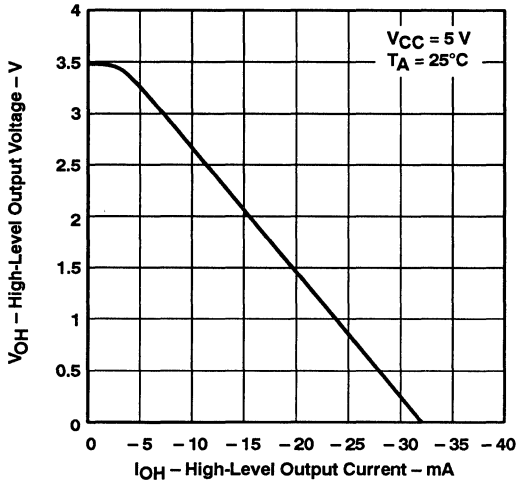


Figure 6

**TERMINAL LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

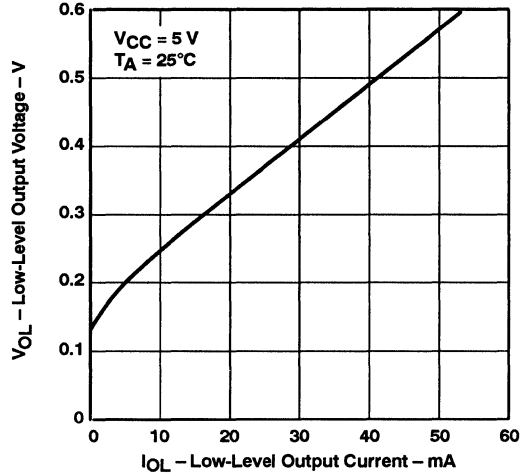


Figure 7

**TERMINAL OUTPUT VOLTAGE  
vs  
BUS INPUT VOLTAGE**

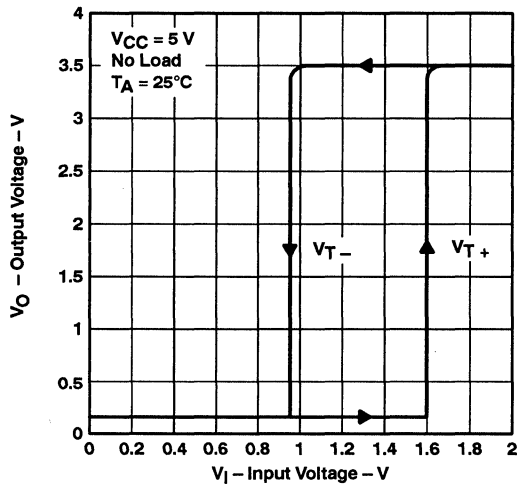


Figure 8

# SN75ALS164 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## TYPICAL CHARACTERISTICS

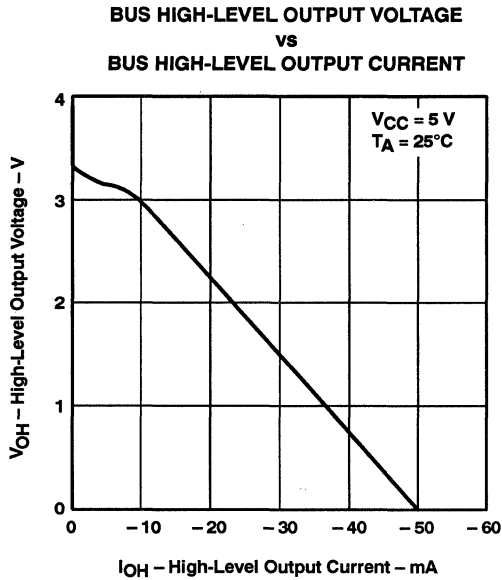


Figure 9

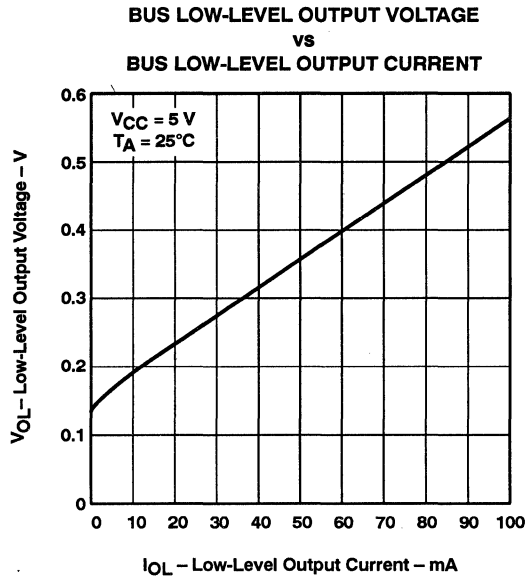


Figure 10

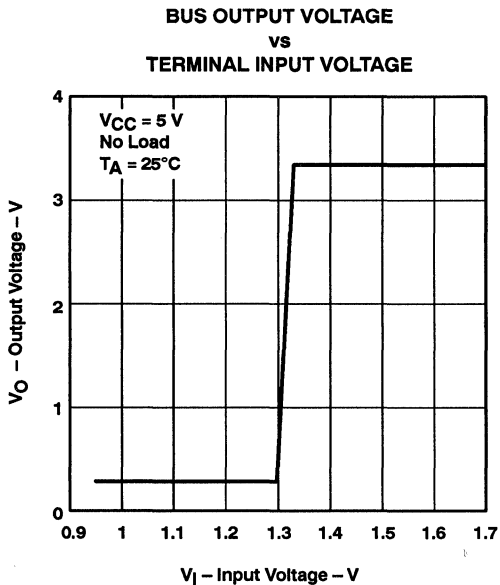


Figure 11

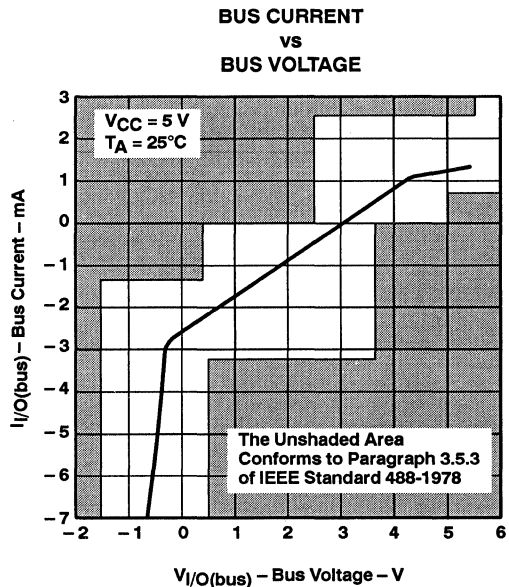


Figure 12

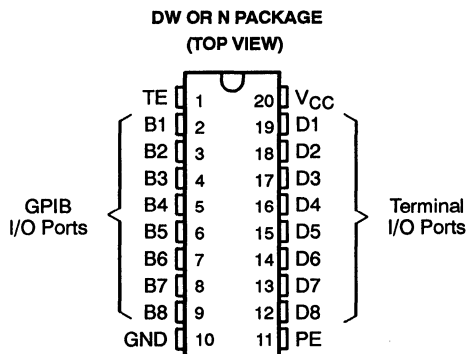
# SN75ALS165

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS023B - D3011, JUNE 1986 - REVISED AUGUST 1989

### MEETS IEEE STANDARD 488-1978 (GPIB)

- **8-Channel Bidirectional Transceiver**
- **High-Speed Advanced Low-Power Schottky Circuitry**
- **Low Power Dissipation . . . 46 mW Max Per Channel**
- **Fast Propagation Times . . . 20 ns Max**
- **High-Impedance PNP Inputs**
- **Receiver Hysteresis . . . 650 mV Typ**
- **No Loading of Bus When Device Is Powered Down ( $V_{CC} = 0$ )**
- **Power-Up/Power-Down Protection (Glitch Free)**
- **Driver and Receiver Can Be Disabled Simultaneously**



**NOT RECOMMENDED FOR NEW DESIGN**

#### Function Tables

EACH DRIVER				EACH RECEIVER			
INPUTS			OUTPUT	INPUTS			OUTPUT
D	TE	PE	B	B	TE	PE	D
H	H	H	H	L	L	H	L
L	H	X	L	H	L	H	H
H	X	L	Z†	X	H	X	Z
X	L	X	Z†	X	X	L	Z

H = high level, L = low level, X = irrelevant, Z = high-impedance state

† This is the high-impedance state of a normal 3-state output modified by the internal resistors to  $V_{CC}$  and GND.

#### description

The SN75ALS165 eight-channel general-purpose interface bus transceiver is a monolithic, high-speed, advanced low-power Schottky device designed for two-way data communications over single-ended transmission lines. It is designed to meet the requirements of IEEE Standard 488-1978. The transceiver features driver outputs that can be operated in either the passive-pullup or 3-state mode. If talk enable (TE) is high, these ports have the characteristics of passive-pullup outputs when pullup enable (PE) is low and of 3-state outputs when PE is high. Taking TE low places these ports in the high-impedance state. Taking TE and PE low places both the drivers and receivers in the high-impedance state. The driver outputs are designed to handle loads up to 48 mA of sink current.

An active turn-off feature is incorporated into the bus-terminating resistors so that the device exhibits a high impedance to the bus when  $V_{CC} = 0$ . When combined with the SN75ALS161 or SN75ALS162 management bus transceiver, the pair provides the complete 16-wire interface for the IEEE 488 bus.

The SN75ALS165 is manufactured in a 20-pin package and is characterized for operation from 0°C to 70°C.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



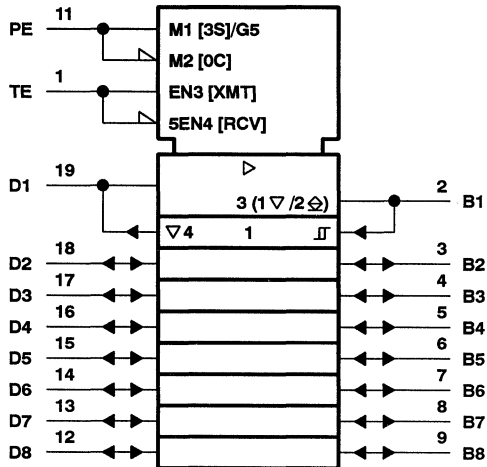
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# SN75ALS165 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## logic symbol†

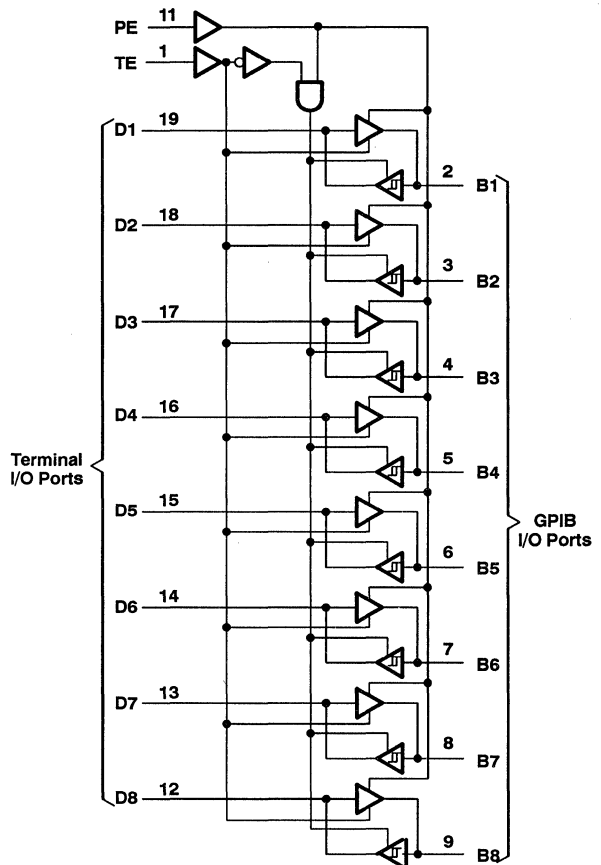


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

▽ Designates 3-state outputs

⊞ Designates passive-pullup outputs

## logic diagram (positive logic)



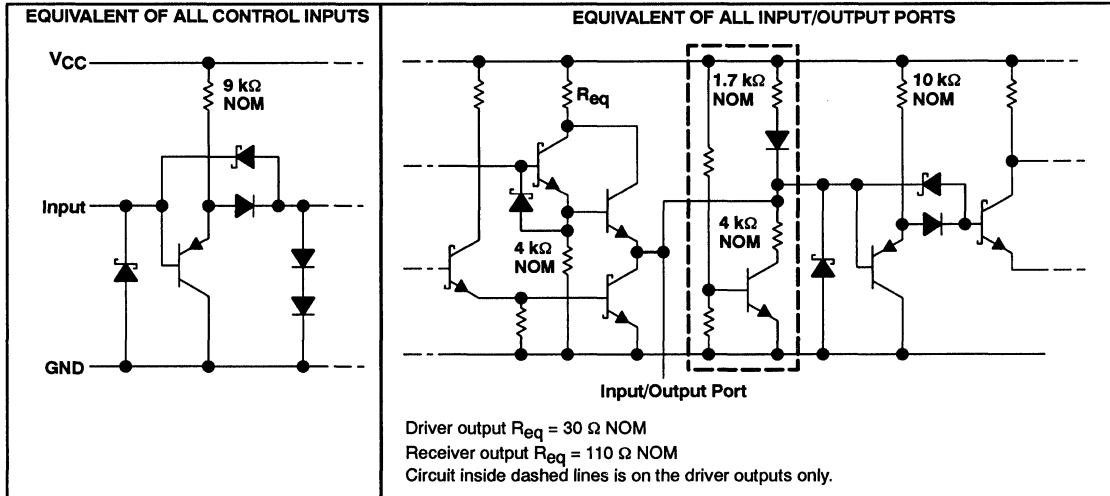
Terminal I/O Ports

GPIB I/O Ports

# SN75ALS165 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS023B – D3011, JUNE 1986 – REVISED AUGUST 1989

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage	5.5 V
Low-level driver output current	100 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1025 mW	8.2 mW/°C	656 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$		2			V
Low-level input voltage, $V_{IL}$		0.8			V
High-level output current, $I_{OH}$	Bus ports with pullups active	-5.2			mA
	Terminal ports	-800			μA
Low-level output current, $I_{OL}$	Bus ports	48			mA
	Terminal ports	16			mA
Operating free-air temperature, $T_A$		0	70		°C

# SN75ALS165 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS023B - D3011, JUNE 1986 - REVISED AUGUST 1989

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage		$I_I = -18 \text{ mA}$	-0.8		1.5	V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	Bus		0.4	0.65		V
$V_{OH}‡$	High-level output voltage	Terminal	$I_{OH} = -800 \mu\text{A}$ , TE at 0.8 V	2.7	3.5		V
		Bus	$I_{OH} = -5.2 \text{ mA}$ , PE and TE at 2 V	2.5	3.3		
$V_{OL}$	Low-level output voltage	Terminal	$I_{OL} = 16 \text{ mA}$ , TE at 0.8 V		0.3	0.5	V
		Bus	$I_{OL} = 48 \text{ mA}$ , TE at 2 V		0.35	0.5	
$I_I$	Input current at maximum input voltage	Terminal	$V_I = 5.5 \text{ V}$		0.2	100	$\mu\text{A}$
$I_{IH}$	High-level input current	Terminal and control inputs	$V_I = 2.7 \text{ V}$		0.1	20	$\mu\text{A}$
$I_{IL}$	Low-level input current	Terminal and control inputs	$V_I = 0.5 \text{ V}$		-10	-100	$\mu\text{A}$
$V_{I/O(\text{bus})}$	Voltage at bus port	Driver disabled	$I_I(\text{bus}) = 0$	2.5	3	3.7	V
			$I_I(\text{bus}) = -12 \text{ mA}$			-1.5	
$I_{I/O(\text{bus})}$	Current into bus port	Power on	Driver disabled	$V_I(\text{bus}) = -1.5 \text{ V to } 0.4 \text{ V}$	-1.3		mA
				$V_I(\text{bus}) = 0.4 \text{ V to } 2.5 \text{ V}$	0	-3.2	
				$V_I(\text{bus}) = 2.5 \text{ V to } 3.7 \text{ V}$		2.5	
				$V_I(\text{bus}) = 3.7 \text{ V to } 5 \text{ V}$	0	2.5	
				$V_I(\text{bus}) = 5 \text{ V to } 5.5 \text{ V}$	0.7	2.5	
		Power off	$V_{CC} = 0$ , $V_I(\text{bus}) = 0 \text{ to } 2.5 \text{ V}$		40	$\mu\text{A}$	
$I_{OS}$	Short-circuit output current	Terminal		-15	-35	-75	mA
		Bus		-25	-50	-125	
$I_{CC}$	Supply current	No load	Terminal outputs low and enabled		42	65	mA
			Bus outputs low and enabled		52	80	
$C_{I/O(\text{bus})}$	Bus-port capacitance		$V_{CC} = 5 \text{ V to } 0$ , $V_{I/O} = 0 \text{ to } 2 \text{ V}$ , $f = 1 \text{ MHz}$		30		pF

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡  $V_{OH}$  applies for 3-state outputs only.

TEXAS  
INSTRUMENTS

# SN75ALS165

## OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS023B – D3011, JUNE 1986 – REVISED AUGUST 1989

switching characteristics over recommended range of operating free-air temperature (unless otherwise noted),  $V_{CC} = 5\text{ V}$

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	Terminal	Bus	$C_L = 30\text{ pF}$ , See Figure 1		7	20	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output					8	20	
$t_{PLH}$ Propagation delay time, low-to-high-level output	Bus	Terminal	$C_L = 30\text{ pF}$ , See Figure 2		7	14	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output					9	14	
$t_{PZH}$ Output enable time to high level	TE	Bus	$C_L = 15\text{ pF}$ , See Figure 3		19	30	ns
$t_{PHZ}$ Output disable time from high level					5	12	
$t_{PZL}$ Output enable time to low level					16	35	
$t_{PLZ}$ Output disable time from low level					9	20	
$t_{PZH}$ Output enable time to high level	TE	Terminal	$C_L = 15\text{ pF}$ , See Figure 4		13	30	ns
$t_{PHZ}$ Output disable time from high level					12	20	
$t_{PZL}$ Output enable time to low level					12	20	
$t_{PLZ}$ Output disable time from low level					11	20	
$t_{en}$ Output pullup enable time	PE	Terminal	$C_L = 15\text{ pF}$ , See Figure 5		11	22	ns
$t_{dis}$ Output pullup disable time					6	12	

† All typical values are at  $T_A = 25^\circ\text{C}$ .



# SN75ALS165 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

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## PARAMETER MEASUREMENT INFORMATION

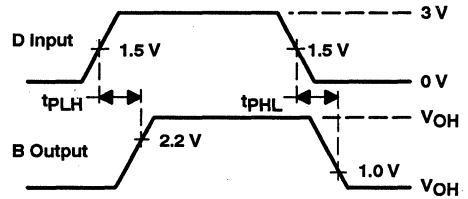
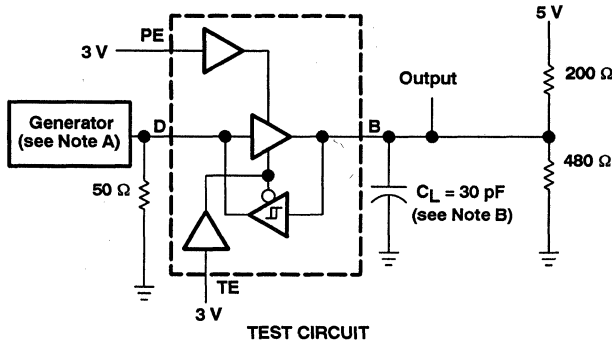


Figure 1. Terminal-to-Bus Test Circuit and Voltage Waveforms

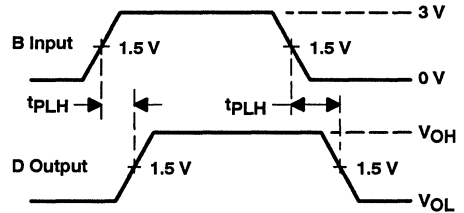
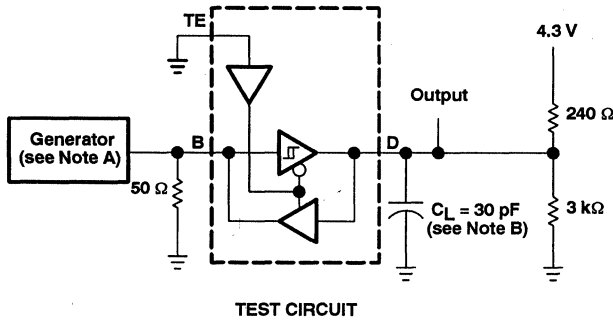


Figure 2. Bus-to-Terminal Test Circuit and Voltage Waveforms

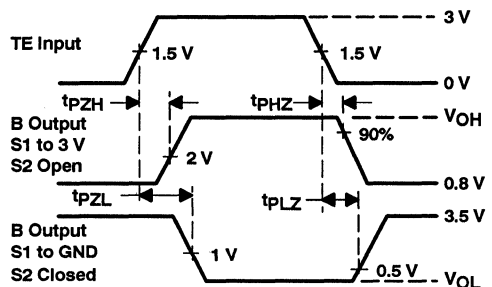
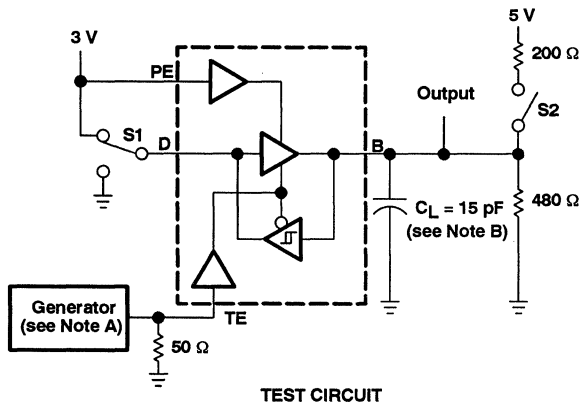


Figure 3. TE-to-Bus Test Circuit and Voltage Waveforms

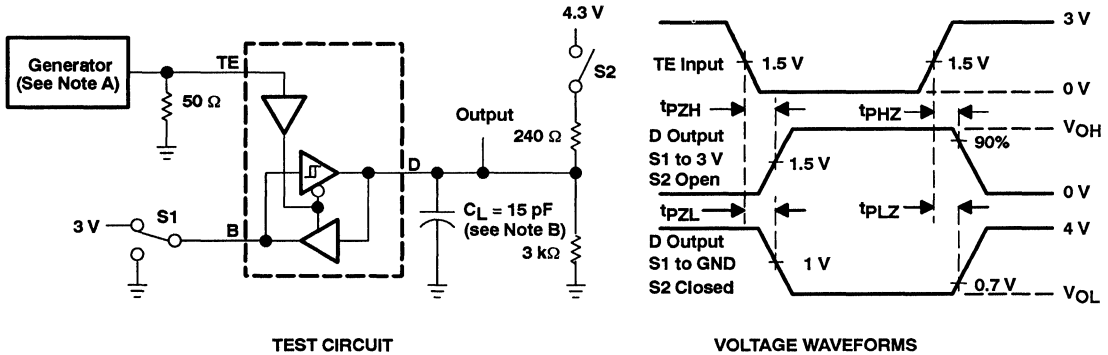
NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

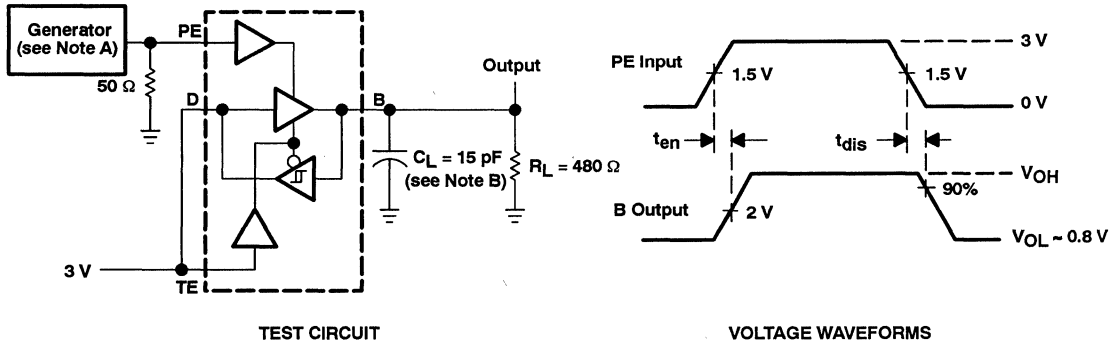
# SN75ALS165 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS023B - D3011, JUNE 1986 - REVISED AUGUST 1989

## PARAMETER MEASUREMENT INFORMATION



**Figure 4. TE-to-Terminal Test Circuit and Voltage Waveforms**



**Figure 5. PE-to-Bus Test Circuit and Voltage Waveforms**

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

# SN75ALS165 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS023B – D3011, JUNE 1986 – REVISED AUGUST 1989

## TYPICAL CHARACTERISTICS

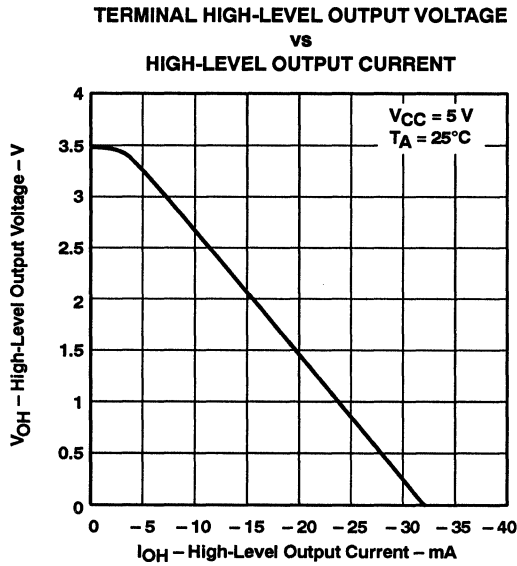


Figure 6

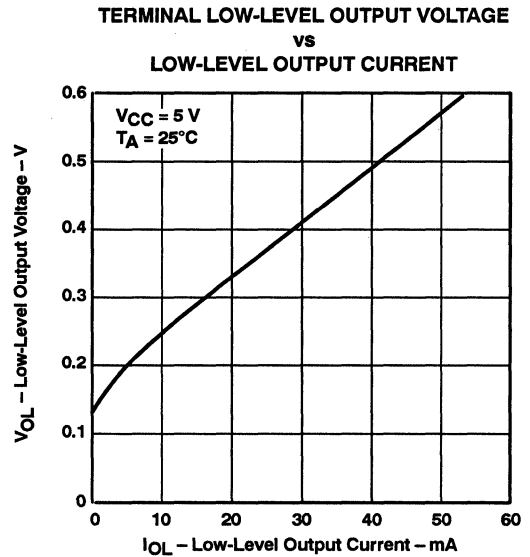


Figure 7

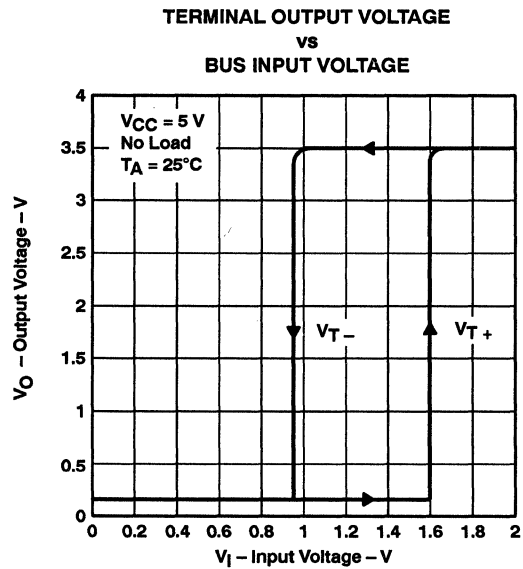


Figure 8

# SN75ALS165 OCTAL GENERAL-PURPOSE INTERFACE BUS TRANSCEIVER

SLLS023B - D3011, JUNE 1986 - REVISED AUGUST 1989

## TYPICAL CHARACTERISTICS

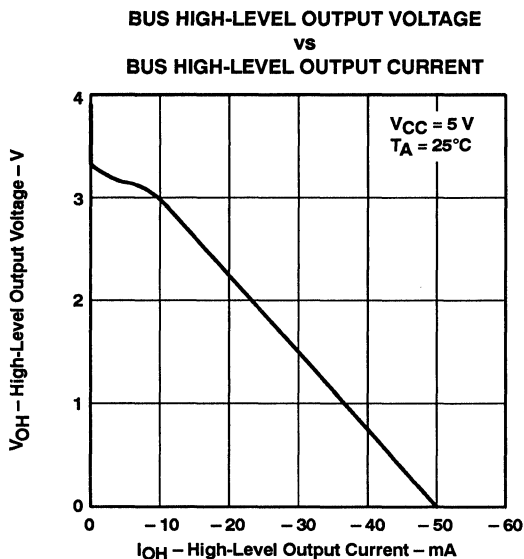


Figure 9

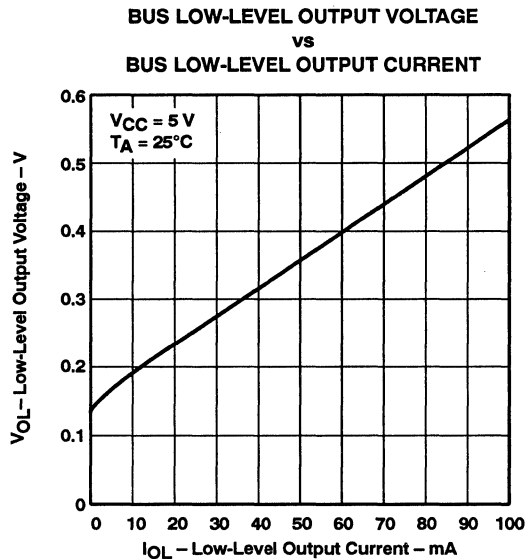


Figure 10

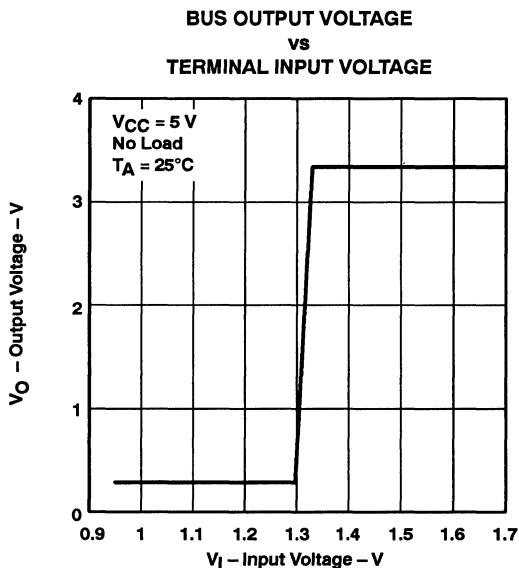


Figure 11



# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

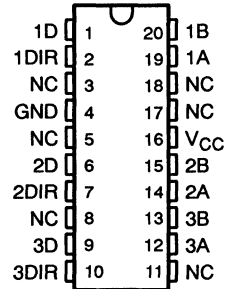
SLLS055B – D3040, AUGUST 1987 – REVISED FEBRUARY 1993

- Three Bidirectional Transceivers
- Driver Meets EIA Standards RS-422A and RS-485 and CCITT Recommendations V.11 and X.27 and ANSI Standard X3.131-1986
- High-Speed Advanced Low-Power Schottky Circuitry
- Designed for 25-MBaud Operation in Both Serial and Parallel Applications
- Low Skew . . . 6 ns Max
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Low Supply Current Requirements  
90 mA Max
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capacity . . .  $\pm 60$  mA
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Impedances . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 300$  mV Max
- Receiver Input Hysteresis . . . 60 mV Typ
- Operates From a Single 5-V Supply
- Glitch-Free Power-Up and Power-Down Protection
- Features Independent Direction Controls for Each Channel

## description

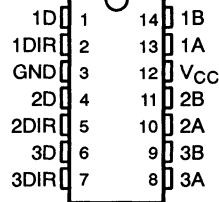
The SN75ALS170 triple differential bus transceiver is a monolithic integrated circuit designed for bidirectional data communication on multipoint bus transmission lines. It is designed for balanced transmission lines and the driver meets EIA Standards RS-422-A and RS-485 and CCITT recommendations V.11 and X.27 and ANSI Standard X3.131-1986.

**DW PACKAGE  
(TOP VIEW)**



NC – No internal connection

**J PACKAGE  
(TOP VIEW)**



## Function Tables

**EACH DRIVER**

INPUT D	DIR	OUTPUTS	
		A	B
H	H	H	L
L	H	L	H
X	L	Z	Z

**EACH RECEIVER**

DIFFERENTIAL INPUTS A – B	DIR	OUTPUT R
$V_{ID} \geq 0.3$ V	L	H
$-0.3$ V < $V_{ID} < 0.3$ V	L	?
$V_{ID} \leq -0.3$ V	L	L
X	H	Z
Open	L	H

H = high level, L = low level, ? = indeterminate;  
X = irrelevant, Z = high impedance (off)

# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

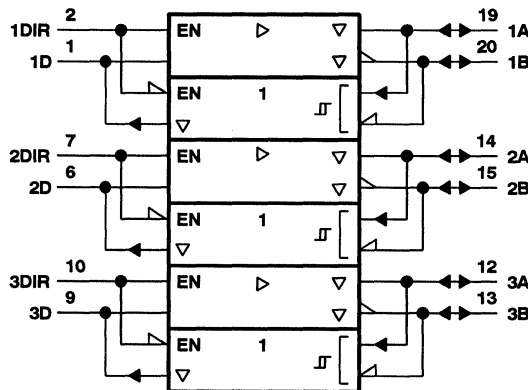
SLLS055B – D3040, AUGUST 1987 – REVISED FEBRUARY 1993

## description (continued)

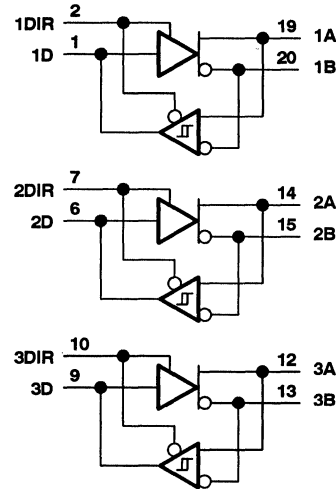
The SN75ALS170 operates from a single 5-V power supply. The drivers and receivers have active-high and active-low enables, respectively, which are internally connected together to function as a direction control. The driver differential outputs and the receiver differential inputs are connected internally to form differential input/output (I/O) bus ports that are designed to offer minimum loading to the bus when the driver is disabled or  $V_{CC} = 0$ . These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

The SN75ALS170 is characterized for operation from 0°C to 70°C.

## logic symbol†



## logic diagram (positive logic)



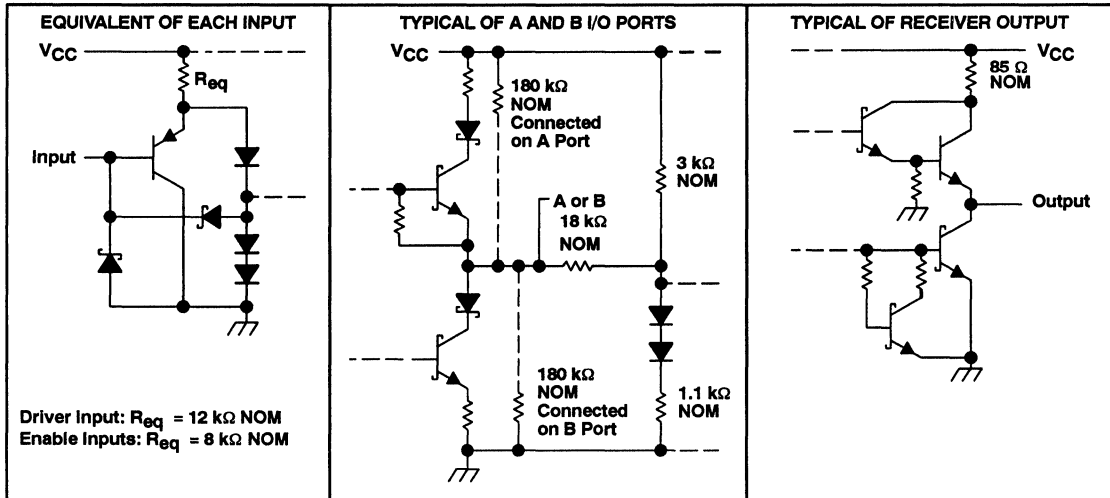
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the DW package.

# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS055B – D3040, AUGUST 1987 – REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	-10 V to 15 V
Enable input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: DW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

NOTE 1: All voltage values, except differential I/O bus voltage, are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
J	1025 mW	8.2 mW/°C	656 mW

**TEXAS**  
**INSTRUMENTS**

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# SN75ALS170

## TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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### recommended operating conditions

		MIN	TYP	MAX	UNIT	
Supply voltage, $V_{CC}$		4.75	5	5.25	V	
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$					12	V
					-7	
High-level input voltage, $V_{IH}$	D, DIR	2			V	
Low-level input voltage, $V_{IL}$	D, DIR	0.8			V	
Differential input voltage, $V_{ID}$ (see Note 2)		±12			V	
High-level output current, $I_{OH}$	Driver	-60			mA	
	Receiver	-400			μA	
Low-level output current, $I_{OL}$	Driver	60			mA	
	Receiver	8				
Operating free-air temperature, $T_A$		0		70	°C	

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.



# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS055B – D3040, AUGUST 1987 – REVISED FEBRUARY 1993

## DRIVER SECTION

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT		
$V_{IK}$	Input clamp voltage	$I_I = -19 \text{ mA}$				-1.5	V		
$V_O$	Output voltage	$I_O = 0$		0		6	V		
$V_{OH}$	High-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ ,	$V_{IH} = 2 \text{ V}$ , $I_{OH} = -55 \text{ mA}$	2.7			V		
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ ,	$V_{IH} = 2 \text{ V}$ , $I_{OL} = 55 \text{ mA}$			1.7	V		
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5		6	V		
$ V_{OD2} $	Differential output voltage	$R_L = 100 \Omega$ ,	See Figure 1	$1/2 V_{OD1}$ or 2 <sup>#</sup>			V		
		$R_L = 54 \Omega$ ,	See Figure 1	1.5	2.5	5	V		
$V_{OD3}$	Differential output voltage	$V_{test} = -7 \text{ V to } 12 \text{ V}$ , See Figure 2		1.5		5	V		
$\Delta  V_{OD} $	Change in magnitude of differential output voltage§	$R_L = 540 \Omega \text{ or } 100 \Omega$ , See Figure 1				$\pm 0.2$	V		
$V_{OC}$	Common-mode output voltage					3		V	
$\Delta  V_{OC} $	Change in magnitude of common-mode output voltage§							-1	V
								$\pm 0.2$	V
$I_O$	Output current	Output disabled, See Note 3	$V_O = 12 \text{ V}$			1	mA		
			$V_O = -7 \text{ V}$			-0.8			
$I_{IH}$	High-level input current	$V_I = 2.4 \text{ V}$				20	$\mu\text{A}$		
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$				-400	$\mu\text{A}$		
$I_{OS}$	Short-circuit output current¶	$V_O = -6 \text{ V}$				-250	mA		
		$V_O = 0 \text{ V}$				-150			
		$V_O = V_{CC}$				250			
		$V_O = 8 \text{ V}$				250			
$I_{CC}$	Supply current	No load	Outputs enabled		69	90	mA		
			Outputs disabled		57	78			

† The power-off measurement in EIA Standard RS-422-A applies to disabled outputs only and is not applied to combined inputs and outputs.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

§  $\Delta |V_{OD}|$  and  $\Delta |V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$  respectively, that occur when the input is changed from a high level to a low level.

¶ Duration of the short-circuit current should not exceed one second.

# The minimum  $V_{OD2}$  with a 100- $\Omega$  load is either  $1/2 V_{OD1}$  or 2 V, whichever is greater.

NOTE 3: This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions. The RS-422-A limit does not apply for a combined driver and receiver terminal.



# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS055B – D3040, AUGUST 1987 – REVISED FEBRUARY 1993

switching characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYPT	MAX	UNIT
t <sub>dD</sub>	Differential-output delay time	R <sub>L</sub> = 54 Ω, C <sub>L</sub> = 50 pF, See Figure 3	3	8	13	ns
		R <sub>L1</sub> = R <sub>L3</sub> = 165 Ω, R <sub>L2</sub> = 75 Ω, C <sub>L</sub> = 60 pF, See Figure 4	3	8	13	
	Skew ( t <sub>dDH</sub> - t <sub>dDL</sub>  )	R <sub>L</sub> = 54 Ω, C <sub>L</sub> = 50 pF, See Figure 3		1	6	ns
		R <sub>L1</sub> = R <sub>L3</sub> = 165 Ω, R <sub>L2</sub> = 75 Ω, C <sub>L</sub> = 60 pF, See Figure 4		1	6	
t <sub>tD</sub>	Differential-output transition time	R <sub>L</sub> = 54 Ω, C <sub>L</sub> = 50 pF, See Figure 3	3	8	13	ns
		R <sub>L1</sub> = R <sub>L3</sub> = 165 Ω, R <sub>L2</sub> = 75 Ω, C <sub>L</sub> = 60 pF, See Figure 4	3	8	13	

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

### SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
V <sub>O</sub>	V <sub>oa</sub> , V <sub>ob</sub>	V <sub>oa</sub> , V <sub>ob</sub>
V <sub>OD1</sub>	V <sub>O</sub>	V <sub>O</sub>
V <sub>OD2</sub>	V <sub>t</sub> (R <sub>L</sub> = 100 Ω)	V <sub>t</sub> (R <sub>L</sub> = 54 Ω)
V <sub>OD3</sub>		V <sub>t</sub> (Test Termination Measurement 2)
V <sub>test</sub>		V <sub>tst</sub>
Δ V <sub>OD</sub>	V <sub>t</sub>   -  V̄ <sub>t</sub>	V <sub>t</sub>   -  V̄ <sub>t</sub>
V <sub>OC</sub>	V <sub>os</sub>	V <sub>os</sub>
Δ V <sub>OC</sub>	V <sub>os</sub> - V̄ <sub>os</sub>	V <sub>os</sub> - V̄ <sub>os</sub>
I <sub>OS</sub>	I <sub>sa</sub>  ,  I <sub>sb</sub>	
I <sub>O</sub>	I <sub>xa</sub>  ,  I <sub>xb</sub>	I <sub>ia</sub> , I <sub>ib</sub>

# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS055B – D3040, AUGUST 1987 – REVISED FEBRUARY 1993

## RECEIVER SECTION

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7\text{ V}$ ,	$I_O = -0.4\text{ mA}$			0.3	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5\text{ V}$ ,	$I_O = 8\text{ mA}$	-0.3‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )				60		mV
$V_{IK}$	Enable-input clamp voltage	$I_I = -18\text{ mA}$				-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 300\text{ mV}$ , See Figure 5	$I_{OH} = -400\text{ }\mu\text{A}$ ,			2.7	V
$V_{OL}$	Low-level output voltage	$V_{ID} = -300\text{ mV}$ , See Figure 5	$I_{OL} = 8\text{ mA}$ ,			0.45	V
$I_{OZ}$	High-impedance-state output current	$V_O = 2.4\text{ V}$				20	$\mu\text{A}$
		$V_O = 0.4\text{ V}$				-400	
$I_I$	Line input current	Other input = 0 V, See Note 4	$V_I = 12\text{ V}$			1	mA
			$V_I = -7\text{ V}$			-0.8	
$I_{IH}$	High-level enable-input current	$V_{IH} = 2.7\text{ V}$				20	$\mu\text{A}$
$I_{IL}$	Low-level enable-input current	$V_{IL} = 0.4\text{ V}$				-100	$\mu\text{A}$
$r_i$	Input resistance				12		k $\Omega$
$I_{OS}$	Short-circuit output current	$V_{ID} = 300\text{ mV}$ ,	$V_O = 0$	-15		-85	mA
$I_{CC}$	Supply current	No load	Outputs enabled		69	90	mA
			Outputs disabled		57	78	

† All typical values are at  $V_{CC} = 5\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 4: This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions.

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature range**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -1.5\text{ V to }1.5\text{ V}$ , See Figure 6	$C_L = 15\text{ pF}$ ,	9	14	19	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output			9	14	19	ns
Skew ( $ t_{PLH} - t_{PHL} $ )				2	6		ns

† All typical values are at  $V_{CC} = 5\text{ V}$  and  $T_A = 25^\circ\text{C}$ .



# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS055B – D3040, AUGUST 1987 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

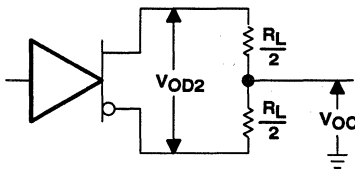


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

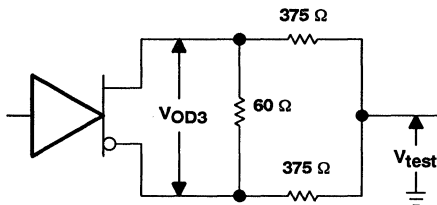
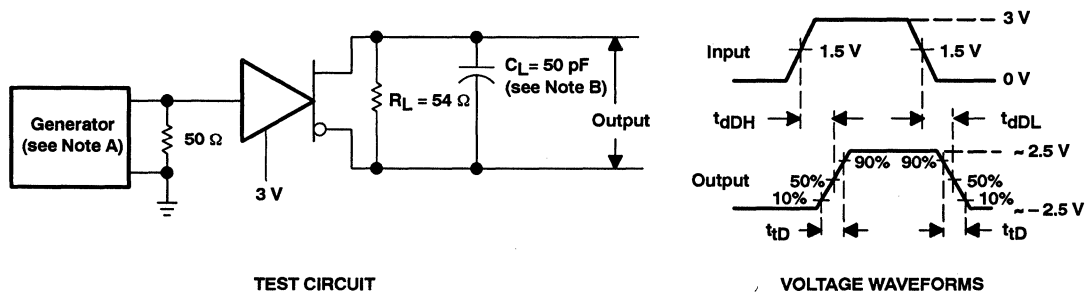


Figure 2. Driver  $V_{OD3}$



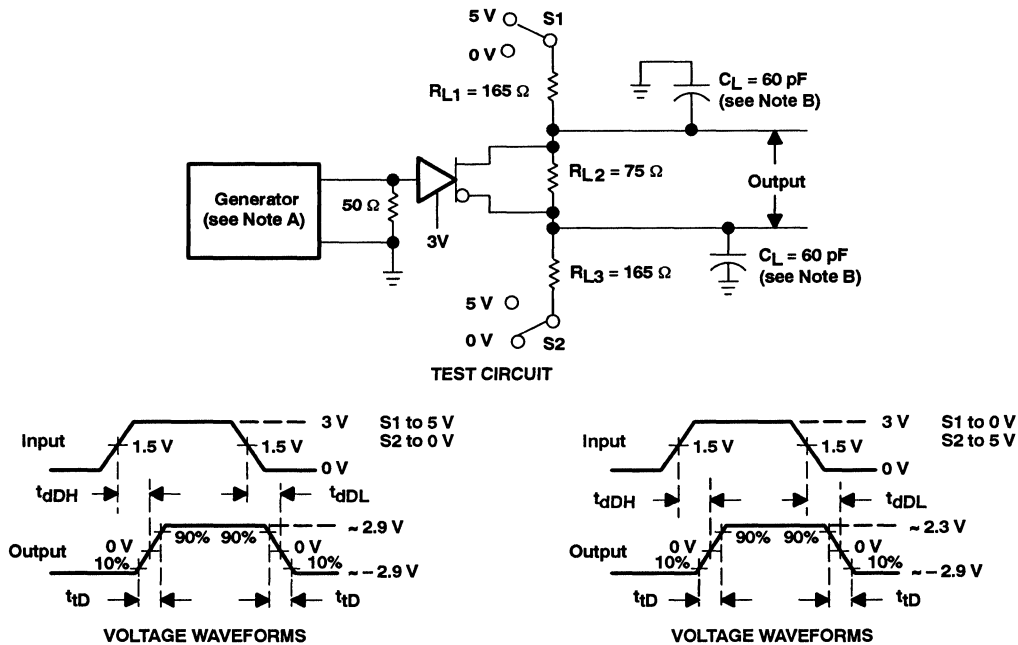
TEST CIRCUIT

VOLTAGE WAVEFORMS

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

Figure 3. Driver Test Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION



NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 4. Driver Test Circuit and Voltage Waveforms With Double-Differential-SCSI Termination for the Load

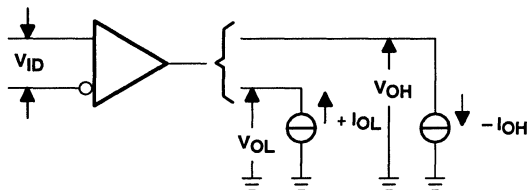
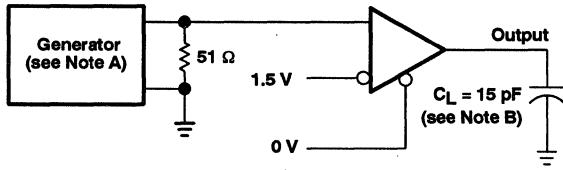


Figure 5. Receiver  $V_{OH}$  and  $V_{OL}$

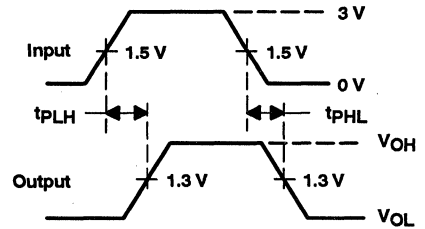
# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS055B - D3040, AUGUST 1987 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1 \text{ MHz}$ , 50% duty cycle,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ ,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 6. Receiver Test Circuit and Voltage Waveforms

## TYPICAL CHARACTERISTICS

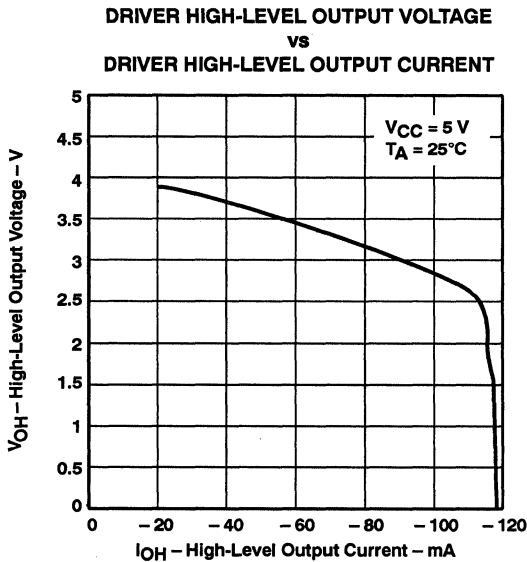


Figure 7

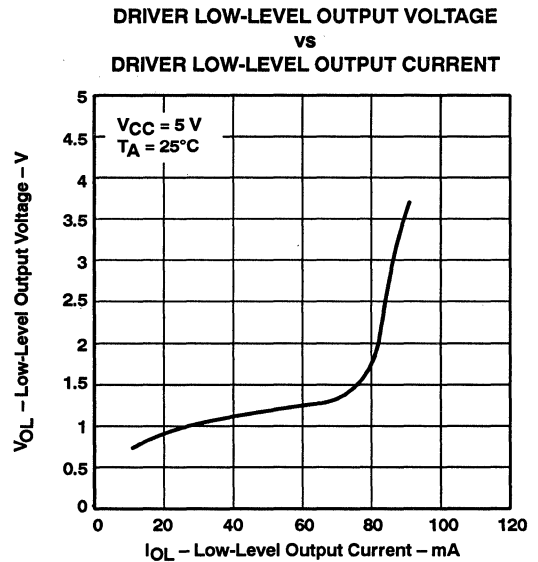


Figure 8

TYPICAL CHARACTERISTICS

DRIVER DIFFERENTIAL OUTPUT VOLTAGE  
vs  
DRIVER OUTPUT CURRENT

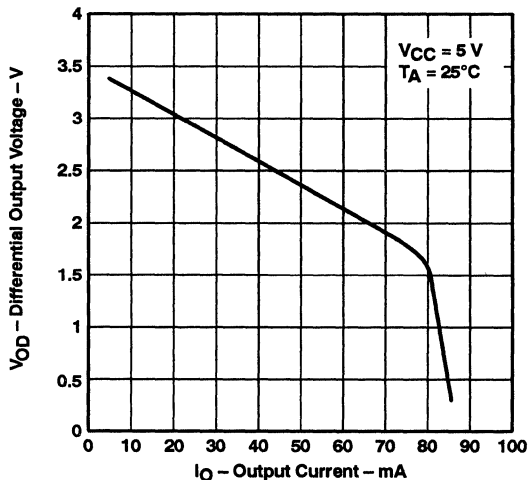


Figure 9

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

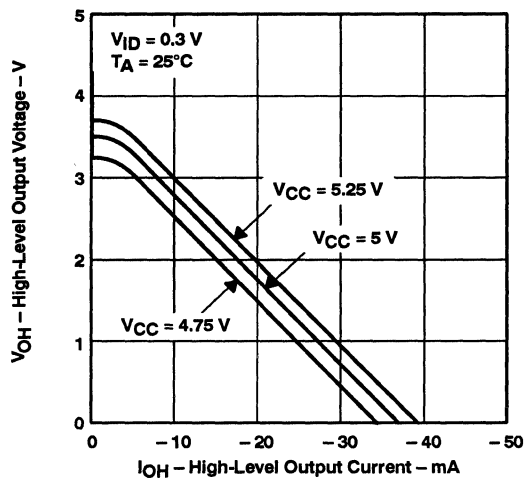


Figure 10

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

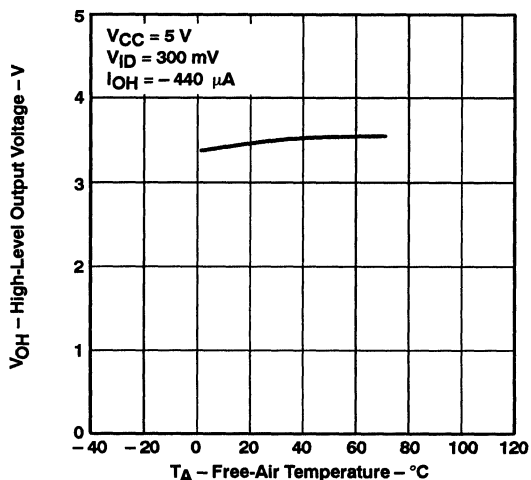


Figure 11

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
RECEIVER LOW-LEVEL OUTPUT CURRENT

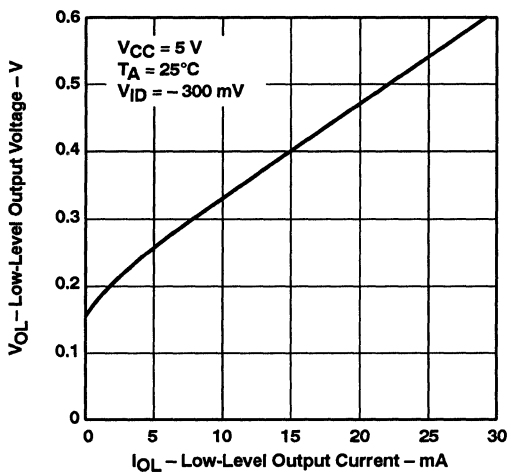


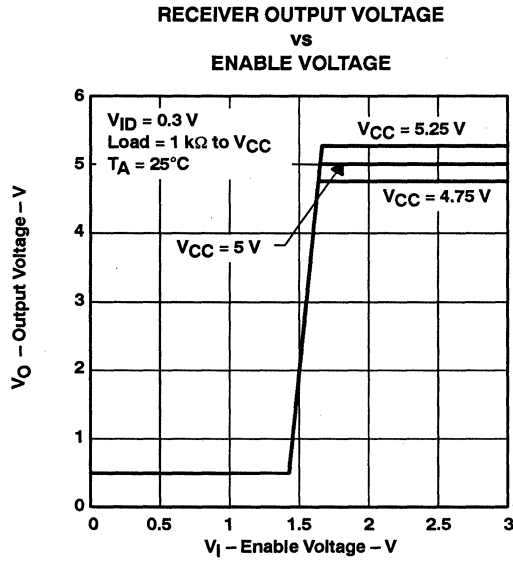
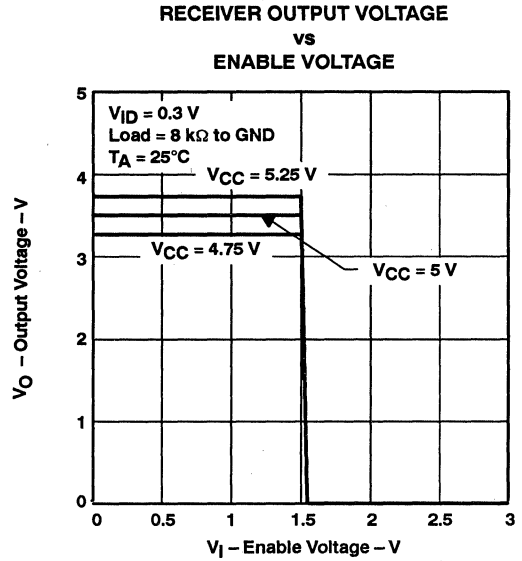
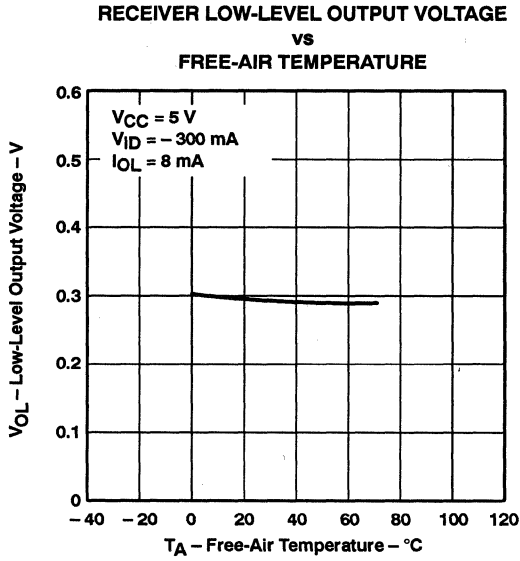
Figure 12



# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS055B - D3040, AUGUST 1987 - REVISED FEBRUARY 1993

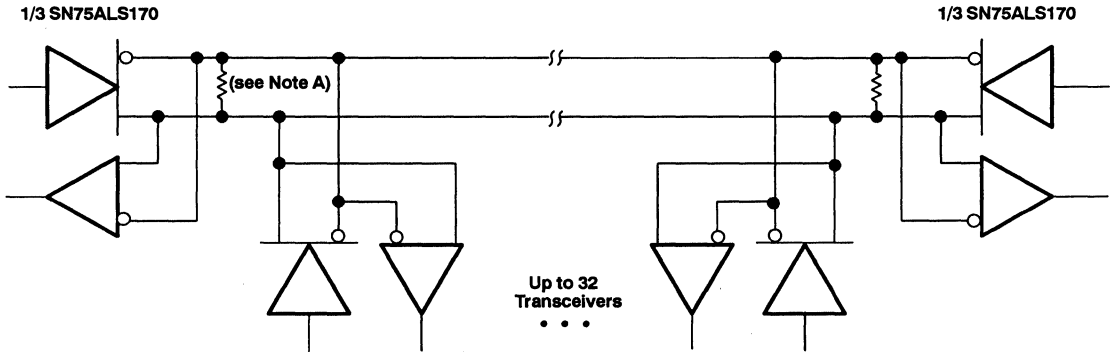
## TYPICAL CHARACTERISTICS



# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

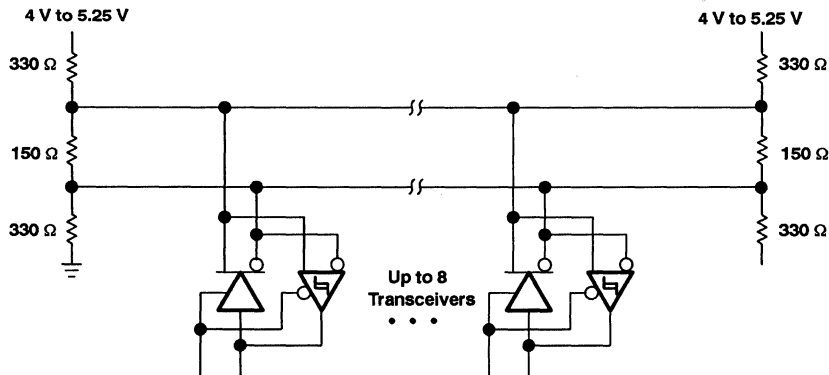
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## APPLICATION INFORMATION



NOTE A: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

**Figure 16. Typical Application Circuit**



**Figure 17. Typical Differential SCSI Application Circuit**

# SN75ALS170 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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## APPLICATION INFORMATION

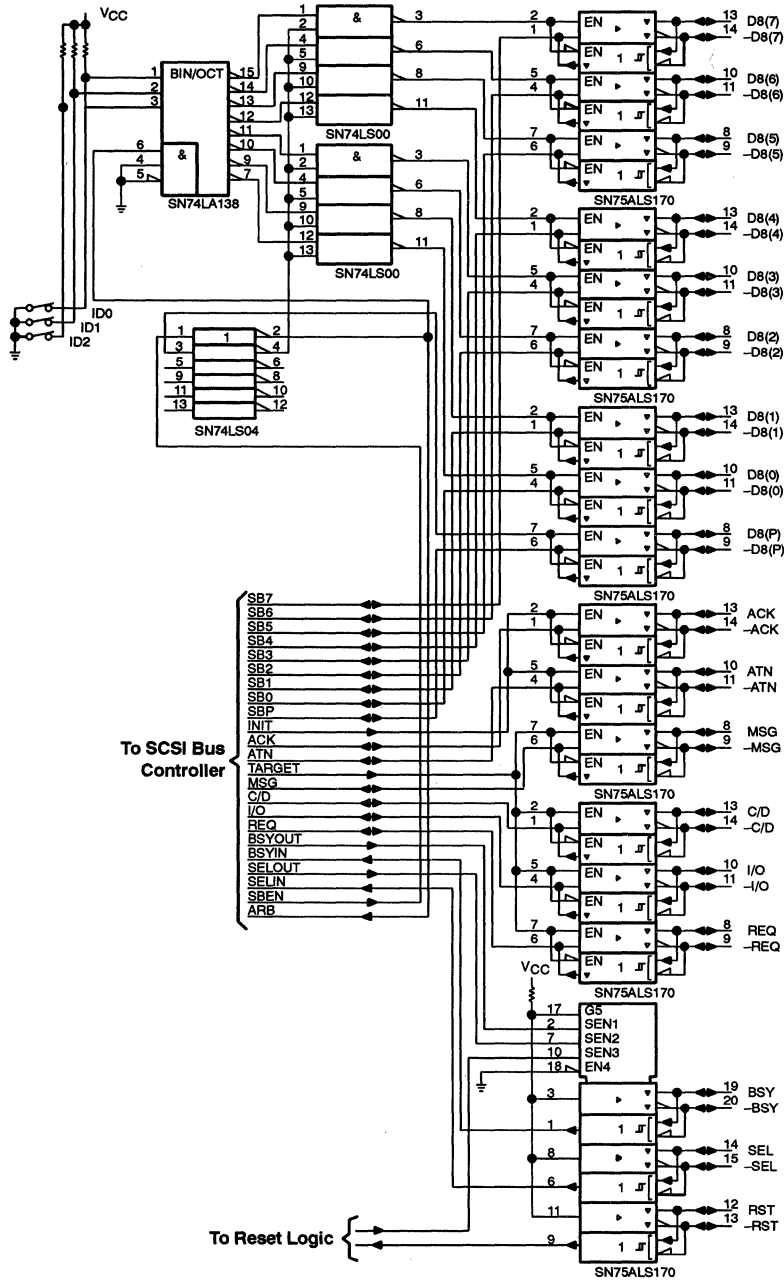


Figure 18. Typical Differential SCSI Bus Interface Implementation

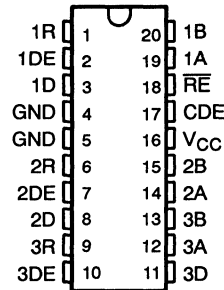
**TEXAS**  
**INSTRUMENTS**

# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS056B – D3041, AUGUST 1987 – REVISED FEBRUARY 1993

- Three Bidirectional Transceivers
- Driver Meets EIA Standards RS-422A and RS-485 and CCITT Recommendations V.11 and X.27 and ANSI Standard X3.131-1986
- High-Speed Advanced Low-Power Schottky Circuitry
- Designed for 25-MBaud Operation in Both Serial and Parallel Applications
- Low Skew . . . 6 ns Max
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Features Independent Driver Enables and Combined Receiver Enables
- Wide Positive and Negative Input/Output Bus Voltages Ranges
- Driver Output Capacity . . .  $\pm 60$  mA
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Impedances . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 300$  mV Max
- Receiver Input Hysteresis . . . 60 mV Typ
- Operates From a Single 5-V Supply
- Glitch-Free Power-Up and Power-Down Protection
- Low Supply Current Requirements  
90 mA Max

DW OR J PACKAGE  
(TOP VIEW)



Function Tables

EACH DRIVER

INPUT D	ENABLE		OUTPUTS	
	DE	CDE	A	B
H	H	H	H	L
L	H	H	L	H
X	L	X	Z	Z
X	X	L	Z	Z

EACH RECEIVER

DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.3$ V	L	H
$-0.3$ V < $V_{ID}$ < 0.3 V	L	?
$V_{ID} \leq -0.3$ V	L	L
X	H	Z
Open	L	H

H = high level, L = low level, ? = indeterminate;  
X = irrelevant, Z = high impedance (off)

## description

The SN75ALS171 triple differential bus transceiver is a monolithic integrated circuit designed for bidirectional data communication on multipoint bus transmission lines. It is designed for balanced transmission lines, and the driver meets EIA Standards RS-422-A and RS-485 and CCITT recommendations V.11 and X.27 and ANSI Standard X3.131-1986.

The SN75ALS171 operates from a single 5-V power supply. The drivers and receivers have individual active-high and active-low enables, respectively, which can be externally connected together to function as a direction control. The driver differential output and the receiver differential input pairs are connected internally to form differential input/output (I/O) bus ports that are designed to offer minimum loading to the bus when the driver is disabled or  $V_{CC}$  is at 0 V. These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

The SN75ALS171 is characterized for operation from 0°C to 70°C.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



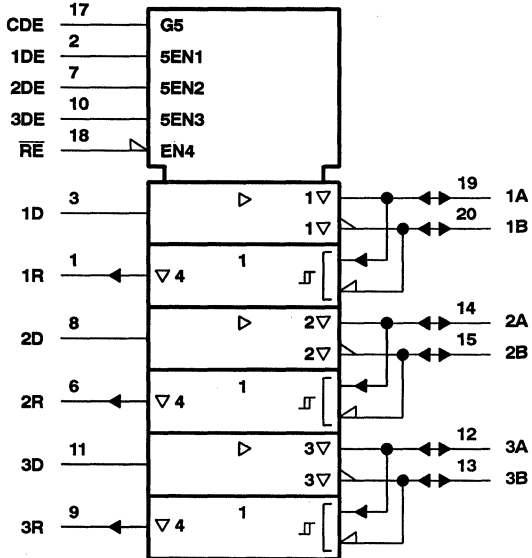
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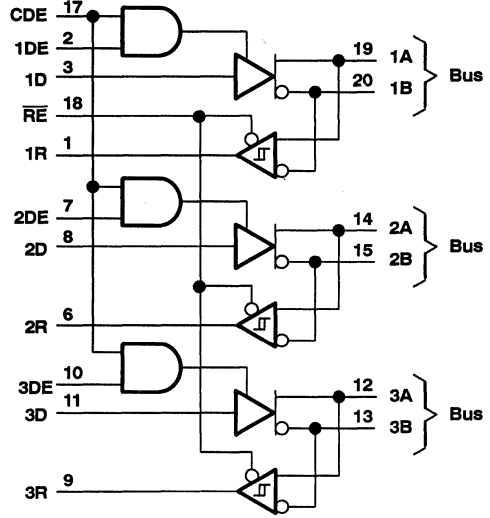
# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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## logic symbol†

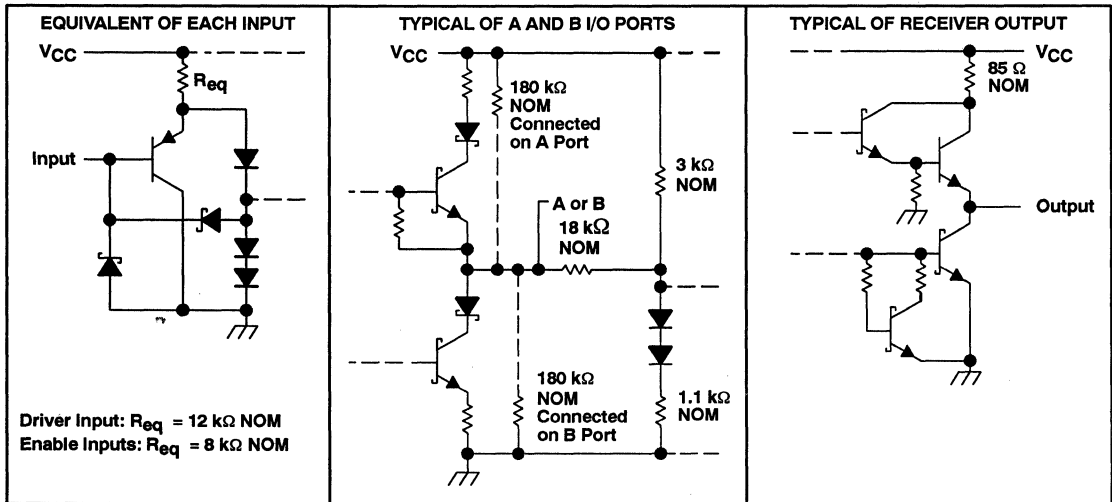


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematics of inputs and outputs



# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	–10 V to 15 V
Enable input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: DW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

NOTE 1: All voltage values, except differential I/O bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
J	1025 mW	8.2 mW/°C	656 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$	–7		12	V
High-level input voltage, $V_{IH}$	D, CDE, DE, and $\overline{RE}$		2	V
Low-level input voltage, $V_{IL}$	D, CDE, DE, and $\overline{RE}$		0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)			±12	V
High-level output current, $I_{OH}$	Driver		–60	mA
	Receiver		–400	µA
Low-level output current, $I_{OL}$	Driver		60	mA
	Receiver		8	
Operating free-air temperature, $T_A$	0		70	°C

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.



# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$				-1.5	V
$V_O$	Output voltage	$I_O = 0$		0		6	V
$V_{OH}$	High-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ ,	$V_{IH} = 2 \text{ V}$ , $I_{OH} = -55 \text{ mA}$	2.7			V
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$	$V_{IH} = 2 \text{ V}$ , $I_{OL} = 55 \text{ mA}$			1.7	V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5		6	V
$ V_{OD2} $	Differential output voltage	$R_L = 100 \Omega$ ,	See Figure 1	$1/2 V_{OD1}$ or 2#	2.5	5	V
		$R_L = 54 \Omega$ ,	See Figure 1	1.5	2.5	5	
$V_{OD3}$	Differential output voltage	$V_{test} = -7 \text{ V to } 12 \text{ V}$ , See Figure 2		1.5		5	V
$\Delta V_{OD} $	Change in magnitude of differential output voltage§	$R_L = 54 \Omega \text{ or } 100 \Omega$ , See Figure 1				$\pm 0.2$	V
$V_{OC}$	Common-mode output voltage					3	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage§					$\pm 0.2$	V
$I_O$	Output current	Output disabled, See Note 3		$V_O = 12 \text{ V}$		1	mA
				$V_O = -7 \text{ V}$		-0.8	
$I_{IH}$	High-level enable-input current	D and DE	$V_{IH} = 2.7 \text{ V}$			20	$\mu\text{A}$
		CDE				60	
$I_{IL}$	Low-level enable-input current	D and DE	$V_{IL} = 0.4 \text{ V}$			-100	$\mu\text{A}$
		CDE				-900	
$I_{OS}$	Short-circuit output current¶	$V_O = -6 \text{ V}$				-250	mA
		$V_O = 0$				-150	
		$V_O = V_{CC}$				250	
		$V_O = 8 \text{ V}$				250	
$I_{CC}$	Supply current	No load		Outputs enabled		69 90	mA
				Outputs disabled		57 78	

† The power-off measurement in EIA Standard RS-422-A applies to disabled outputs only and is not applied to combined inputs and outputs.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

§  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$  respectively, that occur when the input is changed from a high level to a low level.

¶ Duration of the short-circuit current should not exceed one second.

# The minimum  $V_{OD2}$  with 100- $\Omega$  load is either  $1/2 V_{OD2}$  or 2 V, whichever is greater.

NOTE 3: This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions. The RS-422-A limit does not apply for a combined driver and receiver terminal.

TEXAS  
INSTRUMENTS

# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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**switching characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
t <sub>dD</sub>	Differential-output delay time	R <sub>L</sub> = 54 Ω, See Figure 5	C <sub>L</sub> = 50 pF,	3	8	13	ns
		R <sub>L1</sub> = R <sub>L3</sub> = 165 Ω, C <sub>L</sub> = 60 pF, See Figure 6	R <sub>L2</sub> = 75 Ω, V <sub>TERM</sub> = 5 V,	3	8	13	
Skew ( t <sub>dDH</sub> - t <sub>dDL</sub>  )		R <sub>L</sub> = 54 Ω, See Figure 3	C <sub>L</sub> = 50 pF,		1	6	ns
		R <sub>L1</sub> = R <sub>L3</sub> = 165 Ω, C <sub>L</sub> = 60 pF,	R <sub>L2</sub> = 75 Ω, See Figure 6		1	6	
t <sub>tD</sub>	Differential-output transition time	R <sub>L</sub> = 54 Ω, See Figure 3	C <sub>L</sub> = 50 pF,	3	8	13	ns
		R <sub>L1</sub> = R <sub>L3</sub> = 165 Ω, C <sub>L</sub> = 60 pF, See Figure 6	R <sub>L2</sub> = 75 Ω, V <sub>TERM</sub> = 5 V,	3	8	13	
t <sub>PZH</sub>	Output enable time to high level	R <sub>L</sub> = 110 Ω,	See Figure 4		30	50	ns
t <sub>PZL</sub>	Output enable time to low level	R <sub>L</sub> = 110 Ω,	See Figure 5		30	50	ns
t <sub>PHZ</sub>	Output disable time from high level	R <sub>L</sub> = 110 Ω,	See Figure 4	3	8	13	ns
t <sub>PLZ</sub>	Output disable time from low level	R <sub>L</sub> = 110 Ω,	See Figure 5	3	8	13	ns
t <sub>PDE</sub>	Differential-output enable time	R <sub>L1</sub> = R <sub>L3</sub> = 165 Ω,	R <sub>L2</sub> = 75 Ω,	8	30	45	ns
t <sub>PDZ</sub>	Differential-output disable time	C <sub>L</sub> = 60 pF,	See Figure 7	5	10	15	ns

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

### SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
V <sub>O</sub>	V <sub>Oa</sub> , V <sub>Ob</sub>	V <sub>Oa</sub> , V <sub>Ob</sub>
V <sub>OD1</sub>	V <sub>O</sub>	V <sub>O</sub>
V <sub>OD2</sub>	V <sub>t</sub> (R <sub>L</sub> = 100 Ω)	V <sub>t</sub> (R <sub>L</sub> = 54 Ω)
V <sub>OD3</sub>		V <sub>t</sub> (Test Termination Measurement 2)
V <sub>test</sub>		V <sub>tst</sub>
Δ V <sub>OD</sub>	V <sub>t</sub>   -  V̄ <sub>t</sub>	V <sub>t</sub>   -  V̄ <sub>t</sub>
V <sub>OC</sub>	V <sub>os</sub>	V <sub>os</sub>
Δ V <sub>OC</sub>	V <sub>os</sub> - V̄ <sub>os</sub>	V <sub>os</sub> - V̄ <sub>os</sub>
I <sub>OS</sub>	I <sub>sa</sub>  ,  I <sub>sb</sub>	
I <sub>O</sub>	I <sub>xa</sub>  ,  I <sub>xb</sub>	I <sub>ia</sub> , I <sub>ib</sub>





# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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## RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7$ V,	$I_O = -0.4$ mA			0.3	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5$ V,	$I_O = 8$ mA	-0.3‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )				60		mV
$V_{IK}$	Enable-input clamp voltage		$I_I = -18$ mA			-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 300$ mV, See Figure 8	$I_{OH} = -400$ $\mu$ A,		2.7		V
$V_{OL}$	Low-level output voltage	$V_{ID} = -300$ mV, See Figure 8	$I_{OL} = 8$ mA			0.45	V
$I_{OZ}$	High-impedance-state output current	$V_O = 0.4$ V to 2.4 V				$\pm 20$	$\mu$ A
$I_I$	Line input current	Other input = 0 V, See Note 4	$V_I = 12$ V			1	mA
			$V_I = -7$ V			-0.8	
$I_{IH}$	High-level enable-input current	$V_{IH} = 2.7$ V				60	$\mu$ A
$I_{IL}$	Low-level enable-input current	$V_{IL} = 0.4$ V				-300	$\mu$ A
$r_i$	Input resistance				12		k $\Omega$
$I_{OS}$	Short-circuit output current	$V_{ID} = 300$ mV,	$V_O = 0$	-15		-85	mA
$I_{CC}$	Supply current	No load	Outputs enabled		69	90	mA
			Outputs disabled		57	78	

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 4: This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions.

switching characteristics over recommended ranges of supply voltage and operating free-air temperature range

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -1.5$ V to 1.5 V, See Figure 9	$C_L = 15$ pF,	9	14	19	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output			9	14	19	ns
	Skew ( $ t_{PLH} - t_{PHL} $ )				2	6	ns
$t_{PZH}$	Output enable time to high level	$C_L = 15$ pF,	See Figure 10		7	14	ns
$t_{PZL}$	Output enable time to low level				7	14	ns
$t_{PHZ}$	Output disable time from high level	$C_L = 15$ pF,	See Figure 10		20	35	ns
$t_{PLZ}$	Output disable time from low level				8	17	ns

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .



PARAMETER MEASUREMENT INFORMATION

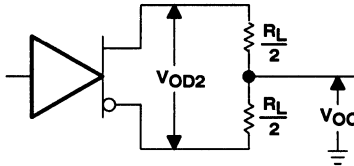


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

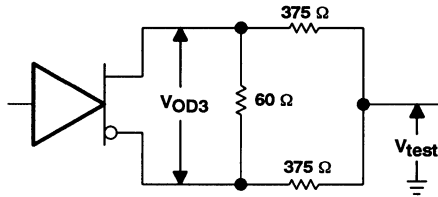
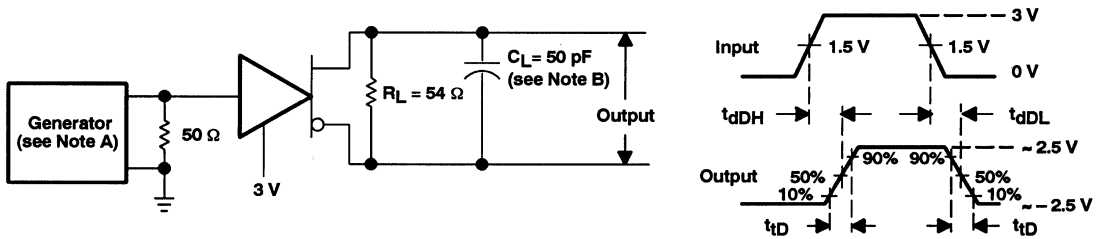


Figure 2. Driver  $V_{OD3}$



TEST CIRCUIT

VOLTAGE WAVEFORMS

Figure 3. Driver Test Circuit and Voltage Waveforms

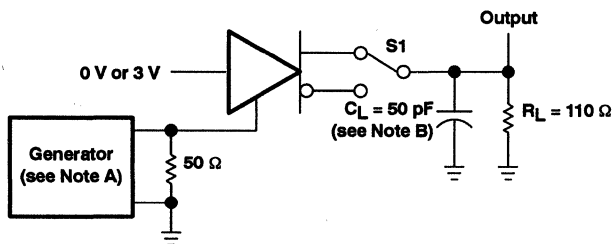
NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

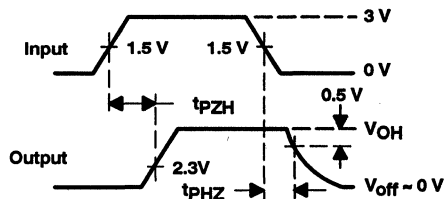
# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

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## PARAMETER MEASUREMENT INFORMATION

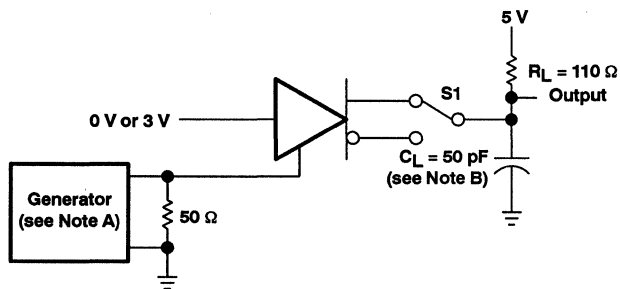


TEST CIRCUIT

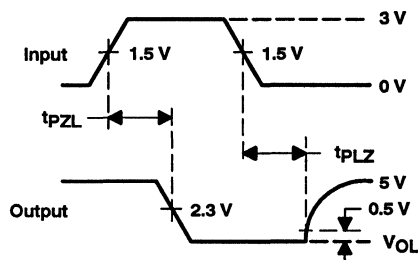


VOLTAGE WAVEFORMS

Figure 4. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT

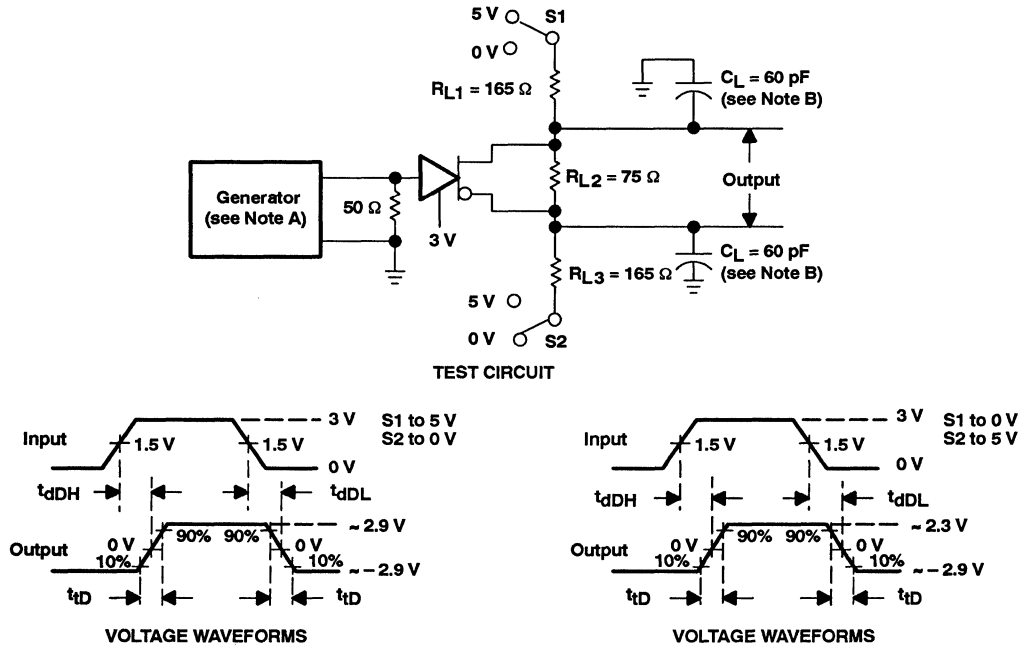


VOLTAGE WAVEFORMS

Figure 5. Driver Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

PARAMETER MEASUREMENT INFORMATION



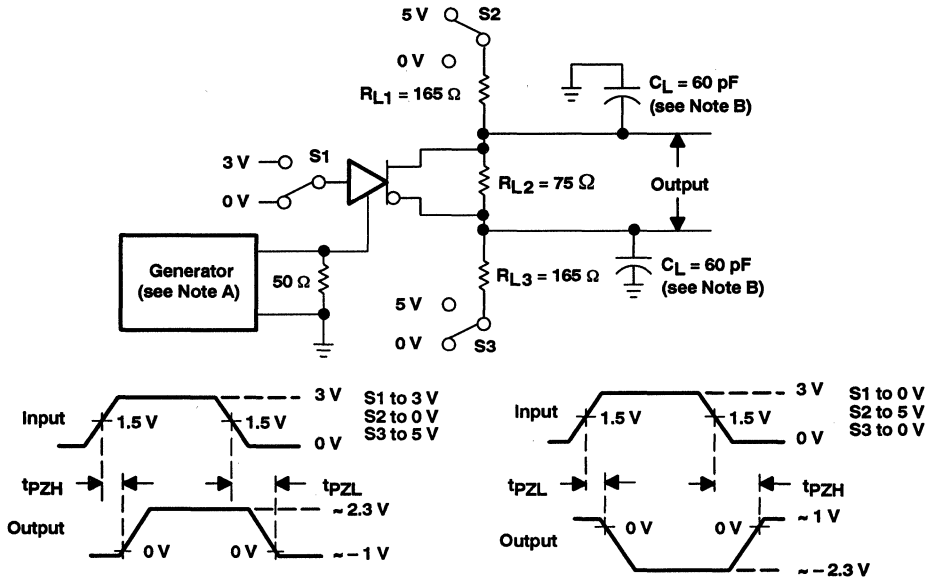
- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $\text{PRR} \leq 1 \text{ MHz}$ , 50% duty cycle,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ ,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 6. Driver Test Circuit and Voltage Waveforms With Double-Differential-SCSI Termination for the Load

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## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

Figure 7. Driver Differential-Enable and Disable Times With a Double-SCI Termination

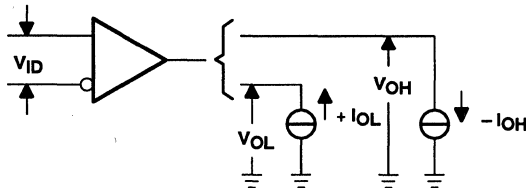
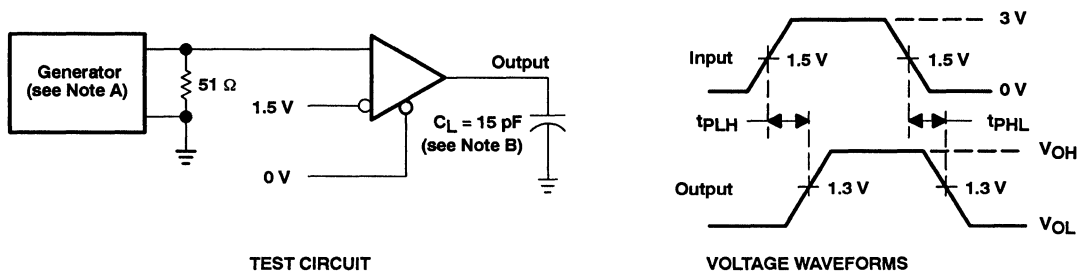


Figure 8. Receiver  $V_{OH}$  and  $V_{OL}$

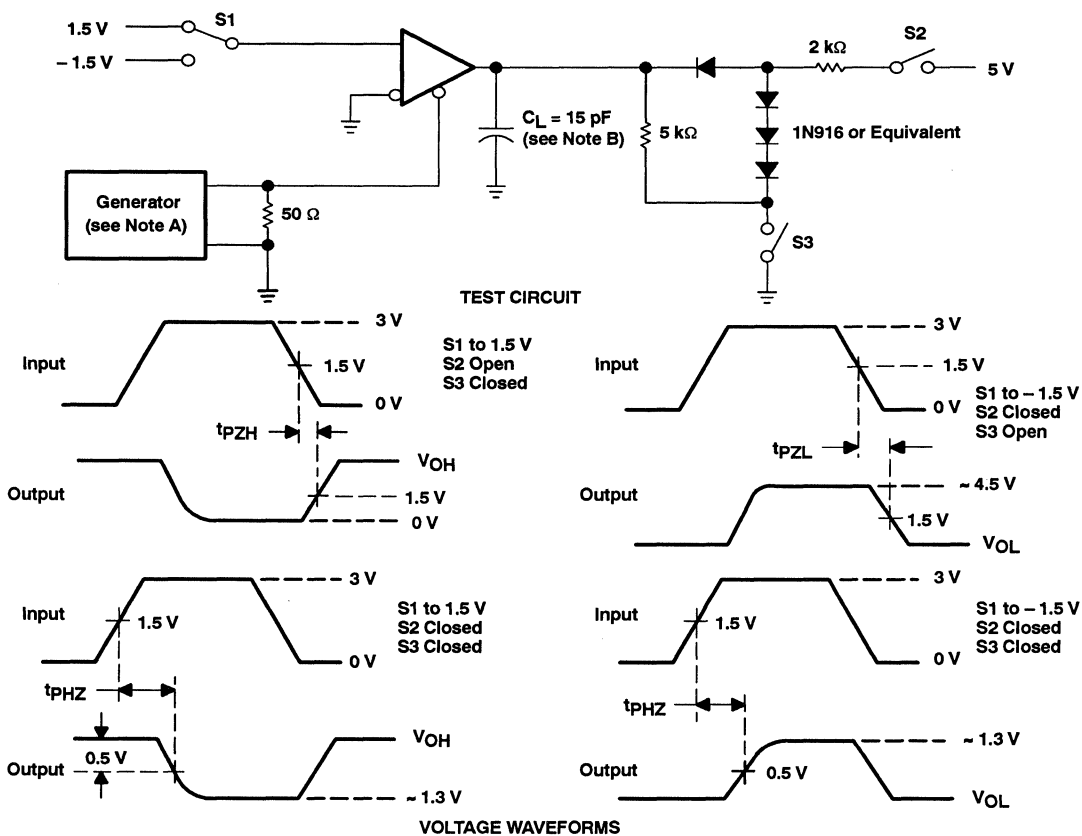
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## PARAMETER MEASUREMENT INFORMATION



**Figure 9. Receiver Test Circuit and Voltage Waveforms**



**Figure 10. Receiver Test Circuit and Voltage Waveforms**

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

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## TYPICAL CHARACTERISTICS

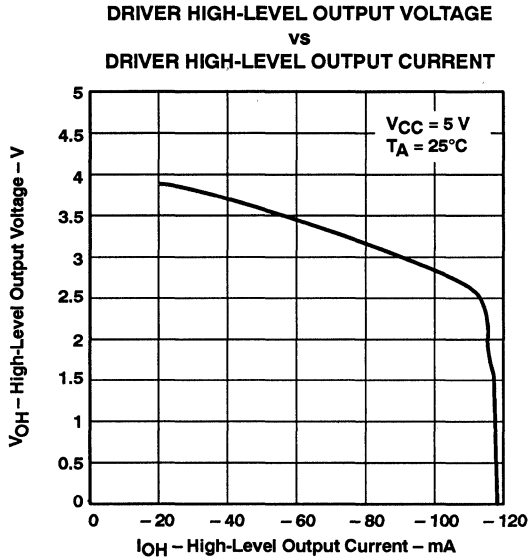


Figure 11

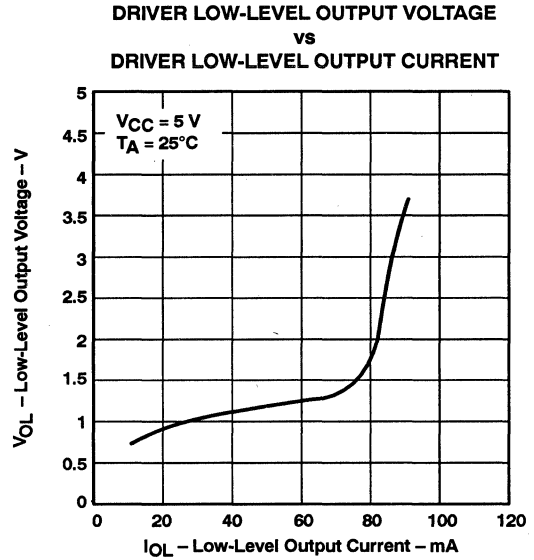


Figure 12

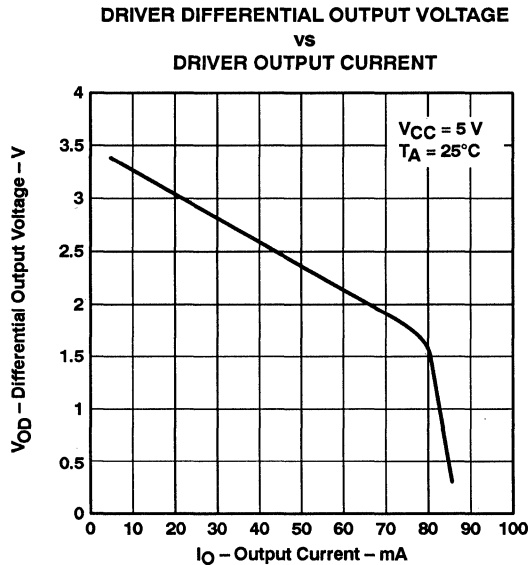


Figure 13

TYPICAL CHARACTERISTICS

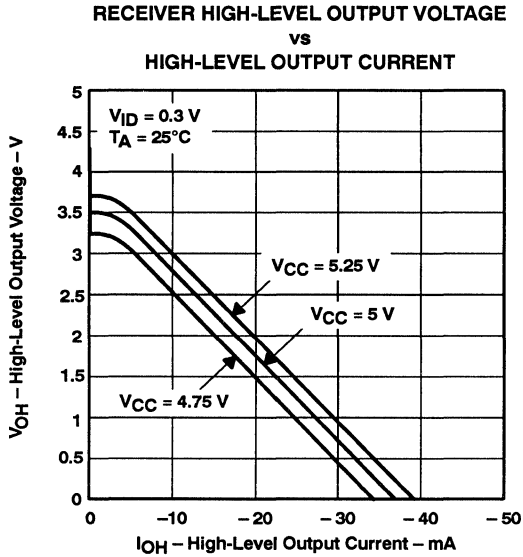


Figure 14

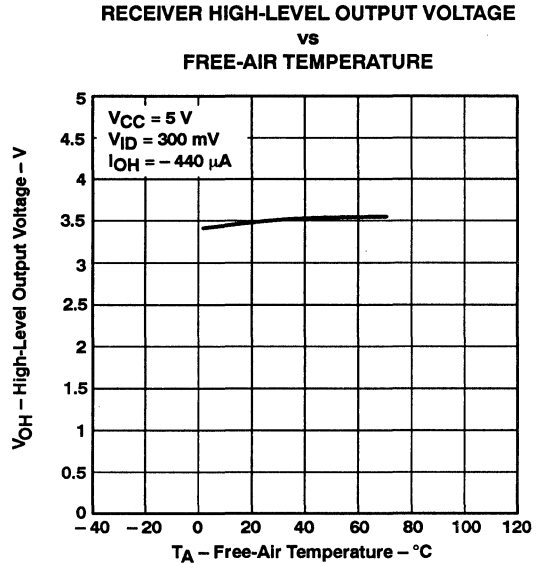


Figure 15

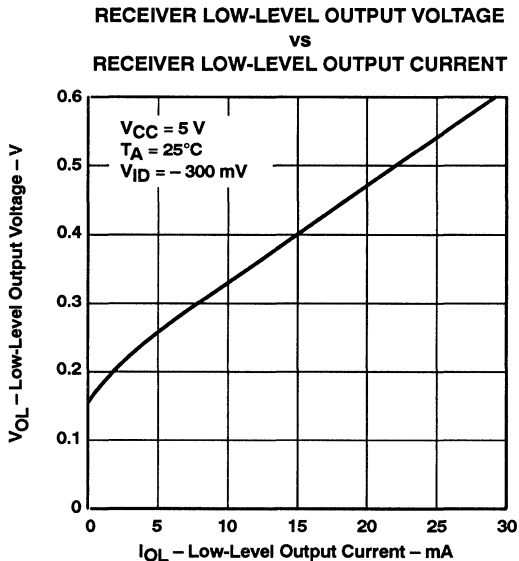


Figure 16

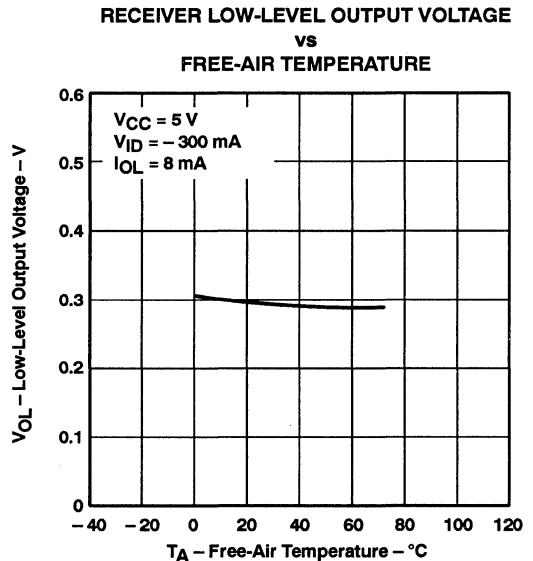


Figure 17



# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS056B - D3041, AUGUST 1987 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

RECEIVER OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE

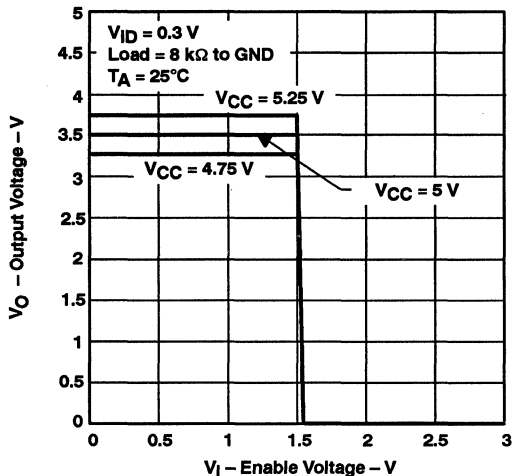


Figure 18

RECEIVER OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE

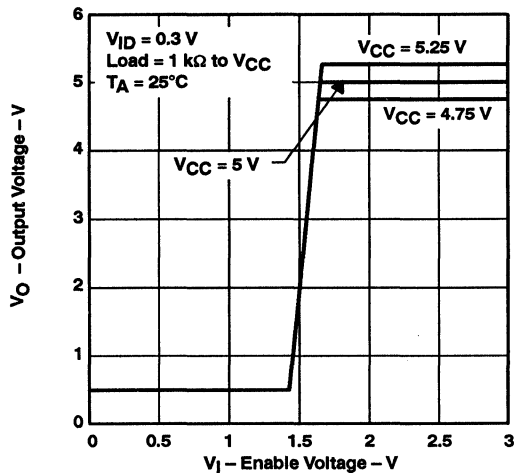
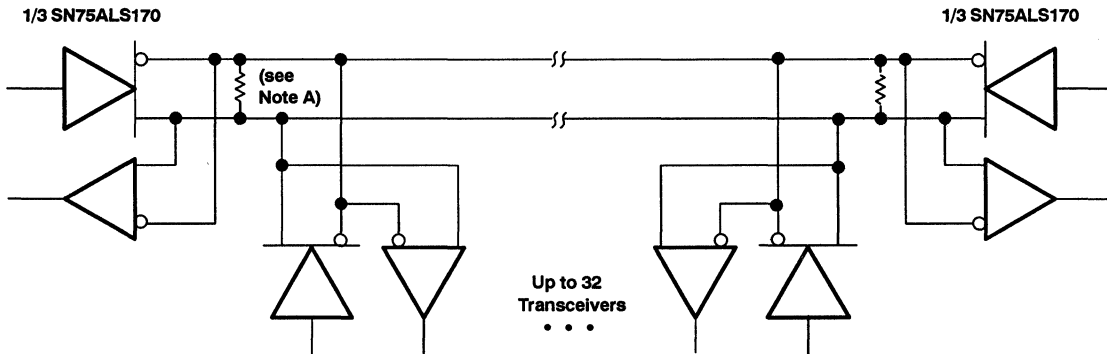


Figure 19

## APPLICATION INFORMATION



NOTE A: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

Figure 20. Typical Application Circuit

# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS056B - D3041, AUGUST 1987 - REVISED FEBRUARY 1993

## APPLICATION INFORMATION

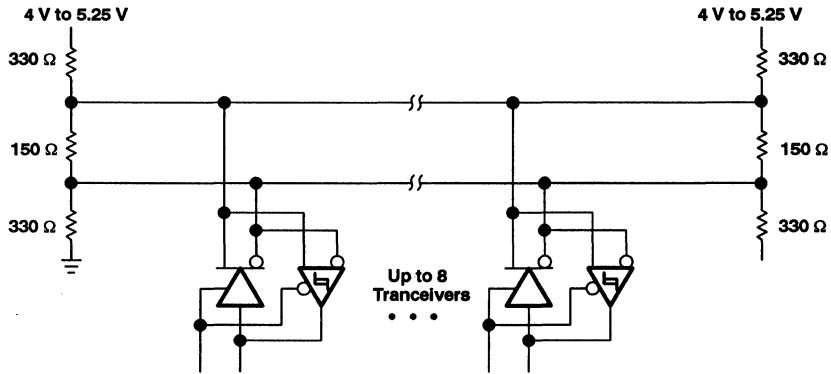


Figure 21. Typical Differential SCSI Application Circuit

# SN75ALS171 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS056B - D3041, AUGUST 1987 - REVISED FEBRUARY 1993

## APPLICATION INFORMATION

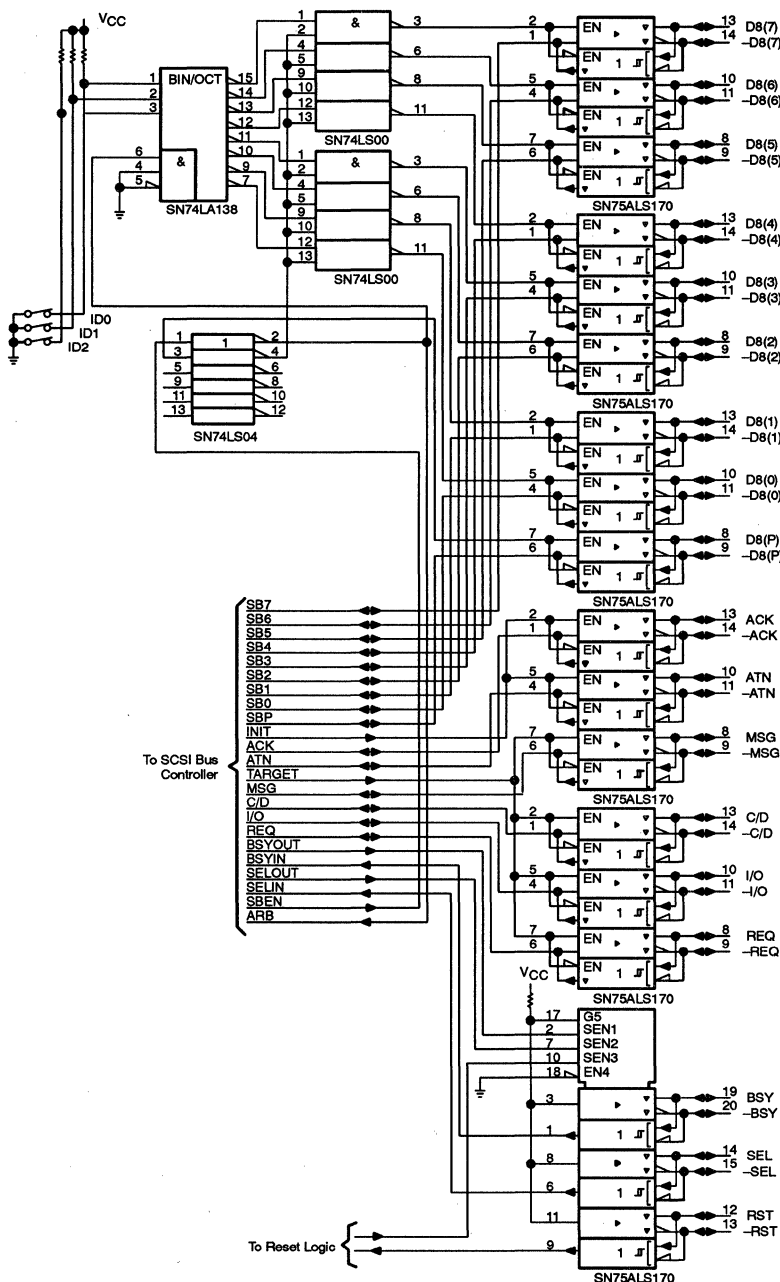


Figure 22. Typical Differential SCSI Bus Interface Implementation

# SN75172 QUAD DIFFERENTIAL LINE DRIVER

SLLS038A – D2596, OCTOBER 1980 – REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A and RS-485
- Meets CCITT Recommendations V.11 and X.27
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- 3-State Outputs
- Common-Mode Output Voltage Range of -7 V to 12 V
- Active-High and Active-Low Enables
- Thermal Shutdown Protection
- Positive- and Negative-Current Limiting
- Operates From Single 5-V Supply
- Low Power Requirements
- Functionally Interchangeable With AM26LS31

## description

The SN75172 is a monolithic quad differential line driver with 3-state outputs. It is designed to meet the requirements of EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27. The device is optimized for balanced multipoint bus transmission at rates of up to 4 megabaud. Each driver features wide positive and negative common-mode output voltage ranges making it suitable for party-line applications in noisy environments.

The SN75172 provides positive- and negative-current limiting and thermal shutdown for protection from line fault conditions on the transmission bus line. Shutdown occurs at a junction temperature of approximately 150°C. This device offers optimum performance when used with the SN75173 or SN75175 quadruple differential line receivers.

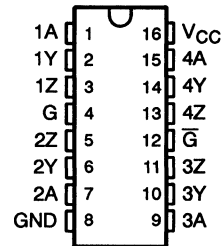
The SN75172 is characterized for operation from 0°C to 70°C.

**FUNCTION TABLE**  
(each driver)

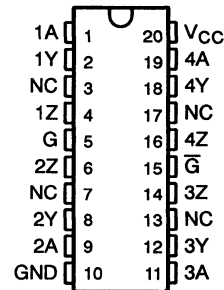
INPUT	ENABLES		OUTPUTS	
	G	$\bar{G}$	Y	Z
H	H	X	H	L
L	H	X	L	H
H	X	L	H	L
L	X	L	L	H
X	L	H	Z	Z

H = high level, L = low level,  
X = irrelevant, Z = high impedance (off)

**N PACKAGE**  
(TOP VIEW)



**DW PACKAGE**  
(TOP VIEW)

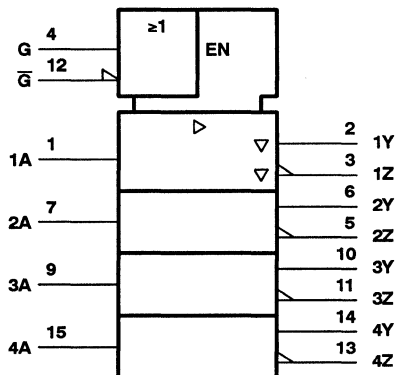


NC – No internal connection

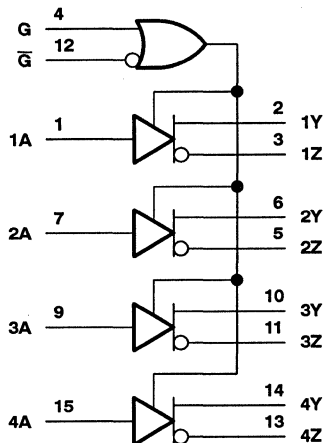
# SN75172 QUAD DIFFERENTIAL LINE DRIVER

SLLS038A - D2596, OCTOBER 1980 - REVISED FEBRUARY 1993

## logic symbol†



## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12. Pin numbers shown are for the N package.

## absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
Common-mode output voltage, $V_{OC}$			-7 to 12	V
High-level output current, $I_{OH}$			-60	mA
Low-level output current, $I_{OL}$			60	mA
Operating free-air temperature, $T_A$	0		70	°C



# SN75172 QUAD DIFFERENTIAL LINE DRIVER

SLLS038A - D2596, OCTOBER 1980 - REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V
$V_O$ Output voltage	$I_O = 0$	0		6	V
$V_{OH}$ High-level output voltage	$V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ , $I_{OH} = -33 \text{ mA}$		3.7		V
$V_{OL}$ Low-level output voltage	$V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ , $I_{OH} = 33 \text{ mA}$		1.1		V
$ V_{OD1} $ Differential output voltage	$I_O = 0$	1.5		6	V
$ V_{OD2} $ Differential output voltage	$R_L = 100 \Omega$ , See Figure 1	$1/2 V_{OD1}$ or $2^{\ddagger}$			V
	$R_L = 54 \Omega$ , See Figure 1	1.5	2.5	5	V
$V_{OD3}$ Differential output voltage	See Note 2	1.5		5	V
$\Delta V_{OD} $ Change in magnitude of differential output voltage <sup>‡</sup>				$\pm 0.2$	V
$V_{OC}$ Common-mode output voltage <sup>§</sup>	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1			+3 -1	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage <sup>‡</sup>				$\pm 0.2$	V
$I_O$ Output current with power off	$V_{CC} = 0$ , $V_O = -7 \text{ V to } 12 \text{ V}$			$\pm 100$	$\mu\text{A}$
$I_{OZ}$ High-impedance-state output current	$V_O = -7 \text{ V to } 12 \text{ V}$			$\pm 100$	$\mu\text{A}$
$I_{IH}$ High-level input current	$V_I = 2.7 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_I = 0.5 \text{ V}$			-360	mA
$I_{OS}$ Short-circuit output current	$V_O = -7 \text{ V}$			-180	mA
	$V_O = V_{CC}$			180	
	$V_O = 12 \text{ V}$			500	
$I_{CC}$ Supply current (all drivers)	No load	Outputs enabled	38	60	mA
		Outputs disabled	18	40	

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

§ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

¶ The minimum  $V_{OD2}$  with a  $100\text{-}\Omega$  load is either  $1/2 V_{OD1}$  or  $2 \text{ V}$ , whichever is greater.

NOTE 2: See Figure 3-5 of EIA Standard RS-485.

### SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{oa}, V_{ob}$	$V_{oa}, V_{ob}$
$ V_{OD1} $	$V_o$	$V_o$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD2} $		$V_t$ (Test Termination Measurement 2)
$\Delta V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

**TEXAS**  
**INSTRUMENTS**

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2-537

# SN75172 QUAD DIFFERENTIAL LINE DRIVER

SLLS038A - D2596, OCTOBER 1980 - REVISED FEBRUARY 1993

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{dD}$ Differential-output delay time	$R_L = 54\ \Omega$ , See Figure 2		45	65	ns
$t_tD$ Differential-output transition time			80	120	ns
$t_{pZH}$ Output enable time to high level	$R_L = 110\ \Omega$ , See Figure 3		80	120	ns
$t_{pZL}$ Output enable time to low level	$R_L = 110\ \Omega$ , See Figure 4		45	80	ns
$t_{pHZ}$ Output disable time from high level	$R_L = 110\ \Omega$ , See Figure 3		78	115	ns
$t_{pLZ}$ Output disable time from low level	$R_L = 110\ \Omega$ , See Figure 3		18	30	ns

## PARAMETER MEASUREMENT INFORMATION

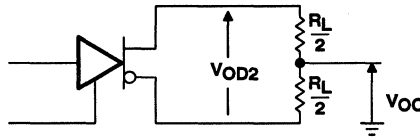


Figure 1. Differential and Common-Mode Output Voltages

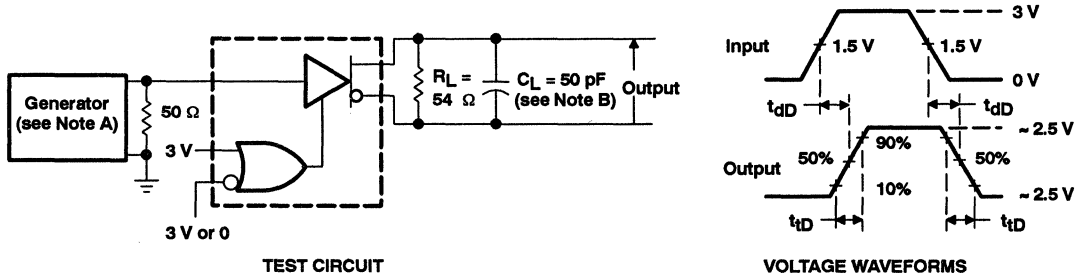


Figure 2. Differential-Output Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5\text{ ns}$ ,  $t_f \leq 5\text{ ns}$ ,  $\text{PRR} \leq 1\text{ MHz}$ , duty cycle = 50%,  $Z_O = 50\ \Omega$ .  
B.  $C_L$  includes probe and stray capacitance.

PARAMETER MEASUREMENT INFORMATION

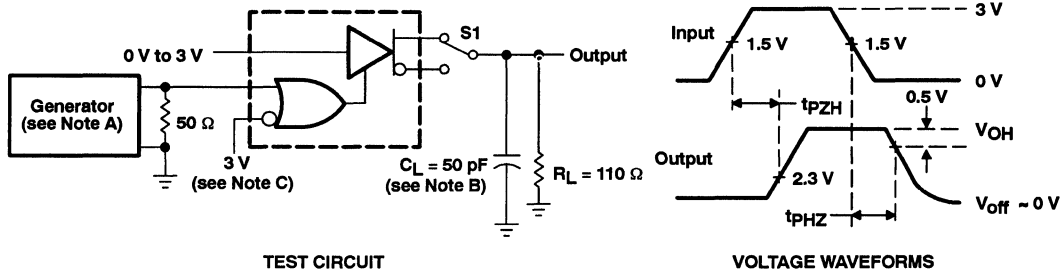


Figure 3. Test Circuit and Voltage Waveforms

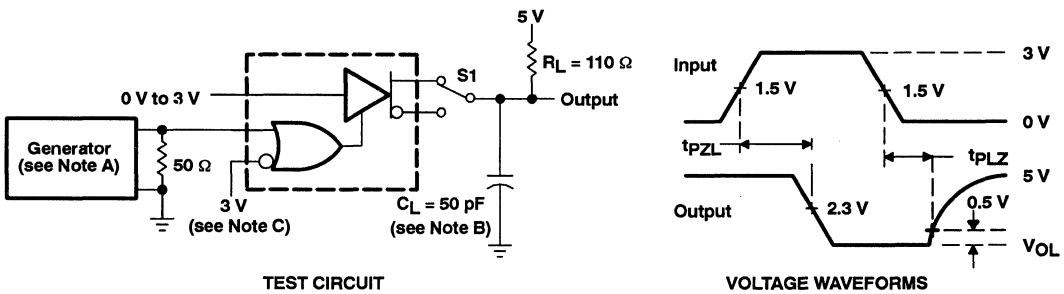


Figure 4. Test Circuit and Voltage Waveforms

NOTES: C. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle = 50%,  $t_r \leq$  5 ns,  $t_f \leq$  5 ns,  $Z_0 = 50 \Omega$ .

D.  $C_L$  includes probe and stray capacitance.

E. To test the active-low enable  $\overline{G}$ , ground G and apply an inverted waveform to  $\overline{G}$ .



# SN75172 QUAD DIFFERENTIAL LINE DRIVER

SLLS038A - D2596, OCTOBER 1980 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

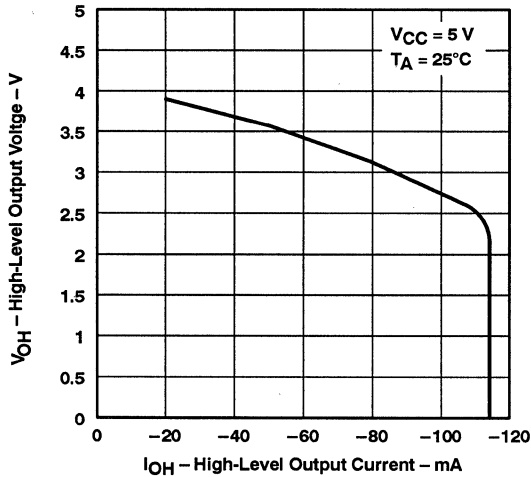


Figure 5

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

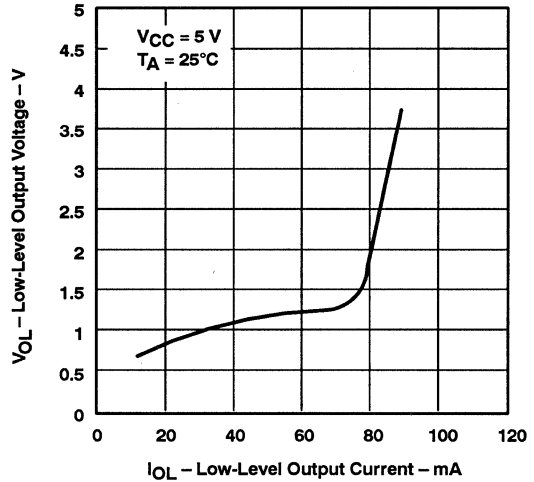


Figure 6

DIFFERENTIAL OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT

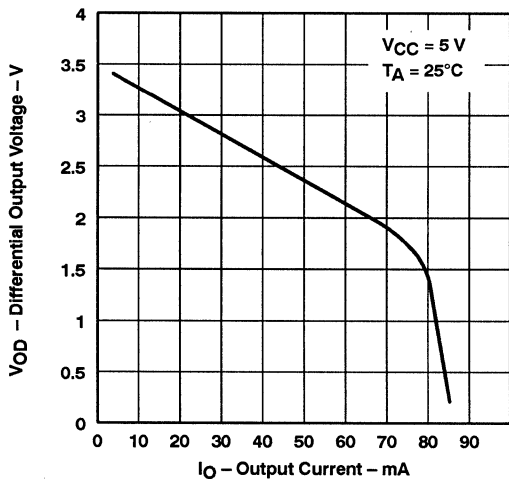


Figure 7

OUTPUT CURRENT  
vs  
OUTPUT VOLTAGE

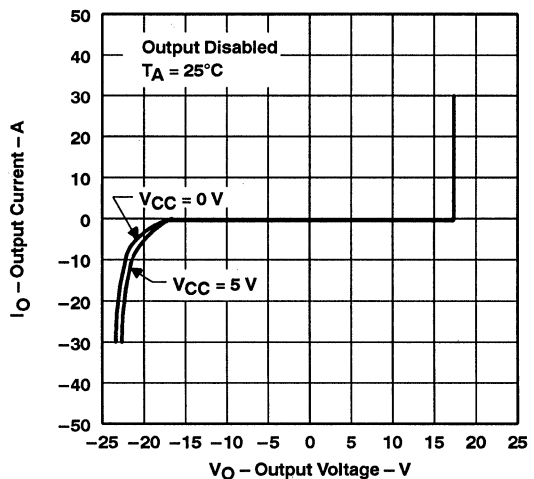


Figure 8

TYPICAL CHARACTERISTICS

SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE

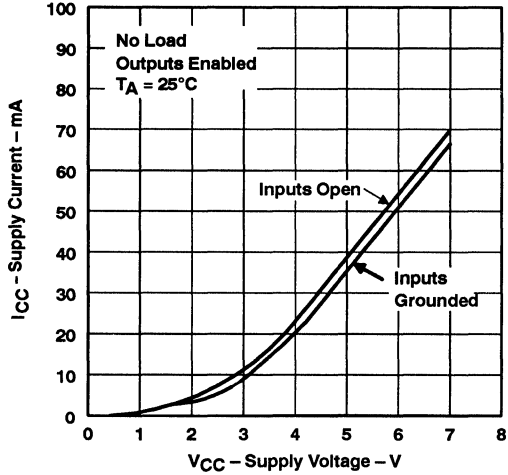


Figure 9

SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE

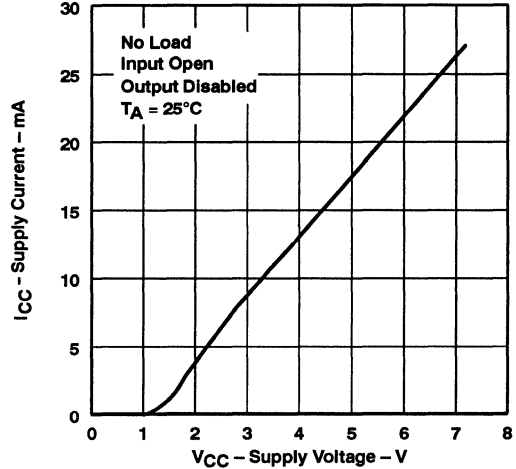
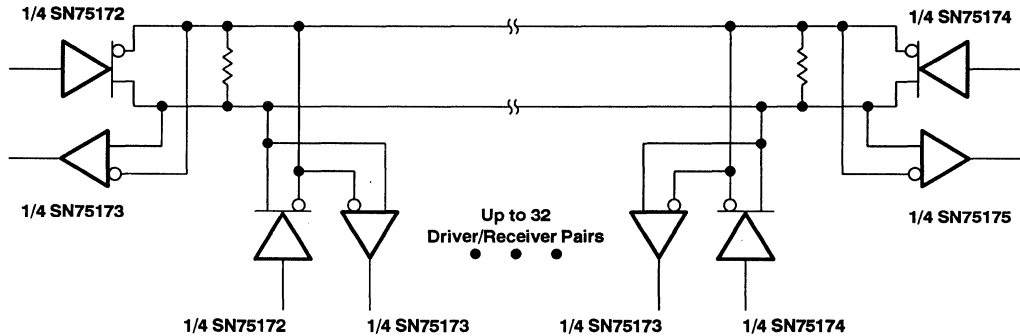


Figure 10

APPLICATION INFORMATION



NOTE: The line length should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

Figure 11



# SN65ALS172A, SN75ALS172A QUAD DIFFERENTIAL LINE DRIVERS

SLLS121B – D3554, AUGUST 1990 – REVISED MARCH 1993

- Meets EIA Standards RS-422-A and RS-485
- Meets CCITT Recommendations V.11 and X.27
- High-Speed Advanced Low-Power Schottky Circuitry
- Designed for 20-MBaud Operation in Both Serial and Parallel Applications
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Low Supply Current Requirements  
55 mA Max
- Wide Positive and Negative Input/Output Bus Voltages Ranges
- Driver Output Capacity . . .  $\pm 60$  mA
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Functionally Interchangeable With SN75172

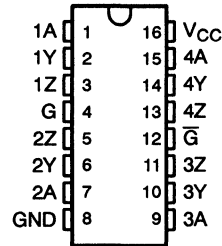
## description

The SN65ALS172A and SN75ALS172A are quad line drivers with 3-state differential outputs. They are designed to meet the requirements of EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27. These devices are optimized for balanced multipoint bus transmission at rates of up to 20-Mbaud. Each driver features wide positive and negative common-mode output voltage ranges making them suitable for party-line applications in noisy environments.

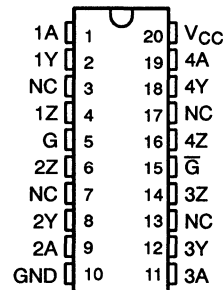
The SN65ALS172A and SN75ALS172A provide positive- and negative-current limiting and thermal shutdown for protection from line fault conditions on the transmission bus line. Shutdown occurs at a junction temperature of approximately 150°C.

The SN65ALS172A is characterized for operation from -40°C to 85°C and the SN75ALS172A is characterized for operation from 0°C to 70°C.

SN75ALS172A . . . N PACKAGE  
(TOP VIEW)



DW PACKAGE  
(TOP VIEW)



NC—No internal connection

FUNCTION TABLE  
(each driver)

INPUT A	ENABLES		OUTPUTS	
	G	Ḡ	Y	Z
H	H	X	H	L
L	H	X	L	H
H	X	L	H	L
L	X	L	L	H
X	L	H	Z	Z

H = high level, L = low level, X = irrelevant, Z = high impedance (off)

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

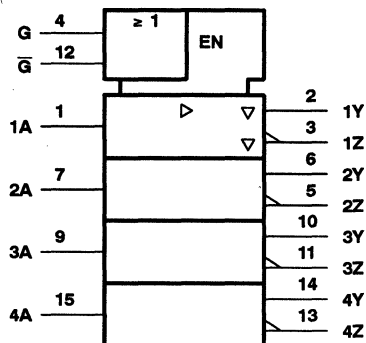
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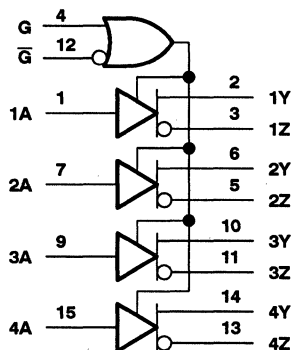
# SN65ALS172A, SN75ALS172A QUAD DIFFERENTIAL LINE DRIVERS

SLLS121B – D3554, AUGUST 1990 – REVISED MARCH 1993

## logic symbol†



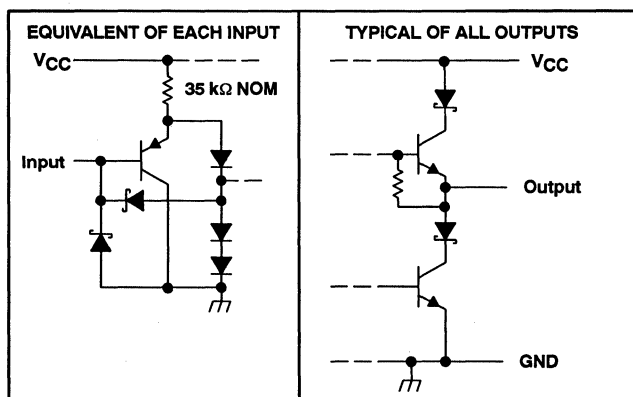
## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the N package.

## schematics of inputs and outputs



# SN65ALS172A, SN75ALS172A QUAD DIFFERENTIAL LINE DRIVERS

SLLS121B – D3554, AUGUST 1990 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	7 V
Output voltage range, $V_O$	-9 V to 14 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65ALS172A	-40°C to 85°C
SN75ALS172A	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
DW	1125 mW	9 mW/°C	720 mW	585 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
Common-mode output voltage, $V_{OC}$			+12 -7	V
High-level output current, $I_{OH}$			-60	mA
Low-level output current, $I_{OL}$			60	mA
Operating free-air temperature, $T_A$	SN65ALS172A	-40	85	°C
	SN75ALS172A	0	70	



# SN65ALS172A, SN75ALS172A QUAD DIFFERENTIAL LINE DRIVERS

SLLS121B – D3554, AUGUST 1990 – REVISED MARCH 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$				-1.5	V
$V_O$	Output voltage	$I_O = 0$		0		6	V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5		6	V
$ V_{OD2} $	Differential output voltage	$V_{CC} = 5 \text{ V},$ $R_L = 100 \Omega$	See Figure 1	$1/2 V_{OD1}$ or $2^\ddagger$			V
		$R_L = 54 \Omega$		1.5	2.5	5	
$ V_{OD3} $	Differential output voltage	See Note 2		1.5		5	V
$\Delta V_{OD} $	Change in magnitude of differential output voltage <sup>‡</sup>	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1				$\pm 0.2$	V
$V_{OC}$	Common-mode output voltage <sup>§</sup>					+3 -1	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage <sup>‡</sup>					$\pm 0.2$	V
$I_O$	Output current with power off	$V_{CC} = 0,$	$V_O = -7 \text{ V to } 12 \text{ V}$			$\pm 100$	$\mu\text{A}$
$I_{OZ}$	High-impedance-state output current	$V_O = -7 \text{ V to } 12 \text{ V}$				$\pm 100$	$\mu\text{A}$
$I_{IH}$	High-level input current	$V_I = 2.7 \text{ V}$				20	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$				-100	$\mu\text{A}$
$I_{OS}$	Short-circuit output current	$V_O = -7 \text{ V to } 12 \text{ V}$				$\pm 250$	mA
$I_{CC}$	Supply current (all drivers)	No load	Outputs enabled		36	55	mA
			Outputs disabled		15	30	

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$  respectively, that occur when the input is changed from a high level to a low level.

§ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

¶ The minimum  $V_{OD2}$  with a 100- $\Omega$  load is either  $1/2 V_{OD1}$  or 2 V, whichever is greater.

NOTE 2: See EIA Standard RS-485, Figure 3-5, Test Termination Measurement 2.

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature,  $C_L = 50 \text{ pF}$**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$t_{dD}$	Differential-output delay time	$R_L = 54 \Omega,$	See Figure 2	9	15	22	ns
$t_{pZH}$	Output enable time to high level	$R_L = 110 \Omega,$	See Figure 3	30	45	70	ns
$t_{pZL}$	Output enable time to low level	$R_L = 110 \Omega,$	See Figure 4	25	40	65	ns
$t_{pHZ}$	Output disable time from high level	$R_L = 110 \Omega,$	See Figure 3	10	20	35	ns
$t_{pLZ}$	Output disable time from low level	$R_L = 110 \Omega,$	See Figure 4	10	30	45	ns

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .



# SN65ALS172A, SN75ALS172A QUAD DIFFERENTIAL LINE DRIVERS

SLLS121B – D3554, AUGUST 1990 – REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

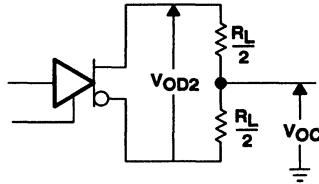
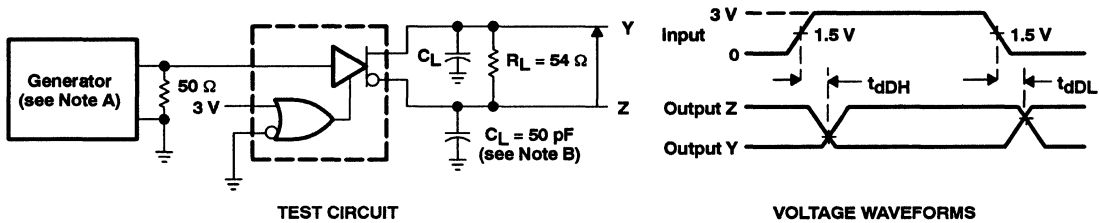


Figure 1. Differential and Common-Mode Output Voltages



- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz,  $Z_O = 50 \Omega$ , duty cycle = 50%,  $t_f \leq 5$  ns,  $t_r \leq 5$  ns.  
 B.  $C_L$  includes probe and stray capacitance.

Figure 2. Differential Output Test Circuit and Voltage Waveforms



# SN65ALS172A, SN75ALS172A QUAD DIFFERENTIAL LINE DRIVERS

SLLS121B – D3554, AUGUST 1990 – REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

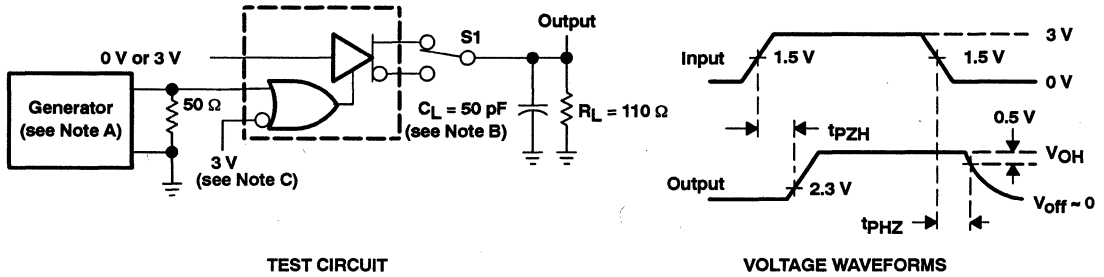


Figure 3. Test Circuit and Voltage Waveforms,  $t_{pZH}$  and  $t_{pHZ}$

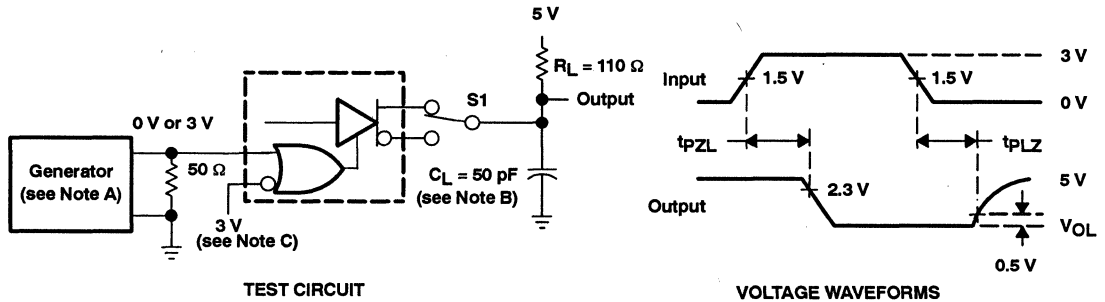


Figure 4. Test Circuit and Voltage Waveforms,  $t_{pZL}$  and  $t_{pLZ}$

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz,  $Z_O = 50 \Omega$ , duty cycle = 50%,  $t_r \leq 5$  ns,  $t_f \leq 5$  ns.  
 B.  $C_L$  includes probe and stray capacitance.  
 C. To test the active-low enable  $\overline{G}$ , ground G and apply an inverted input waveform to  $\overline{G}$ .

# SN55173, SN65173, SN75173 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS144A – D2600, OCTOBER 1980 – REVISED FEBRUARY 1993

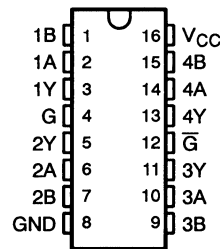
- Meets EIA Standards RS-422-A, RS-423-A, and RS-485
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- Designed for Multipoint Bus Transmission on Long Bus Lines in Noisy Environments
- 3-State Outputs
- Common-Mode Input Voltage Range of –12 V to 12 V
- Input Sensitivity . . . ±200 mV
- Input Hysteresis . . . 50 mV Typ
- High Input Impedance . . . 12 kΩ Min
- Operates From Single 5-V Supply
- Low Power Requirements
- Plug In Replacement for AM26LS32

## description

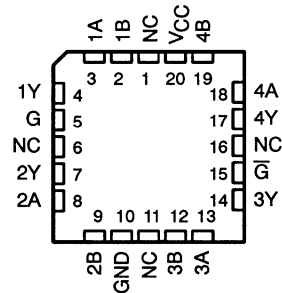
The SN55173, SN65173, and SN75173 are monolithic quad differential line receivers with 3-state outputs. They are designed to meet the requirements of EIA Standards RS-422-A, RS-423-A, RS-485, and several CCITT recommendations. The devices are optimized for balanced multipoint bus transmission at rates up to 10 megabits per second. The four receivers share two ORed enable inputs, one active when high, the other active when low. The '173 devices feature high input impedance, input hysteresis for increased noise immunity, and input sensitivity of ±200 mV over a common-mode input voltage range of –12 to 12 V. Fail-safe design ensures that if the inputs are open circuited, the outputs are always high. The SN65173 and SN75173 are designed for optimum performance when used with the SN75172 or SN75174 quad differential line drivers, and the SN55173 is designed for optimum performance when used with the SN55172 or SN55174 quad differential line drivers.

The SN55173 is characterized over the full military temperature range of –55°C to 125°C. The SN65173 is characterized for operation from –40°C to 85°C. The SN75173 is characterized for operation from 0°C to 70°C.

D, J, OR N PACKAGE  
(TOP VIEW)



SN55173 . . . FK PACKAGE  
(TOP VIEW)

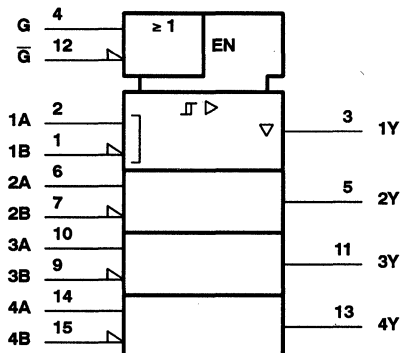


NC—No internal connection

# SN55173, SN65173, SN75173 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS144A - D2600, OCTOBER 1980 - REVISED FEBRUARY 1993

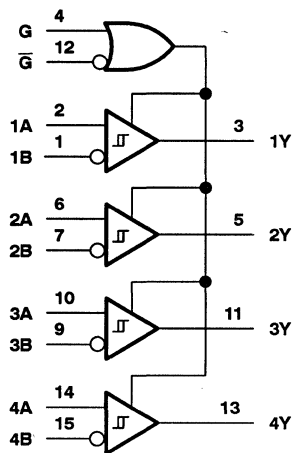
## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, and N packages.

## logic diagram (positive logic)

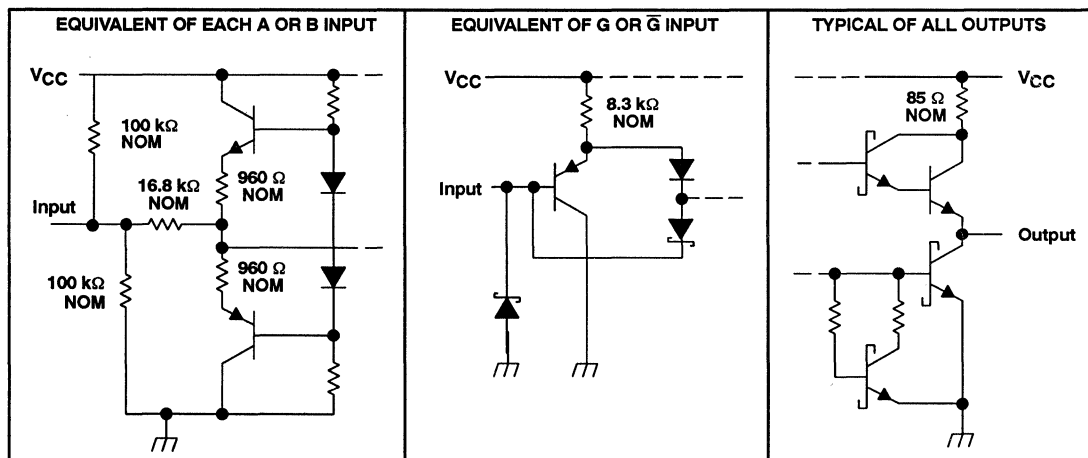


FUNCTION TABLE  
(each receiver)

DIFFERENTIAL A - B	ENABLES G $\bar{G}$	OUTPUT Y
$V_{ID} \geq 0.2 V$	H X X L	H H
$-0.2 V < V_{ID} < 0.2 V$	H X X L	? ?
$V_{ID} \leq -0.2 V$	H X X L	L L
X	L H	Z

H = high level, L = low level, ? = indeterminate,  
X = irrelevant, Z = high impedance (off)

## schematics of inputs and outputs



TEXAS  
INSTRUMENTS

# SN55173, SN65173, SN75173 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS144A – D2600, OCTOBER 1980 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, A or B inputs	$\pm 25$ V
Differential input voltage (see Note 2)	$\pm 25$ V
Enable input voltage	7 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ :	
SN55173	–55°C to 125°C
SN65173	–40°C to 85°C
SN75173	0°C to 70°C
Storage temperature range	–65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$		$T_A = 70^\circ\text{C}$		$T_A = 85^\circ\text{C}$		$T_A = 125^\circ\text{C}$	
	POWER RATING	DERATING FACTOR	POWER RATING	POWER RATING	POWER RATING	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	—	—	—	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW	—	—	—
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW	—	—	—
N	1150 mW	9.2 mW/°C	736 mW	598 mW	—	—	—	—

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	SN55173	4.5	5	5.5	V
	SN65173, SN75173	4.75	5	5.25	
Common-mode input voltage, $V_{IC}$				$\pm 12$	V
Differential input voltage, $V_{ID}$				$\pm 12$	V
High-level enable-input voltage, $V_{IH}$		2			V
Low-level enable-input voltage, $V_{IL}$			0.8		V
High-level output current, $I_{OH}$			–400		$\mu\text{A}$
Low-level output current, $I_{OL}$			16		mA
Operating free-air temperature, $T_A$	SN55173	–55		125	°C
	SN65173	–40		85	
	SN75173	0		70	



# SN55173, SN65173, SN75173 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS144A – D2600, OCTOBER 1980 – REVISED FEBRUARY 1993

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going input threshold voltage	$V_O = 2.7$ V,	$I_O = -0.4$ mA			0.2	V
$V_{T-}$	Negative-going input threshold voltage	$V_O = 0.5$ V,	$I_O = 16$ mA	-0.2‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	See Figure 4			50		mV
$V_{IK}$	Enable-input clamp voltage	$I_I = -18$ mA				-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 200$ mV,	$I_{OH} = -400$ $\mu$ A	SN55173	2.5		V
				SN65173, SN75173	2.7		
$V_{OL}$	Low-level output voltage	$V_{ID} = -200$ mV,	See Figure 1	$I_{OL} = 8$ mA		0.45	V
				$I_{OL} = 16$ mA		0.5	
$I_{OZ}$	High-impedance-state output current	$V_O = 0.4$ V to 2.4 V				$\pm 20$	$\mu$ A
$I_I$	Line input current	Other input at 0 V,	See Note 3	$V_I = 12$ V		1	mA
				$V_I = -7$ V		-0.8	
$I_{IH}$	High-level enable-input current	$V_{IH} = 2.7$ V				20	$\mu$ A
$I_{IL}$	Low-level enable-input current	$V_{IL} = 0.4$ V				-100	$\mu$ A
$r_i$	Input resistance				12		k $\Omega$
$I_{OS}$	Short-circuit output current§			-15		-85	mA
$I_{CC}$	Supply current	Outputs disabled				70	mA

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold voltage levels only.

§ Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

NOTE 3: Refer to EIA Standards RS-422A and RS423-A for exact conditions.

## switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -1.5$ V to 1.5 V, $C_L = 15$ pF,			20	35	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output	See Figure 1			22	35	ns
$t_{PZH}$	Output enable time to high level	$C_L = 15$ pF,	See Figure 2		17	22	ns
$t_{PZL}$	Output enable time to low level	$C_L = 15$ pF,	See Figure 3		20	25	ns
$t_{PHZ}$	Output disable time from high level	$C_L = 5$ pF,	See Figure 2		21	30	ns
$t_{PLZ}$	Output disable time from low level	$C_L = 5$ pF,	See Figure 3		30	40	ns

TEXAS  
INSTRUMENTS

# SN55173, SN65173, SN75173 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS144A - D2600, OCTOBER 1980 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

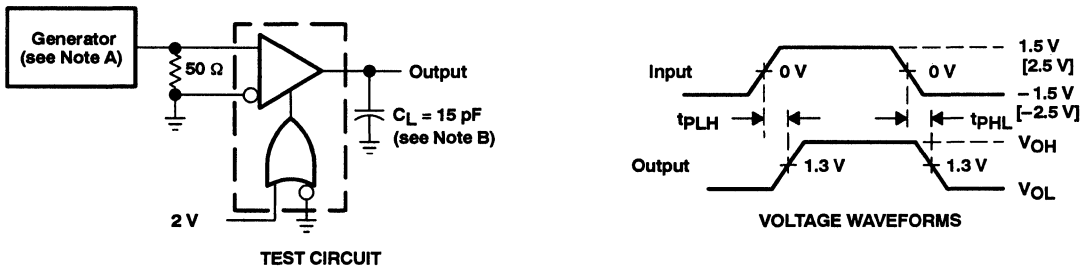


Figure 1.  $t_{pLH}$ ,  $t_{pHL}$  Test Circuit and Voltage Waveforms

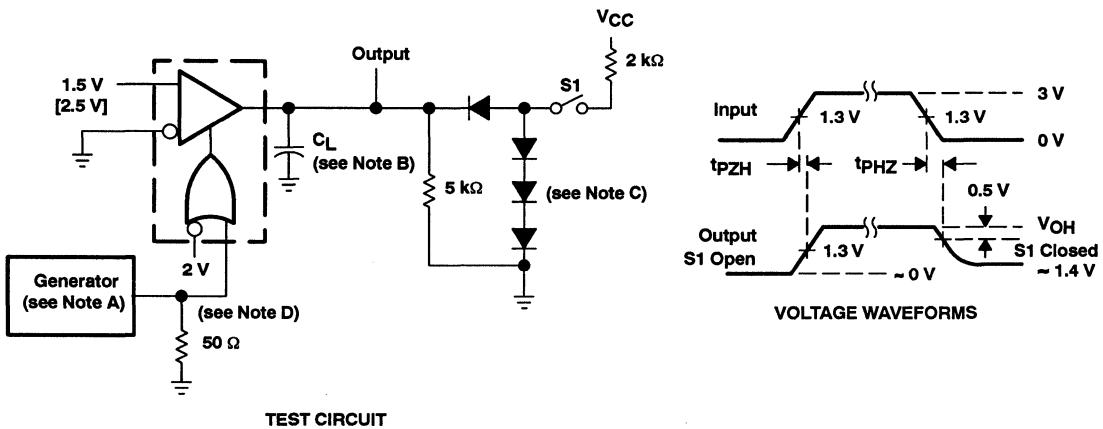


Figure 2.  $t_{pZH}$ ,  $t_{pZH}$  Test Circuit and Voltage Waveforms

[ ] represent voltages on the SN55173 only.

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

C. All diodes are 1N916 or equivalent.

D. To test the active-low enable  $\bar{G}$ , ground G and apply an inverted input waveform to  $\bar{G}$ .

# SN55173, SN65173, SN75173 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS144A - D2600, OCTOBER 1980 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

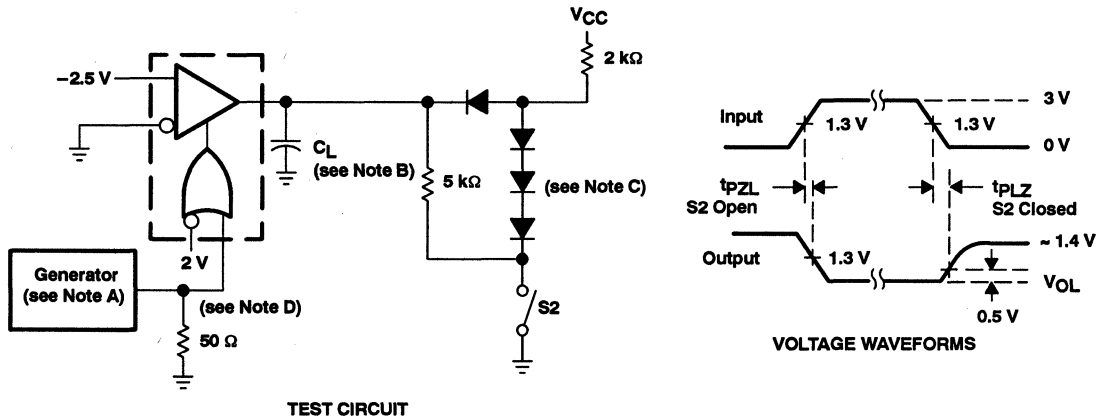


Figure 3.  $t_{pZL}$ ,  $t_{pLZ}$  Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_O = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N916 or equivalent.  
 D. To test the active-low enable  $\bar{G}$ , ground G and apply an inverted input waveform to  $\bar{G}$ .

## TYPICAL CHARACTERISTICS

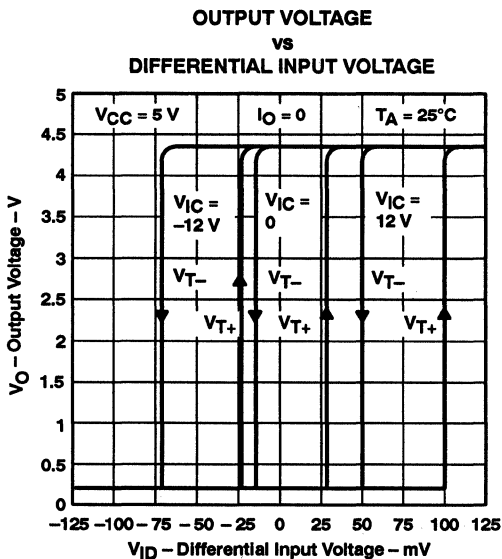


Figure 4

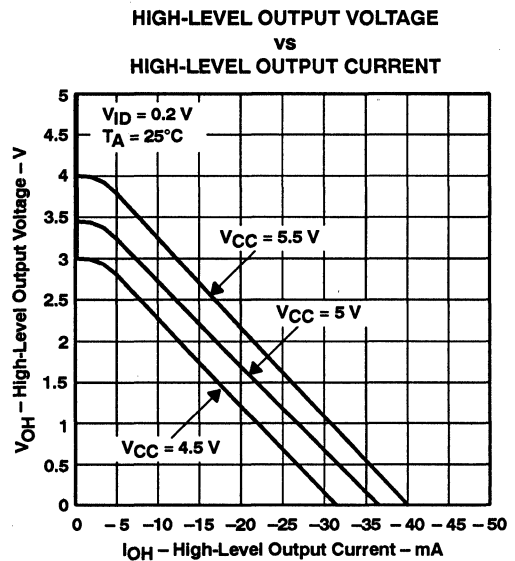


Figure 5

TYPICAL CHARACTERISTICS

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

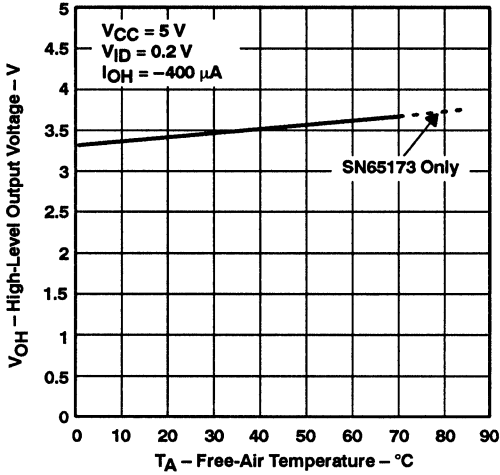


Figure 6

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

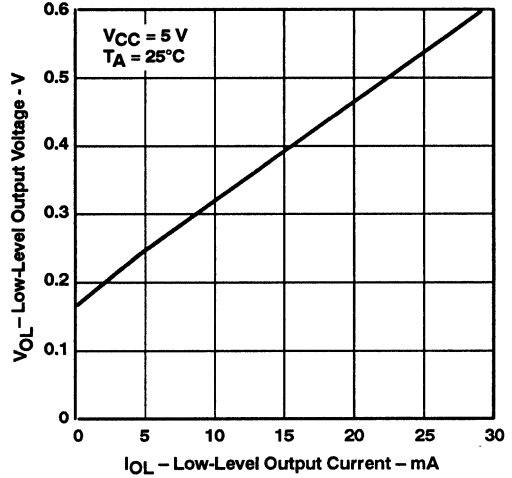


Figure 7

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

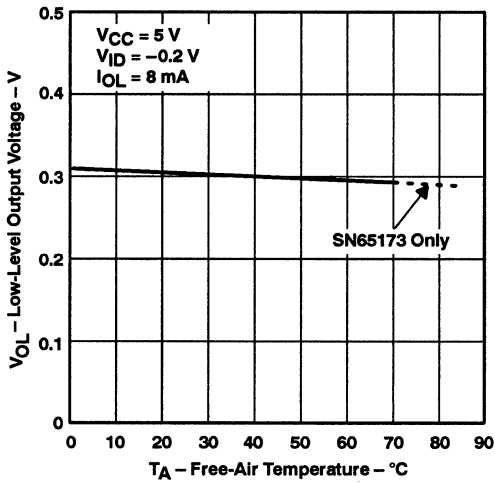


Figure 8

OUTPUT VOLTAGE  
 vs  
 ENABLE G VOLTAGE

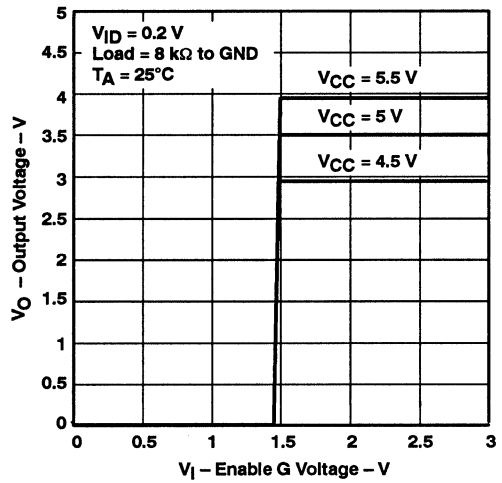


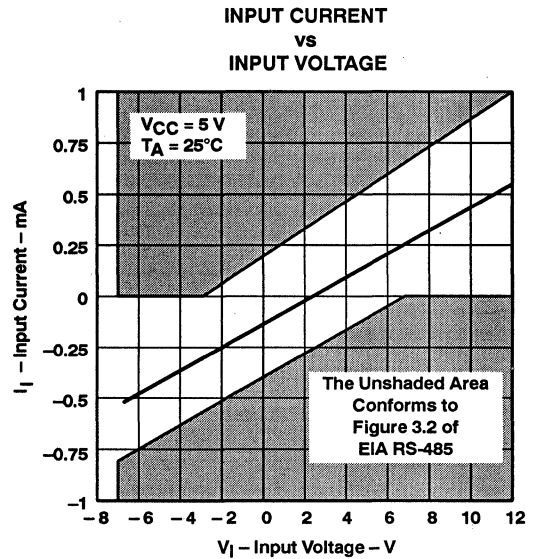
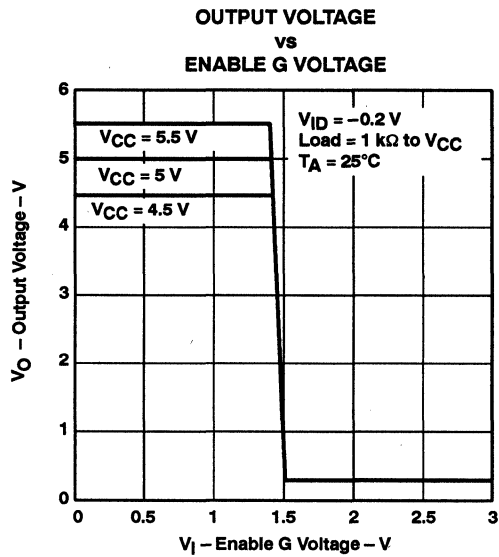
Figure 9



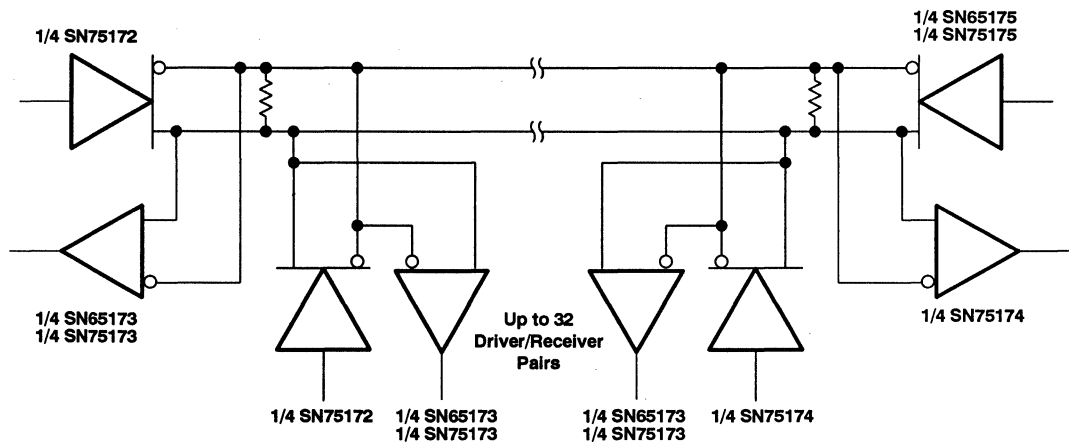
# SN55173, SN65173, SN75173 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS144A - D2600, OCTOBER 1980 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS



## APPLICATION INFORMATION



NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

**Figure 12. Typical Application Circuit**

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# SN75ALS173 QUAD DIFFERENTIAL LINE RECEIVER

SLLS132B – D3886, SEPTEMBER 1991 – REVISED JANUARY 1993

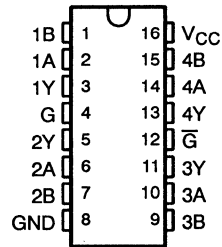
- Meets EIA Standards RS-422-A, RS-423-A, and RS-485
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- Designed for Multipoint Bus Transmission on Log Bus Lines in Noisy Environments
- 3-State Outputs
- Common-Mode Input Voltage Range of -12 V to 12 V
- Input Sensitivity . . .  $\pm 200$  mV
- Input Hysteresis . . . 50 mV Typ
- High Input Impedance . . . 12 k $\Omega$  Min
- Operates From Single 5-V Supply
- Low Supply Current Requirement  
27 mA Max

## description

The SN75ALS173 is a monolithic quad differential line receiver with 3-state outputs. It is designed to meet the requirements of EIA Standards RS-422-A, RS-423-A, and RS-485 and several CCITT recommendations. Advanced low-power Schottky technology provides high speed without the usual power penalty. The four receivers have an ORed pair of enables in common. Either G being high or  $\bar{G}$  being low enables all of the receivers. The device features high input impedance, input hysteresis for increased noise immunity, and input sensitivity of  $\pm 200$  mV over a common-mode input voltage range of -12 V to 12 V.

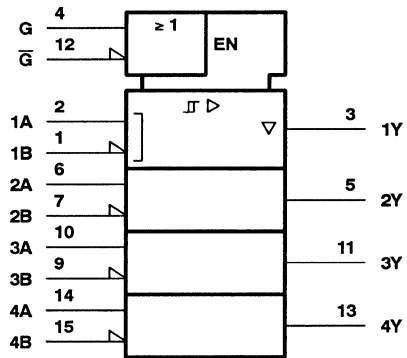
The SN75ALS173 is characterized for operation from 0°C to 70°C.

N OR NS† PACKAGE  
(TOP VIEW)



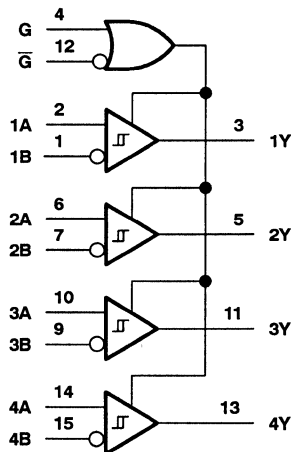
† The NS package is only available left-end taped and reeled (order device SN75ALS173 NSLE).

## logic symbol‡



‡ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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2-557

# SN75ALS173 QUAD DIFFERENTIAL LINE RECEIVER

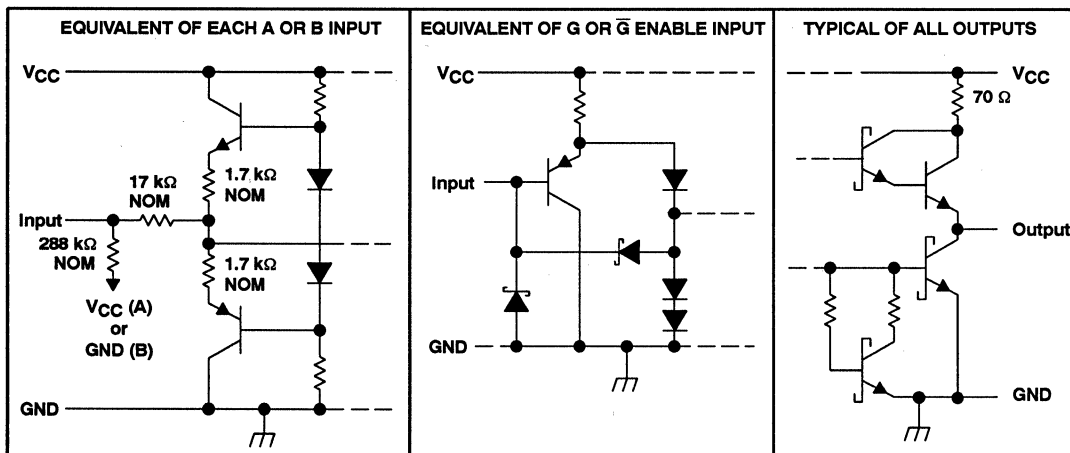
SLLS132B – D3886, SEPTEMBER 1991 – REVISED JANUARY 1993

**FUNCTION TABLE**  
(each receiver)

DIFFERENTIAL INPUTS A – B	ENABLES		OUTPUT Y
	G	$\bar{G}$	
$V_{ID} \geq 0.2 V$	H X	X L	H H
$-0.2 V < V_{ID} < 0.2 V$	H X	X L	? ?
$V_{ID} \leq -0.2 V$	H X	X L	L L
X	L	H	Z
Open Circuit	H X	X L	H H

H = high level, L = low level, ? = indeterminate,  
X = irrelevant, Z = high impedance (off)

## schematics of inputs and outputs



# SN75ALS173 QUAD DIFFERENTIAL LINE RECEIVER

SLLS132B – D3886, SEPTEMBER 1991 – REVISED JANUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, A or B inputs, $V_I$	$\pm 14$ V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 14$ V
Enable input voltage	7 V
Low-level output current, $I_{OL}$	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	$0^\circ\text{C}$ to $70^\circ\text{C}$
Storage temperature range	$-65^\circ\text{C}$ to $150^\circ\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	$260^\circ\text{C}$

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
N	1150 mW	9.2 mW/ $^\circ\text{C}$	736 mW
NS	625 mW	5.0 mW/ $^\circ\text{C}$	400 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$				$\pm 12$	V
Differential input voltage, $V_{ID}$				$\pm 12$	V
High-level input voltage, $V_{IH}$	G, $\bar{G}$	2			V
Low-level input voltage, $V_{IL}$	G, $\bar{G}$			0.8	V
High-level output current, $I_{OH}$				-400	$\mu\text{A}$
Low-level output current, $I_{OL}$				8	mA
Operating free-air temperature, $T_A$		0		70	$^\circ\text{C}$



# SN75ALS173 QUAD DIFFERENTIAL LINE RECEIVER

SLLS132B – D3886, SEPTEMBER 1991 – REVISED JANUARY 1993

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage and operating free-air temperature (unless otherwise noted) (see Note 3)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going threshold voltage				200	mV
$V_{T-}$ Negative-going threshold voltage		-200‡			mV
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )			50		mV
$V_{IK}$ Input clamp voltage	$I_I = -18$ mA			-1.5	V
$V_{OH}$ High-level output voltage	$V_{ID} = 200$ mV, $I_{OH} = -400$ $\mu$ A, See Figure 1		2.7		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -200$ mV, $I_{OL} = 8$ mA, See Figure 1			0.45	V
$I_{OZ}$ High-impedance-state output current	$V_O = 0.4$ V to 2.4 V			$\pm 20$	$\mu$ A
$I_I$ Line input current	Other input at 0 V			1	mA
	$V_I = 12$ V			-0.8	
$I_{IH}$ High-level input current	$V_{IH} = 2.7$ V			20	$\mu$ A
$I_{IL}$ Low-level input current	$V_{IL} = 0.4$ V			-100	$\mu$ A
$r_i$ Input resistance			12		k $\Omega$
$I_{OS}$ Short-circuit output current	See Note 4	-15		-85	mA
$I_{CC}$ Supply current (total package)	No load, Outputs enabled		16	24	
	No load, Outputs disabled		18	27	

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold voltage levels only.

NOTES: 3. Refer to EIA Standards RS-485 for exact conditions.

4. Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

switching characteristics,  $V_{CC} = 5$  V,  $C_L = 15$  pF,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PHL}$ Propagation delay time, high-to-low-level output	$V_{ID} = -2.5$ V to 2.5 V, See Figure 2	9	18	27	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	See Figure 2	9	18	27	ns
$t_{PZH}$ Output enable time to high level	See Figure 3	4	12	18	ns
$t_{PZL}$ Output enable time to low level	See Figure 4	6	13	21	ns
$t_{PHZ}$ Output disable time from high level	See Figure 3	10	21	27	ns
$t_{PLZ}$ Output disable time from low level	See Figure 4	8	15	25	ns

## PARAMETER MEASUREMENT INFORMATION

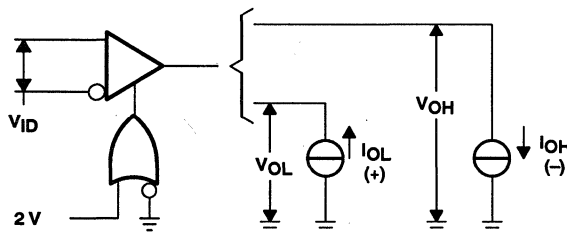


Figure 1.  $V_{OH}$ ,  $V_{OL}$

PARAMETER MEASUREMENT INFORMATION

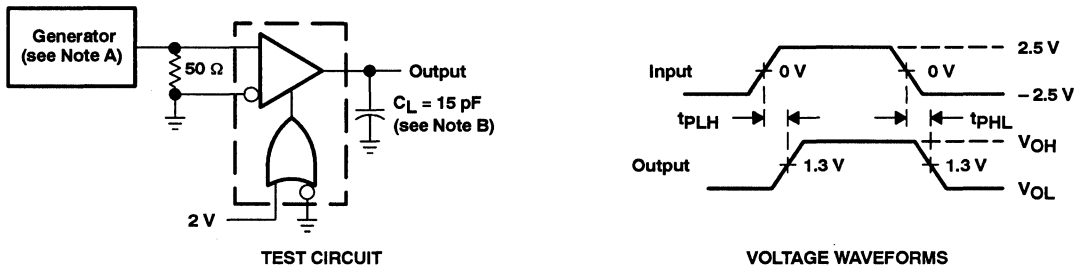


Figure 2. Test Circuit and Voltage Waveforms

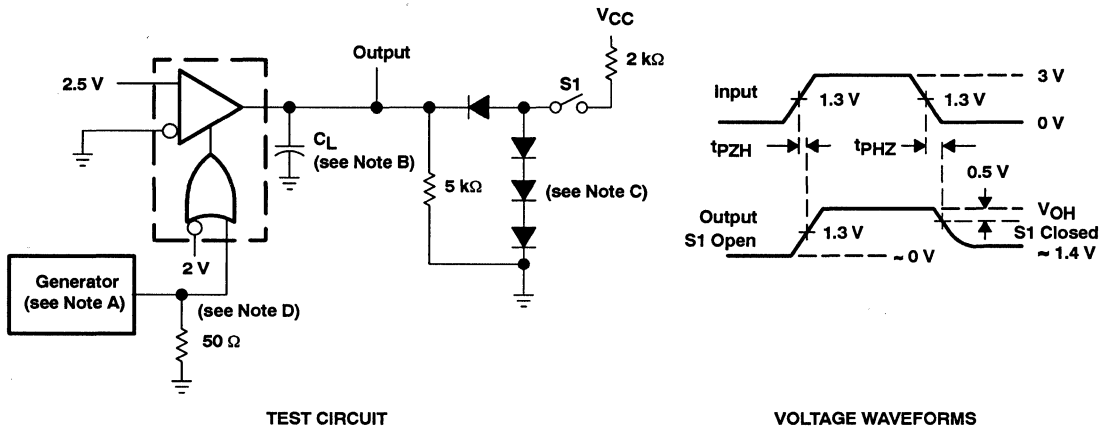


Figure 3. Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N916 or equivalent.  
 D. To test the active-low enable  $\overline{G}$ , ground G and apply an inverted input waveform to  $\overline{G}$ .

# SN75ALS173 QUAD DIFFERENTIAL LINE RECEIVER

SLLS132B – D3886, SEPTEMBER 1991 – REVISED JANUARY 1993

## PARAMETER MEASUREMENT INFORMATION

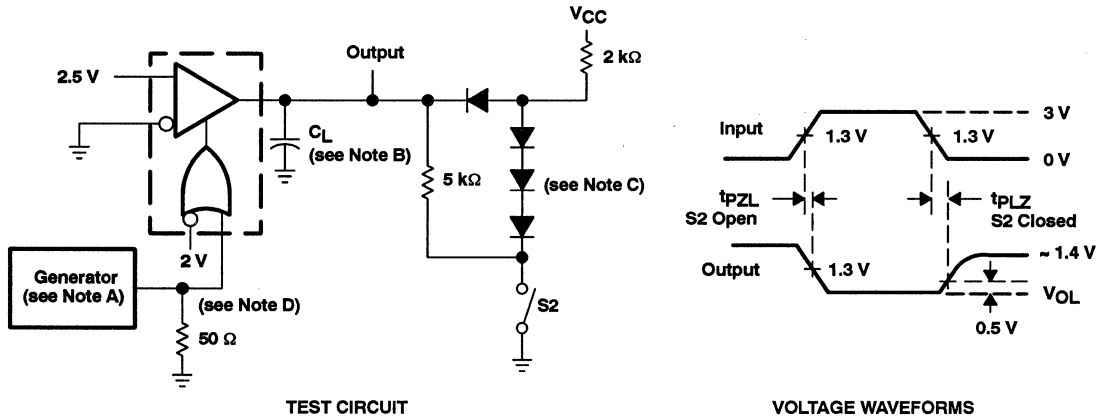


Figure 4. Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .
- B.  $C_L$  includes probe and jig capacitance.
- C. All diodes are 1N916 or equivalent.
- D. To test the active-low enable  $\overline{G}$ , ground G and apply an inverted input waveform to  $\overline{G}$ .

# SN75174 QUAD DIFFERENTIAL LINE DRIVER

SLLS039A – D2601, OCTOBER 1980 – REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- 3-State Outputs
- Common-Mode Output Voltage Range of -7 V to 12 V
- Active-High Enable
- Thermal Shutdown Protection
- Positive- and Negative-Current Limiting
- Operates From Single 5-V Supply
- Low Power Requirements
- Functionally Interchangeable With MC3487

## description

The SN75174 is a monolithic quad differential line driver with 3-state outputs. It is designed to meet the requirements of EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27. The device is optimized for balanced multipoint bus transmission at rates up to 4 megabaud. Each driver features wide positive and negative common-mode output voltage ranges making it suitable for party-line applications in noisy environments.

The SN75174 provides positive- and negative-current limiting and thermal shutdown for protection from line fault conditions on the transmission bus line. Shutdown occurs at a junction temperature of approximately 150°C. This device offers optimum performance when used with the SN75173 or SN75175 quadruple differential line receivers.

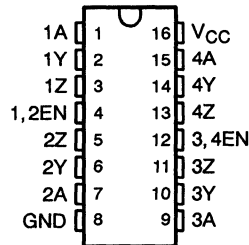
The SN75174 is characterized for operation from 0°C to 70°C.

**FUNCTION TABLE**  
(each driver)

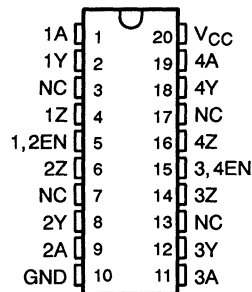
INPUT	ENABLE	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	Z	Z

H = TTL high level, X = irrelevant,  
L = TTL low level,  
Z = high impedance (off)

**N PACKAGE**  
(TOP VIEW)

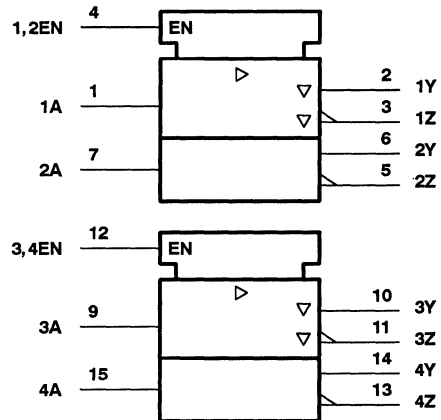


**DW PACKAGE**  
(TOP VIEW)



NC – No internal connection

## logic symbol



† This symbol is in accordance with ANSI/IEEE Std 91-984 and IEC Publication 617-12.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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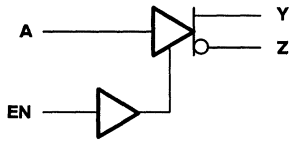
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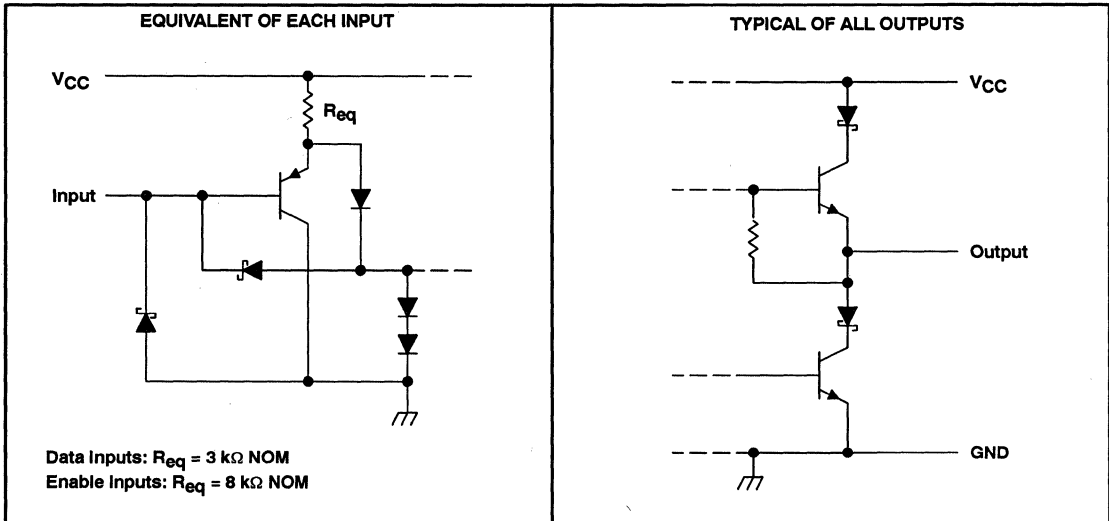
# SN75174 QUAD DIFFERENTIAL LINE DRIVER

SLLS039A - D2601, OCTOBER 1980 - REVISED FEBRUARY 1983

logic diagram, each driver (positive logic)



schematics of inputs and outputs



absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	T <sub>A</sub> = 25°C POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
N	1150 mW	9.2 mW/°C	736 mW

# SN75174 QUAD DIFFERENTIAL LINE DRIVER

SLLS039A – D2601, OCTOBER 1980 – REVISED FEBRUARY 1993

## recommended, operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
Common-mode output voltage, $V_{OC}$			-7 to 12	V
High-level output current, $I_{OH}$			-60	mA
Low-level output current, $I_{OL}$			60	mA
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$I_I = -18$ mA			-1.5	V
$V_{OH}$ High-level output voltage	$V_{IH} = 2$ V, $I_{OH} = -33$ mA $V_{IL} = 0.8$ V,		3.7		V
$V_{OL}$ Low-level output voltage	$V_{IH} = 2$ V, $I_{OL} = 33$ mA $V_{IL} = 0.8$ V,		1.1		V
$V_O$ Output voltage	$I_O = 0$	0		6	V
$ V_{OD1} $ Differential output voltage	$I_O = 0$	1.5	6	6	V
$ V_{OD2} $ Differential output voltage	$R_L = 100 \Omega$ ,      See Figure 1			$1/2 V_{OD1}$ or $2^{\S}$	V
	$R_L = 54 \Omega$ ,      See Figure 1	1.5	2.5	5	V
$V_{OD3}$ Differential output voltage	See Note 2	1.5		5	V
$\Delta V_{OD} $ Change in magnitude of differential output voltage‡				$\pm 0.2$	V
$V_{OC}$ Common-mode output voltage§	$R_L = 54 \Omega$ or $100 \Omega$ ,      See Figure 1			+3 -1	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage‡				$\pm 0.2$	V
$I_O$ Output current with power off	$V_{CC} = 0$ , $V_O = -7$ V to $12$ V			$\pm 100$	$\mu$ A
$I_{OZ}$ High-impedance-state output current	$V_O = -7$ V to $12$ V			$\pm 100$	$\mu$ A
$I_{IH}$ High-level input current	$V_I = 2.7$ V			20	$\mu$ A
$I_{IL}$ Low-level input current	$V_I = 0.5$ V			-360	$\mu$ A
$I_{OS}$ Short-circuit output current	$V_O = -7$ V			-180	mA
	$V_O = V_{CC}$			180	
	$V_O = 12$ V			500	
$I_{CC}$ Supply current (all drivers)	No load	Outputs enabled		38	mA
		Outputs disabled		18	

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

§ The minimum  $V_{OD2}$  with a  $100\text{-}\Omega$  load is either  $1/2 V_{OD1}$  or  $2$  V, whichever is greater.

NOTE 2: See Figure 3.5 of EIA Standard RS-485.



# SN75174 QUAD DIFFERENTIAL LINE DRIVER

SLLS039A - D2601, OCTOBER 1980 - REVISED FEBRUARY 1993

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{dD}$ Differential-output delay time	$R_L = 54\ \Omega$ , See Figure 2		45	65	ns
$t_{tD}$ Differential-output transition time			80	120	ns
$t_{PZH}$ Output enable time to high level	$R_L = 110\ \Omega$ , See Figure 3		80	120	ns
$t_{PZL}$ Output enable time to low level	$R_L = 110\ \Omega$ , See Figure 4		55	80	ns
$t_{PHZ}$ Output disable time from high level	$R_L = 110\ \Omega$ , See Figure 3		75	115	ns
$t_{PLZ}$ Output disable time from low level	$R_L = 110\ \Omega$ , See Figure 3		18	30	ns

## SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{Oa}, V_{Ob}$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_o$	$V_o$
$ V_{OD2} $	$V_t (R_L = 100\ \Omega)$	$V_t (R_L = 54\ \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$\Delta V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

## PARAMETER MEASUREMENT INFORMATION

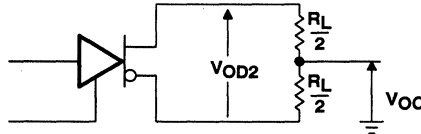
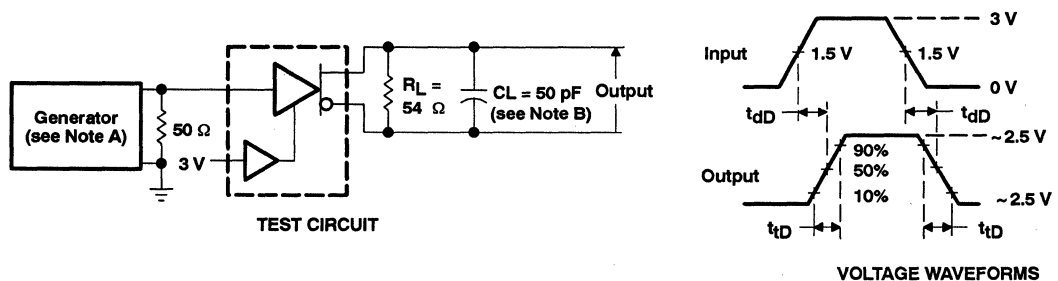


Figure 1. Differential and Common-Mode Output Voltages



- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5\text{ ns}$ ,  $t_f \leq 5\text{ ns}$ ,  $PRR \leq 1\text{ MHz}$ , duty cycle = 50%,  $Z_0 = 50\ \Omega$ .  
B.  $C_L$  includes probe and stray capacitance.

Figure 2. Differential-Output Test Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION

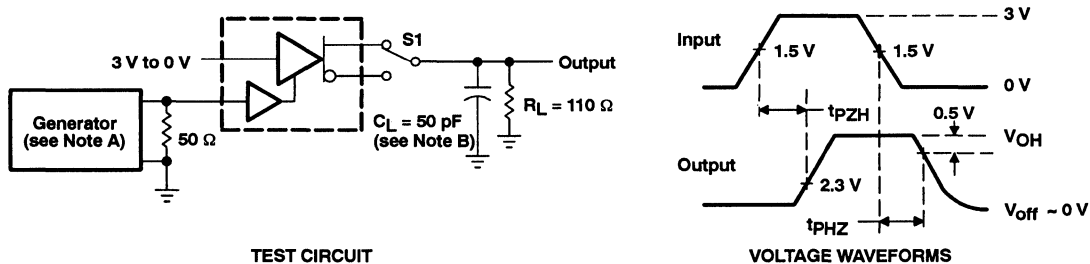


Figure 3. Test Circuit and Voltage Waveforms

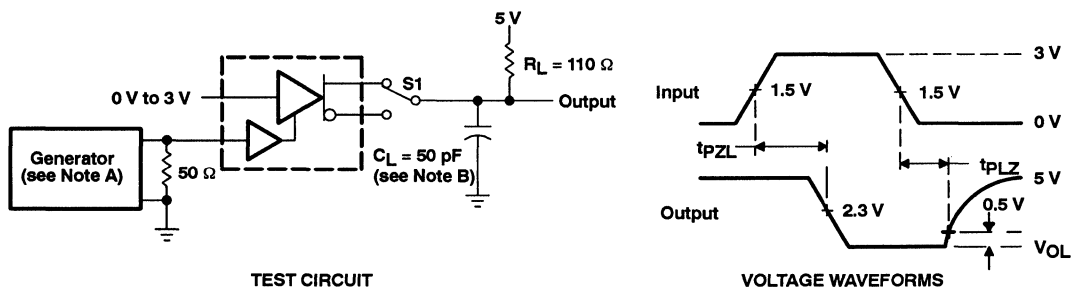


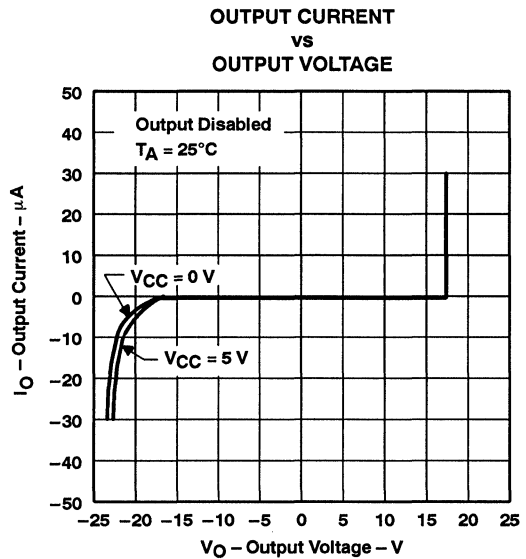
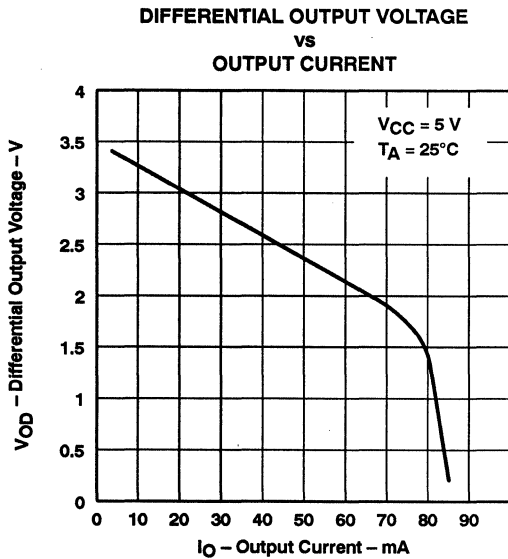
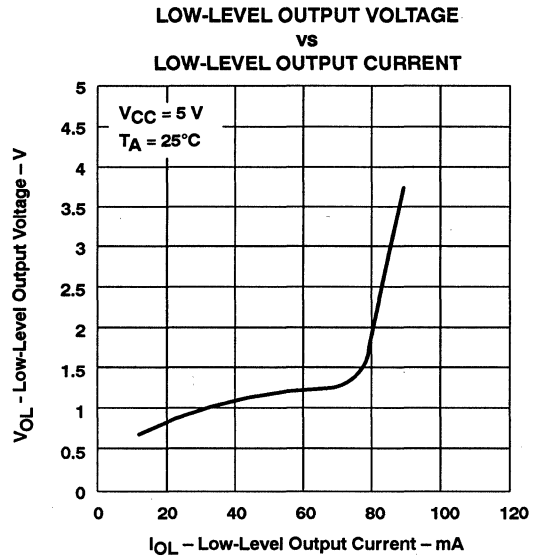
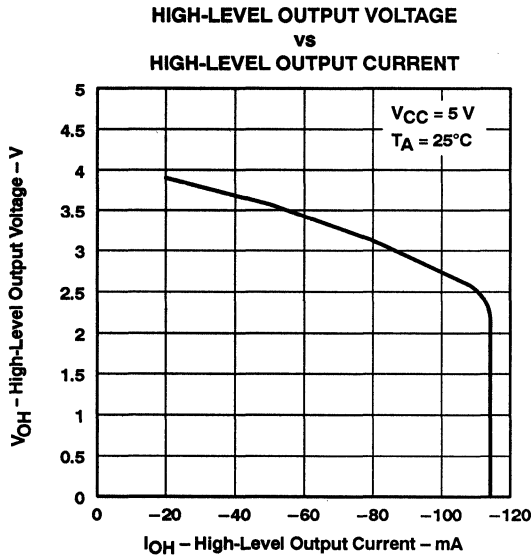
Figure 4. Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle = 50%,  $t_r \leq$  5 ns,  $t_f \leq$  5 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and stray capacitance.

**SN75174**  
**QUAD DIFFERENTIAL LINE DRIVER**

SLLS039A – D2601, OCTOBER 1980 – REVISED FEBRUARY 1993

**TYPICAL CHARACTERISTICS**



TYPICAL CHARACTERISTICS

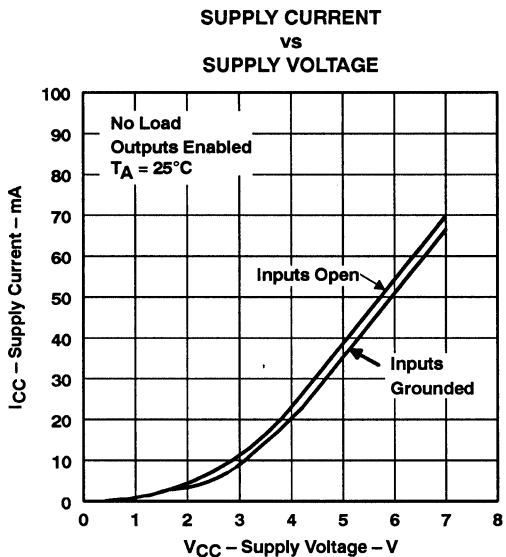


Figure 9

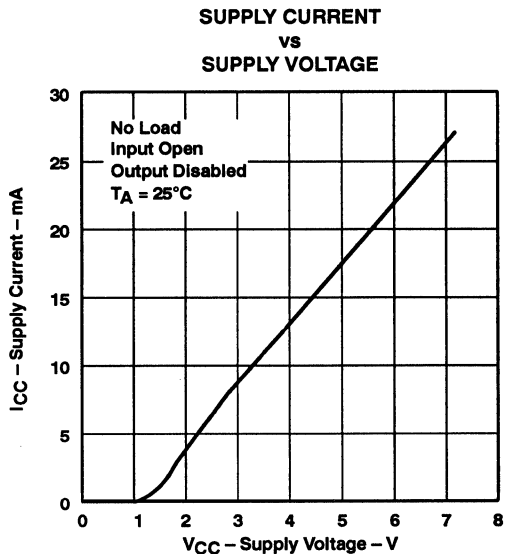
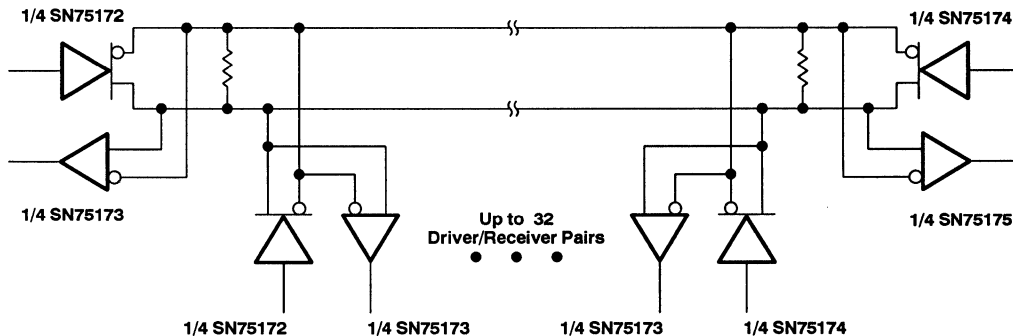


Figure 10

APPLICATION INFORMATION



NOTE: The line length should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

Figure 11. Typical Application Circuit



# SN65ALS174A, SN75ALS174A QUAD DIFFERENTIAL LINE DRIVERS

SLLS122B – D3865, JULY 1991 – REVISED MARCH 1993

- Meets EIA Standards RS-422-A and RS-485
- Meets CCITT Recommendations V.11 and X.27
- High-Speed Advanced Low-Power Schottky Circuitry
- Designed for 20-MBaud Operation in Both Serial and Parallel Applications
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Low Supply Current Requirements  
55 mA Max
- Wide Positive and Negative Input/Output Bus Voltages Ranges
- Driver Output Capacity . . .  $\pm 60$  mA
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Functionally Interchangeable With SN75174

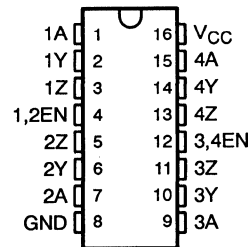
## description

The SN65ALS174A and SN75ALS174A are quad line drivers with 3-state differential outputs. They are designed to meet the requirements of EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27. These devices are optimized for balanced multipoint bus transmission at rates of up to 20 Mbaud. Each driver features wide positive and negative common-mode output voltage ranges making them suitable for party-line applications in noisy environments.

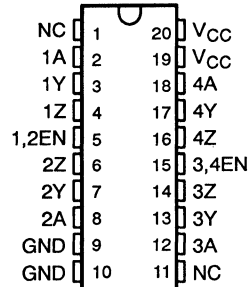
The SN65ALS174A and SN75ALS174A provide positive- and negative-current limiting and thermal shutdown for protection from line fault conditions on the transmission bus line. Shutdown occurs at a junction temperature of approximately 150°C.

The SN65ALS174A is characterized for operation from -40°C to 85°C and the SN75ALS174A is characterized for operation from 0°C to 70°C.

SN75ALS174A . . . N PACKAGE  
(TOP VIEW)



SN65ALS174A, SN75ALS174A . . . DW PACKAGE  
(TOP VIEW)



NC—No internal connection

FUNCTION TABLE  
(each driver)

INPUT A	ENABLES	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	Z	Z

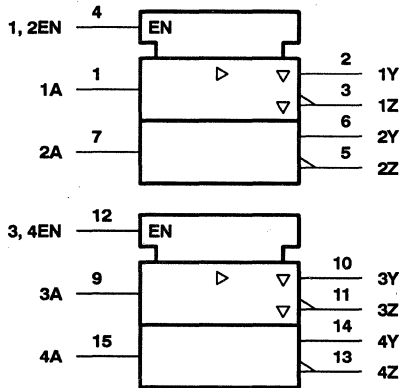
H = high level, L = low level, X = irrelevant,  
Z = high impedance (off)



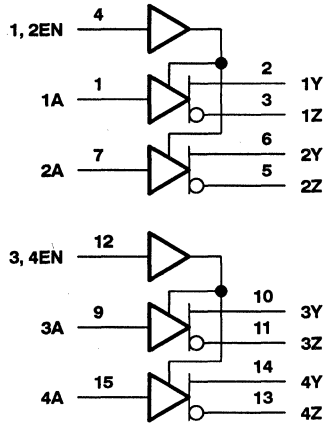
# SN65ALS174A, SN75ALS174A QUAD DIFFERENTIAL LINE DRIVERS

SLLS122B - D3865, JULY 1991 - REVISED MARCH 1993

## logic symbol†



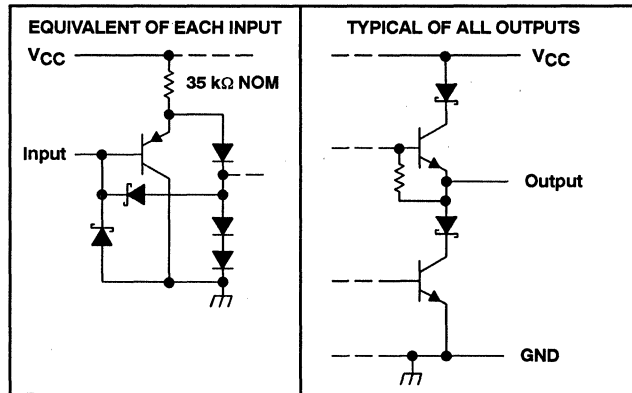
## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the N package.

## schematics of inputs and outputs



# SN65ALS174A, SN75ALS174A QUAD DIFFERENTIAL LINE DRIVERS

SLLS122B – D3865, JULY 1991 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	7 V
Output voltage range, $V_O$	–9 V to 14 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65ALS174A	–40°C to 85°C
SN75ALS174A	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
DW	1125 mW	9 mW/°C	720 mW	585 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
Common-mode output voltage, $V_{OC}$			12 –7	V
High-level output current, $I_{OH}$			–60	mA
Low-level output current, $I_{OL}$			60	mA
Operating free-air temperature, $T_A$			–40	°C
			85	
			0	70

# SN65ALS174A, SN75ALS174A QUAD DIFFERENTIAL LINE DRIVERS

SLLS122B – D3865, JULY 1991 – REVISED MARCH 1993

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$I_I = -18$ mA			-1.5	V
$V_O$ Output voltage	$I_O = 0$	0		6	V
$ V_{OD1} $ Differential output voltage	$I_O = 0$	1.5		6	V
$ V_{OD2} $ Differential output voltage	$R_L = 100 \Omega$	See Figure 1	$1/2 V_{OD1}$ or $2^{\ddagger}$		V
	$R_L = 54 \Omega$		1.5	2.5	
$ V_{OD3} $ Differential output voltage	See Note 2	1.5		5	V
$\Delta V_{OD} $ Change in magnitude of differential output voltage‡				$\pm 0.2$	V
$V_{OC}$ Common-mode output voltage§	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1			$\begin{matrix} 3 \\ -1 \end{matrix}$	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage‡				$\pm 0.2$	V
$I_O$ Output current with power off	$V_{CC} = 0$ , $V_O = -7$ V to $12$ V			$\pm 100$	$\mu$ A
$I_{OZ}$ High-impedance-state output current	$V_O = -7$ V to $12$ V			$\pm 100$	$\mu$ A
$I_{IH}$ High-level input current	$V_I = 2.7$ V			20	$\mu$ A
$I_{IL}$ Low-level input current	$V_I = 0.4$ V			-100	$\mu$ A
$I_{OS}$ Short-circuit output current	$V_O = -7$ V to $12$ V			$\pm 250$	mA
$I_{CC}$ Supply current (all drivers)	No load	Outputs enabled		36	55
		Outputs disabled		16	30

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$  respectively, that occur when the input is changed from a high level to a low level.

§ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

¶ The minimum  $V_{OD2}$  with a  $100\text{-}\Omega$  load is either  $1/2 V_{OD1}$  or  $2$  V whichever is greater.

NOTE 2: See EIA Standard RS-485, Figure 3-5, Test Termination Measurement 2.

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature,  $C_L = 50$  pF**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{dD}$ Differential-output delay time	$R_L = 54 \Omega$ , See Figure 2	9	15	22	ns
$t_{PZH}$ Output enable time to high level	$R_L = 110 \Omega$ , See Figure 3	30	45	70	ns
$t_{PZL}$ Output enable time to low level	$R_L = 110 \Omega$ , See Figure 4	25	40	65	ns
$t_{PHZ}$ Output disable time from high level	$R_L = 110 \Omega$ , See Figure 3	10	20	35	ns
$t_{PLZ}$ Output disable time from low level	$R_L = 110 \Omega$ , See Figure 4	10	30	45	ns

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .



PARAMETER MEASUREMENT INFORMATION

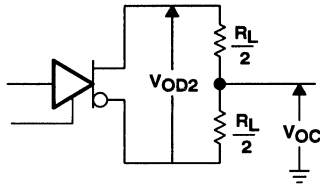
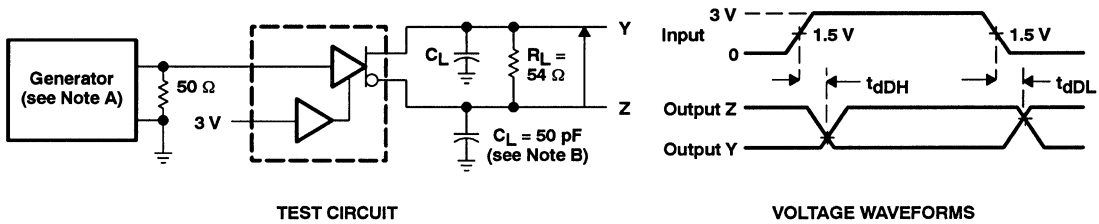


Figure 1. Differential and Common-Mode Output Voltages



NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz,  $Z_O = 50 \Omega$ , duty cycle = 50%,  $t_f \leq 5$  ns,  $t_r \leq 5$  ns.  
 B.  $C_L$  includes probe and stray capacitance.

Figure 2. Differential-Output Test Circuit and Voltage Waveforms Delay and Transition Times

# SN65ALS174A, SN75ALS174A QUAD DIFFERENTIAL LINE DRIVERS

SLLS122B - D3865, JULY 1991 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

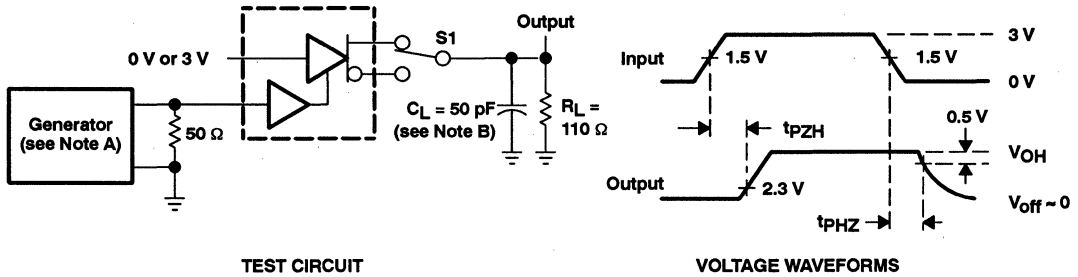


Figure 3. Test Circuit and Voltage Waveforms,  $t_{pZH}$  and  $t_{pHZ}$

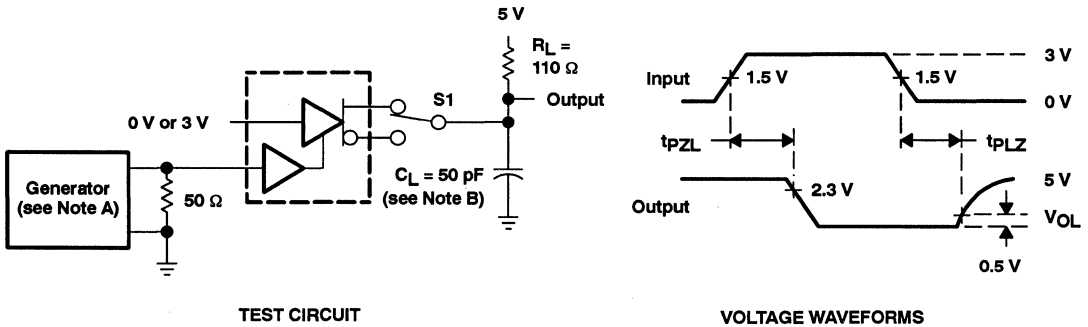


Figure 4. Test Circuit and Voltage Waveforms,  $t_{pZL}$  and  $t_{pLZ}$

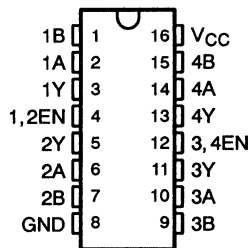
- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz,  $Z_0 = 50 \Omega$ , duty cycle = 50%,  $t_f \leq 5$  ns.  
B.  $C_L$  includes probe and stray capacitance.

# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A – D2602, OCTOBER 1990 – REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A, RS-423-A, and RS-485
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- Designed for Multipoint Bus Transmission on Long Bus Lines in Noisy Environments
- 3-State Outputs
- Common-Mode Input Voltage Range  
–12 V to 12 V
- Input Sensitivity . . .  $\pm 200$  mV
- Input Hysteresis . . . 50 mV Typ
- High Input Impedance . . . 12 k $\Omega$  Min
- Operates From Single 5-V Supply
- Low-Power Requirements
- Plug-In Replacement for MC3486

D OR N PACKAGE  
(TOP VIEW)



FUNCTION TABLE  
(each receiver)

DIFFERENTIAL INPUTS A – B	ENABLE	OUTPUT Y
$V_{ID} \geq 0.2$ V	H	H
$-0.2$ V < $V_{ID}$ < 0.2 V	H	?
$V_{ID} \leq -0.2$ V	H	L
X	L	Z

H = high level, L = low level, ? = indeterminate;  
X = irrelevant, Z = high impedance (off)

## description

The SN65175 and SN75175 are monolithic quad differential line receivers with 3-state outputs.

They are designed to meet the requirements of EIA Standards RS-422-A, RS-423A, RS-485, and several CCITT recommendations. These devices are optimized for balanced multipoint bus transmission at rates up to 10 megabits per second. Each of the two pairs of receivers has a common active-high enable.

The receivers feature high input impedance, input hysteresis for increased noise immunity, and input sensitivity of  $\pm 200$  mV over a common-mode input voltage range of  $\pm 12$  V. The SN65175 and SN75175 are designed for optimum performance when used with the SN75172 or SN75174 quadruple differential line drivers.

The SN65175 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75175 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

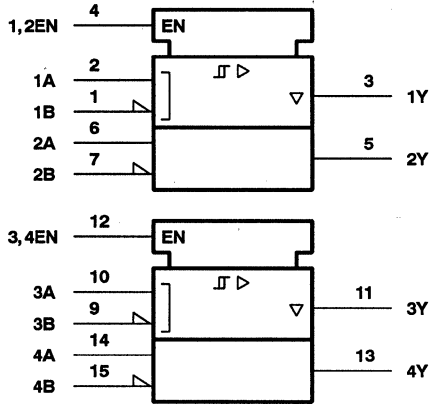
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# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

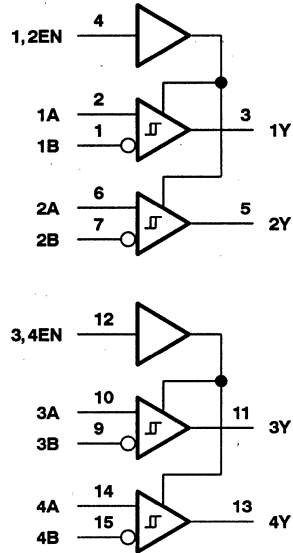
SLLS145A—D2602, OCTOBER 1990—REVISED FEBRUARY 1993

## logic symbol†

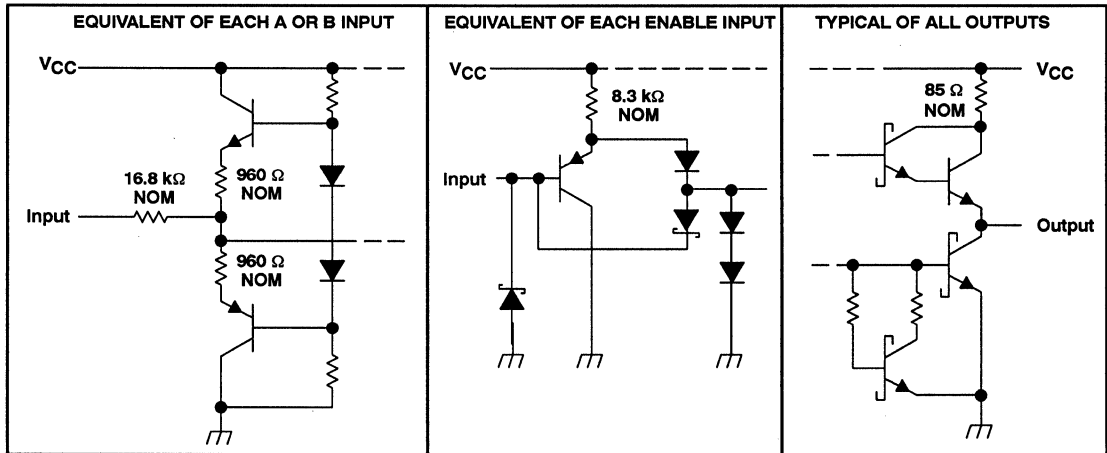


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## schematics of inputs and outputs



# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A—D2602, OCTOBER 1990—REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, A or B inputs	$\pm 25$ V
Differential input voltage (see Note 2)	$\pm 25$ V
Enable input voltage	7 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : SN65175	-40°C to 85°C
SN75175	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	-260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 12$	V
Differential input voltage, $V_{ID}$			$\pm 12$	V
High-level enable-input voltage, $V_{IH}$		2		V
Low-level enable-input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-400	$\mu\text{A}$
Low-level output current, $I_{OL}$			16	mA
Operating free-air temperature, $T_A$	SN65175	-40	85	°C
	SN75175	0	70	





# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A - D2602, OCTOBER 1990 - REVISED FEBRUARY 1993

## electrical characteristics over recommended ranges of common-mode input voltage, supply voltage and operating free-air temperature

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going input threshold voltage	$V_O = 2.7$ V, $I_O = -0.4$ mA			0.2	V
$V_{T-}$ Negative-going input threshold voltage	$V_O = 0.5$ V, $I_O = 16$ mA	-0.2‡			V
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )	See Figure 4		50		mV
$V_{IK}$ Enable-input clamp voltage	$I_I = -18$ mA			-1.5	V
$V_{OH}$ High-level output voltage	$V_{ID} = 200$ mV, $I_{OH} = -400$ $\mu$ A, See Figure 1		2.7		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -200$ mV, See Figure 1			0.45	V
				0.5	
$I_{OZ}$ High-impedance-state output current	$V_O = 0.4$ V to 2.4 V			$\pm 20$	$\mu$ A
$I_I$ Line input current	Other input at 0 V, See Note 3			1	mA
				-0.8	
$I_{IH}$ High-level enable-input current	$V_{IH} = 2.7$ V			20	$\mu$ A
$I_{IL}$ Low-level enable-input current	$V_{IL} = 0.4$ V			-100	$\mu$ A
$r_i$ Input resistance			12		k $\Omega$
$I_{OS}$ Short-circuit output current§		-15		-85	mA
$I_{CC}$ Supply current	Outputs disabled			70	mA

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ$ C.

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold voltage levels only.

§ Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

NOTE 3: Refer to EIA Standards RS-422A, RS423-A, and RS-485 for exact conditions.

## switching characteristics, $V_{CC} = 5$ V, $C_L = 15$ pF, $T_A = 25^\circ$ C

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	See Figure 2		22	35	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			25	35	ns
$t_{PZH}$ Output enable time to high level	See Figure 3		13	30	ns
$t_{PZL}$ Output enable time to low level			19	30	ns
$t_{PHZ}$ Output disable time from high level	See Figure 3		26	35	ns
$t_{PLZ}$ Output disable time from low level			25	35	ns

TEXAS  
INSTRUMENTS

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# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A - D2602, OCTOBER 1990 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

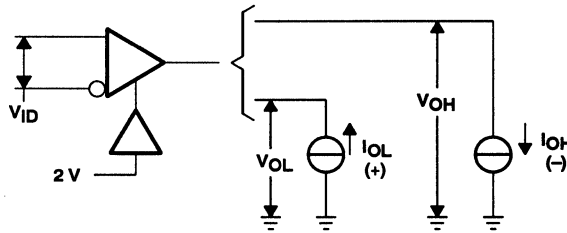


Figure 1.  $V_{OH}$ ,  $V_{OL}$

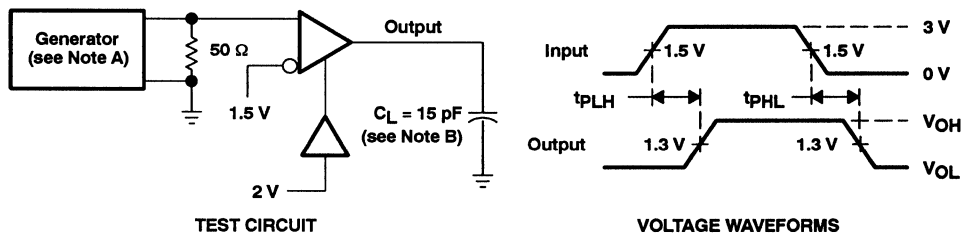


Figure 2. Test Circuit and Voltage Waveforms

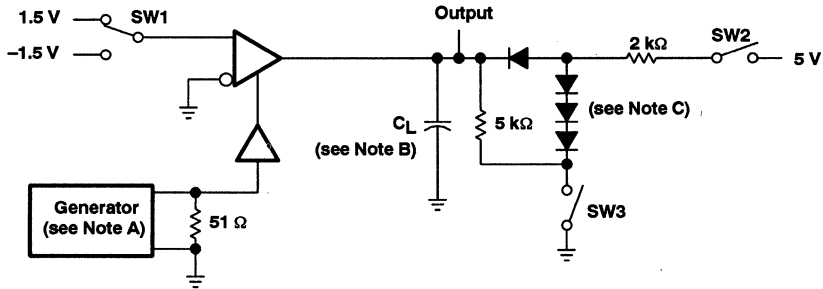
NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle = 50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and stray capacitance.

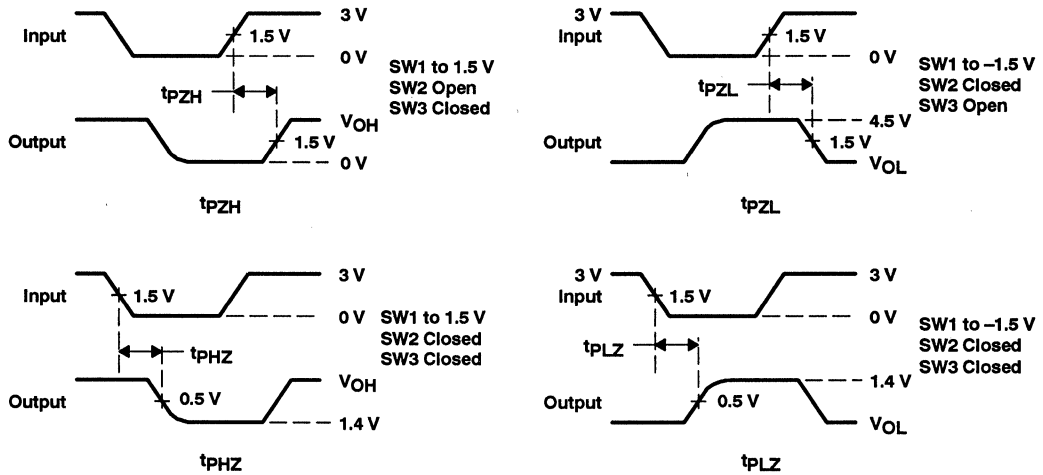
# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A - D2602, OCTOBER 1990 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION



## TEST CIRCUIT



## VOLTAGE WAVEFORMS

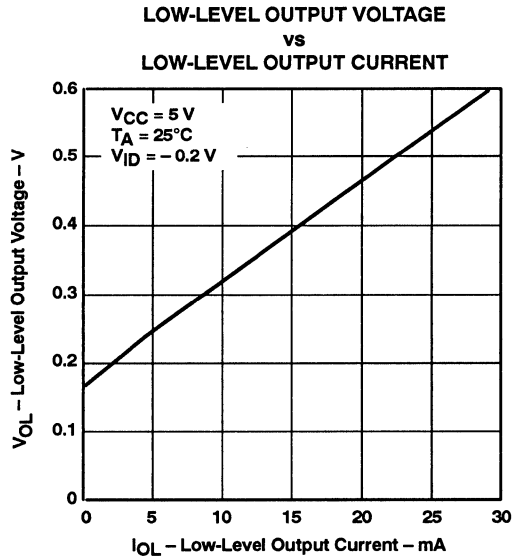
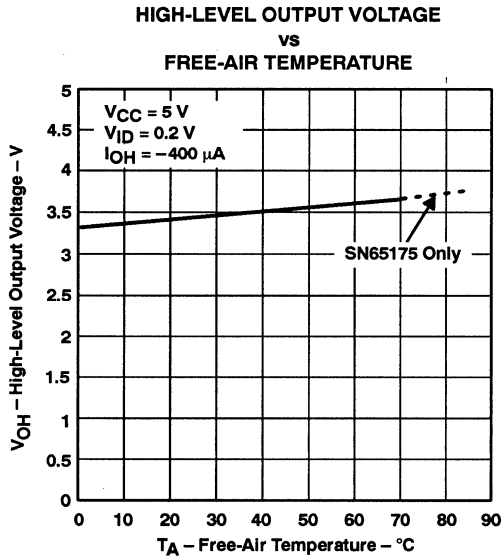
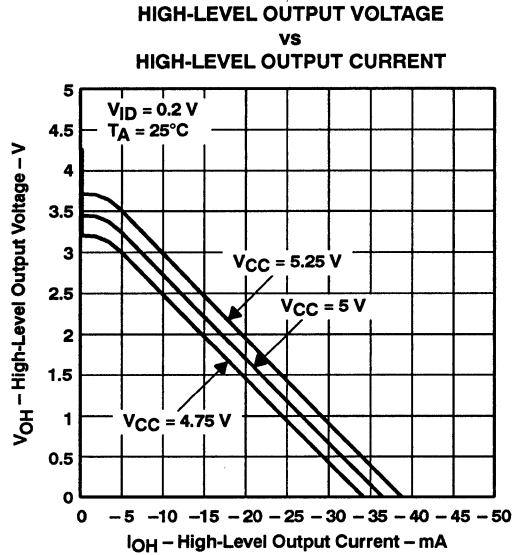
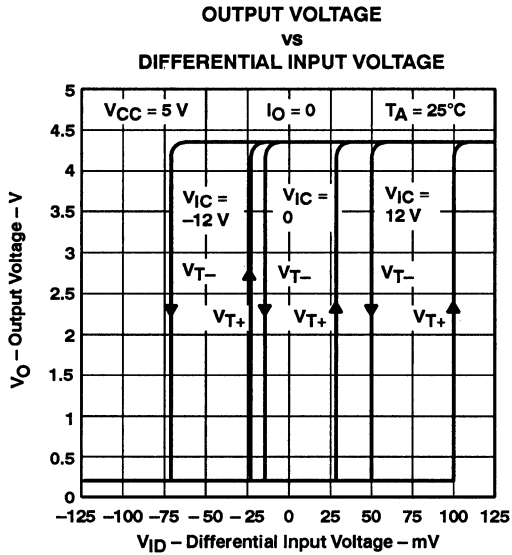
Figure 3. Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle = 50%,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and stray capacitance.  
 C. All diodes are 1N916 or equivalent.

# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A - D2602, OCTOBER 1990 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS



# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A – D2602, OCTOBER 1990 – REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

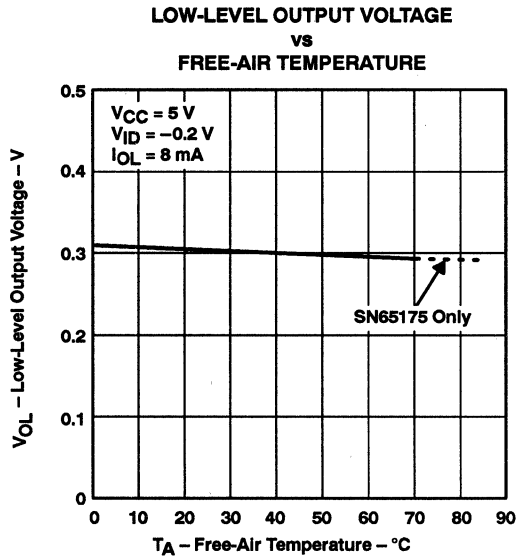


Figure 8

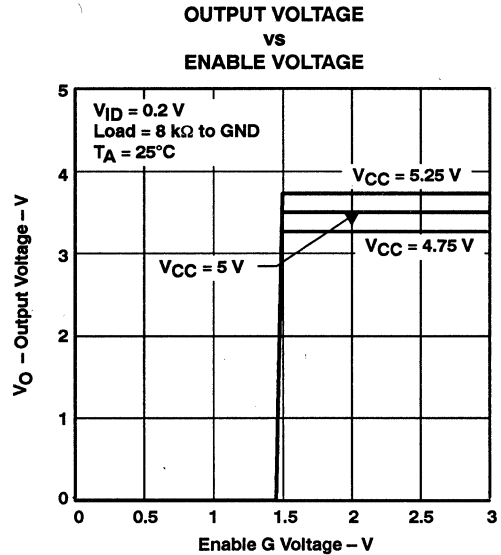


Figure 9

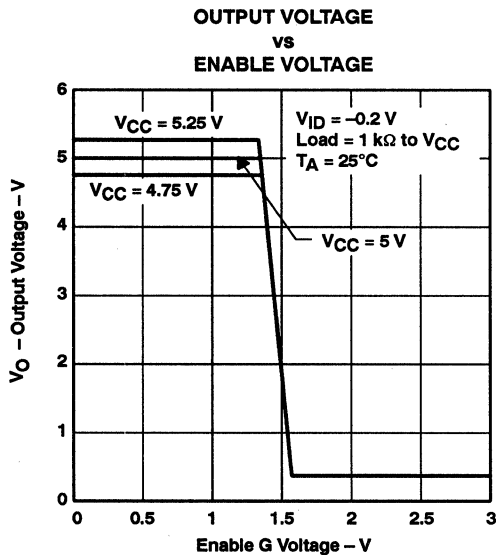


Figure 10

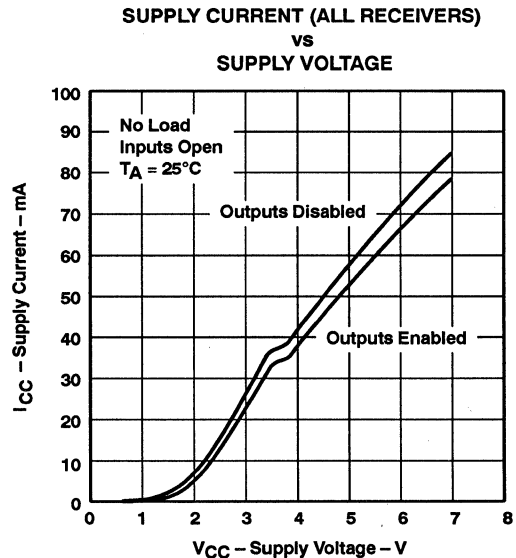


Figure 11

# SN65175, SN75175 QUAD DIFFERENTIAL LINE RECEIVERS

SLLS145A - D2602, OCTOBER 1990 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

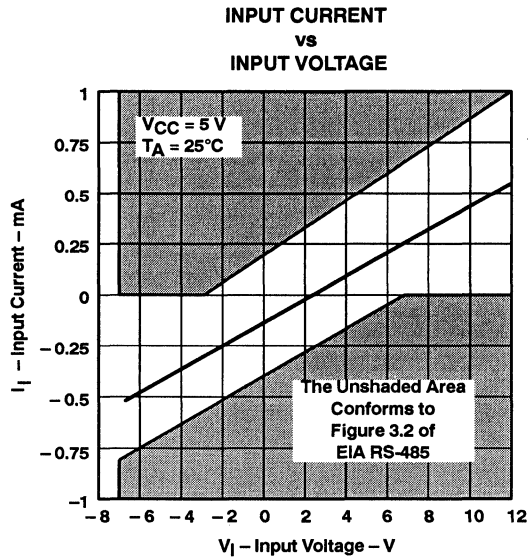


Figure 12

## APPLICATION INFORMATION

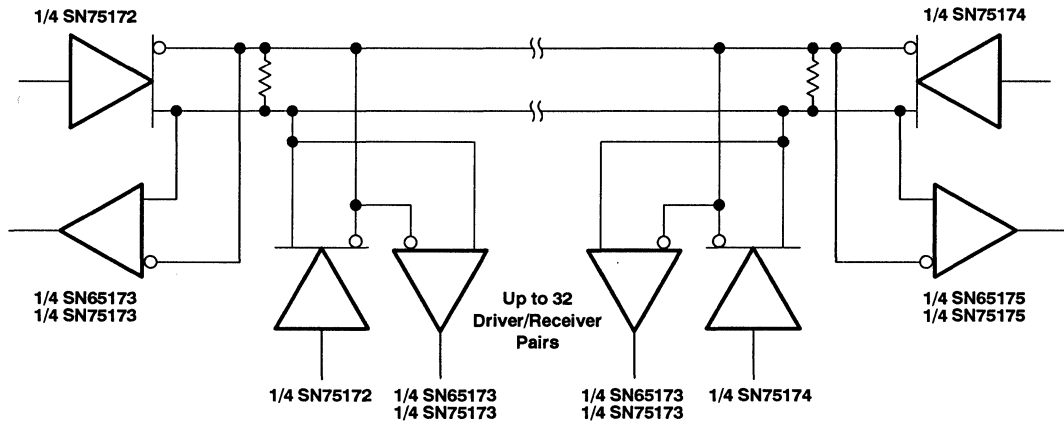


Figure 13. Typical Application Circuit

NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.



# SN75ALS175 QUAD DIFFERENTIAL LINE RECEIVER

SLLS131B – D3910, SEPTEMBER 1991 – REVISED JANUARY 1993

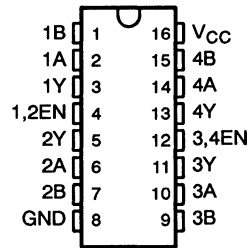
- Meets EIA Standards RS-422-A, RS-423-A, and RS-485
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- Designed for Multipoint Bus Transmission on Long Bus Lines in Noisy Environments
- Low Supply Current Requirement  
27 mA Max
- Common-Mode Input Voltage Range of  
–12 V to 12 V
- Input Sensitivity . . . ±200 mV
- Input Hysteresis . . . 50 mV Typ
- High Input Impedance . . . 12 kΩ Min
- Operates From Single 5-V Supply

## description

The SN75ALS175 is a monolithic quadruple differential line receiver with 3-state outputs. It is designed to meet the requirements of EIA Standards RS-422-A, RS-423-A, and RS-485 and several CCITT recommendations. Advanced low-power Schottky technology provides high speed without the usual power penalty. Each of the two pairs of receivers has a common active-high enable. The device features high input impedance, input hysteresis for increased noise immunity, and input sensitivity of ±200 mV over a common-mode input voltage range of –12 V to 12 V.

The SN75ALS175 is characterized for operation from 0°C to 70°C.

N OR NS† PACKAGE  
(TOP VIEW)



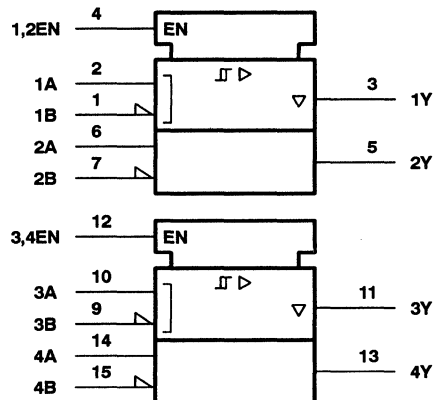
† The NS package is only available left-ende taped and reeled (order device SN75ALS175NSLE).

FUNCTION TABLE (EACH RECEIVER)

DIFFERENTIAL INPUTS A – B	ENABLE EN	OUTPUT Y
$V_{ID} \geq 0.2 \text{ V}$	H	H
$-0.2 \text{ V} < V_{ID} < 0.2 \text{ V}$	H	?
$V_{ID} \leq -0.2 \text{ V}$	H	L
X	L	Z
Open Circuit	H	H

H = high level, L = low level, ? = indeterminate;  
X = irrelevant, Z = high impedance (off)

## logic symbol‡



‡ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

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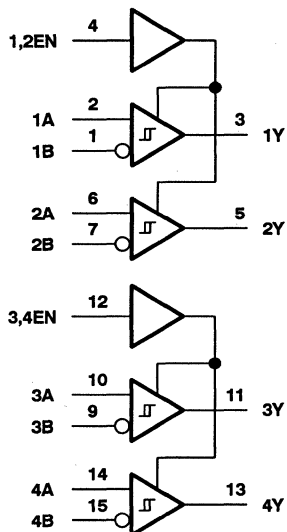
2-587



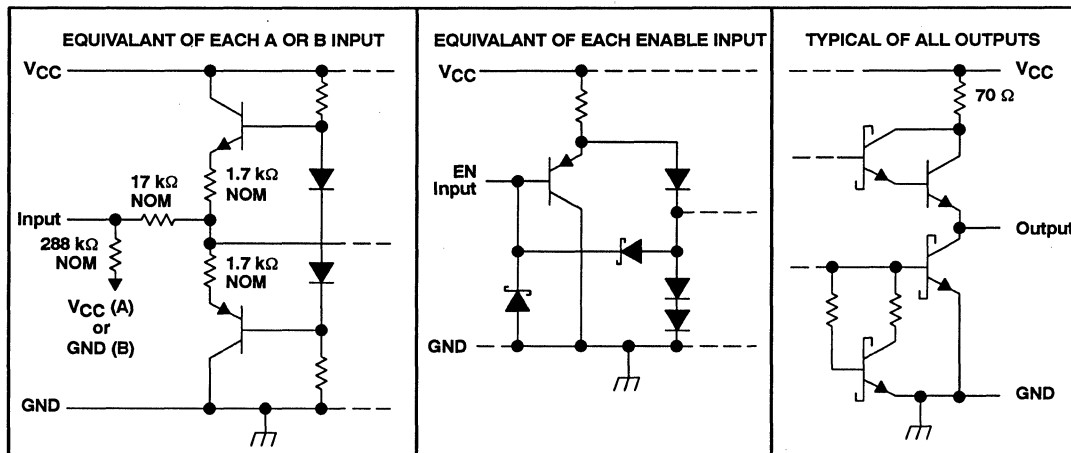
# SN75ALS175 QUAD DIFFERENTIAL LINE RECEIVER

SLLS131B - D3910, SEPTEMBER 1991 - REVISED JANUARY 1993

## logic diagram (positive logic)



## schematics of inputs and outputs



# SN75ALS175 QUAD DIFFERENTIAL LINE RECEIVER

SLLS131B – D3910, SEPTEMBER 1991 – REVISED JANUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, A or B inputs	$\pm 14$ V
Differential input voltage (see Note 2)	$\pm 14$ V
Enable input voltage	7 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
N	1150 mW	9.2 mW/°C	736 mW
NS	625 mW	5.0 mW/°C	400 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 12$	V
Differential input voltage, $V_{ID}$			$\pm 12$	V
High-level enable-input voltage, $V_{IH}$	2			V
Low-level enable-input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-400	$\mu\text{A}$
Low-level output current, $I_{OL}$			8	mA
Operating free-air temperature, $T_A$	0		70	°C



# SN75ALS175 QUAD DIFFERENTIAL LINE RECEIVER

SLLS131B – D3910, SEPTEMBER 1991 – REVISED JANUARY 1993

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage and operating free-air temperature (unless otherwise noted) (see Note 3)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$V_{T+}$ Positive-going threshold voltage				200	mV	
$V_{T-}$ Negative-going threshold voltage		-200‡			mV	
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )			50		mV	
$V_{IK}$ Enable-input clamp voltage	$I_I = -18$ mA			-1.5	V	
$V_{OH}$ High-level output voltage	$V_{ID} = 200$ mV, $I_{OH} = -400$ $\mu$ A, See Figure 1		2.7		V	
$V_{OL}$ Low-level output voltage	$V_{ID} = -200$ mV, $I_{OL} = 8$ mA, See Figure 1			0.45	V	
$I_{OZ}$ High-impedance-state output current	$V_O = 0.4$ V to 2.4 V			$\pm 20$	$\mu$ A	
$I_I$ Line input current	Other input at 0 V, See Note 3			$V_I = 12$ V	1	mA
				$V_I = -7$ V	-0.8	
$I_{IH}$ High-level enable-input current	$V_{IH}(E) = 2.7$ V			20	$\mu$ A	
$I_{IL}$ Low-level enable-input current	$V_{IL}(E) = 0.4$ V			-100	$\mu$ A	
$r_i$ Input resistance			12		k $\Omega$	
$I_{OS}$ Short-circuit output current	$V_O = 0$ , See Note 4		-15	-85	mA	
$I_{CC}$ Supply current (total package)	No load, Outputs enabled		16	24	mA	
	No load, Outputs disabled		18	27		

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold voltage levels only.

NOTES: 3. Refer to EIA Standards RS-485 for exact conditions.

4. Not more than one output should be shorted at a time, and the duration of the short-circuit should not exceed one second.

## switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PHL}$ Propagation delay time, high-to-low-level output	$V_{ID} = -2.5$ V to 2.5 V,	9	18	27	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 15$ pF, See Figure 2	9	18	27	ns
$t_{PZH}$ Output enable time to high level	$C_L = 15$ pF, See Figure 3	4	12	18	ns
$t_{PZL}$ Output enable time to low level		6	13	21	ns
$t_{PHZ}$ Output disable time from high level	$C_L = 15$ pF, See Figure 3	10	21	27	ns
$t_{PLZ}$ Output disable time from low level		8	15	25	ns

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

TEXAS  
INSTRUMENTS

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PARAMETER MEASUREMENT INFORMATION

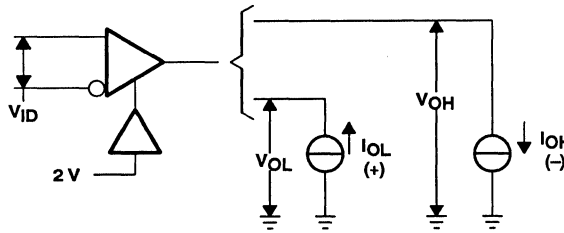
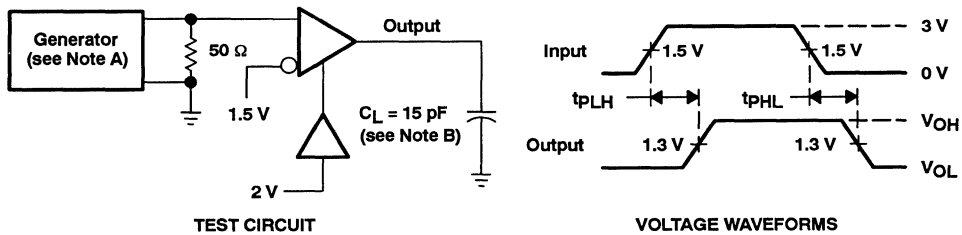


Figure 1.  $V_{OH}$ ,  $V_{OL}$



TEST CIRCUIT

VOLTAGE WAVEFORMS

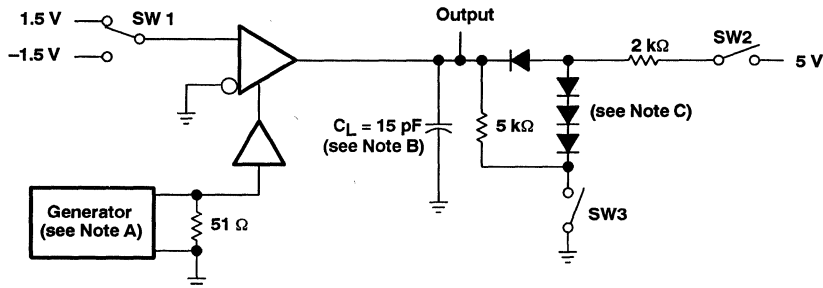
NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r = t_f = 6$  ns.  
 B.  $C_L$  includes probe and jig capacitance.

Figure 2. Propagation Delay Times

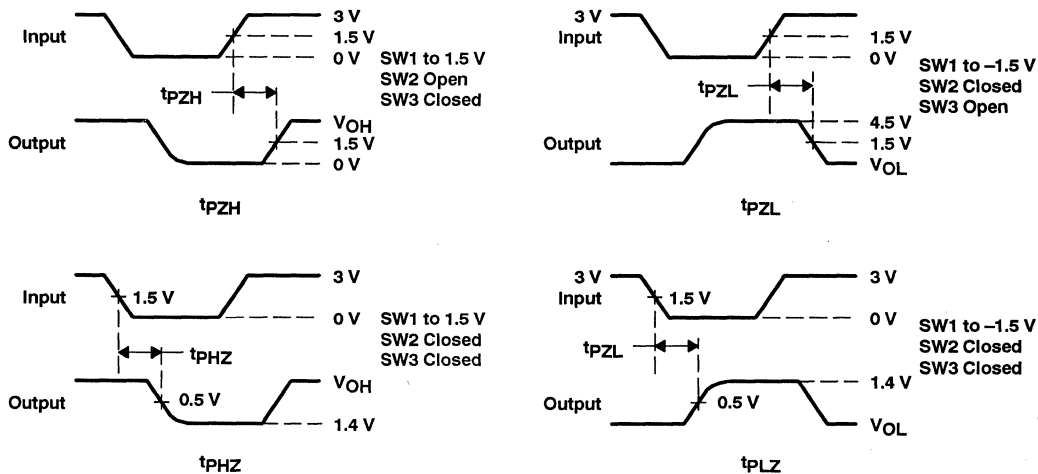
# SN75ALS175 QUAD DIFFERENTIAL LINE RECEIVER

SLLS131B – D3910, SEPTEMBER 1991 – REVISED JANUARY 1993

## PARAMETER MEASUREMENT INFORMATION



## TEST CIRCUIT



## VOLTAGE WAVEFORMS

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r = t_f = 6$  ns.  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N916 or equivalent.

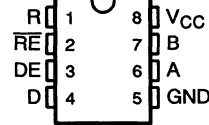
Figure 3. Enable and Disable Times

# SN75176A DIFFERENTIAL BUS TRANSCEIVER

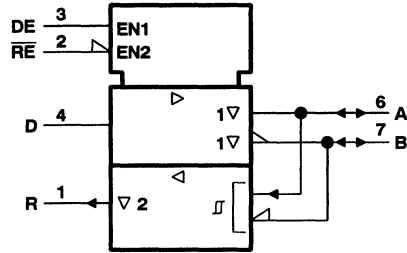
SLLS100 – D2619, JUNE 1984 – REVISED AUGUST 1989

- Bidirectional Transceiver
- Meets EIA Standards RS-422-A and CCITT Recommendations V.11 and X.27
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- 3-State Driver and Receiver Outputs
- Individual Driver and Receiver Enables
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capability . . .  $\pm 60$  mA Max
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Impedance . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From Single 5-V Supply
- Low Power Requirements

**D OR P PACKAGE  
(TOP VIEW)**



**logic symbol†**



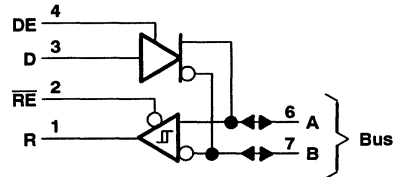
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## description

The SN75176A differential bus transceiver is a monolithic integrated circuit designed for bidirectional data communication on multipoint bus transmission lines. It is designed for balanced transmission lines and meet EIA Standard RS-422-A and CCITT Recommendations V.11 and X.27.

The SN75176A combines a 3-state differential line driver and a differential input line receiver, both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, that can be externally connected together to function as a direction control. The driver differential outputs and the receiver differential inputs are connected internally to form differential input/output (I/O) bus ports that are designed to offer minimum loading to the bus whenever the driver is disabled or  $V_{CC} = 0$ . These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

**logic diagram (positive logic)**



## Function Tables

DRIVER			
INPUT D	ENABLE DE	OUTPUTS	
		A	B
H	H	H	L
L	H	L	H
X	L	Z	Z

RECEIVER		
DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2$ V	L	H
$-0.2$ V $< V_{ID} < 0.2$ V	L	?
$V_{ID} \leq -0.2$ V	L	L
X	H	Z

H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = high impedance (off)

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# SN75176A DIFFERENTIAL BUS TRANSCEIVER

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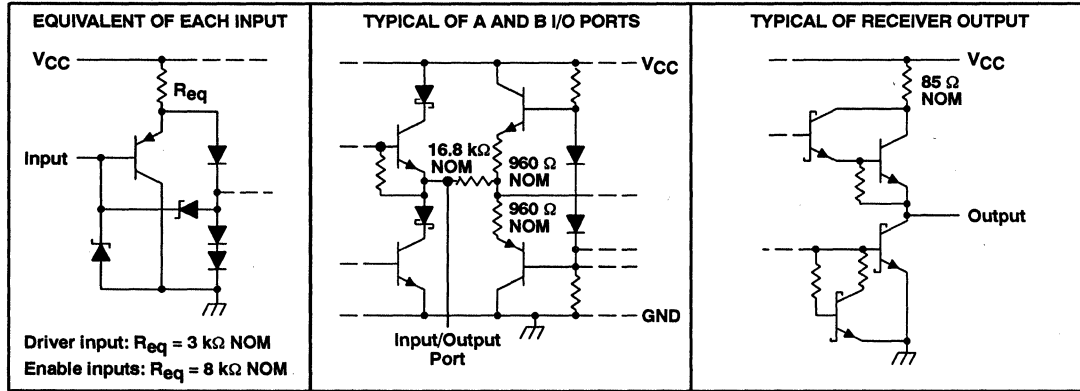
## description (continued)

The driver is designed to handle loads up to 60 mA of sink or source current. The driver features positive- and negative-current limiting and thermal shutdown for protection from line fault conditions. Thermal shutdown is designed to occur at a junction temperature of approximately 150°C. The receiver features a minimum input impedance of 12 kΩ, an input sensitivity of ±200 mV, and a typical input hysteresis of 50 mV.

The SN75176A can be used in transmission line applications employing the SN75172 and SN75174 quad differential line drivers and SN75173 and SN75175 quad differential line receivers.

The SN75176A is characterized for operation from 0°C to 70°C.

## schematics of inputs and outputs



# SN75176A DIFFERENTIAL BUS TRANSCEIVER

SLLS100 – D2619, JUNE 1984 – REVISED AUGUST 1989

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Voltage range at any bus terminal .....	–10 V to 15 V
Enable input voltage .....	5.5 V
Continuous total power dissipation .....	See Dissipation Rating Table
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

NOTE 1: All voltage values, except differential input/output bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 105^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	261 mW
P	1100 mW	8.8 mW/°C	702 mW	396 mW

## recommended operating conditions

		MIN	TYP	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$		–7		12	V
High-level input voltage, $V_{IH}$	D, DE, and $\overline{RE}$	2			V
Low-level input voltage, $V_{IL}$	D, DE, and $\overline{RE}$			0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)				±12	V
High-level output current, $I_{OH}$	Driver			–60	mA
	Receiver			–400	μA
Low-level output current, $I_{OL}$	Driver			60	mA
	Receiver			8	mA
Operating free-air temperature, $T_A$		0		70	°C

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.





# SN75176A DIFFERENTIAL BUS TRANSCEIVER

SLLS100 – D2619, JUNE 1984 – REVISED AUGUST 1989

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V
$V_{OH}$ High-level output voltage	$V_{IH} = 2 \text{ V}$ , $I_{OH} = -33 \text{ mA}$		3.7		V
$V_{OL}$ Low-level output voltage	$V_{IH} = 2 \text{ V}$ , $I_{OH} = 33 \text{ mA}$		1.1		V
$ V_{OD1} $ Differential output voltage	$I_O = 0$			$2V_{OD2}$	V
$ V_{OD2} $ Differential output voltage	$R_L = 100 \Omega$ , See Figure 1	2	2.7		V
	$R_L = 54 \Omega$ , See Figure 1	1.5	2.4		
$\Delta V_{OD} $ Change in magnitude of differential output voltage‡				$\pm 0.2$	V
$V_{OC}$ Common-mode output voltage§	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1			3	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage‡				$\pm 0.2$	V
$I_O$ Output current	Output disabled, See Note 3	$V_O = 12 \text{ V}$		1	mA
		$V_O = -7 \text{ V}$		-0.8	
$I_{IH}$ High-level input current	$V_I = 2.4 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_I = 0.4 \text{ V}$			-400	$\mu\text{A}$
$I_{OS}$ Short-circuit output current	$V_O = -7 \text{ V}$			-250	mA
	$V_O = V_{CC}$			250	
	$V_O = 12 \text{ V}$			500	
$I_{CC}$ Supply current (total package)	No load	Outputs enabled	35	50	mA
		Outputs disabled	26	40	

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$  respectively, that occur when the input is changed from a high level to a low level.

§ In EIA Standard RS-422A,  $V_{OC}$ , which is the average of the two output voltages with respect to GND, is called output offset voltage,  $V_{OS}$ .

NOTE 3: This applies for both power on and off; refer to EIA Standard RS-422A for exact conditions.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{dD}$ Differential-output delay time	$R_L = 60 \Omega$ , See Figure 3		40	60	ns
$t_{tD}$ Differential-output transition time			65	95	
$t_{pZH}$ Output enable time to high level	$R_L = 110 \Omega$ , See Figure 4		55	90	ns
$t_{pZL}$ Output enable time to low level	$R_L = 110 \Omega$ , See Figure 5		30	50	ns
$t_{pHZ}$ Output disable time from high level	$R_L = 110 \Omega$ , See Figure 4		85	130	ns
$t_{pLZ}$ Output disable time from low level	$R_L = 110 \Omega$ , See Figure 5		20	40	ns

TEXAS  
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# SN75176A DIFFERENTIAL BUS TRANSCEIVER

SLLS100 – D2619, JUNE 1984 – REVISED AUGUST 1989

## RECEIVER SECTION

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going input threshold voltage	$V_O = 2.7\text{ V}$ ,	$I_O = -0.4\text{ mA}$			0.2	V
$V_{T-}$	Negative-going input threshold voltage	$V_O = 0.5\text{ V}$ ,	$I_O = 8\text{ mA}$	$-0.2‡$			V
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )				50		mV
$V_{IK}$	Enable clamp voltage	$I_I = -18\text{ mA}$				-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 200\text{ mV}$ , See Figure 2	$I_{OH} = -400\text{ }\mu\text{A}$ ,	2.7			V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200\text{ mV}$ , See Figure 2	$I_{OL} = 8\text{ mA}$ ,			0.45	V
$I_{OZ}$	High-impedance-state output current	$V_O = 0.4\text{ V to } 2.4\text{ V}$				$\pm 20$	$\mu\text{A}$
$I_I$	Line input current	Other input = 0 V, See Note 3	$V_I = 12\text{ V}$			1	mA
			$V_I = -7\text{ V}$			-0.8	
$I_{IH}$	High-level enable input current	$V_{IH} = 2.7\text{ V}$				20	$\mu\text{A}$
$I_{IL}$	Low-level enable input current	$V_{IL} = 0.4\text{ V}$				-100	$\mu\text{A}$
$r_i$	Input resistance			12			k $\Omega$
$I_{OS}$	Short-circuit output current			-15		-85	mA
$I_{CC}$	Supply current (total package)	No load	Outputs enabled		35	50	mA
			Outputs disabled		26	40	

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 3: This applies for both power on and power off. Refer to EIA Standard RS-422A for exact conditions.

### switching characteristics, $V_{CC} = 5\text{ V}$ , $C_L = 15\text{ pF}$ , $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -1.5\text{ V to } 1.5\text{ V}$ ,	See Figure 6		21	35	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output				23	35	
$t_{PZH}$	Output enable time to high level	See Figure 7			10	30	ns
$t_{PZL}$	Output enable time to low level				12	30	
$t_{PHZ}$	Output disable time from high level	See Figure 7			20	35	ns
$t_{PLZ}$	Output disable time from low level				17	25	



# SN75176A DIFFERENTIAL BUS TRANSCEIVER

SLLS100 – D2619, JUNE 1984 – REVISED AUGUST 1989

## PARAMETER MEASUREMENT INFORMATION

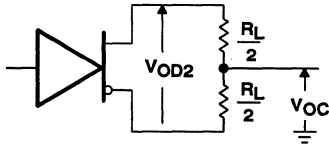


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

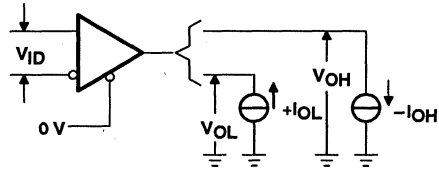


Figure 2. Receiver  $V_{OH}$  and  $V_{OL}$

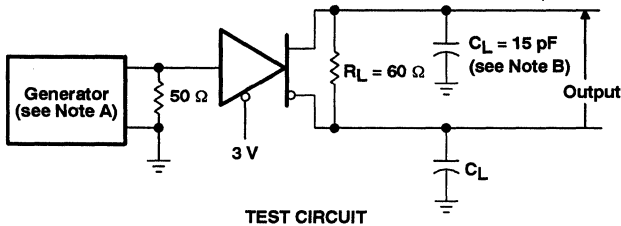


Figure 3. Driver Test Circuit and Voltage Waveforms

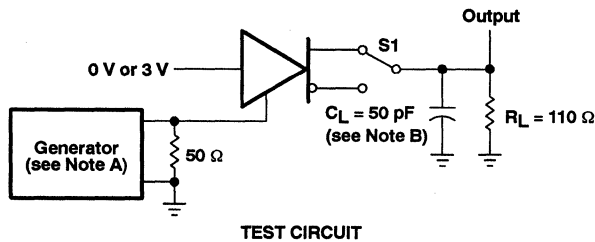


Figure 4. Driver Test Circuit and Voltage Waveforms

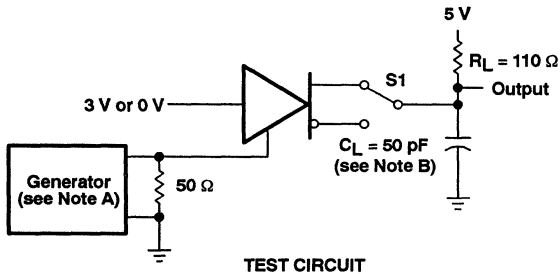


Figure 5. Driver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

PARAMETER MEASUREMENT INFORMATION

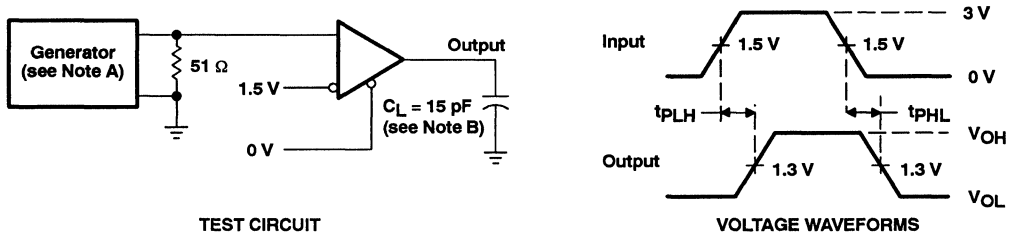


Figure 6. Receiver Test Circuit and Voltage Waveforms

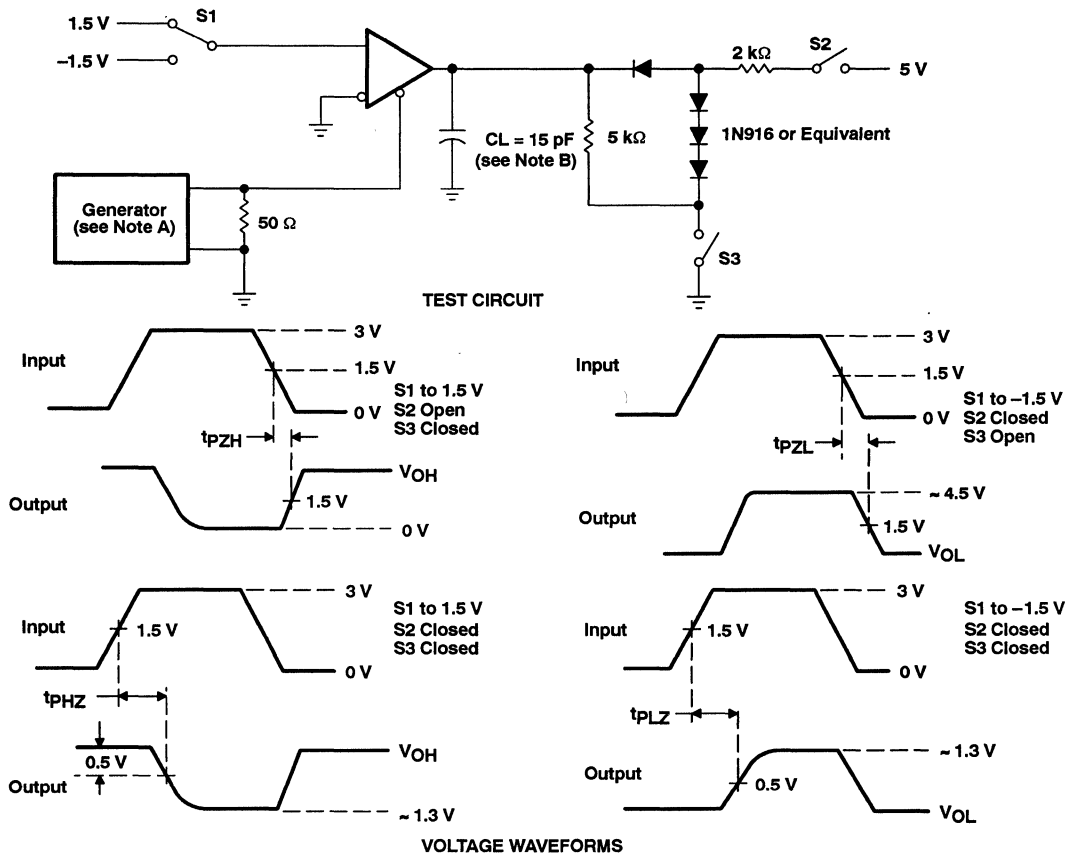


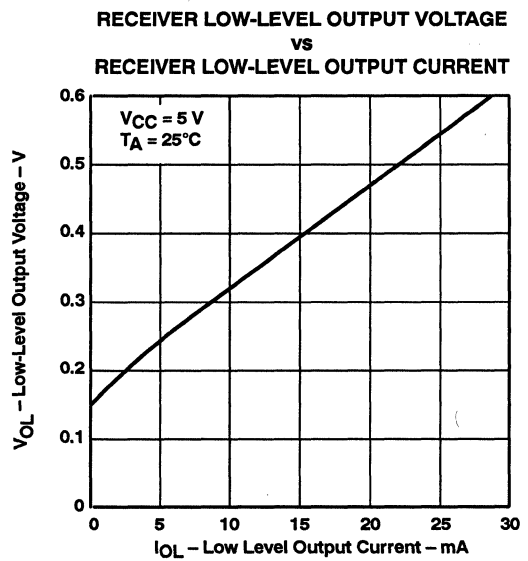
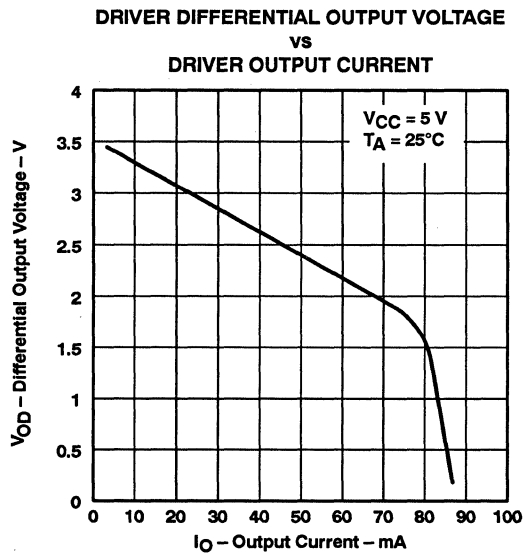
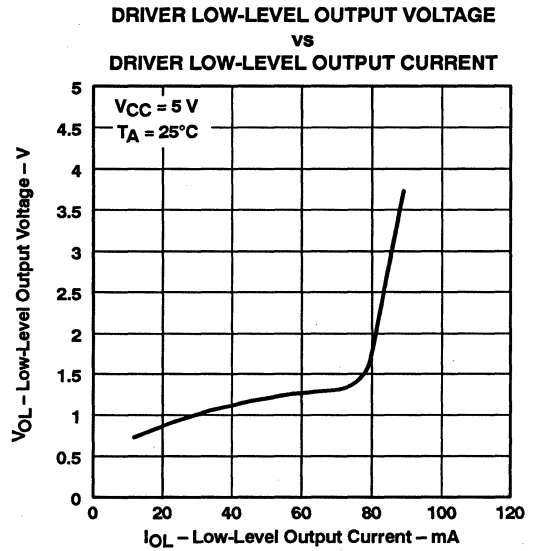
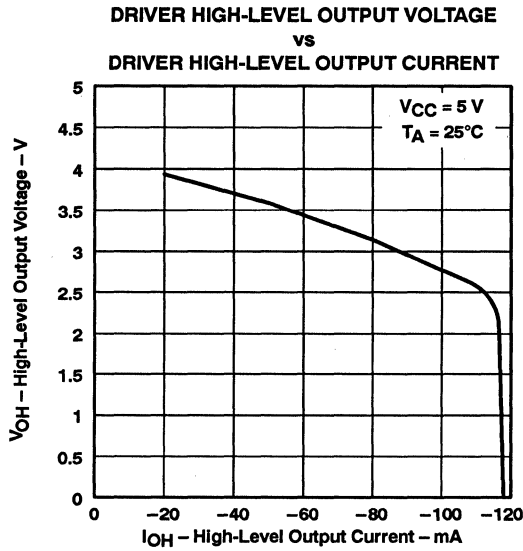
Figure 7. Receiver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

# SN75176A DIFFERENTIAL BUS TRANSCEIVER

SLLS100 – D2619, JUNE 1984 – REVISED AUGUST 1989

## TYPICAL CHARACTERISTICS



TYPICAL CHARACTERISTICS

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

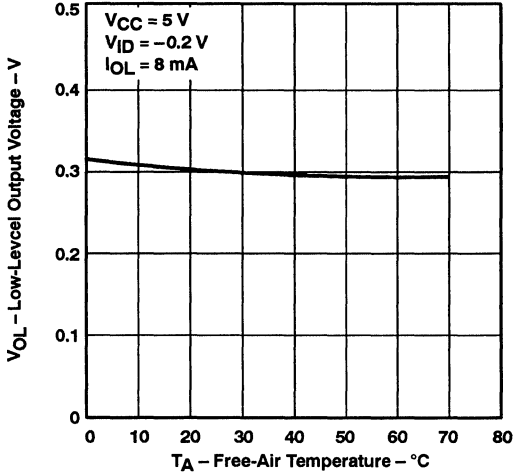


Figure 12

RECEIVER OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE

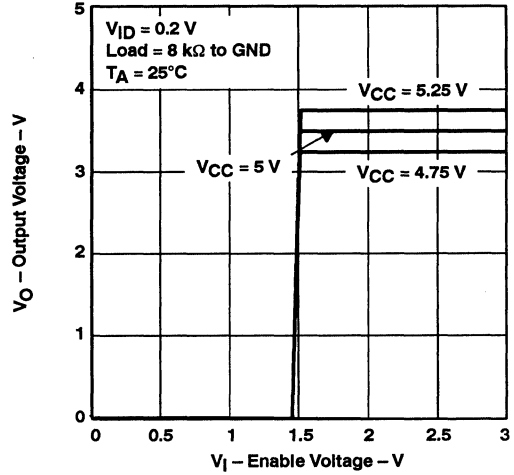


Figure 13

RECEIVER OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE

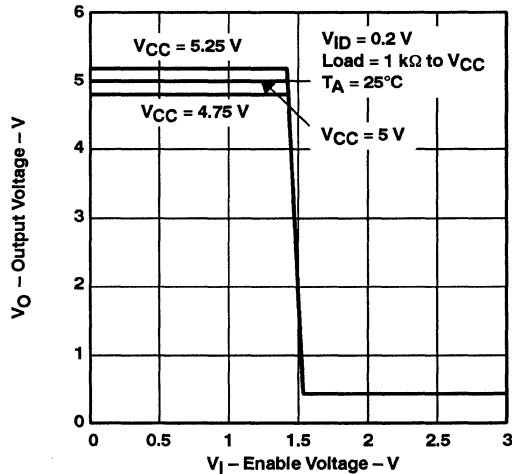


Figure 14

# SN75176A DIFFERENTIAL BUS TRANSCEIVER

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## APPLICATION INFORMATION

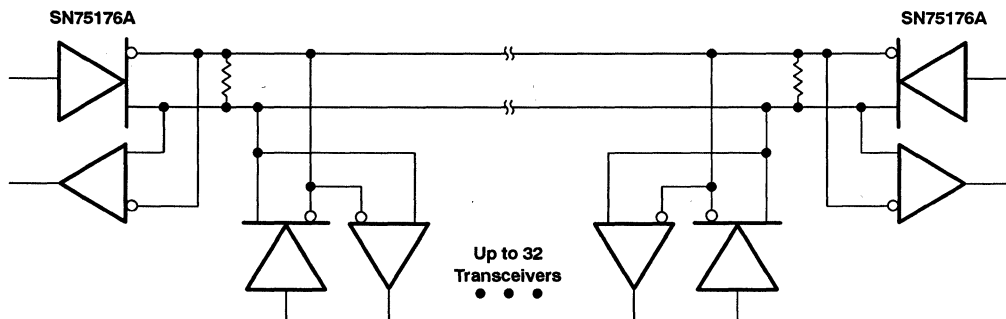


Figure 15. Typical Application Circuit

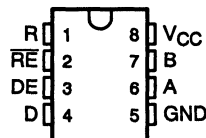
NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

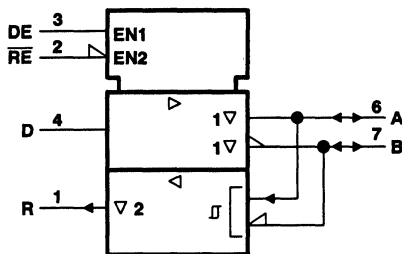
SLLS101 – D2619, JULY 1985 – REVISED SEPTEMBER 1989

- Bidirectional Transceiver
- Meets EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- 3-State Driver and Receiver Outputs
- Individual Driver and Receiver Enables
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capability . . .  $\pm 60$  mA Max
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Impedance . . .  $12\text{ k}\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . .  $50$  mV Typ
- Operates From Single 5-V Supply
- Low Power Requirements

D OR P PACKAGE  
(TOP VIEW)



logic symbol†



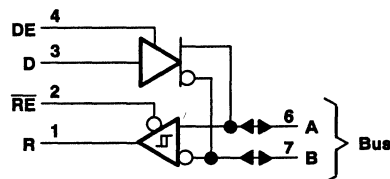
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## description

The SN65176B and SN75176B differential bus transceivers are monolithic integrated circuits designed for bidirectional data communication on multipoint bus transmission lines. They are designed for balanced transmission lines and meet EIA Standard RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27.

The SN65176B and SN75176B combine a 3-state differential line driver and a differential input line receiver, both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, that can be externally connected together to function as a direction control. The driver differential outputs and the receiver differential inputs are connected internally to form differential input/output (I/O) bus ports that are designed to offer minimum loading to the bus whenever the driver is disabled or  $V_{CC} = 0$ . These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

logic diagram (positive logic)



## Function Tables

DRIVER			
INPUT D	ENABLE DE	OUTPUTS	
		A	B
H	H	H	L
L	H	L	H
X	L	Z	Z

RECEIVER		
DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2\text{ V}$	L	H
$-0.2\text{ V} < V_{ID} < 0.2\text{ V}$	L	?
$V_{ID} \leq -0.2\text{ V}$	L	L
X	H	Z
Open	L	H

H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = high impedance (off)

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

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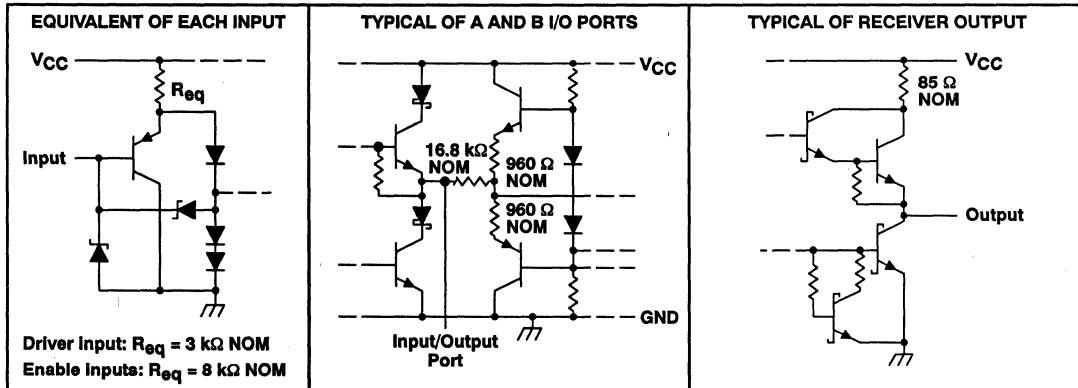
## description (continued)

The driver is designed to handle loads up to 60 mA of sink or source current. The driver features positive- and negative-current limiting and thermal shutdown for protection from line fault conditions. Thermal shutdown is designed to occur at a junction temperature of approximately 150°C. The receiver features a minimum input impedance of 12 kΩ, an input sensitivity of ±200 mV, and a typical input hysteresis of 50 mV.

The SN65176B and SN75176B can be used in transmission line applications employing the SN75172 and SN75174 quadruple differential line drivers and SN75173 and SN75175 quadruple differential line receivers.

The SN65176B characterized for operation from -40°C to 105°C and the SN75176B is characterized for operation from 0°C to 70°C.

## schematics of inputs and outputs



# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS101 – D2619, JULY 1985 – REVISED SEPTEMBER 1989

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Voltage at any bus terminal .....	–10 V to 15 V
Enable input voltage .....	5.5 V
Continuous total power dissipation .....	See Dissipation Rating Table
Operating free-air temperature range: SN65176B .....	–40°C to 105°C
SN75176B .....	0°C to 70°C
Storage temperature range .....	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

NOTE 1: All voltage values, except differential input/output bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 105^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	261 mW
P	1100 mW	8.8 mW/°C	702 mW	396 mW

## recommended operating conditions

		MIN	TYP	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$				12	V
				–7	
High-level input voltage, $V_{IH}$	D, DE, and $\overline{RE}$	2			V
Low-level input voltage, $V_{IL}$	D, DE, and $\overline{RE}$			0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)				±12	V
High-level output current, $I_{OH}$	Driver			–60	mA
	Receiver			–400	µA
Low-level output current, $I_{OL}$	Driver			60	mA
	Receiver			8	
Operating free-air temperature, $T_A$	SN65176B	–40		105	°C
	SN75176B	0		70	

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.

# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS101 – D2619, JULY 1985 – REVISED SEPTEMBER 1989

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA				-1.5	V
$V_O$	Output voltage	$I_O = 0$		0		6	V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5	6	6	V
$ V_{OD2} $	Differential output voltage	$R_L = 100 \Omega$ ,	See Figure 1	$1/2 V_{OD1}$ or 2V			V
		$R_L = 54 \Omega$ ,	See Figure 1	1.5	2.5	5	V
$V_{OD3}$	Differential output voltage	See Note 4		1.5		5	V
$\Delta V_{OD} $	Change in magnitude of differential output voltage§	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1				$\pm 0.2$	V
$V_{OC}$	Common-mode output voltage					+3 -1	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage§					$\pm 0.2$	V
$I_O$	Output current	Output disabled, See Note 5	$V_O = 12$ V			1	mA
			$V_O = -7$ V			-0.8	
$I_{IH}$	High-level input current	$V_I = 2.4$ V				20	$\mu$ A
$I_{IL}$	Low-level input current	$V_I = 0.4$ V				-400	$\mu$ A
$I_{OS}$	Short-circuit output current	$V_O = -7$ V				-250	mA
		$V_O = 0$				150	
		$V_O = V_{CC}$				250	
		$V_O = 12$ V				250	
$I_{CC}$	Supply current (total package)	No load	Outputs enabled		42	70	mA
			Outputs disabled		26	35	

† The power-off measurement in EIA Standard RS-422-A applies to disabled outputs only and is not applied to combined inputs and outputs.

‡ All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .

§  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

¶ The minimum  $V_{OD2}$  with a  $100\text{-}\Omega$  load is either  $1/2 V_{OD1}$  or 2 V, whichever is greater.

NOTES: 1. See EIA Standard RS-485 Figure 3.5, Test Termination Measurement 2.

2. This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions. The RS-422-A limit does not apply for a combined driver and receiver terminal.

switching characteristics,  $V_{CC} = 5$  V,  $R_L = 110$  k $\Omega$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{dD}$	Differential-output delay time	$R_L = 54 \Omega$ See Figure 3			15	20	ns
$t_{tD}$	Differential-output transition time				20	30	ns
$t_{PZH}$	Output enable time to high level	See Figure 4			85	120	ns
$t_{PZL}$	Output enable time to low level	See Figure 5			40	60	ns
$t_{PHZ}$	Output disable time from high level	See Figure 4			150	250	ns
$t_{PLZ}$	Output disable time from low level	See Figure 5			20	30	ns

# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS101 – D2619, JULY 1985 – REVISED SEPTEMBER 1989

## SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{oa}, V_{ob}$	$V_{oa}, V_{ob}$
$ V_{OD1} $	$V_o$	$V_o$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$\Delta V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

## RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going input threshold voltage	$V_O = 2.7 \text{ V}$ ,	$I_O = -0.4 \text{ mA}$			0.2	V
$V_{T-}$	Negative-going input threshold voltage	$V_O = 0.5 \text{ V}$ ,	$I_O = 8 \text{ mA}$	-0.2‡			V
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )				50		mV
$V_{IK}$	Enable clamp voltage	$I_I = -18 \text{ mA}$				-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 200 \text{ mV}$ , See Figure 2	$I_{OH} = -400 \mu\text{A}$ ,	2.7			V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200 \text{ mV}$ , See Figure 2	$I_{OL} = 8 \text{ mA}$ ,			0.45	V
$I_{OZ}$	High-impedance-state output current	$V_O = 0.4 \text{ V to } 2.4 \text{ V}$				$\pm 20$	$\mu\text{A}$
$I_I$	Line input current	Other input = 0 V, See Note 5	$V_I = 12 \text{ V}$ $V_I = -7 \text{ V}$			1 -0.8	mA
$I_{IH}$	High-level enable input current	$V_{IH} = 2.7 \text{ V}$				20	$\mu\text{A}$
$I_{IL}$	Low-level enable input current	$V_{IL} = 0.4 \text{ V}$				-100	$\mu\text{A}$
$r_i$	Input resistance	$V_I = 12 \text{ V}$		12			k $\Omega$
$I_{OS}$	Short-circuit output current			-15		-85	mA
$I_{CC}$	Supply current (total package)	No load	Outputs enabled Outputs disabled		42 26	55 35	mA

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 3: This applies for both power on and power off. Refer to EIA Standard RS-485 for exact conditions.



# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

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switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 15\text{ pF}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$V_{ID} = 0\text{ to }3\text{ V}$ , See Figure 6		21	35	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			23	35	ns
$t_{PZH}$ Output enable time to high level	See Figure 7		10	20	ns
$t_{PZL}$ Output enable time to low level			12	20	ns
$t_{PHZ}$ Output disable time from high level	See Figure 7		20	35	ns
$t_{PLZ}$ Output disable time from low level			17	25	ns

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

PARAMETER MEASUREMENT INFORMATION

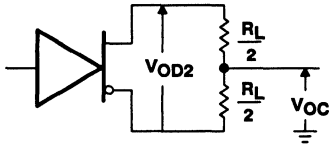


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

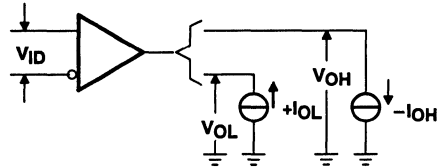
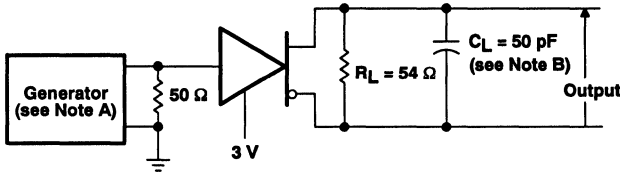


Figure 2. Receiver  $V_{OH}$  and  $V_{OL}$



TEST CIRCUIT

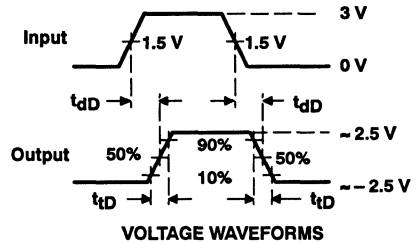
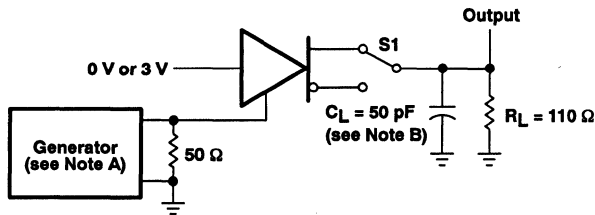


Figure 3. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT

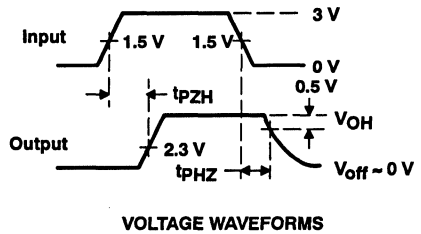
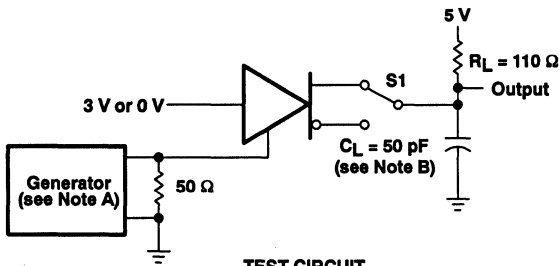


Figure 4. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT

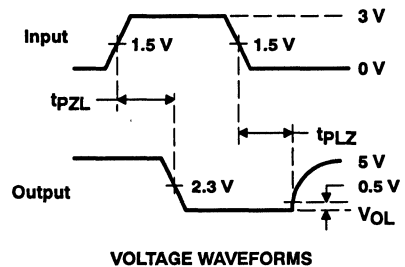


Figure 5. Driver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION

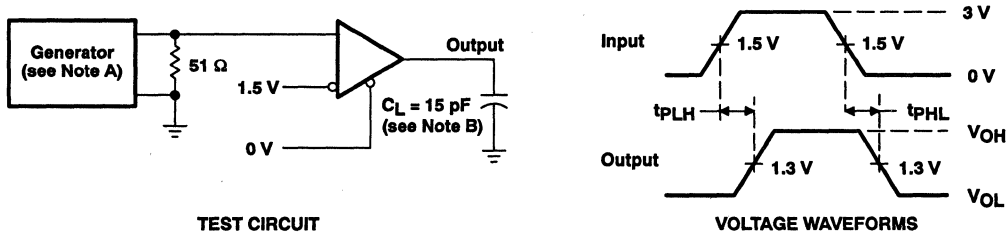


Figure 6. Receiver Test Circuit and Voltage Waveforms

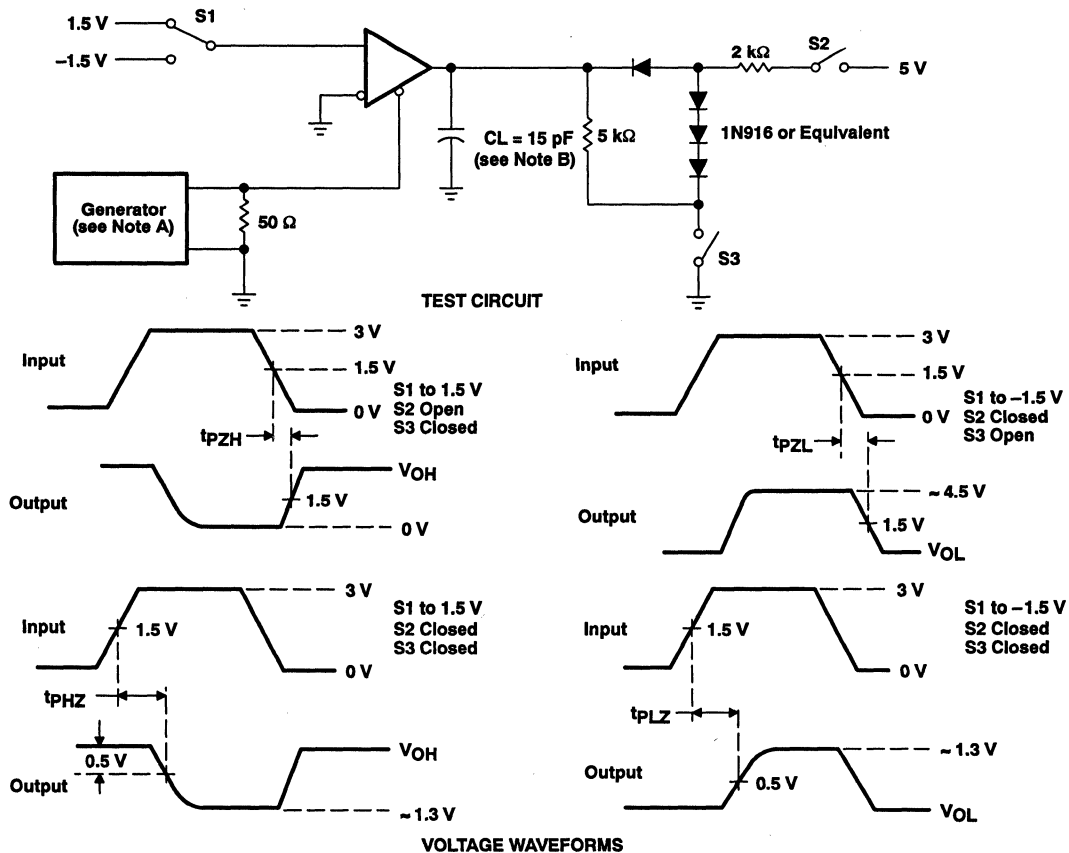


Figure 7. Receiver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

TYPICAL CHARACTERISTICS

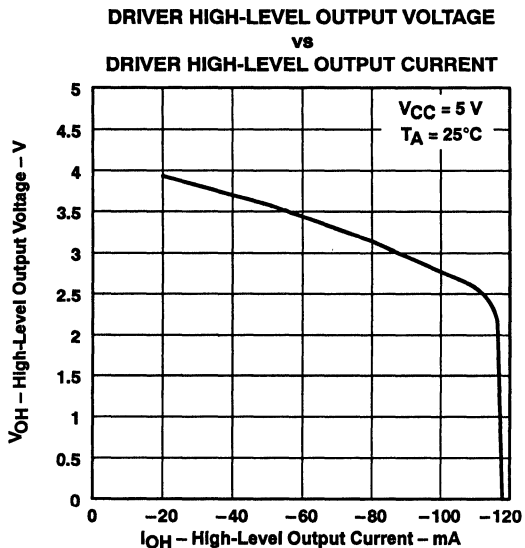


Figure 8

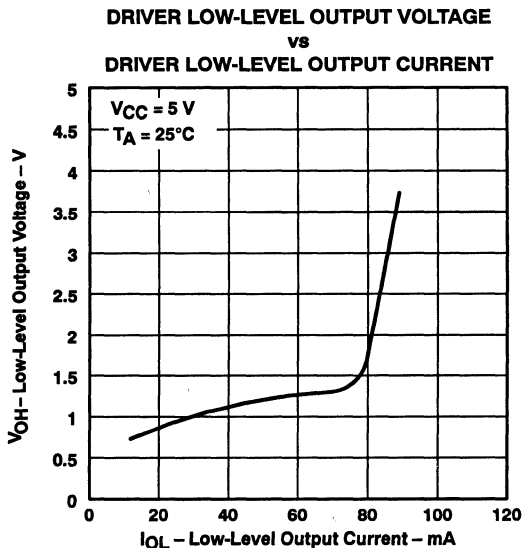


Figure 9

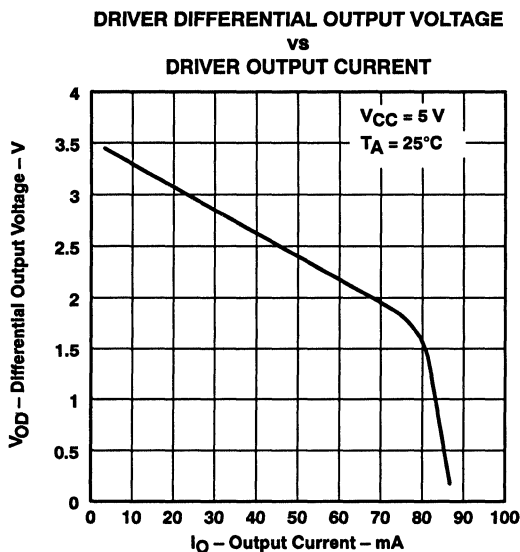


Figure 10



**SN65176B, SN75176B**  
**DIFFERENTIAL BUS TRANSCEIVERS**

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**TYPICAL CHARACTERISTICS**

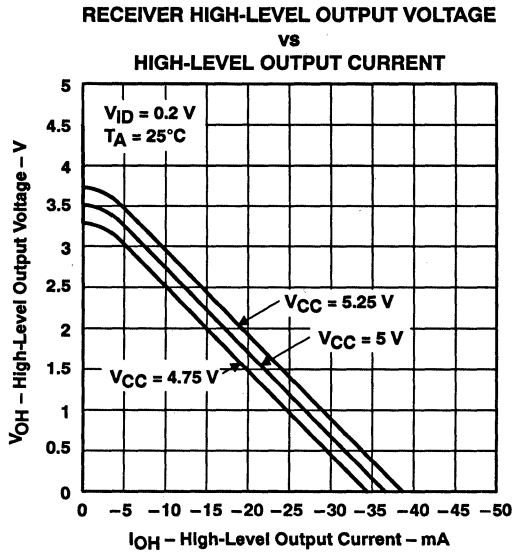


Figure 11

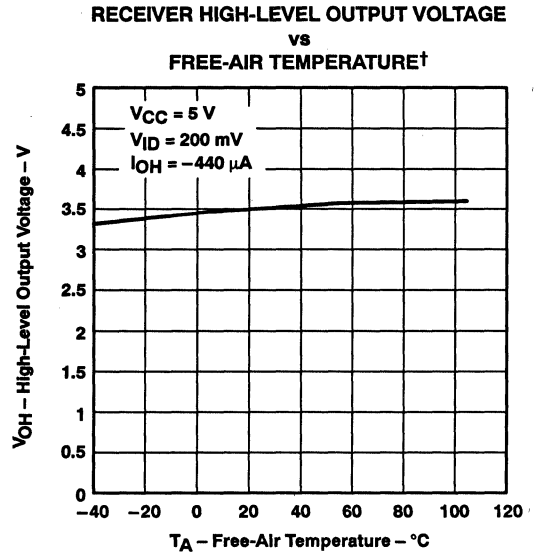


Figure 12

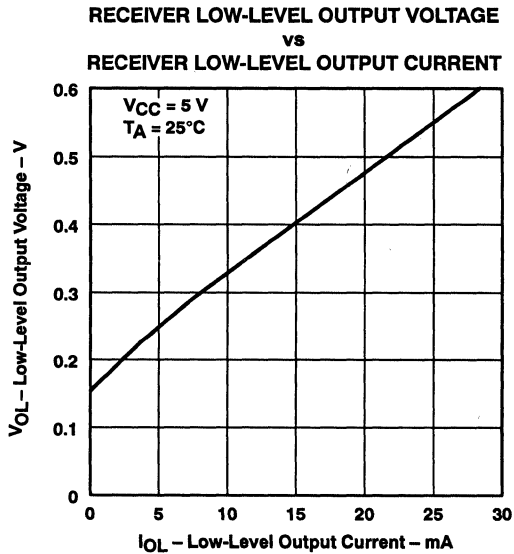


Figure 13

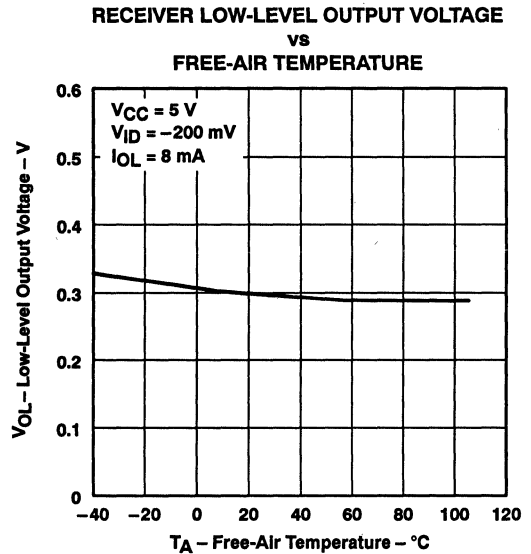


Figure 14

† Only the 0°C to 70°C portion of the curve applies to the SN75176B.

# SN65176B, SN75176B DIFFERENTIAL BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

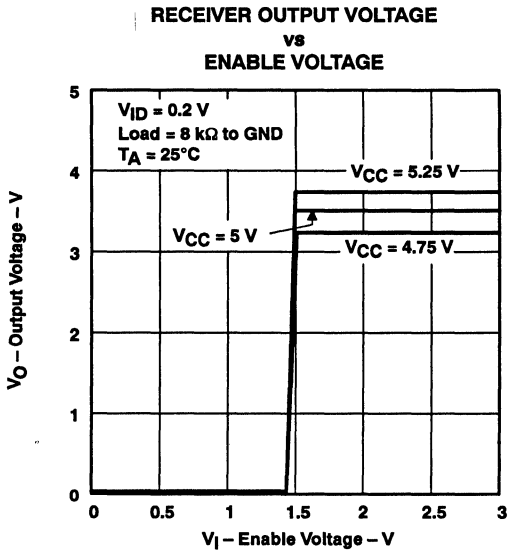


Figure 15

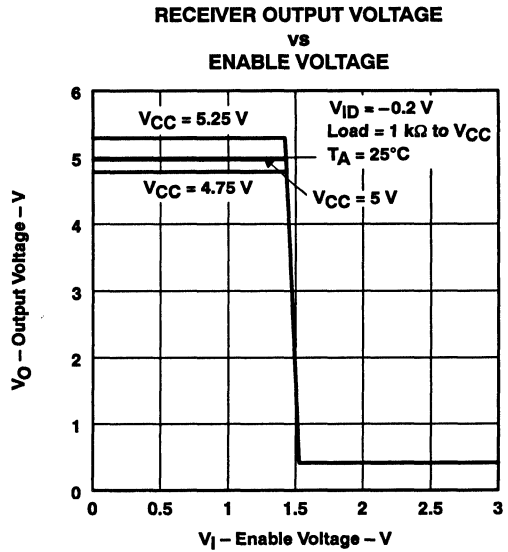


Figure 16

## APPLICATION INFORMATION

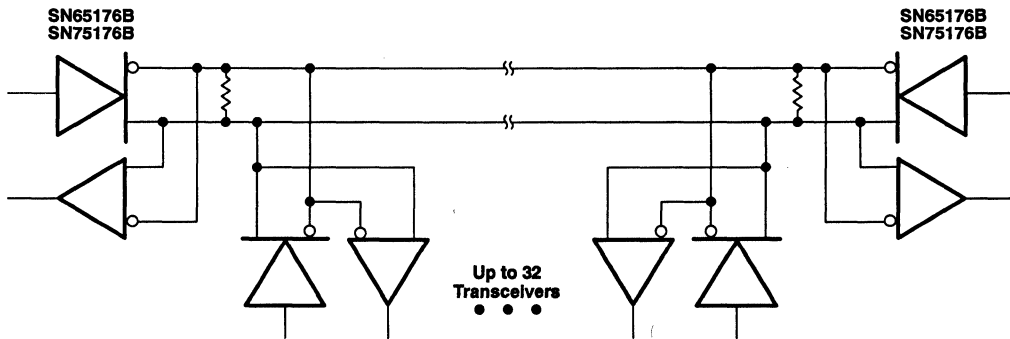


Figure 17. Typical Application Circuit

NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.



# SN95176B DIFFERENTIAL BUS TRANSCEIVER

SGLS026 – D3272, MARCH 1989

- Bidirectional Transceiver
- Suitable for Most EIA Standards RS-422-A and RS-485 Applications
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- 3-State Driver and Receiver Outputs
- Individual Driver and Receiver Enables
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capability . . .  $\pm 60$  mA Max
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From Single 5-V Supply
- Low Power Requirements

## description

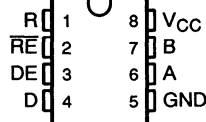
The SN95176B differential bus transceiver is a monolithic integrated circuit designed for bi-directional data communication on multipoint bus transmission lines. The transceiver is suitable for most RS-422-A and RS-485 applications to the extent of the specified data sheet characteristics and operating conditions.

The SN95176B combines a 3-state differential line driver and a differential input line receiver, both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, that can be externally connected together to function as a direction control. The driver differential outputs and the receiver differential inputs are connected internally to form differential input/output (I/O) bus ports that are designed to offer minimum loading to the bus whenever the driver is disabled or  $V_{CC} = 0$ . These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

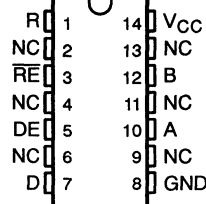
The driver is designed to handle loads up to 60 mA of sink or source current. The driver features positive- and negative-current limiting and thermal shutdown for protection from line fault conditions. Thermal shutdown is designed to occur at a junction temperature of approximately 150°C. The receiver features a minimum input impedance of 12 k $\Omega$ , an input sensitivity of  $\pm 200$  mV, and a typical input hysteresis of 50 mV.

The SN95176B is characterized for operation from  $-40^{\circ}\text{C}$  to  $110^{\circ}\text{C}$ .

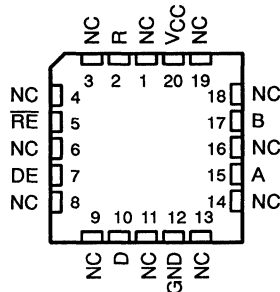
JG PACKAGE  
(TOP VIEW)



W PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC – No internal connection

# SN95176B DIFFERENTIAL BUS TRANSCEIVER

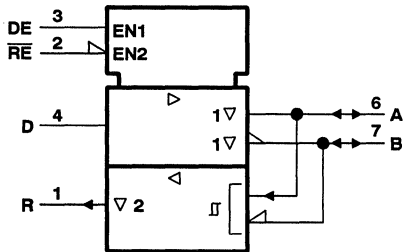
SGLS026 - D3272, MARCH 1989

## Function Tables

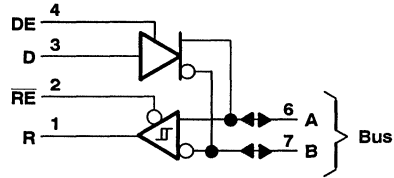
DRIVER				RECEIVER		
INPUT D	ENABLE DE	OUTPUTS A B		DIFFERENTIAL INPUTS A - B	ENABLE RE	OUTPUT R
H	H	H	L	$V_{ID} \geq 0.2V$	L	H
L	H	L	H	$-0.2V < V_{ID} < 0.2V$	L	?
X	L	Z	Z	$V_{ID} \leq -0.2V$	L	L
				X	H	Z

H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = high impedance (off)

## logic symbol†

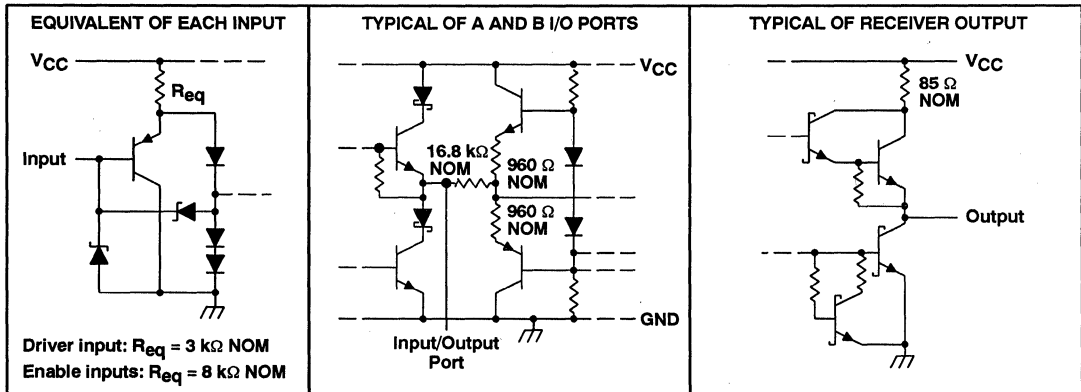


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12. Pin numbers shown are for the JG package.

## schematics of inputs and outputs



# SN95176B DIFFERENTIAL BUS TRANSCEIVER

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage at any bus terminal	–10 V to 15 V
Enable input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	–40°C to 110°C
Storage temperature range	–65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or W package	300°C

NOTE 1: All voltage values, except differential input/output bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 110^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	440 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	336 mW
W	1000 mW	8.0 mW/°C	640 mW	520 mW	320 mW

## recommended operating conditions

		MIN	TYP	MAX	UNIT	
Supply voltage, $V_{CC}$		4.75	5	5.25	V	
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$					12	V
					–7	
High-level input voltage, $V_{IH}$	D, DE, and $\overline{RE}$	2			V	
Low-level input voltage, $V_{IL}$	D, DE, and $\overline{RE}$				0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)					±12	V
High-level output current, $I_{OH}$	Driver				–60	mA
	Receiver				–400	µA
Low-level output current, $I_{OL}$	Driver				60	mA
	Receiver				8	
Operating free-air temperature, $T_A$		–40			110	°C

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.

# SN95176B DIFFERENTIAL BUS TRANSCEIVER

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## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT	
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$				-1.5	V	
$V_O$	Output voltage	$I_O = 0$		0		6	V	
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5		6	V	
$ V_{OD2} $	Differential output voltage	$R_L = 100 \Omega$ ,	See Figure 1	2			V	
		$R_L = 54 \Omega$ ,	See Figure 1	1.5	2.5	5	V	
$V_{OD3}$	Differential output voltage	See Note 3			4		V	
$\Delta V_{OD} $	Change in magnitude of differential output voltage§	$R_L = 54 \Omega$ , See Figure 1				$\pm 0.2$	V	
$V_{OC}$	Common-mode output voltage					3		V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage§							$\pm 0.2$
$I_O$	Output current	Output disabled, See Note 4	$V_O = 12 \text{ V}$			1	mA	
			$V_O = -7 \text{ V}$			-0.8		
$I_{IH}$	High-level input current	$V_I = 2.4 \text{ V}$				20	$\mu\text{A}$	
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$				-400	$\mu\text{A}$	
$I_{OS}$	Short-circuit output current	$V_O = -7 \text{ V}$				-250	mA	
		$V_O = 0$				-150		
		$V_O = V_{CC}$				250		
		$V_O = 12 \text{ V}$				250		
$I_{CC}$	Supply current (total package)	No load	Outputs enabled		42	70	mA	
			Outputs disabled		26	35		

† The power-off measurement in EIA Standard RS-422-A applies to disabled outputs only and is not applied to combined inputs and outputs.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

§  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

NOTES: 3. See EIA Standard RS-485 Figure 3.5, Test Termination Measurement 2.

4. This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions. The RS-422-A limit does not apply for a combined driver and receiver terminal.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{dD}$	Differential-output delay time	$R_L = 54 \Omega$ , See Figure 3		15	22	ns
$t_{tD}$	Differential-output transition time			20	30	ns
$t_{PZH}$	Output enable time to high level	$R_L = 110 \Omega$ , See Figure 4		85	120	ns
$t_{PZL}$	Output enable time to low level	$R_L = 110 \Omega$ , See Figure 5		40	60	ns
$t_{PHZ}$	Output disable time from high level	$R_L = 110 \Omega$ , See Figure 4		150	250	ns
$t_{PLZ}$	Output disable time from low level	$R_L = 110 \Omega$ , See Figure 5		20	30	ns

# SN95176B DIFFERENTIAL BUS TRANSCEIVER

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## SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{oa}, V_{ob}$	$V_{oa}, V_{ob}$
$ V_{OD1} $	$V_o$	$V_o$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$\Delta V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

## RECEIVER SECTION

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going input threshold voltage	$V_O = 2.7 \text{ V}$ , $I_O = -0.4 \text{ mA}$			0.2	V
$V_{T-}$ Negative-going input threshold voltage	$V_O = 0.5 \text{ V}$ , $I_O = 8 \text{ mA}$	-0.2‡			V
$V_{hys}$ Input hysteresis ( $V_{T+} - V_{T-}$ )			50		mV
$V_{IK}$ Enable clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V
$V_{OH}$ High-level output voltage	$V_{ID} = 200 \text{ mV}$ , See Figure 2 $I_{OH} = -400 \mu\text{A}$ ,	2.7			V
$V_{OL}$ Low-level output voltage	$V_{ID} = -200 \text{ mV}$ , See Figure 2 $I_{OL} = 8 \text{ mA}$ ,			0.45	V
$I_{OZ}$ High-impedance-state output current	$V_O = 0.4 \text{ V to } 2.4 \text{ V}$			$\pm 20$	$\mu\text{A}$
$I_I$ Line input current	Other input = 0 V, See Note 5 $V_I = 12 \text{ V}$ $V_I = -7 \text{ V}$			1 -0.8	mA
$I_{IH}$ High-level enable input current	$V_{IH} = 2.7 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$ Low-level enable input current	$V_{IL} = 0.4 \text{ V}$			-100	$\mu\text{A}$
$r_i$ Input resistance	$V_I = 12 \text{ V}$		12		k $\Omega$
$I_{OS}$ Short-circuit output current		-15		-85	mA
$I_{CC}$ Supply current (total package)	No load		42	70	mA
			26	35	

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 5: This applies for both power on and power off. Refer to EIA Standard RS-485 for exact conditions.





# SN95176B DIFFERENTIAL BUS TRANSCEIVER

SGLS026 - D3272, MARCH 1989

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 15\text{ pF}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$V_{ID} = 0$ to 3 V, See Figure 6		21	35	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			23	35	ns
$t_{PZH}$ Output enable time to high level	See Figure 7		10	20	ns
$t_{PZL}$ Output enable time to low level			12	20	ns
$t_{PHZ}$ Output disable time from high level	See Figure 7		20	35	ns
$t_{PLZ}$ Output disable time from low level			17	25	ns

  
TEXAS  
INSTRUMENTS

PARAMETER MEASUREMENT INFORMATION

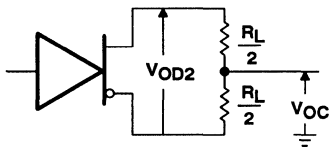


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

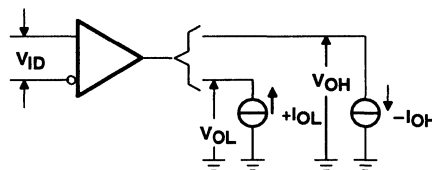
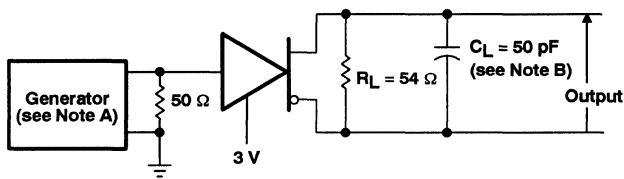
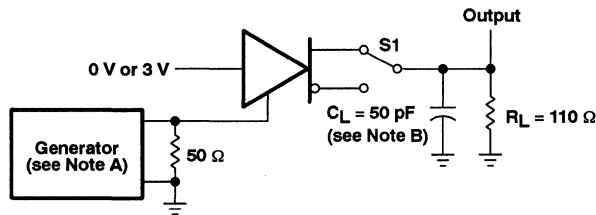


Figure 2. Receiver  $V_{OH}$  and  $V_{OL}$



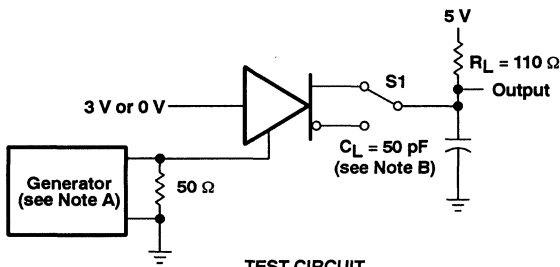
TEST CIRCUIT

Figure 3. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT

Figure 4. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT

Figure 5. Driver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.  
C. Equivalent test circuits may be substituted for actual testing.

# SN95176B DIFFERENTIAL BUS TRANSCEIVER

SGLS026 – D3272, MARCH 1989

## PARAMETER MEASUREMENT INFORMATION

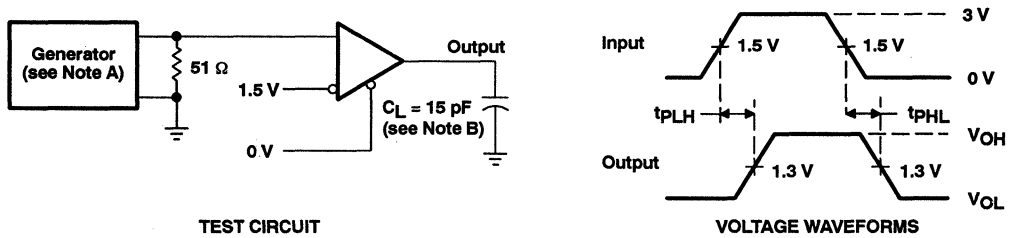


Figure 6. Receiver Test Circuit and Voltage Waveforms

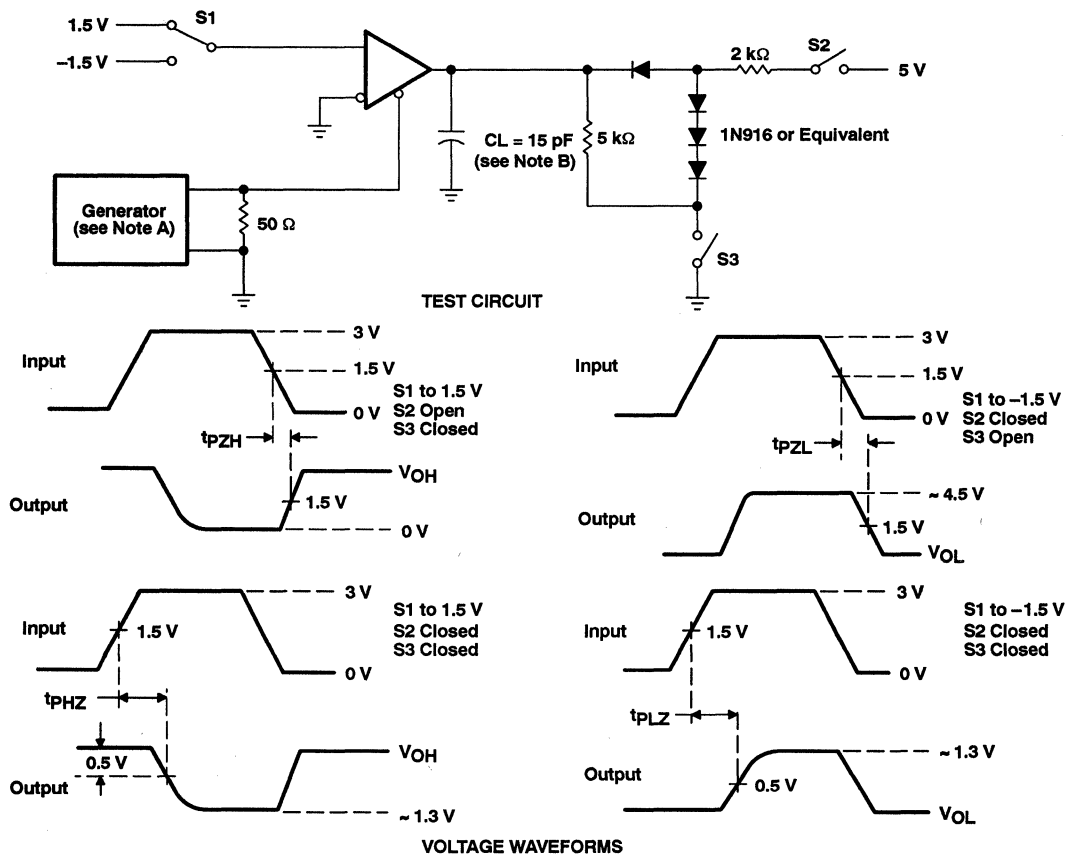


Figure 7. Receiver Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. Equivalent test circuits may be substituted for actual testing.

TYPICAL CHARACTERISTICS

DRIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
DRIVER HIGH-LEVEL OUTPUT CURRENT

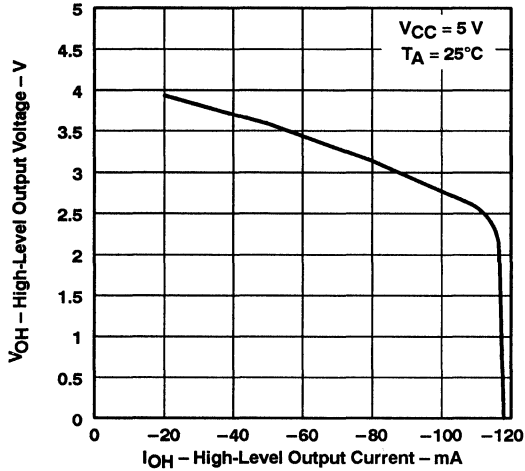


Figure 8

DRIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
DRIVER LOW-LEVEL OUTPUT CURRENT

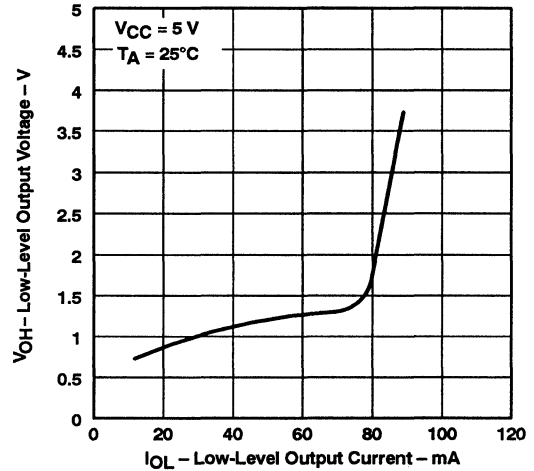


Figure 9

DRIVER DIFFERENTIAL OUTPUT VOLTAGE  
vs  
DRIVER OUTPUT CURRENT

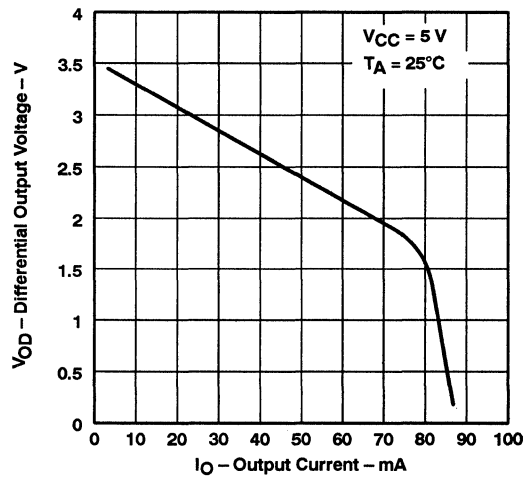


Figure 10

# SN95176B DIFFERENTIAL BUS TRANSCEIVER

SGLS026 - D3272, MARCH 1989

## TYPICAL CHARACTERISTICS

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

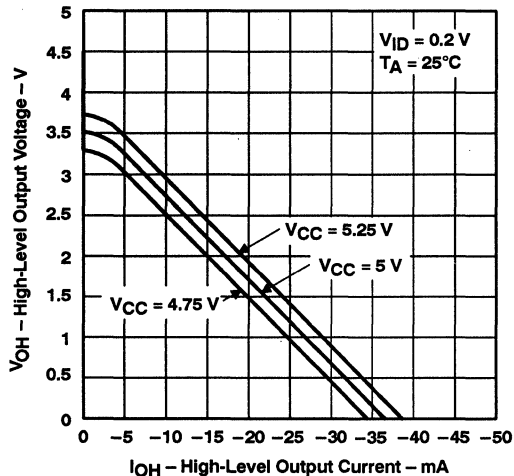


Figure 11

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE†

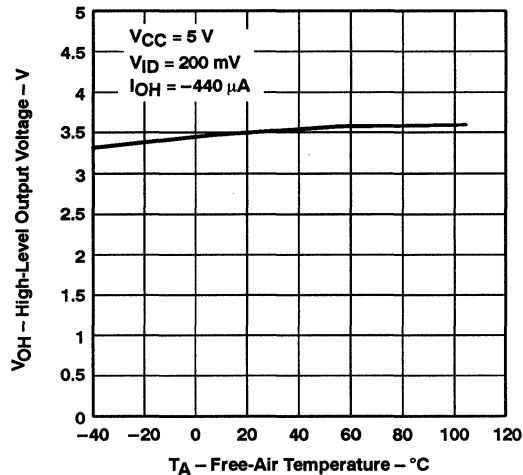


Figure 12

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
RECEIVER LOW-LEVEL OUTPUT CURRENT

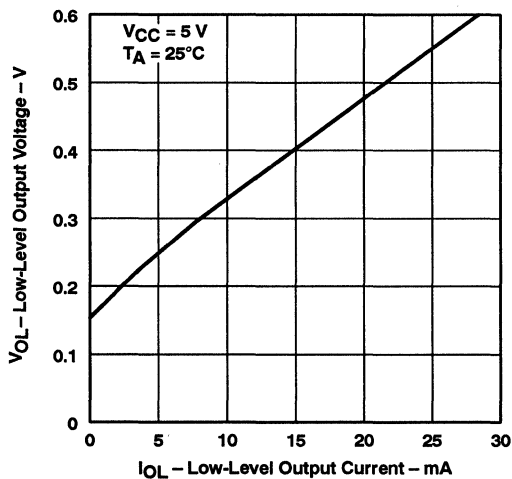


Figure 13

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

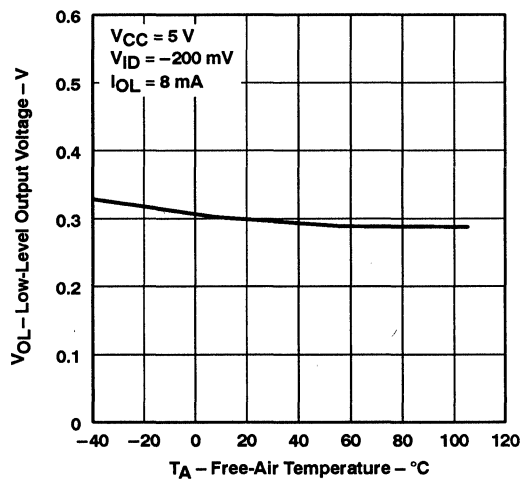


Figure 14



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TYPICAL CHARACTERISTICS

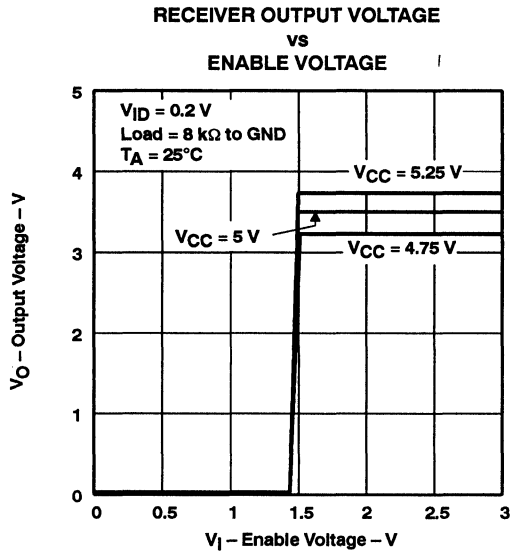


Figure 15

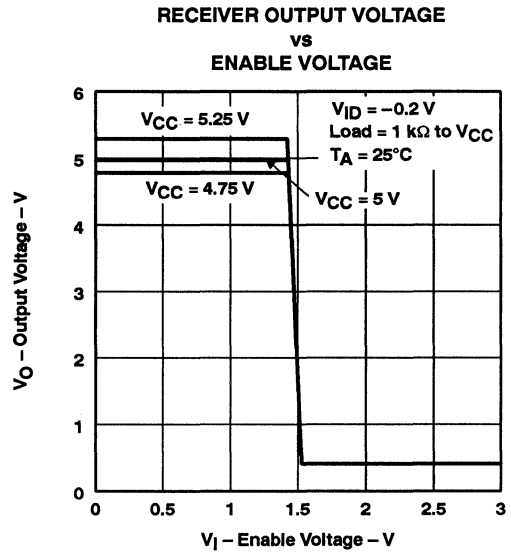


Figure 16

APPLICATION INFORMATION

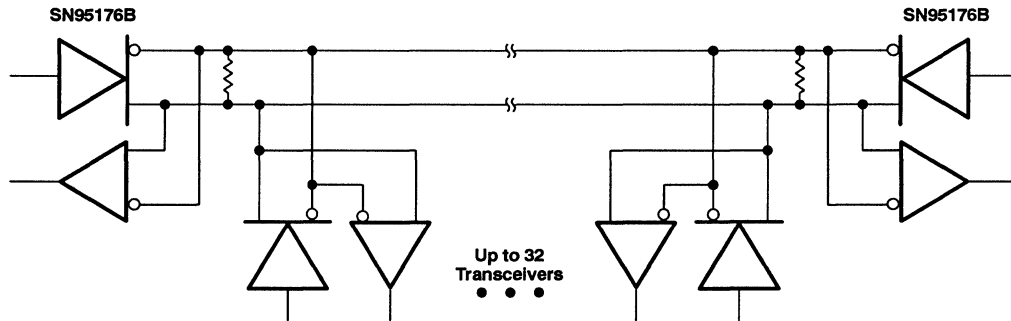


Figure 17. Typical Application Circuit

NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.



# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D – D3042, AUGUST 1987 – REVISED AUGUST 1991

- Meets EIA Standards RS-422A and RS-485 and CCITT Recommendations V.11 and X.27
- Designed and Tested for Data Rates up to 35 MBaud
- SN65ALS176 Operating Temperature  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$
- Three Skew Limits Available:
  - 'ALS176 . . . 10 ns
  - 'ALS176A . . . 7.5 ns
  - 'ALS176B . . . 5 ns
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Low Supply Current Requirements  
30 mA Max
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Hysteresis
- Glitch-Free Power-Up and Power-Down Protection
- Receiver Open-Circuit Fail-Safe Design

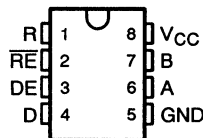
## description

The SN65ALS176 and SN75ALS176 series differential bus transceivers are monolithic integrated circuits designed for bidirectional data communication on multipoint bus transmission lines. They are designed for balanced transmission lines and meet EIA Standards RS-422-A and RS-485 and CCITT recommendations V.11 and X.27.

The SN65ALS176 and SN75ALS176 series combine a 3-state differential line driver and a differential input line receiver, both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, which can be externally connected together to function as a direction control. The driver differential outputs and the receiver differential inputs are connected internally to form a differential input/output (I/O bus port that is designed to offer minimum loading to the bus whenever the driver is disabled or  $V_{CC} = 0$ . This port features wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

The SN65ALS176 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ , and the SN75ALS176 series is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

D OR P PACKAGE  
(TOP VIEW)



Function Tables

DRIVER

INPUT D	ENABLE DE	OUTPUTS	
		A	B
H	H	H	L
L	H	L	H
X	L	Z	Z

RECEIVER

DIFFERENTIAL INPUTS A-B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2\text{ V}$	L	H
$-0.2\text{ V} < V_{ID} < 0.2\text{ V}$	L	?
$V_{ID} \leq -0.2\text{ V}$	L	L
X	H	Z
Inputs open	L	H

H = high level, L = low level, ? = indeterminate,  
X = irrelevant, Z = high impedance (off)

AVAILABLE OPTIONS

$T_A$	$t_{sk}(\text{LIM})^\ddagger$	PACKAGE	
		SMALL OUTLINE (D)†	PLASTIC DIP (P)
$0^{\circ}\text{C}$ to $70^{\circ}\text{C}$	10 7.5 5	SN75ALS176D SN75ALS176AD SN75ALS176BD	SN75ALS176P SN75ALS176AP SN75ALS176BP
$-40^{\circ}\text{C}$ to $85^{\circ}\text{C}$	10	SN65ALS176D	SN65ALS176P

† The D package is available taped and reeled. Add the suffix R to the device type, (e.g., SN75ALS176DR).

‡  $t_{sk}(\text{LIM})$  is the greater of 1) the difference between the maximum and minimum specified values of  $t_{PLH}$  (or  $t_{DHL}$ ), and 2) the difference between the maximum and minimum specified values of  $t_{PHL}$  (or  $t_{DDL}$ ). This is the maximum range that the driver or receiver delay time will vary over temperature,  $V_{CC}$ , and device to device.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

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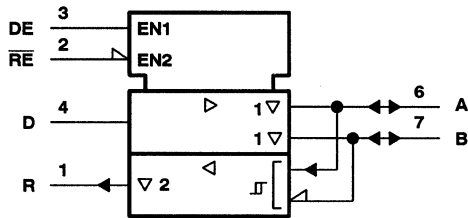
Copyright © 1991, Texas Instruments Incorporated



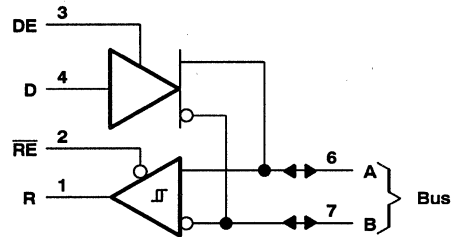
# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D - D3042, AUGUST 1987 - REVISED AUGUST 1991

## logic symbol†

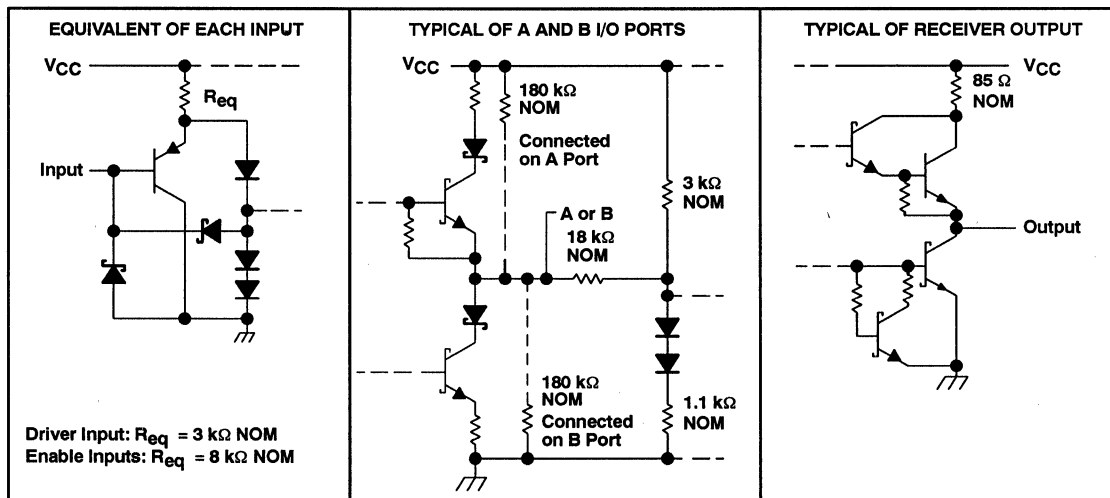


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematics of inputs and outputs



# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D – D3042, AUGUST 1987 – REVISED AUGUST 1991

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Voltage range at any bus terminal .....	–7 V to 12 V
Enable input voltage .....	5.5 V
Continuous total power dissipation .....	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : SN65ALS176 .....	–40°C to 85°C
SN75ALS176 series .....	0°C to 70°C
Storage temperature range .....	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

NOTE 1: All voltage values, except differential I/O bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Input voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$		12			V
		–7			
High-level input voltage, $V_{IH}$	D, DE, and $\overline{RE}$	2			V
Low-level input voltage, $V_{IL}$	D, DE, and $\overline{RE}$	0.8			V
Differential input voltage, $V_{ID}$ (see Note 2)		±12			V
High-level output current, $I_{OH}$	Driver	–60			mA
	Receiver	–400			µA
Low-level output current, $I_{OL}$	Driver	60			mA
	Receiver	8			
Operating free-air temperature, $T_A$	SN65ALS176	–40			°C
	SN75ALS176	0	70		

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.

# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D – D3042, AUGUST 1987 – REVISED AUGUST 1991

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$				-1.5	V
$V_O$	Output voltage	$I_O = 0$		0		6	V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5		6	V
$ V_{OD2} $	Differential output voltage	$R_L = 100 \Omega$ ,	See Figure 1	$1/2 V_{OD1}$ or 2‡			V
		$R_L = 54 \Omega$ ,	See Figure 1	1.5	2.5	5	V
$V_{OD3}$	Differential output voltage	$V_{test} = -7 \text{ V to } 12 \text{ V}$ , See Figure 2		1.5		5	V
$\Delta  V_{OD} $	Change in magnitude of differential output voltage§	$R_L = 54 \Omega \text{ or } 100 \Omega$ , See Figure 1				$\pm 0.2$	V
$V_{OC}$	Common-mode output voltage					3 -1	V
$\Delta  V_{OC} $	Change in magnitude of common-mode output voltage§					$\pm 0.2$	V
$I_O$	Output current	Outputs disabled, See Note 3	$V_O = 12 \text{ V}$			1	mA
			$V_O = -7 \text{ V}$			-0.8	
$I_{IH}$	High-level input current	$V_I = 2.4 \text{ V}$				20	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$				-400	$\mu\text{A}$
$I_{OS}$	Short-circuit output current	$V_O = -4 \text{ V}$	SN65ALS176			-250	mA
		$V_O = -6 \text{ V}$	SN75ALS176				
		$V_O = 0$				-150	
		$V_O = V_{CC}$				250	
		$V_O = 8 \text{ V}$					
$I_{CC}$	Supply current	No load	Outputs enabled		23	30	mA
			Outputs disabled		19	26	

† The power-off measurement in EIA Standard RS-422-A applies to disabled outputs only and is not applied to combined inputs and outputs.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

§  $\Delta |V_{OD}|$  and  $\Delta |V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from one logic state to the other.

¶ The minimum  $V_{OD2}$  with a 100- $\Omega$  load is either  $1/2 V_{OD1}$  or 2 V, whichever is greater.

NOTE 3: This applies for both power on and power off; refer to EIA standard RS-485 for exact conditions. The RS-422-A limit does not apply for a combined driver and receiver terminal.



# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D – D3042, AUGUST 1987 – REVISED AUGUST 1991

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

## SN65ALS176

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{dD}$	Differential output delay time	$R_L = 54 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 3			15	ns
$t_{sk(p)}$	Pulse skew ( $ t_{dDL} - t_{dDH} $ )			0	2	ns
$t_{TD}$	Differential output transition time			8		ns
$t_{PZH}$	Output enable time to high level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 4			80	ns
$t_{PZL}$	Output enable time to low level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 5			30	ns
$t_{PHZ}$	Output disable time from high level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 4			50	ns
$t_{PLZ}$	Output disable time from low level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 5			30	ns

## SN75ALS176, SN75ALS176A, SN75ALS176B

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$t_{dD}$	Differential output delay time	$R_L = 54 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 3	'ALS176	3	8	13	ns
			'ALS176A	4	7	11.5	
			'ALS176B	5	8	10	
$t_{sk(p)}$	Pulse skew ( $ t_{dDL} - t_{dDH} $ )			0	2	ns	
$t_{TD}$	Differential output transition time			8		ns	
$t_{PZH}$	Output enable time to high level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 4			23	ns	
$t_{PZL}$	Output enable time to low level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 5			14	ns	
$t_{PHZ}$	Output disable time from high level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 4			20	ns	
$t_{PLZ}$	Output disable time from low level	$R_L = 110 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 5			8	ns	

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

### SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{Oa}, V_{Ob}$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_o$	$V_o$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$\Delta  V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta  V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$



# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D – D3042, AUGUST 1987 – REVISED AUGUST 1991

## RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7\text{ V}$ , $I_O = -0.4\text{ mA}$			0.2	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5\text{ V}$ , $I_O = 8\text{ mA}$	-0.2‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )			60		mV
$V_{IK}$	Enable-input clamp voltage	$I_I = -18\text{ mA}$			-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 200\text{ mV}$ , See Figure 6 $I_{OH} = -400\text{ }\mu\text{A}$ ,	2.7			V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200\text{ mV}$ , See Figure 6 $I_{OL} = 8\text{ mA}$ ,			0.45	V
$I_{OZ}$	High-impedance-state output current	$V_O = 0.4\text{ V to } 2.4\text{ V}$			$\pm 20$	$\mu\text{A}$
$I_I$	Line input current	Other input = 0 V, $V_I = 12\text{ V}$ See Note 4 $V_I = -7\text{ V}$			1 -0.8	mA
$I_{IH}$	High-level-enable input current	$V_{IH} = 2.7\text{ V}$			20	$\mu\text{A}$
$I_{IL}$	Low-level-enable input current	$V_{IL} = 0.4\text{ V}$			-100	$\mu\text{A}$
$r_i$	Input resistance		12	20		k $\Omega$
$I_{OS}$	Short-circuit output current	$V_{ID} = 200\text{ mV}$ , $V_O = 0$	-15		-85	mA
$I_{CC}$	Supply current	No load				mA
		Outputs enabled		23	30	
		Outputs disabled		19	26	

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 4: This applies for both power on and power off. Refer to EIA Standard RS-485 for exact conditions.

switching characteristics over recommended ranges of supply voltage and operating free-air temperature

### SN65ALS176

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{pd}$	Propagation time	$V_{ID} = -1.5\text{ V to } 1.5\text{ V}$ , $C_L = 15\text{ pF}$ , See Figure 7			25	ns
$t_{sk(p)}$	Pulse skew ( $ t_{pHL} - t_{pLH} $ )			0	2	ns
$t_{pZH}$	Output enable time to high level	$C_L = 15\text{ pF}$ , See Figure 8		11	18	ns
$t_{pZL}$	Output enable time to low level			11	18	ns
$t_{pHZ}$	Output disable time from high level				50	ns
$t_{pLZ}$	Output disable time from low level				30	ns

### SN75ALS176, SN75ALS176A, SN75ALS176B

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$t_{pd}$	Propagation time	$V_{ID} = -1.5\text{ V to } 1.5\text{ V}$ , $C_L = 15\text{ pF}$ , See Figure 7		9	14	19	ns
			'ALS176A	10.5	14	18	
			'ALS176B	11.5	13	16.5	
$t_{sk(p)}$	Pulse skew ( $ t_{pHL} - t_{pLH} $ )	$C_L = 15\text{ pF}$ , See Figure 8		0	2	ns	
$t_{pZH}$	Output enable time to high level			7	14	ns	
$t_{pZL}$	Output enable time to low level			20	35	ns	
$t_{pHZ}$	Output disable time from high level			20	35	ns	
$t_{pLZ}$	Output disable time from low level			8	17	ns	

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .



# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D – D3042, AUGUST 1987 – REVISED AUGUST 1991

## PARAMETER MEASUREMENT INFORMATION

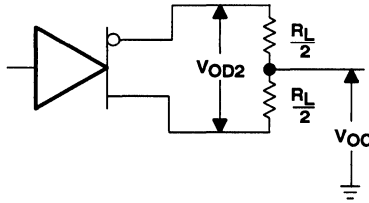


Figure 1. Driver  $V_{OD2}$  and  $V_{OC}$

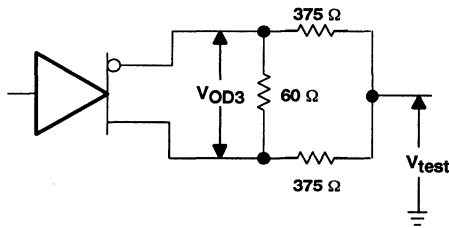
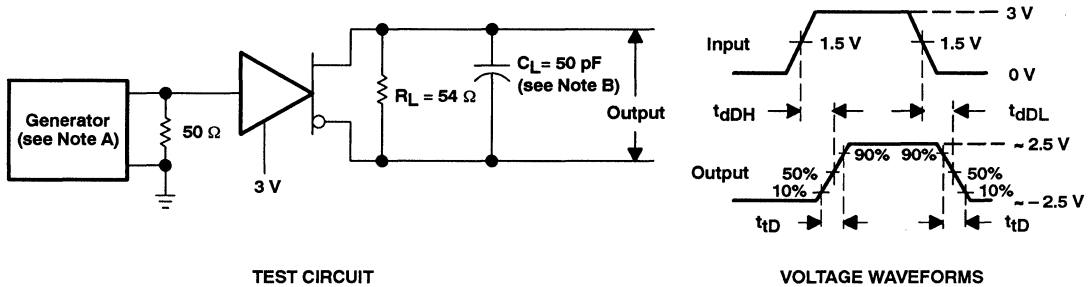


Figure 2. Driver  $V_{OD3}$



TEST CIRCUIT

VOLTAGE WAVEFORMS

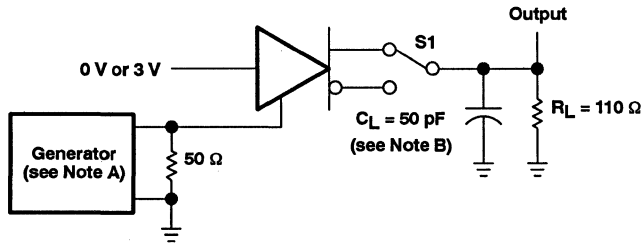
- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C.  $t_{dD} = t_{dDH}$  or  $t_{dDL}$

Figure 3. Driver Test Circuit and Voltage Waveforms

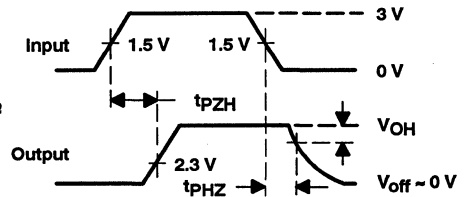
# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D - D3042, AUGUST 1987 - REVISED AUGUST 1991

## PARAMETER MEASUREMENT INFORMATION

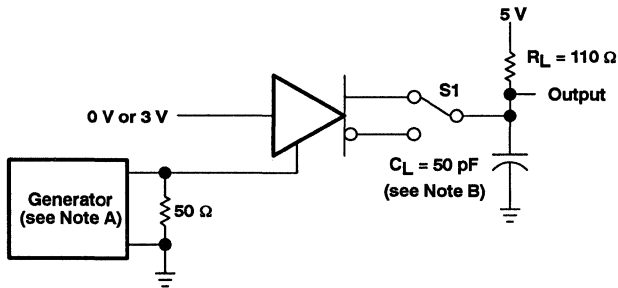


TEST CIRCUIT

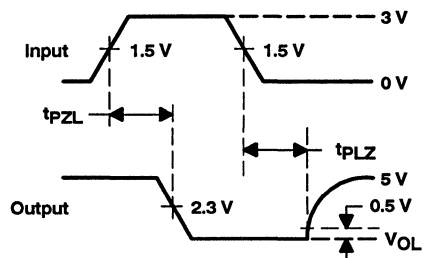


VOLTAGE WAVEFORMS

Figure 4. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT



VOLTAGE WAVEFORMS

Figure 5. Driver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

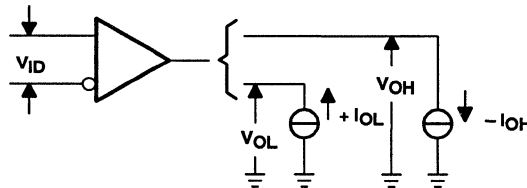


Figure 6. Receiver  $V_{OH}$  and  $V_{OL}$  Test Circuit

# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

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## PARAMETER MEASUREMENT INFORMATION

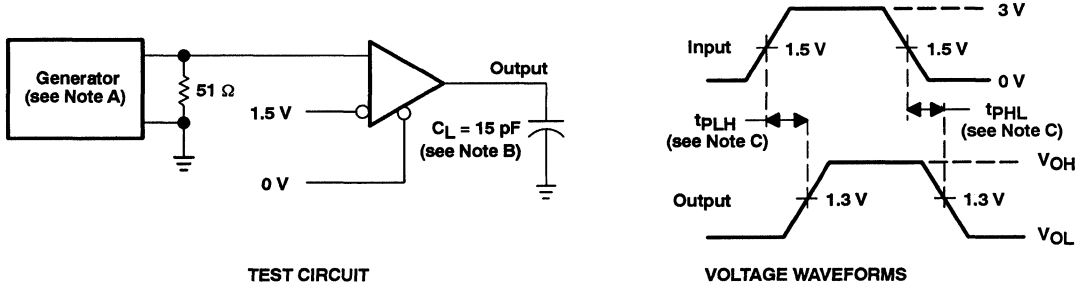


Figure 7. Receiver Test Circuit and Voltage Waveforms

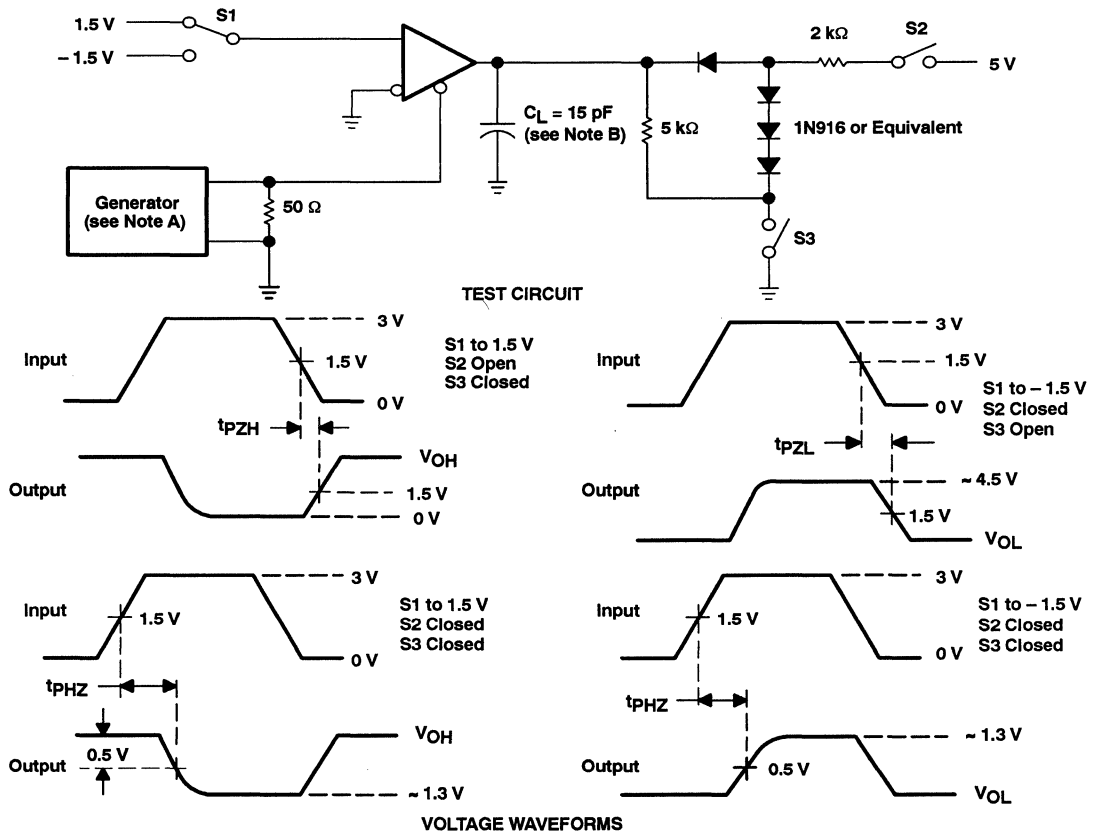


Figure 8. Receiver Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_O = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C.  $t_{pd} = t_{PLH}$  or  $t_{PHL}$



# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D - D3042, AUGUST 1987 - REVISED AUGUST 1991

## TYPICAL CHARACTERISTICS

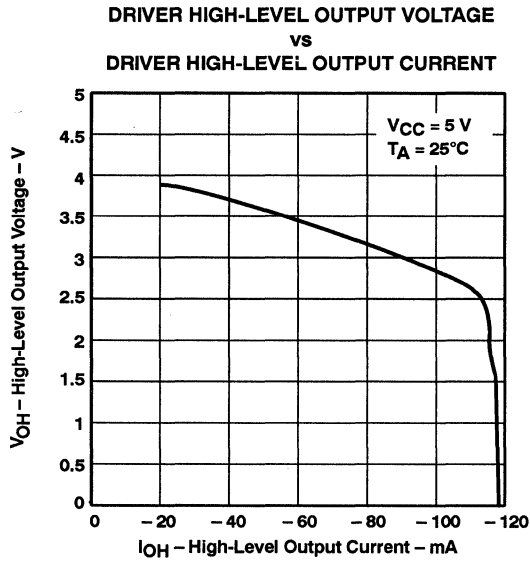


Figure 9

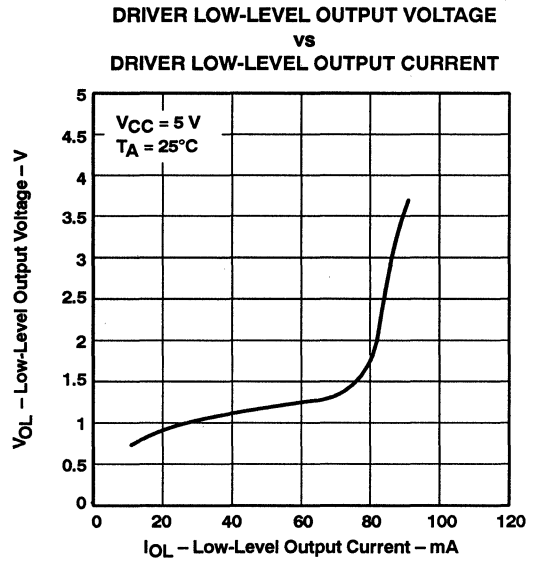


Figure 10

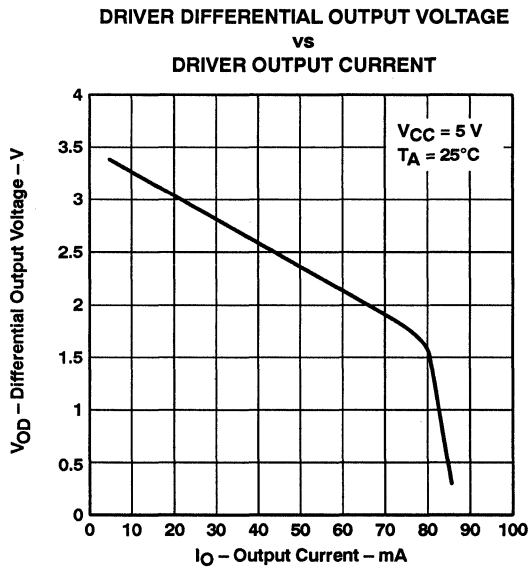


Figure 11

SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B  
DIFFERENTIAL BUS TRANSCEIVERS

SLLS040D - D3042, AUGUST 1987 - REVISED AUGUST 1991

TYPICAL CHARACTERISTICS

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

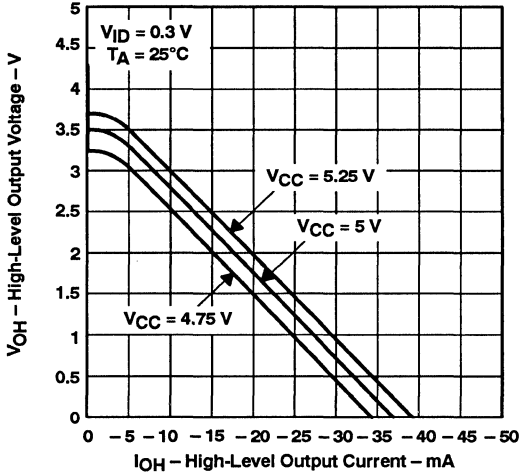


Figure 12

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

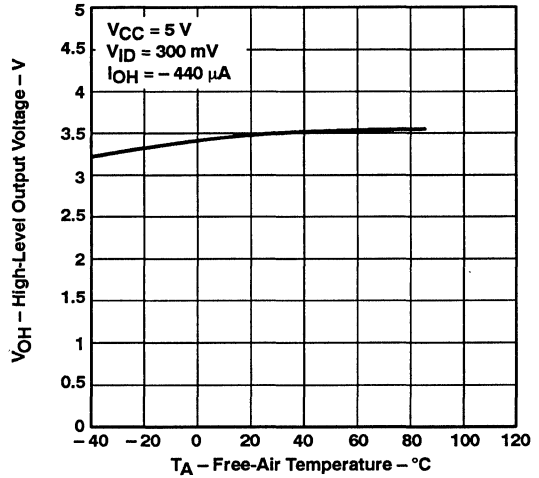


Figure 13

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
RECEIVER LOW-LEVEL OUTPUT CURRENT

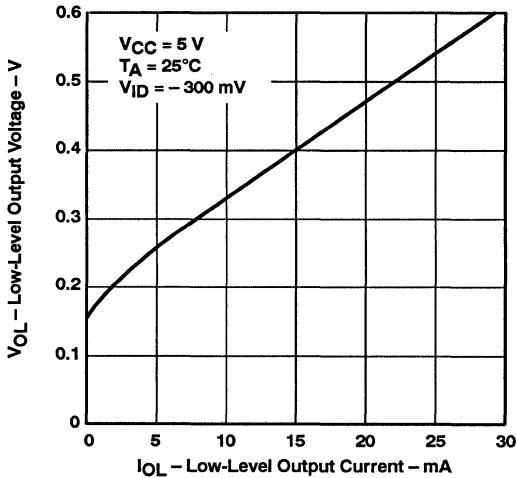


Figure 14

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

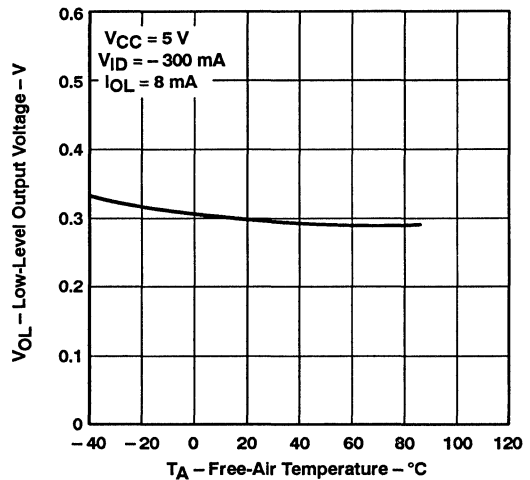


Figure 15



# SN65ALS176, SN75ALS176, SN75ALS176A, SN75ALS176B DIFFERENTIAL BUS TRANSCEIVERS

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## TYPICAL CHARACTERISTICS

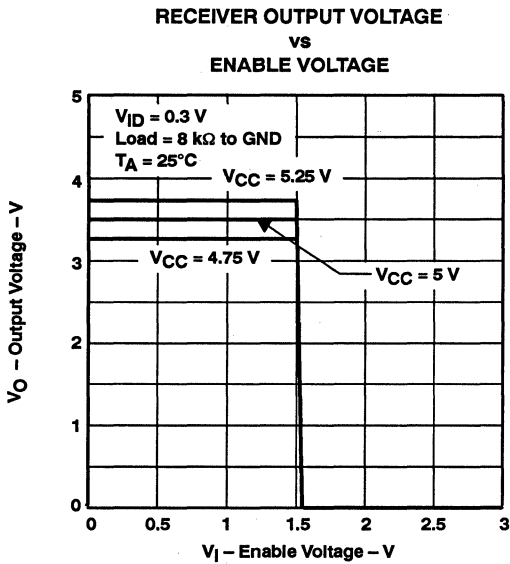


Figure 16

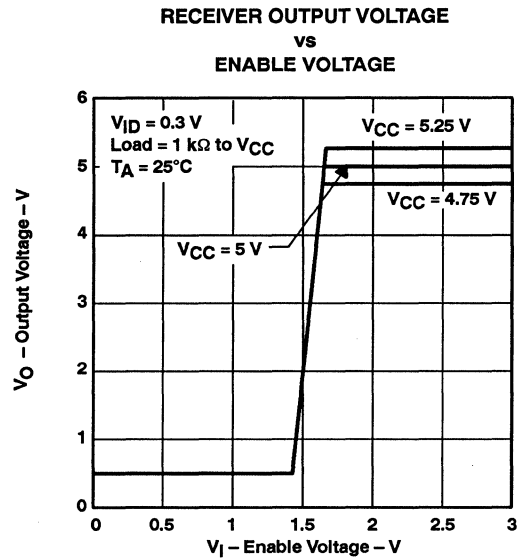
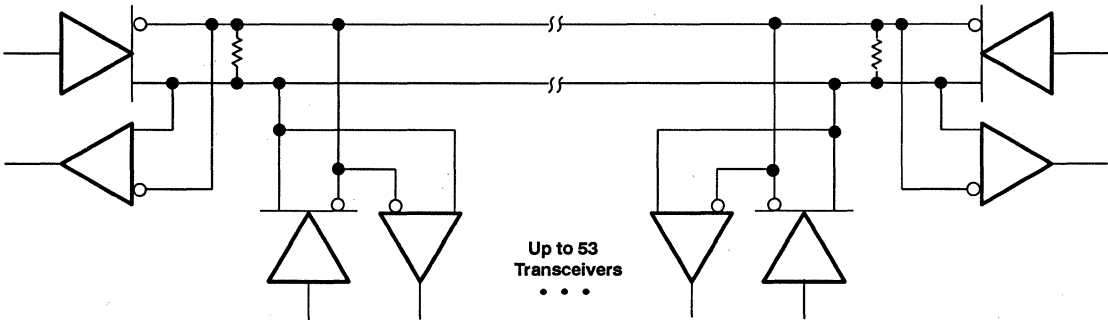


Figure 17

## APPLICATION INFORMATION



NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

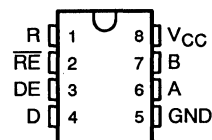
Figure 18. Typical Application Circuit

# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B – D3502, AUGUST 1990 – REVISED FEBRUARY 1993

- Bidirectional Transceiver
- Meets EIA Standard RS-485 and ISO 8482:1987(E)
- High-Speed Low-Power LinBiCMOS™ Circuitry
- Designed for High-Speed Operation in Both Serial and Parallel Applications
- Low Skew
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Very Low Disabled Supply Current Requirements . . . 200  $\mu$ A Maximum
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capacity . . .  $\pm 60$  mA
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Open-Circuit Fail-Safe Receiver Design
- Receiver Input Sensitivity . . .  $\pm 200$  mV Max
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From a Single 5-V Supply
- Glitch-Free Power-Up and Power-Down Protection

D, JG, OR P PACKAGE  
(TOP VIEW)



Function Tables

DRIVER			
INPUT D	ENABLE DE	OUTPUTS	
		A	B
H	H	H	L
L	H	L	H
X	L	Z	Z

RECEIVER		
DIFFERENTIAL INPUTS A-B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2$ V	L	H
$-0.2$ V $< V_{ID} < 0.2$ V	L	?
$V_{ID} \leq -0.2$ V	L	L
X	H	Z
Open	L	H

H = high level, L = low level, ? = indeterminate,  
X = irrelevant, Z = high impedance (off)

## description

The SN55LBC176, SN65LBC176, and SN75LBC176 differential bus transceivers are monolithic integrated circuits designed for bidirectional data communication on multipoint bus transmission lines. They are designed for balanced transmission lines and meet EIA Standard RS-485 and ISO 8482:1987(E).

The SN65LBC176 and SN75LBC176 combine a 3-state differential-line driver and a differential-input line receiver, both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, which can be externally connected together to function as a direction control. The driver differential outputs and the receiver differential inputs are connected internally to form a differential input/output (I/O) bus port that is designed to offer minimum loading to the bus whenever the driver is disabled or  $V_{CC} = 0$ . This port features wide positive and negative common-mode voltage ranges, making the device suitable for party line applications. Very low device supply current can be achieved by disabling the driver and the receiver. Both the driver and receiver are available as cells in the Texas Instruments LinASIC™ Library.

These transceivers are suitable for RS-485 and ISO 8482:1987 (E) applications to the extent that they are specific in the operating conditions and characteristics section of this data sheet. Certain limits contained in the RS-485 and ISO 8482:1987 (E) are not met or cannot be tested over the entire military temperature range.

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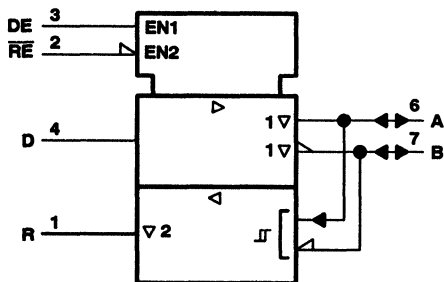
# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B - D3502, AUGUST 1990 - REVISED FEBRUARY 1993

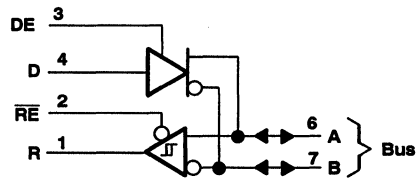
## description (continued)

The SN55LBC176 is characterized for operation from  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN65LBC176 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ , and the SN75LBC176 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## logic symbol†

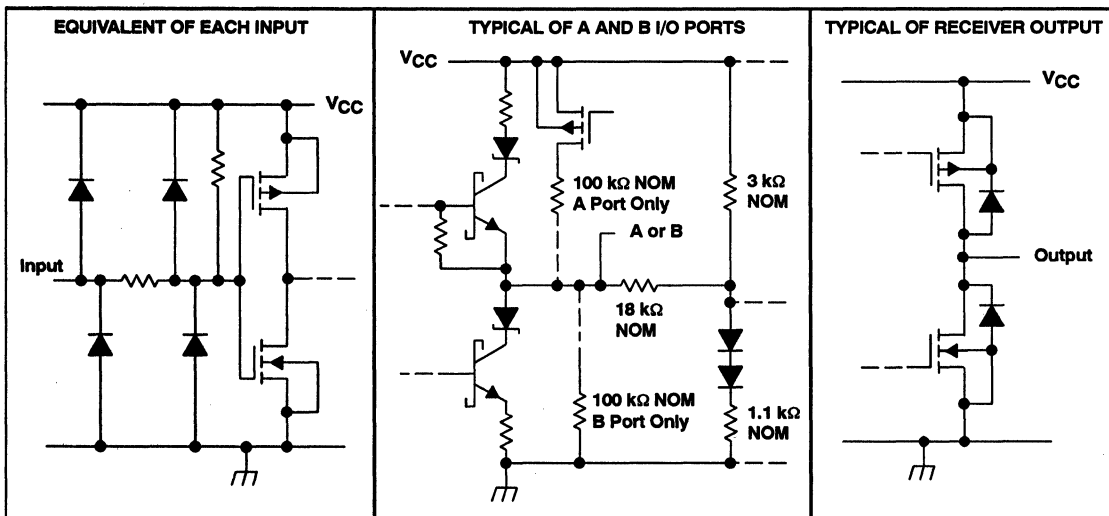


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematics of inputs and outputs



# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B – D3502, AUGUST 1990 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	–10 V to 15 V
Enable input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : SN55LBC176	–55°C to 125°C
SN65LBC176	–40°C to 85°C
SN75LBC176	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values, except differential I/O bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	—
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	336 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	—

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	SN55LBC176	4.5	5	5.5	V
	SN65/75LBC176	4.75	5	5.25	
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$				12 –7	V
High-level input voltage, $V_{IH}$	D, DE, and $\overline{RE}$	2			V
Low-level input voltage, $V_{IL}$	D, DE, and $\overline{RE}$			0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)				±12	V
High-level output current, $I_{OH}$	Driver			–60	mA
	Receiver			–400	µA
Low-level output current, $I_{OL}$	Driver			60	mA
	Receiver			8	
Operating free-air temperature, $T_A$	SN55LBC176	–55		125	°C
	SN65LBC176	–40		85	
	SN75LBC176	0		70	

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.



# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B - D3502, AUGUST 1990 - REVISED FEBRUARY 1993

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V
$V_O$	Output voltage	$I_O = 0$		0	6	V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5	6	V
$ V_{OD2} $	Differential output voltage	$R_L = 54 \Omega$ , See Note 3	See Figure 1,	55LBC176	1.1	V
				65LBC176	1.1	
				75LBC176	1.5	
$V_{OD3}$	Differential output voltage	$V_{test} = -7 \text{ V to } 12 \text{ V}$ , See Note 3	See Figure 2,	55LBC176	1.1	V
				65LBC176	1	
				75LBC176	1.5	
$\Delta V_{OD} $	Change in magnitude of differential output voltage †	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1			$\pm 0.2$	V
$V_{OC}$	Common-mode output voltage				3	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage †				$\pm 0.2$	V
$I_O$	Output current	Output disabled, See Note 4	$V_O = 12 \text{ V}$		1	mA
			$V_O = -7 \text{ V}$		-0.8	
$I_{IH}$	High-level input current	$V_I = 2.4 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$			-100	$\mu\text{A}$
$I_{OS}$	Short-circuit output current	$V_O = -7 \text{ V}$			-250	mA
		$V_O = 0$			-150	
		$V_O = V_{CC}$			250	
		$V_O = 12 \text{ V}$				
$I_{CC}$	Supply current	$V_I = 0$ or $V_{CC}$ , No load	Receiver disabled and driver enabled	55LBC176	1.75	mA
				65LBC176	1.5	
				75LBC176		
			Receiver and driver disabled	55LBC176	0.25	
				65LBC176	0.2	
				75LBC176		

†  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$  respectively, that occur when the input is changed from a high level to a low level.

NOTES: 3. This device meets the RS485  $V_{OD}$  requirements above  $0^\circ\text{C}$  only.

4. This applies for both power on and off; refer to EIA standard RS-485 for exact conditions.

# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B – D3502, AUGUST 1990 – REVISED FEBRUARY 1993

**switching characteristics over recommended ranges of supply voltage, operating free-air temperature**

PARAMETER	TEST CONDITIONS	SN55LBC176		SN65LBC176 SN75LBC176			UNIT
		MIN	MAX	MIN	TYP†	MAX	
$t_{dD}$ Differential-output delay time	$R_L = 54 \Omega, C_L = 50 \text{ pF}$ , See Figure 3	8	31	8		25	ns
$t_{sk(p)}$ Pulse skew ( $ t_{dDH} - t_{dDL} $ )		6		0		6	ns
$t_{PLH}$ Propagation time, low-to-high-level single-ended output						26	ns
$t_{PHL}$ Propagation time, high-to-low-level single-ended output						26	ns
$t_{PZH}$ Output enable time to high level	$R_L = 110 \Omega$ , See Figure 4	65				60	ns
$t_{PZL}$ Output enable time to low level	$R_L = 110 \Omega$ , See Figure 5	65				60	ns
$t_{PHZ}$ Output disable time from high level	$R_L = 110 \Omega$ , See Figure 4	105				60	ns
$t_{PLZ}$ Output disable time from low level	$R_L = 110 \Omega$ , See Figure 5	105				60	ns

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

### SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-485
$V_O$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_O$
$ V_{OD2} $	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $	$V_t$ (Test Termination Measurement 2)
$\Delta  V_{OD} $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{Os} $
$\Delta  V_{OC} $	$ V_{Os} - \bar{V}_{Os} $
$I_{OS}$	
$I_O$	$I_{ia}, I_{ib}$



# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B – D3502, AUGUST 1990 – REVISED FEBRUARY 1993

## RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7\text{ V}$ ,	$I_O = -0.4\text{ mA}$			0.2	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5\text{ V}$ ,	$I_O = 8\text{ mA}$	-0.2‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ ) (see Figure 4)				50		mV
$V_{IK}$	Enable-input clamp voltage	$I_I = -18\text{ mA}$				-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 200\text{ mV}$ , See Figure 6	$I_{OH} = -400\text{ }\mu\text{A}$ ,		2.7		V
$V_{OL}$	Low-level output voltage	$V_{ID} = 200\text{ mV}$ , See Figure 6	$I_{OL} = 8\text{ mA}$ ,			0.45	V
$I_{OZ}$	High-impedance-state output current	$V_O = 0.4\text{ V to } 2.4\text{ V}$				$\pm 20$	$\mu\text{A}$
$I_I$	Line input current	Other input = 0 V, See Note 5	$V_I = 12\text{ V}$ $V_I = -7\text{ V}$			1 -0.8	mA
$I_{IH}$	High-level enable-input current	$V_{IH} = 2.7\text{ V}$				20	$\mu\text{A}$
$I_{IL}$	Low-level enable-input current	$V_{IL} = 0.4\text{ V}$				-100	$\mu\text{A}$
$r_i$	Input resistance				12		k $\Omega$
$I_{CC}$	Supply current	$V_I = 0$ or $V_{CC}$ , No load	Receiver enabled and driver disabled			3.9	mA
			Receiver and driver disabled	SN55LBC176		0.25	mA
				SN65LBC176 SN75LBC176		0.2	

switching characteristics over recommended ranges of supply voltage and operating free-air temperature,  $C_L = 15\text{ pF}$

PARAMETER	TEST CONDITIONS	SN55LBC176		SN65LBC176 SN75LBC176			UNIT
		MIN	MAX	MIN	TYP†	MAX	
$t_{PLH}$	Propagation time, low-to-high-level single-ended output	$V_{ID} = -1.5\text{ V to } 1.5\text{ V}$ , See Figure 7	20	37	20	30	ns
$t_{PHL}$	Propagation time, high-to-low-level single-ended output		20	55	28	45	ns
$t_{sk(p)}$	Pulse skew ( $ t_{dDH} - t_{dDL} $ )		22		10	18	ns
$t_{PZH}$	Output enable time to high level	See Figure 8	34		30		ns
$t_{PZL}$	Output enable time to low level		34		30		ns
$t_{PHZ}$	Output disable time from high level	See Figure 8	34		30		ns
$t_{PLZ}$	Output disable time from low level		34		30		ns

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 5: This applies for both power on and power off. Refer to EIA Standard RS-485 for exact conditions.



# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B – D3502, AUGUST 1990 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

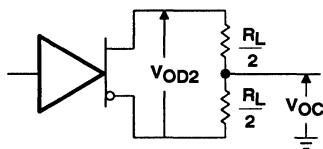


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

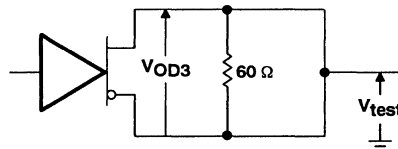
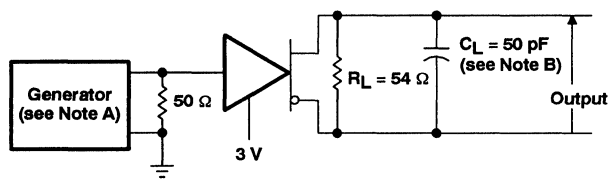
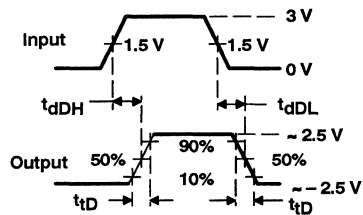


Figure 2. Driver  $V_{OD3}$

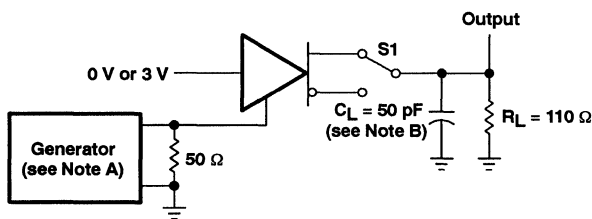


TEST CIRCUIT

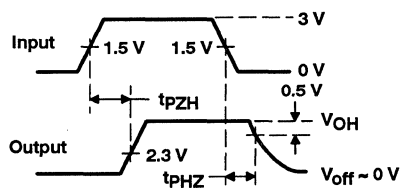


VOLTAGE WAVEFORMS

Figure 3. Driver Test Circuit and Voltage Waveforms

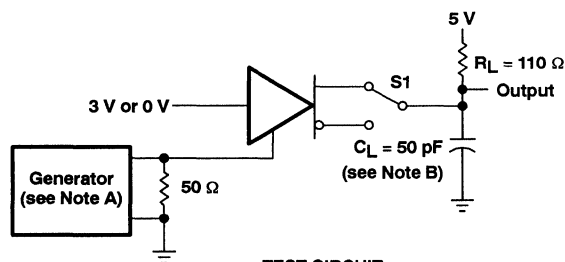


TEST CIRCUIT

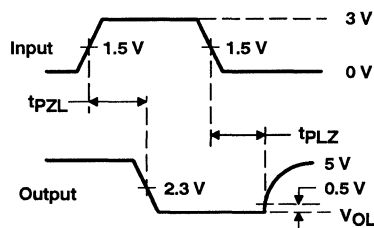


VOLTAGE WAVEFORMS

Figure 4. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT



VOLTAGE WAVEFORMS

Figure 5. Driver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B – D3502, AUGUST 1990 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

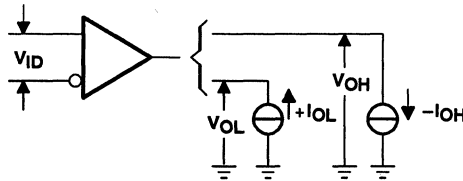


Figure 6. Receiver  $V_{OH}$  and  $V_{OL}$

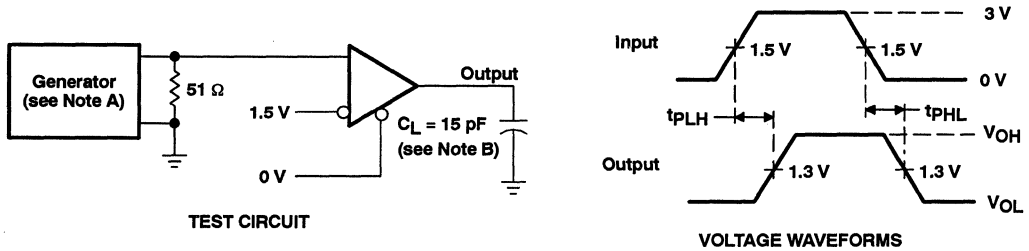


Figure 7. Receiver Test Circuit and Voltage Waveforms

# SN55LBC176, SN65LBC176, SN75LBC176 DIFFERENTIAL BUS TRANSCEIVERS

SLLS067B – D3502, AUGUST 1990 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

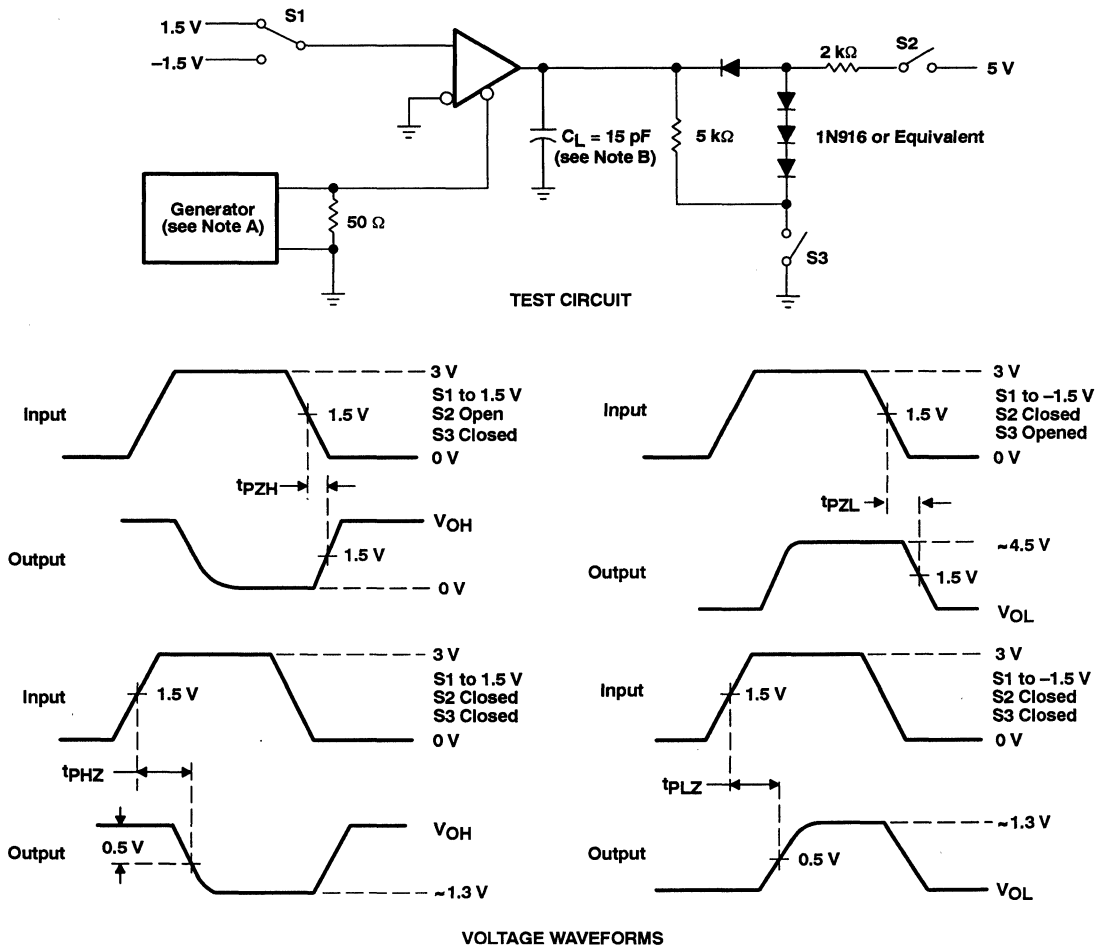


Figure 8. Receiver Test Circuit and Voltage Waveforms

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

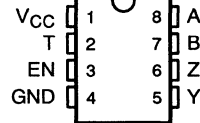


# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

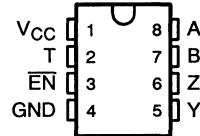
SLLS002C - D2606, JULY 1985 - REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- 3-State Outputs
- Bus Voltage Range . . . -7 V to 12 V
- Positive and Negative Current Limiting
- Driver Output Capability . . . 60 mA Max
- Driver Thermal Shutdown Protection
- Receiver Input Impedance . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From Single 5-V Supply
- Low Power Requirements

SN75177B . . . D OR P PACKAGE  
(TOP VIEW)



SN75178B . . . P PACKAGE  
(TOP VIEW)



THE SN75177B IS NOT  
RECOMMENDED FOR NEW DESIGN

## description

The SN75177B and SN75178B differential bus repeaters are monolithic integrated devices each designed for one-way data communication on multipoint bus transmission lines. These devices are designed for balanced transmission bus line applications and meet EIA Standard RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27. Each device is designed to improve the performance of the data communication over long bus lines. The SN75177B and SN75178B are identical except for the complementary enable inputs, which allow the devices to be used in pairs for bidirectional communication.

The SN75177B and SN75178B feature positive- and negative-current limiting 3-state outputs for the receiver and driver. The receiver features high input impedance, input hysteresis for increased noise immunity, and input sensitivity of  $\pm 200$  mV over a common-mode input voltage range of -7 V to 12 V. The driver features thermal shutdown for protection from line fault conditions. Thermal shutdown is designed to occur at a junction temperature of approximately 150°C. The driver is designed to drive current loads up to 60 mA maximum.

The SN75177B and SN75178B are designed for optimum performance when used on transmission buses employing the SN75172 and SN75174 differential line drivers, SN75173 and SN75175 differential line receivers, or SN75176B bus transceiver.

## Function Tables

SN75177B

DIFFERENTIAL INPUTS A - B	ENABLE EN	OUTPUTS		
		T	Y	Z
$V_{ID} \geq 0.2$ V	H	H	H	L
$-0.2$ V < $V_{ID} < 0.2$ V	H	?	?	?
$V_{ID} \leq 0.2$ V	H	L	L	H
X	L	Z	Z	Z

SN75178B

DIFFERENTIAL INPUTS A - B	ENABLE $\overline{EN}$	OUTPUTS		
		T	Y	Z
$V_{ID} \geq 0.2$ V	L	H	H	L
$-0.2$ V < $V_{ID} < 0.2$ V	L	?	?	?
$V_{ID} \leq 0.2$ V	L	L	L	H
X	H	Z	Z	Z

H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = impedance (off)

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

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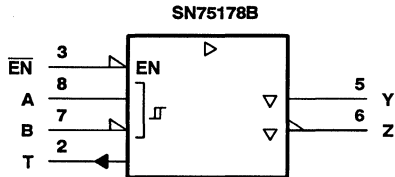
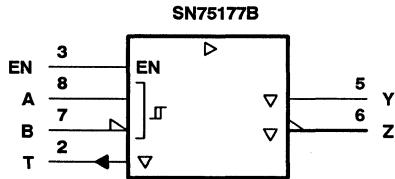
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# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

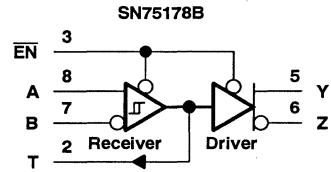
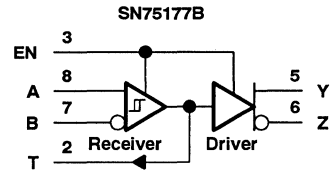
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## logic symbols†

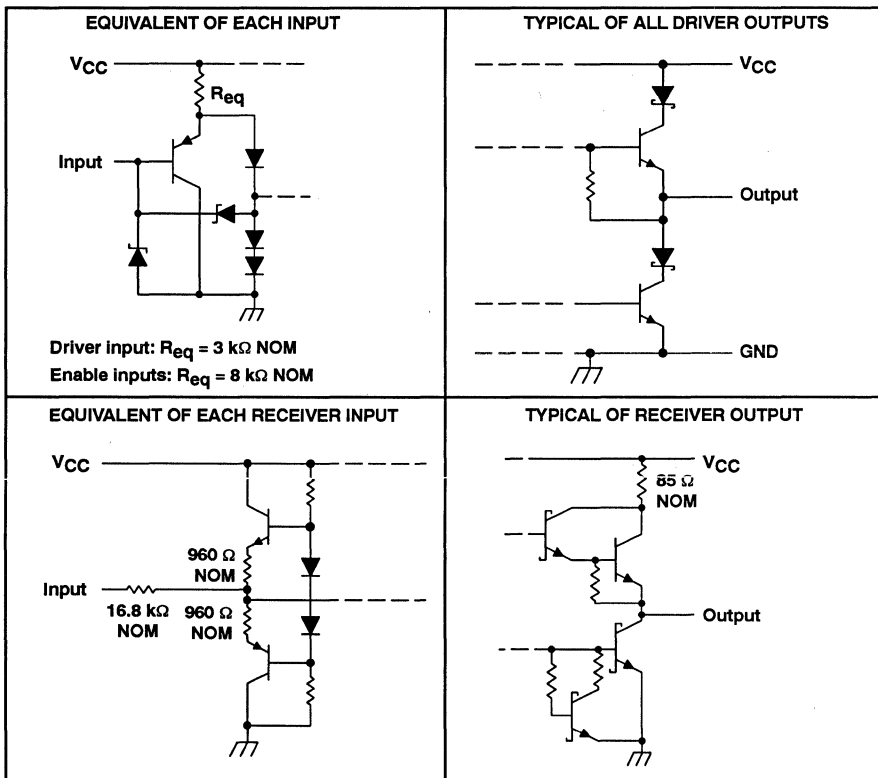


† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagrams (positive logic)



## schematics of inputs and outputs



# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Voltage range at any bus terminal .....	–10 V to 15 V
Differential input voltage (see Note 2) .....	±25 V
Enable input voltage .....	5.5 V
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential input voltage is measured at the noninverting input with respect to the corresponding inverting input.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	EN or $\overline{EN}$	2			V
low-level input voltage, $V_{IL}$	EN or $\overline{EN}$			0.8	V
Common-mode input voltage, $V_{IC}$		–7†		12	V
Differential input voltage, $V_{ID}$				±12	V
High-level output current, $I_{OH}$	Driver			–60	mA
	Receiver			–400	µA
Low-level output current, $I_{OL}$	Driver			60	mA
	Receiver			8	
Operating free-air temperature, $T_A$		0		70	°C

† The algebraic convention, where the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage.



# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

SLLS002C – D2606, JULY 1985 – REVISED FEBRUARY 1993

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$V_{IK}$ Input clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V	
$V_O$ Output voltage	$I_O = 0$	0		6	V	
$ V_{OD1} $ Differential output voltage	$I_O = 0$	1.5		6	V	
$ V_{OD2} $ Differential output voltage	$R_L = 100 \Omega$ , See Figure 1	$1/2 V_{OD1}$ or $2^{\S}$			V	
	$R_L = 54 \Omega$ , See Figure 1	1.5	2.5	5		
$ V_{OD3} $ Differential output voltage	See Note 3	1.5		5	V	
$\Delta V_{OD} $ Change in magnitude of differential output voltage‡	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1			$\pm 0.2$	V	
$V_{OC}$ Common-mode output voltage				3 -1	V	
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage‡				$\pm 0.2$	V	
$I_O$ Output current	$V_{CC} = 0$ , $V_O = -7 \text{ V to } 12 \text{ V}$			$\pm 100$	$\mu\text{A}$	
$I_{OZ}$ High-impedance-state output current	$V_O = -7 \text{ V to } 12 \text{ V}$			$\pm 100$	$\mu\text{A}$	
$I_{IH}$ High-level input current	$V_I = 2.4 \text{ V}$			20	$\mu\text{A}$	
$I_{IL}$ Low-level input current	$V_I = 0.4 \text{ V}$			-400	$\mu\text{A}$	
$I_{OS}$ Short-circuit output current	$V_O = -7 \text{ V}$			-250	mA	
	$V_O = V_{CC}$			250		
	$V_O = 12 \text{ V}$			250		
$I_{CC}$ Supply current (total package)	No load	Outputs enabled		57	70	mA
		Outputs disabled		26	35	

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

§ The minimum  $V_{OD2}$  with a  $100\text{-}\Omega$  load is either  $1/2 V_{OD1}$  or 2, whichever is greater.

NOTE 3: See Figure 3.5 of EIA Standard RS-485.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{dD}$ Differential-output delay time	$R_L = 54 \Omega$ , See Figure 3		15	20	ns
$t_{tD}$ Differential-output transition time				20	30
$t_{PZH}$ Output enable time to high level	$R_L = 110 \Omega$ , See Figure 4		85	120	ns
$t_{PZL}$ Output enable time to low level	$R_L = 110 \Omega$ , See Figure 5		40	60	ns
$t_{PHZ}$ Output disable time from high level	$R_L = 110 \Omega$ , See Figure 4		150	250	ns
$t_{PLZ}$ Output disable time from low level	$R_L = 110 \Omega$ , See Figure 5		20	30	ns

# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

SLLS002C - D2606, JULY 1985 - REVISED FEBRUARY 1993

## SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{Oa}, V_{Ob}$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_o$	$V_o$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$\Delta V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{OS} $	$ V_{OS} $
$\Delta V_{OC} $	$ V_{OS} - \bar{V}_{OS} $	$ V_{OS} - \bar{V}_{OS} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

## RECEIVER SECTION

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going input threshold voltage	$V_O = 2.7 V, I_O = -0.4 mA$			0.2	V
$V_{T-}$ Negative-going input threshold voltage	$V_O = 0.5 V, I_O = 8 mA$	-0.2‡			V
$V_{hys}$ Input hysteresis ( $V_{T+} - V_{T-}$ )			50		mV
$V_{IK}$ Input clamp voltage at EN	$I_I = -18 mA$			-1.5	V
$V_{OH}$ High-level output voltage	$V_{ID} = 200 mV, I_{OH} = -400 \mu A$ See Figure 2		2.7		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -200 mV, I_{OL} = 8 mA$ See Figure 2			0.45	V
$I_{OZ}$ High-impedance-state output current	$V_O = 0.4 V$ to $2.4 V$			20	$\mu A$
				-400	
$I_I$ Line input current	Other input at 0 V, See Note 4	$V_I = 12 V$		1	mA
		$V_I = -7 V$		-0.8	
$I_{IH}$ High-level enable-input current	$V_{IH} = 2.7 V$			20	$\mu A$
$I_{IL}$ Low-level enable-input current	$V_{IL} = 0.4 V$			-200	$\mu A$
$r_i$ Input resistance			12		k $\Omega$
$I_{OS}$ Short-circuit output current		-15		-85	mA
$I_{CC}$ Supply current (total package)	No load	Outputs enabled	57	70	mA
		Outputs disabled	26	35	

† All typical values are at  $V_{CC} = 5 V, T_A = 25^\circ C$ .

‡ The algebraic convention, where the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 4: Refer to EIA Standard RS-422 for exact conditions.

## switching characteristics, $V_{CC} = 5 V, T_A = 25^\circ C$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high level output	$V_{ID} = -1.5 V$ to $1.5 V,$ $C_L = 15 pF,$ See Figure 6		19	35	ns
$t_{PHL}$ Propagation delay time, high-to-low level output			30	40	
$t_{PZH}$ Output enable time to high level	$C_L = 15 pF,$ See Figure 7		10	20	ns
$t_{PZL}$ Output enable time to high level			12	20	
$t_{PHZ}$ Output disable time from high level	$C_L = 15 pF,$ See Figure 8		25	35	ns
$t_{PLZ}$ Output disable time from low level			17	25	



# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

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## PARAMETER MEASUREMENT INFORMATION

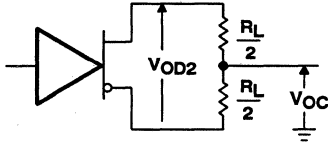


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

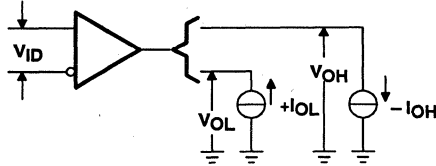
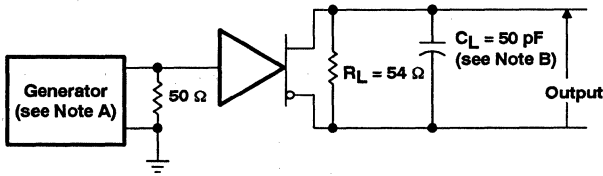
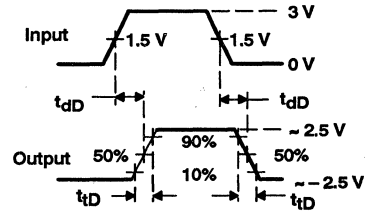


Figure 2. Receiver  $V_{OH}$  and  $V_{OL}$

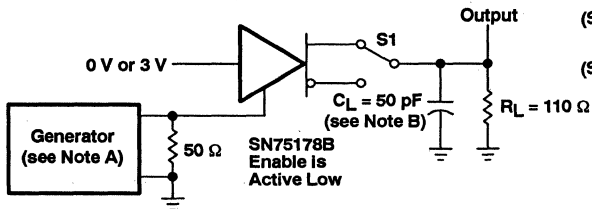


TEST CIRCUIT

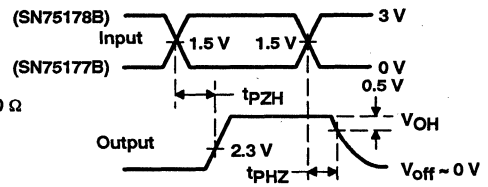


VOLTAGE WAVEFORMS

Figure 3. Driver Differential-Output Test Circuit and Voltage Waveforms

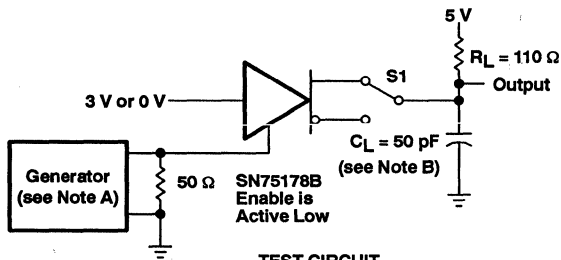


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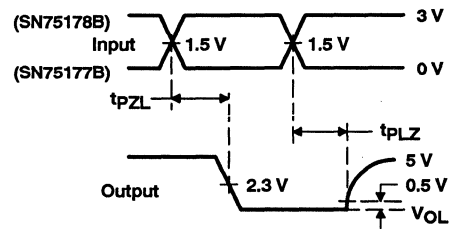


VOLTAGE WAVEFORMS

Figure 4. Driver Enable and Disable Times



TEST CIRCUIT



VOLTAGE WAVEFORMS

Figure 5. Driver Enable and Disable Times

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

SLLS002C - D2606, JULY 1985 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

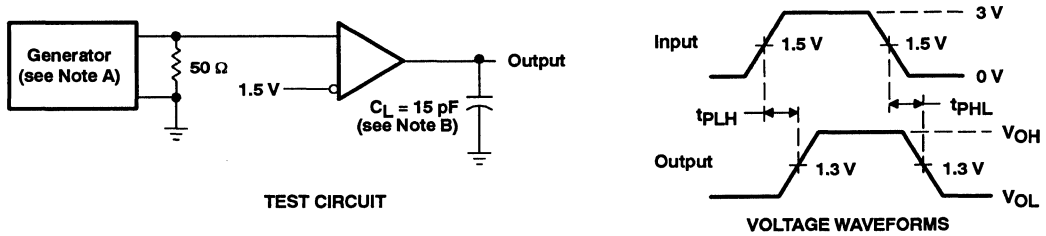


Figure 6. Receiver Propagation Delay Times

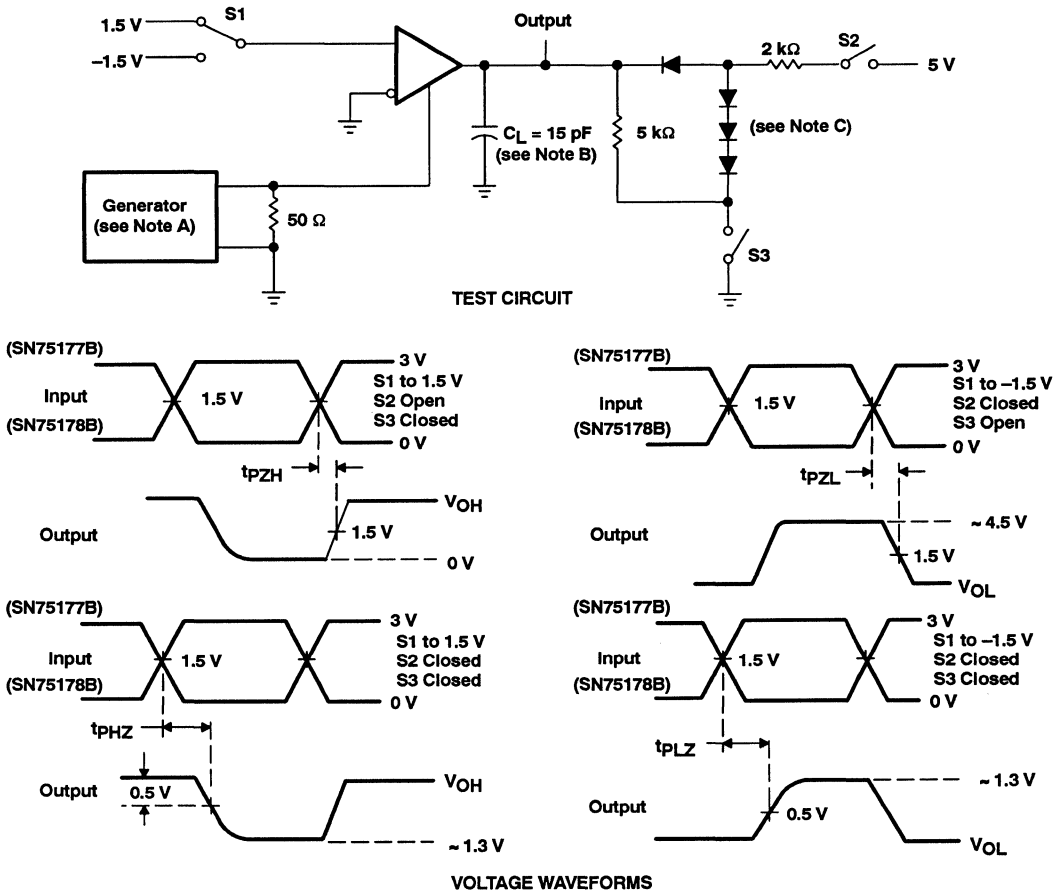


Figure 7. Receiver Output Enable and Disable Times

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1 \text{ MHz}$ , 50% duty cycle,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ ,  $Z_0 = 50 \Omega$ .
- B.  $C_L$  includes probe and jig capacitance.
- C. All diodes are 1N916 or equivalent.

# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

SLLS002C – D2606, JULY 1985 – REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

DRIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

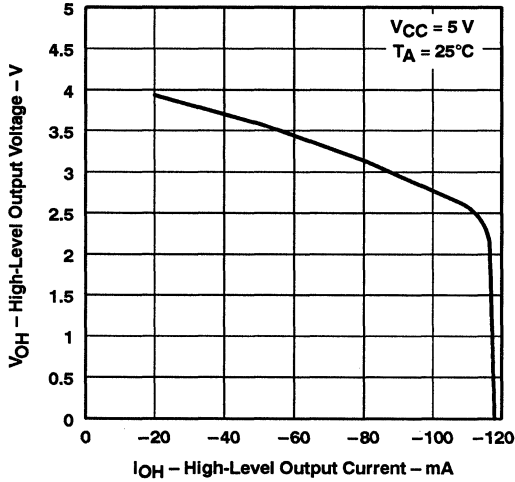


Figure 8

DRIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

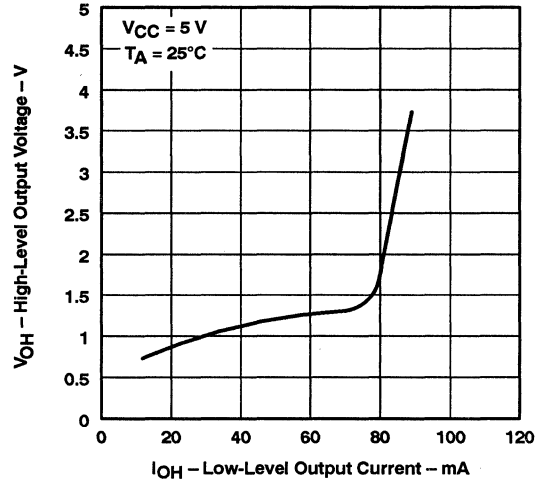


Figure 9

DRIVER DIFFERENTIAL OUTPUT VOLTAGE  
vs  
DRIVER OUTPUT CURRENT

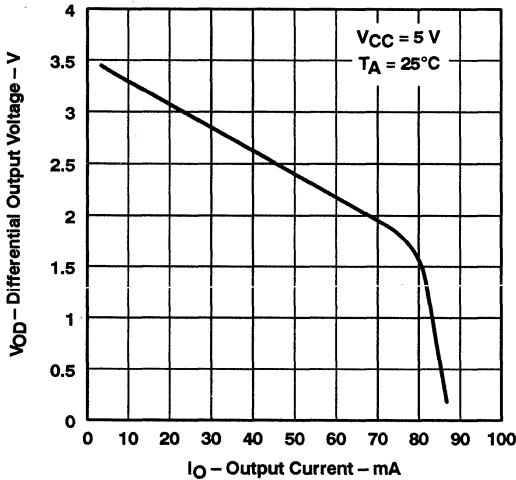


Figure 10

RECEIVER OUTPUT VOLTAGE  
vs  
DIFFERENTIAL INPUT VOLTAGE

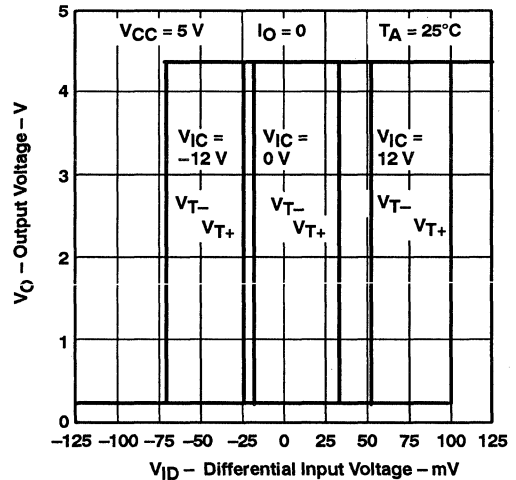


Figure 11

TYPICAL CHARACTERISTICS

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

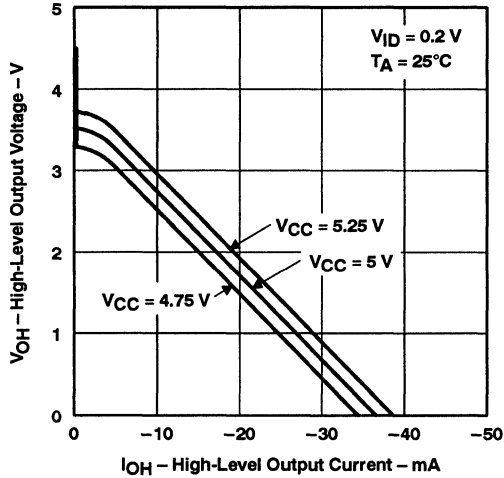


Figure 12

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

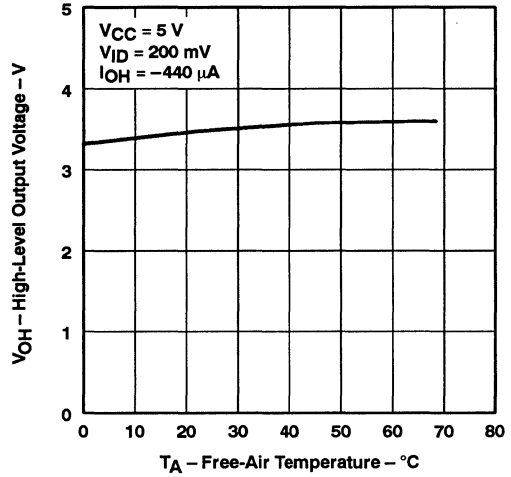


Figure 13

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

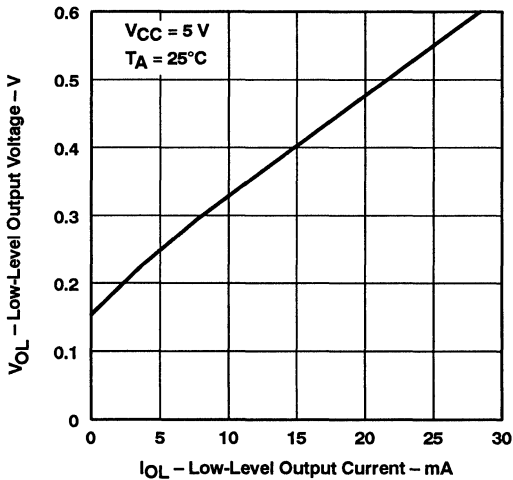


Figure 14

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

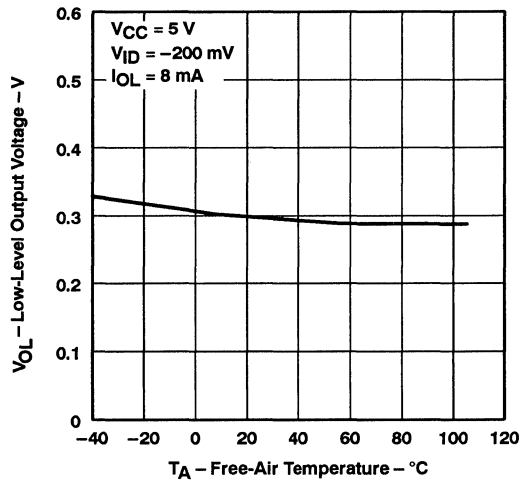
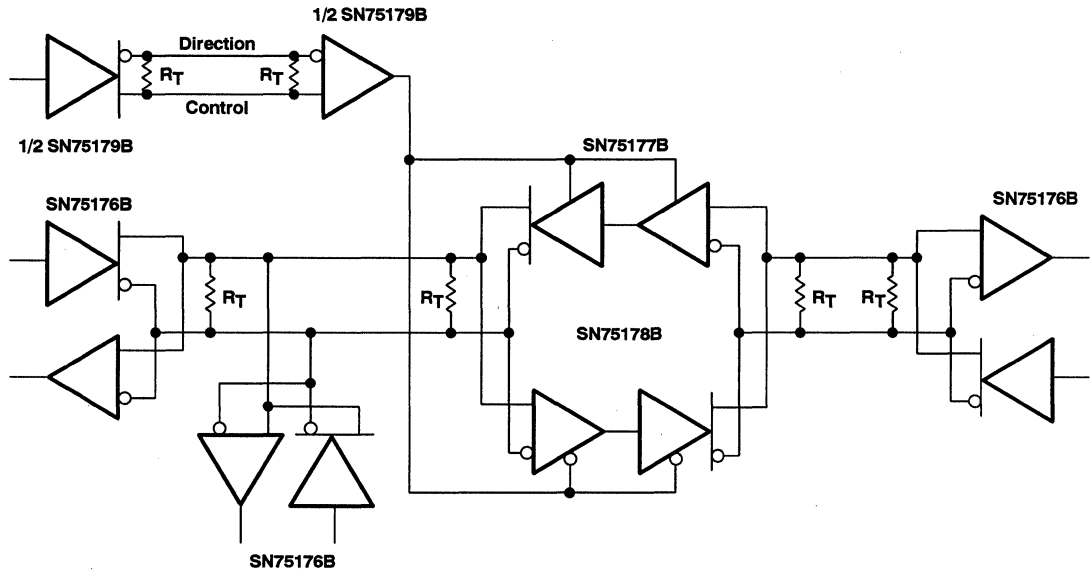


Figure 15

# SN75177B, SN75178B DIFFERENTIAL BUS REPEATERS

SLLS002C - D2606, JULY 1985 - REVISED FEBRUARY 1993

## APPLICATION INFORMATION



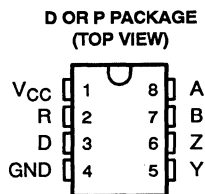
NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

Figure 16. Typical Application Circuit

# SN75179A DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B – D2845, JUNE 1984 – REVISED FEBRUARY 1993

- Meets EIA Standards RS-422A, RS423A, and CCITT Recommendations V.11 and X.27
- Bus Voltage Range . . . -7 V to 12 V
- Positive and Negative Current Limiting
- Driver Output Capability . . . 60 mA Max
- Driver Thermal Shutdown Protection
- Receiver Input Impedance . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From Single 5-V Supply
- Low Power Requirements



NOT RECOMMENDED FOR NEW DESIGN

## description

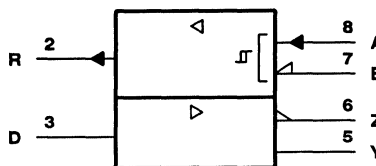
The SN75179A driver and bus receiver circuit is a monolithic integrated device designed for balanced transmission line applications, and meets EIA Standards RS-422A, RS-423A, and CCITT Recommendations V.11 and X.27. It is designed to improve the performance of data communications over long bus lines.

The SN75179A features positive- and negative-current limiting for the driver and receiver. The receiver features high input impedance, input hysteresis for increased noise immunity, and input sensitivity of  $\pm 200$  mV over a common-mode input voltage range of -12 V to 12 V.

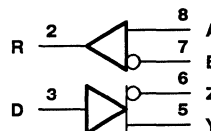
The driver provides thermal shutdown for protection from line fault conditions. Thermal shutdown is designed to occur at a junction temperature of approximately 150°C. The device is designed to drive current loads of up to 60 mA maximum.

The SN75179A is characterized for operation from 0°C to 70°C.

## logic symbol



## logic diagram



## Function Tables

DRIVER		
INPUT D	OUTPUTS Y Z	
H	H	L
L	L	H

RECEIVER	
DIFFERENTIAL INPUTS A - B	OUTPUT R
$V_{ID} \geq 0.2$ V	H
$-0.2$ V $< V_{ID} < 0.2$ V	?
$V_{ID} \leq -0.2$ V	L

H = high level, L = low level, ? = indeterminate

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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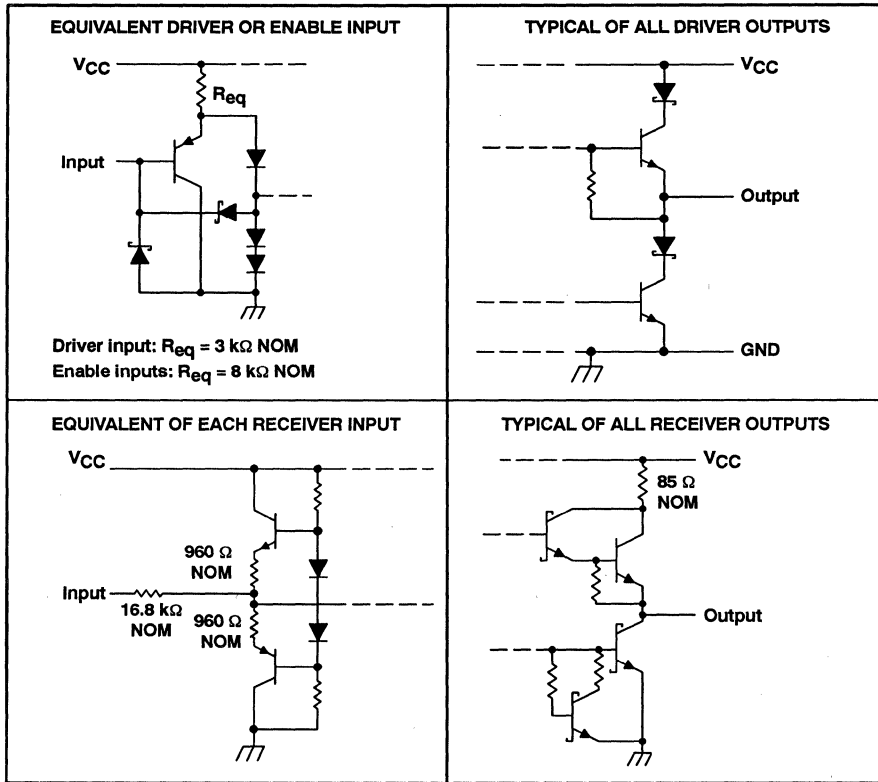
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# SN75179A DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B – D2845, JUNE 1984 – REVISED FEBRUARY 1993

## schematics of inputs and outputs



### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	-10 V to 15 V
Differential input voltage (see Note 2)	$\pm 25$ V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

# SN75179A DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B – D2845, JUNE 1984 – REVISED FEBRUARY 1993

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.5	5	5.25	V
High-level input voltage, $V_{IH}$	Driver	2			V
Low-level input voltage, $V_{IL}$	Driver	0.8			V
Common-mode input voltage, $V_{IC}$		-7 †			V
Differential input voltage, $V_{ID}$		±12			V
High-level output current, $I_{OH}$	Driver	-60			mA
	Receiver	-400			μA
Low-level output current, $I_{OL}$	Driver	60			mA
	Receiver	8			mA
Operating free-air temperature, $T_A$		0	70		°C

† The algebraic convention, where the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage.

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP ‡	MAX	UNIT	
$V_{IK}$ Input clamp voltage	$I_I = -18$ mA	-1.5			V	
$V_{OH}$ High-level output voltage	$V_{IH} = 2$ V, $I_{OH} = -33$ mA	$V_{IL} = 0.8$ V,		3.7	V	
$V_{OL}$ Low-level output voltage	$V_{IH} = 2$ V, $I_{OH} = 33$ mA	$V_{IL} = 0.8$ V,		1.1	V	
$ V_{OD1} $ Differential output voltage	$I_O = 0$			$2 V_{OD2}$	V	
$ V_{OD2} $ Differential output voltage	$R_L = 100$ Ω, See Figure 13	2	2.7	V		
	$R_L = 54$ Ω, See Figure 13	1.5	2.4			
$\Delta V_{OD} $ Change in magnitude of differential output voltage §	$R_L = 54$ Ω or $100$ Ω, See Figure 13				±0.2	V
$V_{OC}$ Common-mode output voltage ¶					3	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage §					±0.2	V
$I_O$ Output current with power off	$V_{CC} = 0$ , $V_O = -7$ V to $12$ V				±100	μA
$I_{IH}$ High-level input current	$V_I = 2.4$ V				20	μA
$I_{IL}$ Low-level input current	$V_I = 0.4$ V				-400	μA
$I_{OS}$ Short-circuit output current	$V_O = -7$ V				-250	mA
	$V_O = V_{CC}$				250	
	$V_O = 12$ V				500	
$I_{CC}$ Supply current (total package)	No load				50	mA

‡ All typical values are at  $V_{CC} = 5$  V and  $T_A = 25$ °C.

§  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

¶ In EIA Standard RS-422A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

## switching characteristics, $V_{CC} = 5$ V, $T_A = 25$ °C

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{dD}$ Differential-output delay time	$R_L = 60$ Ω, See Figure 3	40		60	ns
$t_{TD}$ Differential-output transition time		65		95	ns



# SN75179A

## DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B – D2845, JUNE 1984 – REVISED FEBRUARY 1993

### RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7\text{ V}$ , $I_O = -0.4\text{ mA}$			0.2	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5\text{ V}$ , $I_O = 8\text{ mA}$	-0.2‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	See Figure 9		50		mV
$V_{OH}$	High-level output voltage	$V_{ID} = 200\text{ mV}$ , See Figure 2		2.7		V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200\text{ mV}$ , $I_{OL} = 8\text{ mA}$ , See Figure 2			0.45	V
$I_I$	Line input current	Other input at 0 V, See Note 3		$V_I = 12\text{ V}$ $V_I = -7\text{ V}$	1 -0.8	mA
$r_i$	Input resistance			12		k $\Omega$
$I_{OS}$	Short-circuit output current		-15		-85	mA
$I_{CC}$	Supply current (total package)	No load			50	mA

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, where the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 3: Refer to EIA Standard RS-422A for exact conditions.

### switching characteristics, $V_{CC} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -1.5\text{ V}$ to $1.5\text{ V}$ , $C_L = 15\text{ pF}$ ,		26	35	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output	See Figure 5		27	35	ns

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PARAMETER MEASUREMENT INFORMATION

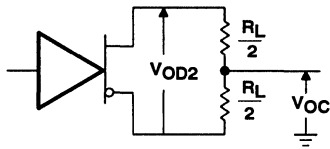


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

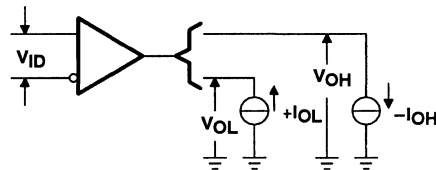


Figure 2. Receiver  $V_{OH}$  and  $V_{OL}$

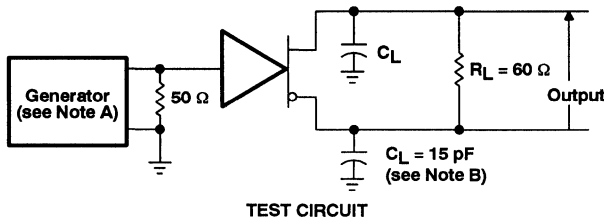


Figure 3. Driver Differential-Output Delay and Transition Times

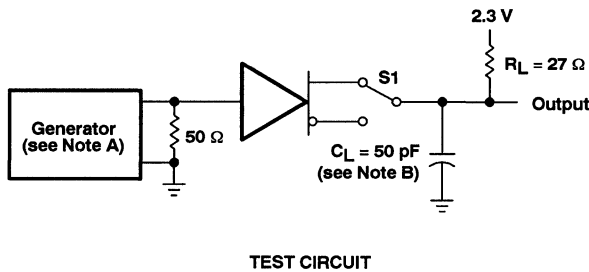
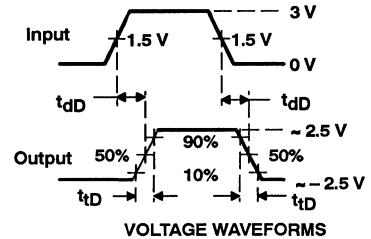


Figure 4. Driver Test Circuit and Voltage Waveforms

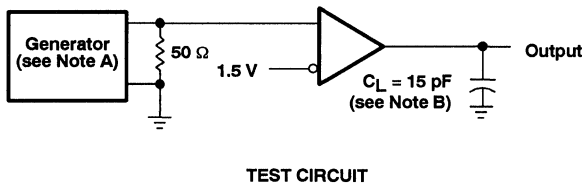
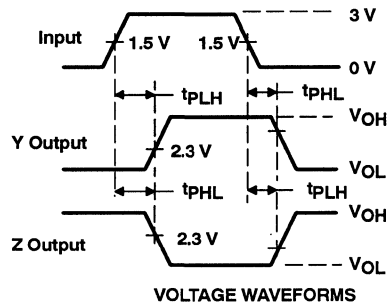
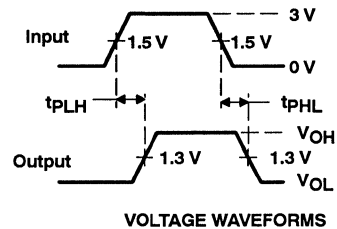


Figure 5. Receiver Test Circuit and Voltage Waveforms



NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, 50% duty cycle,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

# SN75179A DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B - D2845, JUNE 1984 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

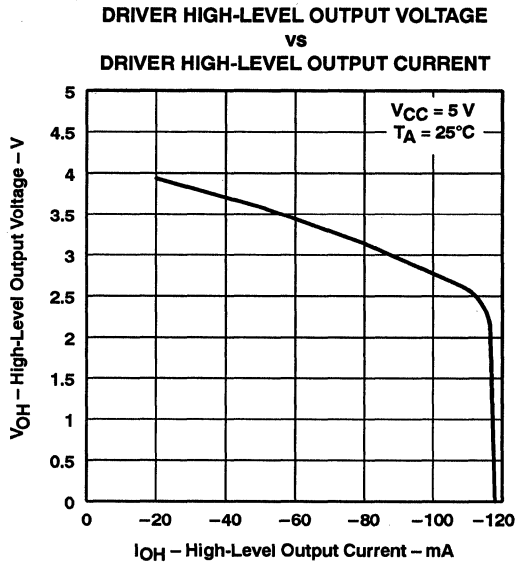


Figure 6

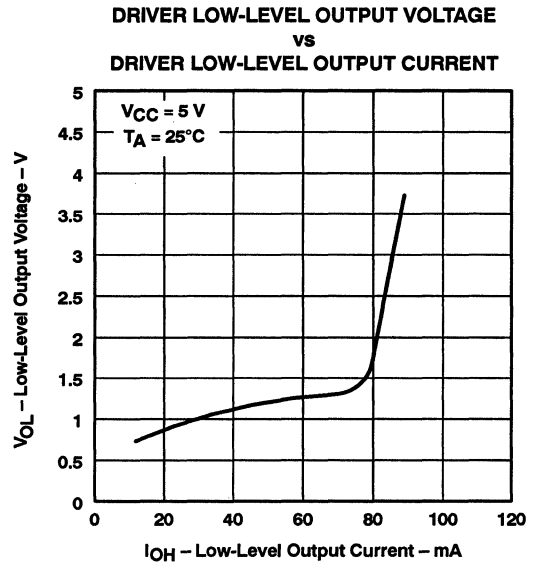


Figure 7

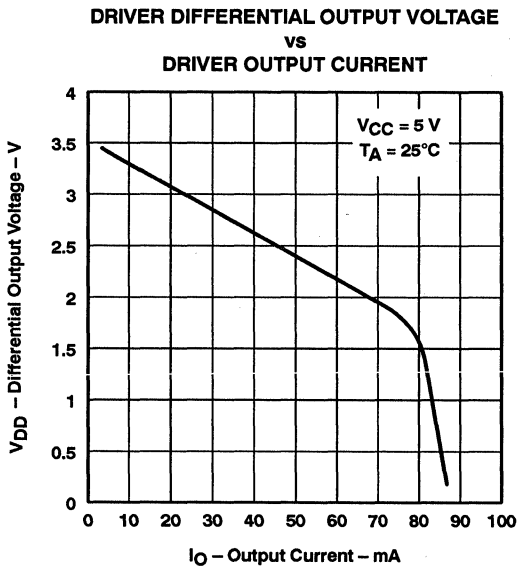


Figure 8

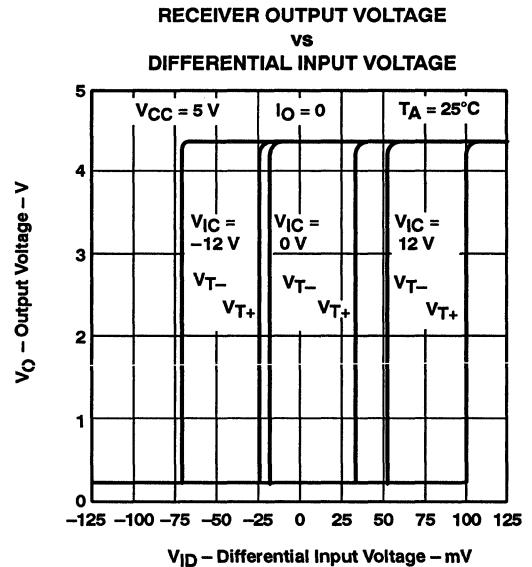


Figure 9

TYPICAL CHARACTERISTICS

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

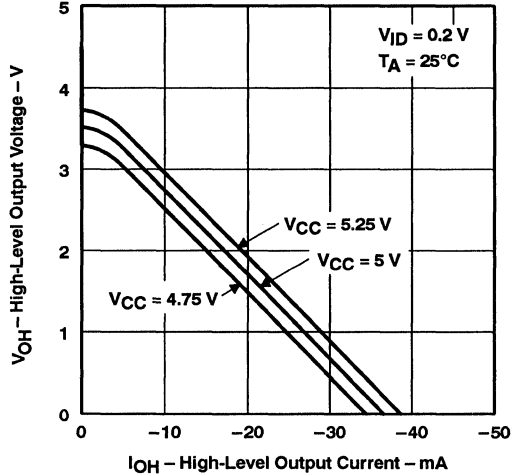


Figure 10

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

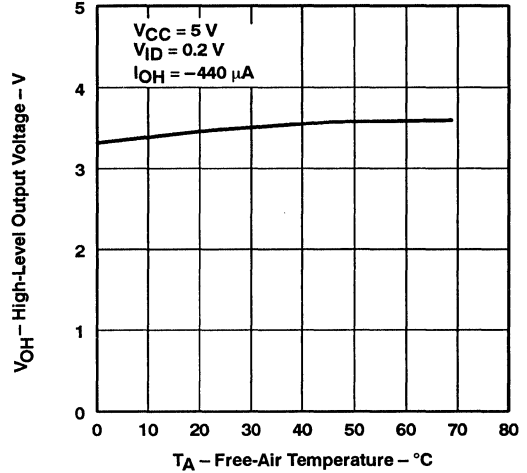


Figure 11

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
RECEIVER LOW-LEVEL OUTPUT CURRENT

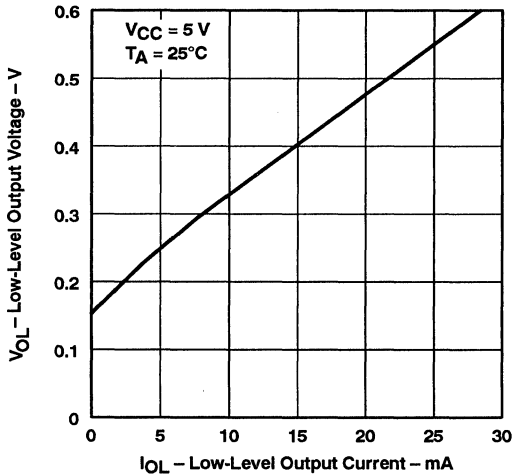


Figure 12

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

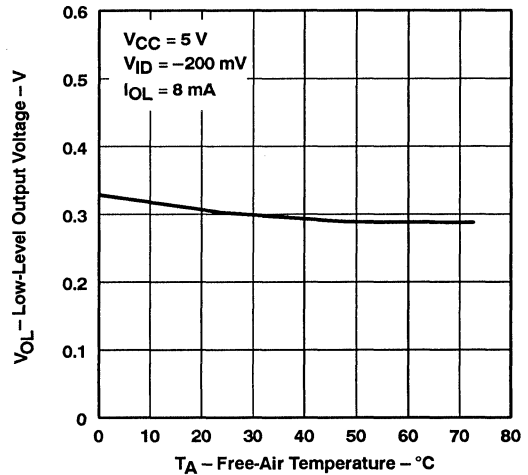


Figure 13



# SN65179B, SN75179B DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B – D2845, OCTOBER 1985 – REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A, RS-485 and CCITT Recommendations V.11 and X.27
- Bus Voltage Range . . . -7 V to 12 V
- Positive and Negative Current Limiting
- Driver Output Capability . . . 60 mA Max
- Driver Thermal Shutdown Protection
- Receiver Input Impedance . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From Single 5-V Supply
- Low Power Requirements

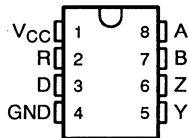
## description

The SN65179B and SN75179B differential driver and receiver pair are monolithic integrated devices designed for balanced transmission line applications and meet EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27. They are designed to improve the performance of full-duplex data communications over long bus lines.

The SN65179B and SN75179B driver outputs provide limiting for both positive and negative currents. The receiver features high input impedance, input hysteresis for increased noise immunity, and input sensitivity of  $\pm 200$  mV over a common-mode input voltage range of -7 V to 12 V. The driver provides thermal shutdown for protection from line fault conditions. Thermal shutdown is designed to occur at a junction temperature of approximately 150°C. The SN65179B and SN75179B are designed to drive current loads of up to 60 mA maximum.

The SN65179B is characterized for operation from -40°C to 85°C. The SN75179B is characterized for operation from 0°C to 70°C.

D OR P PACKAGE  
(TOP VIEW)



Function Tables

DRIVER

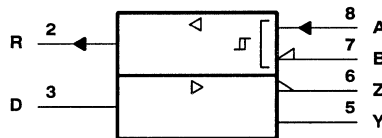
INPUT D	OUTPUTS Y Z	
H	H	L
L	L	H

RECEIVER

DIFFERENTIAL INPUTS A - B	OUTPUT R
$V_{ID} \geq 0.2$ V	H
$-0.2$ V < $V_{ID}$ < 0.2 V	?
$V_{ID} \leq -0.2$ V	L

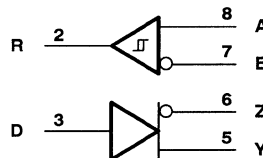
H = high level, L = low level,  
? = indeterminate

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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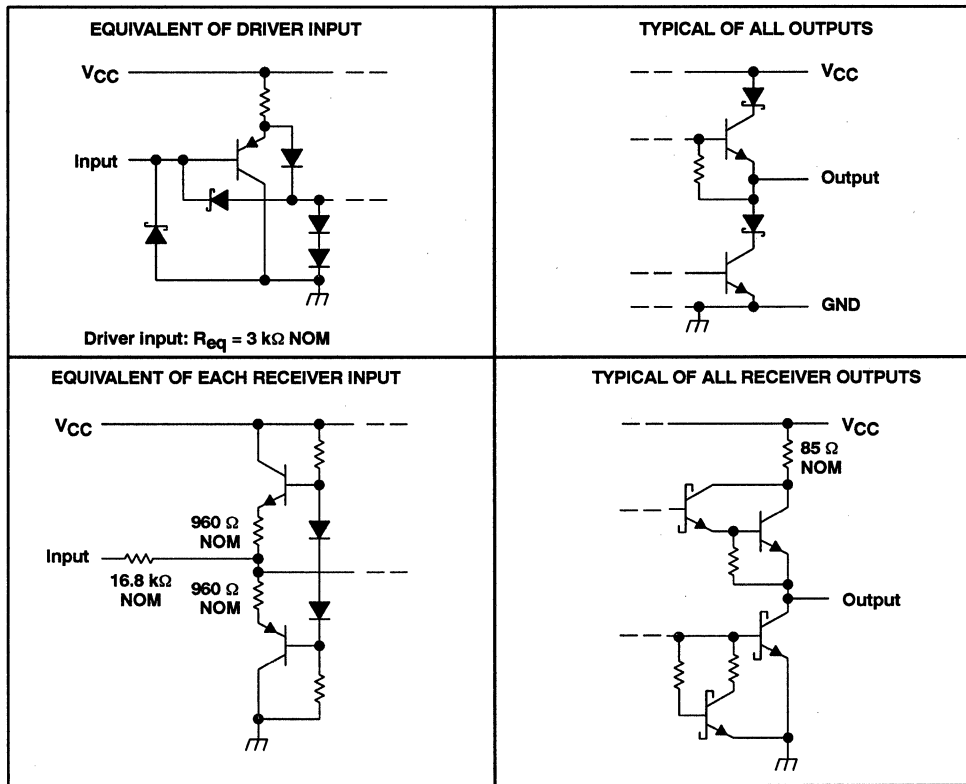
2-667



# SN65179B, SN75179B DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B – D2845, OCTOBER 1985 – REVISED FEBRUARY 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	-10 V to 15 V
Differential input voltage (see Note 2)	$\pm 25$ V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65179B	-40°C to 85°C
SN75179B	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW

# SN65179B, SN75179B DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B – D2845, OCTOBER 1985 – REVISED FEBRUARY 1993

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	Driver	2			V
Low-level input voltage, $V_{IL}$	Driver	0.8			V
Common-mode input voltage, $V_{IC}$		-7 <sup>†</sup>			V
Differential input voltage, $V_{ID}$		±12			V
High-level output current, $I_{OH}$	Driver	-60			mA
	Receiver	-400			μA
Low-level output current, $I_{OL}$	Driver	60			mA
	Receiver	8			
Operating free-air temperature, $T_A$	SN65179B	-40			°C
	SN75179B	0			

<sup>†</sup> The algebraic convention, where the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage.

## DRIVER SECTION

### electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP <sup>†</sup>	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA		-1.5			V
$V_O$	Output voltage	$I_O = 0$		0	6		V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5	6		V
$ V_{OD2} $	Low-level output voltage	$R_L = 100 \Omega$ ,	See Figure 1	$1/2 V_{OD1}$ or 2 <sup>††</sup>			V
		$R_L = 54 \Omega$ ,	See Figure 1	1.5	2.5	5	V
$ V_{OD3} $	Differential output voltage	See Note 3		1.5	5		V
$\Delta V_{OD} $	Change in magnitude of common-mode output voltage <sup>§</sup>					±0.2	V
$V_{OC}$	Common-mode output voltage	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1				<sup>3</sup> -1	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage <sup>§</sup>					±0.2	V
$I_O$	Output current	$V_{CC} = 0$ , $V_O = -7$ V to 12 V		±100			μA
$I_{IH}$	High-level input current	$V_I = 2.4$ V		20			μA
$I_{IL}$	Low-level input current	$V_I = 0.4$ V		-200			μA
$I_{OS}$	Short-circuit output current	$V_O = -7$ V		-250			mA
		$V_O = V_{CC}$ or 12 V		250			
$I_{CC}$	Supply current (total package)	No load		57	70		mA

<sup>†</sup> All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ\text{C}$ .

<sup>§</sup>  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

<sup>††</sup> The minimum  $V_{OD2}$  with 100-Ω load is either  $1/2 V_{OD2}$  or 2 V, whichever is greater.

NOTE 3: See EIA Standard RS-485, Figure 3.5, Test Termination Measurement 2.

### switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{dD}$	Differential-output delay time	$R_L = 54 \Omega$ ,	See Figure 3	15	22		ns
$t_{tD}$	Differential-output transition time			20	30		ns



# SN65179B, SN75179B DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B - D2845, OCTOBER 1985 - REVISED FEBRUARY 1993

## Symbol Equivalents

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{Oa}, V_{Ob}$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_o$	$V_o$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$\Delta V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

## RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going input threshold voltage	$V_O = 2.7 \text{ V}$ , $I_O = -0.4 \text{ mA}$			0.2	V
$V_{T-}$ Negative-going input threshold voltage	$V_O = 0.5 \text{ V}$ , $I_O = 8 \text{ mA}$	-0.2‡			V
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )			50		mV
$V_{OH}$ High-level output voltage	$V_{ID} = 200 \text{ mV}$ , $I_{OH} = -400 \mu\text{A}$ , See Figure 2		2.7		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -200 \text{ mV}$ , $I_{OL} = 8 \text{ mA}$ , See Figure 2			0.45	V
$I_I$ Line input current	Other input at 0 V, See Note 4			1 -0.8	mA
$r_i$ Input resistance			12		k $\Omega$
$I_{OS}$ Short-circuit output current		-15		-85	mA
$I_{CC}$ Supply current (total package)	No load		57	70	mA

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, where the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 4: Refer to EIA Standard RS-422A for exact conditions.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{pLH}$ Propagation delay time, low-to-high-level output	$V_{ID} = -1.5 \text{ V to } 1.5 \text{ V}$ , $C_L = 15 \text{ pF}$ , See Figure 4		19	35	ns
$t_{pHL}$ Propagation delay time, high-to-low-level output			30	40	ns



PARAMETER MEASUREMENT INFORMATION

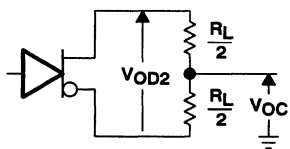


Figure 1. Driver  $V_{DD}$  and  $V_{OC}$

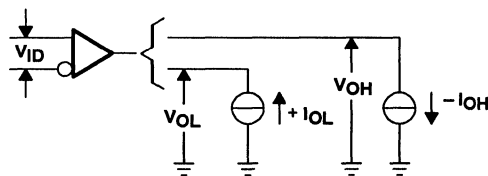
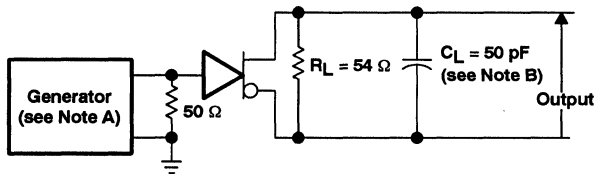


Figure 2. Receiver  $V_{OH}$  and  $V_{OL}$



TEST CIRCUIT

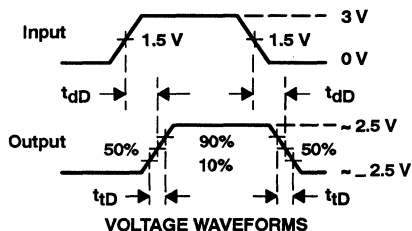
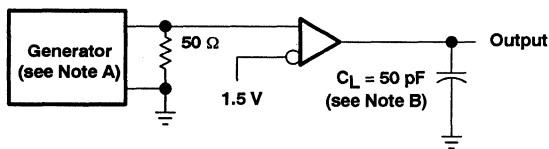


Figure 3. Driver Test Circuit and Voltage Waveforms



TEST CIRCUIT

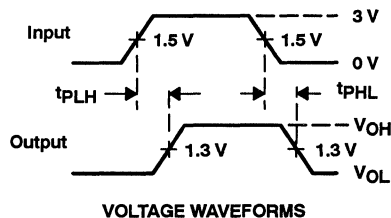


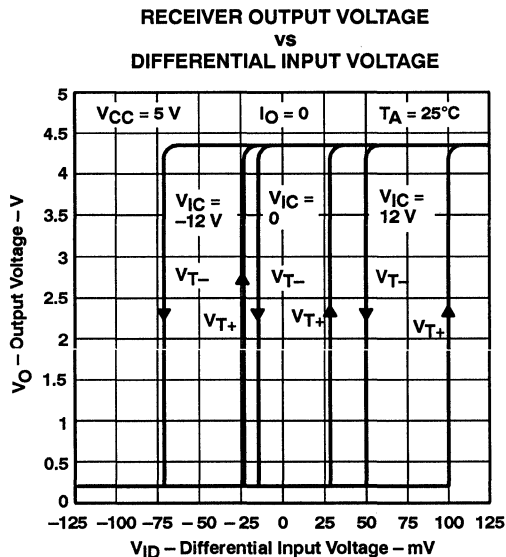
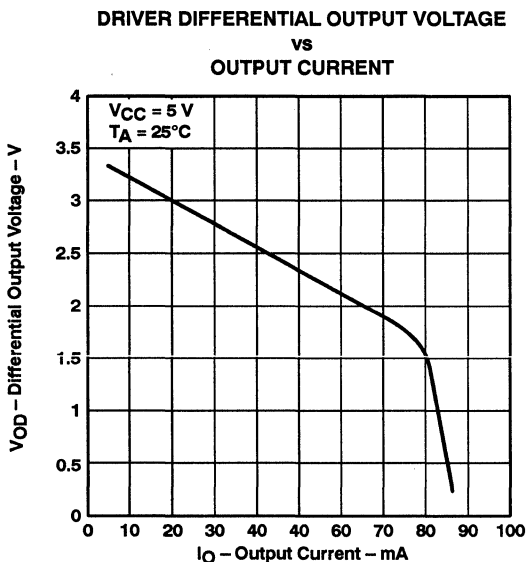
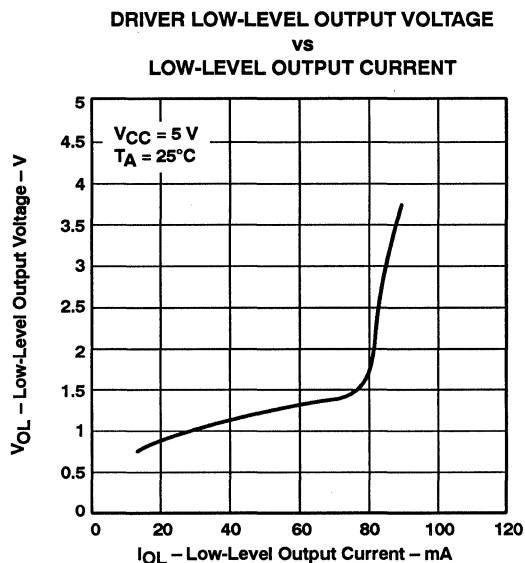
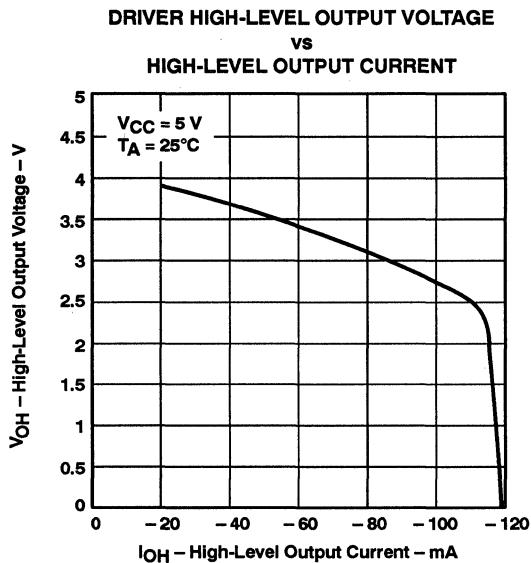
Figure 4. Receiver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

# SN65179B, SN75179B DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS123B - D2845, OCTOBER 1985 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS



TYPICAL CHARACTERISTICS

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

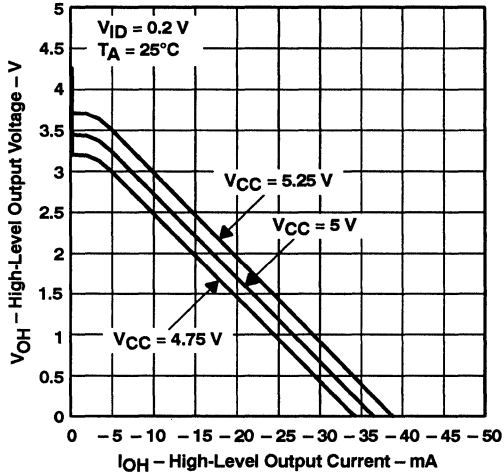


Figure 9

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

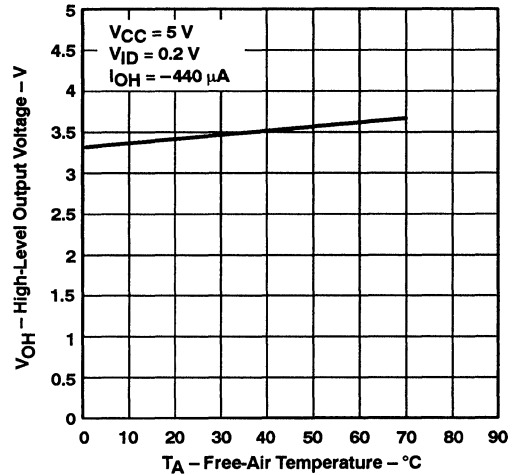


Figure 10

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

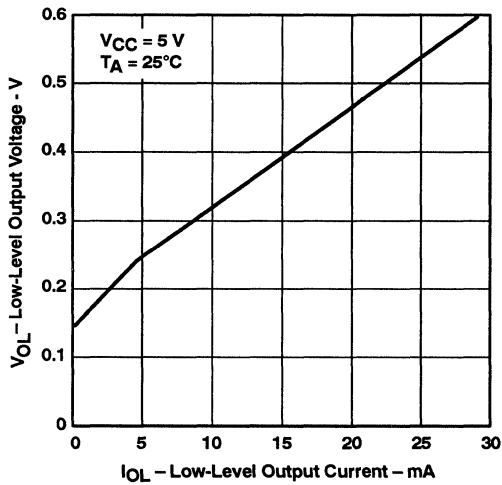


Figure 11

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

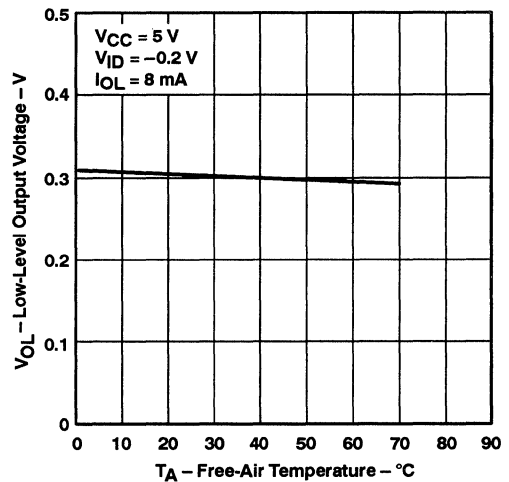


Figure 12



# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C – D3043, AUGUST 1987 – REVISED FEBRUARY 1992

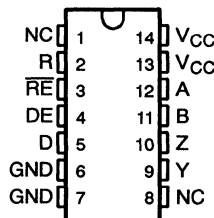
- Meets EIA Standards RS-422A and RS-485† and CCITT Recommendations V.11 and X.27
- High-Speed Advanced Low-Power Schottky Circuitry
- Designed for 25-MBaud Operation in Both Serial and Parallel Applications
- Low Skew Between Devices . . . 6 ns Max
- Low Supply Current Requirements  
30 mA Max
- Individual Driver and Receiver I/O Pins With Dual V<sub>CC</sub> and Dual GND
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capacity . . . ±60 mA
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Impedances . . . 12 kΩ Min
- Receiver Input Sensitivity . . . ±200 mV Max
- Receiver Input Hysteresis . . . 60 mV Typ
- Operates From a Single 5-V Supply
- Glitch-Free Power-Up and Power-Down Protection

## description

The SN65ALS180 and SN75ALS180 differential driver and receiver pairs are monolithic integrated circuits designed for bidirectional data communication on multipoint bus transmission lines. They are designed for balanced transmission lines and meet EIA Standards RS-422-A and RS-485 and CCITT recommendations V.11 and X.27.

The SN65ALS180 and SN75ALS180 combine a 3-state differential line driver and a differential input line receiver, both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, which can be externally connected together to function as a direction control. The driver differential outputs and the receiver

D OR N PACKAGE  
(TOP VIEW)



NC—No internal connection

## Function Tables

### DRIVER

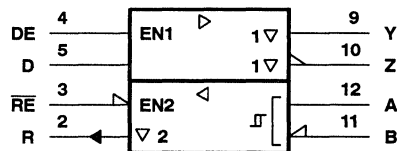
INPUT D	ENABLE DE	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	Z	Z

### RECEIVER

DIFFERENTIAL INPUTS A–B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2\text{ V}$	L	H
$-0.2\text{ V} < V_{ID} < 0.2\text{ V}$	L	?
$V_{ID} \leq -0.2\text{ V}$	L	L
X	H	Z
Open	L	H

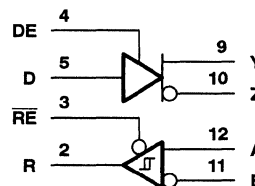
H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = high impedance (off)

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



† These devices meet or exceed the requirements of EIA RS485 except for the Generator Contention Test (para. 3.4.2) and the Generator Current Limit (para. 3.4.3). The applied test voltage ranges are –6 V to 8 V for the SN75ALS180 and –4.5 V to 8 V for the SN65ALS180.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

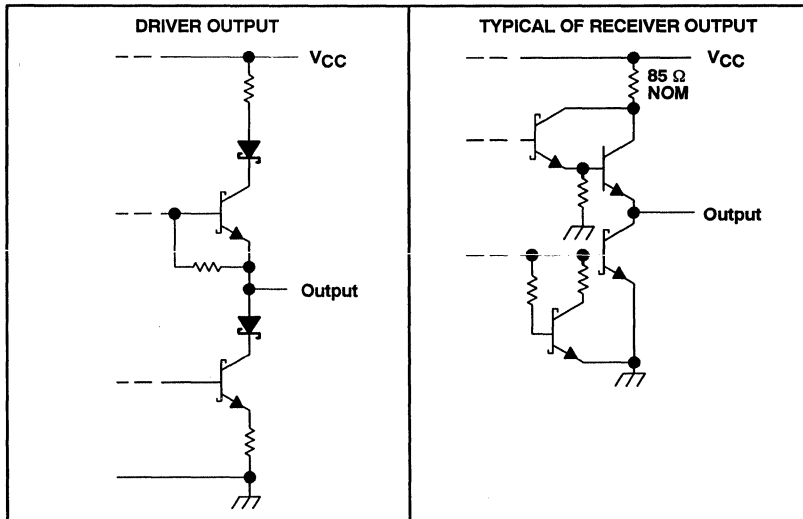
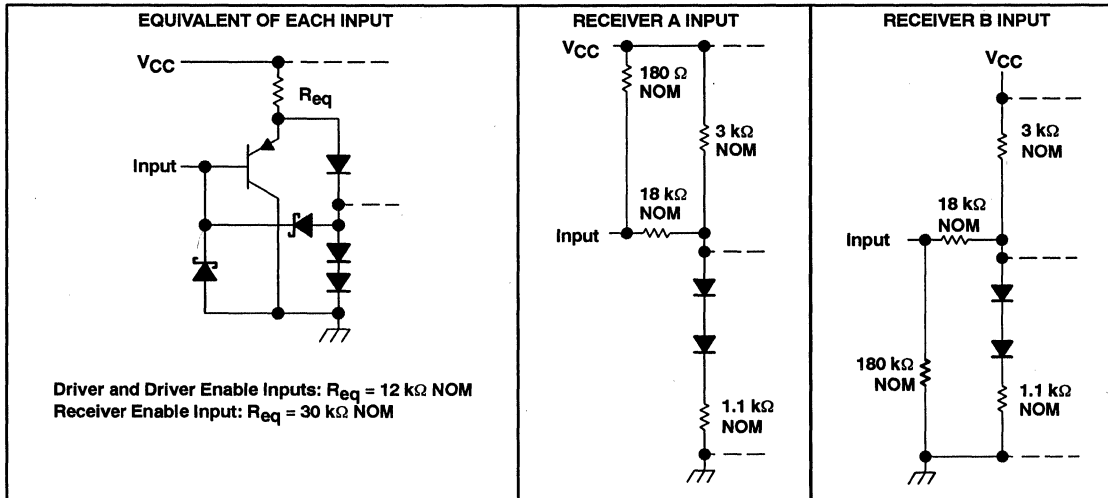
SLLS052C - D3043, AUGUST 1987 - REVISED FEBRUARY 1992

## description (continued)

differential inputs are connected to separate pins for greater flexibility and are designed to offer minimum loading to the bus when the driver is disabled or  $V_{CC} = 0$ . These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

The SN65ALS180 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75ALS180 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## schematics of inputs and outputs



# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C – D3043, AUGUST 1987 – REVISED FEBRUARY 1992

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	–10 V to 15 V
Enable input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : SN65ALS180	–40°C to 85°C
SN75ALS180	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values, except differential I/O bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_O$			12 –7	V
High-level input voltage, $V_{IH}$	D, DE, and RE		2	V
Low-level input voltage, $V_{IL}$	D, DE, and RE		0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)			±12	V
High-level output current, $I_{OH}$	Driver		–60	mA
	Receiver		–400	µA
Low-level output current, $I_{OL}$	Driver		60	mA
	Receiver		8	mA
Operating free-air temperature, $T_A$	SN65ALS180		–40	85
	SN75ALS180		0	70

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A/Y with respect to the inverting terminal B/Z.



# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C - D3043, AUGUST 1987 - REVISED FEBRUARY 1992

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT	
V <sub>IK</sub>	Input clamp voltage	I <sub>I</sub> = -18 mA				-1.5	V	
V <sub>O</sub>	Output voltage	I <sub>O</sub> = 0		0		6	V	
V <sub>OD1</sub>	Differential output voltage	I <sub>O</sub> = 0		1.5		6	V	
V <sub>OD2</sub>	Differential output voltage	R <sub>L</sub> = 100 Ω, See Figure 1	1/2 V <sub>OD1</sub> or 2 <sup>S</sup>				V	
		R <sub>L</sub> = 54 Ω, See Figure 1	1.5	2.5	5			
V <sub>OD3</sub>	Differential output voltage	V <sub>test</sub> = -7 V to 12 V, See Figure 2		1.5		5	V	
Δ V <sub>OD</sub>	Change in magnitude of differential output voltage¶	R <sub>L</sub> = 54 Ω or 100 Ω, See Figure 1				±0.2	V	
V <sub>OC</sub>	Common-mode output voltage						3 -1	V
Δ V <sub>OC</sub>	Change in magnitude of common-mode output voltage¶						±0.2	V
I <sub>O</sub>	Output current	Output disabled, See Note 3	V <sub>O</sub> = 12 V			1	mA	
			V <sub>O</sub> = -7 V			-0.8		
I <sub>IH</sub>	High-level input current	V <sub>I</sub> = 2.4 V				20	μA	
I <sub>IL</sub>	Low-level input current	V <sub>I</sub> = 0.4 V				-400	μA	
I <sub>OS</sub>	Short-circuit output current#	V <sub>O</sub> = -6 V	SN75ALS180			-250	mA	
		V <sub>O</sub> = -4 V	SN65ALS180					
		V <sub>O</sub> = 0	All			-150		
		V <sub>O</sub> = V <sub>CC</sub>	All					
		V <sub>O</sub> = 8 V	All					
I <sub>CC</sub>	Supply current	No load	Outputs enabled		23	30	mA	
			Outputs disabled		19	26		

† The power-off measurement in EIA Standard RS-422-A applies to disabled outputs only and is not applied to combined inputs and outputs.

‡ All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

§ The minimum V<sub>OD2</sub> with 100-Ω load is either 1/2 V<sub>OD2</sub> or 2 V, whichever is greater.

¶ Δ|V<sub>OD</sub>| and Δ|V<sub>OC</sub>| are the changes in magnitude of V<sub>OD</sub> and V<sub>OC</sub> respectively, that occur when the input is changed from a high level to a low level.

# Duration of the short circuit does not exceed one second for this test.

NOTE 3: This applies for both power on and off; refer to EIA standard RS-485 for exact conditions. The RS-422-A limit does not apply for a combined driver and receiver terminal.

switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		TEST CONDITIONS		MIN	TYP‡	MAX	UNIT	
t <sub>dD</sub>	Differential output delay time	R <sub>L</sub> = 54 Ω, C <sub>L</sub> = 50 pF, See Figure 3		3	8	13	ns	
	Pulse skew ( t <sub>dDH</sub> - t <sub>dDL</sub>  )					1	6	ns
t <sub>D</sub>	Differential output transition time				3	8	13	ns
t <sub>PZH</sub>	Output enable time to high level	R <sub>L</sub> = 110 Ω, See Figure 4			23	50	ns	
t <sub>PZL</sub>	Output enable time to low level	R <sub>L</sub> = 110 Ω, See Figure 5			19	24	ns	
t <sub>PHZ</sub>	Output disable time from high level	R <sub>L</sub> = 110 Ω, See Figure 4			8	13	ns	
t <sub>PLZ</sub>	Output disable time from low level	R <sub>L</sub> = 110 Ω, See Figure 5			8	13	ns	

‡ All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.



# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C – D3043, AUGUST 1987 – REVISED FEBRUARY 1992

## SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{Oa}, V_{Ob}$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_O$	$V_O$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$V_{test}$		$V_{tst}$
$\Delta  V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta  V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

## RECEIVER SECTION

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7 \text{ V}, I_O = -0.4 \text{ mA}$			0.2	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5 \text{ V}, I_O = 8 \text{ mA}$	-0.2‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )			60		mV
$V_{IK}$	Enable-input clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = -200 \text{ mV}, I_{OH} = -400 \mu\text{A}$ , See Figure 6	2.7			V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200 \text{ mV}, I_{OL} = 8 \text{ mA}$ , See Figure 6			0.45	V
$I_{OZ}$	High-impedance-state output current	$V_O = 0.4 \text{ V to } 2.4 \text{ V}$			±20	μA
$I_I$	Line input current	Other input = 0 V, See Note 4	$V_I = 12 \text{ V}$		1	mA
			$V_I = -7 \text{ V}$		-0.8	
$I_{IH}$	High-level enable-input current	$V_{IH} = 2.7 \text{ V}$			20	μA
$I_{IL}$	Low-level enable-input current	$V_{IL} = 0.4 \text{ V}$			-100	μA
$r_i$	Input resistance		12			kΩ
$I_{OS}$	Short-circuit output current	$V_{ID} = 200 \text{ mV}, V_O = 0$	-15		-85	mA
$I_{CC}$	Supply current	No load			23	mA
					30	
					19	26

† All typical values are at  $V_{CC} = 5 \text{ V}, T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 4: This applies for both power on and power off. Refer to EIA Standard RS-485 for exact conditions.



# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C – D3043, AUGUST 1987 – REVISED FEBRUARY 1992

switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$V_{ID} = -1.5 \text{ V to } 1.5 \text{ V}$ , See Figure 7	9	14	19	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output					
Skew ( $ t_{PHL} - t_{PLH} $ )					
$t_{PZH}$ Output enable time to high level	$C_L = 15 \text{ pF}$ , See Figure 8	7	14	ns	
$t_{PZL}$ Output enable time to low level					
$t_{PHZ}$ Output disable time from high level					
$t_{PLZ}$ Output disable time from low level					

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

## PARAMETER MEASUREMENT INFORMATION

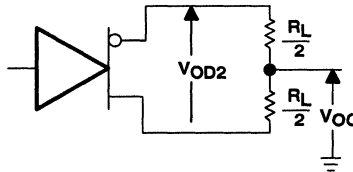


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

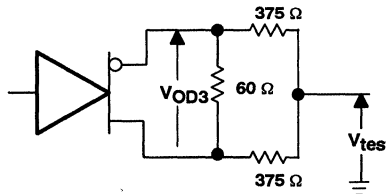
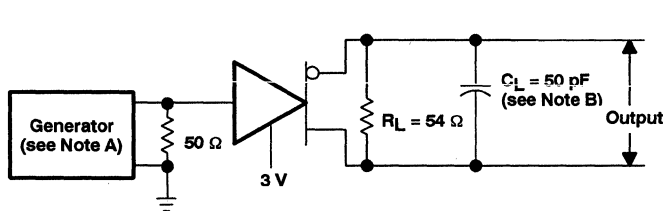
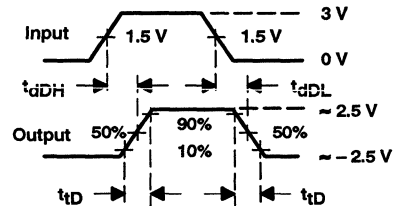


Figure 2. Driver  $V_{OD3}$



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1 \text{ MHz}$ , 50% duty cycle,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ ,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 3. Driver Test Circuit and Voltage Waveforms

# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C – D3043, AUGUST 1987 – REVISED FEBRUARY 1992

## PARAMETER MEASUREMENT INFORMATION

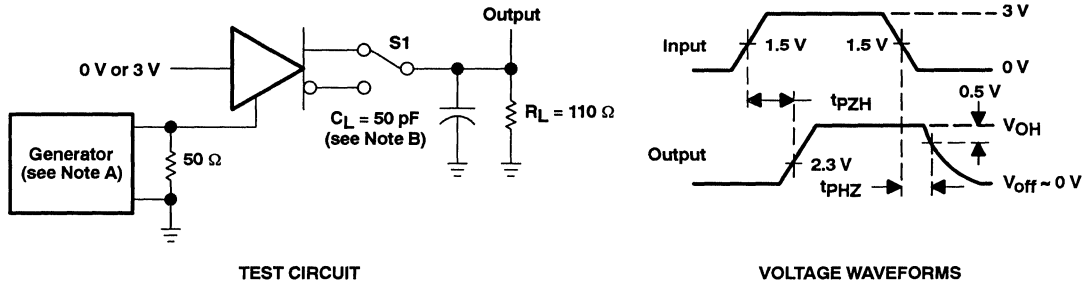


Figure 4. Driver Test Circuit and Voltage Waveforms

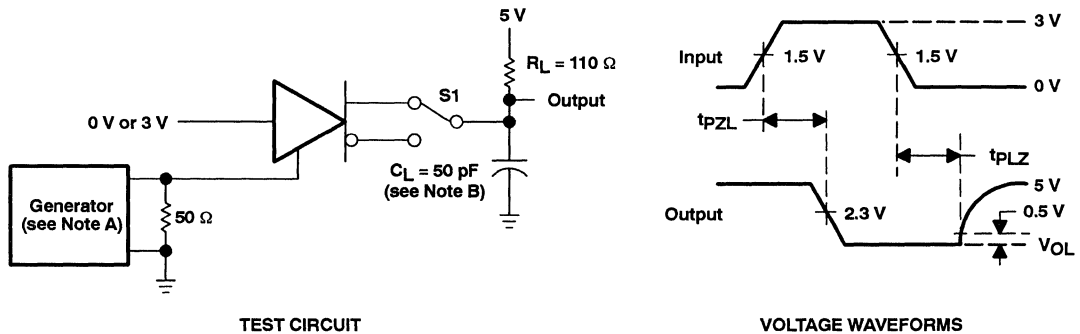


Figure 5. Driver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1\ \text{MHz}$ , 50% duty cycle,  $t_r \leq 6\ \text{ns}$ ,  $t_f \leq 6\ \text{ns}$ ,  $Z_0 = 50\ \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

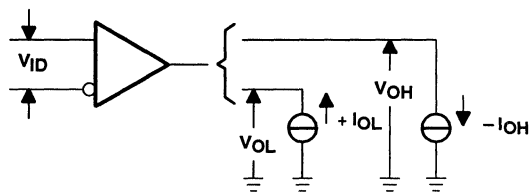


Figure 6. Receiver  $V_{OH}$  and  $V_{OL}$

# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C - D3043, AUGUST 1987 - REVISED FEBRUARY 1992

## PARAMETER MEASUREMENT INFORMATION

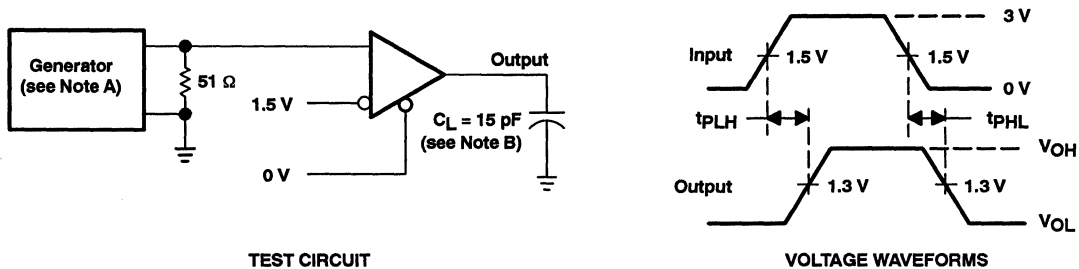


Figure 7. Receiver Test Circuit and Voltage Waveforms

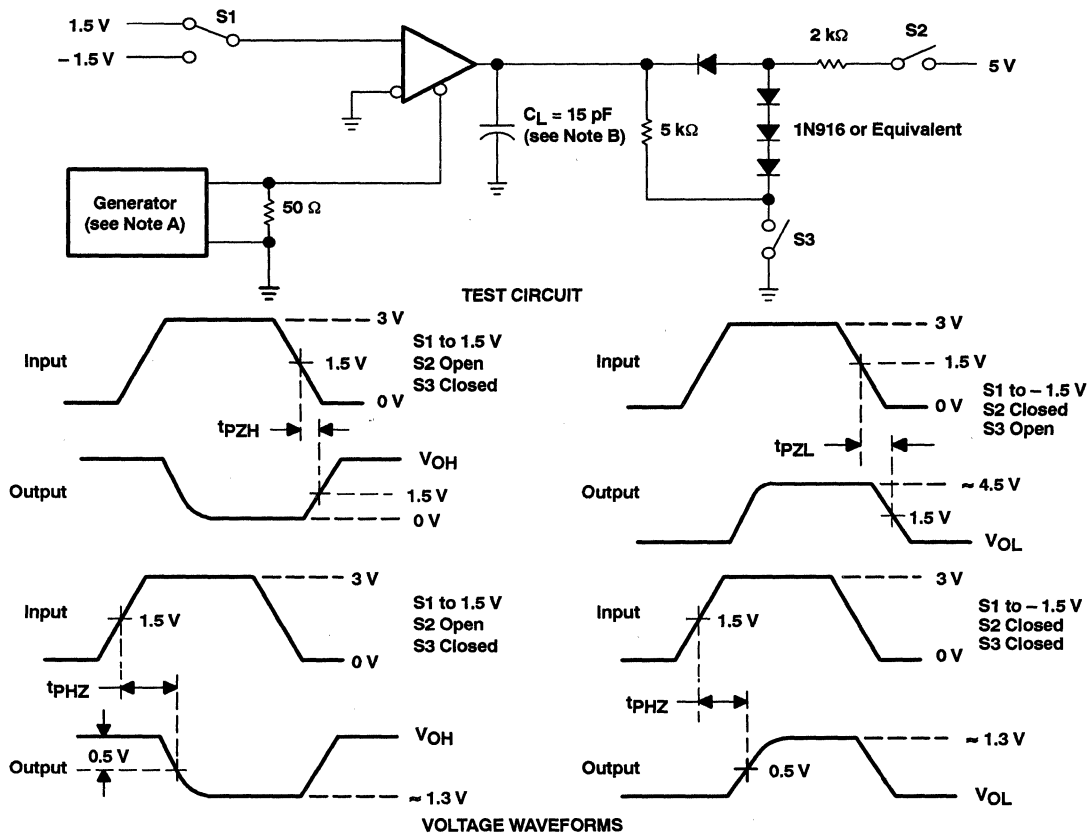


Figure 8. Receiver Test Circuit and Voltage Waveforms

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

TYPICAL CHARACTERISTICS

DRIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
DRIVER HIGH-LEVEL OUTPUT CURRENT

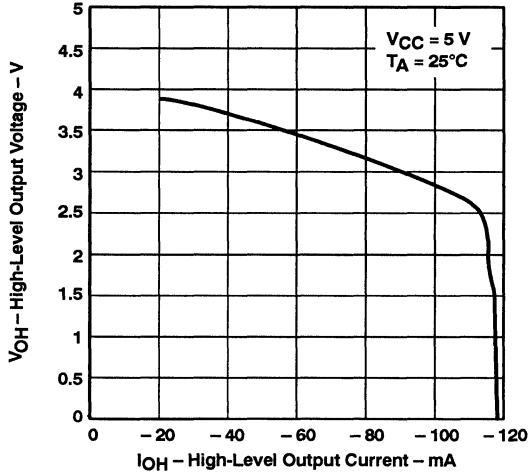


Figure 9

DRIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
DRIVER LOW-LEVEL OUTPUT CURRENT

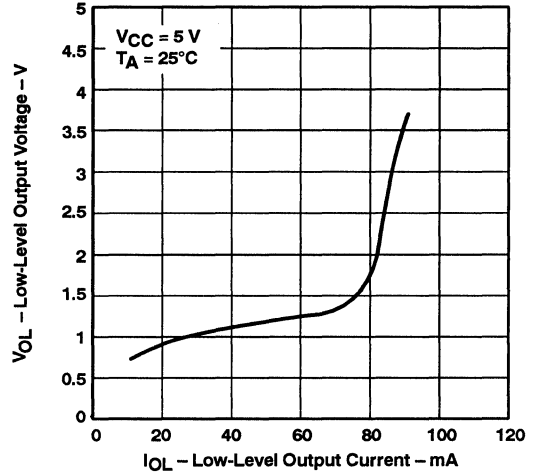


Figure 10

DRIVER DIFFERENTIAL OUTPUT VOLTAGE  
vs  
DRIVER OUTPUT CURRENT

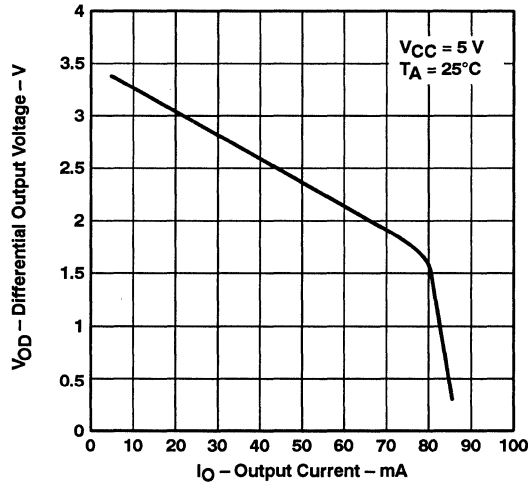


Figure 11



# SN65ALS180, SN75ALS180 DIFFERENTIAL DRIVER AND RECEIVER PAIRS

SLLS052C – D3043, AUGUST 1987 – REVISED FEBRUARY 1992

## TYPICAL CHARACTERISTICS

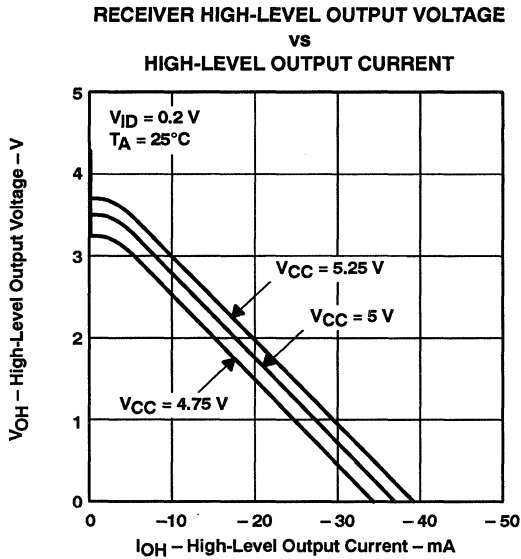


Figure 12

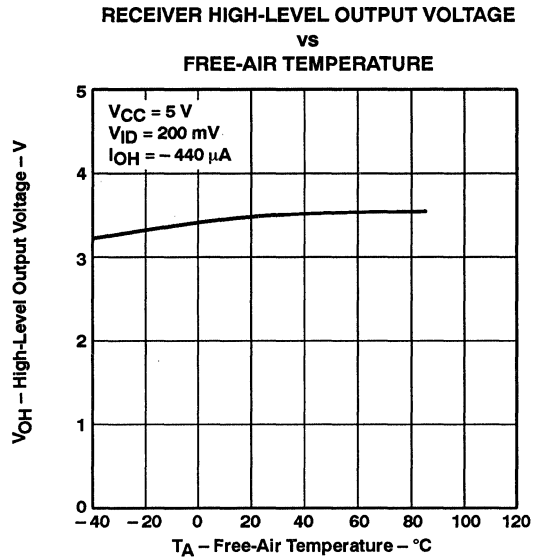


Figure 13

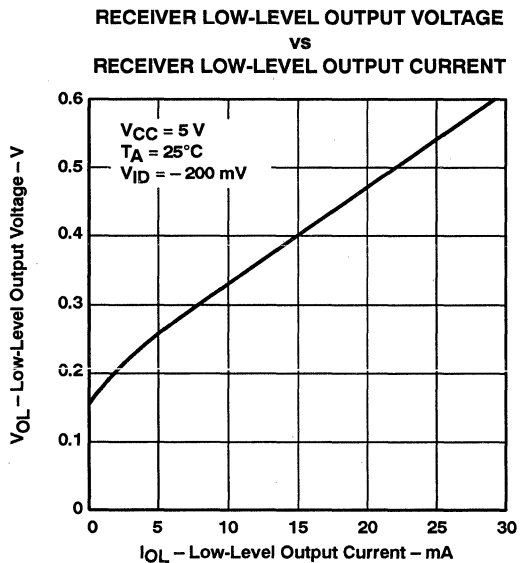


Figure 14

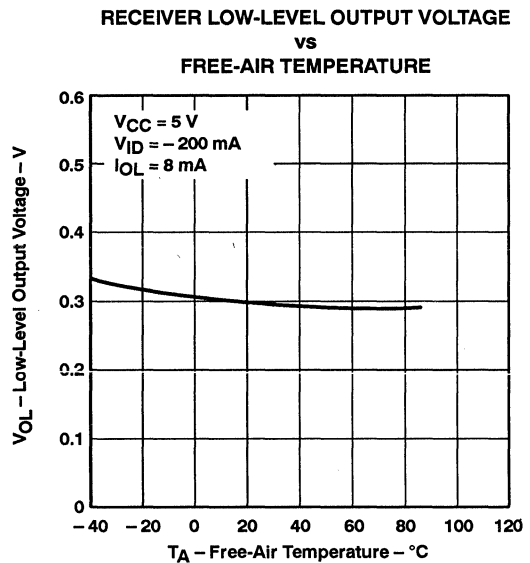


Figure 15

TEXAS  
INSTRUMENTS

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TYPICAL CHARACTERISTICS

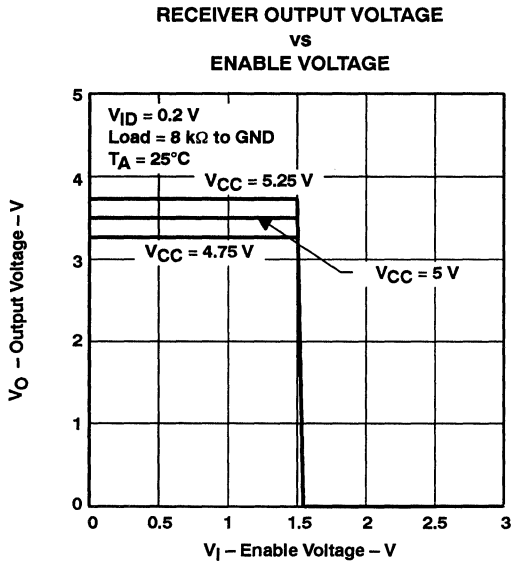


Figure 16

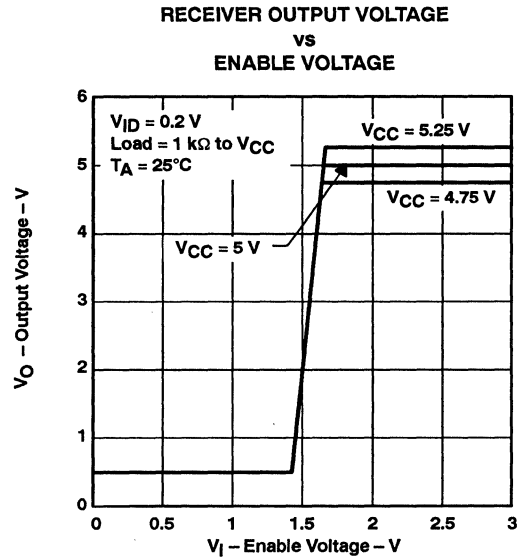
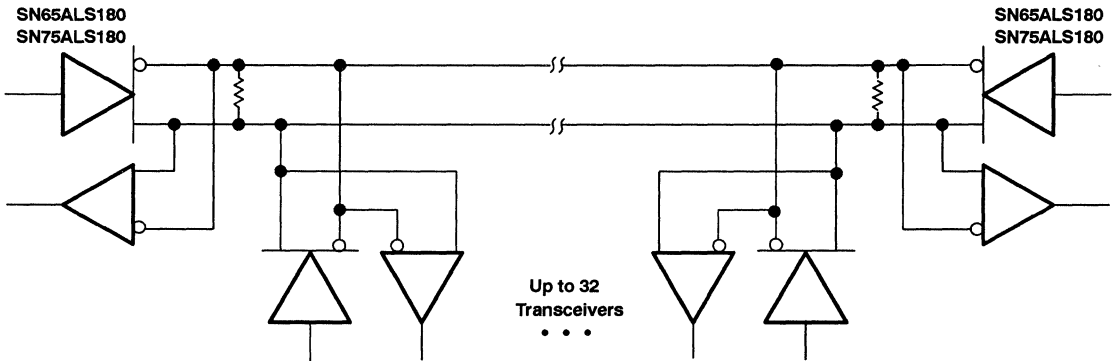


Figure 17

APPLICATION INFORMATION



NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

Figure 18. Typical Application Circuit



# SN75ALS181 DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS152 – D4060, DECEMBER 1992

- Meets EIA Standards RS-422-A, RS-485, and CCITT Recommendations V.11 and X.27
- Low Supply Current Requirements  
30 mA Max
- Driver Output Capacity . . .  $\pm 60$  mA
- Thermal Shutdown Protection
- Driver Common-Mode Output Voltage  
Range of  $-7$  V to  $12$  V
- Receiver Input Impedance . . .  $12$  k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Input Hysteresis . . .  $60$  mV Typ
- Receiver Common-Mode Input Voltage  
Range of  $\pm 12$  V
- Operates From Single 5-V Supply
- Glitch-Free Power-Up and Power-Down  
Protection

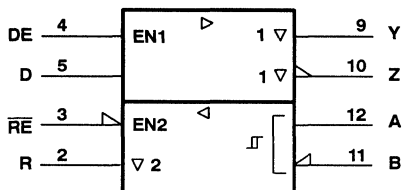
## description

The SN75ALS181 differential driver and receiver pair are monolithic integrated circuits designed for bidirectional data communication on multipoint bus transmission lines. The design provides for balanced transmission lines and meets EIA Standards RS-422-A and RS-485, and CCITT recommendations V.10, V.11, X.26, and X.27.

The SN75ALS181 combines a 3-state differential line driver and a differential input line receiver that operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, which can be externally connected together to function as a direction control. The driver differential outputs and the receiver differential inputs are connected to separate pins for greater flexibility and are designed to offer minimum loading to the bus when the driver is disabled or  $V_{CC} = 0$ . These ports feature wide positive and negative common-mode voltage changes making the device suitable for party-line applications.

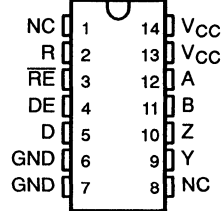
The SN75ALS181 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## N OR NS† PACKAGE (TOP VIEW)



NC – No internal connection

† The NS package is only available in left-end taped and reeled (order device SN75ALS181NSLE).

## Function Tables

### EACH DRIVER

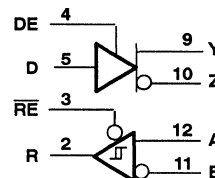
INPUT D	ENABLE DE	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	Z	Z

### EACH RECEIVER

DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2$ V	L	H
$-0.2$ V $< V_{ID} < 0.2$ V	L	?
$V_{ID} \leq -0.2$ V	L	L
X	H	Z

H = high level, L = low level, ? = indeterminate,  
X = irrelevant, Z = high impedance (off)

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
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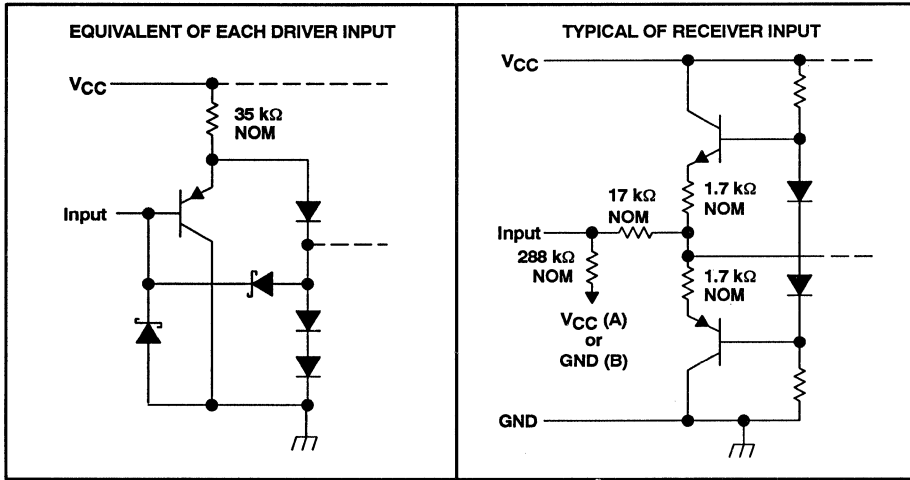
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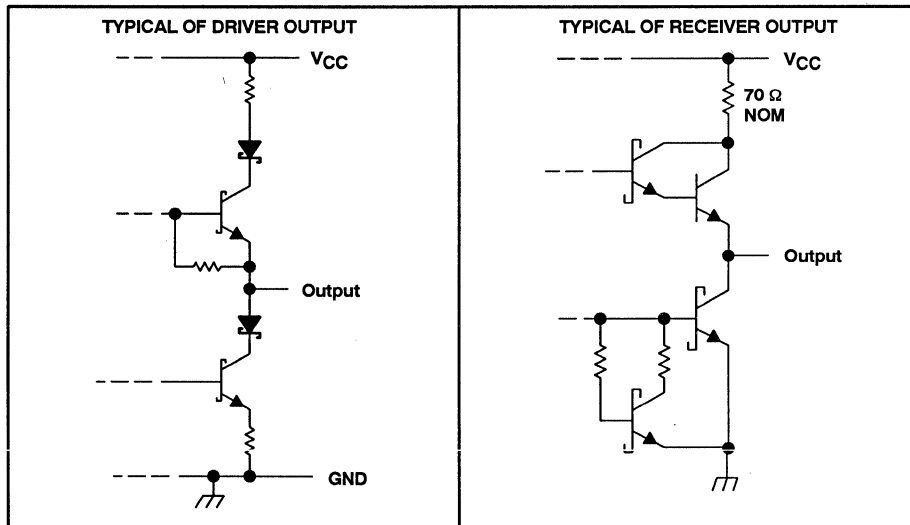
# SN75ALS181 DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS152 - D4060, DECEMBER 1992

## schematics of inputs



## schematics of outputs



# SN75ALS181 DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS152 – D4060, DECEMBER 1992

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, DE, $\overline{RE}$ , and D inputs	7 V
Output voltage range, driver	–9 V to 14 V
Input voltage range, receiver	–14 V to 14 V
Receiver differential input voltage range (see Note 2)	–14 V to 14 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential input voltage is measured at the noninverting terminal with respect to the inverting terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
N	1150 mW	9.2 mW/°C	736 mW
NS	625 mW	4.0 mW/°C	445 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Common-mode output voltage, $V_{OC}$ (see Note 3)	Driver	–7		12	V
Common-mode input voltage, $V_{IC}$ (see Note 3)	Receiver	–12		12	V
High-level input current, $V_{IH}$	D, DE, and $\overline{RE}$	2			V
Low-level input current, $V_{IL}$	D, DE, and $\overline{RE}$			0.8	V
Differential input voltage, $V_{ID}$				±12	V
High-level output current, $I_{OH}$	Driver			–60	mA
	Receiver			–400	µA
Low-level output current, $I_{OL}$	Driver			60	mA
	Receiver			8	mA
Operating free-air temperature, $T_A$		0		70	°C

NOTE 3: The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this table for common-mode output voltage level only.



# SN75ALS181 DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS152 – D4060, DECEMBER 1992

## DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V
$V_O$ Output voltage	$I_O = 0$	0		6	V
$ V_{OD1} $ Differential output voltage	$I_O = 0$	1.5		6	V
$ V_{OD2} $ Differential output voltage	$V_{CC} = 5 \text{ V}$ , $R_L = 100 \Omega$	See Figure 1		$1/2 V_{OD1}$	V
$ V_{OD2} $ Differential output voltage	$R_L = 54 \Omega$			2	
$ V_{OD2} $ Differential output voltage	$R_L = 54 \Omega$	1.5	2.3	5	V
$ V_{OD3} $ Differential output voltage	$V_{test} = -7 \text{ V to } 12 \text{ V}$ , See Figure 2	1.5		5	V
$\Delta V_{OD} $ Change in magnitude of differential output voltage (see Note 4)	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1			$\pm 0.2$	V
$V_{OC}$ Common-mode output voltage				3	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage (see Note 4)				-1	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage (see Note 4)				$\pm 0.2$	V
$I_{OZ}$ High-impedance-state output current	$V_O = -7 \text{ V to } 12 \text{ V}$ , See Note 5			$\pm 100$	$\mu\text{A}$
$I_{IH}$ High-level input current	$V_{IH} = 2.4 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_{IL} = 0.4 \text{ V}$			-100	$\mu\text{A}$
$I_{OS}$ Short-circuit output current	$V_O = -7 \text{ V}$			-250	mA
	$V_O = V_{CC}$			250	
	$V_O = 12 \text{ V}$			250	
	$V_O = 0 \text{ V}$			-150	
$I_{CC}$ Supply current (total package)	No load	Outputs enabled	21	30	mA
		Outputs disabled	14	21	

NOTES: 4.  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

5. This applies for both power on and power off. Refer to EIA Standard RS-485 for exact conditions.

switching characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{dD}$ Differential output delay time, $t_{dDH}$ or $t_{dDL}$	$R_L = 54 \Omega$ , $C_L = 50 \text{ pF}$ , See Figure 3	9	13	20	ns
$t_{sk(p)}$ Pulse skew ( $ t_{dDH} - t_{dDL} $ )		1	8		
$t_t$ Differential output transition time		3	10	16	
$t_{pZH}$ Output enable time to high level	$R_L = 110 \Omega$ , See Figure 4		36	53	ns
$t_{pZL}$ Output enable time to low level	$R_L = 110 \Omega$ , See Figure 5		39	56	ns
$t_{pHZ}$ Output disable time from high level	$R_L = 110 \Omega$ , See Figure 4		20	31	ns
$t_{pLZ}$ Output disable time from low level	$R_L = 110 \Omega$ , See Figure 5		9	20	ns

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .



# SN75ALS181 DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS152 – D4060, DECEMBER 1992

## RECEIVER SECTION

**electrical characteristics over recommended ranges of supply voltage, common-mode input voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage, differential input $V_O = 2.7\text{ V}$ , $I_O = -0.4\text{ mA}$			0.2	V
$V_{T-}$	Negative-going threshold voltage, differential input $V_O = 0.5\text{ V}$ , $I_O = 8\text{ mA}$ See Note 8	-0.2			V
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )		60		mV
$V_{IK}$	Input clamp voltage, $\overline{RE}$ $I_I = -18\text{ mA}$			-1.5	V
$V_{OH}$	High-level output voltage $V_{ID} = 200\text{ mV}$ , $I_{OH} = -400\text{ }\mu\text{A}$ , See Figure 6	2.7			V
$V_{OL}$	Low-level output voltage $V_{ID} = -200\text{ mV}$ , $I_{OL} = 8\text{ mA}$ , See Figure 6			0.45	V
$I_{OZ}$	High-impedance-state output current $V_O = 0.4\text{ V to } 2.4\text{ V}$			$\pm 20$	$\mu\text{A}$
$I_I$	Line input current Other input at 0 V, See Note 5			1	mA
				-0.8	
$I_{IH}$	High-level input current, $\overline{RE}$ $V_{IH} = 2.7\text{ V}$			20	$\mu\text{A}$
$I_{IL}$	Low-level input current, $\overline{RE}$ $V_{IL} = 0.4\text{ V}$			-100	$\mu\text{A}$
$r_i$	Input resistance		12		k $\Omega$
$I_{OS}$	Short-circuit output current $V_{ID} = 200\text{ mV}$ , $V_O = 0\text{ V}$	-15		-85	mA
$I_{CC}$	Supply current (total package) No load			21	mA
				30	
				14	21

NOTE 5: This applies for both power on and power off. Refer to EIA Standards RS-485 for exact conditions.

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature,  $C_L = 15\text{ pF}$  (unless otherwise noted) (see Figure 7)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PHL}$	Propagation delay time, high-to-low-level output $V_{ID} = -1.5\text{ V to } 1.5\text{ V}$	10	16	25	ns
$t_{PLH}$	Propagation delay time, low-to-high-level output $V_{ID} = -1.5\text{ V to } 1.5\text{ V}$	10	16	25	ns
$t_{sk(p)}$	Pulse skew ( $ t_{PLH} - t_{PHL} $ ) $V_{ID} = -1.5\text{ V to } 1.5\text{ V}$		1	8	ns
$t_{PZH}$	Output enable time to high level		7	15	ns
$t_{PZL}$	Output enable time to low level		9	19	ns
$t_{PHZ}$	Output disable time from high level		18	27	ns
$t_{PLZ}$	Output disable time from low level		10	15	ns

† All typical values are at  $V_{CC} = 5\text{ V}$  and  $T_A = 25^\circ\text{C}$ .





# SN75ALS181 DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS152 – D4060, DECEMBER 1992

## PARAMETER MEASUREMENT INFORMATION

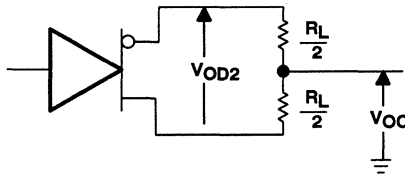


Figure 1. Driver Test Circuit,  $V_{OD}$  and  $V_{OC}$

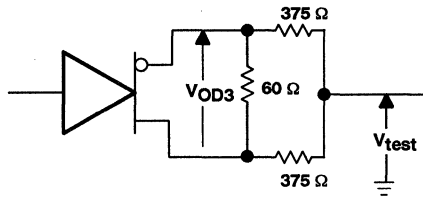
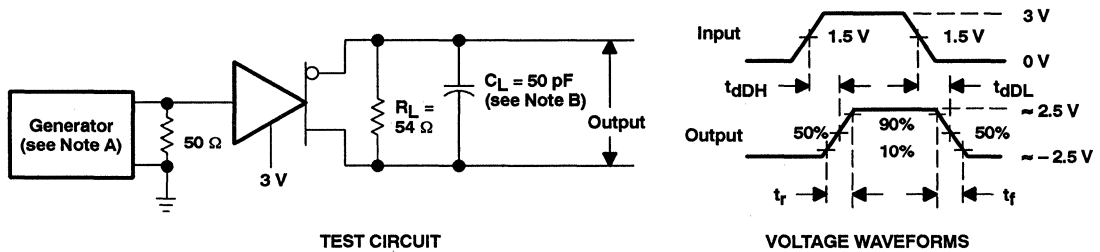


Figure 2. Driver Circuit,  $V_{OD3}$



- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

Figure 3. Driver Differential-Output Delay and Transition Times

PARAMETER MEASUREMENT INFORMATION

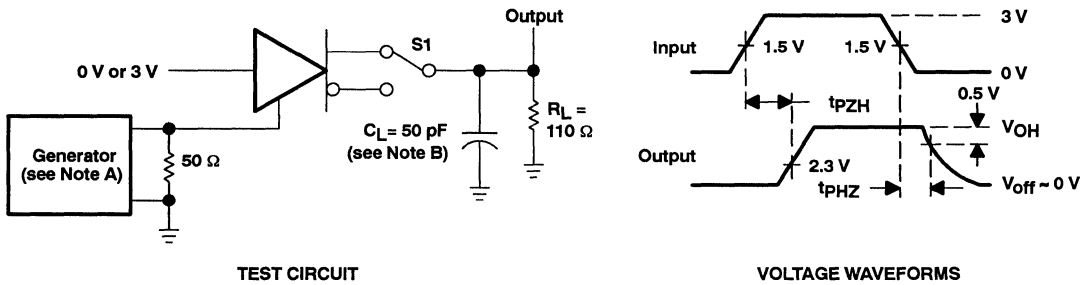


Figure 4. Driver Enable and Disable Times

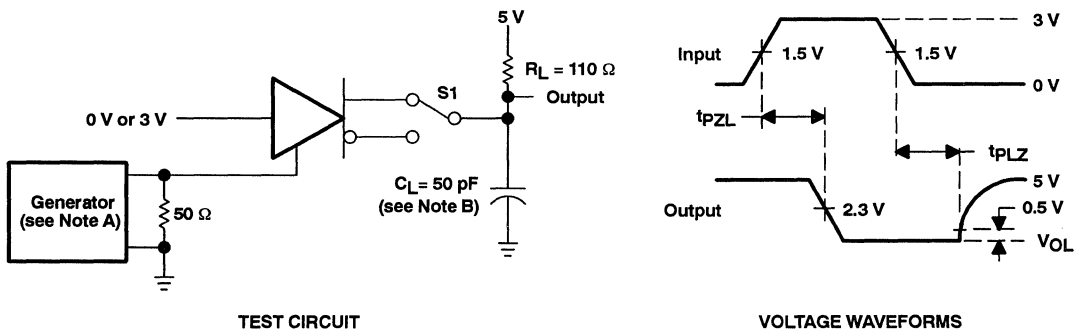


Figure 5. Driver Enable and Disable Times

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

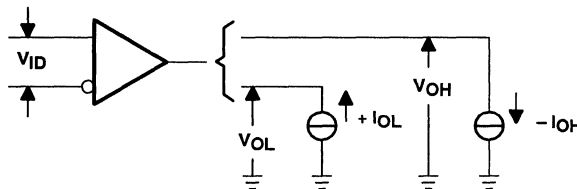


Figure 6. Receiver,  $V_{OH}$  and  $V_{OL}$

# SN75ALS181 DIFFERENTIAL DRIVER AND RECEIVER PAIR

SLLS152 – D4060, DECEMBER 1992

## PARAMETER MEASUREMENT INFORMATION

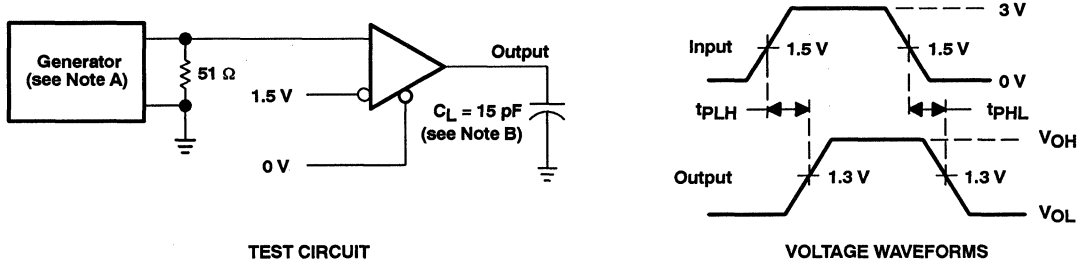


Figure 7. Receiver Propagation Delay Times

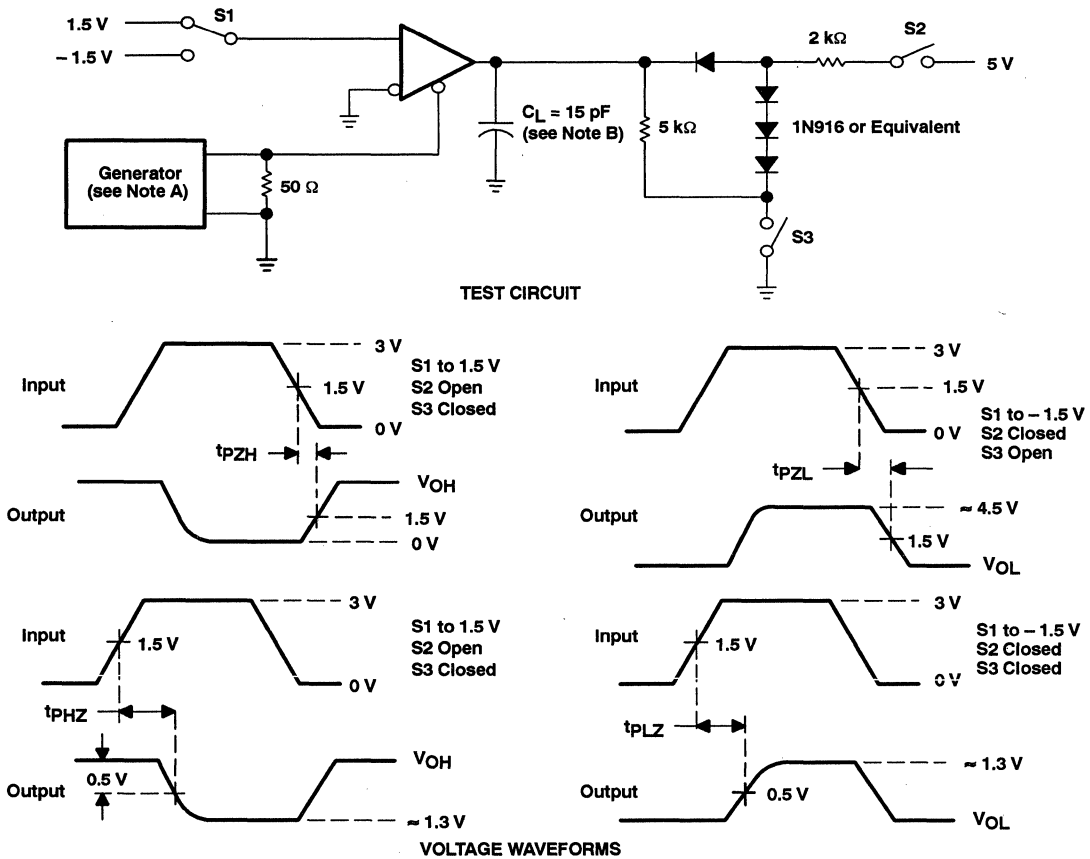


Figure 8. Receiver Output Enable and Disable Times

NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  3.6 ns,  $Z_0 = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A – D1292, OCTOBER 1972 – REVISED MARCH 1993

- Single 5-V Supply
- Differential Line Operation
- Dual Channels
- TTL Compatibility
- $\pm 15\text{-V}$  Common-Mode Input Voltage Range
- $\pm 15\text{-V}$  Differential Input Voltage Range
- Individual Channel Strobes
- Built-In Optional Line-Termination Resistor
- Individual Frequency Response Controls
- Designed for Use With Dual Differential Drivers SN55183 and SN75183
- Designed to Be Interchangeable With National Semiconductor DS7820A and DS8820A

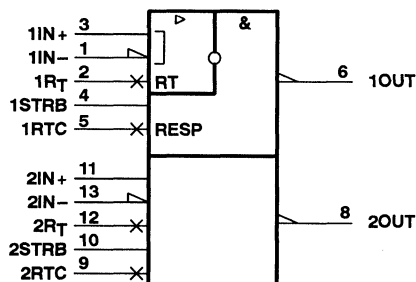
## description

The DS8820A, SN55182, and SN75182 dual differential line receivers are designed to sense small differential signals in the presence of large common-mode noise. These devices give TTL-compatible output signals as a function of the polarity of the differential input voltage. The frequency response of each channel may be easily controlled by a single external capacitor to provide immunity to differential noise spikes. The output goes to a high level when the inputs are open circuited. A strobe input is provided which, when in the low level, disables the receiver and forces the output to a high level.

The receiver is of monolithic single-chip construction, and both halves of the dual circuits use common power supply and ground terminals.

The SN55182 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The DS8820A and SN75182 are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

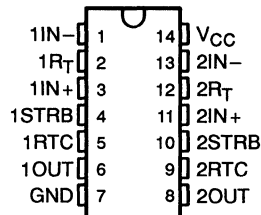
## logic symbol†



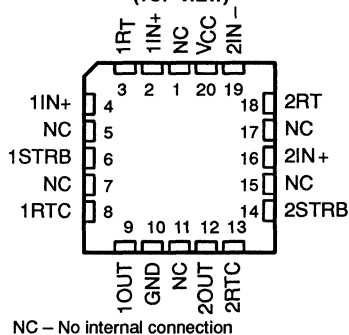
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, N, and W packages.

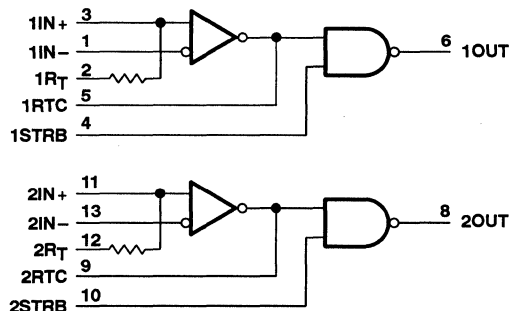
SN55182 . . . J OR W PACKAGE  
DS8820A, SN75182 . . . D OR N PACKAGE  
(TOP VIEW)



SN55182 . . . FK PACKAGE  
(TOP VIEW)



## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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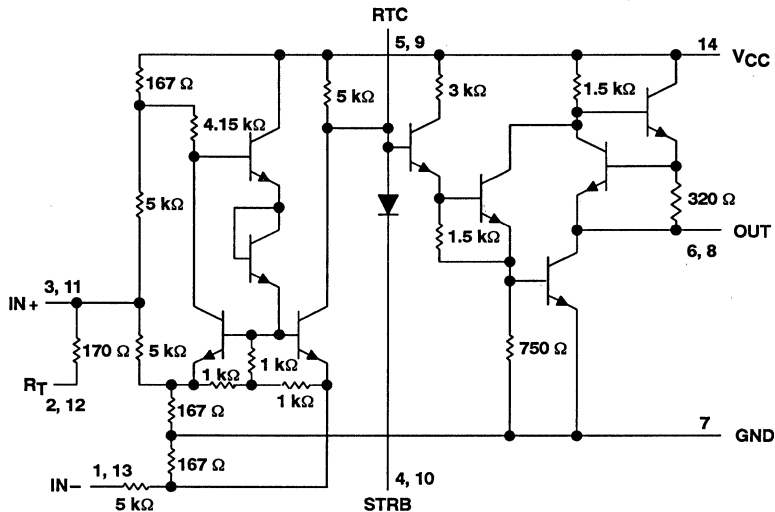
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# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A—D1292, OCTOBER 1972—REVISED MARCH 1993

## schematic (each receiver)



Resistor values shown are nominal.  
Pin numbers shown are for the D, J, N, and W packages.

FUNCTION TABLE

STRB	$V_{ID}$	OUT
L	X	H
H	H	H
H	L	L

H =  $V_I \geq V_{IH \text{ min}}$  or  $V_{ID}$  more positive than  $V_{TH \text{ max}}$   
L =  $V_I \leq V_{IL \text{ max}}$  or  $V_{ID}$  more negative than  $V_{TL \text{ max}}$   
X = irrelevant

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55182	DS8820A SN75182	UNIT
Supply voltage, $V_{CC1}$ (see Note 1)	8	8	V
Common-mode input voltage	$\pm 20$	$\pm 20$	V
Differential input voltage (see Note 2)	$\pm 20$	$\pm 20$	V
Strobe input voltage	8	8	V
Output sink current	50	50	mA
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package		260	°C
Case temperature for 60 seconds: FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	300	300	°C

NOTES: 1. All voltage values, except differential voltages, are with respect to network ground terminal.  
2. Differential voltage values are at the noninverting terminal with respect to the inverting terminal.

TEXAS  
INSTRUMENTS

# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A - D1292, OCTOBER 1972 - REVISED MARCH 1993

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/ $^\circ\text{C}$	608 mW	—
FK†	1375 mW	11.0 mW/ $^\circ\text{C}$	880 mW	275 mW
J†	1375 mW	11.0 mW/ $^\circ\text{C}$	880 mW	275 mW
N	1150 mW	9.2 mW/ $^\circ\text{C}$	736 mW	—
W†	1000 mW	8.0 mW/ $^\circ\text{C}$	640 mW	200 mW

† In the FK, J, and W packages, SN55182 chips are alloy mounted.

## recommended operating conditions

	SN55182			DS8820A, SN75182			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.5	5	5.5	V
Common-mode input voltage, $V_{IC}$	$\pm 15$			$\pm 15$			V
High-level strobe input voltage, $V_{IH}(\text{STRB})$	2.1		5.5	2.1		5.5	V
Low-level strobe input voltage, $V_{IL}(\text{STRB})$	0		0.9	0		0.9	V
High-level output current, $I_{OH}$	-400			-400			$\mu\text{A}$
Low-level output current, $I_{OL}$	16			16			mA
Operating free-air temperature, $T_A$	-55		125	0		70	$^\circ\text{C}$

# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A - D1292, OCTOBER 1972 - REVISED MARCH 1993

electrical characteristics over recommended ranges of  $V_{CC}$ ,  $V_{IC}$ , and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{T+}$	Positive-going input threshold voltage	$V_O = 2.5$ V, $I_{OH} = -400$ $\mu$ A	$V_{IC} = -3$ V to 3 V			0.5	V
			$V_{IC} = -15$ V to 15 V			1	
$V_{T-}$	Negative-going input threshold voltage	$V_O = 0.4$ V, $I_{OL} = 16$ mA	$V_{IC} = -3$ V to 3 V			-0.5	V
			$V_{IC} = -15$ V to 15 V			-1	
$V_{OH}$	High-level output voltage	$V_{ID} = 1$ V, $I_{OH} = -400$ $\mu$ A	$V_{STRB} = 2.1$ V,	2.5	4.2	5.5	V
			$V_{STRB} = 0.4$ V,	2.5	4.2	5.5	
$V_{OL}$	Low-level output voltage	$V_{ID} = -1$ V, $I_{OL} = 16$ mA	$V_{STRB} = 2.1$ V,		0.25	0.4	V
$I_i$	Inverting input		$V_{IC} = 15$ V		3	4.2	mA
			$V_{IC} = 0$		0	-0.5	
			$V_{IC} = -15$ V		-3	-4.2	
	Noninverting input		$V_{IC} = 15$ V		5	7	mA
			$V_{IC} = 0$		-1	-1.4	
			$V_{IC} = -15$ V		-7	-9.8	
$I_{STRB(H)}$	High-level strobe current	$V_{STRB} = 5.5$ V				5	$\mu$ A
$I_{STRB(L)}$	Low-level strobe current	$V_{STRB} = 0$			-1	-1.4	mA
$r_i$	Input resistance		Inverting input	3.6	5		k $\Omega$
			Noninverting input	1.8	2.5		k $\Omega$
$R_T$	Line terminating resistance	$T_A = 25^\circ\text{C}$		120	170	250	$\Omega$
$I_{OS}$	Short-circuit output current	$V_{CC} = 5.5$ V,	$V_O = 0$	-2.8	-4.5	-6.7	mA
$I_{CC}$	Supply current (average per receiver)		$V_{IC} = 15$ V, $V_{ID} = -1$ V		4.2	6	mA
			$V_{IC} = 0$ , $V_{ID} = -0.5$ V		6.8	10.2	
			$V_{IC} = -15$ V, $V_{ID} = -1$ V		9.4	14	

† Unless otherwise noted,  $V_{STRB} \geq 2.1$  V or open.

‡ All typical values are at  $V_{CC} = 5$  V,  $V_{IC} = 0$ , and  $T_A = 25^\circ\text{C}$ .

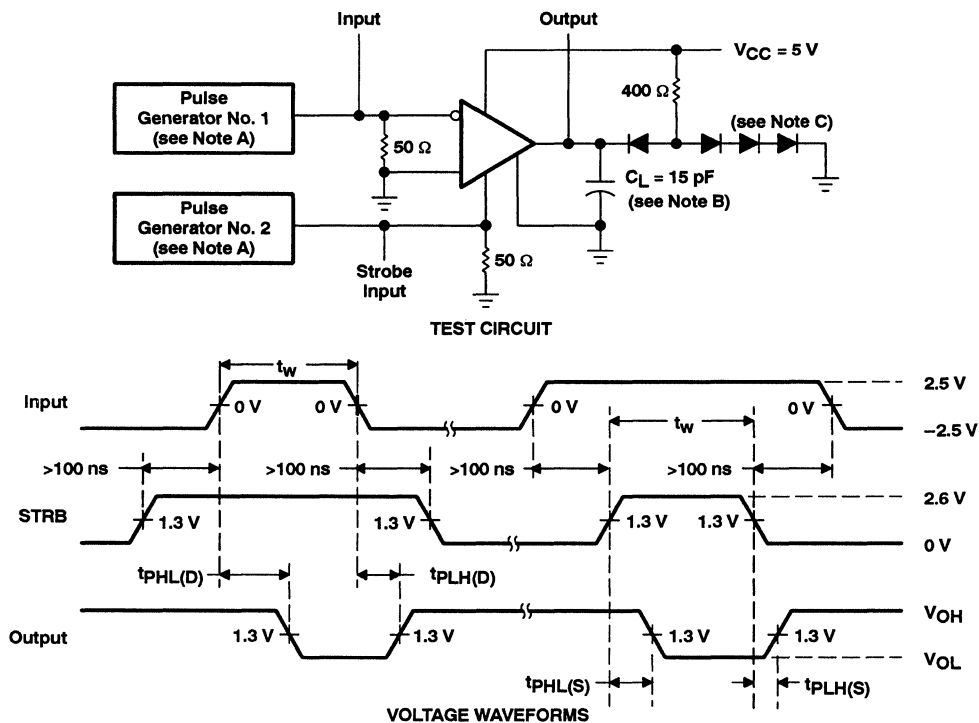
## switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH(D)}$	Propagation delay time, low-to-high-level output from differential input	$R_L = 400$ $\Omega$ , $C_L = 15$ pF, See Figure 1		18	40	ns
$t_{PHL(D)}$	Propagation delay time, high-to-low-level output from differential input			31	45	ns
$t_{PLH(S)}$	Propagation delay time, low-to-high-level output from STRB input			9	30	ns
$t_{PHL(S)}$	Propagation delay time, high-to-low-level output from STRB input			15	25	ns

# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A - D1292, OCTOBER 1972 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The pulse generators have the following characteristics:  $Z_O = 50 \Omega$ ,  $t_r \leq 10 \text{ ns}$ ,  $t_f \leq 10 \text{ ns}$ ,  $t_W = 0.5 \pm 0.1 \mu\text{s}$ ,  $\text{PRR} \leq 1 \text{ MHz}$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.

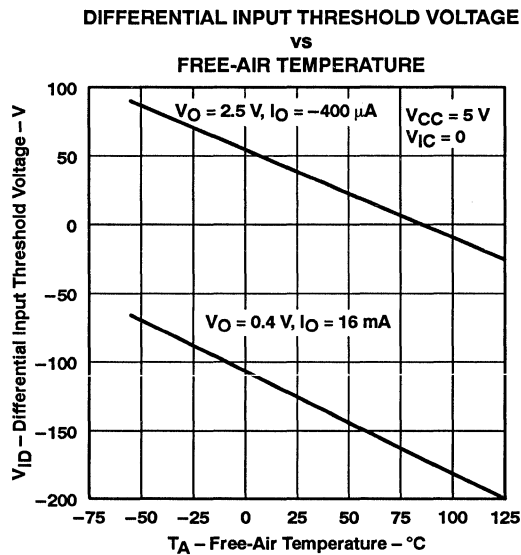
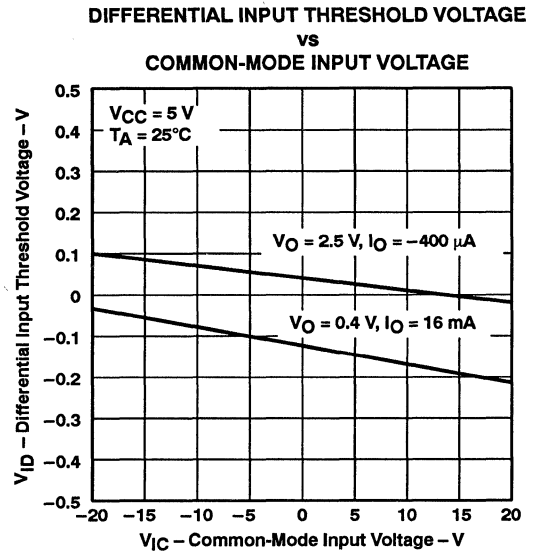
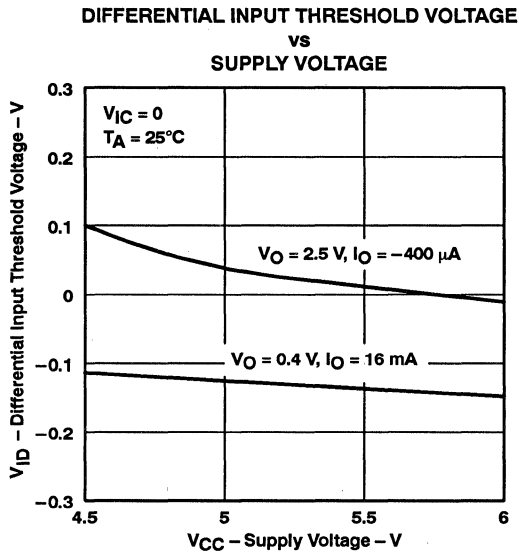
**Figure 1. Test Circuit and Voltage Waveforms**



# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A - D1292, OCTOBER 1972 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS†



† Data for temperatures below 0°C and above 70°C are applicable to SN55182 circuits only.

TYPICAL CHARACTERISTICS†

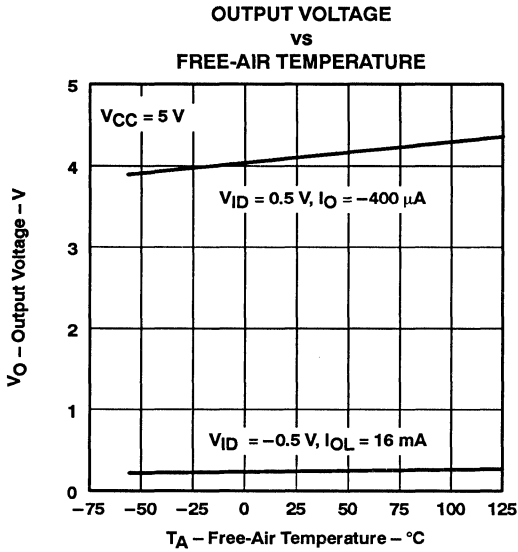


Figure 5

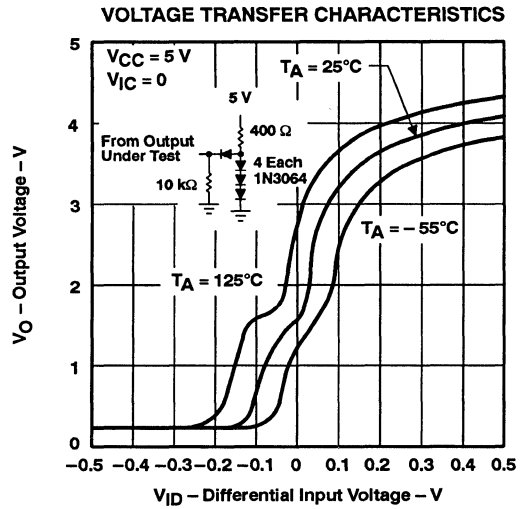


Figure 6

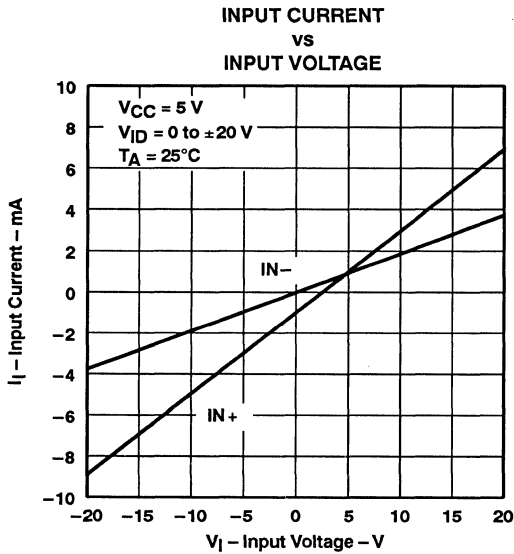


Figure 7

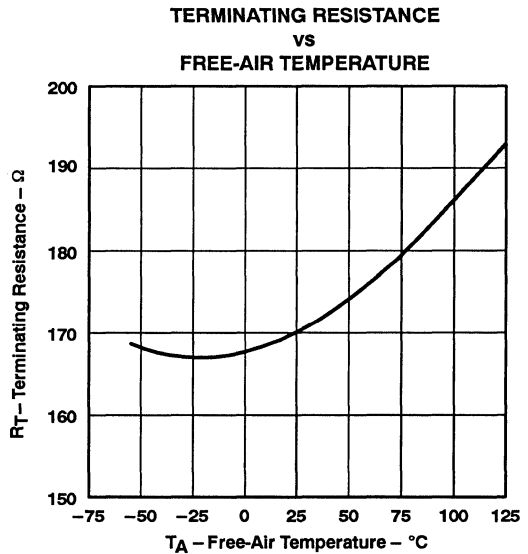


Figure 8

† Data for temperatures below 0°C and above 70°C are applicable to SN55182 circuits only.

# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A - D1292, OCTOBER 1972 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS†

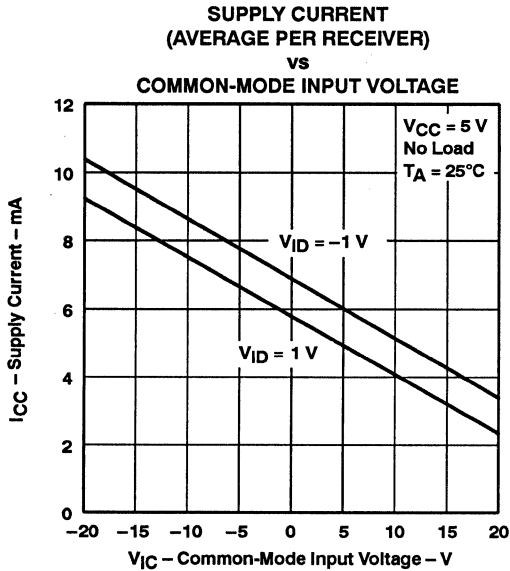


Figure 9

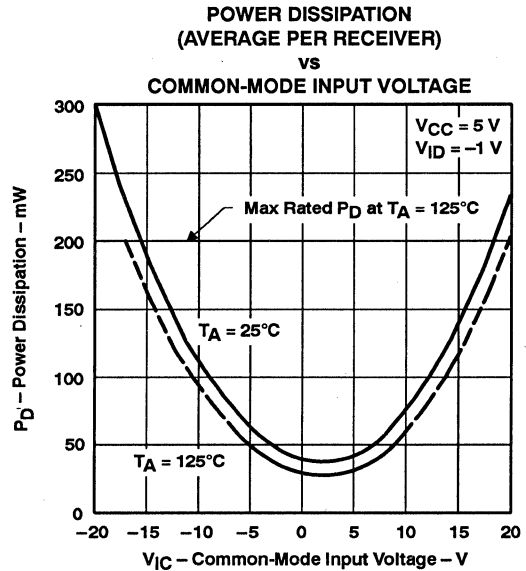
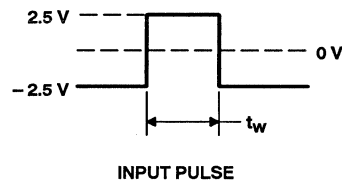
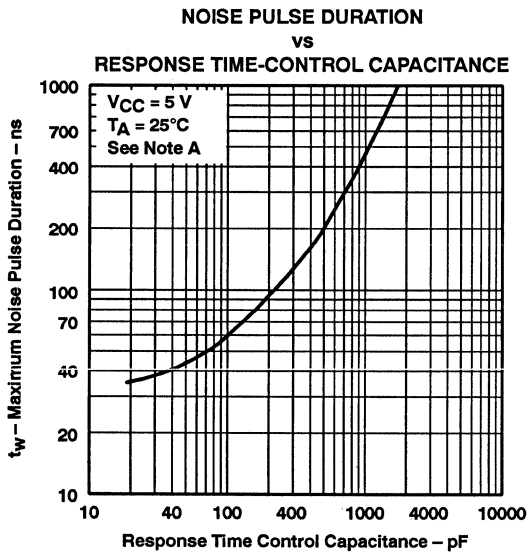


Figure 10



NOTE A: Figure 11 shows the maximum duration of the illustrated pulse that can be applied differently without the output changing from the low to high level.

Figure 11

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  are applicable to SN55182 circuits only.

# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A – D1292, OCTOBER 1972 – REVISED MARCH 1993

## TYPICAL CHARACTERISTICS†

PROPAGATION DELAY TIMES FROM  
DIFFERENTIAL INPUT  
vs  
FREE-AIR TEMPERATURE

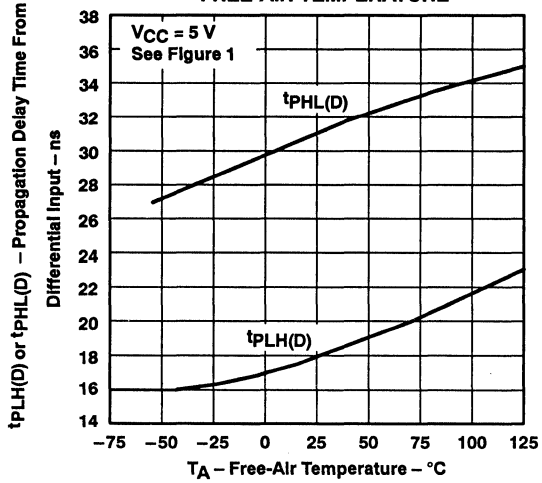


Figure 12

PROPAGATION DELAY TIMES FROM  
STROBE INPUT  
vs  
FREE-AIR TEMPERATURE

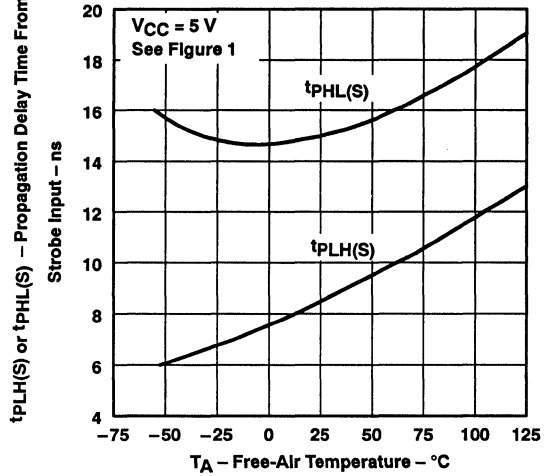


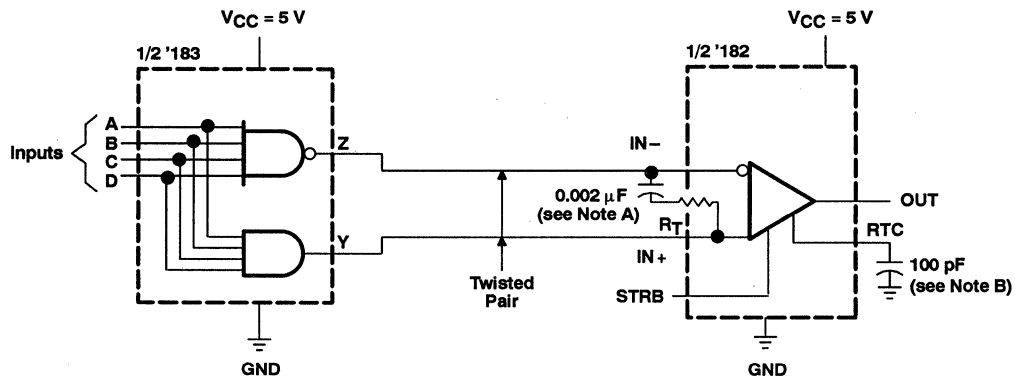
Figure 13

† Data for temperatures below 0°C and above 70°C are applicable to SN55182 circuits only.

# DS8820A, SN55182, SN75182 DUAL DIFFERENTIAL LINE RECEIVERS

SLLS092A - D1292, OCTOBER 1972 - REVISED MARCH 1993

## APPLICATION INFORMATION



NOTES: A. When the inputs are open circuited, the output will be high. A capacitor may be used for dc isolation of the line-terminating resistor. At the frequency of operation, the impedance of the capacitor should be relatively small.

Example: let  $f = 5 \text{ MHz}$   
 $C = 0.002 \mu\text{F}$

$$Z_C = \frac{1}{2\pi f C} = \frac{1}{2\pi (5 \times 10^6) (0.002 \times 10^{-6})}$$

$$Z_C \approx 16 \Omega$$

B. Use of a capacitor to control response time is optional.

**Figure 14. Transmission of Digital Data Over Twisted-Pair Line**

# DS8830, SN55183, SN75183 DUAL DIFFERENTIAL LINE DRIVERS

SLLS093A - D1292, OCTOBER 1972 - REVISED MARCH 1993

- Single 5-V Supply
- Differential Line Operation
- Dual Channels
- TTL Compatibility
- Short-Circuit Protection of Outputs
- Output Clamp Diodes to Terminate Line Transients
- High-Current Outputs
- Quad Inputs
- Single-Ended or Differential AND/NAND Outputs
- Designed for Use With Dual Differential Drivers SN55182 and SN75182
- Designed to Be Interchangeable With National Semiconductor DS7830 and DS8830

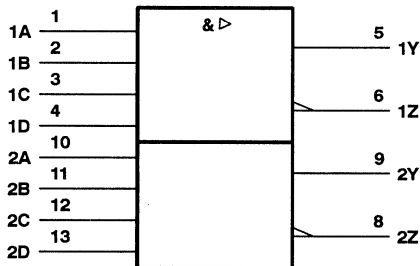
## description

The DS8830, SN55183, and SN75183 dual differential line drivers are designed to provide differential output signals with high-current capability for driving balanced lines, such as twisted-pair, at normal line impedances without high power dissipation. These devices may be used as TTL expander/phase splitters, as the output stages are similar to TTL totem-pole outputs.

The driver is of monolithic single-chip construction, and both halves of the dual circuits use common power supply and ground terminals.

The SN55183 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The DS8830 and SN75183 are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

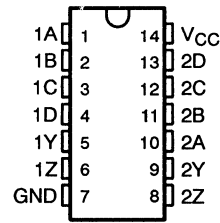
## logic symbol†



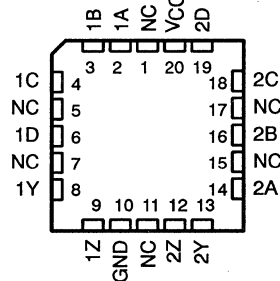
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, N, and W packages.

SN55183 . . . J OR W PACKAGE  
DS8830, SN75183 . . . D OR N PACKAGE  
(TOP VIEW)

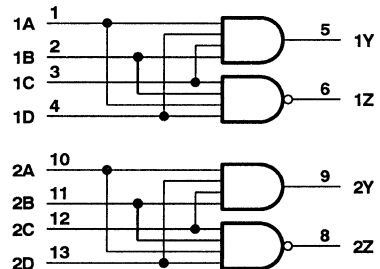


SN55183 . . . FK PACKAGE  
(TOP VIEW)



NC - No internal connection

## logic diagram (positive logic)



positive logic:  $Y = ABCD$ ,  $Z = \overline{ABCD}$

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

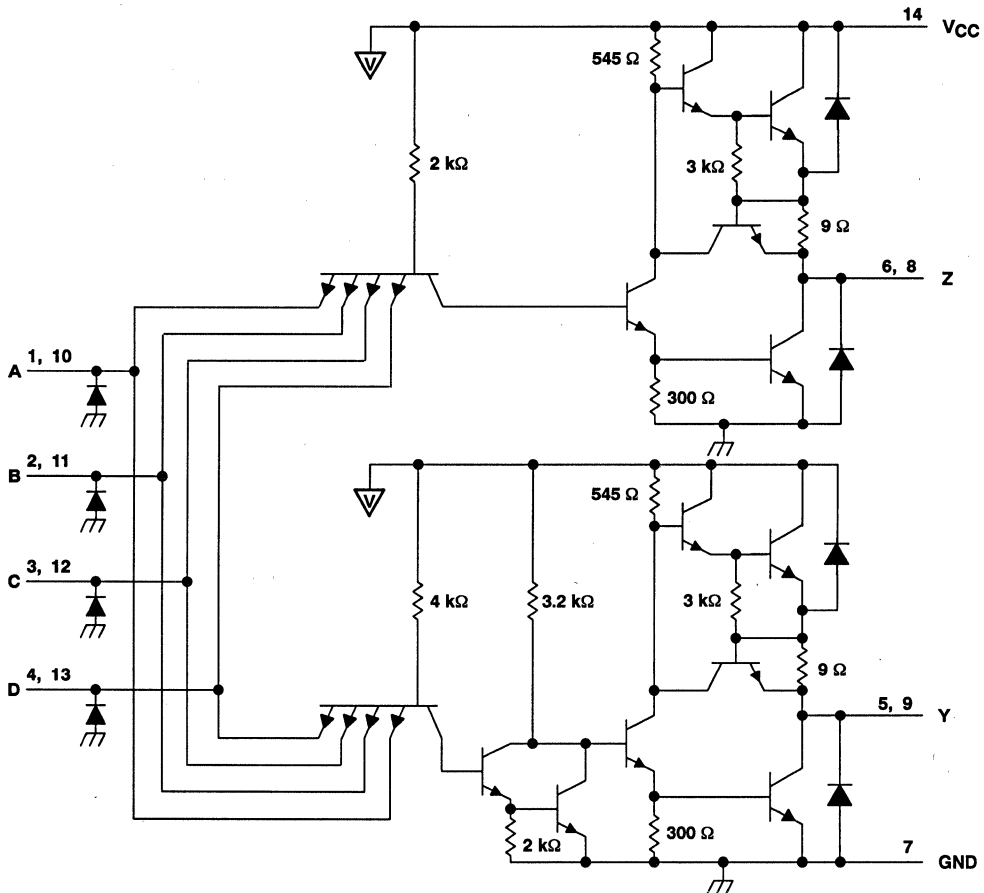
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2-705

# DS8830, SN55183, SN75183 DUAL DIFFERENTIAL LINE DRIVERS

SLLS093A - D1292, OCTOBER 1972 - REVISED MARCH 1993

## schematic (each driver)



Resistor values shown are nominal.  
Pin numbers shown are for the D, N, J, and W packages.

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# DS8830, SN55183, SN75183 DUAL DIFFERENTIAL LINE DRIVERS

SLLS093A – D1292, OCTOBER 1972 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55183	DS8830 SN75183	UNIT
Supply voltage, $V_{CC}$ (see Note 1)	7	7	V
Input voltage	5.5	5.5	V
Duration of output short circuit (see Note 2)	1	1	s
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package		260	°C
Case temperature for 60 seconds: FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300	300	°C

NOTES: 1. All voltage values, except differential voltages, are with respect to network ground terminal.

2. Not more than one output should be shorted to ground at a time.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	–
FK†	1375 mW	11.0 mW/°C	880 mW	275 mW
J†	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	–
W	1000 mW	8.0 mW/°C	640 mW	200 mW

† In the FK and J packages, SN55183 chips are alloy mounted and SN75183 chips are glass mounted.

## recommended operating conditions

	SN55183			DS8830, SN75183			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			2			V
Low-level input voltage, $V_{IL}$				0.8			V
High-level output current, $I_{OH}$				-40			mA
Low-level output current, $I_{OL}$				40			mA
Operating free-air temperature, $T_A$	-55			0			°C



**DS8830, SN55183, SN75183**  
**DUAL DIFFERENTIAL LINE DRIVERS**

SLLS083A - D1292, OCTOBER 1972 - REVISED MARCH 1993

**electrical characteristics over recommended ranges of  $V_{CC}$  and operating free-air temperature (unless otherwise noted)**

PARAMETER			TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage		Y (AND) output	$V_{IH} = 2\text{ V}$ ,	$I_{OH} = -0.8\text{ mA}$	2.4		V
		$V_{IH} = 2\text{ V}$ ,		$I_{OH} = -40\text{ mA}$	1.8	3.3		
$V_{OL}$	Low-level output voltage	$V_{IL} = 0.8\text{ V}$ ,		$I_{OL} = 32\text{ mA}$	0.2		V	
		$V_{IL} = 0.8\text{ V}$ ,		$I_{OL} = 40\text{ mA}$	0.22	0.4		
$V_{OH}$	High-level output voltage	Z (NAND) output	$V_{IL} = 0.8\text{ V}$ ,	$I_{OH} = -0.8\text{ mA}$	2.4		V	
			$V_{IL} = 0.8\text{ V}$ ,	$I_{OH} = -40\text{ mA}$	1.8	3.3		
$V_{OL}$	Low-level output voltage		$V_{IH} = 2\text{ V}$ ,	$I_{OL} = 32\text{ mA}$	0.2		V	
			$V_{IH} = 2\text{ V}$ ,	$I_{OL} = 40\text{ mA}$	0.22	0.4		
$I_{IH}$	High-level input current		$V_{IH} = 2.4\text{ V}$		120		$\mu\text{A}$	
$I_I$	Input current at maximum input voltage		$V_{IH} = 5.5\text{ V}$		2		mA	
$I_{IL}$	Low-level input current		$V_{IL} = 0.4\text{ V}$		-4.8		mA	
$I_{OS}$	Short-circuit output current‡		$V_{CC} = 5\text{ V}$ ,	$T_A = 125^\circ\text{C}$	-40	-100	-120	mA
$I_{CC}$	Supply current (average per driver)		$V_{CC} = 5\text{ V}$ , No load	All inputs at 5 V,	10		18	mA

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

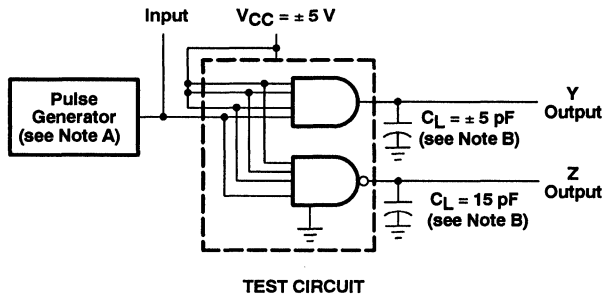
‡ Not more than one output should be shorted to ground at a time, and duration of the short circuit should not exceed one second.

**switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

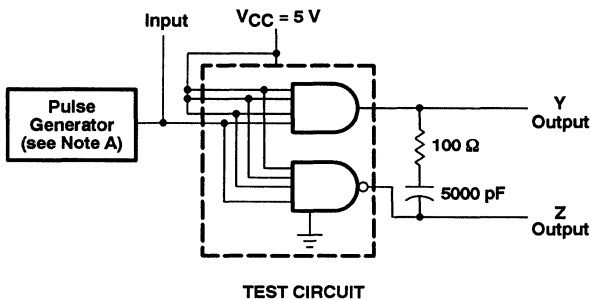
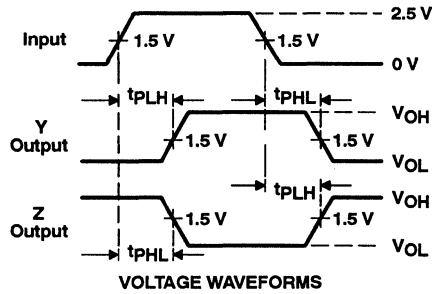
PARAMETER			TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level Y output		AND gates	$C_L = 15\text{ pF}$ , See Figure 1 (a)		8		12
$t_{PHL}$	Propagation delay time, high-to-low-level Y output	12				18		
$t_{PLH}$	Propagation delay time, low-to-high-level Z output	NAND gates	6			12	ns	
$t_{PHL}$	Propagation delay time, high-to-low-level Z output		6			8		
$t_{PLH}$	Propagation delay time, low-to-high-level differential output	Y output with respect to Z output	$Z_L = 100\ \Omega$ in series with 5000 pF, See Figure 1 (b)		9		16	ns
$t_{PHL}$	Propagation delay time, high-to-low-level differential output				8		16	



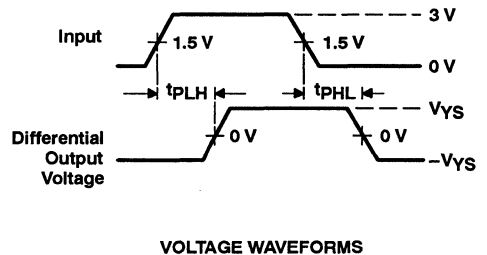
PARAMETER MEASUREMENT INFORMATION



(a) OUTPUTS Y AND Z



(b) DIFFERENTIAL OUTPUT



- NOTES: A. The pulse generators have the following characteristics:  $Z_O = 50 \Omega$ ,  $t_r \leq 10 \text{ ns}$ ,  $t_f \leq 10 \text{ ns}$ ,  $t_w = 0.5 \mu\text{s}$ ,  $\text{PRR} \leq 1 \text{ MHz}$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. Waveforms are monitored on an oscilloscope with  $R_{in} \geq 1 \text{ M}\Omega$ .

Figure 1. Test Circuits and Voltage Waveforms

# DS8830, SN55183, SN75183 DUAL DIFFERENTIAL LINE DRIVERS

SLLS093A—D1292, OCTOBER 1972—REVISED MARCH 1993

## TYPICAL CHARACTERISTICS†

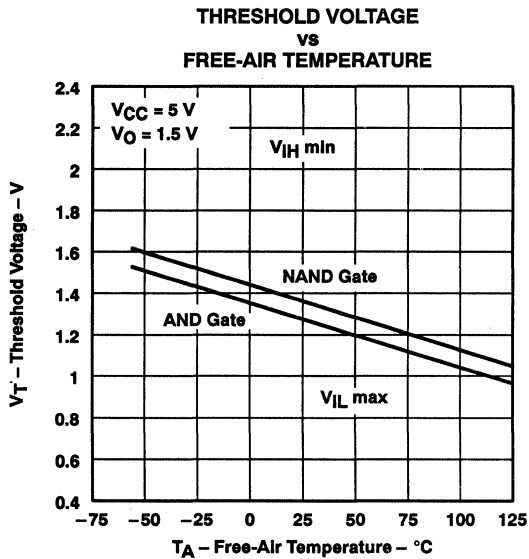


Figure 2

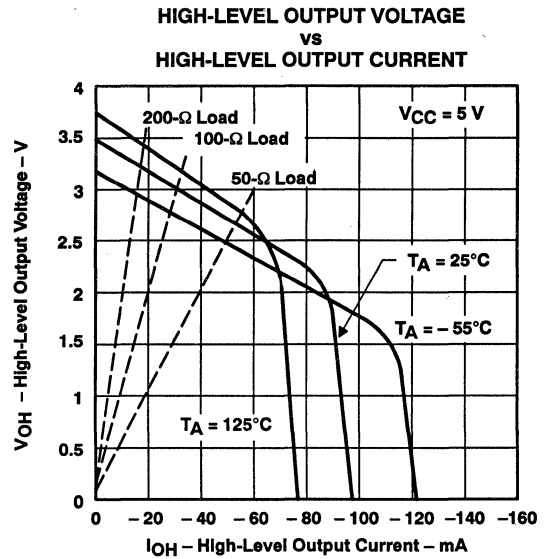


Figure 3

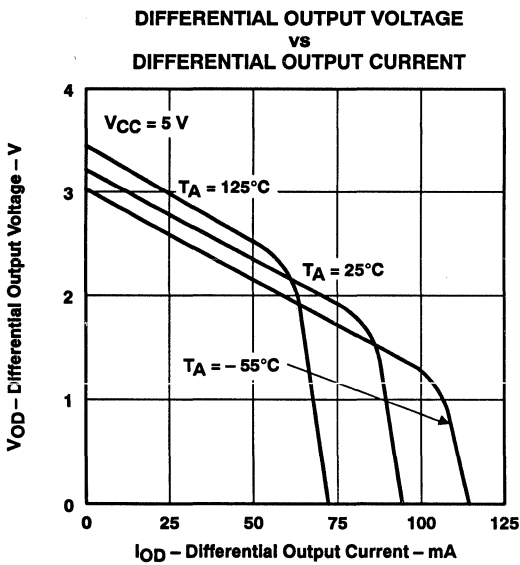


Figure 4

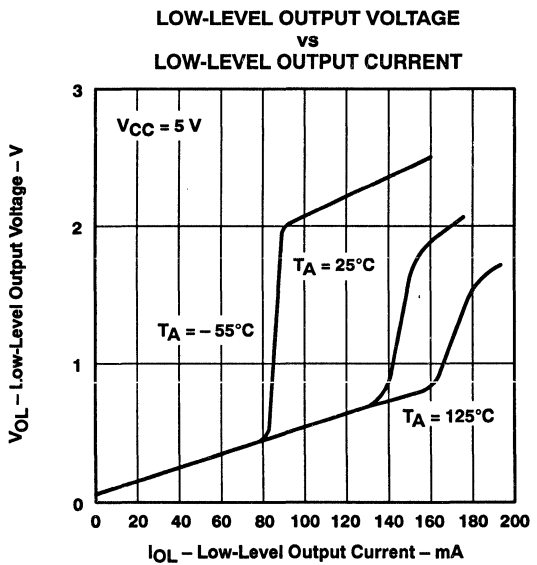
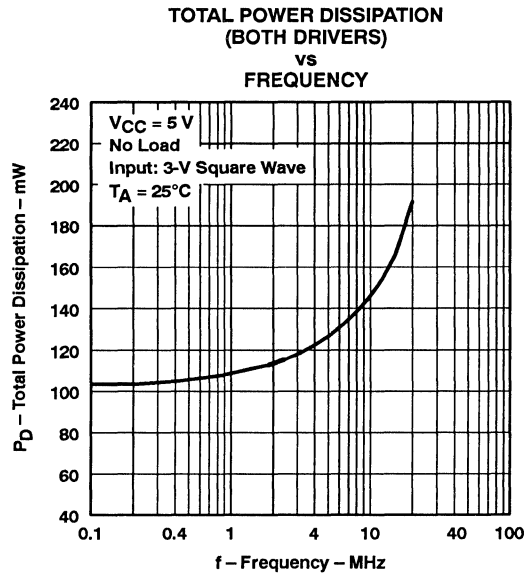
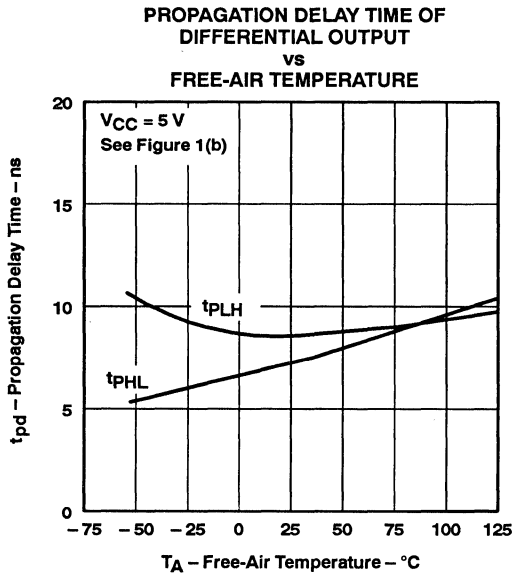


Figure 5

† Data for temperatures below 0°C and above 70°C are applicable to SN55183 circuits only.

TYPICAL CHARACTERISTICS†

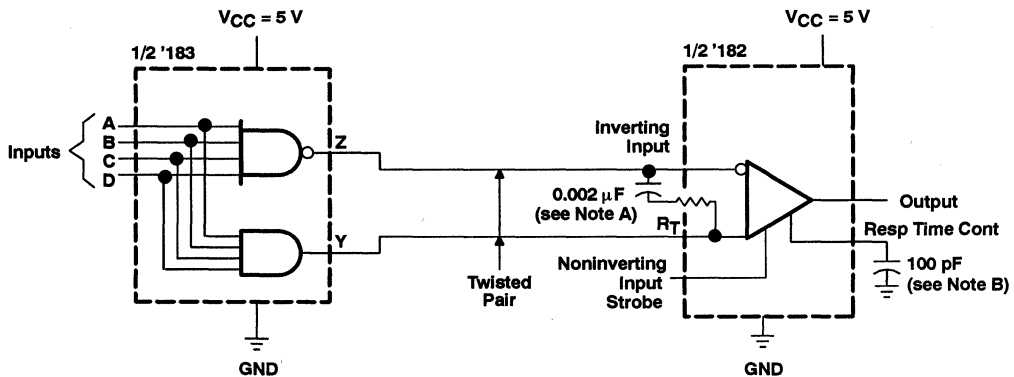


† Data for temperatures below 0°C and above 70°C are applicable to SN55183 circuits only.

# DS8830, SN55183, SN75183 DUAL DIFFERENTIAL LINE DRIVERS

SLLS093A - D1292, OCTOBER 1972 - REVISED MARCH 1993

## APPLICATION INFORMATION



NOTES: A. When the inputs are open circuited, the output will be high. A capacitor may be used for dc isolation of the line-terminating resistor. At the frequency of operation, the impedance of the capacitor should be relatively small.

Example: let  $f = 5 \text{ MHz}$   
 $C = 0.002 \mu\text{F}$

$$Z_C = \frac{1}{2\pi f C} = \frac{1}{2\pi (5 \times 10^6) (0.002 \times 10^{-6})}$$

$$Z_C \sim 16 \Omega$$

B. Use of a capacitor to control response time is optional.

**Figure 8. Transmission of Digital Data Over Twisted-Pair Line**

# SN65C185, SN75C185 LOW-POWER MULTIPLE DRIVERS AND RECEIVERS

SLLS065B, D3325, AUGUST 1989 – REVISED MARCH 1993

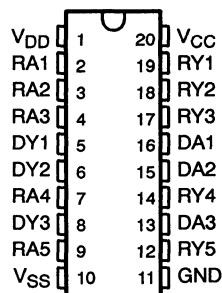
- Meets Standard EIA-232-D (Revision of RS-232-C)
- Single Chip With Easy Interface Between UART and Serial Port Connector
- Less Than 9-mW Power Consumption
- Wide Driver Supply Voltage . . . 4.5 V to 13.2 V
- Driver Output Slew Rate Limited to 30 V/ $\mu$ s Max
- Receiver Input Hysteresis . . . 1100 mV Typ
- Push-Pull Receiver Outputs
- On-Chip Receiver 1- $\mu$ s Noise Filter
- ESD Protection Exceeds 500 V Per MIL-STD-883C, Method 3015

## description

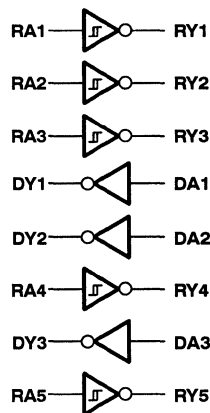
The SN65C185 and SN75C185 are low-power BI-MOS devices containing three independent drivers and five receivers that are used to interface data terminal equipment (DTE) with data circuit-terminating equipment (DCE). The SN65C185 and SN75C185 will typically replace one SN75188 and two SN75189 devices. These devices have been designed to conform to Standards ANSI/EIA-232-D-1986, which supersedes RS-232-C. The three drivers and five receivers of the SN65C185 and SN75C185 are similar to those of the SN75C188 quad drivers and SN75C189A quad receivers, respectively. The drivers have a controlled output slew rate that is limited to a maximum of 30 V/ $\mu$ s and the receivers have filters that reject input noise pulses that are shorter than 1  $\mu$ s. Both these features eliminate the need for external components.

The SN65C185 and SN75C185 have been designed using low-power techniques in a BI-MOS technology. In most applications, the receivers contained in these devices will interface to single inputs of peripheral devices such as ACEs, UARTS, or microprocessors. By using sampling, such peripheral devices are usually insensitive to the transition times of the input signals. If this is not the case, or for other uses, it is recommended that the SN65C185 and SN75C185 receiver outputs be buffered by single Schmitt input gates or single gates of the HCMOS, ALS, or 74F logic families.

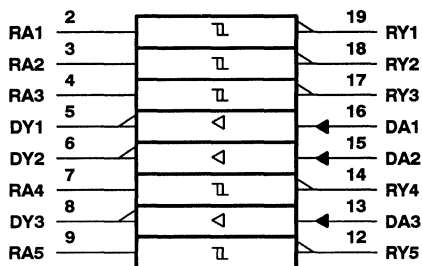
DW OR N PACKAGE  
(TOP VIEW)



logic diagram (positive logic)



logic symbol



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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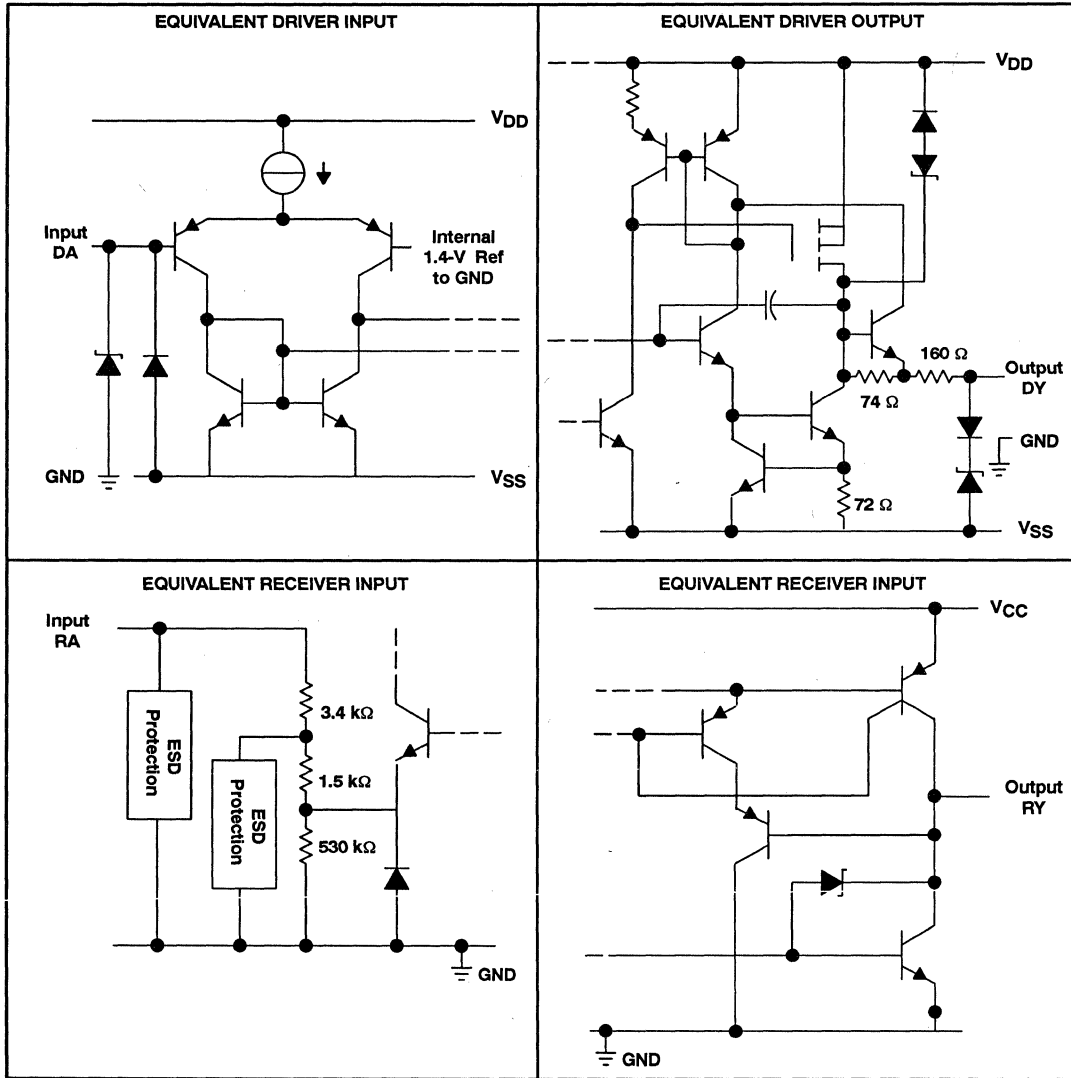
# SN65C185, SN75C185 LOW-POWER MULTIPLE DRIVERS AND RECEIVERS

SLLS065B, D3325, AUGUST 1989 – REVISED MARCH 1993

## description (continued)

The SN65C185 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75C185 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## equivalent schematics of inputs and outputs



All resistor values are nominal.

# SN65C185, SN75C185 LOW-POWER MULTIPLE DRIVERS AND RECEIVERS

SLLS065B, D3325, AUGUST 1989 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{DD}$ (see Note 1)	13.5 V
Supply voltage, $V_{SS}$	-13.5 V
Supply voltage, $V_{CC}$	7 V
Input voltage range: Driver	$V_{SS}$ to $V_{DD}$
Receiver	-30 V to 30 V
Output voltage range: Driver	$V_{SS} - 6$ V to $V_{DD} + 6$ V
Receiver	-0.3 V to $V_{CC} + 0.3$ V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : SN65C185	-40°C to 85°C
SN75C185	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 85^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	585 mW
N	1150 mW	9.2 mW/°C	598 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{DD}$		4.5	12	13.2	V
Supply voltage, $V_{SS}$		-4.5	-12	-13.2	V
Supply voltage, $V_{CC}$		4.5	5	6	V
Input voltage, $V_I$ (see Note 2)	Driver	$V_{SS} + 2$		$V_{DD}$	V
	Receiver	-25		25	
High-level input voltage, $V_{IH}$	Driver	2		0.8	V
Low-level input voltage, $V_{IL}$					
High-level output current, $I_{OH}$	Receiver			-1	mA
High-level output current, $I_{OL}$				3.2	mA
Operating free-air temperature, $T_A$	SN65C185	-40		85	°C
	SN75C185	0		70	

NOTE 2: The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only, e.g., if -10 V is a maximum, the typical value is a more negative voltage.

## supply currents

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$I_{DD}$	Supply current from $V_{DD}$	No load, All inputs at 2 V or 0.8 V	$V_{DD} = 5$ V, $V_{SS} = -5$ V		115	200	$\mu\text{A}$
			$V_{DD} = 12$ V, $V_{SS} = -12$ V		115	200	
$I_{SS}$	Supply current from $V_{SS}$	No load, All inputs at 2 V or 0.8 V	$V_{DD} = 5$ V, $V_{SS} = -5$ V		-115	-200	$\mu\text{A}$
			$V_{DD} = 12$ V, $V_{SS} = -12$ V		-115	-200	
$I_{CC}$	Supply current from $V_{CC}$	No load All inputs at 0 or 5 V	$V_{DD} = 5$ V, $V_{SS} = -5$ V			750	$\mu\text{A}$
			$V_{DD} = 12$ V, $V_{SS} = -12$ V			750	





# SN65C185, SN75C185 LOW-POWER MULTIPLE DRIVERS AND RECEIVERS

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## DRIVER SECTION

electrical characteristics over operating free-air temperature range,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage	$V_{IL} = 0.8\text{ V}$ , $R_L = 3\text{ k}\Omega$ , See Figure 1	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$	4	4.5		V
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$	10	10.8		
$V_{OL}$	Low-level output voltage (see Note 2)	$V_{IH} = 0.8\text{ V}$ , $R_L = 3\text{ k}\Omega$ , See Figure 1	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$		-4.4	-4	V
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$		-10.7	-10	
$I_{IH}$	High-level input current	$V_I = 5\text{ V}$ , See Figure 2				1	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0$ , See Figure 2				-1	$\mu\text{A}$
$I_{OS(H)}$	High-level short-circuit output current (see Note 3)	$V_I = 0.8\text{ V}$ , See Figure 1	$V_O = 0$ or $V_O = V_{SS}$ ,	-4.5	-12	-19.5	mA
$I_{OS(L)}$	Low-level short-circuit output current (see Note 3)	$V_I = 2\text{ V}$ , See Figure 1	$V_O = 0$ or $V_O = V_{DD}$ ,	4.5	12	19.5	mA
$r_o$	Output resistance	$V_{DD} = V_{SS} = V_{CC} = 0$ , See Note 4	$V_O = -2\text{ V}$ to $2\text{ V}$ ,	300	400		$\Omega$

† All typical values are at  $T_A = 25^\circ\text{C}$ .

NOTES: 2. The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only, e.g., if  $-10\text{ V}$  is a maximum, the typical value is a more negative voltage.

3. Not more than one output should be shorted at one time.

4. Test conditions are those specified by EIA-232-D.

switching characteristics,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$ ,  $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT	
$t_{PLH}$	Propagation delay time, high-to-low-level output (see Note 5)	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Figure 3	$C_L = 15\text{ pF}$ ,		1.2	3	$\mu\text{s}$	
$t_{PHL}$	Propagation delay time, low-to-high-level output (see Note 5)				2.5	3.5	$\mu\text{s}$	
$t_{TLH}$	Transition time, low-to-high-level output				0.53	2	3.2	$\mu\text{s}$
$t_{THL}$	Transition time, high-to-low-level output				0.53	2	3.2	$\mu\text{s}$
$t_{TLH}$	Transition time, low-to-high-level output (see Note 6)			$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Figure 3	$C_L = 2500\text{ pF}$ ,		1	3
$t_{THL}$	Transition time, high-to-low-level output (see Note 6)		1			3	$\mu\text{s}$	
SR	Output slew rate (see Note 6)	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Figure 3	$C_L = 15\text{ pF}$ ,	4	10	30	$\text{V}/\mu\text{s}$	

NOTES: 5.  $t_{PHL}$  and  $t_{PLH}$  include the additional time due to on-chip slew rate and is measured at the 50% points.

6. Measured between 3-V and  $-3\text{ V}$  points of output waveform (EIA-232-D conditions), all unused inputs tied either high or low.



# SN65C185, SN75C185 LOW-POWER MULTIPLE DRIVERS AND RECEIVERS

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## RECEIVER SECTION

**electrical characteristics over operating free-air temperature range,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	See Figure 5	1.6	2.1	2.55	V
$V_{T-}$	Negative-going threshold voltage	See Figure 5	0.65	1	1.25	V
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )		600	1100		mV
$V_{OH}$	High-level output voltage	$V_I = 0.75\text{ V}$ , $I_{OH} = -20\text{ }\mu\text{A}$ , See Figure 5 and Note 7	3.5			V
		$V_I = 0.75\text{ V}$ , $I_{OH} = -1\text{ mA}$ , See Figure 5	2.8	4.4		
		$V_{CC} = 4.5\text{ V}$ $V_{CC} = 5\text{ V}$ $V_{CC} = 5.5\text{ V}$	3.8	4.9	5.4	
$V_{OL}$	Low-level output voltage	$V_I = 3\text{ V}$ , $I_{OL} = 3.2\text{ mA}$ , See Figure 5		0.17	0.4	V
$I_{IH}$	High-level output current	$V_I = 3\text{ V}$	0.43	0.55	1	mA
		$V_I = 25\text{ V}$	3.6	4.6	8.3	
$I_{IL}$	Low-level output current	$V_I = -3\text{ V}$	-0.43	-0.55	-1	mA
		$V_I = -25\text{ V}$	-3.6	-5.0	-8.3	
$I_{OS(H)}$	Short-circuit output at high level	$V_I = 0.75\text{ V}$ , $V_O = 0$ , See Figure 4		-8	-15	mA
$I_{OS(L)}$	Short-circuit output at low level	$V_I = V_{CC}$ , $V_O = V_{CC}$ , See Figure 4		13	25	mA

† All typical values are at  $T_A = 25^\circ\text{C}$ .

NOTE 7: If the inputs are left unconnected, the receiver interprets this as an input low, and the receiver outputs will remain in the high state.

**switching characteristics,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$R_L = 5\text{ k}\Omega$ , $C_L = 50\text{ pF}$ , See Figure 6		3	4	$\mu\text{s}$
$t_{PHL}$	Propagation delay time, high-to-low-level output			3	4	$\mu\text{s}$
$t_{TLH}$	Transition time, low-to-high-level output			300	450	ns
$t_{THL}$	Transition time, high-to-low-level output			100	300	ns
$t_w(N)$	Duration of longest pulse rejected as noise (see Note 8)	$R_L = 5\text{ k}\Omega$ , $C_L = 50\text{ pF}$ , See Figure 6	1		4	$\mu\text{s}$

NOTE 8: The intent of this specification is that any input pulse of less than  $1\text{ }\mu\text{s}$  will have no effect on the output, and any pulse duration of greater than  $4\text{ }\mu\text{s}$  will cause the output to change state twice. Reaction to a pulse duration between  $1\text{ }\mu\text{s}$  and  $4\text{ }\mu\text{s}$  is uncertain.



# SN65C185, SN75C185 LOW-POWER MULTIPLE DRIVERS AND RECEIVERS

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## PARAMETER MEASUREMENT INFORMATION

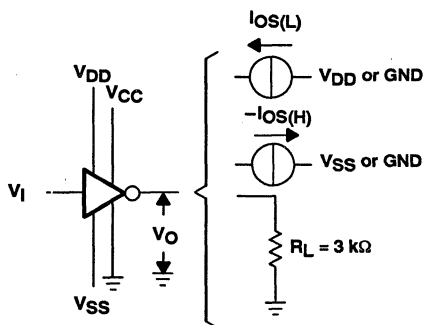


Figure 1. Driver Test Circuit for  $V_{OH}$ ,  $V_{OL}$ ,  $I_{OS(H)}$ , and  $I_{OS(L)}$

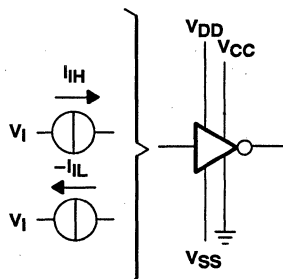
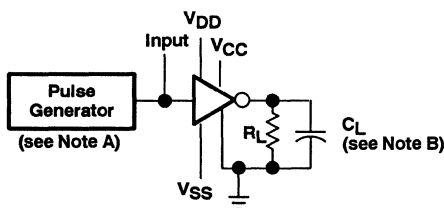


Figure 2. Driver Test Circuit for  $I_{IH}$  and  $I_{IL}$



TEST CIRCUIT

VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $t_W = 25\ \mu\text{s}$ ,  $\text{PRR} = 20\ \text{kHz}$ ,  $Z_O = 50\ \Omega$ ,  $t_r = t_f < 50\ \text{ns}$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 3. Driver Test Circuit and Voltage Waveforms

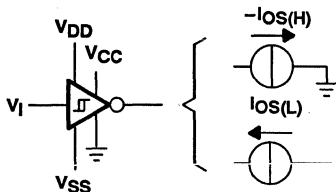


Figure 4. Receiver Test Circuit for  $I_{OS(H)}$  and  $I_{OS(L)}$

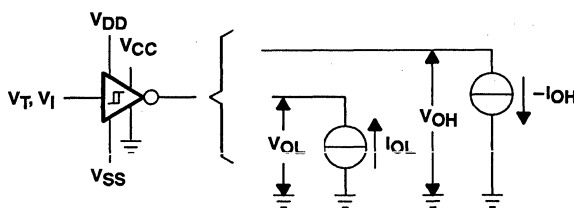
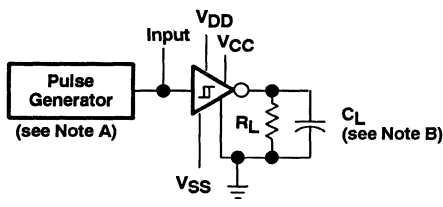


Figure 5. Receiver Test Circuit for  $V_T$ ,  $V_{OH}$ , and  $V_{OL}$

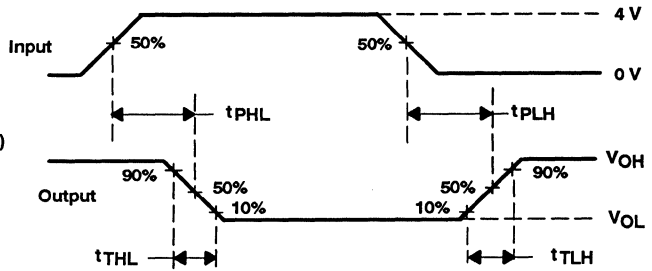
# SN65C185, SN75C185 LOW-POWER MULTIPLE DRIVERS AND RECEIVERS

SLLS065B, D3325, AUGUST 1989 – REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

- NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu s$ ,  $PRR = 20 \text{ kHz}$ ,  $Z_0 = 50 \Omega$ ,  $t_r = t_f < 50 \text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.

Figure 6. Receiver Propagation and Transition Times

## APPLICATION INFORMATION

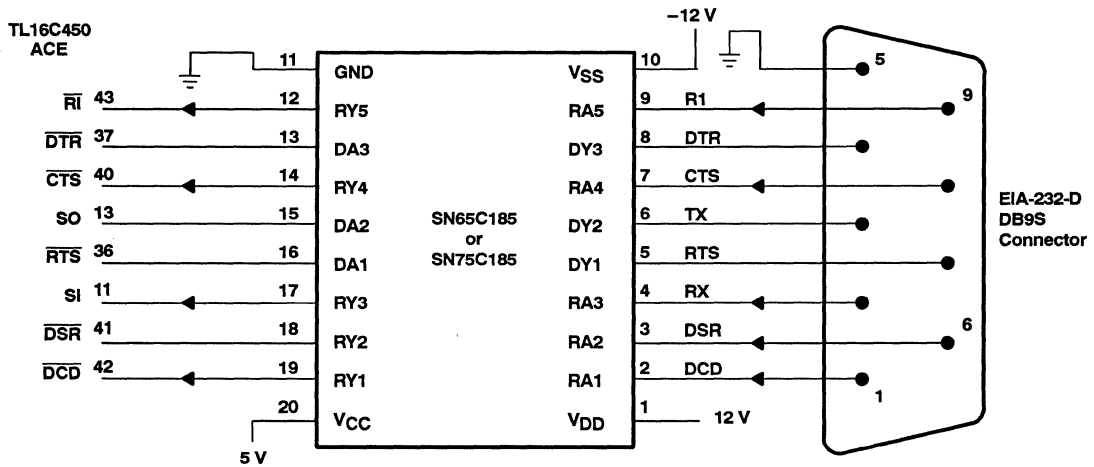


Figure 7. Typical Connection

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# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

SLLS068B – D3389, FEBRUARY 1990 – REVISED MARCH 1993

- Meets Standards RS-232-C, EIA-232-D, and CCITT V.28
- Four Independent Drivers and Receivers
- Loopback Mode Functionally Self-Tests Drivers and Receivers Without Disconnection From Line
- Driver Slew Rate Limited to 30 V/μs Max
- Built-In Receiver 1-μs Noise Filter
- Internal Thermal Overload Protection
- EIA-232-D Inputs and Outputs Withstand ±30 V
- Low Supply Current . . . 2.5 mA Typ
- ESD Protection Exceeds 4000 V Per MIL-STD-883C Method 3015

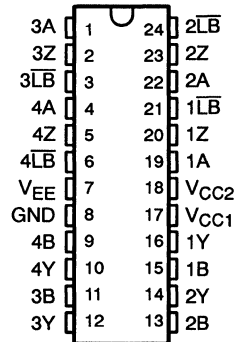
## description

The SN75186 is a low-power bipolar device containing four driver/receiver pairs designed to interface data terminal equipment (DTE) with data circuit-terminating equipment (DCE). Additionally, the SN75186 has a loopback mode that may be used by a data communication system to perform a functional self test on each driver/receiver pair, removing the need to locally disconnect cables and install a loopback connector. Flexibility of control is ensured by each driver/receiver pair having its own loopback control input. The SN75186 is designed to conform to standards RS-232-C, its revision ANSI/EIA-232-D-1986, and CCITT V.28.

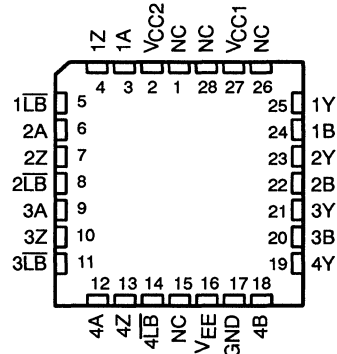
The maximum slew rate is limited to 30 V/μs at the driver outputs and drives a capacitive load of 2500 pF at 20 kBaud. The receivers have input filters that disregard input noise pulses shorter than 1 μs. The SN75186 is a robust device capable of withstanding ±30 V at driver outputs and at receiver inputs whether powered or unpowered. This device has an internal ESD protection rated at 4 kV to prevent functional failures.

The SN75186 is characterized for operation from 0°C to 70°C.

**DW PACKAGE  
(TOP VIEW)**



**FN PACKAGE  
(TOP VIEW)**



NC – No internal connection

## Function Tables

**EACH RECEIVER**

LOOPBACK LB	INPUTS		OUTPUT Z
	A	B†	
H	X	H	L
H	X	L	H
L	L	X	L
L	H	X	H

**EACH DRIVER**

LOOPBACK LB	INPUT A	OUTPUT Y†
H	H	L
H	L	H
L	X	L

† Voltages are RS-232-C, EIA-232-D, and V.28 levels

H = high level, L = low level, X = irrelevant

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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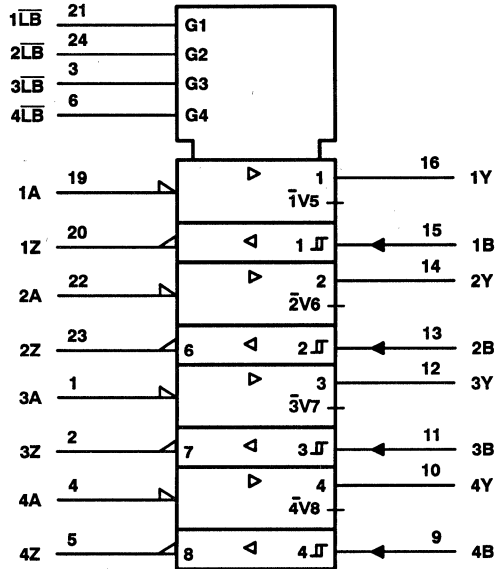
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# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

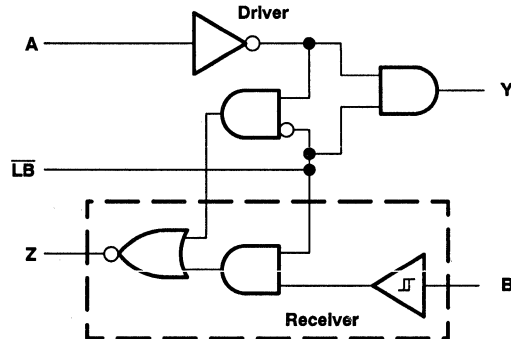
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logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12. Pin numbers shown are for the DW package.

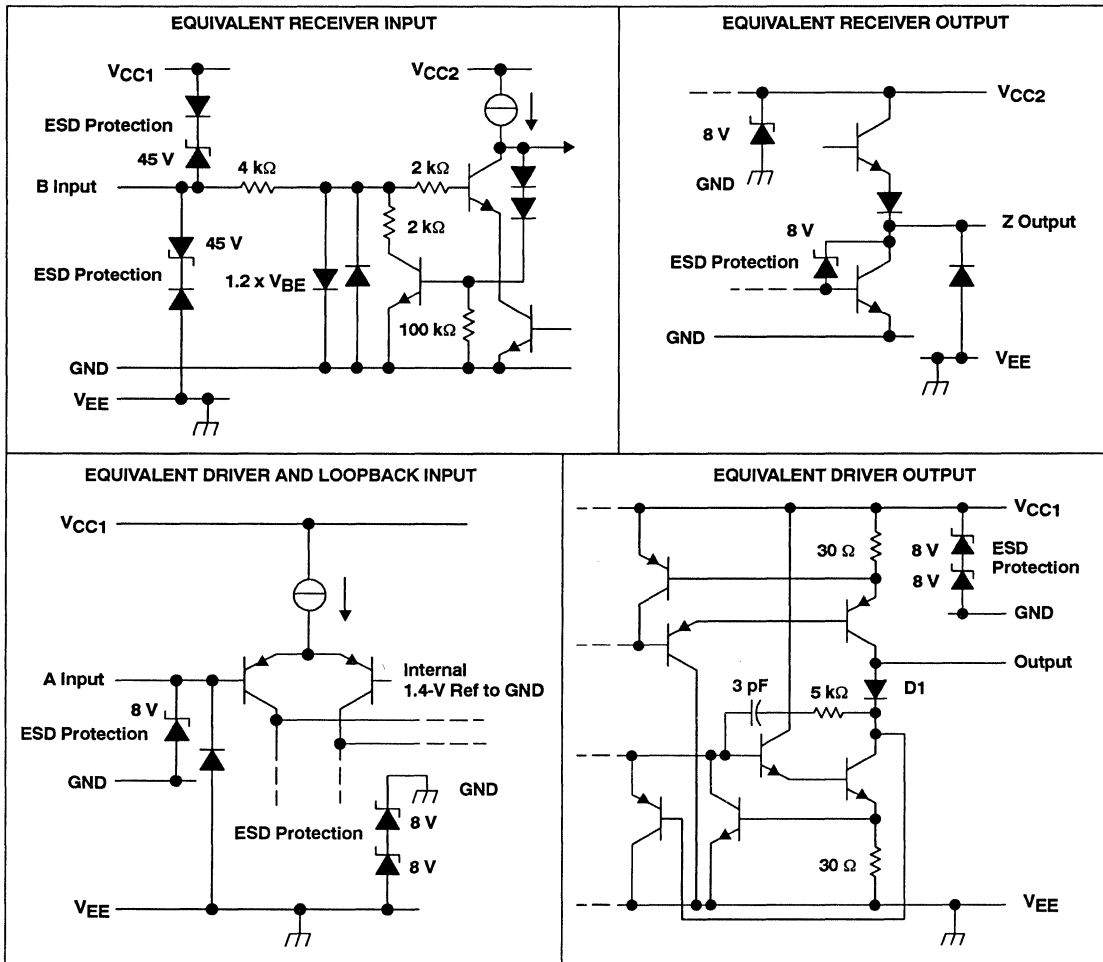
logic diagram, each driver/receiver pair (positive logic)



# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

SLLS068B – D3389, FEBRUARY 1990 – REVISED MARCH 1993

## schematics of inputs and outputs



All component values shown are nominal.



# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

SLLS068B – D3389, FEBRUARY 1990 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC1}$ (see Note 1)	15 V
Supply voltage, $V_{CC2}$	7 V
Supply voltage, $V_{EE}$	-15 V
Receiver input voltage range	-30 V to 30 V
Driver input voltage range	( $V_{EE} + 2 V$ ) to $V_{CC1}$
Loopback input voltage range	0 V to 7 V
Driver output voltage range	-30 V to 30 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 10 seconds: FN package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: DW package	260°C

NOTE 1: All voltages are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1350 mW	10.8 mW/°C	864 mW
FN	1400 mW	11.2 mW/°C	896 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC1}$	10.8	12	13.2	V
Supply voltage, $V_{CC2}$	4.5	5	5.5	V
Supply voltage, $V_{EE}$	-10.8	-12	-13.2	V
Input voltage, $V_I$	Driver and loopback		0	$V_{CC2}$
Input voltage, $V_I$ (see Note 2)	Receiver		$\pm 30$	V
High-level input voltage, $V_{IH}$	Driver and loopback		2	V
Low-level input voltage, $V_{IL}$	Driver and loopback		0.8	V
Output voltage powered on or off, $V_O$	Driver		$\pm 30$	V
High-level output current, $I_{OH}$	Receiver		-4	mA
Low-level output current, $I_{OL}$	Receiver		4	mA
Operating free-air temperature, $T_A$	0		70	°C

NOTE 2: If all receiver inputs are held at  $\pm 30$  V, the thermal dissipation limit of the package may be exceeded. The thermal shutdown may not protect the device, as this dissipation occurs in the receiver input resistors.



# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

SLLS068B – D3389, FEBRUARY 1990 – REVISED MARCH 1993

## DRIVER SECTION

**electrical characteristics over full recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	R <sub>L</sub> = 3 kΩ, V <sub>IL</sub> = 0.8 V, See Figure 1		7		V
V <sub>OL</sub>	Low-level output voltage‡	R <sub>L</sub> = 3 kΩ, V <sub>IH</sub> = 2 V, See Figure 1			-7	V
V <sub>OH(LB)</sub>	High-level output voltage in loopback mode‡§¶	R <sub>L</sub> = 3 kΩ, $\overline{\text{LB}}$ at 0.8 V, V <sub>IL</sub> = 0.8 V			-7	V
I <sub>IH</sub>	High-level input current (driver and loopback inputs)#	V <sub>I</sub> = 5 V, See Figure 2			100	μA
I <sub>IL</sub>	Low-level input current (driver and loopback inputs)#				-100	μA
V <sub>OS(H)</sub>	High-level short-circuit output current	V <sub>I</sub> = 0.8 V, V <sub>O</sub> = 0, See Note 3 and Figure 1	-10	-20	-35	mA
V <sub>OS(L)</sub>	Low-level short-circuit output current	V <sub>I</sub> = 2 V, V <sub>O</sub> = 0, See Note 3 and Figure 1	10	20	35	mA
I <sub>CC1</sub>	Supply current from V <sub>CC1</sub>	No load		2.5	4	mA
I <sub>CC1(LB)</sub>	Supply current from V <sub>CC1</sub> with loopback on	No load, $\overline{\text{LB}}$ at 0.8 V			10	mA
I <sub>EE</sub>	Supply current from V <sub>EE</sub>	No load		-2.5	-4	mA
I <sub>EE(LB)</sub>	Supply current from V <sub>EE</sub> with loopback on	No load, $\overline{\text{LB}}$ at 0.8 V			-10	mA
I <sub>CC2</sub>	Supply current from V <sub>CC2</sub>	No load, V <sub>I</sub> = 0, See Note 5		-10	-100	μA
I <sub>CC2(LB)</sub>	Supply current from V <sub>CC2</sub> with loopback on	No load, $\overline{\text{LB}}$ at 0.8 V, V <sub>I</sub> = 0, See Note 5		-10	-100	μA
		V <sub>CC1</sub> = V <sub>EE</sub> = V <sub>CC2</sub> = 0, V <sub>O</sub> = -2 V to 2 V, See Note 4	0.3	5		kΩ

† All typical values are at T<sub>A</sub> = 25°C.

‡ The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only.

§ This is the most positive level that the driver output will rise to when the device is in the loopback mode and the driver input is at a low level.

¶ The loopback mode should be entered only when the driver output is in the low (marking) state.

# Unused driver inputs should be tied to 0 V or V<sub>CC2</sub>; unused loopback inputs should be tied to V<sub>CC2</sub>.

NOTES: 3. Minimum I<sub>OS(H)</sub> and I<sub>OS(L)</sub> are specified at V<sub>O</sub> = 0, as this more accurately describes the output current needed to dynamically drive capacitive lines. A minimum of ±10 mA is sufficient to drive 2500 pF in parallel with 3 kΩ at a slew rate of 4 V/μs (in accordance with EIA-232-D and V.2B).

4. Test conditions are those specified by EIA-232-D.

5. Without a load and V<sub>I</sub> = 0, the worst-case conditions, V<sub>CC2</sub> sources a small current originating from V<sub>CC1</sub> giving I<sub>CC2</sub> supply current a negative sign. When a receiver has an output load, V<sub>CC2</sub> sinks static and dynamic supply currents to meet load requirements.



# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

SLLS068B – D3389, FEBRUARY 1990 – REVISED MARCH 1993

**switching characteristics over full recommended ranges of supply voltages and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high level output	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$ , See Figure 3		0.6	5	$\mu\text{s}$
$t_{PHL}$	Propagation delay time, high-to-low-level output			0.8	5	$\mu\text{s}$
$t_{skew}$	$ t_{PLH} - t_{PHL} $	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$ to $2500\text{ pF}$		0.2	1	$\mu\text{s}$
SR	Output slew rate	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$ to $2500\text{ pF}$		4	30	$\text{V}/\mu\text{s}$
$t_{pd}(\text{ILB})$	Propagation delay time going into loopback mode‡	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Note 6 and Figure 7		3	50	$\mu\text{s}$
$t_{pd}(\text{OLB})$	Propagation delay time going out of loopback mode§	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Note 6 and Figure 7		3	50	$\mu\text{s}$
$t_{pd}(\text{LB})$	Propagation delay time in loopback mode¶	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Note 6 and Figure 8		3	15	$\mu\text{s}$
$t_{skew}$	Skew time in loopback mode	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Note 6		4	10	$\mu\text{s}$

† All typical values are at  $T_A = 25^\circ\text{C}$ .

‡ This is the delay between entering the loopback mode and when the data on the receiver output becomes valid.

§ This is the worst-case (rising or falling edges) total propagation delay between driver input and receiver output when in the loopback mode.

¶ This is the magnitude of the difference between the propagation delay time of the rising and falling edges of  $t_{pd}(\text{LB})$ .

NOTE 6: Skew time is the magnitude of the difference between  $t_{PHL}$  and  $t_{PLH}$  and is measured with a 0 to 3-V input pulse.



**RECEIVER SECTION**

**electrical characteristics over full recommended ranges of supply voltages and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going threshold voltage	See Figure 5	1.3	2	2.5	V
$V_{T-}$ Negative-going threshold voltage	See Figure 5	0.5	1	1.7	V
$V_{hys}$ Input hysteresis ( $V_{T+} - V_{T-}$ )		0.5	1	1.5	V
$V_{OH}$ High-level output voltage	$V_I = -3$ V or inputs open, $I_{OH} = -20$ $\mu$ A	3.5			V
	$I_{OH} = -4$ mA, See Note 7 and Figure 5	2.4			
$V_{OL}$ Low-level output voltage	$I_{OL} = 4$ mA, $V_I = 3$ V, See Figure 5	0.4			V
$I_{OS(H)}$ Short-circuit output current at high level	$V_{OH} = 0$ , See Figure 4	-20	-60	mA	
$I_{OS(L)}$ Short-circuit output current at low level	$V_{OL} = V_{CC2}$ , See Figure 4	20	60	mA	
$r_{in}$ Input resistance	$ V_I  \leq 25$ V	3			k $\Omega$
	$ V_I  = 3$ V to 25 V	7			

NOTE 7: If the inputs are left unconnected, the receiver interprets this as a low input and the receiver outputs will remain in the high state.

**switching characteristics over full recommended ranges of supply voltages and operating free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$t_{PLH}$ Propagation delay time, low-to-high level output	See Figure 6	2			6	$\mu$ s
$t_{PHL}$ Propagation delay time, high-to-low-level output		2			6	$\mu$ s
$t_{TLH}$ Transition time, low-to-high level output‡	$C_L = 50$ pF, See Figure 6	200	300	ns		
$t_{THL}$ Transition time, high-to-low level output‡		50	300	ns		
$t_{skew}$   $t_{PLH} - t_{PHL}$		0.1	1	$\mu$ s		
$t_w$ Maximum pulse duration assumed to be noise§	Pulse amplitude = 5 V	1	2	4	$\mu$ s	

† All typical values are at  $T_A = 25^\circ\text{C}$ .

‡ Transition times are measured between 10% and 90% points on output waveform.

§ The receiver will ignore any positive- or negative-going pulse whose duration is less than the minimum value of  $t_w$  and accept any positive- or negative-going pulse whose duration is greater than the maximum value of  $t_w$ .

# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

SLLS068B - D3389, FEBRUARY 1990 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

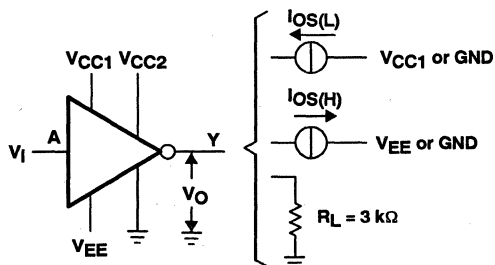


Figure 1. Driver Test Circuit,  $V_{OH}$ ,  $V_{OL}$ ,  $I_{OS(L)}$ ,  $I_{OS(H)}$

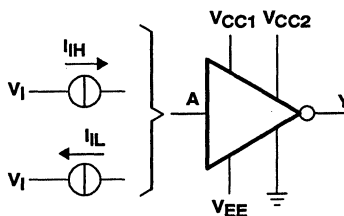


Figure 2. Driver and Loopback Test Circuit,  $I_{iL}$ ,  $I_{iH}$

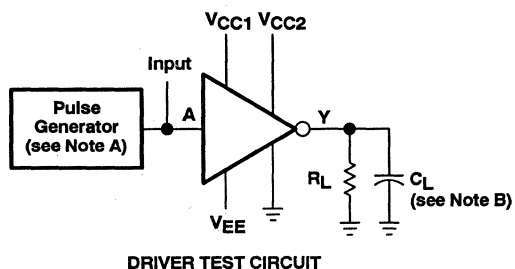


Figure 3. Driver Test Circuit and Voltage Waveforms

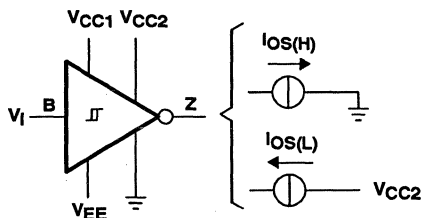
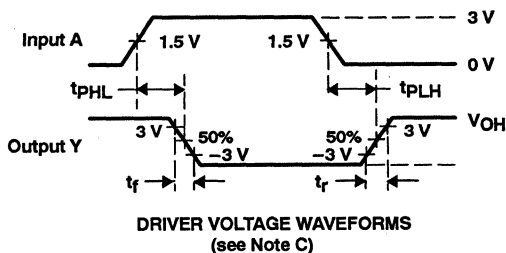


Figure 4. Receiver Test Circuit,  $I_{OS(H)}$ ,  $I_{OS(L)}$

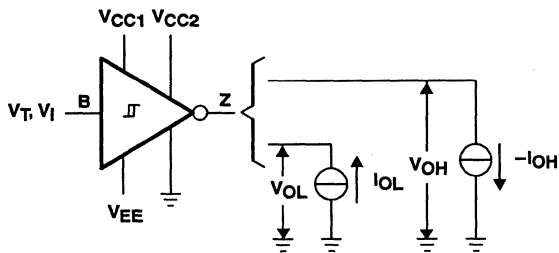


Figure 5. Receiver Test Circuit,  $V_T$ ,  $V_{OL}$ ,  $V_{OH}$

NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu s$ , PRR = 20 kHz,  $Z_O = 50 \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

C. Slew rate =  $\frac{6 V}{t_r \text{ or } t_f}$

PARAMETER MEASUREMENT INFORMATION

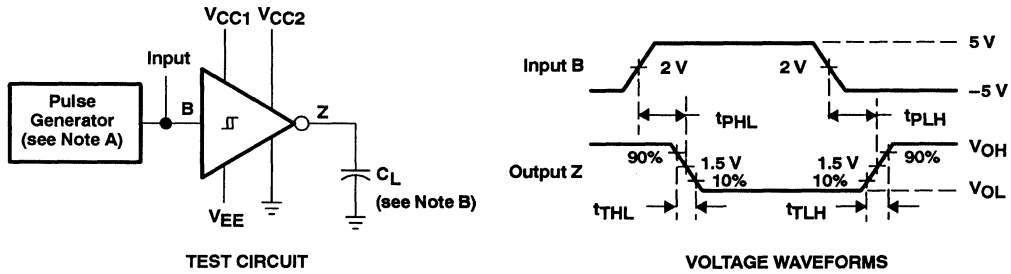


Figure 6. Receiver Propagation and Transition Times

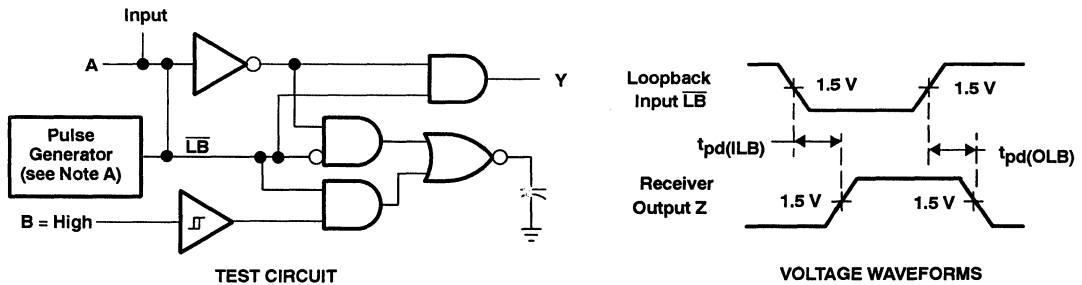


Figure 7. Loopback Entry and Exit Propagation Times

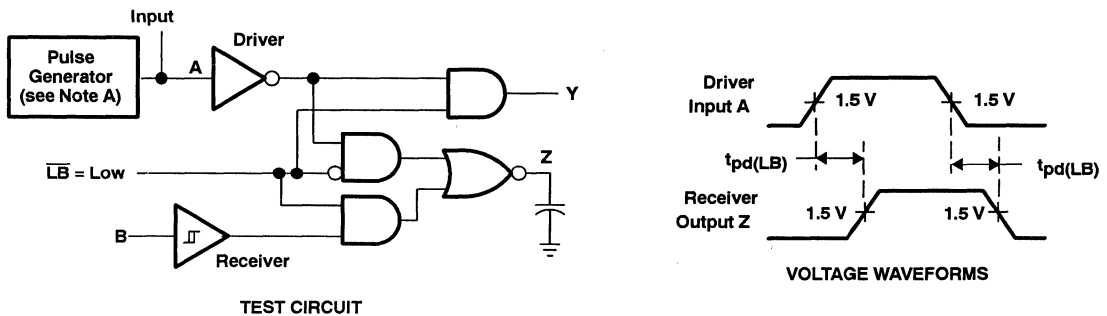


Figure 8. Loop Propagation Times in Loopback Mode

NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu s$ , PRR = 20 kHz,  $Z_O = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

# SN75186 QUAD DRIVER/RECEIVER WITH LOOPBACK

SLLS068B – D3389, FEBRUARY 1990 – REVISED MARCH 1993

## PRINCIPLES OF OPERATION

In normal operation, the SN75186 acts as four independent drivers and receivers; the loopback mode is held off by keeping logic inputs  $\overline{LB}$  high. Taking a particular  $\overline{LB}$  input low activates the loopback mode in the corresponding driver/receiver pair. This causes the output from that driver to be fed back to the input of its receiver through dedicated internal loopback circuitry. Data from the receiver output can then be compared, by a communication system, with the data transmitted to the driver to determine if the functional operation of the driver and receiver together is correct.

In the loopback mode, external data at the input of the receiver is ignored and the driver does not transmit data onto the line. Extraneous data is prevented internally from being sent by the driver in the loopback mode by clamping its output to a level below the maximum interface voltage,  $-5\text{ V}$ , or the EIA-232-D marking state. Below this marking level, a reduced 1.5-V output amplitude is used at the driver output. This signal is detected by an on-chip loopback comparator and fed to the input stage of the receiver to complete the loop.

Line faults external to the SN75186 are detected in addition to device failures. These line faults include short circuits to ground and to external supply voltages that are greater than  $(V_{EE} + 7\text{ V})$  and less than  $V_{EE}$  typically. For example, with  $V_{EE} = -12\text{ V}$ , line short circuits to voltages greater than  $-5\text{ V}$  and less than  $-12\text{ V}$  will be detected. The loopback mode should be entered only when the driver output is low, that is, the marking condition of EIA-232-D. It is recommended that loopback not be entered when the driver output is in a high state as this may cause a low-level, nondamaging oscillation at the driver output.

When in the loopback mode, approximately 95% of the SN75186 circuit is functionally checked. There exists some low probability of fault mechanisms in circuitry not checked in the loopback mode. To reduce the chances of undetected failure, the unchecked circuitry has been designed to be more robust than that within the loopback test loop. The areas where special attention has been paid are the receiver input potential divider and resistors, the driver output blocking diode (D1), and parts of the driver clamp circuit.

Protection of the SN75186 is achieved by means of driver output current limits and a thermal trip. Although this device will withstand  $\pm 30\text{ V}$  at its receiver input, package thermal dissipation limitations have to be taken into consideration if more than one receiver is connected simultaneously. This is due to the possible dissipation in the 3-k $\Omega$  minimum input resistors, which is not under the control of the thermal trip. Although the supply current is higher in the loopback mode than in normal operation, the total power dissipation is not sufficient under normal worst-case conditions (of receiver input  $V_I = 15\text{ V} + 10\%$ , receiver output voltage = 2.4 V at 4 mA, driver load of 3 k $\Omega$ ) to cause the thermal limiting circuitry to trip.

If the SN75186 goes into thermal trip, the output of the driver goes to a high-impedance state and the receiver output is held in a logic-high marking state. Both driver and receiver outputs maintain a marking state for the following circuit and do not allow indeterminate conditions to exist.

The standards specify a minimum driver output resistance to ground of 300  $\Omega$  when the device is powered off. To fully comply with EIA-232-D power-off fault conditions, many drivers need diodes in series with each supply voltage to prevent reverse current flow and driver damage. The SN75186 overcomes this need by providing a high-impedance driver output of typically 5 k $\Omega$  under power-off conditions through the use of the equivalent of these series diodes in the driver output circuit.

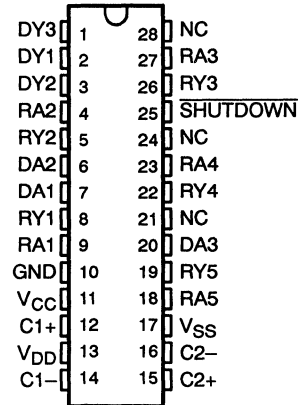


# SN75LBC187 MULTICHANNEL EIA-232 DRIVER/RECEIVER WITH CHARGE PUMP

SLLS130B – D3881, SEPTEMBER 1991 – REVISED JANUARY 1993

- **Single IC and Single 5-V Supply Interface for Serial Communication Ports**
- **Three EIA/TIA-E Drivers and Five Receivers Meet EIA/TIA-232-E-1991, EIA/TIA-562, and CCITT Recommendation V.28 Standards**
- **Switched-Capacitor Voltage Converter Eliminates Need for  $\pm 12$ -V Supplies**
- **Voltage Converter Operates With Low Capacitance . . . 0.1  $\mu$ F Min**
- **ESD Protection on RS-232 Lines Exceeds 6 kV Per MIL-STD-883C, Method 3015**
- **Designed for Data Rates up to 120 kb/s Over 3-m Cable**
- **Available in Shrink Small-Outline 25-mil-Pitch Package**
- **Shutdown Mode to Save Power When Not in Use**
- **$\pm 30$ -V Receiver Input Voltage Range**
- **LinBiCMOS™ Process Technology**
- **Applications**
  - Laptop or Notebook Computers**
  - Portable Terminals**
  - Single-Board Computers**
  - Portable Test Equipment**

DB PACKAGE  
(TOP VIEW)



NC—No internal connection

## description

The SN75LBC187 is a low-power LinBiCMOS™ device containing three independent drivers, five receivers, and a switched-capacitor voltage converter. The SN75LBC187 provides a single integrated circuit (IC) and single 5-V supply interface between the asynchronous communications element and the serial port connector of the data terminal equipment (DTE). This device has been designed to conform to standards EIA/TIA-232-E-1991, EIA/TIA-562, and CCITT recommendation V.28.

The switched-capacitor voltage converter of the SN75LBC187 uses four small external capacitors to generate the positive and negative voltages required by EIA-232 (and V.28) line drivers from a single 5-V input. The drivers feature output slew-rate limiting to eliminate the need for external filter capacitors. The receivers can accept  $\pm 30$  V without damage. The device also features a reduced power or shutdown mode that cuts the quiescent power to the IC when not transmitting data between the CPU and peripheral.

The SN75LBC187 has been designed using LinBiCMOS™ technology and cells contained in the Texas Instruments LinASIC™ library. The SN75LBC187 is characterized for operation from 0°C to 70°C.

NOTE: This device includes circuit designs and process technologies that have patents pending.

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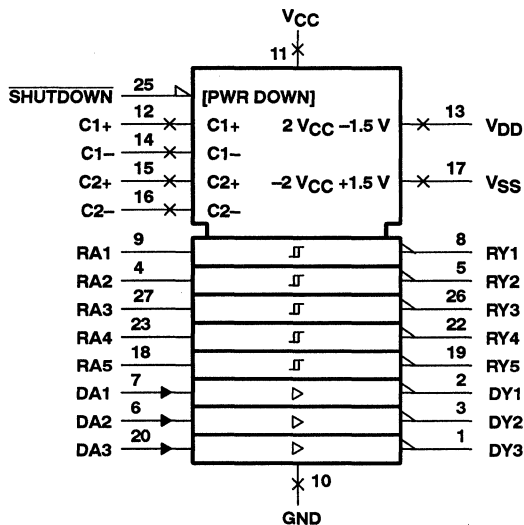
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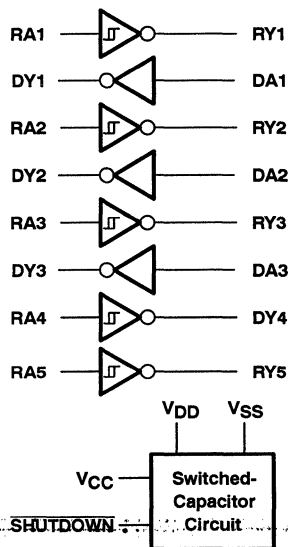
# SN75LBC187 MULTICHANNEL EIA-232 DRIVER/RECEIVER WITH CHARGE PUMP

SLLS130B - D3881, SEPTEMBER 1991 - REVISED JANUARY 1993

## logic symbol†



## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std. 91-1984 and IEC Publication 617-12.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, V <sub>CC</sub> (see Note 1)	0.3 V to 6 V
Positive output supply voltage range, V <sub>DD</sub>	V <sub>CC</sub> - 0.3 V to 15 V
Negative output supply voltage range, V <sub>SS</sub>	0.3 V to -15 V
Input voltage range: RA	±30 V
All other inputs	-0.3 V to V <sub>CC</sub> + 3 V
Output voltage range: DY	-2 V <sub>CC</sub> + 1.2 V to 2 V <sub>CC</sub> - 1.2 V
All other outputs	-0.3 V to V <sub>CC</sub> + 3 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T <sub>A</sub>	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	T <sub>A</sub> = 25°C POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
DB	1025 mW	8.2 mW/°C	656 mW

**SN75LBC187**  
**MULTICHANNEL EIA-232 DRIVER/RECEIVER**  
**WITH CHARGE PUMP**

SLLS130B – D3881, SEPTEMBER 1991 – REVISED JANUARY 1993

**recommended operating conditions**

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.5	5	5.5	V
High-level input voltage, $V_{IH}$	DA	2			V
	RA, SHUTDOWN	2.4			
Low-level input voltage, $V_{IL}$	RA, DA, SHUTDOWN	0.8			V
Receiver input voltage, $V_I$		-25	25		V
High-level output current, $I_{OH}$	RY	-1			mA
Low-level output current, $I_{OL}$	RY	3.2			mA
Output current, $I_O$	$V_{DD}$	$\pm 10$			$\mu A$
	$V_{SS}$	$\pm 10$			$\mu A$
C1, C2, C3, C4 charge pump capacitors		0.1	0.47	$\mu F$	
Operating free-air temperature, $T_A$		0			70 °C

**electrical characteristics over recommended operating conditions (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{OH}$ High-level output voltage	Receiver	$I_O = -1$ mA		3.5			V
	Driver	$R_L = 3$ k $\Omega$ to GND		5	7		
$V_{OL}$ Low-level output voltage	Receiver	$I_O = 3.2$ mA		0.4			V
	Driver	$R_L = 3$ k $\Omega$ to GND		-7	-5		
$V_{T+}$ Receiver positive-going input voltage threshold			1.7		2.4		V
$V_{T-}$ Receiver negative-going input voltage threshold			0.8	1.2		V	
$V_{hys}$ Receiver input hysteresis ( $V_{T+} - V_{T-}$ )			0.5		1		V
$r_i$ Receiver input resistance	$V_{CC} = 5$ V, $T_A = 25^\circ C$		3	5	7		k $\Omega$
$r_o$ Driver output resistance	$V_{CC} = 0$ , $V_O = \pm 2$ V		300			$\Omega$	
$I_I$ Input current (DA, SHUTDOWN)	$V_I = 0$ to $V_{CC}$		$\pm 50$			$\mu A$	
$I_{OS}$ Driver output short-circuit current	$V_O = 0$		$\pm 10$			mA	
$I_{CC}$ Supply current	All outputs open, SHUTDOWN at 2.4 V		15			30	mA
	All outputs open, SHUTDOWN at 0.1 V		10			$\mu A$	

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ C$ .



**SN75LBC187**  
**MULTICHANNEL EIA-232 DRIVER/RECEIVER**  
**WITH CHARGE PUMP**

SLLS130B – D3881, SEPTEMBER 1991 – REVISED JANUARY 1993

**switching characteristics over recommended operating conditions,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	Receiver	R <sub>L</sub> = 5 k $\Omega$ , See Figure 1	C <sub>L</sub> = 50 pF,	1.25	$\mu\text{s}$
		Driver	R <sub>L</sub> = 3 k $\Omega$ , See Figure 2	C <sub>L</sub> = 1200 pF,	1.25	$\mu\text{s}$
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output	Receiver	R <sub>L</sub> = 5 k $\Omega$ , See Figure 1	C <sub>L</sub> = 50 pF,	1.25	$\mu\text{s}$
		Driver	R <sub>L</sub> = 3 k $\Omega$ , See Figure 2	C <sub>L</sub> = 1200 pF,	1.25	$\mu\text{s}$
t <sub>r</sub>	Rise time, driver output		R <sub>L</sub> = 3 k $\Omega$ , V <sub>O</sub> = -3 V to 3 V,	C <sub>L</sub> = 50 pF, See Note 2	200	ns
			R <sub>L</sub> = 3 k $\Omega$ , V <sub>O</sub> = -3.3 V to 3.3 V,	C <sub>L</sub> = 2500 pF, See Note 3	1.5	$\mu\text{s}$
t <sub>f</sub>	Fall time, driver output		R <sub>L</sub> = 3 k $\Omega$ , V <sub>O</sub> = 3 V to -3 V	C <sub>L</sub> = 50 pF,	200	ns
			R <sub>L</sub> = 3 k $\Omega$ , V <sub>O</sub> = 3.3 V to -3.3 V	C <sub>L</sub> = 2500 pF,	1.5	$\mu\text{s}$

NOTES: 2. The 200 ns for the output to change from -3 V to 3 V (or vice versa) corresponds to the 30 V/ $\mu\text{s}$  maximum slew rate of EIA/TIA-232-E, EIA/TIA-562, and CCITT V.28.

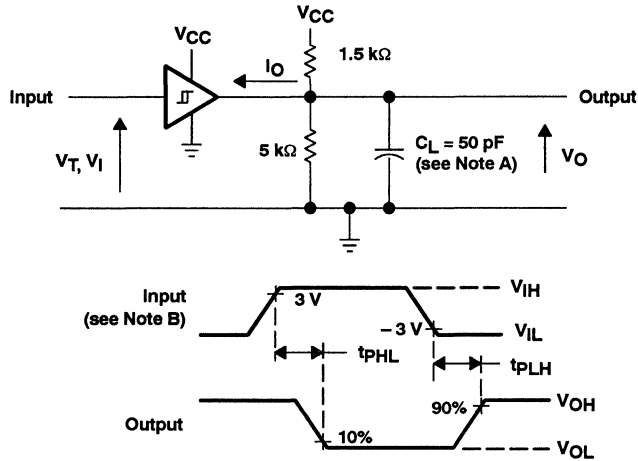
3. The more stringent requirement for transition times comes from the EIA/TIA-562, which requires the rise and fall times to be measured from 3.3 V.



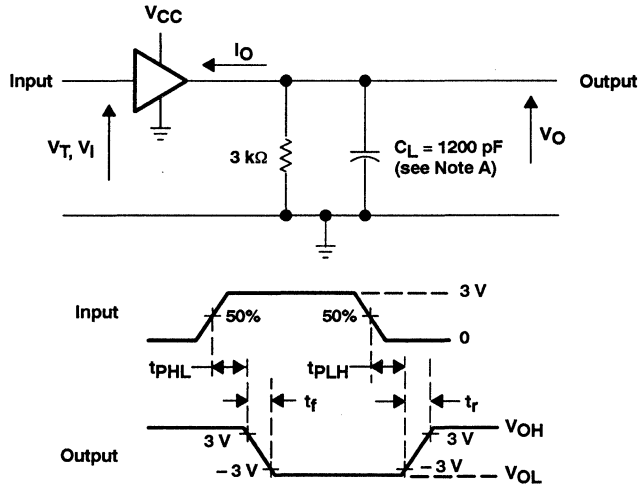
**SN75LBC187**  
**MULTICHANNEL EIA-232 DRIVER/RECEIVER**  
**WITH CHARGE PUMP**

SLLS130B – D3881, SEPTEMBER 1991 – REVISED JANUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



**Figure 1. Receiver Test Circuit and Waveforms**



**Figure 2. Driver Test Circuit and Waveforms**

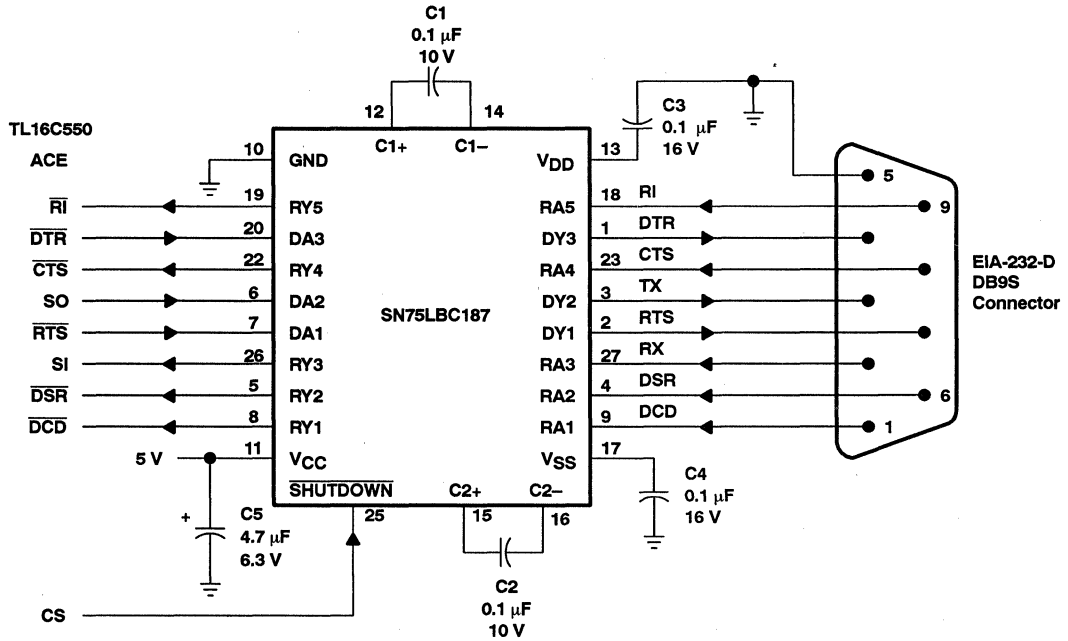
NOTES: A.  $C_L$  includes probe and jig capacitance.

B. The pulse generator has the following characteristics:  $t_w = 8.33 \mu s$ , PRR = 60 kHz,  $t_r = t_f \leq 50$  ns.

**SN75LBC187**  
**MULTICHANNEL EIA-232 DRIVER/RECEIVER**  
**WITH CHARGE PUMP**

SLLS130B - D3881, SEPTEMBER 1991 - REVISED JANUARY 1993

**APPLICATION INFORMATION**



NOTE: C1, C2, C3, and C4 are Z5U-type ceramic-chip capacitors.

**Figure 3. Typical SN75LBC187 Connection**

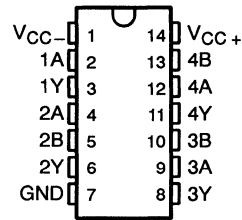


# MC1488, SN55188, SN75188 QUAD LINE DRIVERS

SLLS094A – D1323, SEPTEMBER 1983 – REVISED MARCH 1993

- Meets Specifications of EIA RS-232-C
- Designed to Be Interchangeable With Motorola MC1488
- Current-Limited Output: 10 mA Typ
- Power-Off Output Impedance: 300 Ω Min
- Slew Rate Control by Load Capacitor
- Flexible Supply Voltage Range
- Input Compatible With Most TTL Circuits

SN55188 . . . J OR W PACKAGE  
MC1488, SN75188 . . . D OR N PACKAGE  
(TOP VIEW)



## description

The MC1488, SN55188, and SN75188 are monolithic quad line drivers designed to interface data terminal equipment with data communications equipment in conformance with EIA Standard RS-232-C using a diode in series with each supply-voltage terminal as shown under typical applications.

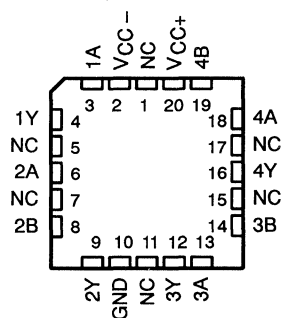
The and SN55188 is characterized for operation over the full military temperature range of -55°C to 125°C. The MC1488 and SN75188 are characterized for operation from 0°C to 70°C.

FUNCTION TABLE  
(drivers 2 through 4)

A	B	Y
H	H	L
L	X	H
X	L	H

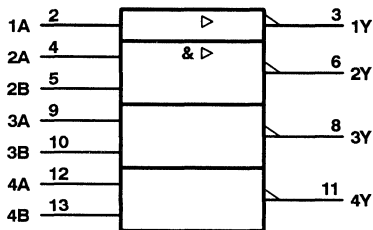
H = high level, L = low level,  
X = irrelevant

SN55188 . . . FK PACKAGE  
(TOP VIEW)



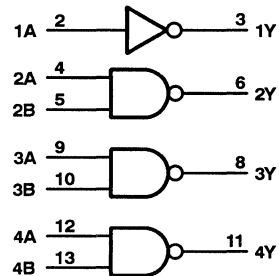
NC – No internal connection

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



Positive logic  
Y =  $\bar{A}$  (driver 1)  
Y = AB or  $\bar{A} + \bar{B}$  (drivers 2 thru 4)

Pin numbers shown are for the D and N packages.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

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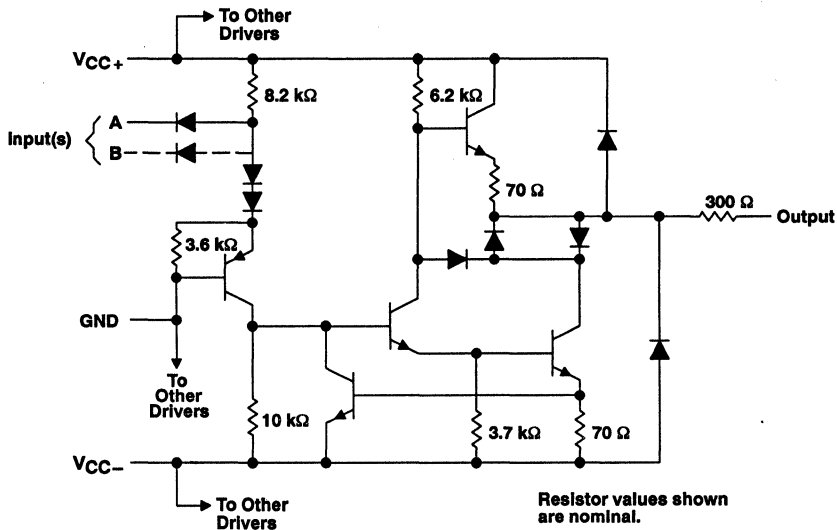
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2-737

# MC1488, SN55188, SN75188 QUAD LINE DRIVERS

SLLS094A - D1323, SEPTEMBER 1983 - REVISED MARCH 1993

## schematic (each driver)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55188	MC1488 SN75188	UNIT
Supply voltage, $V_{CC+}$ , at (or below) 25°C free-air temperature (see Notes 1 and 2)	15	15	V
Supply voltage, $V_{CC-}$ , at (or below) 25°C free-air temperature (see Notes 1 and 2)	-15	-15	V
Input voltage range	-15 to 7	-15 to 7	V
Output voltage range	-15 to 15	-15 to 15	V
Continuous total power dissipation (see Note 2)	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package	260	°C
Case temperature for 60 seconds	FK package	260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or W package	300	

NOTES: 1. All voltage values are with respect to the network ground terminal.

2. For operation above 25°C free-air temperature, refer to the maximum supply voltage curve, Figure 6. In the FK and J packages, SN55188 chips are alloy mounted.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	-
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	-
W	1000 mW	8.0 mW/°C	640 mW	200 mW

# MC1488, SN55188, SN75188 QUAD LINE DRIVERS

SLLS094A – D1323, SEPTEMBER 1983 – REVISED MARCH 1993

## recommended operating conditions

	SN55188			MC1488, SN75188			UNIT		
	MIN	NOM	MAX	MIN	NOM	MAX			
Supply voltage, $V_{CC+}$	7.5	9	15	7.5	9	15	V		
Supply voltage, $V_{CC-}$	-7.5	-9	-15	-7.5	-9	-15	V		
High-level input voltage, $V_{IH}$	1.9			1.9			V		
Low-level input voltage, $V_{IL}$				0.8			V		
Operating free-air temperature, $T_A$	-55			125			0	70	°C

## electrical characteristics over operating free-air temperature range, $V_{CC\pm} = \pm 9\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS		SN55188			MC1488, SN75188			UNIT		
			MIN	TYP†	MAX	MIN	TYP†	MAX			
$V_{OH}$ High-level output voltage	$V_{IL} = 0.8\text{ V}$ , $R_L = 3\text{ k}\Omega$	$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$	6	7		6	7		V		
		$V_{CC+} = 13.2\text{ V}$ , $V_{CC-} = -13.2\text{ V}$	9	10.5		9	10.5				
$V_{OL}$ Low-level output voltage	$V_{IH} = 1.9\text{ V}$ , $R_L = 3\text{ k}\Omega$	$V_{CC+} = 9\text{ V}$ , $V_{CC-} = -9\text{ V}$		-7‡	-6		-7	-6	V		
		$V_{CC+} = 13.2\text{ V}$ , $V_{CC-} = -13.2\text{ V}$		-10.5‡	-9		-10.5	-9			
$I_{IH}$ High-level input current	$V_I = 5\text{ V}$				10			10	$\mu\text{A}$		
$I_{IL}$ Low-level input current	$V_I = 0$				-1	-1.6		-1	-1.6	mA	
$I_{OS(H)}$ Short-circuit output current at high level§	$V_I = 0.8\text{ V}$ , $V_O = 0$				-4.6	-9	-13.5	-6	-9	-12	mA
$I_{OS(L)}$ Short-circuit output current at low level§	$V_I = 1.9\text{ V}$ , $V_O = 0$				4.6	9	13.5	6	9	12	mA
$r_o$ Output resistance, power off	$V_{CC+} = 0$ , $V_O = -2\text{ V to } 2\text{ V}$	$V_{CC-} = 0$ ,			300			300		$\Omega$	
$I_{CC+}$ Supply current from $V_{CC+}$	$V_{CC+} = 9\text{ V}$ , No load	All inputs at 1.9 V			15	20		15	20	mA	
		All inputs at 0.8 V			4.5	6		4.5	6		
		$V_{CC+} = 12\text{ V}$ , No load	All inputs at 1.9 V			19	25		19		25
			All inputs at 0.8 V			5.5	7		5.5		7
		$V_{CC+} = 15\text{ V}$ , No load, $T_A = 25^\circ\text{C}$	All inputs at 1.9 V			34			34		
			All inputs at 0.8 V			12			12		
$I_{CC-}$ Supply current from $I_{CC-}$	$V_{CC-} = -9\text{ V}$ , No load	All inputs at 1.9 V			-13	-17		-13	-17	mA	
		All inputs at 0.8 V				-0.5			-0.015		
	$V_{CC-} = -12\text{ V}$ , No load	All inputs at 1.9 V			-18	-23		-18	-23		
		All inputs at 0.8 V				-0.5			-0.015		
	$V_{CC-} = -15\text{ V}$ , No load, $T_A = 25^\circ\text{C}$	All inputs at 1.9 V			-34			-34			
		All inputs at 0.8 V			-2.5			-2.5			
$P_D$ Total power dissipation	$V_{CC+} = 9\text{ V}$ , No load	$V_{CC-} = -9\text{ V}$ ,			333			333	mW		
		$V_{CC-} = -12\text{ V}$ ,			576			576			

† All typical values are at  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for logic voltage levels only, e.g., if  $-6\text{ V}$  is a maximum, the typical value is a more negative voltage.

§ Not more than one output should be shorted at a time.



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# MC1488, SN55188, SN75188 QUAD LINE DRIVERS

SLLS094A - D1323, SEPTEMBER 1983 - REVISED MARCH 1993

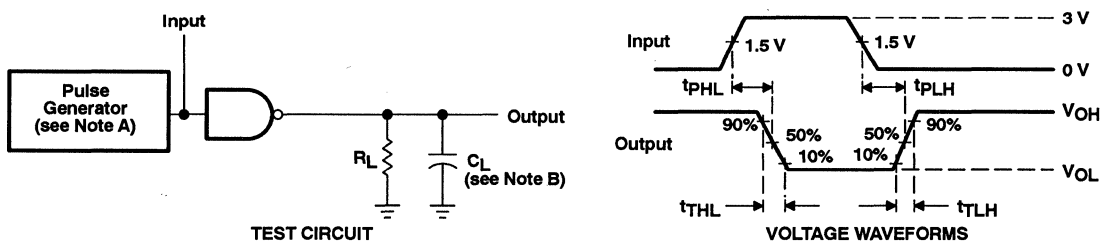
switching characteristics,  $V_{CC\pm} = \pm 9\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 3\text{ k}\Omega$ , $C_L = 15\text{ pF}$ , See Figure 1		220	350	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			100	175	ns
$t_{TLH}$ Transition time, low-to-high-level output†			55	100	ns
$t_{THL}$ Transition time, high-to-low-level output†			45	75	ns
$t_{TLH}$ Transition time, low-to-high-level output‡	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 2500\text{ pF}$ , See Figure 1		2.5		$\mu\text{s}$
$t_{THL}$ Transition time, high-to-low-level output‡			3.0		$\mu\text{s}$

† Measured between 10% and 90% points of output waveform.

‡ Measured between 3 V and -3 V points on the output waveform (EIA RS-232-C conditions).

## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics:  $t_w = 0.5\ \mu\text{s}$ ,  $\text{PRR} \leq 1\ \text{MHz}$ ,  $Z_O = 50\ \Omega$ .

B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

TEXAS  
INSTRUMENTS

TYPICAL CHARACTERISTICS†

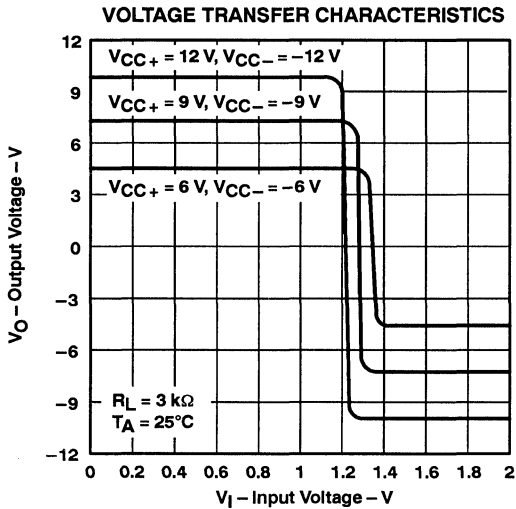


Figure 2

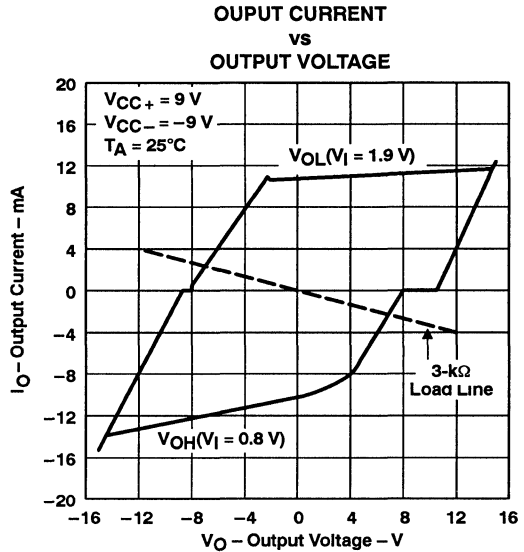


Figure 3

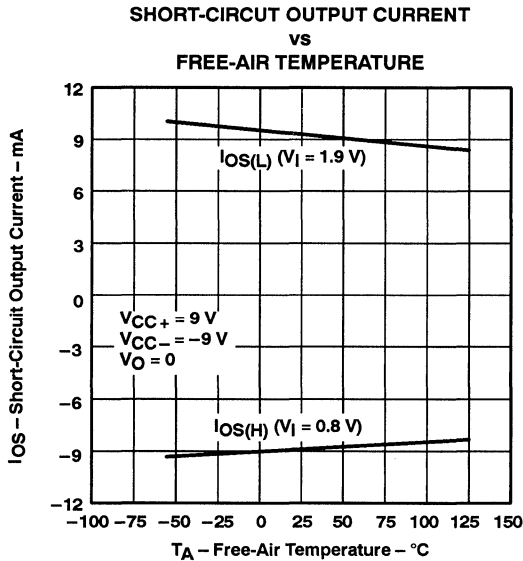


Figure 4

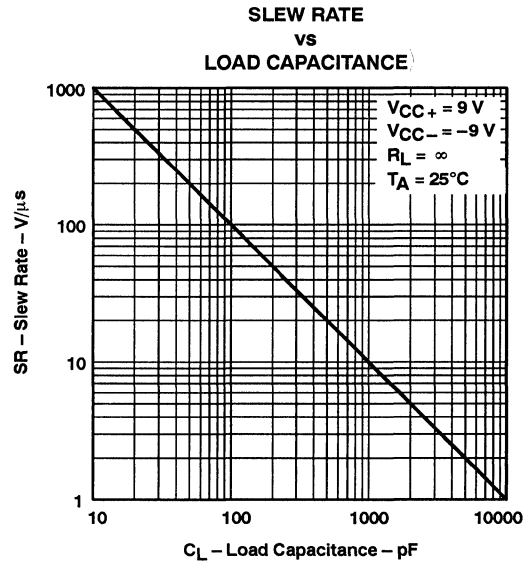


Figure 5

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  are applicable to SN55188 circuit only.

# MC1488, SN55188, SN75188 QUAD LINE DRIVERS

SLLS094A – D1323, SEPTEMBER 1983 – REVISED MARCH 1993

## THERMAL INFORMATION†

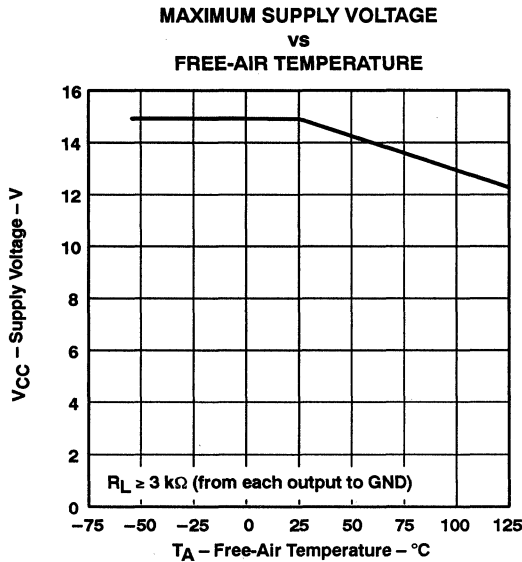


Figure 6

## APPLICATION INFORMATION

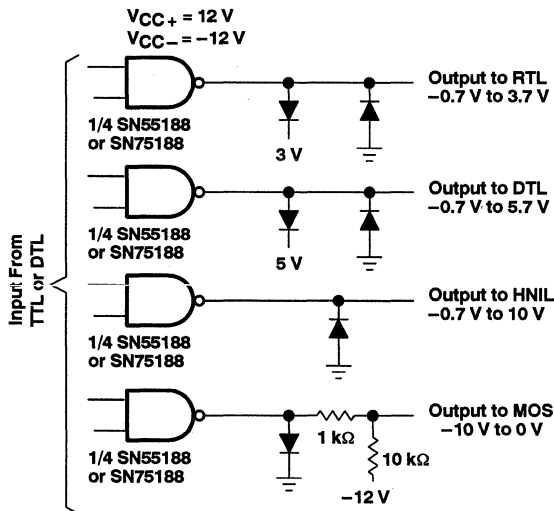


Figure 7. Logic Translator Applications

Diodes placed in series with the  $V_{CC+}$  and  $V_{CC-}$  leads will protect the SN55188/SN75188 in the fault condition in which the device outputs are shorted to  $\pm 15\text{ V}$  and the power supplies are at low and provide low-impedance paths to ground.

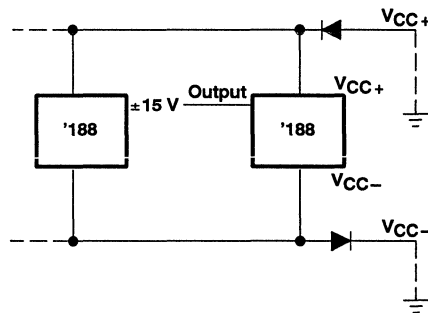


Figure 8. Power Supply Protection to Meet Power-Off Fault Conditions of EIA Standard RS-232-C

# SN65C188, SN75C188 QUAD LOW-POWER LINE DRIVERS

SLLS033D – D3075, JANUARY 1988 – REVISED MARCH 1993

- **Bi-MOS Technology With TTL and CMOS Compatibility**
- **Meets Standard EIA-232-D (Revision of RS-232-C)**
- **Very Low Quiescent Current . . . 95  $\mu$ A Typ**  
 $V_{CC\pm} = \pm 12$  V
- **Current-Limited Output . . . 10 mA Typ**
- **CMOS-and TTL-Compatible Inputs**
- **On-Chip Slew Rate Limited to 30 V/ $\mu$ s max**
- **Flexible Supply Voltage Range**
- **Characterized at  $V_{CC\pm}$  of  $\pm 4.5$  V and  $\pm 15$  V**
- **Functionally Interchangeable With Texas Instruments SN75188, Motorola MC1488, and National Semiconductor DS14C88**
- **ESD Protection Exceeds 1000 V Per MIL-Std-883C Method 3015**

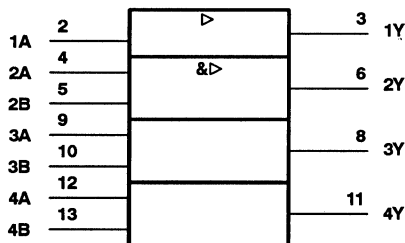
## description

The SN65C188 and SN75C188 are monolithic, low-power, quad line drivers that interface data terminal equipment with data communications equipment. These devices are designed to conform to Standard ANSI/EIA-232-D-1986, which supersedes RS-232-C.

An external diode in series with each supply-voltage terminal is needed to protect the SN65C188 and SN75C188 under certain fault conditions to comply with EIA-232-D (refer to Application Information).

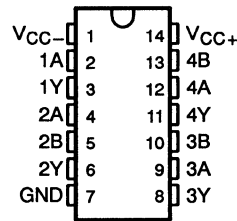
The SN65C188 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75C188 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## D, DB†, OR N PACKAGE (TOP VIEW)



† The DB package is only available left-end taped and reeled, i.e., order device SN75C188DBLE.

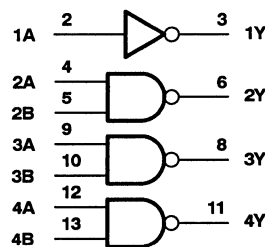
## Function Tables

DRIVER 1	
B	Y
H	L
L	H

DRIVERS 2 THRU 4		
A	B	Y
H	H	L
L	X	H
X	L	H

H = high level, L = low level, X = don't care

## logic diagram (positive logic)



positive logic  
 $Y = \bar{A}$  (driver 1)  
 $Y = AB$  or  $\bar{A} + \bar{B}$  (drivers 2 thru 4)

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
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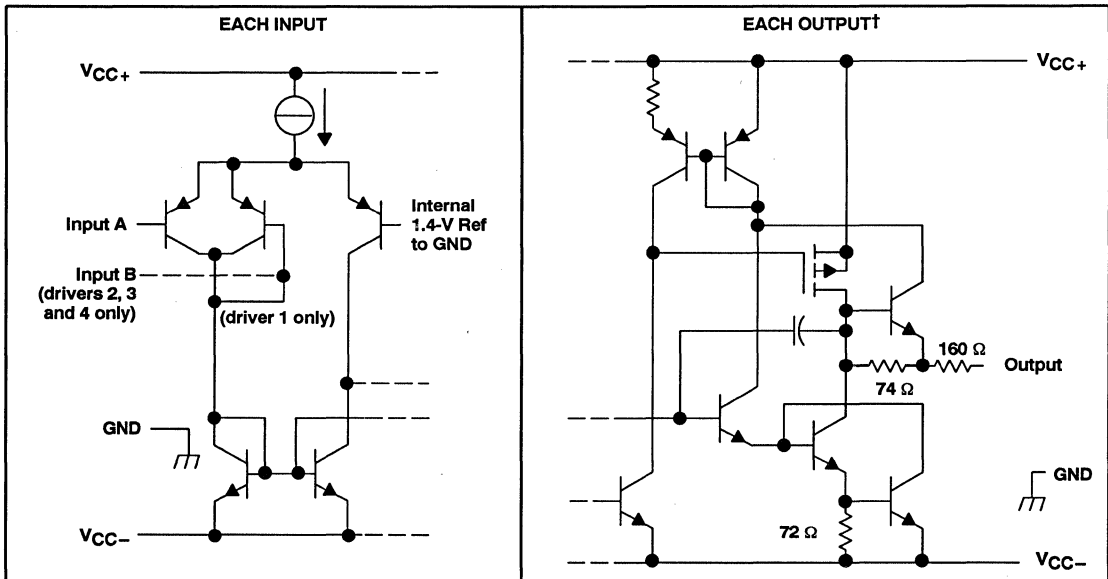
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# SN65C188, SN75C188 QUAD LOW-POWER LINE DRIVERS

SLLS033D – D3075, JANUARY 1988 – REVISED MARCH 1993

## schematics of inputs and outputs



† All resistor values shown are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	15 V
Supply voltage, $V_{CC-}$ (see Note 1)	-15 V
Input voltage range, $V_I$	$V_{CC-}$ to $V_{CC+}$
Output voltage range, $V_O$	$V_{CC-} - 6$ V to $V_{CC+} + 6$ V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : SN65C188	-40°C to 85°C
SN75C188	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 in.) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
DB	525 mW	4.2 mW/°C	336 mW	273 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

# SN65C188, SN75C188 QUAD LOW-POWER LINE DRIVERS

SLLS033D – D3075, JANUARY 1988 – REVISED MARCH 1993

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$		4.5	12	15	V
Supply voltage, $V_{CC-}$		-4.5	-12	-15	V
Input voltage, $V_I$		$V_{CC-}+2$		$V_{CC+}$	V
High-level Input voltage, $V_{IH}$		2			V
Low-level Input voltage, $V_{IL}$				0.8	V
Operating free-air temperature, $T_A$	SN65C188	-40		85	°C
	SN75C188	0		70	

## electrical characteristics over operating free-air temperature range, $V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage	$V_{IL} = 0.8\text{ V}$	$R_L = 3\text{ k}\Omega$	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$	4		V
				$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$	10		
$V_{OL}$	Low-level output voltage (see Note 2)	$V_{IH} = 2\text{ V}$	$R_L = 3\text{ k}\Omega$	$V_{CC+} = 5\text{ V}$ , $V_{CC-} = -5\text{ V}$		-4	V
				$V_{CC+} = 12\text{ V}$ , $V_{CC-} = -12\text{ V}$		-10	
$I_{IH}$	High-level input current	$V_I = 5\text{ V}$				10	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0$				-10	$\mu\text{A}$
$I_{OS(H)}$	High-level short-circuit output current‡	$V_I = 0.8\text{ V}$	$V_O = 0$ or $V_{CC-}$	-5.5	-10	-19.5	mA
$I_{OS(L)}$	Low-level short-circuit output current‡	$V_I = 2\text{ V}$	$V_O = 0$ or $V_{CC+}$	5.5	10	19.5	mA
$I_{OL}$	Output resistance, power off	$V_{CC+} = 0\text{ V}$ , $V_{CC-} = 0$ ,	$V_I = -2\text{ V}$ to $2\text{ V}$	300			$\Omega$
$I_{CC+}$	Supply current from $V_{CC+}$	$V_{CC+} = 5\text{ V}$ , No load	$V_{CC-} = -5\text{ V}$ , All inputs at $2\text{ V}$ or $0.8\text{ V}$		90	160	$\mu\text{A}$
		$V_{CC+} = 12\text{ V}$ , No load	$V_{CC-} = -12\text{ V}$ , All inputs at $2\text{ V}$ or $0.8\text{ V}$		95	160	
$I_{CC-}$	Supply current from $V_{CC-}$	$V_{CC+} = 5\text{ V}$ , No load	$V_{CC-} = -5\text{ V}$ , All inputs at $2\text{ V}$ or $0.8\text{ V}$		-90	-160	$\mu\text{A}$
		$V_{CC+} = 12\text{ V}$ , No load	$V_{CC-} = -12\text{ V}$ , All inputs at $2\text{ V}$ or $0.8\text{ V}$		-95	-160	

† All typical values are at  $T_A = 25^\circ\text{C}$ .

‡ Not more than one output should be shorted at one time.

NOTE 2: The algebraic convention, in which the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only, e.g., if a  $-4\text{ V}$  is a maximum, the typical value is a more negative voltage.



# SN65C188, SN75C188 QUAD LOW-POWER LINE DRIVERS

SLLS033D - D3075, JANUARY 1988 - REVISED MARCH 1993

switching characteristics,  $V_{CC+} = 12\text{ V}$ ,  $V_{CC-} = -12\text{ V}$ ,  $T_A = 25^\circ\text{C}$

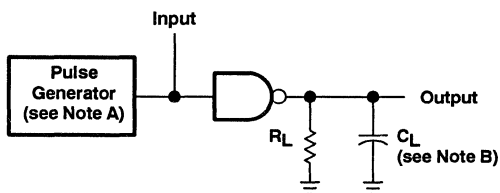
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output†	$R_L = 3\text{ k}\Omega$ , $C_L = 15\text{ pF}$ , See Figure 1			3	$\mu\text{s}$
$t_{PHL}$ Propagation delay time, high-to-low-level output†				3.5	$\mu\text{s}$
$t_{TLH}$ Transition time, low-to-high-level output‡		0.53		3.2	$\mu\text{s}$
$t_{THL}$ Transition time, high-to-low-level output‡		0.53		3.2	$\mu\text{s}$
$t_{TLH}$ Transition time, low-to-high-level output§	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 2500\text{ pF}$ , See Figure 1		1.5	3	$\mu\text{s}$
$t_{THL}$ Transition time, high-to-low-level output§			1.5	3	$\mu\text{s}$
SR Output slew rate§	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$	6	15	30	$\text{V}/\mu\text{s}$

† Measured at the 50% level

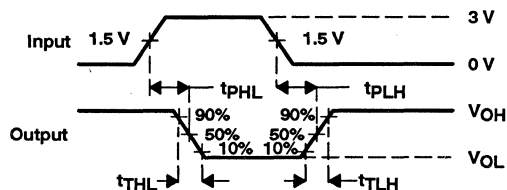
‡ Measured between the 10% and 90% points on the output waveform

§ Measured between the 3 V and -3 V points on the output waveform (EIA-232-D conditions), all unused inputs tied either high or low.

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $t_w = 25\text{ }\mu\text{s}$ ,  $\text{PRR} = 20\text{ kHz}$ ,  $Z_O = 50\text{ }\Omega$ ,  $t_r = t_f \leq 50\text{ ns}$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Test Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

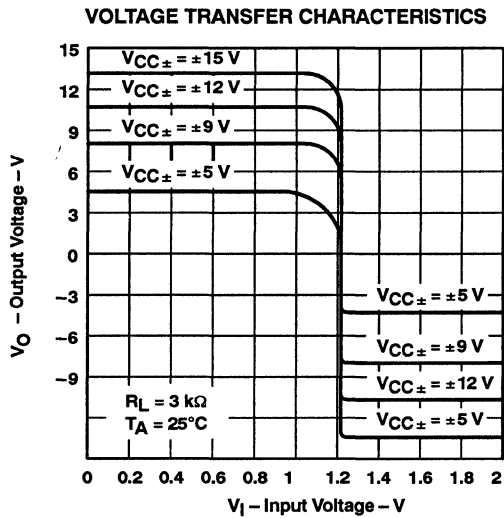


Figure 2

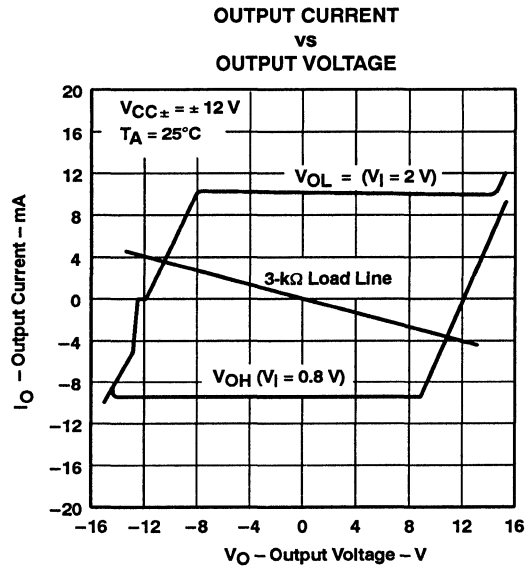


Figure 3

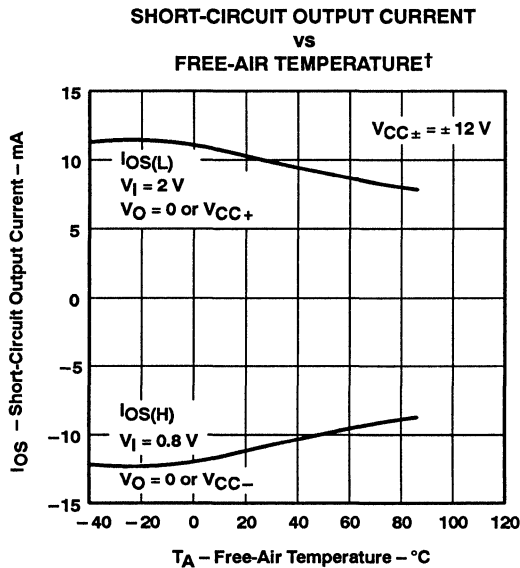


Figure 4

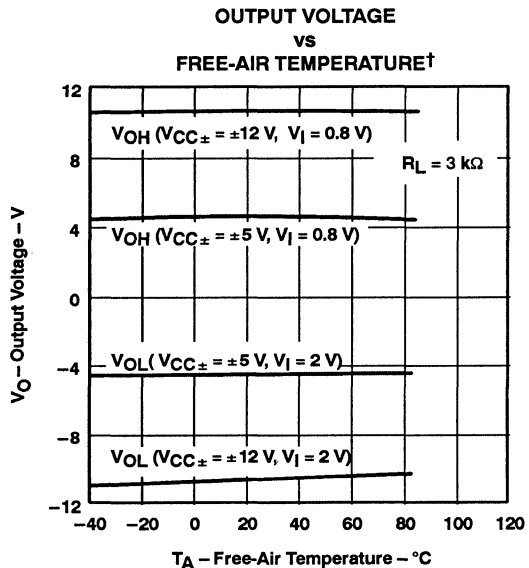


Figure 5

† Only the  $0^\circ\text{C}$  to  $70^\circ\text{C}$  portion of the curves applies to the SN75C188.



# SN65C188, SN75C188 QUAD LOW-POWER LINE DRIVERS

SLLS033D - D3075, JANUARY 1988 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

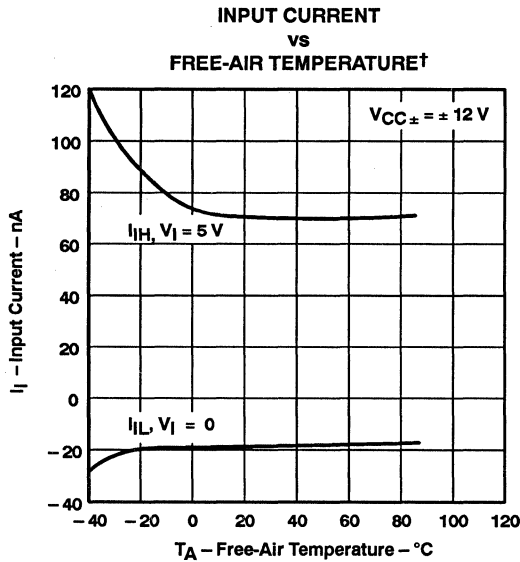


Figure 6

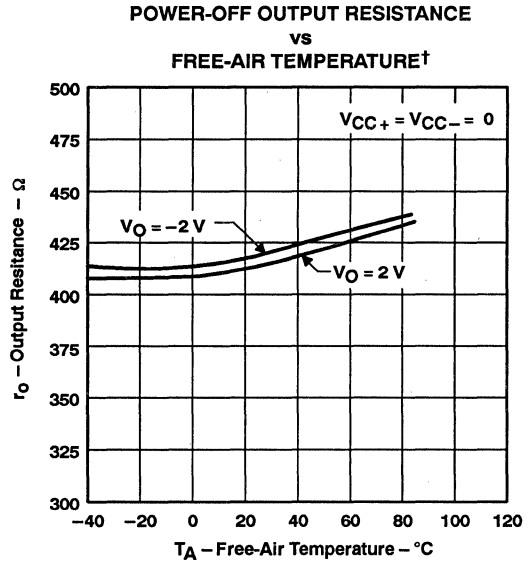


Figure 7

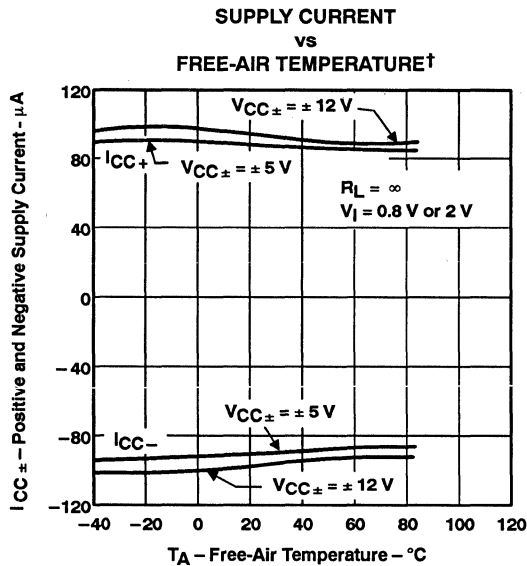


Figure 8

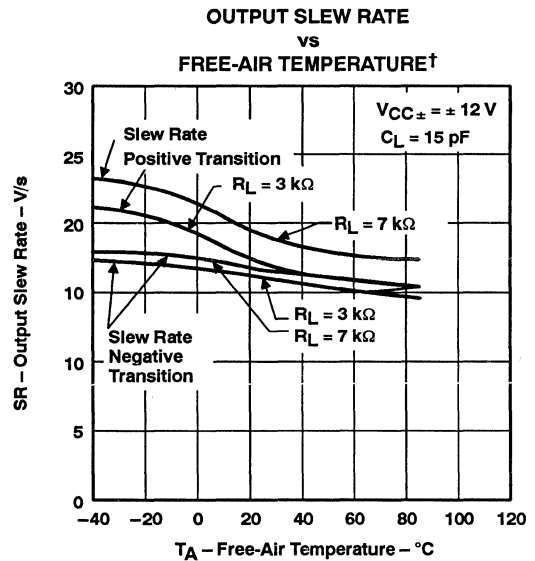


Figure 9

† Only the 0°C to 70°C portion of the curves applies to the SN75C188.

TYPICAL CHARACTERISTICS

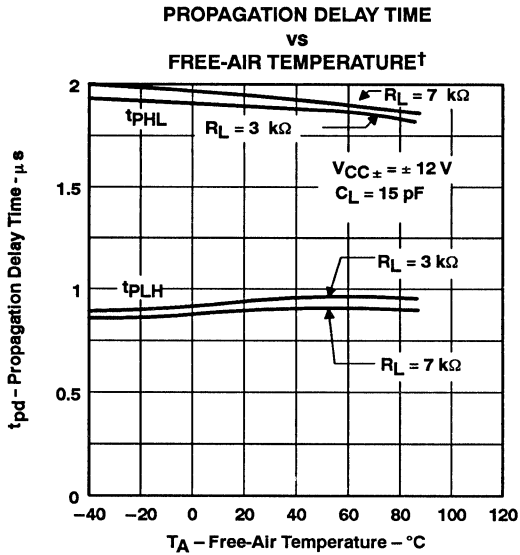


Figure 10

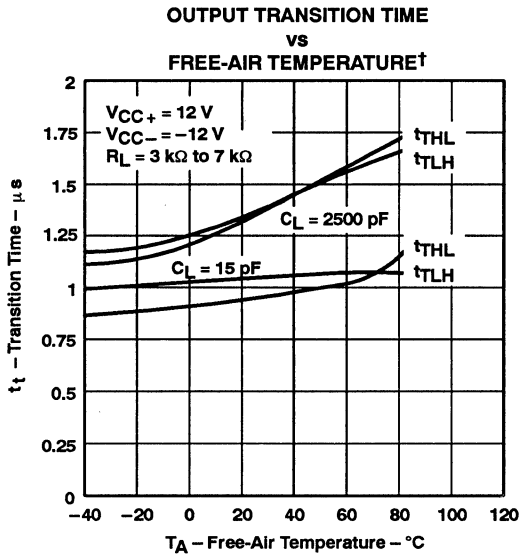


Figure 11

† Only the 0°C to 70°C portion of the curves applies to the SN75C188.

APPLICATION INFORMATION

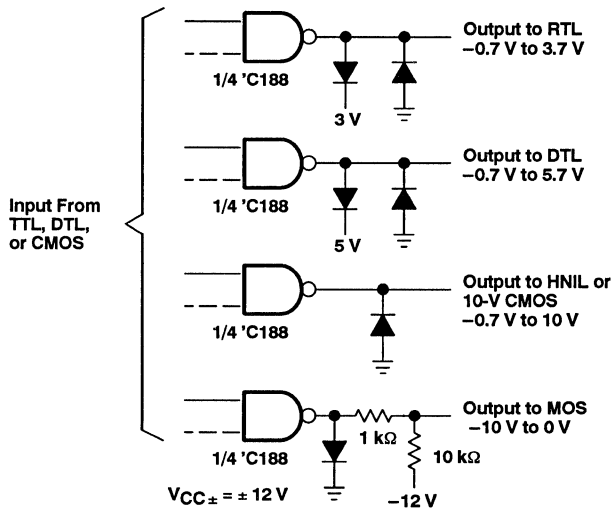
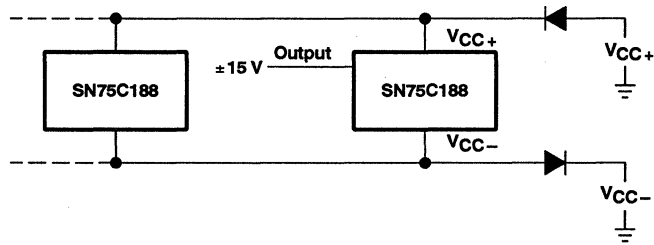


Figure 12. Logic Translator Applications

# SN65C188, SN75C188 QUAD LOW-POWER LINE DRIVERS

SLLS033D – D3075, JANUARY 1988 – REVISED MARCH 1993

## APPLICATION INFORMATION



NOTE: External diodes placed in series with the VCC+ and VCC- leads will protect the SN75C188 in the fault condition where the device outputs are shorted to  $\pm 15$  V and the power supplies are at low voltage and provide low-impedance paths to GND.

**Figure 13. Power Supply Protection to Meet Power-Off  
Fault Conditions of Standard EIA-232-D**

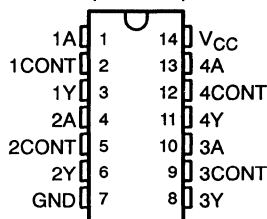
# MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

- Input Resistance . . . 3 k $\Omega$  to 7 k $\Omega$
- Input Signal Range . . .  $\pm 30$  V
- Operates From Single 5-V Supply
- Built-In Input Hysteresis (Double Thresholds)
- Response Control Provides:  
Input Threshold Shifting  
Input Noise Filtering
- Satisfies Requirements of EIA RS-232-C
- Fully Interchangeable With Motorola  
MC1489 and MC1489A

SN55189, SN55189A . . . J OR W PACKAGE  
MC1489, MC1489A, SN75189, SN75189A  
D, N, OR NS<sup>†</sup> PACKAGE

(TOP VIEW)



<sup>†</sup> The NS package is only available left-end taped and reeled and for SN75189, i.e., order SN75189NSLE.

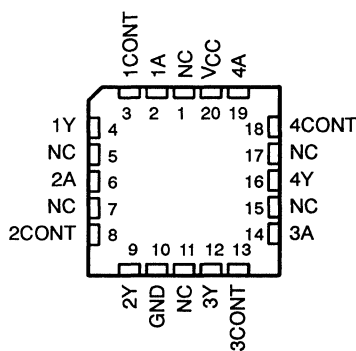
## description

These devices are monolithic low-power Schottky quad line receivers designed to satisfy the requirements of the standard interface between data terminal equipment and data communication equipment as defined by EIA Standard RS-232-C. A separate response control terminal is provided for each receiver. A resistor or a resistor and bias voltage source can be connected between this terminal and ground to shift the input threshold levels. An external capacitor can be connected between this terminal and ground to provide input noise filtering.

The SN55189 and SN55189A are characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The MC1489, MC1489A, SN75189, and SN75189A are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

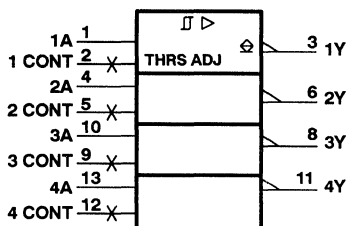
SN55189, SN55189A . . . FK PACKAGE

(TOP VIEW)



NC - No internal connection

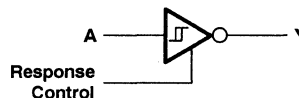
## logic symbol<sup>†</sup>



<sup>†</sup> This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, N, NS, and W packages.

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

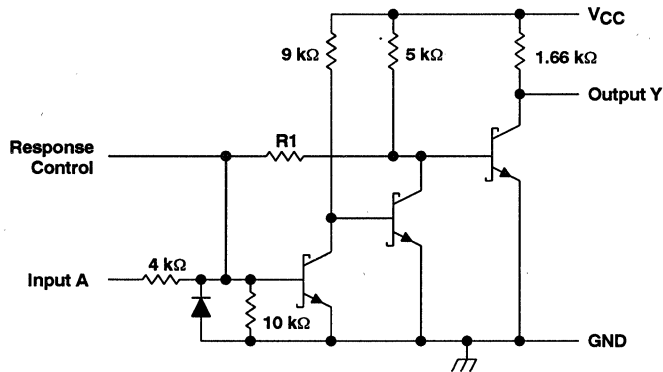
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# MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

## schematic (each receiver)



	MC1489 SN55189 SN75189	MC1489A SN55189A SN75189A
R1	8.4 kΩ	1.84 kΩ

Resistor values shown are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	SN55189 SN55189A	MC1489, MC1489A SN75189 SN75189A	UNIT
Supply voltage, $V_{CC}$ (see Note 1)	10	10	V
Input voltage	$\pm 30$	$\pm 30$	V
Output current	20	20	mA
Continuous total power dissipation	See Dissipation Rating Table		
Operating temperature range	-55 to 125	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds: FK package	260		$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	300	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or NS package		260	$^{\circ}\text{C}$

NOTE 1: All voltage values are with respect to network ground terminals.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW	N/A
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
J†	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW	N/A
NS	625 mW	4.0 mW/ $^{\circ}\text{C}$	445 mW	N/A
W	1000 mW	8.0 mW/ $^{\circ}\text{C}$	640 mW	200 mW

† In the J package, SN55189 and SN55189A chips are either silver glass or alloy mounted.

# MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

**electrical characteristics over operating free-air temperature range,  $V_{CC} = 5\text{ V} \pm 1\%$  (unless otherwise noted)**

PARAMETER	TEST FIGURE	TEST CONDITION†		SN55189 SN55189A			MC1489, MC1489A SN75189 SN75189A			UNIT
				MIN	TYP‡	MAX	MIN	TYP‡	MAX	
$V_{T+}$ Positive-going threshold voltage	1	'89	$T_A = 25^\circ\text{C}$	1	1.3	1.5	1	1.3	1.5	V
			$T_A = 0^\circ\text{C to } 70^\circ\text{C}$				0.9		1.6	
			$T_A = -55^\circ\text{C to } 125^\circ\text{C}$	0.6		1.9				
		'89A	$T_A = 25^\circ\text{C}$	1.75	1.9	2.25	1.75	1.9	2.25	
			$T_A = 0^\circ\text{C to } 70^\circ\text{C}$				1.55		2.25	
			$T_A = -55^\circ\text{C to } 125^\circ\text{C}$	1.30		2.65				
$V_{T-}$ Negative-going threshold voltage	1	'89, '89A	$T_A = 25^\circ\text{C}$	0.75	1.0	1.25	0.75	1.0	1.25	V
			$T_A = 0^\circ\text{C to } 70^\circ\text{C}$				0.65		1.25	
			$T_A = -55^\circ\text{C to } 125^\circ\text{C}$	0.35		1.6				
$V_{OH}$ High-level output voltage	1	$V_I = 0.75\text{ V}, I_{OH} = -0.5\text{ mA}$		2.6	4	5	2.6	4	5	V
		Input open, $I_{OH} = -0.5\text{ mA}$		2.6	4	5	2.6	4	5	
$V_{OL}$ Low-level output voltage	1	$V_I = 3\text{ V}, I_{OL} = 10\text{ mA}$			0.2	0.45		0.2	0.45	V
$I_{IH}$ High-level input current	2	$V_I = 25\text{ V}$		3.6		8.3	3.6		8.3	mA
		$V_I = 3\text{ V}$		0.43			0.43			
$I_{IL}$ Low-level input current	2	$V_I = -25\text{ V}$		-3.6		-8.3	-3.6		-8.3	mA
		$V_I = -3\text{ V}$		-0.43			-0.43			
$I_{OS}$ Short-circuit output current	3				-3			-3		mA
$I_{CC}$ Supply current	2	$V_I = 5\text{ V},$ Outputs open			20	26		20	26	mA

† All characteristics are measured with the response control terminal open.

‡ All typical values are at  $V_{CC} = 5\text{ V}, T_A = 25^\circ\text{C}$ .

**switching characteristics,  $V_{CC} = 5\text{ V}, C_L = 15\text{ pF}, T_A = 25^\circ\text{C}$**

PARAMETER	TEST FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	4	$R_L = 3.9\text{ k}\Omega$		25	85	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output		$R_L = 390\ \Omega$		25	50	
$t_{TLH}$ Transition time, low-to-high-level output		$R_L = 3.9\text{ k}\Omega$		120	175	ns
$t_{THL}$ Transition time, high-to-low-level output		$R_L = 390\ \Omega$		10	20	



# MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION†

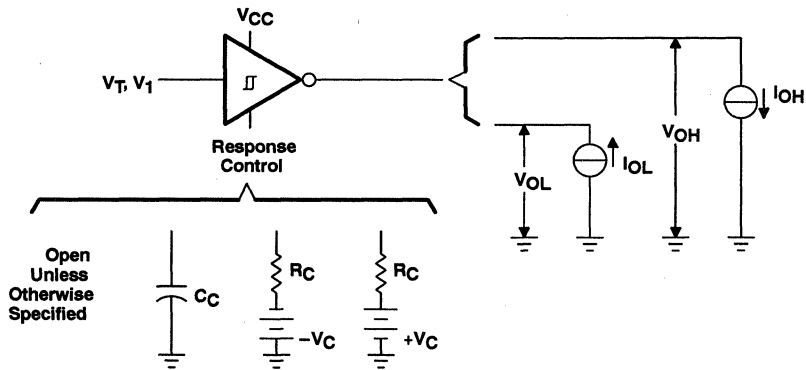
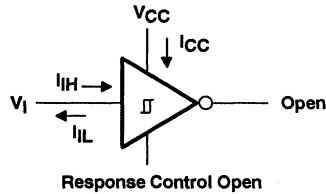


Figure 1.  $V_{T+}$ ,  $V_{T-}$ ,  $V_{OH}$ ,  $V_{OL}$



$I_{CC}$  is tested for all four receivers simultaneously.

Figure 2.  $I_{IH}$ ,  $I_{IL}$ ,  $I_{CC}$

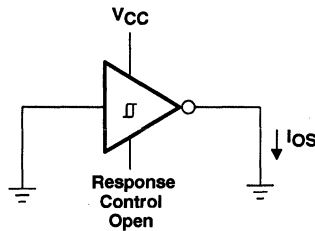


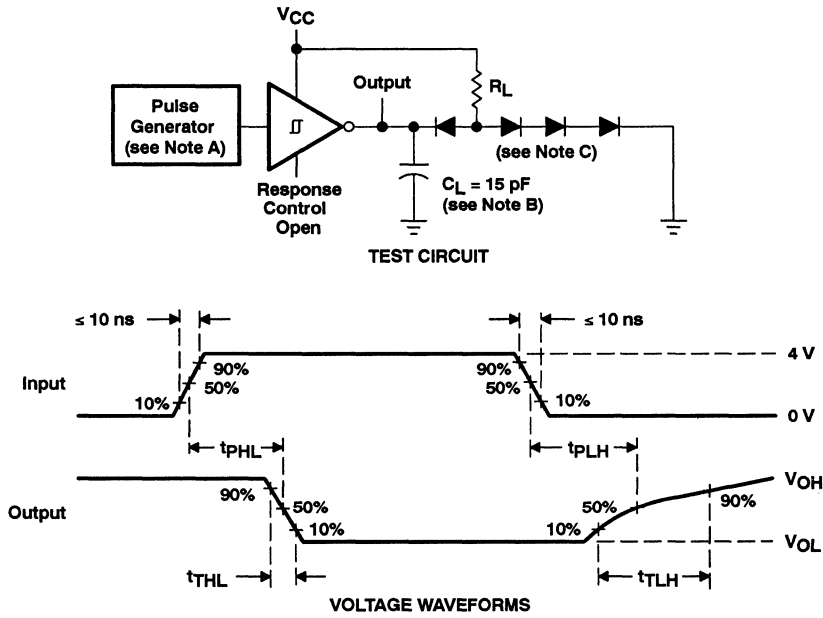
Figure 3.  $I_{OS}$

† Arrows indicate actual direction of current flow. Current into a terminal is a positive value.

MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A  
QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $t_w = 500 \text{ ns}$ .  
B.  $C_L$  includes probe and jig capacitances.  
C. All diodes are 1N3064 or equivalent.

Figure 4. Test Circuit and Voltage Waveforms



# MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

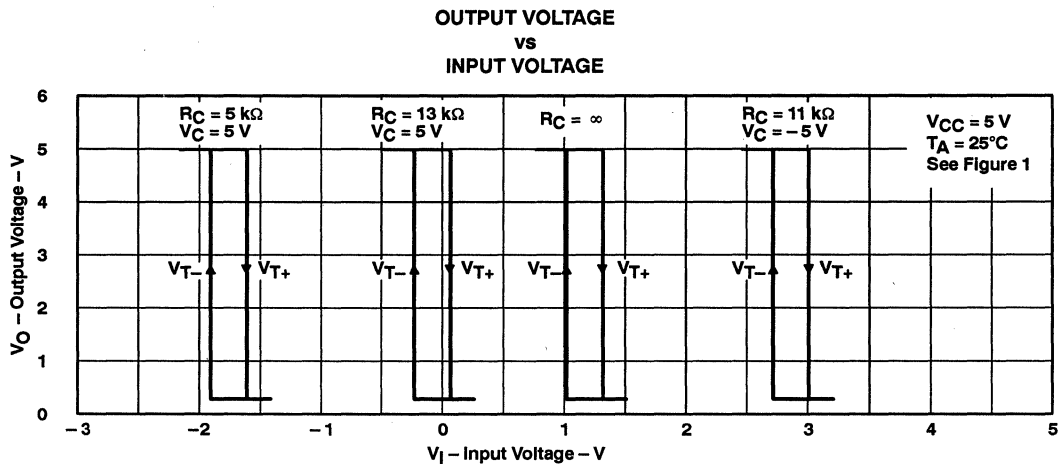


Figure 5

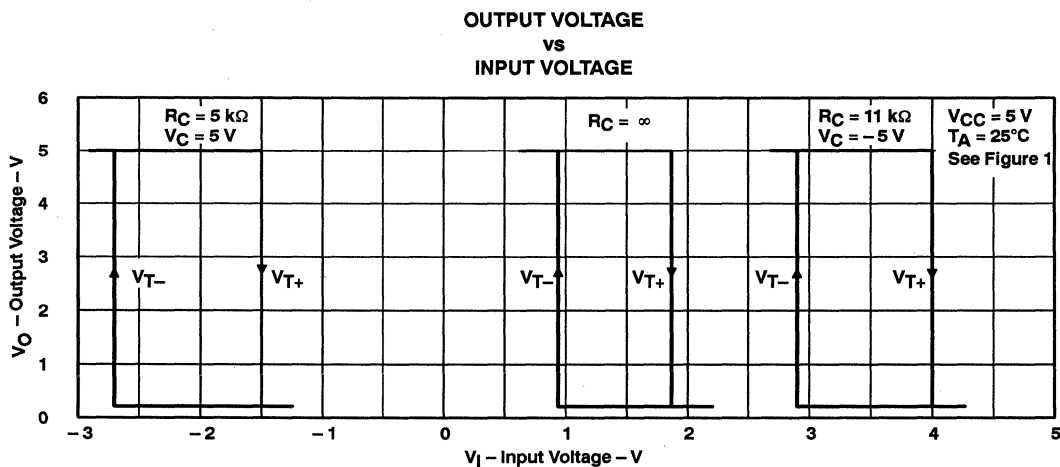


Figure 6

# MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS†

**INPUT THRESHOLD VOLTAGE  
VS  
FREE-AIR TEMPERATURE**

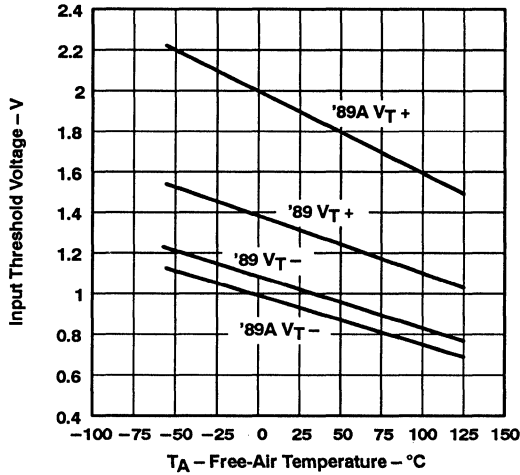


Figure 7

**INPUT THRESHOLD VOLTAGE  
VS  
SUPPLY VOLTAGE**

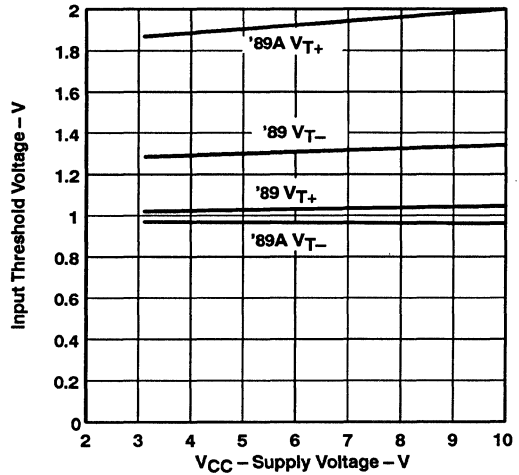


Figure 8

**SN75189  
NOISE REJECTION**

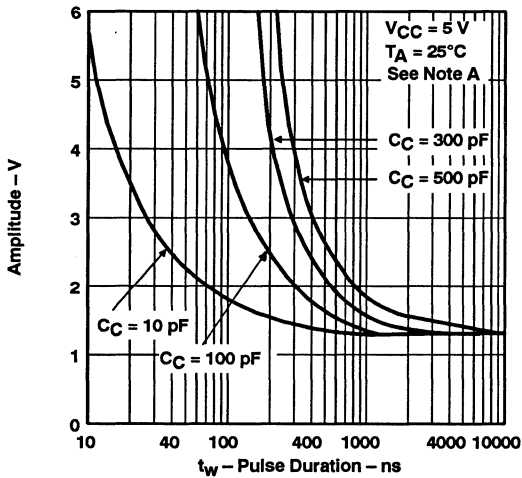


Figure 9

**SN75189A  
NOISE REJECTION**

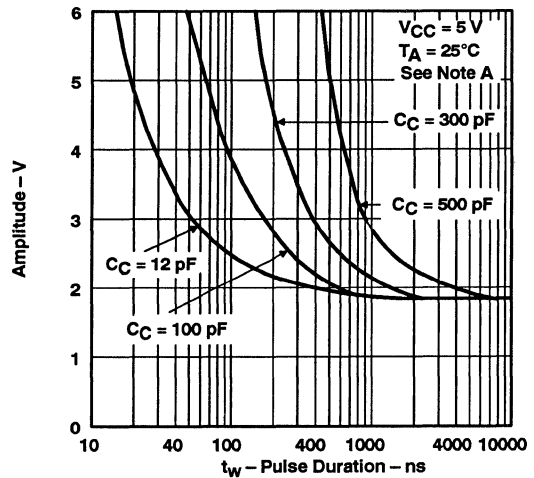


Figure 10

† Data for free-air temperatures below 0°C and above 70°C are applicable to SN55189 and SN55189A circuits only.

NOTE A: This figure shows the maximum amplitude of a positive-going pulse that, starting from 0 V, will not cause a change of the output level.

# MC1489, MC1489A, SN55189, SN55189A, SN75189, SN75189A QUAD LINE RECEIVERS

SLLS095A - D1619, SEPTEMBER 1973 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

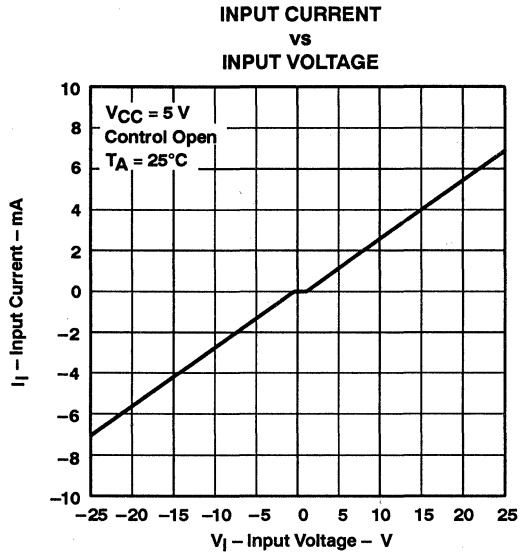


Figure 11

# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C – D3144, OCTOBER 1988 – REVISED MARCH 1993

- Meets Standard EIA-232-D (Revision of RS-232-C)
- Low Supply Current . . . 420  $\mu$ A Typ
- Preset On-Chip Input Noise Filter
- Built-in Input Hysteresis
- Response and Threshold Control Inputs
- Push-Pull Outputs
- ESD Protection Exceeds 500 V Per MIL-STD-883C, Method 3015
- Functionally Interchangeable and Pin Compatible With Texas Instruments SN75189/SN75189A, Motorola MC1489/MC1489A, and National Semiconductor DS14C88A

## description

The SN65C189, SN65C189A, SN75C189, and SN75C189A are low-power bipolar quad line receivers that are used to interface data terminal equipment (DTE) with data circuit-terminating equipment (DCE). These devices have been designed to conform with Standard ANSI/EIA-232-D-1986, which supersedes RS-232-C.

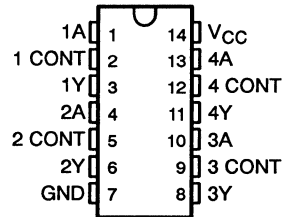
The SN65C189 and SN75C189 have a 0.33 V typical hysteresis compared with 0.97 V for the SN65C189A and SN75C189A. Each receiver has provision for adjustment of the overall input threshold levels. This is achieved by choosing external series resistors and voltages to provide bias levels for the response control pins. The output is in the high logic state if the input is left open circuited or shorted to ground.

These devices have an on-chip filter that rejects input pulses of shorter than 1- $\mu$ s minimum duration. An external capacitor may be connected from the control pins to ground to provide further input noise filtering for each receiver.

The SN65C189, SN75C189, SN65C189A, and SN75C189A have been designed using low-power techniques in a bipolar technology. In most applications, these receivers will interface to single inputs of peripheral devices such as UARTs, ACEs, or microprocessors. By using sampling, such peripheral devices are usually insensitive to the transition times of the input signals. If this is not the case or for other uses, it is recommended that the SN65C189, SN75C189, SN65C189A, and SN75C189A outputs be buffered by single Schmitt input gates or single gates of the HCMOS, ALS, or 74F logic families.

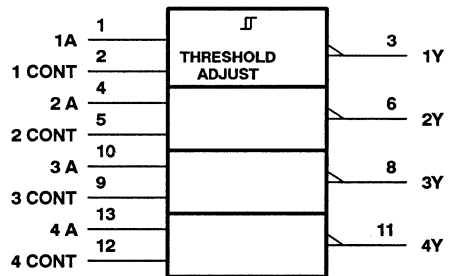
The SN65C189 and SN65C189A are characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75C189 and SN75C189A are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

D, DB†, N, OR NS† PACKAGE  
(TOP VIEW)



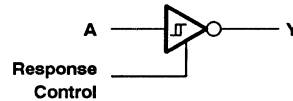
† The DB and NS packages are only available left-end taped and reeled, i.e., order SN\_5C189ADBLE or SN\_5C189ANSLE.

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

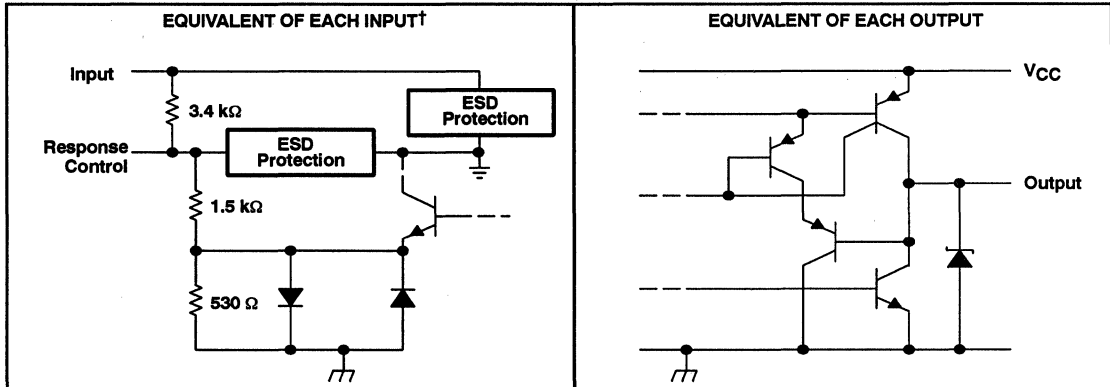
## logic diagram (each receiver)



# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

## schematic of inputs and outputs



† All resistor values shown are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage range, $V_I$	-30 V to 30 V
Output voltage range	-0.3 V to $V_{CC} + 0.3$ V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65C189, SN65C189A	-40°C to 85°C
SN75C189, SN75C189A	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
DB	525 mW	4.2 mW/°C	336 mW	273 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW
NS	500 mW	4.0 mW/°C	320 mW	260 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.5	5	6	V
Input voltage, $V_I$ (see Note 2)	-25		25	V
High-level output current, $I_{OH}$			-3.2	mA
Low-level output current, $I_{OL}$			3.2	mA
Response control current			±1	mA
Operating free-air temperature, $T_A$	SN65C189, SN65C189A		-40	85
	SN75C189, SN75C189A		0	70
				°C

NOTE 2: The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only, e.g., if -10 V is a maximum, the typical value is a more negative voltage.



# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

**electrical characteristics over recommended free-air temperature range,  $V_{CC} = 5 V \pm 10\%$  (unless otherwise noted) (see Note 3)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$V_{T+}$	Positive-going threshold voltage	See Figure 1	1		1.5	V	
					1.6		2.25
$V_{T-}$	Negative-going threshold voltage	See Figure 1	0.75		1.25	V	
					0.75		1
$V_{hys}$	Input hysteresis	See Figure 1	0.15	0.33		V	
					0.65		0.97
$V_{OH}$	High-level output voltage	$V_{CC} = 4.5 V$ to $6 V$ , $I_{OH} = -20 \mu A$	3.5			V	
		$V_I = 0.75 V$ , $I_{OH} = -3.2 mA$			2.5		
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.5 V$ to $6 V$ , $I_{OL} = 3.2 mA$			0.4	V	
$I_{IH}$	High-level input current	See Figure 2		$V_I = 25 V$	8.3	mA	
				$V_I = 3 V$	0.43		1
$I_{IL}$	Low-level input current	See Figure 2		$V_I = -25 V$	-8.3	mA	
				$V_I = -3 V$	-0.43		-1
$I_{OS}$	Short-circuit output current	See Figure 3			-35	mA	
$I_{CC}$	Supply current	$V_I = 5 V$ , See Figure 2	No load,		420	700	$\mu A$

† All typical values are at  $T_A = 25^\circ C$ .

NOTE 3: All characteristics are measured with response control terminal open.

**switching characteristics,  $V_{CC} = 5 V \pm 10\%$ ,  $T_A = 25^\circ C$**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
$t_{PLH}$	Propagation delay time, low-to-high-level output	$R_L = 5 k\Omega$ , $C_L = 50 pF$ , See Figure 4			6	$\mu s$	
$t_{PHL}$	Propagation delay time, high-to-low-level output				6	$\mu s$	
$t_{TLH}$	Transition time, low-to-high-level output‡				500	ns	
$t_{THL}$	Transition time, high-to-low-level output‡				300	ns	
$t_w(N)$	Duration of longest pulse rejected as noise§			1		6	$\mu s$

‡ Measured between 10% and 90% points of output waveform.

§ The intent of this specification is that any input pulse of less than  $1 \mu s$  will have no effect on the output, and any pulse duration of greater than  $6 \mu s$  will cause the output to change state twice. Reaction to a pulse duration between  $1 \mu s$  and  $6 \mu s$  is uncertain.

# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION†

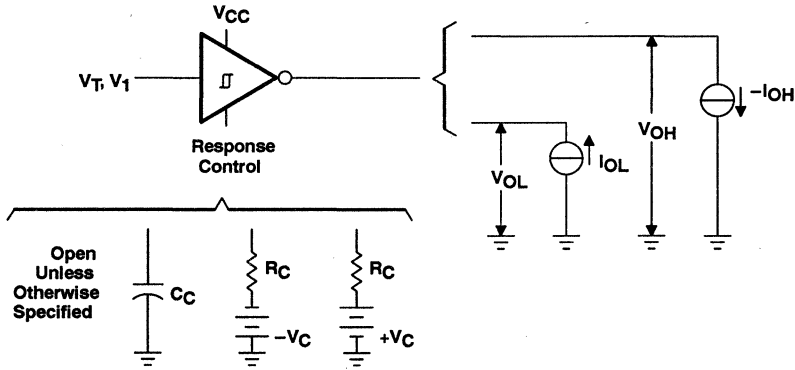


Figure 1.  $V_{T+}$ ,  $V_{T-}$ ,  $V_{OH}$ ,  $V_{OL}$

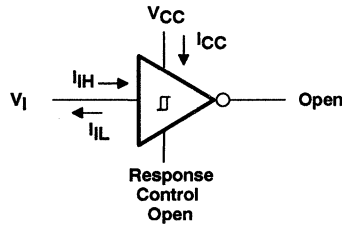


Figure 2.  $I_{IH}$ ,  $I_{IL}$ ,  $I_{CC}$

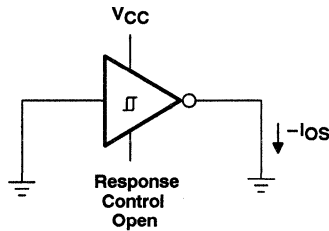


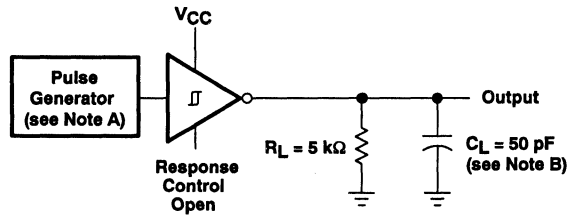
Figure 3.  $I_{OS}$

† Arrows indicate actual direction of current flow. Current into a terminal is a positive value.

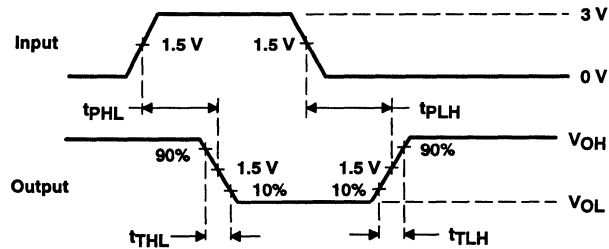
# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

Figure 4. Test Circuit and Voltage Waveforms

- NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50\ \Omega$ ,  $t_W = 25\ \mu\text{s}$ .  
B.  $C_L$  includes probe and jig capacitances.

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# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C – D3144, OCTOBER 1988 – REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

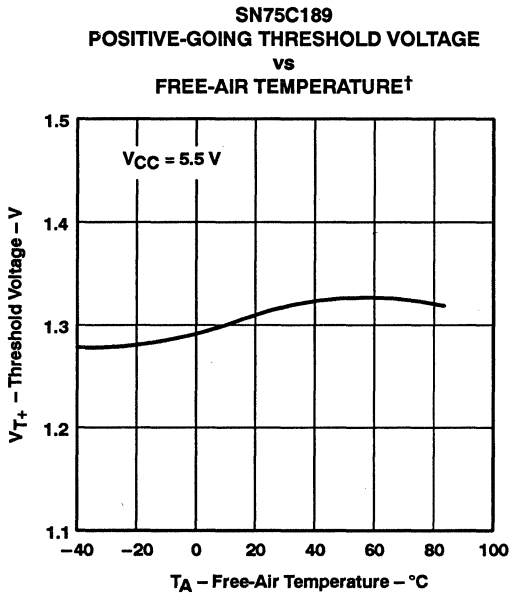


Figure 5

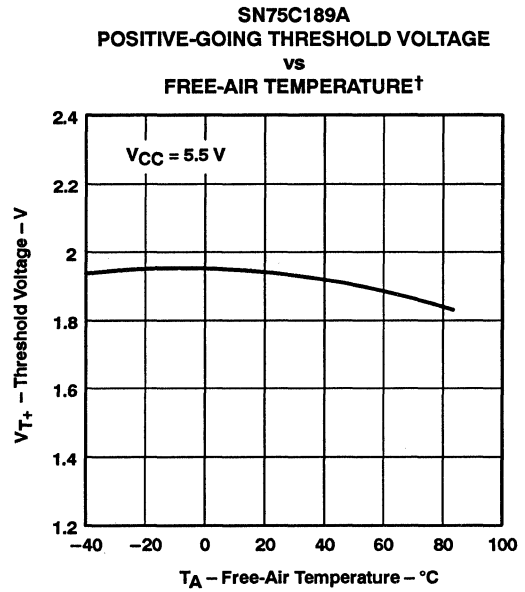


Figure 6

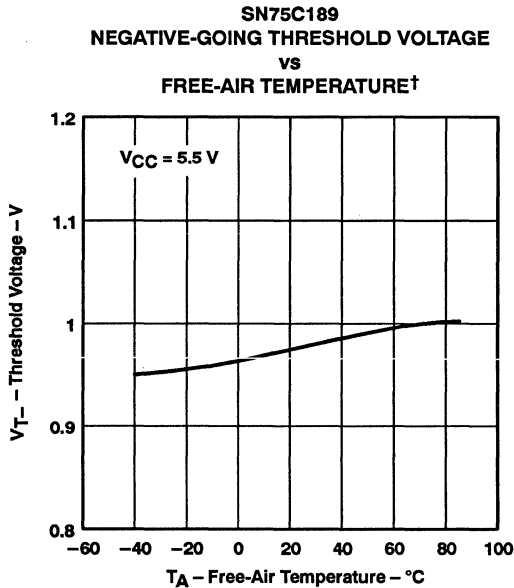


Figure 7

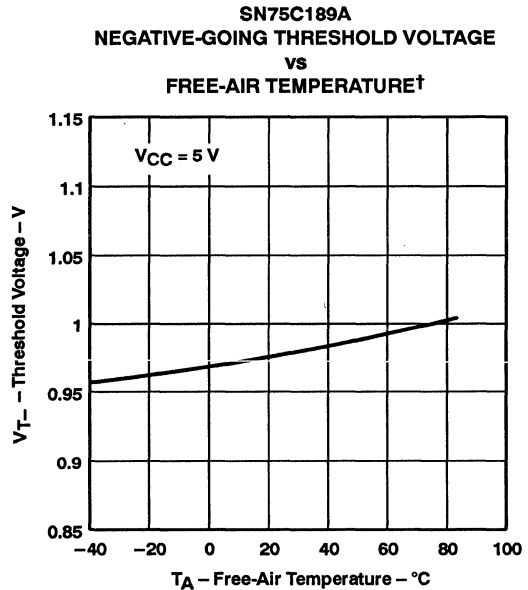


Figure 8

† Only the 0°C to 70°C portion of the curves applies to the SN75'.

# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

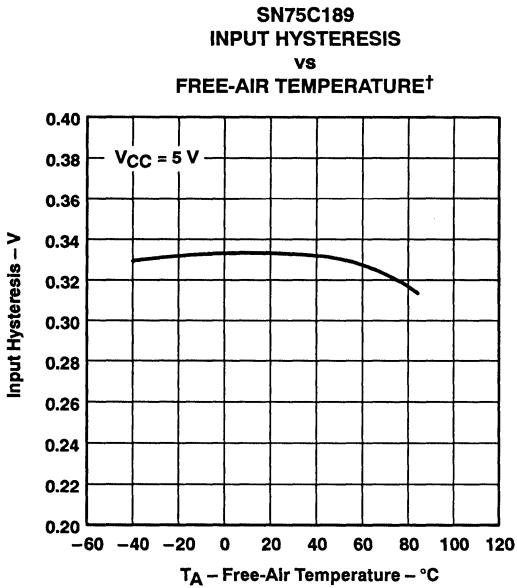


Figure 9

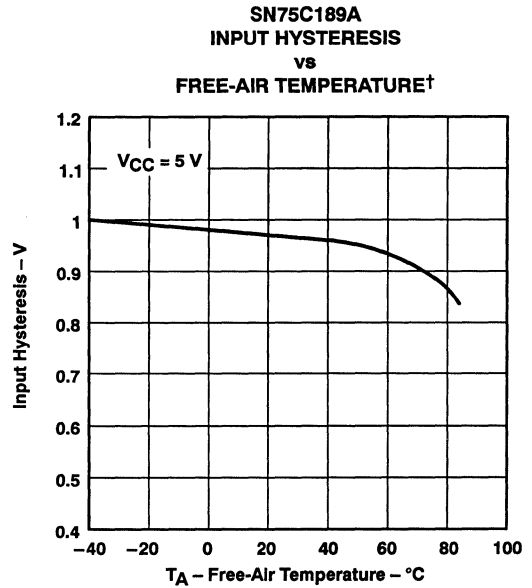


Figure 10

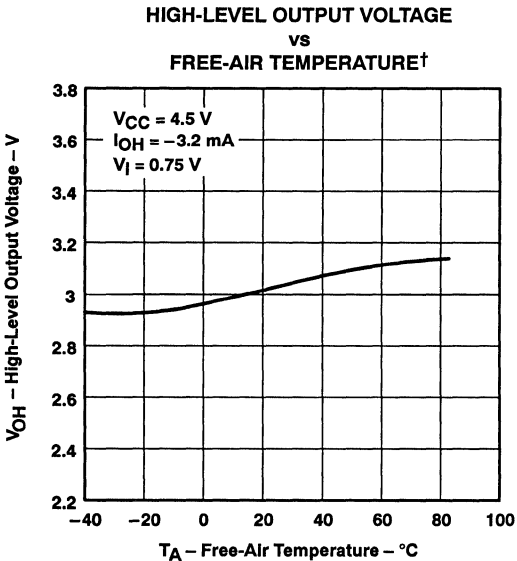


Figure 11

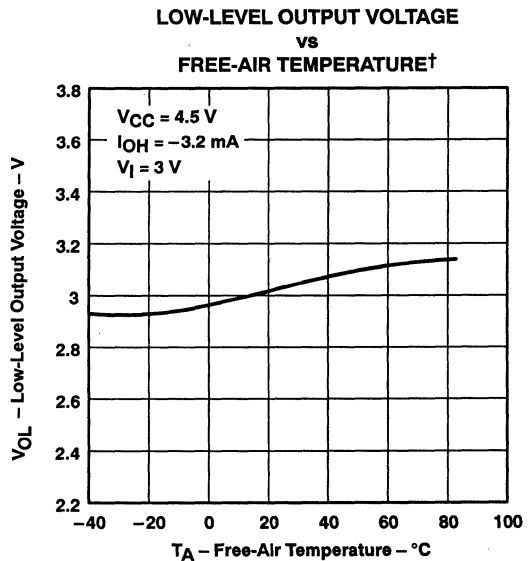


Figure 12

† Only the 0°C to 70°C portion of the curves applies to the SN75'.

# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

**SN75C189**  
HIGH-LEVEL INPUT CURRENT  
vs  
FREE-AIR TEMPERATURE†

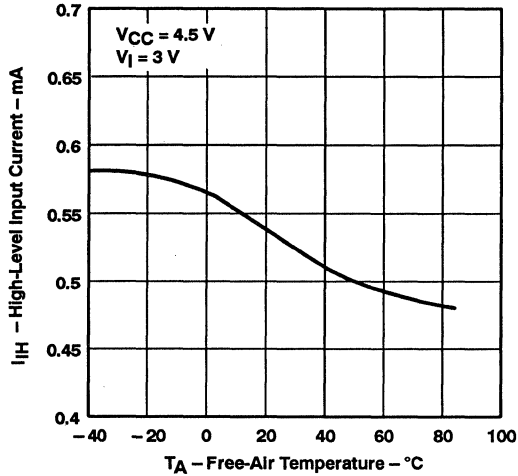


Figure 13

**SN75C189A**  
HIGH-LEVEL INPUT CURRENT  
vs  
FREE-AIR TEMPERATURE†

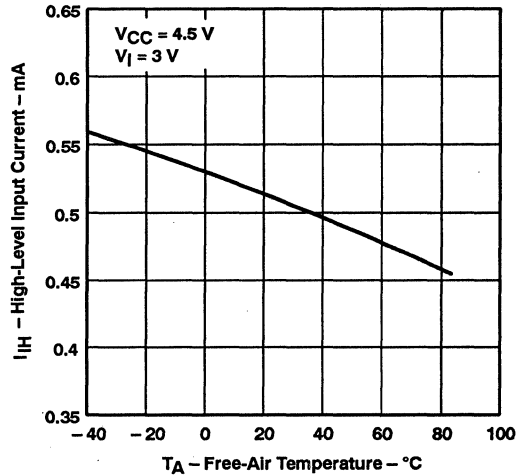


Figure 14

**SN75C189**  
LOW-LEVEL INPUT CURRENT  
vs  
FREE-AIR TEMPERATURE†

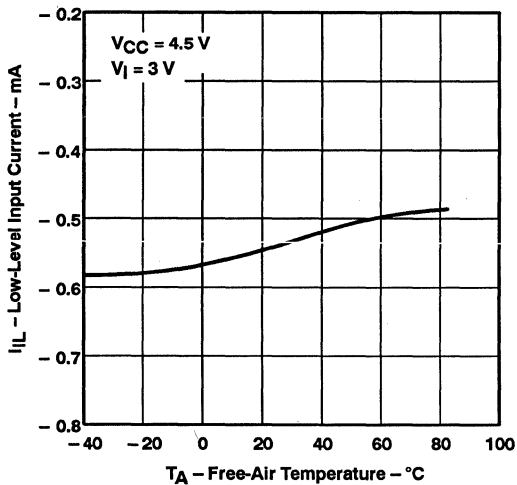


Figure 15

**SN75C189A**  
LOW-LEVEL INPUT CURRENT  
vs  
FREE-AIR TEMPERATURE†

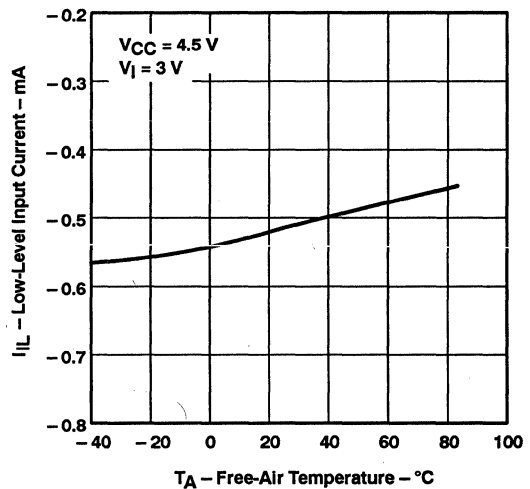


Figure 16

† Only the 0°C to 70°C portion of the curves applies to the SN75'.

# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

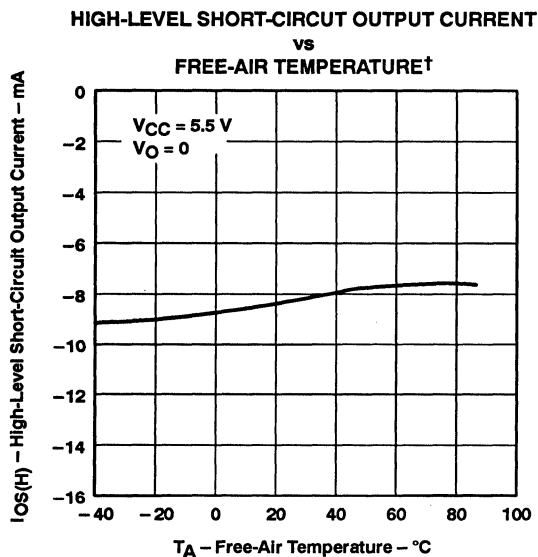


Figure 17

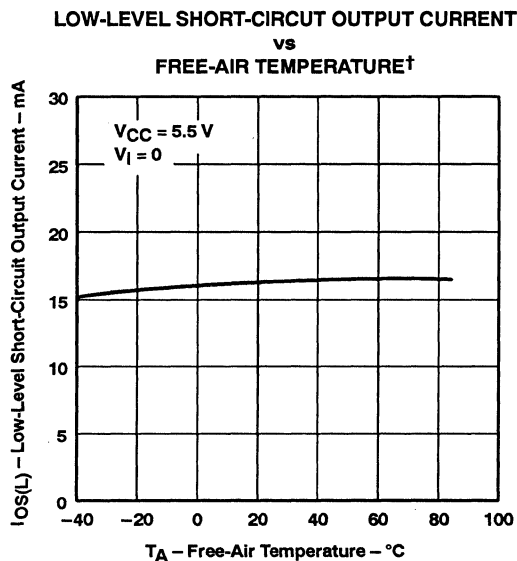


Figure 18

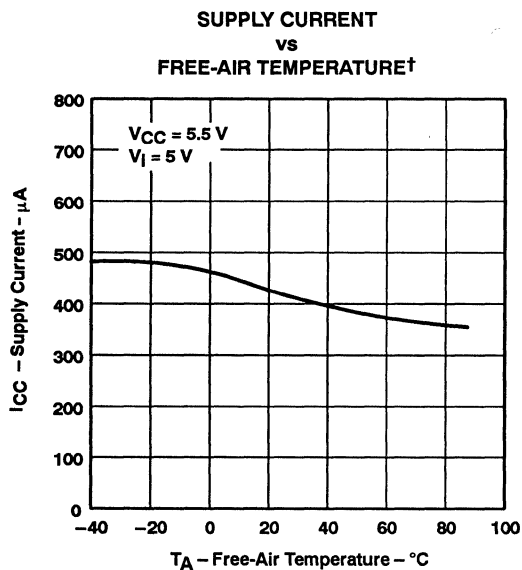


Figure 19

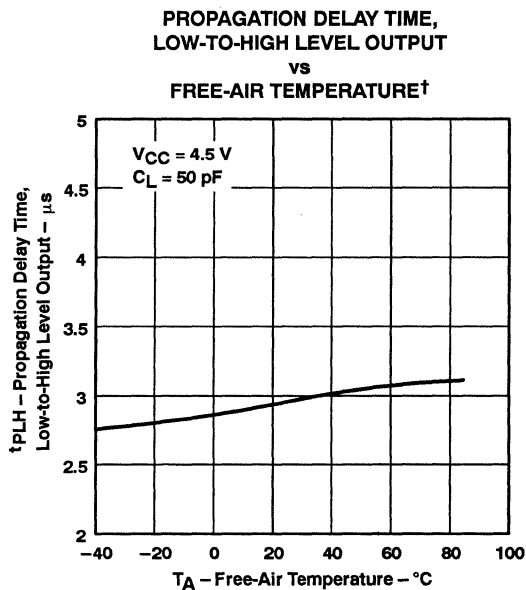


Figure 20

† Only the 0°C to 70°C portion of the curves applies to the SN75'.

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# SN65C189, SN65C189A, SN75C189, SN75C189A QUAD LOW-POWER LINE RECEIVERS

SLLS041C - D3144, OCTOBER 1988 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

PROPAGATION DELAY TIME,  
HIGH-TO-LOW-LEVEL OUTPUT  
vs  
FREE-AIR TEMPERATURE†

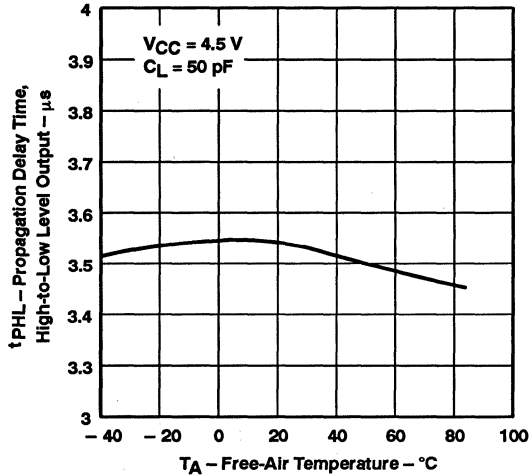


Figure 21

TRANSITION TIME,  
LOW-TO-HIGH-LEVEL  
vs  
FREE-AIR TEMPERATURE†

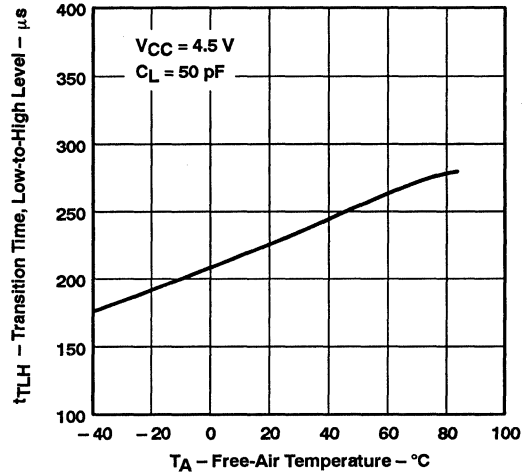


Figure 22

TRANSITION TIME,  
HIGH-TO-LOW-LEVEL OUTPUT  
vs  
FREE-AIR TEMPERATURE†

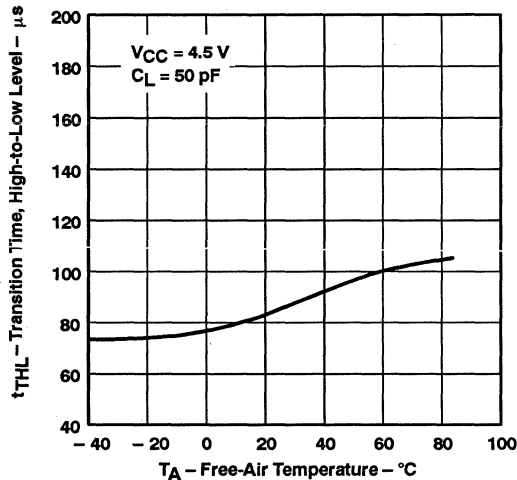


Figure 23

† Only the 0°C to 70°C portion of the curves applies to the SN75'.

# SN75ALS191 DUAL DIFFERENTIAL LINE DRIVER

SLLS032A - D3068, DECEMBER 1987 - REVISED AUGUST 1989

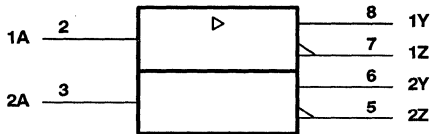
- Meets EIA Standard RS-422-A
- High Speed, Low-Power ALS Design
- TTL-and CMOS-Input Compatibility
- Single 5-V Supply Operation
- Output Short-Circuit Protection
- Improved Replacement for the uA9638

## description

The SN75ALS191 is a dual high-speed differential line driver designed to meet EIA Standard RS-422-A. The inputs are TTL- and CMOS-compatible and have input clamp diodes. Schottky-diode-clamped transistors are used to minimize propagation delay time. This device operates from a single 5-V power supply and is supplied in 8-pin packages.

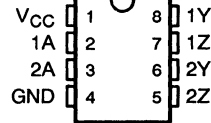
The SN75ALS191 is characterized for operation from 0°C to 70°C.

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

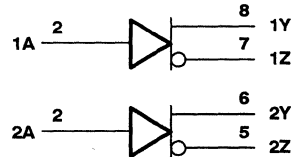
## D OR P PACKAGE (TOP VIEW)



## FUNCTION TABLE (each driver)

INPUT A	OUTPUTS	
	Y	Z
H	H	L
L	L	H

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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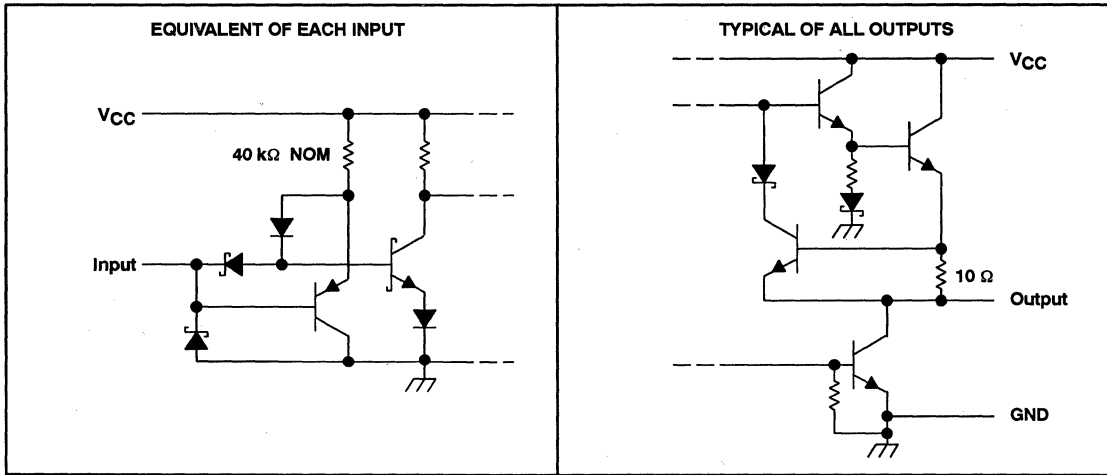
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# SN75ALS191 DUAL DIFFERENTIAL LINE DRIVER

SLLS032A - D3068, DECEMBER 1987 - REVISED AUGUST 1989

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	7 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values except differential output voltage  $V_{OD}$  are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

# SN75ALS191 DUAL DIFFERENTIAL LINE DRIVER

SLLS032A – D3068, DECEMBER 1987 – REVISED AUGUST 1989

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-50	mA
Low-level output current, $I_{OL}$			50	mA
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$V_{CC} = 4.75\text{ V}$ , $I_I = -18\text{ mA}$		-1	-1.2	V
$V_{OH}$ High-level output voltage	$V_{CC} = 4.75\text{ V}$ , $V_{IL} = 0.8\text{ V}$ , $I_{OH} = -10\text{ mA}$	2.5	3.3		
	$I_{OH} = -40\text{ mA}$	2			
$V_{OL}$ Low-level output voltage	$V_{CC} = 4.75\text{ V}$ , $V_{IH} = 2\text{ V}$ , $I_{OL} = 40\text{ mA}$ , $V_{IL} = 0.8\text{ V}$			0.5	V
$ V_{OD1} $ Differential output voltage	$V_{CC} = 5.25\text{ V}$ , $I_O = 0$			$2 V_{OD2}$	V
$ V_{OD2} $ Differential output voltage		2			V
$\Delta  V_{OD} $ Change in magnitude of differential output voltage‡	$V_{CC} = 4.75\text{ V to } 5.25\text{ V}$ , See Figure 1, $R_L = 100\ \Omega$			$\pm 0.4$	V
$V_{OC}$ Common-mode output voltage§				3	V
$\Delta  V_{OC} $ Change in magnitude of common-mode output voltage‡				$\pm 0.4$	V
$I_O$ Output current with power off	$V_{CC} = 0$				
	$V_O = 6\text{ V}$		0.1	100	$\mu\text{A}$
	$V_O = -0.25\text{ V}$		-0.1	-100	$\mu\text{A}$
	$V_O = -0.25\text{ V to } 6\text{ V}$			$\pm 100$	$\mu\text{A}$
$I_I$ Input current	$V_{CC} = 5.25\text{ V}$ , $V_I = 5.5\text{ V}$			50	$\mu\text{A}$
$I_{IH}$ High-level input current	$V_{CC} = 5.25\text{ V}$ , $V_I = 2.7\text{ V}$			25	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_{CC} = 5.25\text{ V}$ , $V_I = 0.5\text{ V}$			200	$\mu\text{A}$
$I_{OS}$ Short-circuit output current¶	$V_{CC} = 5.25\text{ V}$ , $V_O = 0$	-50		-150	mA
$I_{CC}$ Supply current (all drivers)	$V_{CC} = 5.25\text{ V}$ , No load, All inputs at 0 V		32	40	mA

† All typical values are at  $V_{CC} = 5\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡  $|V_{OD}|$  and  $|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

§ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

¶ Only one output at a time should be shorted, and duration of the short circuit should not exceed one second.

## switching characteristics over recommended range of operating free-air temperature, $V_{CC} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	MIN	TYP#	MAX	UNIT
$t_{dD}$ Differential-output delay time	$C_L = 15\text{ pF}$ , $R_L = 100\ \Omega$ , See Figure 2		3.5	7	ns
$t_{rD}$ Differential-output transition time			3.5	7	ns
Skew			1.5	4	ns

# Typical values are at  $T_A = 25^\circ\text{C}$ .





# SN75ALS191 DUAL DIFFERENTIAL LINE DRIVER

SLLS032A - D3068, DECEMBER 1987 - REVISED AUGUST 1989

## PARAMETER MEASUREMENT INFORMATION

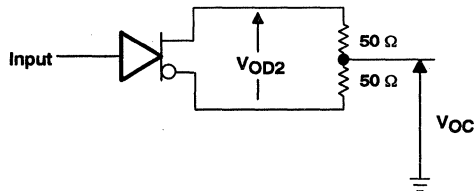
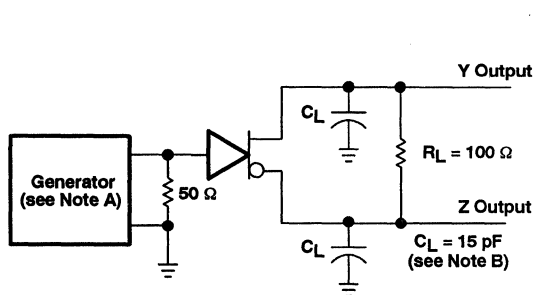
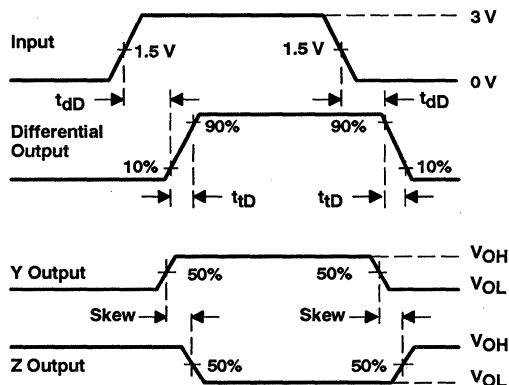


Figure 1. Differential and Common-Mode Output Voltages



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTES: A. The input pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 500$  kHz,  $t_w = 100$  ns,  $t_r = \leq 5$  ns.  
B.  $C_L$  includes probe and jig capacitance.

Figure 2. Test Circuit and Voltage Waveforms

# SN55ALS192, SN75ALS192 QUAD DIFFERENTIAL LINE DRIVERS

SLLS007B - D2904, JULY 1985 - REVISED MARCH 1993

- Meets EIA Standard RS-422-A
- High-Speed, Low-Power ALS Design
- 3-State TTL Compatible
- Single 5-V Supply Operation
- High Output Impedance in Power-Off Condition
- Complementary Output Enable Inputs
- Improved Replacement for the AM26LS31

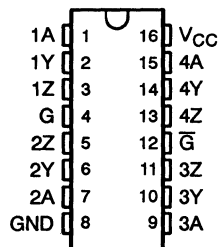
## description

These quad differential line drivers are designed for data transmission over twisted-pair or parallel-wire transmission lines. They meet the requirements of EIA Standard RS-422-A and are compatible with 3-state TTL circuits. Advanced low-power Schottky technology provides high speed without the usual power penalties. Standby supply current is typically only 26 mA, while typical propagation delay time is less than 10 ns.

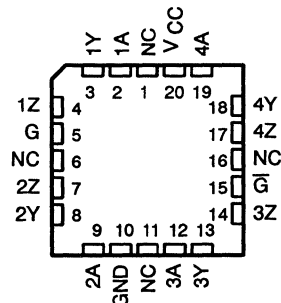
High-impedance inputs maintain input currents low, less than 1  $\mu$ A for a high level and less than 100  $\mu$ A for a low level. Complementary enable inputs, G and  $\bar{G}$ , allow these devices to be enabled at either a high input level or low input level. The SN75ALS192 is capable of data rates in excess of 20 megabits per second and is designed to operate with the SN75ALS193 quad line receiver. The SN55ALS192 is also capable of data rates in excess of 20 megabits per second and designed to operate with the SN55ALS193; however, it may be limited to a lower bit rate based on the temperature. Reference should be made to the Dissipation Rating Table and Figure 15.

The SN55ALS192 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75ALS192 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

SN55ALS192 ... J OR W PACKAGE  
SN75ALS192 ... D OR N PACKAGE  
(TOP VIEW)



SN55ALS192 ... FK PACKAGE  
(TOP VIEW)



NC - No internal connection

FUNCTION TABLE  
(each driver)

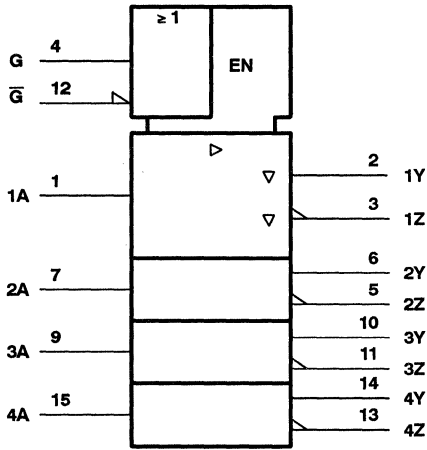
INPUT A	ENABLES		OUTPUTS	
	G	$\bar{G}$	Y	$\bar{Z}$
H	H	X	H	L
L	H	X	L	H
H	X	L	H	L
L	X	L	L	H
X	L	H	Z	Z

H = high level, L = low level,  
Z = high impedance (off), X = irrelevant

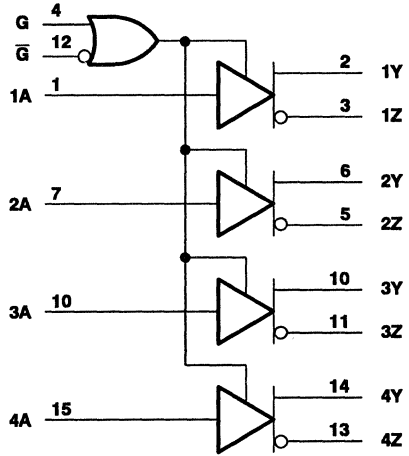
# SN55ALS192, SN75ALS192 QUAD DIFFERENTIAL LINE DRIVERS

SLLS007B – D2904, JULY 1985 – REVISED MARCH 1993

## logic symbol†



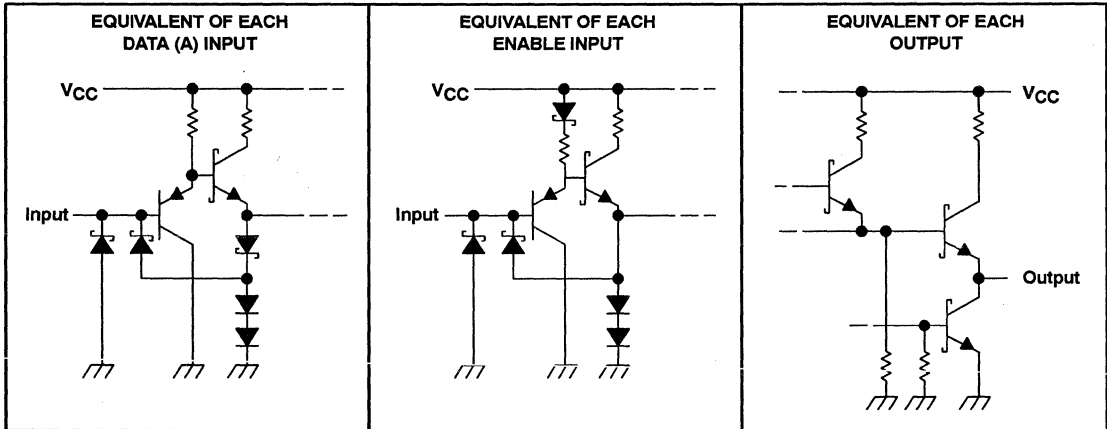
## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, N, and W packages.

## schematics of inputs and outputs



# SN55ALS192, SN75ALS192 QUAD DIFFERENTIAL LINE DRIVERS

SLLS007B – D2904, JULY 1985 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	7 V
Off-state output voltage	6 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN55ALS192	– 55°C to 125°C
SN75ALS192	0°C to 70°C
Storage temperature range	– 65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or W package	260°C

NOTE 1: All voltage values except differential output voltage  $V_{OD}$  are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J†	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	N/A
W	1000 mW	8.0 mW/°C	640 mW	200 mW

† In the J package, the SN55ALS192 chips are either alloy or silver glass mounted.

## recommended operating conditions

	SN55ALS192			SN75ALS192			UNIT	
	MIN	NOM	MAX	MIN	NOM	MAX		
Supply voltage, $V_{CC}$	4.5	5	5.5	4.75	5	5.25	V	
High level input voltage, $V_{IH}$	2			2			V	
Low-level input voltage, $V_{IL}$	0.8			0.8			V	
High-level output current, $I_{OH}$	– 20			– 20			mA	
Low-level output current, $I_{OL}$	20			20			mA	
Operating free-air temperature, $T_A$	– 55			0			70	°C



# SN55ALS192, SN75ALS192 QUAD DIFFERENTIAL LINE DRIVERS

SLLS007B – D2904, JULY 1985 – REVISED MARCH 1993

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	SN55ALS192			SN75ALS192			UNIT
		MIN	TYP‡	MAX	MIN	TYP‡	MAX	
V <sub>IK</sub>	Input clamp voltage	V <sub>CC</sub> = MIN, I <sub>I</sub> = -18 mA			-1.5			V
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN, I <sub>OH</sub> = -20 mA			2.4			V
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN, I <sub>OL</sub> = 20 mA			0.5			V
V <sub>O</sub>	Output voltage	V <sub>CC</sub> = MAX, I <sub>O</sub> = 0			0			V
V <sub>OD1</sub>	Differential output voltage	V <sub>CC</sub> = MIN, I <sub>O</sub> = 0			1.5			V
V <sub>OD2</sub>	Differential output voltage	R <sub>L</sub> = 100 Ω, See Figure 1			1/2 V <sub>OD1</sub> or 2§			V
Δ V <sub>OD</sub>	Change in magnitude of differential output voltage¶	R <sub>L</sub> = 100 Ω, See Figure 1			± 0.2			V
V <sub>OC</sub>	Common-mode output voltage#				± 3			V
Δ V <sub>OC</sub>	Change in magnitude of common-mode output voltage¶				± 0.2			V
I <sub>O</sub>	Output current with power off	V <sub>CC</sub> = 0		V <sub>O</sub> = 6 V		100		μA
				V <sub>O</sub> = -0.25 V		-100		
I <sub>OZ</sub>	Off-state (high-impedance state) output current	V <sub>CC</sub> = MAX		V <sub>O</sub> = 0.5 V		-20		μA
				V <sub>O</sub> = 2.5 V		20		
I <sub>I</sub>	Input current at maximum input voltage	V <sub>CC</sub> = MAX, V <sub>I</sub> = 7 V			100			μA
I <sub>IH</sub>	High-level input current	V <sub>CC</sub> = MAX, V <sub>I</sub> = 2.7 V			20			μA
I <sub>IL</sub>	Low-level input current	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0.4 V			-200			μA
I <sub>OS</sub>	Short-circuit output current	V <sub>CC</sub> = MAX			-30			mA
I <sub>CC</sub>	Supply current (all drivers)	V <sub>CC</sub> = MAX, All outputs disabled			26			mA

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

§ The minimum V<sub>OD2</sub> with a 100-Ω load is either 1/2 V<sub>OD1</sub> or 2 V, whichever is greater.

¶ |V<sub>OD</sub>| and |V<sub>OC</sub>| are the changes in magnitude of V<sub>OD</sub> and V<sub>OC</sub>, respectively, that occur when the input is changed from a high level to a low level.

# In EIA Standard RS-422A, V<sub>OC</sub>, which is the average of the two output voltages with respect to ground, is called output offset voltage, V<sub>OS</sub>.

|| Not more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

## switching characteristics, V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C (see Figure 2)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	S1 and S2 open, C <sub>L</sub> = 30 pF			6	13	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output				9	14	ns
	Output-to-output skew				3	6	ns
t <sub>PZH</sub>	Output enable time to high level	S1 open and S2 closed			11	15	ns
t <sub>PZL</sub>	Output enable time to low level	S1 closed and S2 open			16	20	ns
t <sub>PHZ</sub>	Output disable time from high level	S1 open and S2 closed, C <sub>L</sub> = 10 pF			8	15	ns
t <sub>PLZ</sub>	Output disable time from low level	S1 and S2 closed, C <sub>L</sub> = 10 pF			18	20	ns



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PARAMETER MEASUREMENT INFORMATION

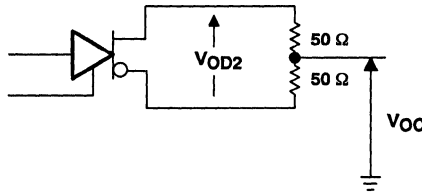
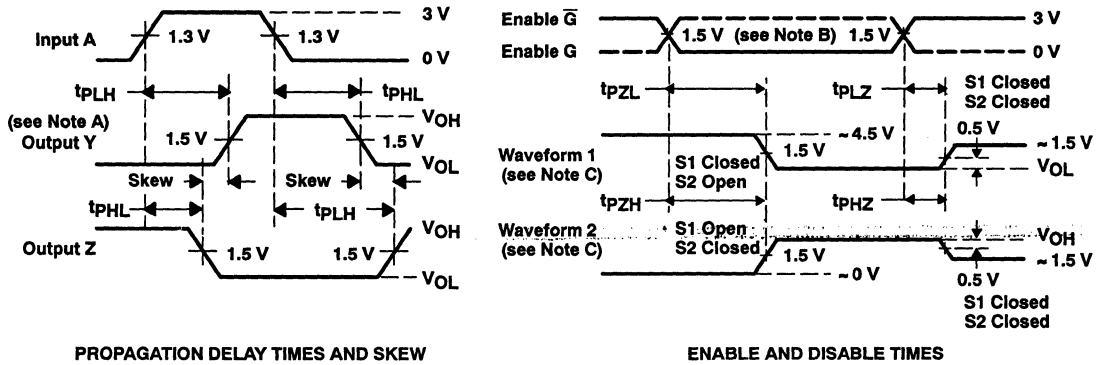


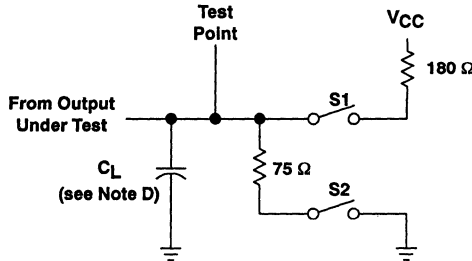
Figure 1. Differential and Common-Mode Output Voltages



PROPAGATION DELAY TIMES AND SKEW

ENABLE AND DISABLE TIMES

VOLTAGE WAVEFORMS



TEST CIRCUIT

- NOTES: A. When measuring propagation delay times and skew, switches S1 and S2 are open.  
 B. Each enable is tested separately.  
 C. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.  
 D.  $C_L$  includes probe and jig capacitance.  
 E. All input pulses are supplied by generators having the following characteristics:  $PRR \leq 1$  MHz,  $Z_O \sim 50 \Omega$ ,  $t_r \leq 15$  ns, and  $t_f \leq 6$  ns.

Figure 2. Test Circuit and Voltage Waveforms

# SN55ALS192, SN75ALS192 QUAD DIFFERENTIAL LINE DRIVERS

SLLS007B - D2904, JULY 1985 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS†

Y OUTPUT VOLTAGE  
vs  
DATA INPUT VOLTAGE

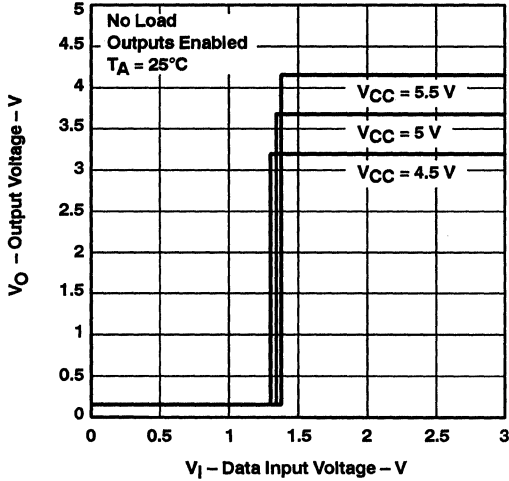


Figure 3

Y OUTPUT VOLTAGE  
vs  
DATA INPUT VOLTAGE

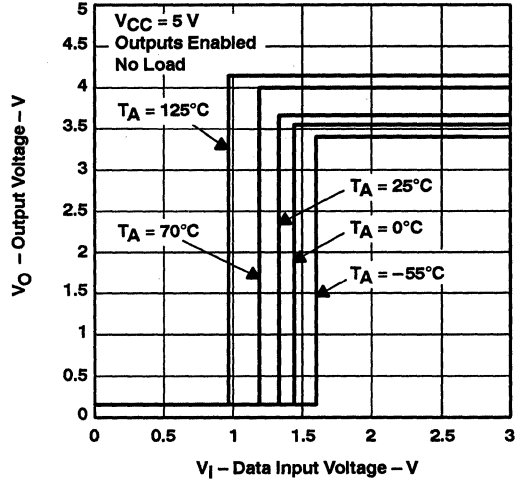


Figure 4

Y OUTPUT VOLTAGE  
vs  
ENABLE G INPUT VOLTAGE

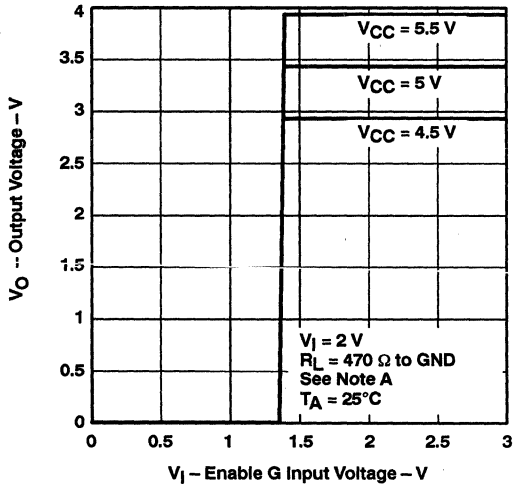


Figure 5

Y OUTPUT VOLTAGE  
vs  
ENABLE G INPUT VOLTAGE

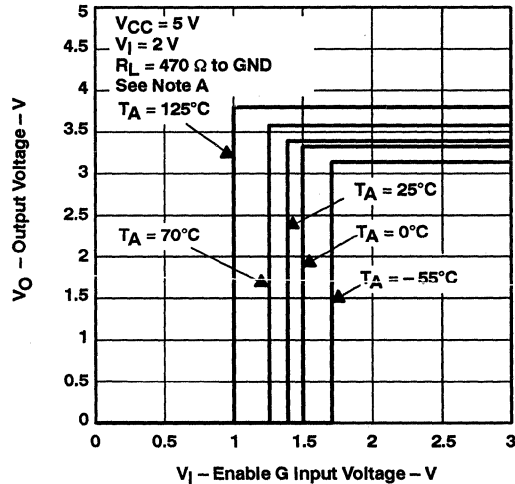


Figure 6

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  and below  $4.75\text{ V}$  and above  $5.25\text{ V}$ , are applicable to SN55ALS192 circuits only.  
NOTE A: The A input is connected to  $V_{CC}$  during the testing of the Y outputs and to ground during the testing of the Z outputs.

TYPICAL CHARACTERISTICS†

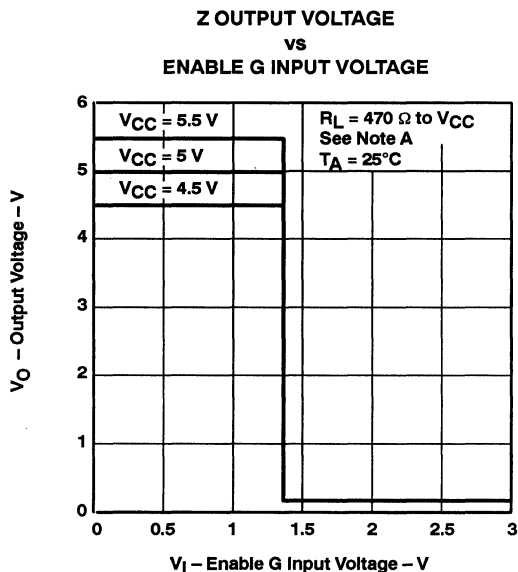


Figure 7

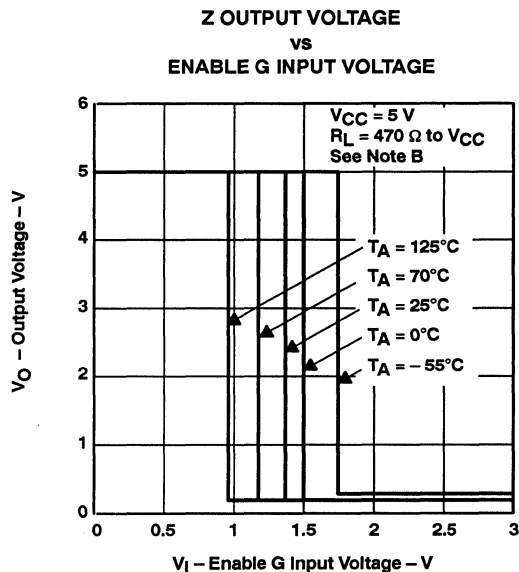


Figure 8

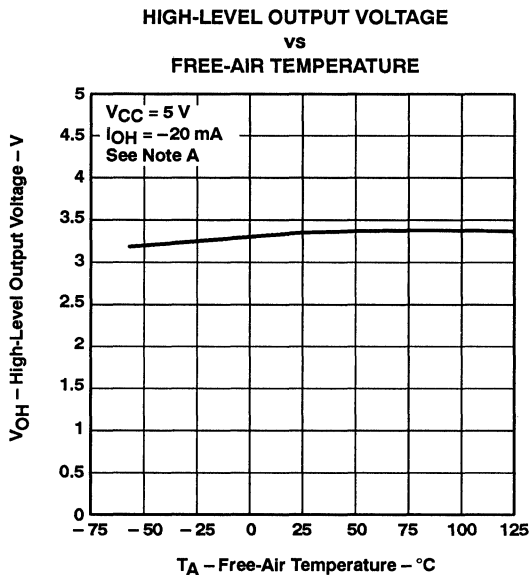


Figure 9

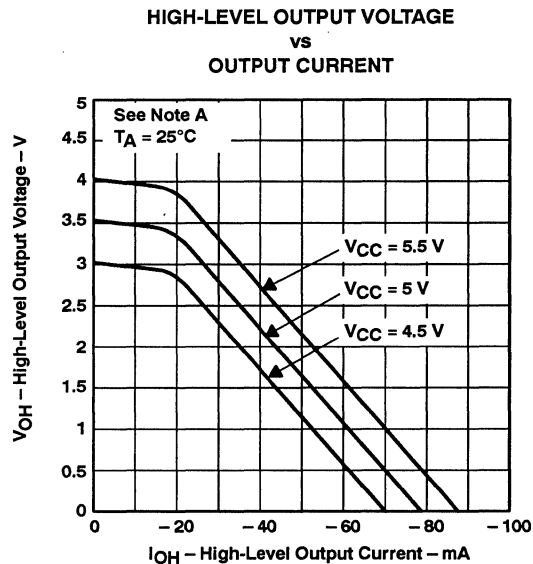


Figure 10

† Data for temperatures below 0°C and above 70°C, and below 4.75 V and above 5.25 V, are applicable to SN55ALS192 circuits only.  
 NOTES: A. The A input is connected to V<sub>CC</sub> during the testing of the Y outputs and to ground during the testing of the Z outputs.  
 B. The A input is connected to GND during the testing of the Y outputs and to V<sub>CC</sub> during the testing of the Z outputs.



# SN55ALS192, SN75ALS192 QUAD DIFFERENTIAL LINE DRIVERS

SLLS007B – D2904, JULY 1985 – REVISED MARCH 1993

## TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

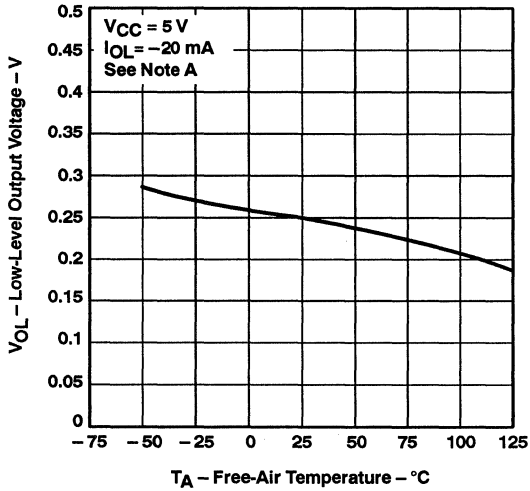


Figure 11

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-OUTPUT CURRENT

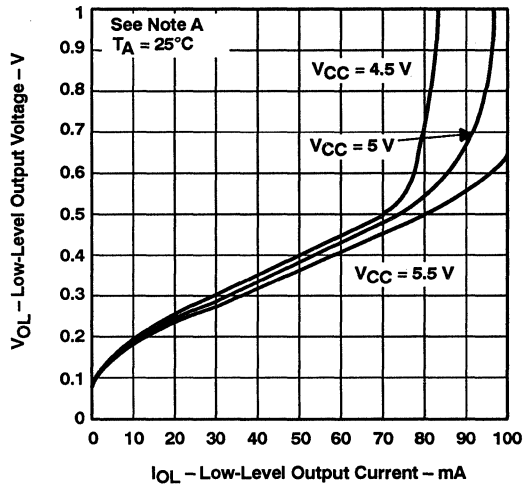


Figure 12

SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

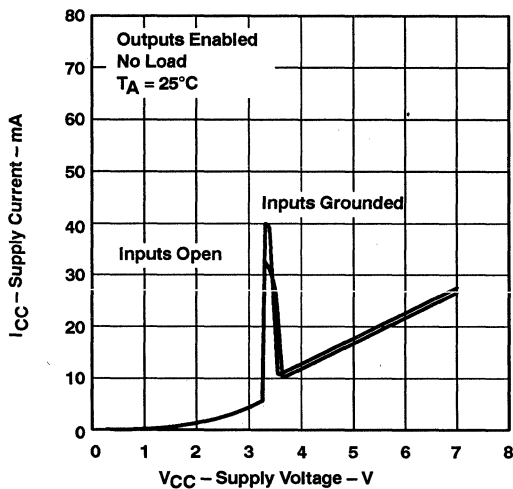


Figure 13

SUPPLY CURRENT  
vs  
SUPPLY CURRENT

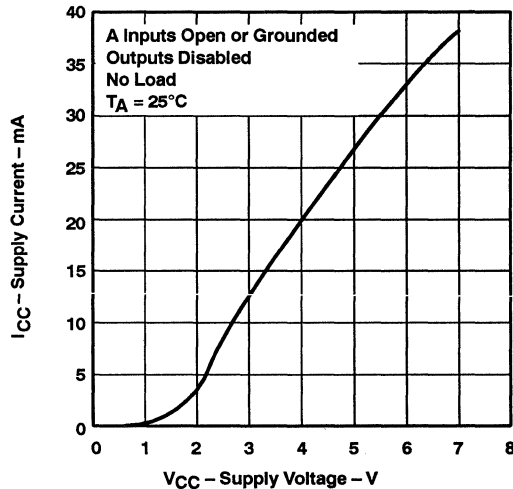


Figure 14

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$ , and below  $4.75\text{ V}$  and above  $5.25\text{ V}$ , are applicable to SN55ALS192 circuits only.  
NOTE A: The A input is connected to GND during the testing of the Y outputs and to  $V_{CC}$  during the testing of the Z outputs.

TYPICAL CHARACTERISTICS

SUPPLY CURRENT  
vs  
FREQUENCY

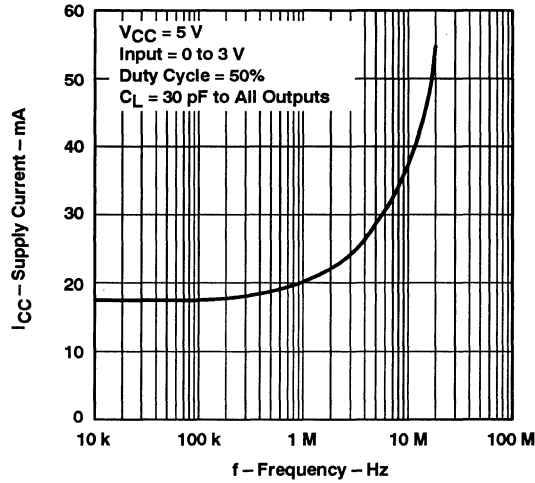


Figure 15

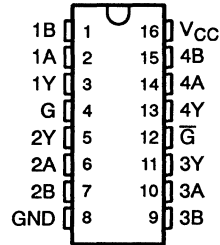


# SN75ALS193 QUAD DIFFERENTIAL LINE RECEIVER WITH 3-STATE OUTPUTS

SLLS008C – D2931, JUNE 1986 – REVISED MARCH 1993

- Meets EIA Standards RS-422-A and RS-423-A
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- Designed for Multipoint Bus Transmission on Long Bus Lines in Noisy Environments
- 3-State Outputs
- Common-Mode Input Voltage Range –7 V to 7 V
- Input Sensitivity . . .  $\pm 200$  mV
- Input Hysteresis . . . 120 mV Typ
- High Input Impedance . . . 12 k $\Omega$  Min
- Operates from Single 5-V Supply
- Low Supply Current Requirement 35 mA Max
- Improved Speed and Power Consumption Compared to AM26LS32A

SN75ALS193 . . . J PACKAGE†  
(TOP VIEW)



† For surface-mount package, see the SN75ALS197.

## description

The SN75ALS193 is a monolithic quad line receiver with 3-state outputs designed using advanced low-power Schottky technology. This technology provides combined improvements in bar design, tooling production, and wafer fabrication. This, in turn, provides significantly less power requirements and permits much higher data throughput than other designs. These devices meet the specifications of EIA Standards RS-422-A and RS-423-A. It features 3-state outputs that permit direct connection to a bus-organized system with a fail-safe design that ensures the outputs will always be high if the inputs are open.

The device is optimized for balanced multipoint bus transmission at rates up to 20 megabits per second. The input features high Input impedance, input hysteresis for increased noise immunity, and an input sensitivity of  $\pm 200$  mV over a common-mode input voltage range of –7 to 7 V. It also features active-high and active-low enable functions that are common to the four channels. The SN75ALS193 is designed for optimum performance when used with the ALS192 quad differential line driver.

The SN75ALS193 is characterized for operation from 0°C to 70°C.

FUNCTION TABLE  
(each receiver)

DIFFERENTIAL INPUTS A – B	ENABLES G $\bar{G}$		OUTPUT Y
$V_{ID} \geq 0.2$ V	H X	X L	H H
$-0.2$ V < $V_{ID}$ < 0.2 V	H X	X L	? ?
$V_{ID} \leq -0.2$ V	H X	X L	L L
X	L	H	Z
Open	H X	X L	H H

H = high level, L = low level, X = irrelevant, ? = indeterminate  
Z = high impedance (off)

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**TEXAS  
INSTRUMENTS**

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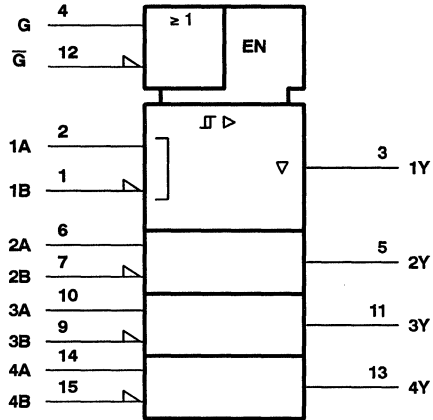
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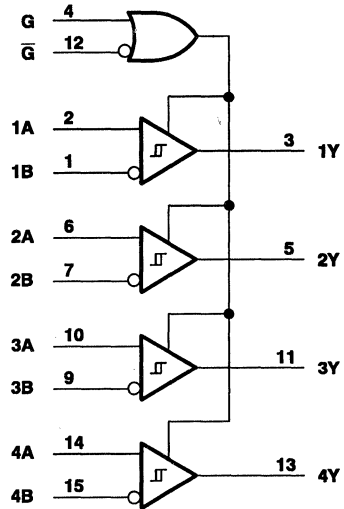
**SN75ALS193**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS008C - D2931, JUNE 1986 - REVISED MARCH 1993

**logic symbol†**

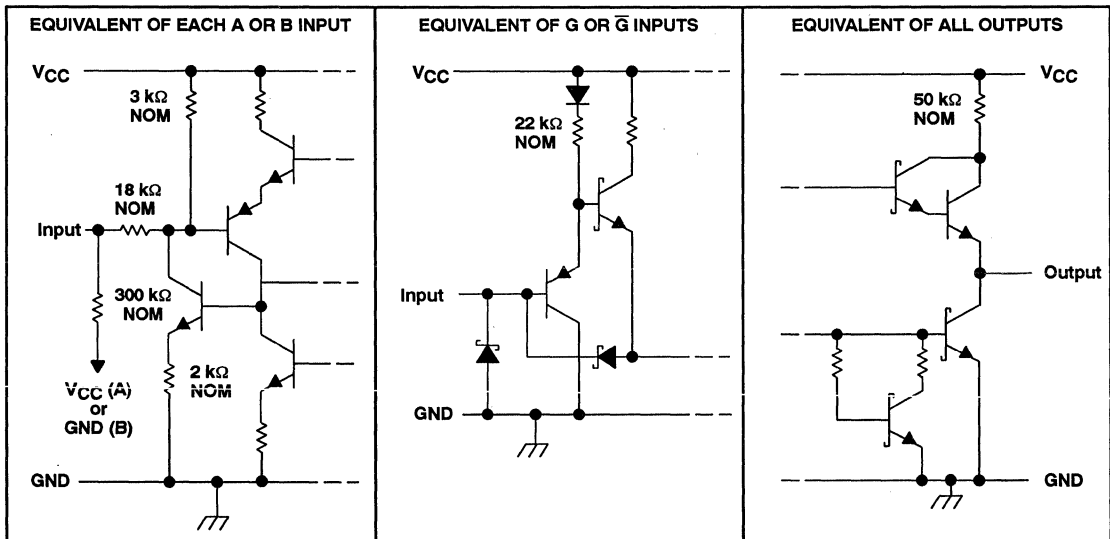


**logic diagram (positive logic)**



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

**schematics of inputs and outputs**



**SN75ALS193**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS008C - D2931, JUNE 1986 - REVISED MARCH 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, A or B, $V_I$	$\pm 15$ V
Differential input voltage (see Note 2)	$\pm 15$ V
Enable input voltage	7 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	- 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	300°C

† Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the, corresponding inverting input.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
J	1025 mW	8.2 mW/°C	656 mW	N/A

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Differential input voltage, $V_{ID}$			$\pm 12$	V
High-level input voltage, $V_{IH}$		2		V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			- 400	$\mu\text{A}$
Low-level output current, $I_{OL}$			16	mA
Operating free-air temperature, $T_A$	0		70	°C



**SN75ALS193**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS008C - D2931, JUNE 1986 - REVISED MARCH 1993

**electrical characteristics over recommended range of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage					200	mV
$V_{T-}$	Negative-going threshold voltage			-200§			mV
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )				120		mV
$V_{IK}$	Enable-input clamp voltage	$V_{CC} = \text{MIN}$ ,	$I_I = -18 \text{ mA}$			-1.5	V
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ , $I_{OH} = -400 \mu\text{A}$ ,	$V_{ID} = 200 \text{ mV}$ , See Figure 1	2.5	3.6		V
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ , $V_{ID} = -200 \text{ mV}$ , See Figure 1	$I_{OL} = 8 \text{ mA}$			0.45	V
			$I_{OL} = 16 \text{ mA}$			0.5	
$I_{OZ}$	High-impedance-state output current	$V_{CC} = \text{MAX}$	$V_O = 2.4 \text{ V}$			20	$\mu\text{A}$
			$V_O = 0.4 \text{ V}$			-20	
$I_I$	Line input current	Other input at 0 V, See Note 3	$V_{CC} = \text{MIN}$ , $V_I = 15 \text{ V}$		0.7	1.2	mA
			$V_{CC} = \text{MIN}$ , $V_I = -15 \text{ V}$		-1.0	-1.7	
$I_{IH}$	High-level enable-input current	$V_{CC} = \text{MAX}$	$V_{IH} = 2.7 \text{ V}$			20	$\mu\text{A}$
			$V_{IH} = \text{MAX}$			100	
$I_{IL}$	Low-level enable-input current	$V_{CC} = \text{MAX}$ ,	$V_{IL} = 0.4 \text{ V}$			-100	$\mu\text{A}$
	Input resistance			12	18		k $\Omega$
$I_{OS}$	Short-circuit output current	$V_{CC} = \text{MAX}$ , $V_O = 0$ ,	$V_{ID} = 3 \text{ V}$ , See Note 4	-15	-78	-130	mA
$I_{CC}$	Supply current	$V_{CC} = \text{MAX}$ ,	Outputs disabled		22	35	mA

† For conditions shown as MIN or MAX, use the appropriate values specified under recommended operating conditions.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

§ The algebraic convention, in which the less positive limit is designated minimum, is used in this data sheet for threshold voltage levels only.

NOTES: 3. Refer to EIA Standard RS-422-A and RS-423-A for exact conditions.

4. Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

**switching characteristics,  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -2.5 \text{ V to } 2.5 \text{ V}$ , $C_L = 15 \text{ pF}$	See Figure 2		15	22	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output				15	22	
$t_{PZH}$	Output enable time to high level	$C_L = 15 \text{ pF}$	See Figure 3		13	25	
$t_{PZL}$	Output enable time to low level				11	25	
$t_{PHZ}$	Output disable time from high level	$C_L = 5 \text{ pF}$	See Figure 3		13	25	
$t_{PLZ}$	Output disable time from low level				15	22	

**SN75ALS193**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS008C - D2931, JUNE 1986 - REVISED MARCH 1993

**PARAMETER MEASUREMENT INFORMATION**

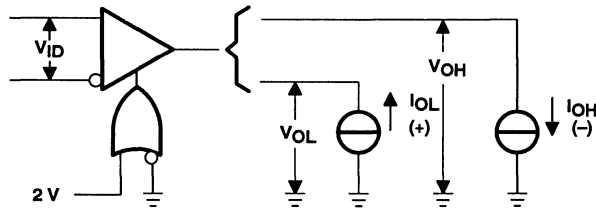
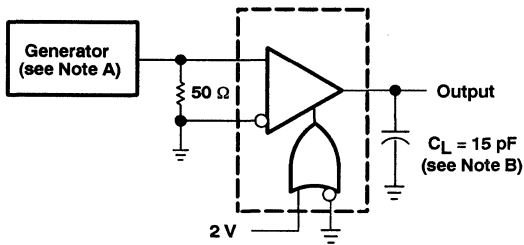
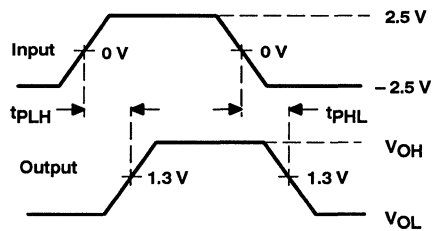


Figure 1.  $V_{OH}$ ,  $V_{OL}$



**TEST CIRCUIT**



**VOLTAGE WAVEFORMS**

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1 \text{ MHz}$ , duty cycle  $\leq 50\%$ ,  $Z_O = 50 \Omega$ ,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ .

B.  $C_L$  includes probe and jig capacitance.

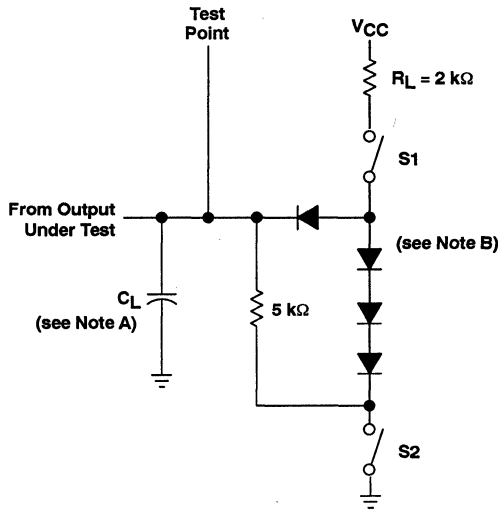
Figure 2. Test Circuit and Voltage Waveforms



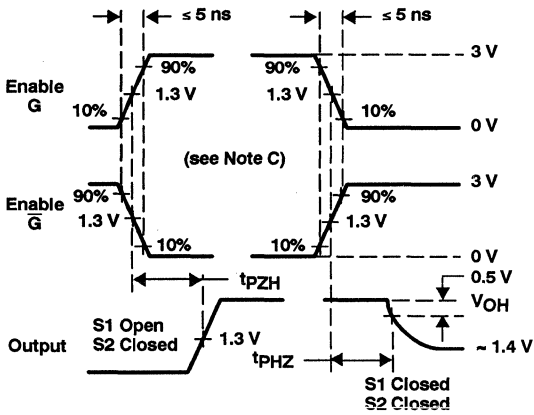
**SN75ALS193**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

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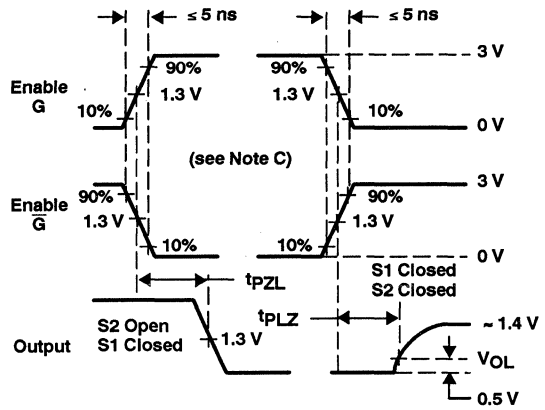
**PARAMETER MEASUREMENT INFORMATION**



**LOAD CIRCUIT**



**VOLTAGE WAVEFORMS FOR  $t_{pHZ}$ ,  $t_{pZH}$**



**VOLTAGE WAVEFORMS FOR  $t_{pLZ}$ ,  $t_{pZL}$**

- NOTES: A.  $C_L$  includes probe and jig capacitance.  
 B. All diodes are 1N3064 or equivalent.  
 C. Enable  $\bar{G}$  is tested with  $\bar{G}$  high;  $\bar{G}$  is tested with  $\bar{G}$  low.

**Figure 3. Load Circuit and Voltage Waveforms**



TYPICAL CHARACTERISTICS

OUTPUT VOLTAGE  
 vs  
 ENABLE VOLTAGE

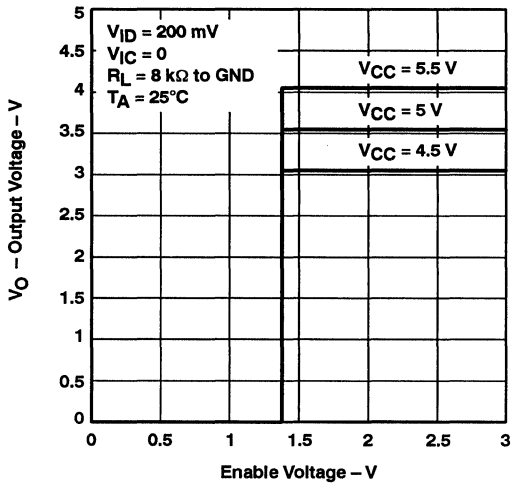


Figure 4

OUTPUT VOLTAGE  
 vs  
 ENABLE VOLTAGE

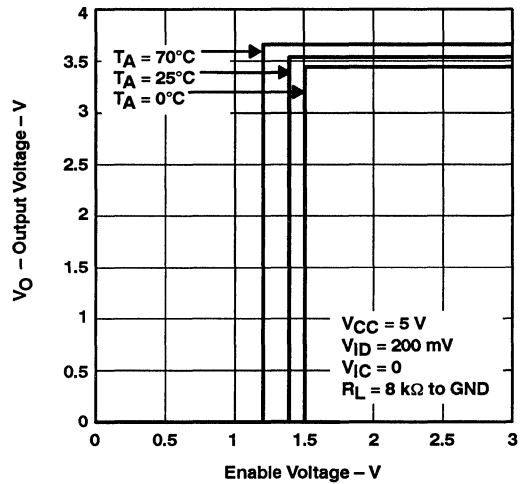


Figure 5

OUTPUT VOLTAGE  
 vs  
 ENABLE VOLTAGE

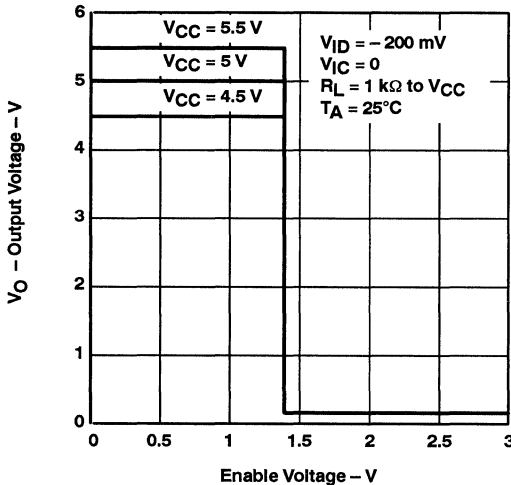


Figure 6

OUTPUT VOLTAGE  
 vs  
 ENABLE VOLTAGE

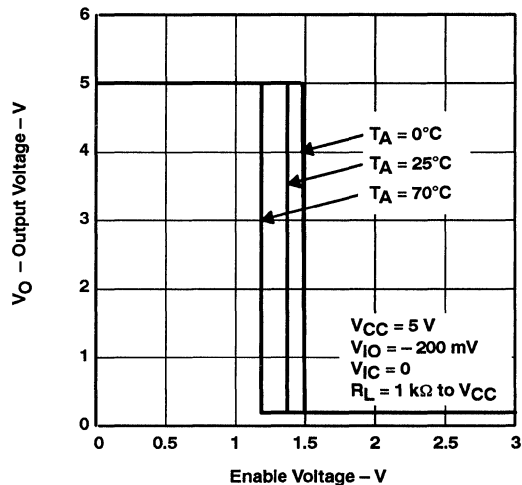
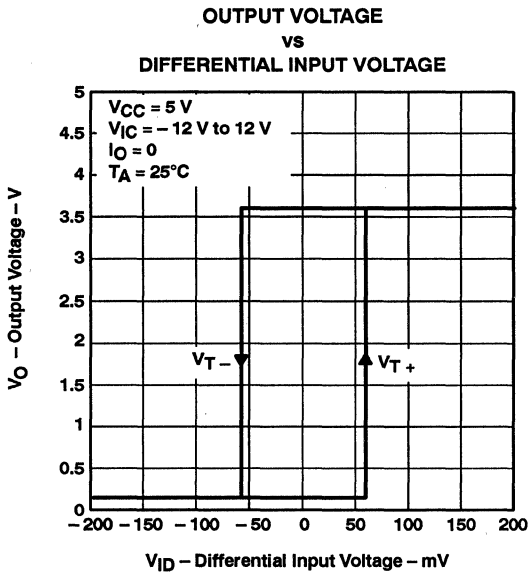


Figure 7

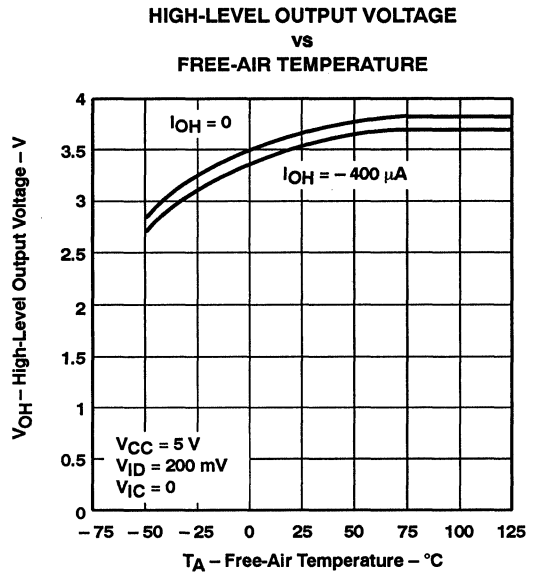
**SN75ALS193**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS008C - D2931, JUNE 1986 - REVISED MARCH 1993

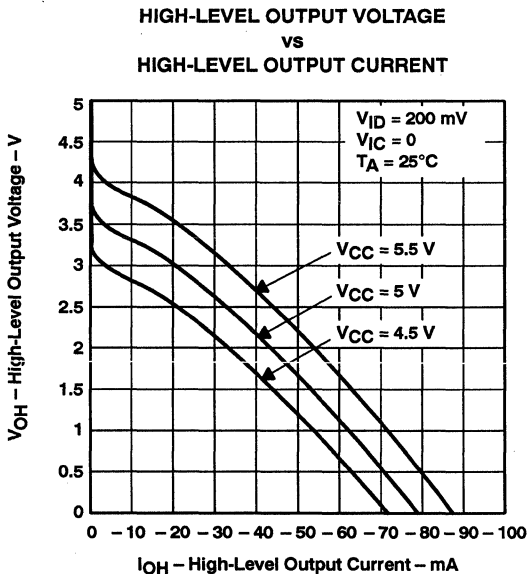
**TYPICAL CHARACTERISTICS**



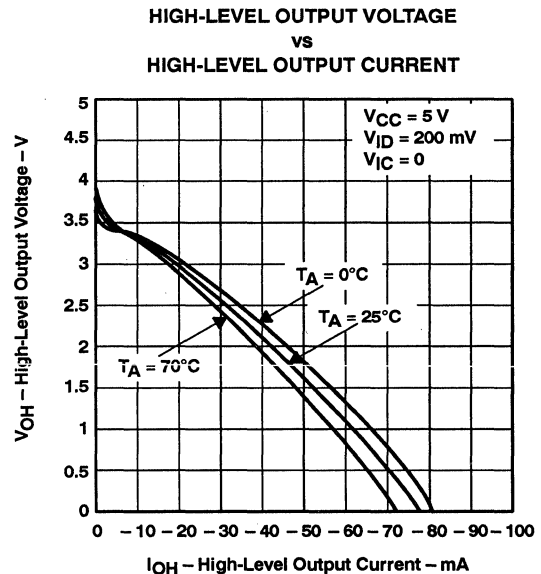
**Figure 8**



**Figure 9**



**Figure 10**



**Figure 11**



TYPICAL CHARACTERISTICS

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

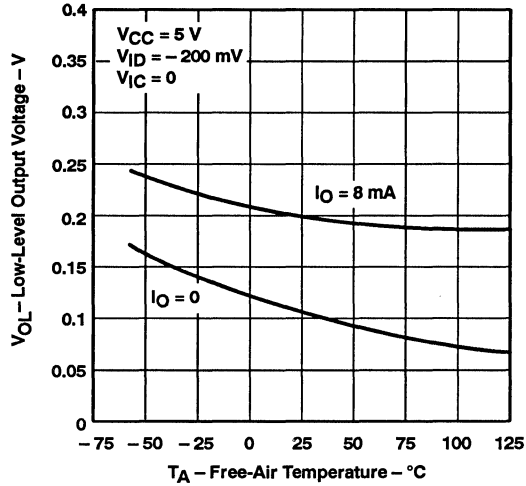


Figure 12

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

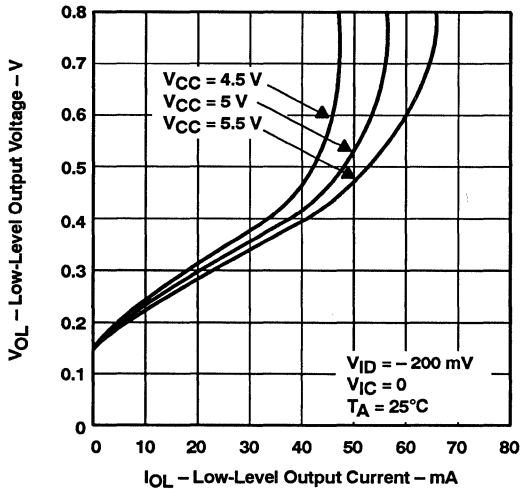


Figure 13

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

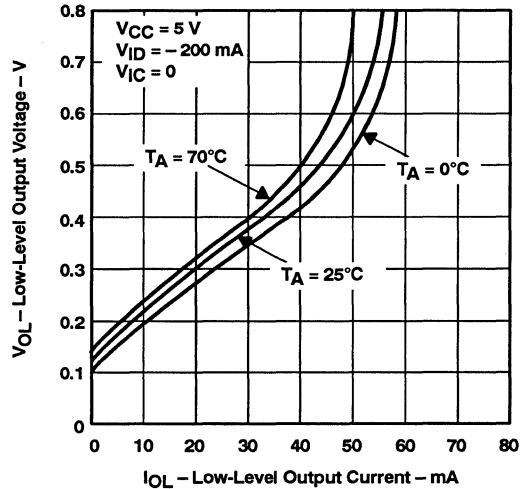


Figure 14

**SN75ALS193**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS008C - D2931, JUNE 1986 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

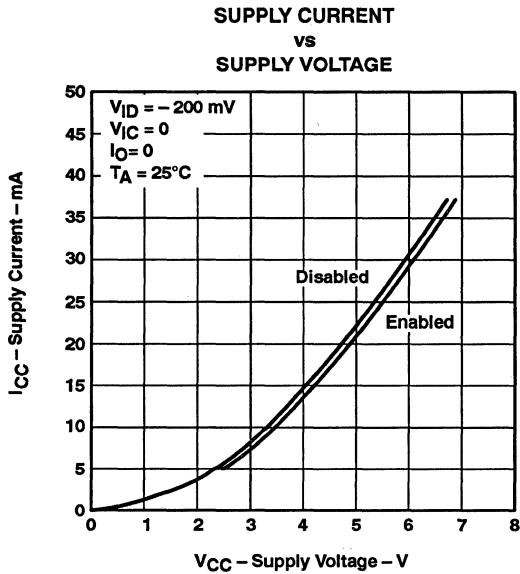


Figure 15

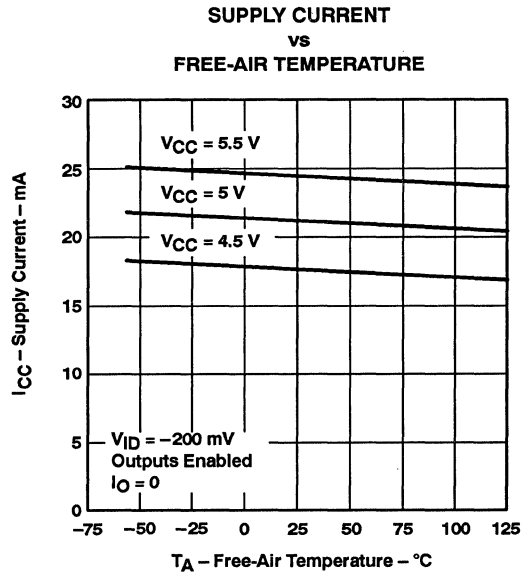


Figure 16

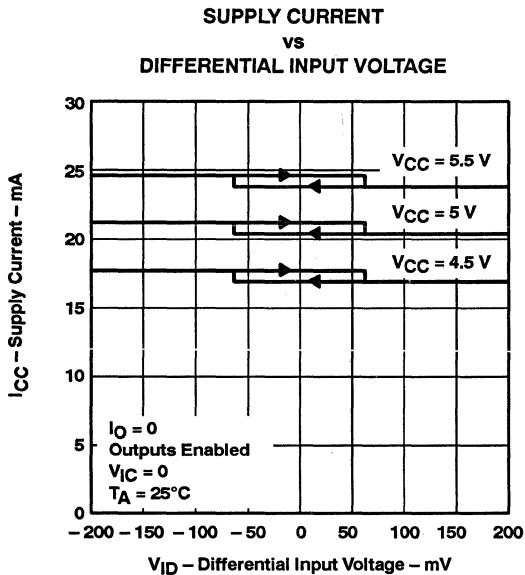


Figure 17

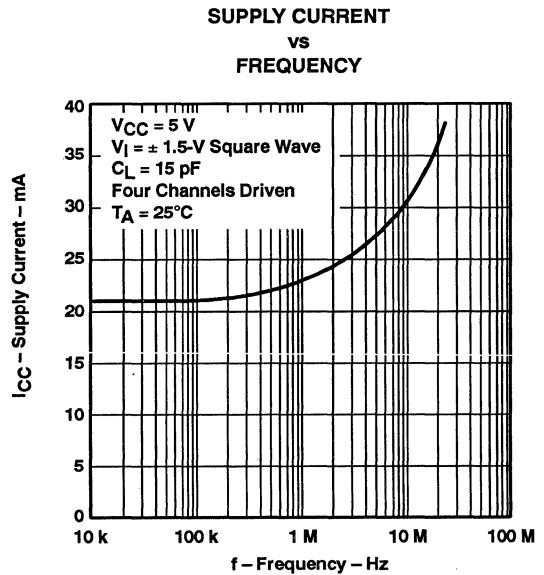


Figure 18

TYPICAL CHARACTERISTICS

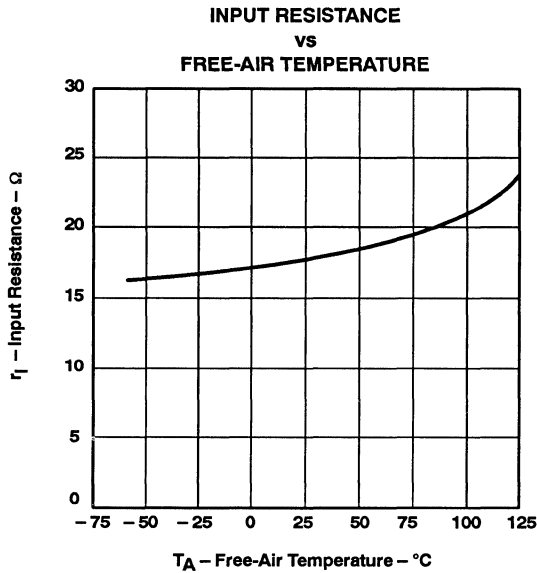


Figure 19

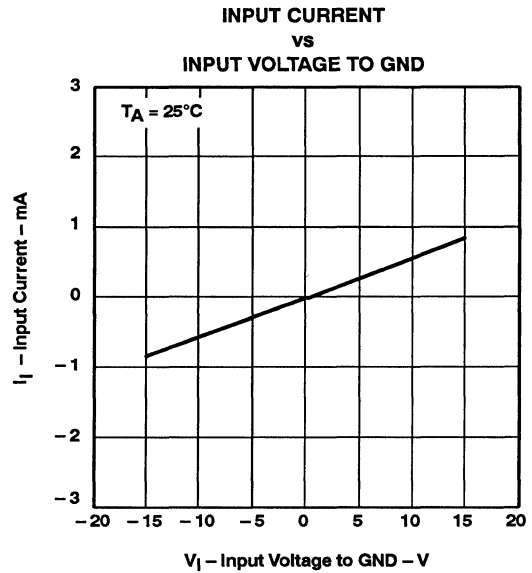


Figure 20

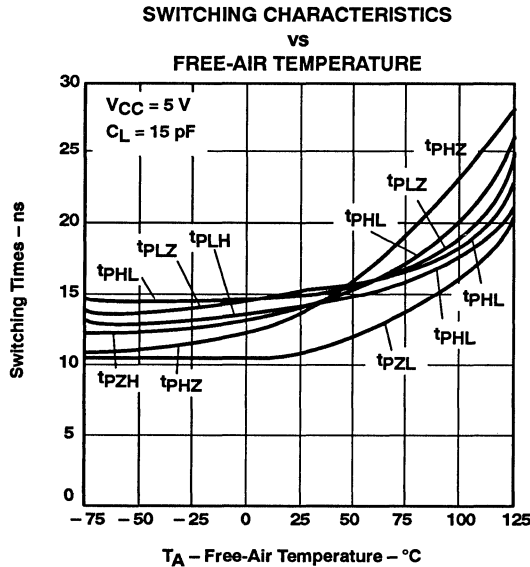


Figure 21

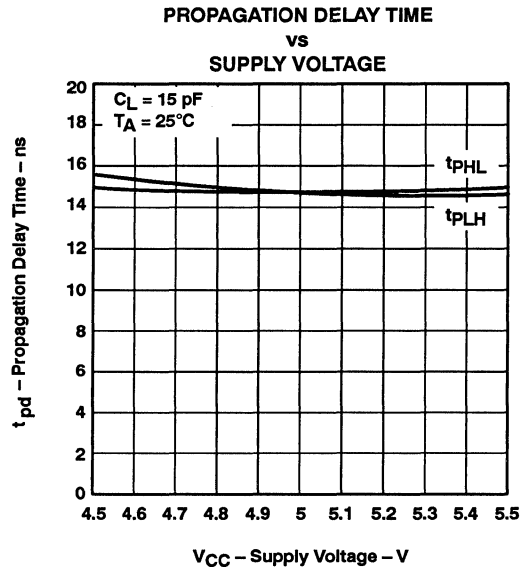


Figure 22

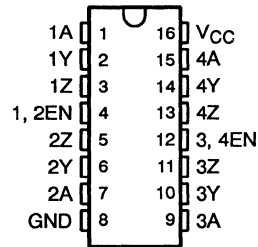


# SN55ALS194, SN75ALS194 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

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- Meets EIA Standard RS-422-A
- High-Speed ALS Design
- 3-State TTL-Compatible Outputs
- Single 5-V Supply Operation
- High Output Impedance in Power-Off Condition
- Two Pairs of Drivers Independently Enabled
- Designed as a Replacement for the MC3487 With Improvements:  $I_{CC}$  50% Lower, Switching Speed 30% Faster, Full-Temperature-Range Version

SN55ALS194 . . . J OR W PACKAGE  
SN75ALS194 . . . D OR N PACKAGE  
(TOP VIEW)



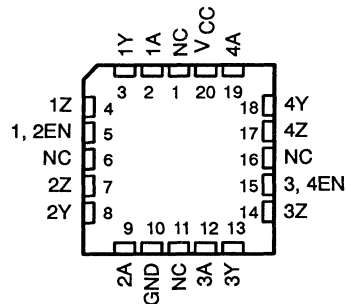
## description

These quad differential line drivers are designed for data transmission over twisted-pair or parallel-wire transmission lines. They meet the requirements of EIA Standard RS-422-A and are compatible with 3-state TTL circuits. Advanced low-power Schottky technology provides high speed without the usual power penalty. Standby supply current is typically only 26 mA, while typical propagation delay time is less than 10 ns and enable/disable times are typically less than 16 ns.

High-impedance inputs keep input currents low, less than 1  $\mu$ A for a high level and less than 100  $\mu$ A for a low level. The driver circuits can be enabled in pairs by separate active-high enable inputs. The SN55ALS194 and SN75ALS194 are capable of data rates in excess of 10 megabits per second and are designed to operate with the SN55ALS195 and SN75ALS195 quadruple line receivers.

The SN55ALS194 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The SN75ALS194 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

SN55ALS194 . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

FUNCTION TABLE  
(each driver)

INPUT A	OUTPUT EN	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	Z	Z

H = high level, L = low level, X = irrelevant,  
Z = high impedance

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
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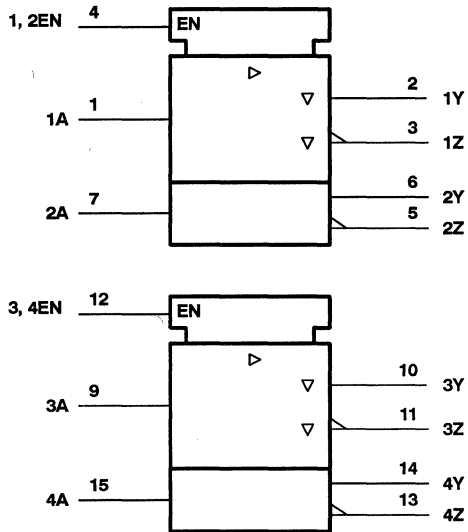
2-795



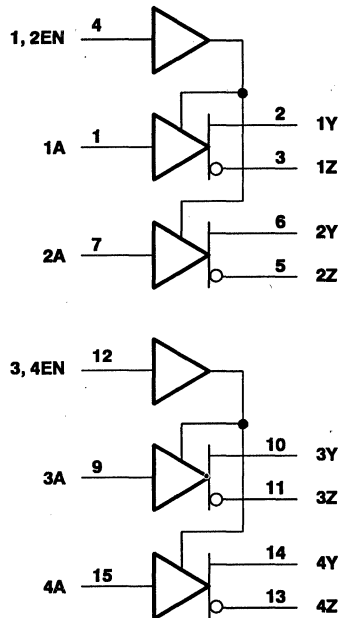
# SN55ALS194, SN75ALS194 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS009C - D2917, OCTOBER 1985 - REVISED MARCH 1993

## logic symbol†



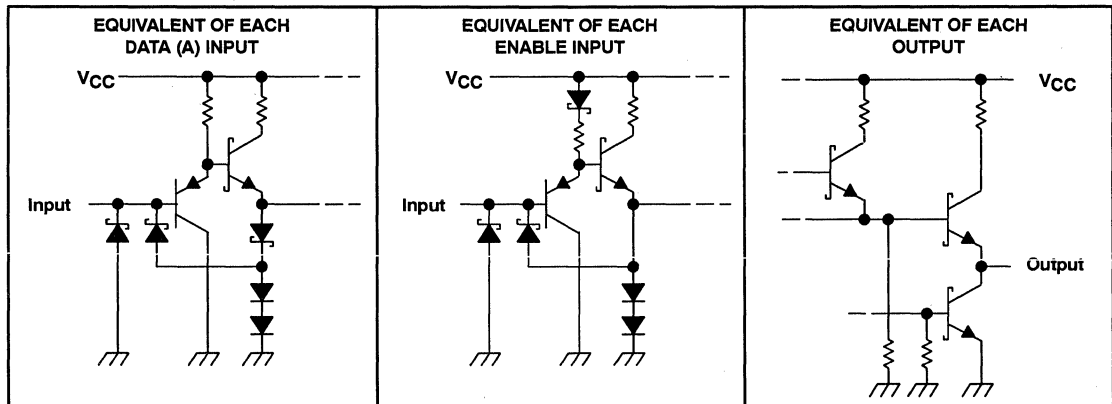
## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the D, J, N, and W packages.

## schematics of inputs and outputs



# SN55ALS194, SN75ALS194 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS009C – D2917, OCTOBER 1985 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, $V_I$	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN55ALS194	– 55°C to 125°C
SN75ALS194	0°C to 70°C
Storage temperature range	– 65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or W package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

NOTE 1: All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	N/A
W	1000 mW	8.0 mW/°C	640 mW	200 mW

## recommended operating conditions†

		SN55ALS194			SN75ALS194			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$		4.5	5	5.5	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	All inputs, $T_A = 25^\circ\text{C}$	2			2			V
	A inputs, $T_A = \text{Full range}$	2			2			
	EN inputs, $T_A = \text{Full range}$	2.1			2			
Low-level input voltage, $V_{IL}$		0.8			0.8			V
High-level output current, $I_{OH}$		– 20			– 20			mA
Low-level output current, $I_{OL}$	$T_A = 25^\circ\text{C}$	48			48			mA
	$T_A = \text{Full range}$	20			48			
Operating free-air temperature, $T_A$		– 55	125		0	70		°C

† Full range is  $T_A = -55^\circ\text{C}$  to  $125^\circ\text{C}$  for SN55ALS194 and  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$  for SN75ALS194.



# SN55ALS194, SN75ALS194 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS009C - D2917, OCTOBER 1985 - REVISED MARCH 1993

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{IK}$	Input clamp voltage	$V_{CC} = \text{MIN}$ ,	$I_I = -18 \text{ mA}$			-1.5	V
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ , $I_{OH} = -20 \text{ mA}$	SN55ALS194	2.4			V
			SN75ALS194	2.5			
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ ,	$I_{OL} = \text{MAX}$			0.5	V
$V_O$	Output voltage	$I_O = 0$		0		6	V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5		6	V
$ V_{OD2} $	Differential output voltage	$R_L = 100 \Omega$ , See Figure 1		$1/2 V_{OD1}$ or $2^S$			V
$\Delta V_{OD} $	Change in magnitude of differential output voltage¶					$\pm 0.4$	V
$V_{OC}$	Common-mode output voltage					$\pm 3$	V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage¶					$\pm 0.4$	V
$I_O$	Output current with power off	$V_{CC} = 0$	$V_O = 6 \text{ V}$			100	$\mu\text{A}$
			$V_O = -0.25 \text{ V}$			-100	
$I_{OZ}$	High-impedance state output current	$V_{CC} = \text{MAX}$ , Output enables at 0.8 V	$V_O = 2.7 \text{ V}$			100	$\mu\text{A}$
			$V_O = 0.5 \text{ V}$			-100	
$I_I$	Input current at maximum input voltage	$V_{CC} = \text{MAX}$ ,	$V_I = 5.5 \text{ V}$			100	$\mu\text{A}$
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$ ,	$V_I = 2.7 \text{ V}$			50	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$ ,	$V_I = 0.5 \text{ V}$			-200	$\mu\text{A}$
$I_{OS}$	Short-circuit output current#	$V_{CC} = \text{MAX}$ ,	$V_I = 2 \text{ V}$	-40		-140	mA
$I_{CC}$	Supply current (all drivers)	$V_{CC} = \text{MAX}$ ,	All outputs disabled		26	45	mA

† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

§ The minimum  $V_{OD2}$  with a  $100\text{-}\Omega$  load is either  $1/2 V_{OD1}$  or  $2 \text{ V}$ , whichever is greater.

¶  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

# Not more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	SN55ALS194			SN75ALS194			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$t_{PLH}$	Propagation delay time, low-to-high-level output	$C_L = 15 \text{ pF}$ , See Figure 2	6	13	6	13	ns	
$t_{PHL}$	Propagation delay time, high-to-low-level output		9	14	9	14	ns	
	Output-to-output skew		3.5	6	3.5	6	ns	
$t_D$	Differential-output transition time	$C_L = 15 \text{ pF}$ , See Figure 3	8	14	8	14	ns	
$t_{PZH}$	Output enable time to high level	$C_L = 15 \text{ pF}$ , See Figure 4	9	12	9	12	ns	
$t_{PZL}$	Output enable time to low level		12	20	12	20	ns	
$t_{PHZ}$	Output disable time from high level		9	15	9	14	ns	
$t_{PLZ}$	Output disable time from low level		12	15	12	15	ns	

# SN55ALS194, SN75ALS194 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS009C - D2917, OCTOBER 1985 - REVISED MARCH 1993

## SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A
$V_O$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_o$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$
$\Delta  V_{OD} $	$  V_{t1}  -  V_{t2}  $
$V_{OC}$	$ V_{Os} $
$\Delta  V_{OC} $	$ V_{Os} - \bar{V}_{Os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $
$I_O$	$ I_{xa} ,  I_{xb} $

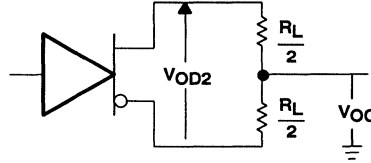


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

## PARAMETER MEASUREMENT INFORMATION

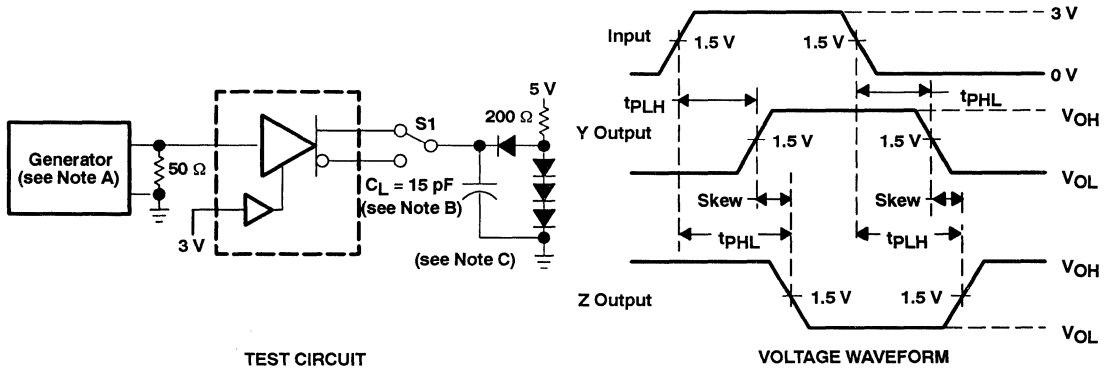


Figure 2. Test Circuit and Voltage Waveform

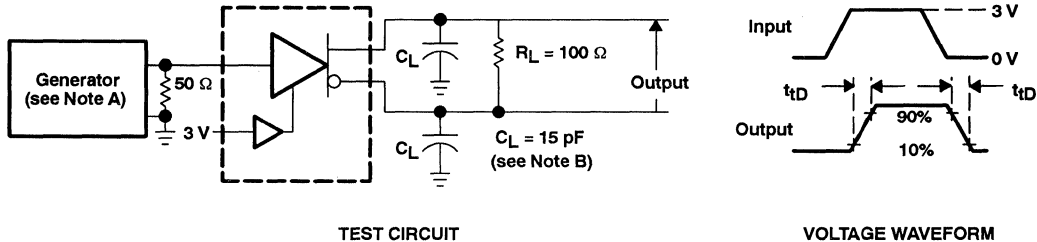


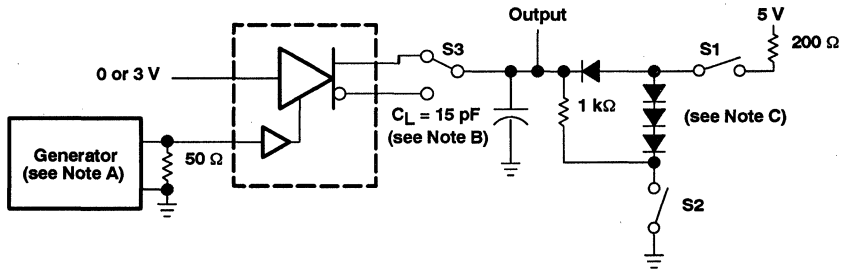
Figure 3. Differential-Output Test Circuit and Voltage Waveform

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5$  ns,  $t_f \leq 5$  ns, PRR  $\leq 1$  MHz, duty cycle  $\leq 50\%$ ,  $Z_O \sim 50 \Omega$ .  
 B.  $C_L$  includes probe and stray capacitance.  
 C. All diodes are 1N916 or 1N3064.

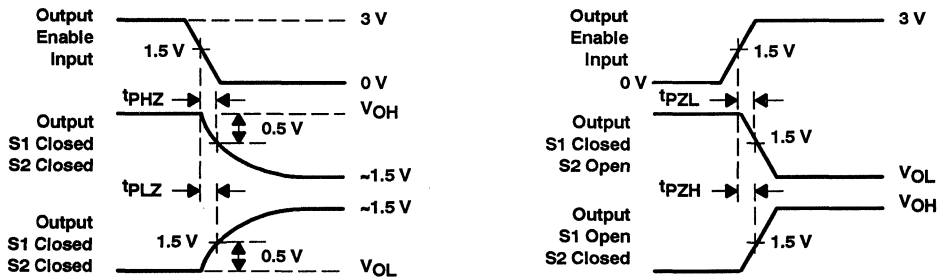
**SN55ALS194, SN75ALS194**  
**QUAD DIFFERENTIAL LINE DRIVERS**  
**WITH 3-STATE OUTPUTS**

SLLS009C - D2917, OCTOBER 1985 - REVISED MARCH 1993

**PARAMETER MEASUREMENT INFORMATION**



**TEST CIRCUIT**



**VOLTAGE WAVEFORMS**

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5$  ns,  $t_f \leq 5$  ns, PRR  $\leq 1$  MHz, duty cycle  $\leq 50\%$ ,  $Z_0 \sim 50 \Omega$ .  
 B.  $C_L$  includes probe and stray capacitance.  
 C. All diodes are 1N916 or 1N3064.

**Figure 4. Driver Test Circuit and Voltage Waveforms**

SN55ALS194, SN75ALS194  
 QUAD DIFFERENTIAL LINE DRIVERS  
 WITH 3-STATE OUTPUTS

SLLS009C – D2917, OCTOBER 1985 – REVISED MARCH 1993

TYPICAL CHARACTERISTICS†

Y OUTPUT VOLTAGE  
 VS  
 DATA INPUT VOLTAGE

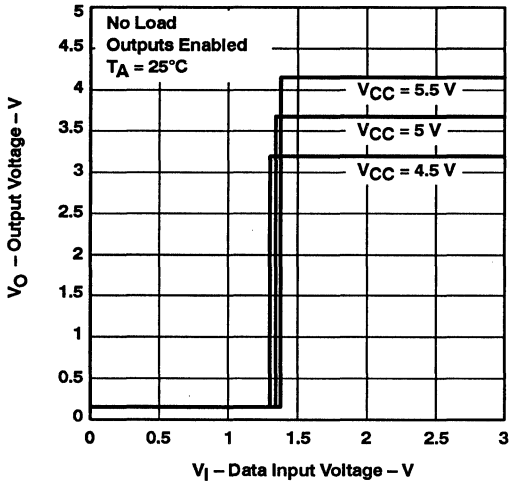


Figure 5

Y OUTPUT VOLTAGE  
 VS  
 DATA INPUT VOLTAGE

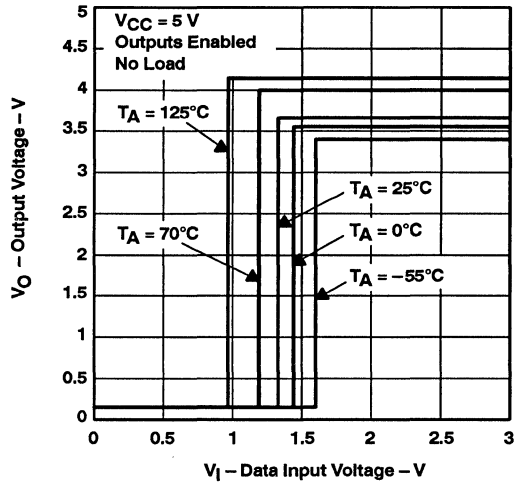


Figure 6

Y OUTPUT VOLTAGE  
 VS  
 ENABLE G INPUT VOLTAGE

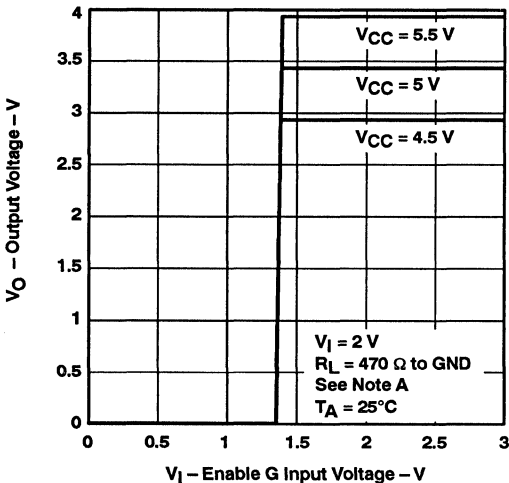


Figure 7

Y OUTPUT VOLTAGE  
 VS  
 ENABLE G INPUT VOLTAGE

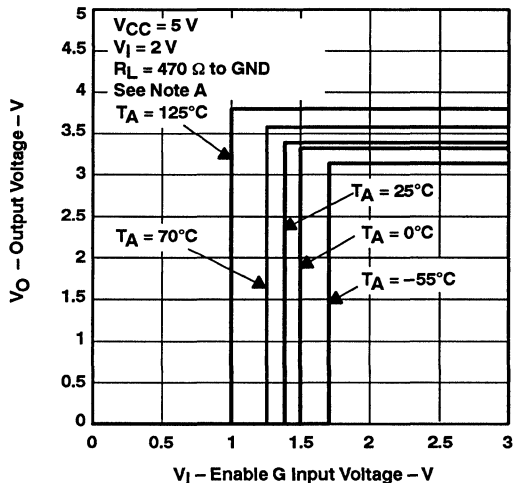


Figure 8

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  are applicable to SN55ALS194 circuits only.

NOTE A: The A input is connected to  $V_{CC}$  during the testing of the Y outputs and to GND during the testing of the Z outputs.

# SN55ALS194, SN75ALS194 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

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## TYPICAL CHARACTERISTICS†

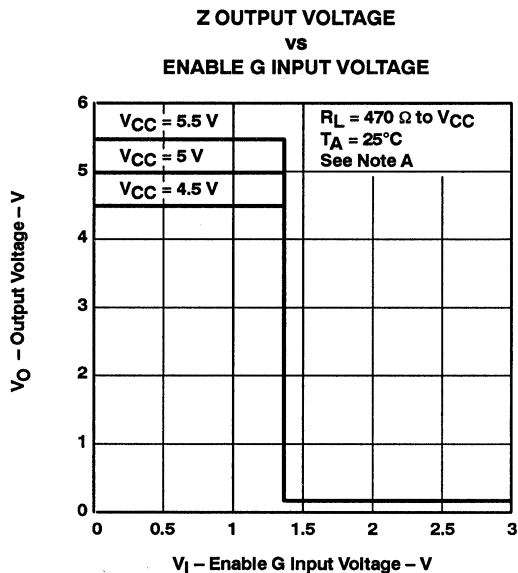


Figure 9

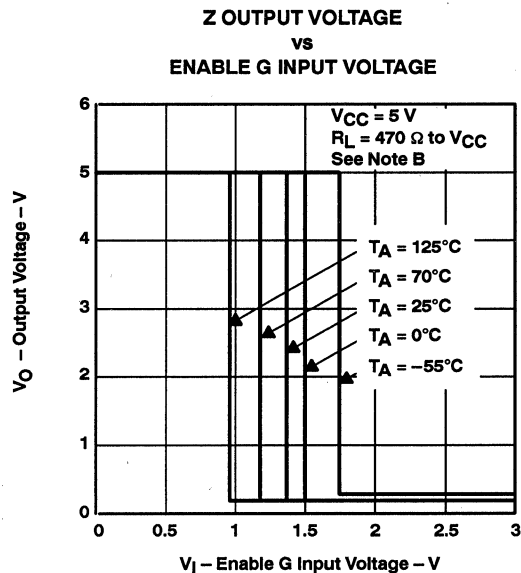


Figure 10

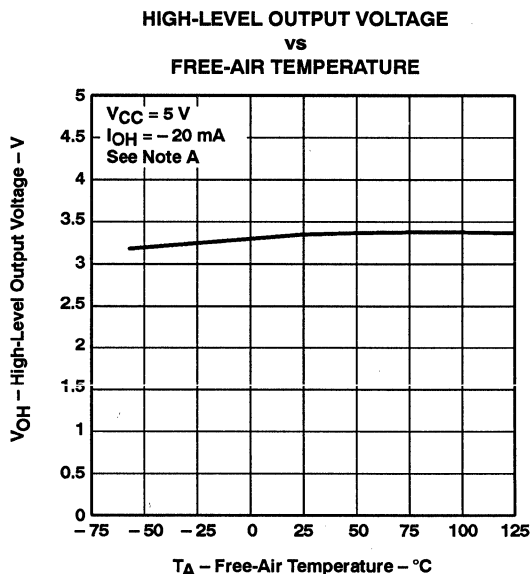


Figure 11

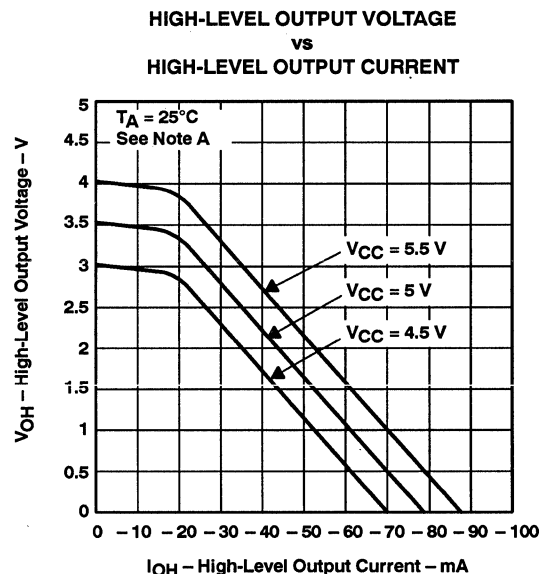


Figure 12

† Data for temperatures below 0°C and above 70°C are applicable to the SN55ALS194 circuits only.

NOTES: A. The A input is connected to V<sub>CC</sub> during the testing of the Y outputs and to GND during the testing of the Z outputs.

B. The A input is connected to ground during the testing of the Y outputs and to V<sub>CC</sub> during the testing of the Z outputs.

# SN55ALS194, SN75ALS194 QUAD DIFFERENTIAL LINE DRIVERS WITH 3-STATE OUTPUTS

SLLS009C - D2917, OCTOBER 1985 - REVISED MARCH 1983

## TYPICAL CHARACTERISTICS

LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

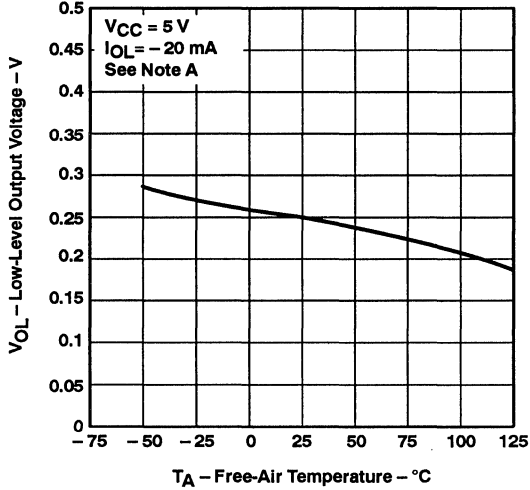


Figure 13

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

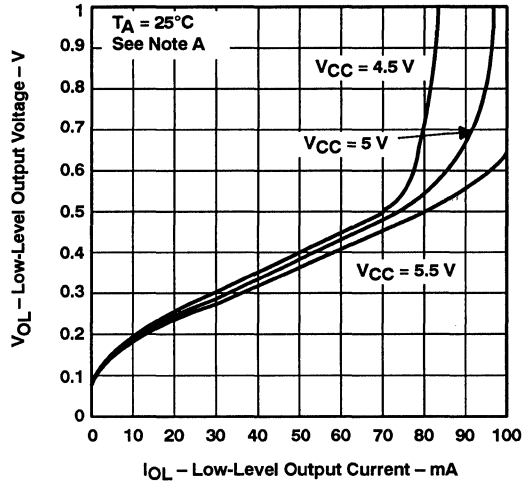


Figure 14

SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

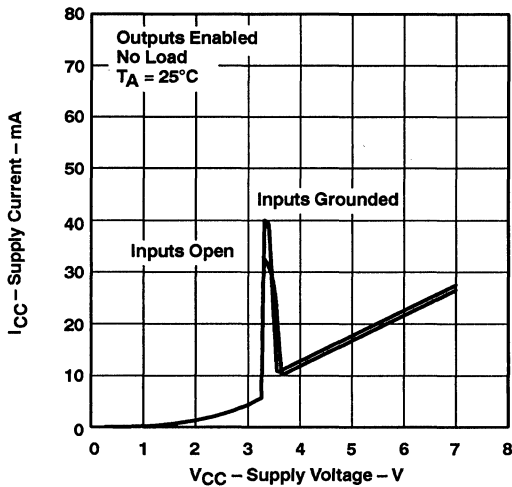


Figure 15

SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

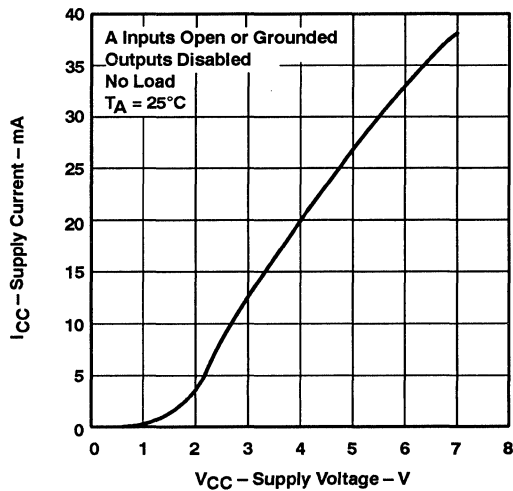


Figure 16

† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$  are applicable to the SN55ALS194 circuits only.

NOTE A: The A input is connected to GND during the testing of the Y outputs and to  $V_{CC}$  during the testing of the Z outputs.

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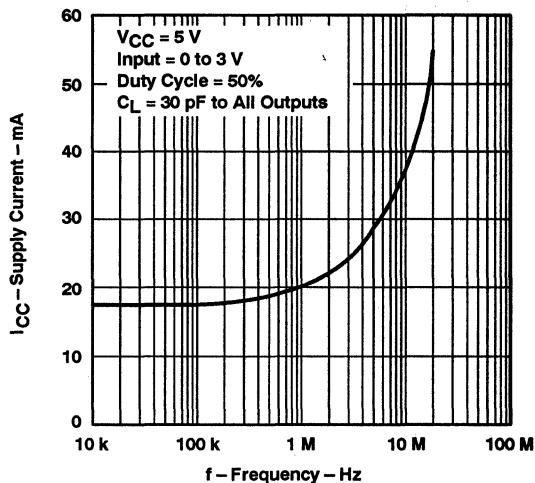


**SN55ALS194, SN75ALS194**  
**QUAD DIFFERENTIAL LINE DRIVERS**  
**WITH 3-STATE OUTPUTS**

SLLS009C - D2917, OCTOBER 1985 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

**SUPPLY CURRENT**  
**vs**  
**FREQUENCY**



**Figure 17**

# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

SLLS010C – D2928, JUNE 1986 – REVISED MARCH 1993

- Meets EIA Standards RS-422-A and RS-423-A
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- -7 V to 7 V Common-Mode Input Voltage Range With 200-mV Sensitivity
- 3-State TTL-Compatible Outputs
- High Input Impedance . . . 12 k $\Omega$  Min
- Input Hysteresis . . . 120 mV Typ
- Single 5-V Supply Operation
- Low Supply Current Requirement  
35 mA Max
- Improved Speed and Power Consumption Compared to MC3486

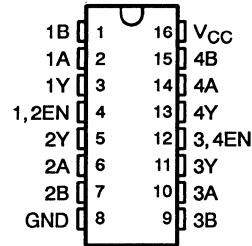
## description

The SN55ALS195 and SN75ALS195 are monolithic quad line receivers with 3-state outputs designed using advanced low-power Schottky technology. This technology provides combined improvements in die design, tooling production, and wafer fabrication, which in turn, provide lower power consumption and permit much higher data throughput than other designs. The devices meet the specifications of EIA Standards RS-422-A and RS-423-A. The 3-state outputs permit direct connection to a bus-organized system with a fail-safe design that ensures the outputs will always be high if the inputs are open.

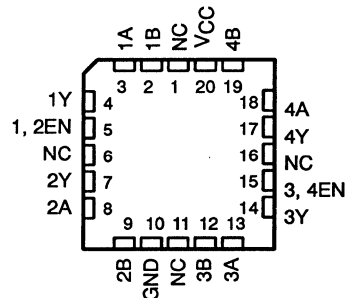
The devices are optimized for balanced multipoint bus transmission at rates up to 20 megabits per second. The input features high input impedance, input hysteresis for increased noise immunity, and an input sensitivity of  $\pm 200$  mV over a common-mode input voltage range of  $\pm 7$  V. The devices also feature an active-high enable function for each of two receiver pairs. The SN55ALS195 and SN75ALS195 are designed for optimum performance when used with the SN55ALS194 and SN75ALS194 quad differential line drivers.

The SN55ALS195 is characterized for operation over the full military temperature range of -55°C to 125°C. The SN75ALS195 is characterized for operation from 0°C to 70°C.

SN55ALS195 . . . J OR W PACKAGE  
SN75ALS195 . . . J PACKAGE  
(TOP VIEW)



SN55ALS195 . . . FK PACKAGE  
(TOP VIEW)

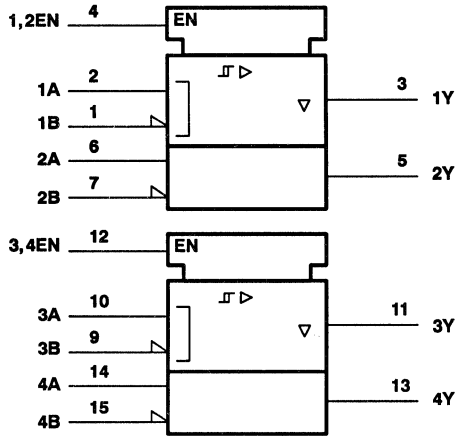


NC – No internal connection

# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

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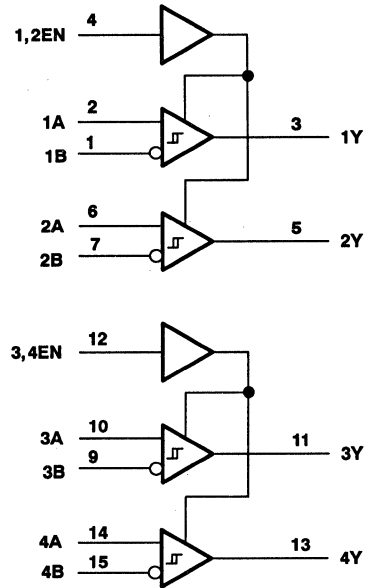
## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for the J and W packages.

## logic diagram



FUNCTION TABLE  
(each receiver)

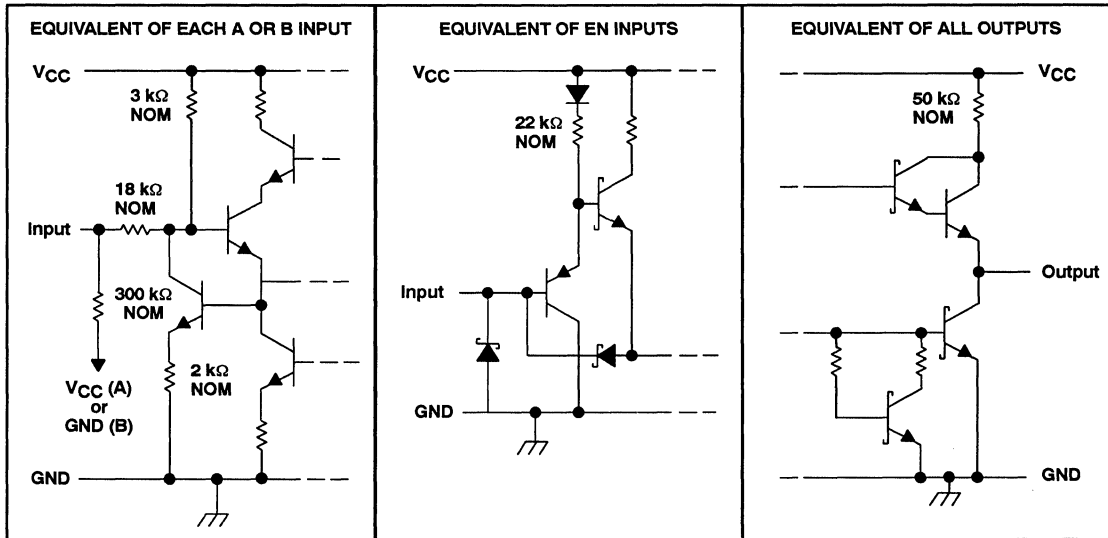
DIFFERENTIAL INPUTS A-B	ENABLE EN	OUTPUT Y
$V_{ID} \geq 0.2V$	H	H
$-0.2V < V_{ID} < 0.2V$	H	?
$V_{ID} \leq -0.2V$	H	L
X	L	Z
Open	H	H

H = high level, L = low level, X = irrelevant, ? = indeterminate, Z = high impedance (off)

# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

SLLS010C – D2928, JUNE 1986 – REVISED MARCH 1993

## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, A or B inputs, $V_I$	$\pm 15$ V
Differential input voltage (see Note 2)	$\pm 15$ V
Enable input voltage	7 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN55ALS195	$-55^{\circ}\text{C}$ to $125^{\circ}\text{C}$
SN75ALS195	$0^{\circ}\text{C}$ to $70^{\circ}\text{C}$
Storage temperature range	$-65^{\circ}\text{C}$ to $150^{\circ}\text{C}$
Case temperature for 60 seconds: FK package	$260^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	$300^{\circ}\text{C}$

† Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.

2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^{\circ}\text{C}$	DERATING FACTOR	$T_A = 70^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING	ABOVE $T_A = 25^{\circ}\text{C}$	POWER RATING	POWER RATING
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
J (SN55ALS195)	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	275 mW
J (SN75ALS195)	1025 mW	8.2 mW/ $^{\circ}\text{C}$	656 mW	N/A
W	1000 mW	8.0 mW/ $^{\circ}\text{C}$	690 mW	200 mW

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# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

SLLS010C - D2928, JUNE 1986 - REVISED MARCH 1993

## recommended operating conditions

	SN55ALS195			SN75ALS195			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$	±7			±7			V
Differential input voltage, $V_{ID}$	±12			±12			V
High-level input voltage, $V_{IH}$	2			2			V
Low-level input voltage, $V_{IL}$	0.8			0.8			V
High-level output current, $I_{OH}$	-400			-400			μA
Low-level output current, $I_{OL}$	16			16			mA
Operating free-air temperature, $T_A$	-55		125	0		70	°C

## electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT	
$V_{T+}$ Positive-going threshold voltage					200	mV	
$V_{T-}$ Negative-going threshold voltage					-200§	mV	
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )					120	mV	
$V_{IK}$ Enable-input clamp voltage	$V_{CC} = \text{MIN},$	$I_I = -18 \text{ mA}$			-1.5	V	
$V_{OH}$ High-level output voltage	$V_{CC} = \text{MIN},$ See Figure 1	$V_{ID} = 200 \text{ mV},$ $I_{OH} = -400 \text{ μA},$	2.5	3.6		V	
$V_{OL}$ Low-level output voltage	$V_{CC} = \text{MIN},$ $V_{ID} = -200 \text{ mV},$ See Figure 1	$I_{OL} = 8 \text{ mA}$			0.45	V	
		$I_{OL} = 16 \text{ mA}$			0.5		
$I_{OZ}$ High-impedance state output current	$V_{CC} = \text{MAX},$ $V_O = 2.7 \text{ V}$	$V_{IL} = 0.8 \text{ V},$ $V_{ID} = -3 \text{ V},$			20	μA	
	$V_{CC} = \text{MAX},$ $V_O = 0.5 \text{ V}$	$V_{IL} = 0.8 \text{ V},$ $V_{ID} = 3 \text{ V},$			-20		
$I_I$ Line input current	Other input at 0 V, See Note 3	$V_{CC} = \text{MIN},$ $V_I = 15 \text{ V}$			0.7	mA	
		$V_{CC} = \text{MAX},$ $V_I = -15 \text{ V}$			-1		
$I_{IH}$ High-level enable-input current	$V_{CC} = \text{MAX}$	$V_{IH} = 2.7 \text{ V}$			20	μA	
		$V_{IH} = 5.25 \text{ V}$			100		
$I_{IL}$ Low-level enable-input current	$V_{CC} = \text{MAX},$	$V_{IL} = 0.4 \text{ V}$			-100	μA	
Input resistance			12	18		kΩ	
$I_{OS}$ Short-circuit output current	$V_{CC} = \text{MAX},$ See Note 4	$V_{ID} = 3 \text{ V},$ $V_O = 0,$	-15	-78	130	mA	
$I_{CC}$ Supply current	$V_{CC} = \text{MAX},$	Outputs disabled			22	35	mA

† For conditions shown as MIN or MAX, use the appropriate values specified under recommended operating conditions.

‡ All typical values are at  $V_{CC} = 5 \text{ V}, T_A = 25^\circ\text{C}.$

§ The algebraic convention, in which the less positive limit is designated minimum, is used in this data sheet for threshold voltage levels only.

NOTES: 3. Refer to EIA Standards RS-422-A and RS-423-A for exact conditions.

4. Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

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switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $C_L = 15\text{ pF}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$V_{ID} = 0\text{ V to }3\text{ V}$ , See Figure 2	15	22		ns
$t_{PHL}$ Propagation delay time, high-to-low-level output		15	22		ns
$t_{PZH}$ Output enable time to high level	See Figure 3	13	25		ns
$t_{PZL}$ Output enable time to low level		10	25		
$t_{PHZ}$ Output disable time from high level	See Figure 3	19	25		ns
$t_{PLZ}$ Output disable time from low level		17	22		

## PARAMETER MEASUREMENT INFORMATION

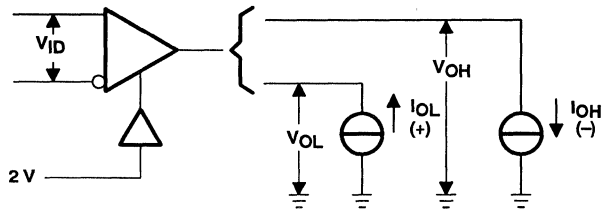
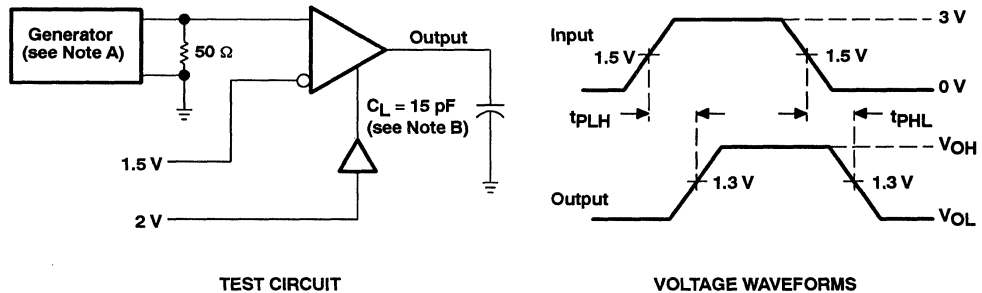


Figure 1.  $V_{OH}$ ,  $V_{OL}$



TEST CIRCUIT

VOLTAGE WAVEFORMS

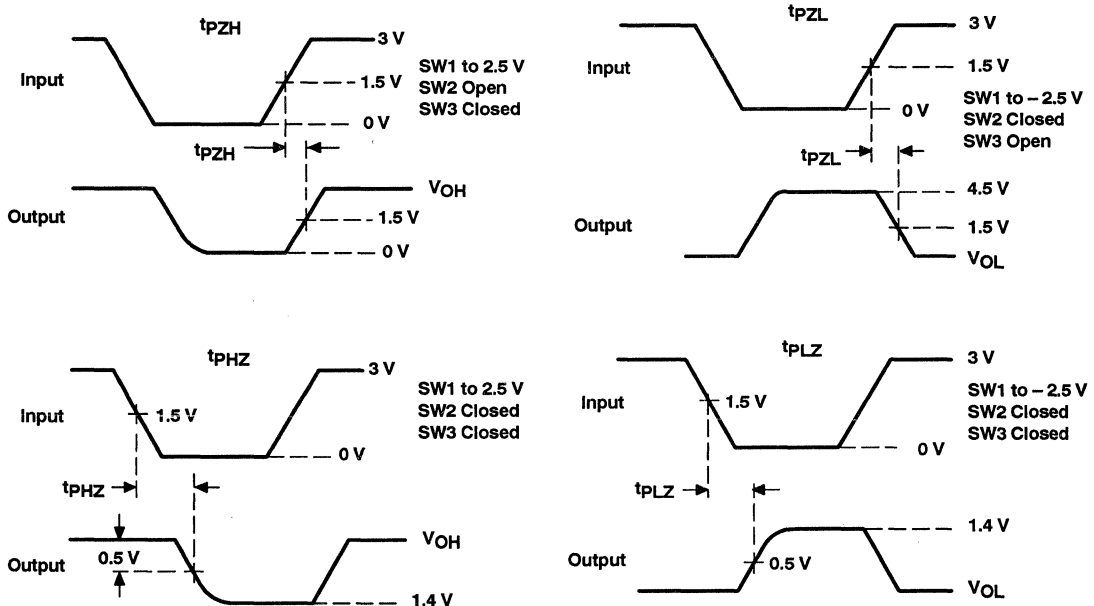
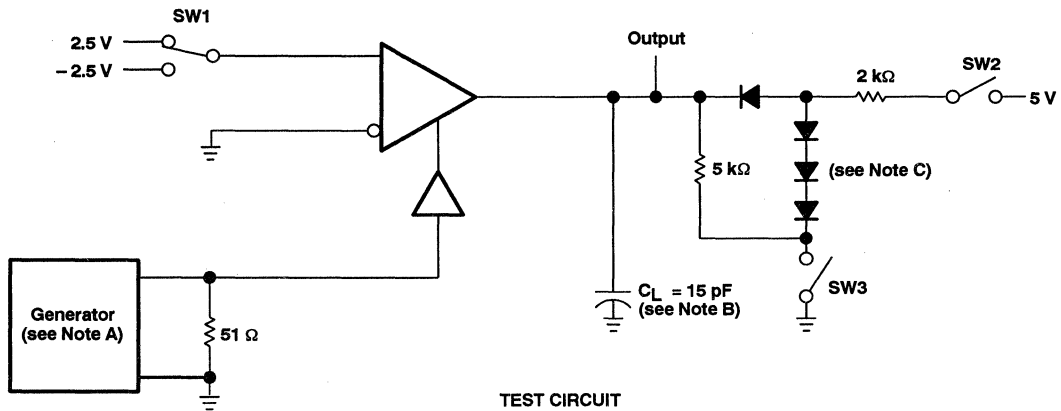
NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1\text{ MHz}$ , duty cycle  $\leq 50\%$ ,  $Z_0 = 50\ \Omega$ ,  $t_r \leq 6\text{ ns}$ ,  $t_f \leq 6\text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.

Figure 2. Test Circuit and Voltage Waveforms

# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

SLLS010C - D2928, JUNE 1986 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



## VOLTAGE WAVEFORMS

- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, duty cycle  $\leq 50\%$ ,  $Z_0 = 50 \Omega$ ,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns.  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.

Figure 3. Test Circuit and Voltage Waveforms

# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

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## TYPICAL CHARACTERISTICS†

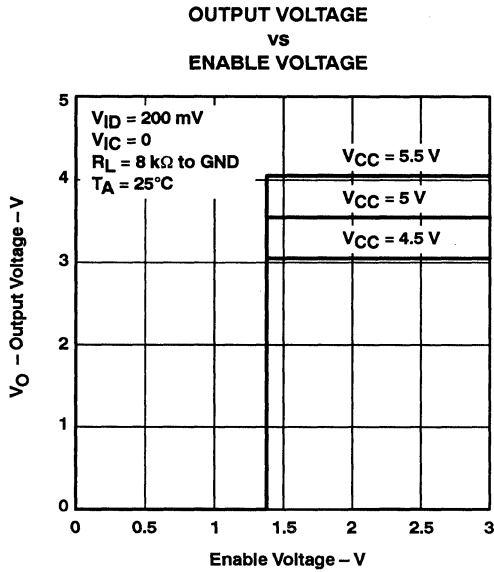


Figure 4

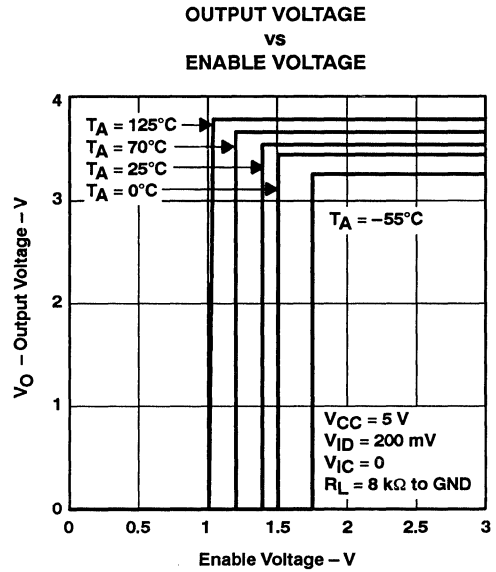


Figure 5

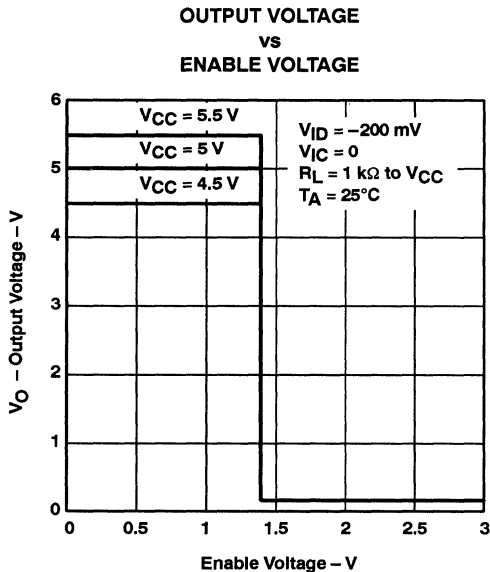


Figure 6

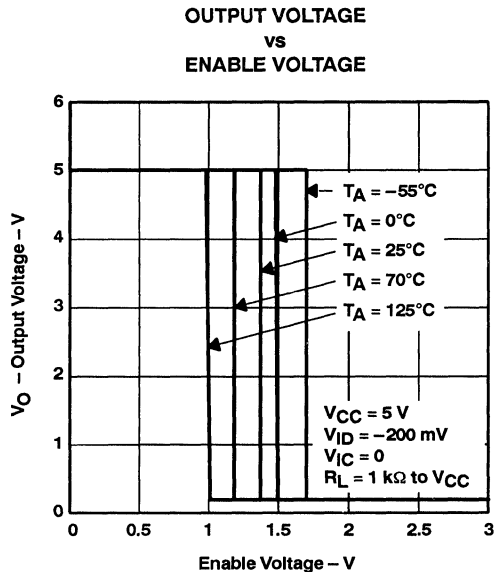


Figure 7

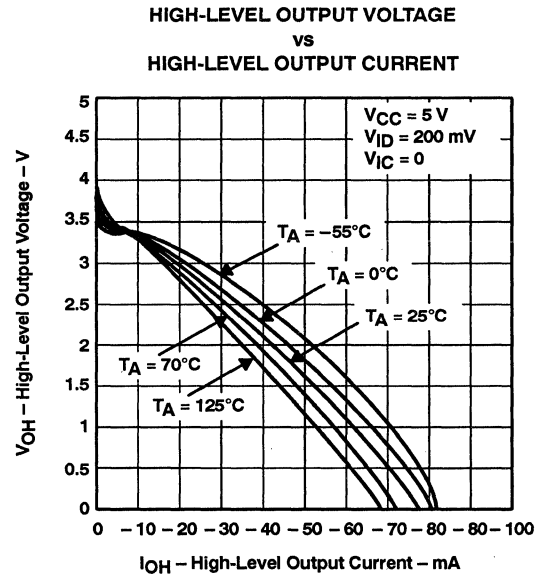
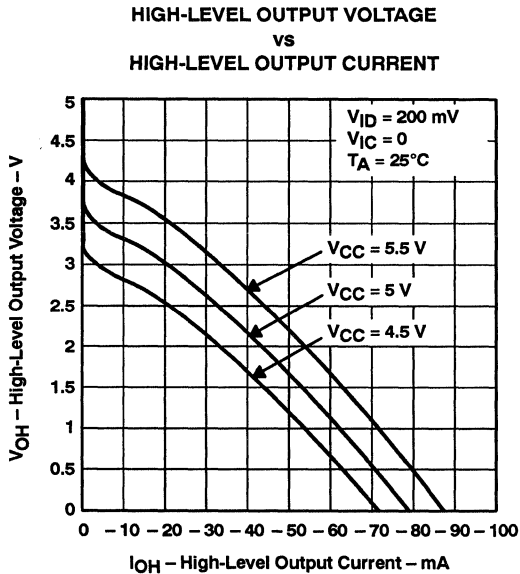
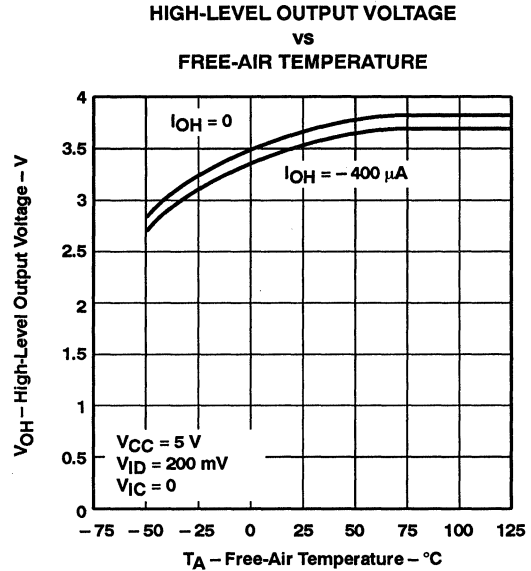
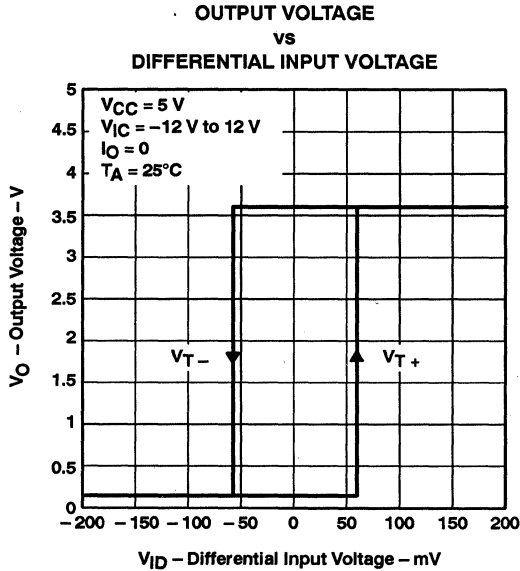
† Data for temperatures below  $0^\circ\text{C}$  and above  $70^\circ\text{C}$ , and below  $4.75 \text{ V}$  and above  $5.25 \text{ V}$ , are applicable to SN55ALS195 circuits only.



**SN55ALS195, SN75ALS195**  
**QUAD DIFFERENTIAL LINE RECEIVERS**  
**WITH 3-STATE OUTPUTS**

SLLS010C - D2928, JUNE 1986 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS†**



† Data for temperatures below 0°C and above 70°C, and below 4.75 V and above 5.25 V, are applicable to SN55ALS195 circuits only.



SN55ALS195, SN75ALS195  
 QUAD DIFFERENTIAL LINE RECEIVERS  
 WITH 3-STATE OUTPUTS

SLLS010C - D2928, JUNE 1986 - REVISED MARCH 1993

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

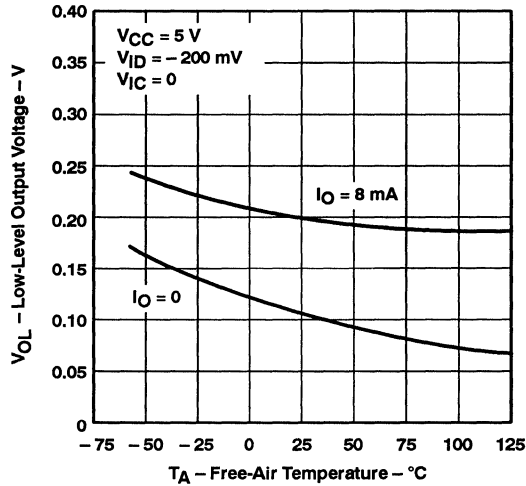


Figure 12

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

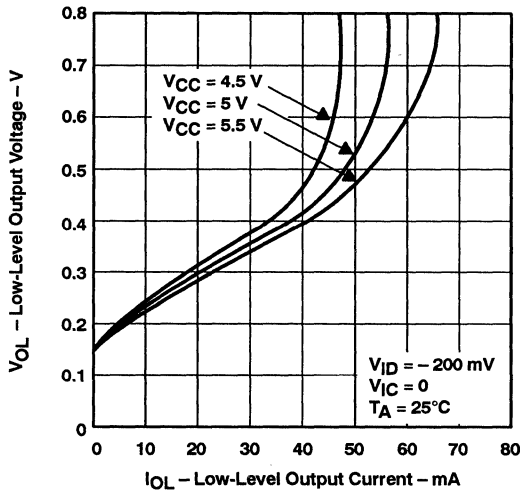


Figure 13

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

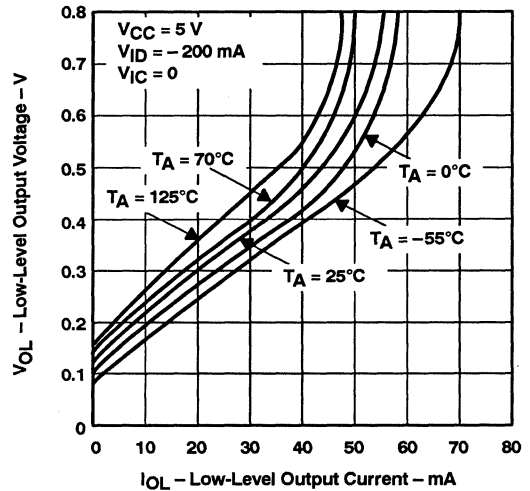


Figure 14

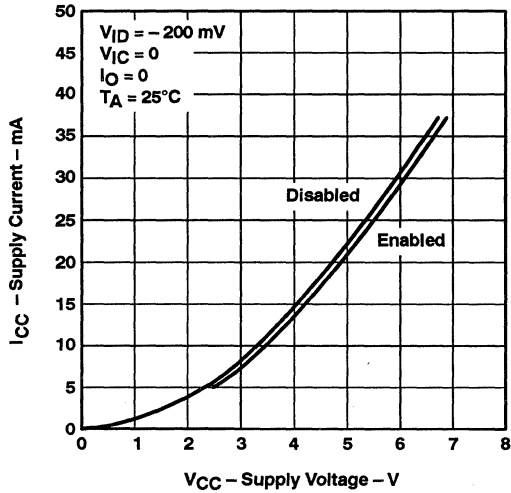
† Data for temperatures below 0°C and above 70°C, and below 4.75 V and above 5.25 V, are applicable to SN55ALS195 circuits only.

**SN55ALS195, SN75ALS195**  
**QUAD DIFFERENTIAL LINE RECEIVERS**  
**WITH 3-STATE OUTPUTS**

SLLS010C - D2928, JUNE 1986 - REVISED MARCH 1993

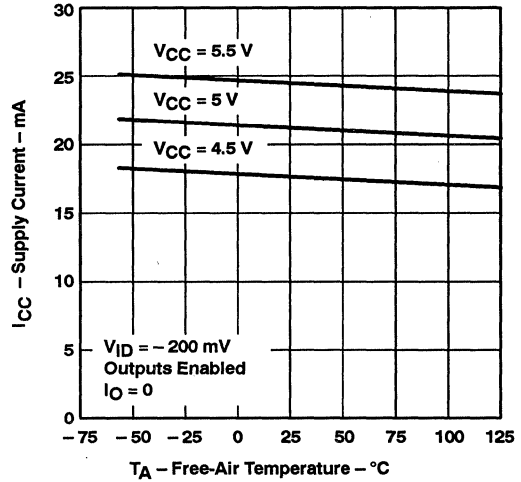
**TYPICAL CHARACTERISTICS†**

**SUPPLY CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



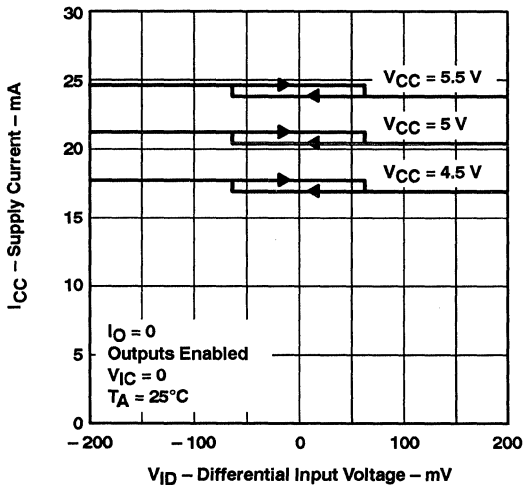
**Figure 15**

**SUPPLY CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



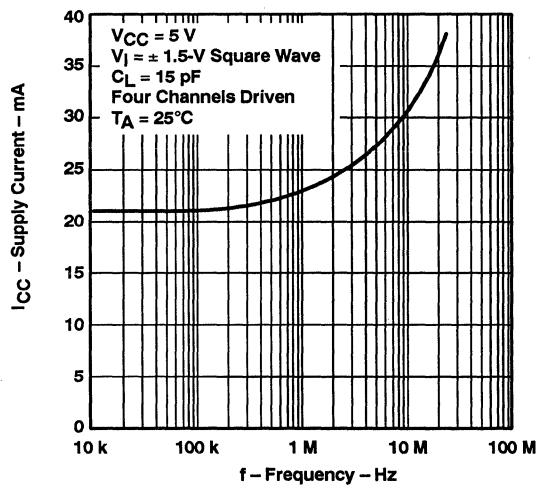
**Figure 16**

**SUPPLY CURRENT**  
**vs**  
**DIFFERENTIAL INPUT VOLTAGE**



**Figure 17**

**SUPPLY CURRENT**  
**vs**  
**FREQUENCY**



**Figure 18**

† Data for temperatures below 0°C and above 70°C, and below 4.75 V and above 5.25 V, are applicable to SN55ALS195 circuits only.



# SN55ALS195, SN75ALS195 QUAD DIFFERENTIAL LINE RECEIVERS WITH 3-STATE OUTPUTS

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## TYPICAL CHARACTERISTICS†

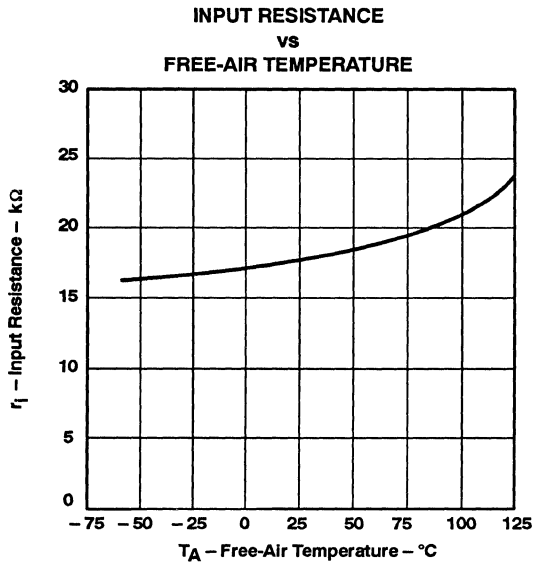


Figure 19

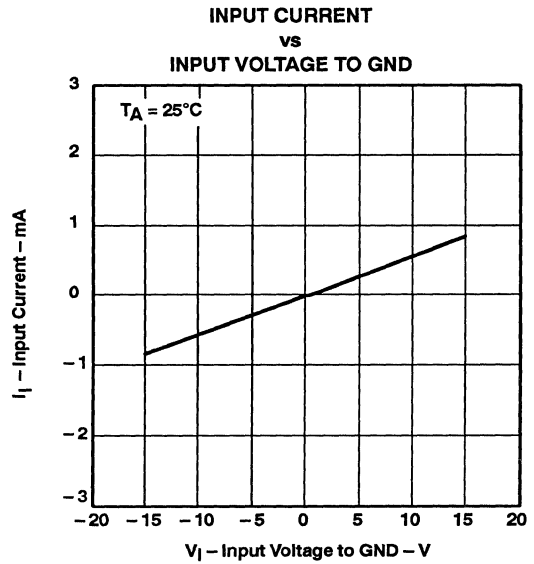


Figure 20

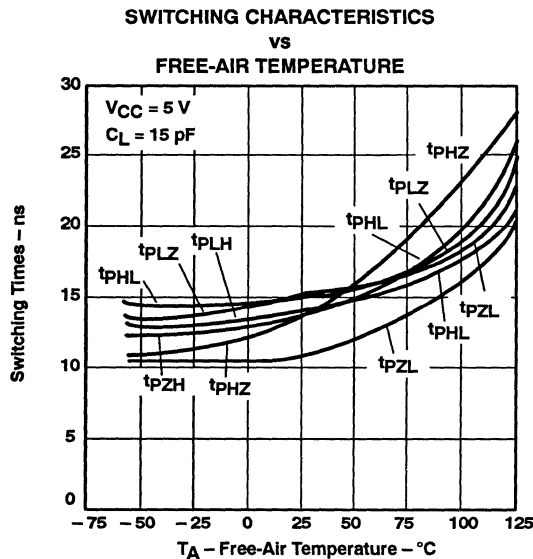


Figure 21

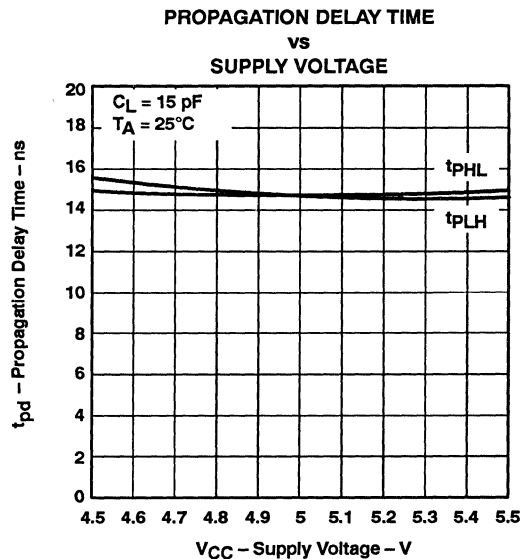


Figure 22

† Data for temperatures below  $0^{\circ}\text{C}$  and above  $70^{\circ}\text{C}$ , and below 4.75 V and above 5.25 V, are applicable to SN55ALS195 circuits only.



# SN75ALS197 QUAD DIFFERENTIAL LINE RECEIVER WITH 3-STATE OUTPUTS

SLLS045A - D3203, JANUARY 1989 - REVISED MARCH 1993

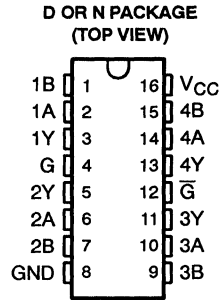
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- Designed for Multipoint Bus Transmission on Long Bus Lines in Noisy Environments
- 3-State Outputs
- Common-Mode Input Voltage Range - 7 V to 7 V
- Input Sensitivity . . .  $\pm 300$  mV
- Input Hysteresis . . . 120 mV Typ
- High Input Impedance . . . 12 k $\Omega$  Min
- Operates from Single 5-V Supply
- Low Supply Current Requirement 35 mA Max
- Improved Speed and Power Consumption Compared to AM26LS32A

## description

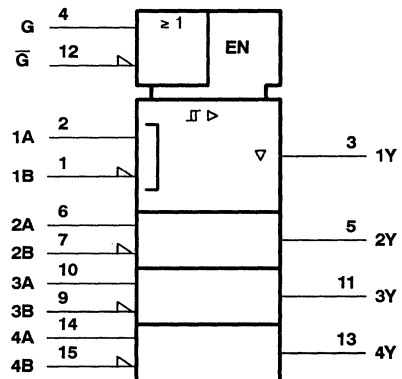
The SN75ALS197 is a monolithic quad line receiver with 3-state outputs designed using advanced low-power Schottky technology. This technology provides combined improvements in bar design, tooling production, and wafer fabrication. This, in turn, provides significantly less power requirements and permits much higher data throughput than other designs. The device meets the specifications of CCITT Recommendations V.10, V.11, X.26, and X.27. It features 3-state outputs that permit direct connection to a bus-organized system with a fail-safe design that ensures the outputs will always be high if the inputs are open.

The device is optimized for balanced multipoint bus transmission at rates up to 10 megabits per second. The input features high input impedance, input hysteresis for increased noise immunity, and an input sensitivity of  $\pm 300$  mV over a common-mode input voltage range of -7 V to 7 V. It also features active-high and active-low enable functions that are common to the four channels. The SN75ALS197 is designed for optimum performance when used with the SN75ALS192 quad differential line driver.

The SN75ALS197 is characterized for operation from 0°C to 70°C.

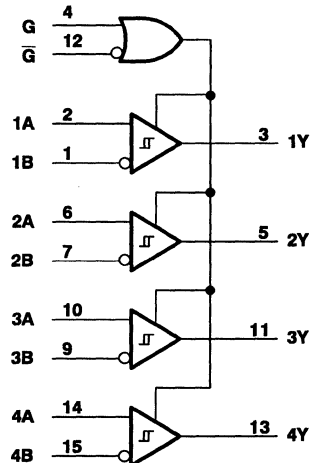


## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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**SN75ALS197**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

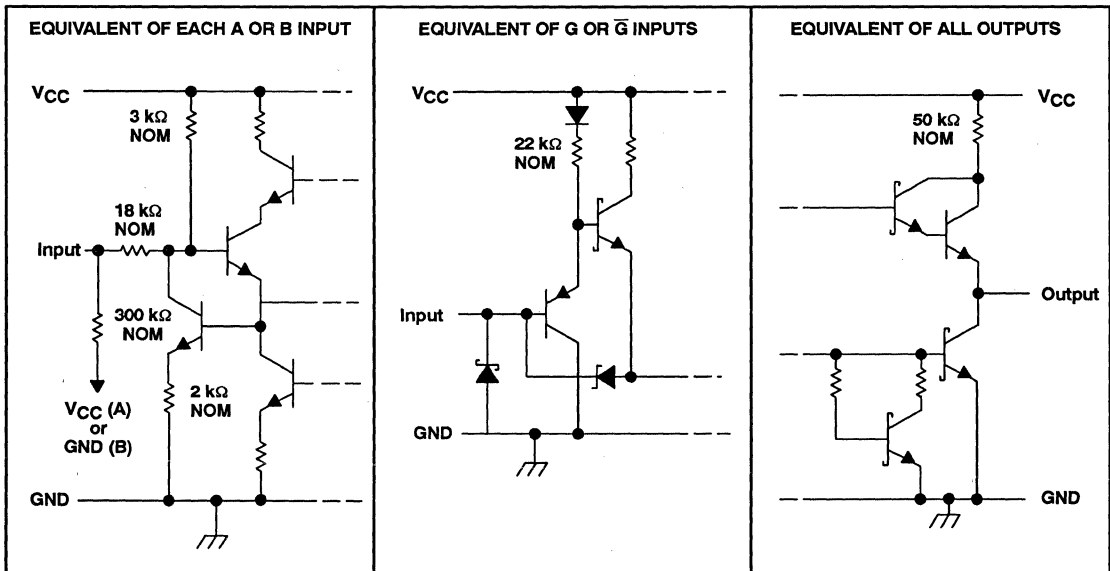
SLLS045A—D3203, JANUARY 1989—REVISED MARCH 1993

**FUNCTION TABLE**  
 (each receiver)

DIFFERENTIAL INPUTS A-B	ENABLES		OUTPUT Y
	G	$\bar{G}$	
$V_{ID} \geq 0.3 V$	H	X	H
	X	L	H
$-0.3 V < V_{ID} < 0.3 V$	H	X	?
	X	L	?
$V_{ID} \leq -0.3 V$	H	X	L
	X	L	L
X	L	H	Z
Open	H	X	H
	X	L	H

H = high level, L = low level, X = irrelevant, ? = indeterminate,  
 Z = high impedance (off)

**schematics of inputs and outputs**



**SN75ALS197**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS045A – D3203, JANUARY 1989 – REVISED MARCH 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Input voltage, A or B inputs, $V_I$ .....	$\pm 15$ V
Differential input voltage (see Note 2) .....	$\pm 15$ V
Enable input voltage .....	7 V
Low-level output current .....	50 mA
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the, corresponding inverting input.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Differential input voltage, $V_{ID}$			$\pm 12$	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			-400	$\mu\text{A}$
Low-level output current, $I_{OL}$			16	mA
Operating free-air temperature, $T_A$	0		70	°C





**SN75ALS197**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS045A - D3203, JANUARY 1989 - REVISED MARCH 1993

**electrical characteristics over recommended range of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage					300	mV
$V_{T-}$	Negative-going threshold voltage			-300‡			mV
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )	See Figure 4			120		mV
$V_{IK}$	Enable-input clamp voltage	$I_I = -18$ mA				-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 300$ mV,	$I_{OH} = -400$ $\mu$ A	2.7	3.6		V
$V_{OL}$	Low-level output voltage	$V_{ID} = -300$ mV	$I_{OL} = 8$ mA			0.45	V
			$I_{OL} = 16$ mA			0.5	
$I_{OZ}$	High-impedance-state output current	$V_{CC} = 5.25$ V	$V_O = 2.4$ V			20	$\mu$ A
			$V_{OH} = 0.4$ V			-20	
$I_I$	Line input current	Other input at 0 V, See Note 3	$V_I = 15$ V		0.7	1.2	mA
			$V_I = -15$ V		-1.0	-1.7	
$I_H$	High-level enable-input current		$V_{IH} = 2.7$ V			20	$\mu$ A
			$V_{IH} = 5.25$ V			100	
$I_{IL}$	Low-level enable-input current	$V_{IL} = 0.4$ V				-100	$\mu$ A
Input resistance				12	18		k $\Omega$
$I_{OS}$	Short-circuit output current§	$V_{ID} = 3$ V,	$V_O = 0$	-15	-78	-130	mA
$I_{CC}$	Supply current	Outputs disabled			22	35	mA

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ$ C.

‡ The algebraic convention, in which the less positive limit is designated minimum, is used in this data sheet for threshold voltage levels only.

§ Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

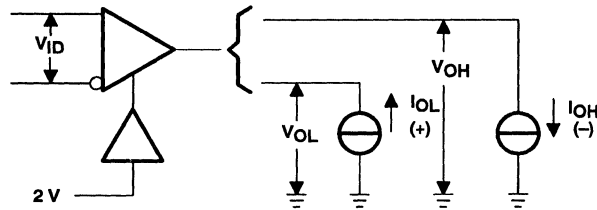
NOTE 3: Refer to EIA Standard RS-422-A and RS-423-A for exact conditions.

**switching characteristics,  $V_{CC} = 5$  V,  $T_A = 25^\circ$ C**

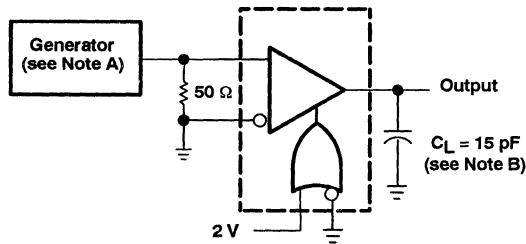
PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -2.5$ V to 2.5 V, $C_L = 15$ pF, See Figure 2			15	22	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output				15	22	
$t_{PZH}$	Output enable time to high level	$C_L = 15$ pF, See Figure 2			13	25	ns
$t_{PZL}$	Output enable time to low level				11	25	
$t_{PHZ}$	Output disable time from high level	$C_L = 15$ pF, See Figure 2			13	25	ns
$t_{PLZ}$	Output disable time from low level				15	22	



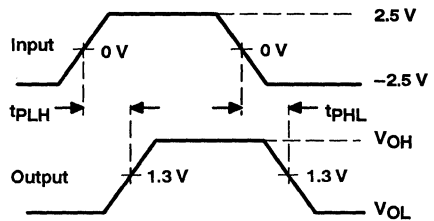
**PARAMETER MEASUREMENT INFORMATION**



**Figure 1.  $V_{OH}$ ,  $V_{OL}$  Test Circuit**



**TEST CIRCUIT**



**VOLTAGE WAVEFORMS**

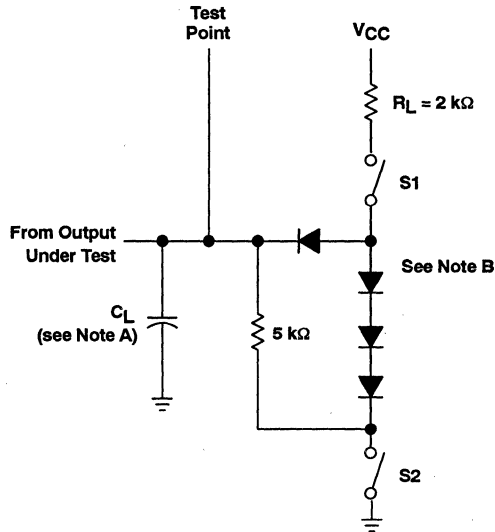
- NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, duty cycle  $\leq 50\%$ ,  $Z_0 = 50 \Omega$ ,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns.  
 B.  $C_L$  includes probe and jig capacitance.

**Figure 2.  $t_{PLH}$ ,  $t_{PHL}$  Test Circuit and Voltage Waveforms**

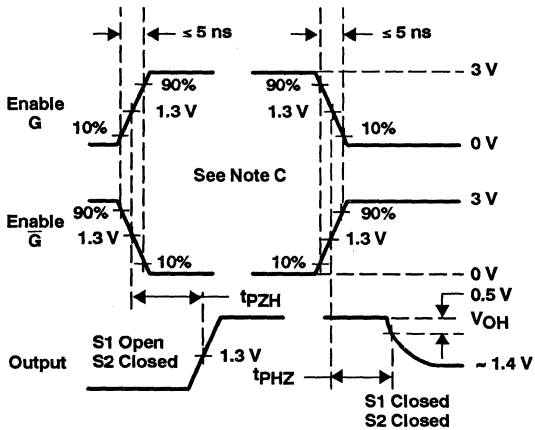
**SN75ALS197**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS045A - D3203, JANUARY 1989 - REVISED MARCH 1993

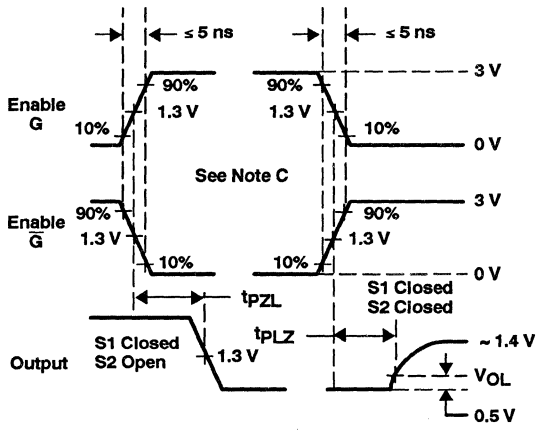
**PARAMETER MEASUREMENT INFORMATION**



**LOAD CIRCUIT**



**VOLTAGE WAVEFORMS FOR  $t_{pZH}$ ,  $t_{pZL}$**



**VOLTAGE WAVEFORMS FOR  $t_{pLZ}$ ,  $t_{pLH}$**

- NOTES: A.  $C_L$  includes probe and jig capacitance.  
 B. All diodes are 1N3064 or equivalent.  
 C. Enable G is tested with  $\bar{G}$  high;  $\bar{G}$  is tested with G low.

**Figure 3.  $t_{pZH}$ ,  $t_{pZL}$ ,  $t_{pLZ}$ ,  $t_{pLH}$  Load Circuit and Voltage Waveforms**

TYPICAL CHARACTERISTICS

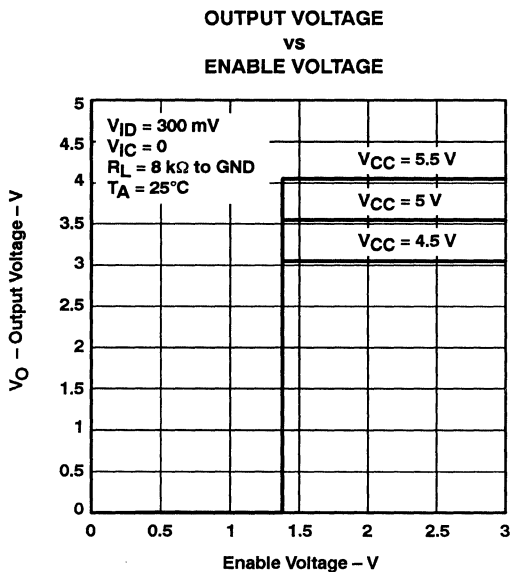


Figure 4

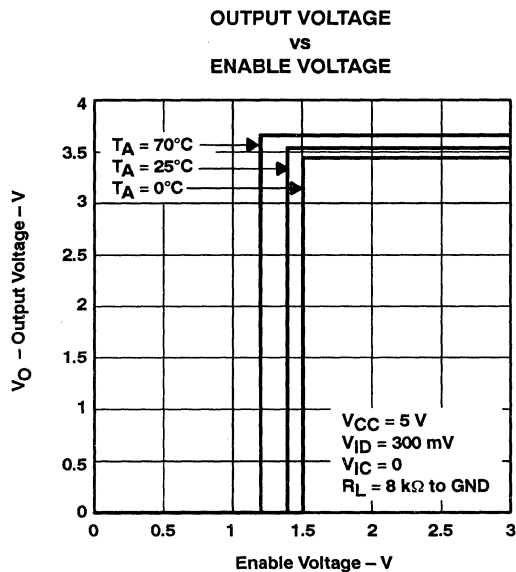


Figure 5

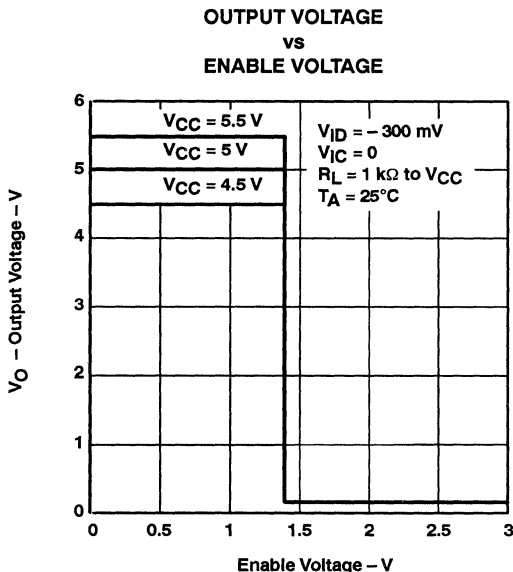


Figure 6

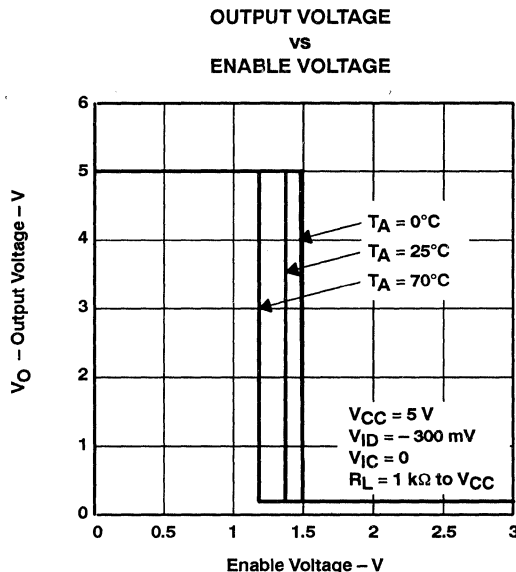


Figure 7

**SN75ALS197**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS045A - D3203, JANUARY 1989 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

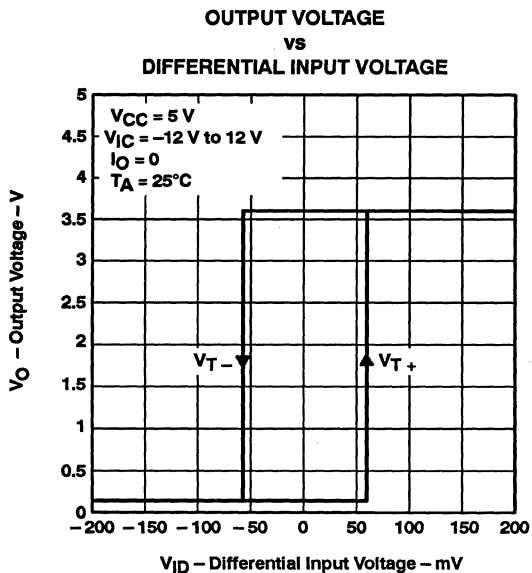


Figure 8

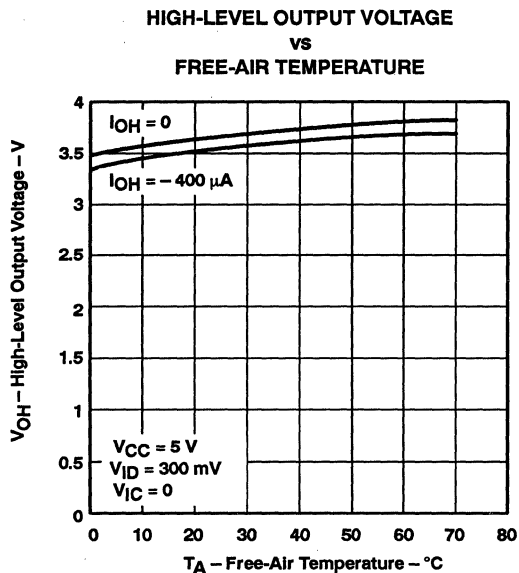


Figure 9

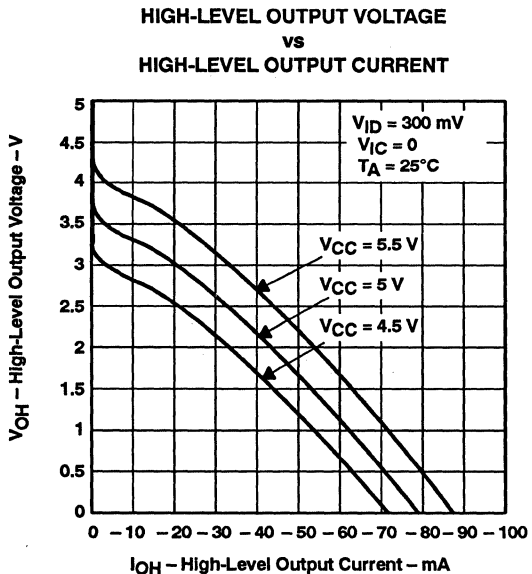


Figure 10

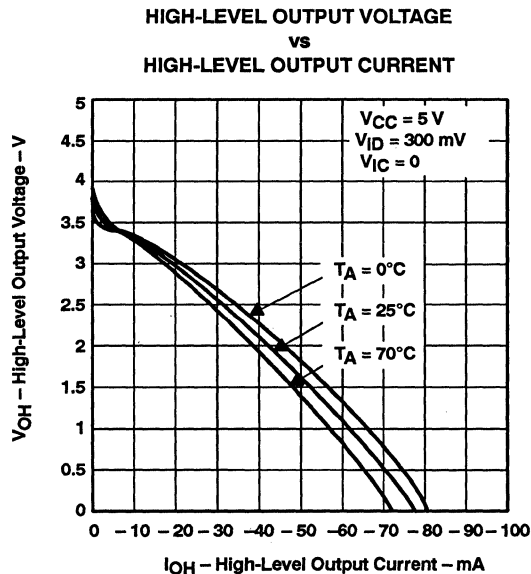


Figure 11

TYPICAL CHARACTERISTICS

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

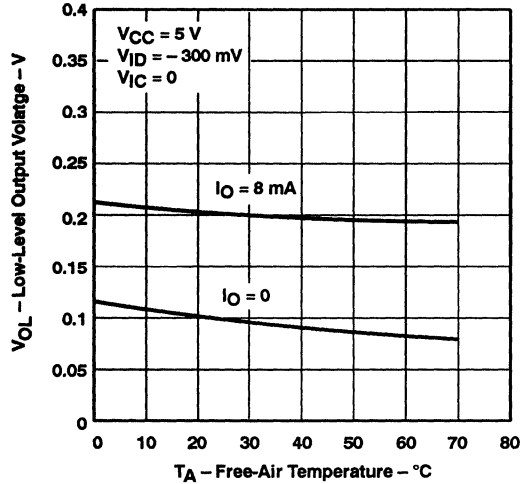


Figure 12

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

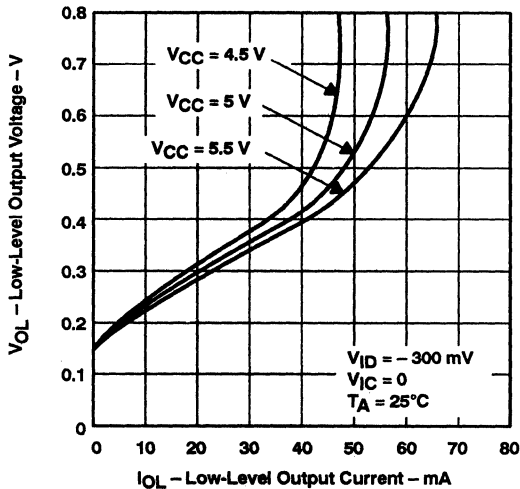


Figure 13

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

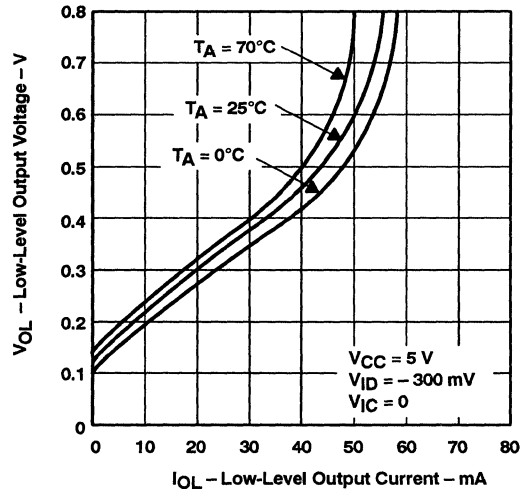


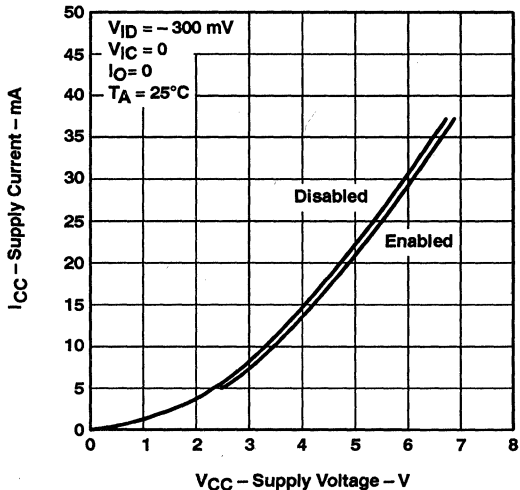
Figure 14

**SN75ALS197**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS045A - D3203, JANUARY 1989 - REVISED MARCH 1993

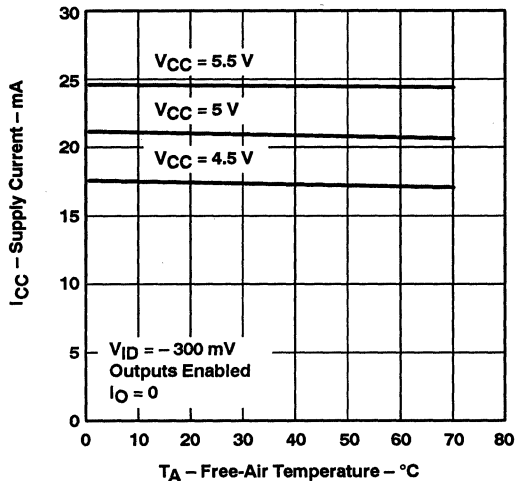
**TYPICAL CHARACTERISTICS**

**SUPPLY CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



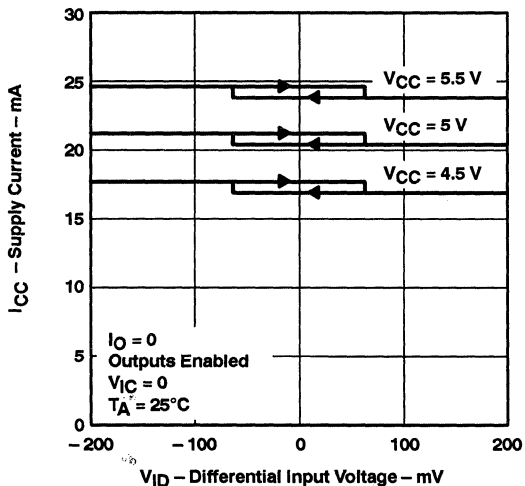
**Figure 15**

**SUPPLY CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



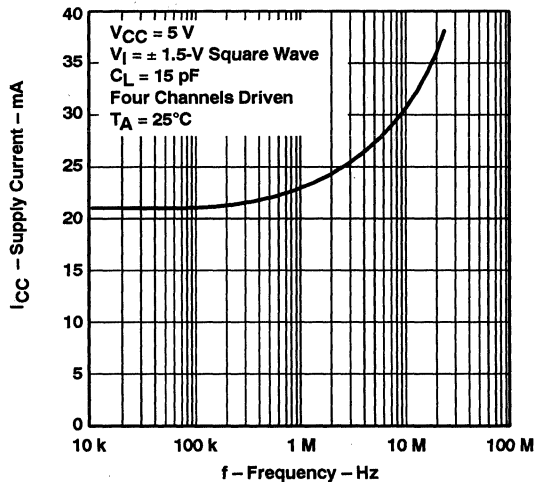
**Figure 16**

**SUPPLY CURRENT**  
**vs**  
**DIFFERENTIAL INPUT VOLTAGE**



**Figure 17**

**SUPPLY CURRENT**  
**vs**  
**FREQUENCY**



**Figure 18**



**SN75ALS197**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS045A - D3203, JANUARY 1989 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

**INPUT RESISTANCE**  
**vs**  
**FREE-AIR TEMPERATURE**

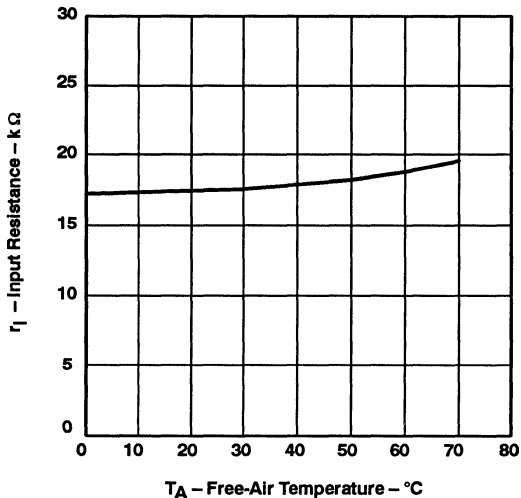


Figure 19

**INPUT CURRENT**  
**vs**  
**INPUT VOLTAGE TO GND**

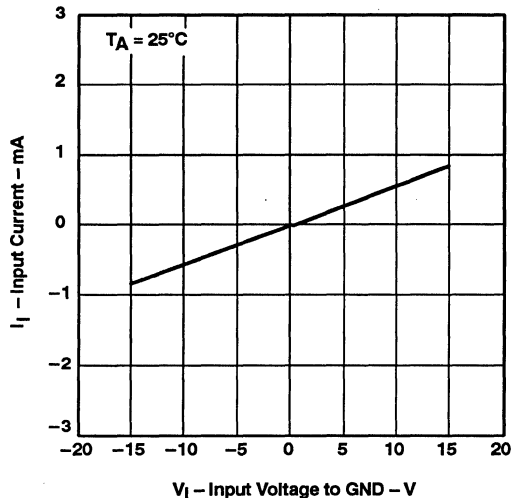


Figure 20

**SWITCHING CHARACTERISTICS**  
**vs**  
**FREE-AIR TEMPERATURE**

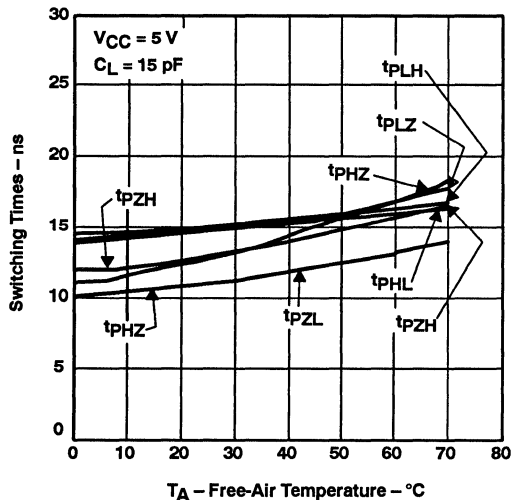


Figure 21

**PROPAGATION DELAY TIME**  
**vs**  
**SUPPLY VOLTAGE**

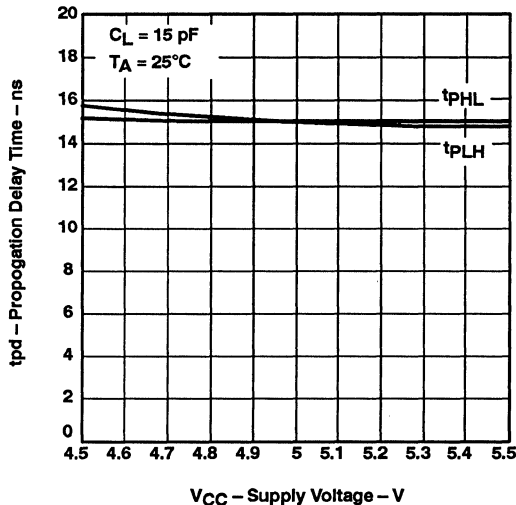


Figure 22







# SN65C198, SN75C198 QUAD LOW-POWER LINE DRIVERS

SLLS051A – D3472, JULY 1990 – REVISED MARCH 1993

- Meets EIA-232-D (Revision of RS-232-C)
- Very Low Supply Current
- Sleep Mode:  
3-State Outputs in High-Impedance State  
Ultra-Low Supply Current . . . 17  $\mu$ A Typ
- Improved Functional Replacement for:  
SN75188,  
Motorola MC1488,  
National Semiconductor DS14C88, and  
DS1488
- CMOS- and TTL-Compatible Data Inputs
- On-Chip Slew-Rate Limit . . . 30 V/ $\mu$ s
- Output Current Limit . . . 10 mA Typ
- Wide Supply Voltage Range . . .  $\pm 4.5$  V  
to  $\pm 15$  V
- ESD Protection Exceeds 500 V Per  
MIL-STD-883C, Method 3015.2

## description

The SN65C198 and SN75C198 are monolithic low-power BI-MOS quad low-power line drivers designed to interface data terminal equipment (DTE) with data circuit-terminating equipment (DCE) in conformance with the specifications of ANSI/EIA-232-D-1986.

The sleep-mode input,  $\overline{SM}$ , can be used to switch the outputs to high impedance, which avoids the transmission of corrupted data during power up and allows significant system power savings during data-off periods.

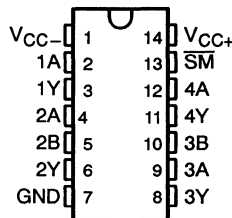
The SN65C198 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75C198 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

FUNCTION TABLE

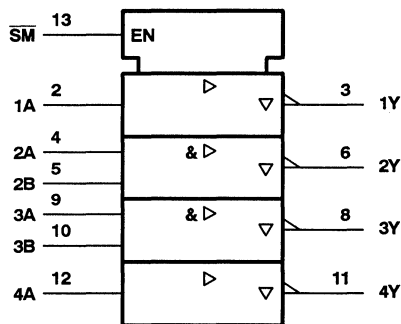
INPUTS			OUTPUT
$\overline{SM}$	A	B	Y
H	H	H	L
H	L	X	H
H	X	L	H
L	X	X	Z

H = high level, L = low level,  
X = irrelevant,  
Z = high impedance

D OR N PACKAGE  
(TOP VIEW)

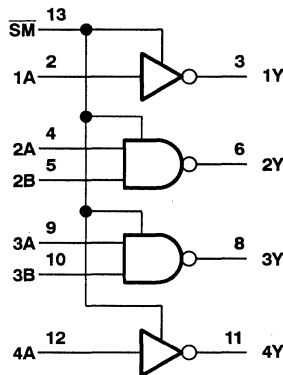


## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

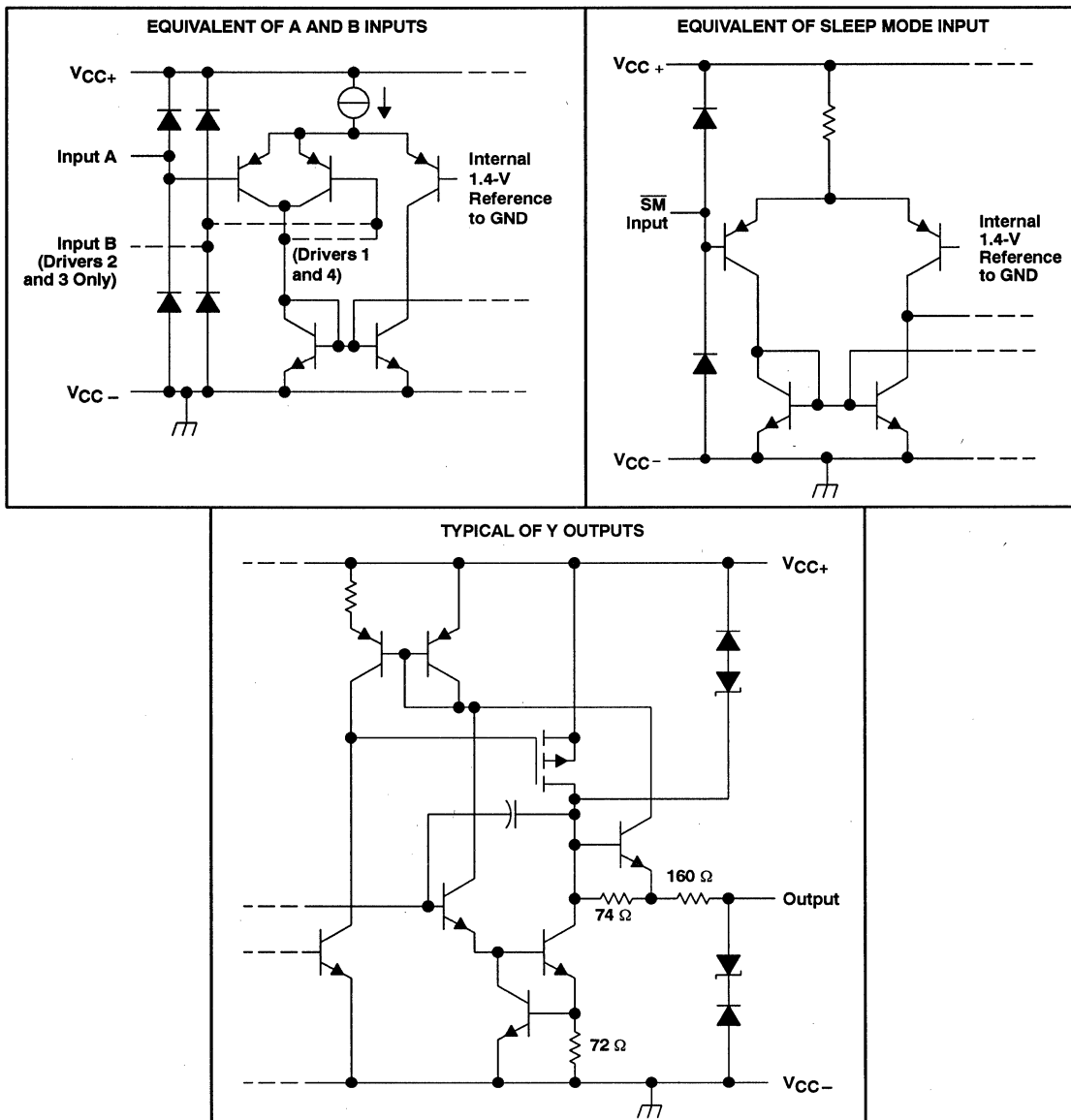
## logic diagram (positive logic)



# SN65C198, SN75C198 QUAD LOW-POWER LINE DRIVERS

SLLS051A - D3472, JULY 1990 - REVISED MARCH 1993

## schematics of inputs and outputs



All resistor values shown are nominal.

# SN65C198, SN75C198 QUAD LOW-POWER LINE DRIVERS

SLLS051A – D3472, JULY 1990 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	15 V
Supply voltage, $V_{CC-}$	-15 V
Input voltage range	-15 V to 15 V
Output voltage range	$V_{CC-} - 6\text{ V}$ to $V_{CC+} + 6\text{ V}$
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65C198	-40°C to 85°C
SN75C198	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	494 mW
N	1150 mW	9.2 mW/°C	598 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$		4.5	12	15	V
Supply voltage, $V_{CC-}$		-4.5	-12	-15	V
Input voltage, $V_I$ (see Figure 2)		$V_{CC-} + 2$		$V_{CC+}$	V
High-level input voltage, $V_{IH}$		2			V
Low-level input voltage, $V_{IL}$	A and B inputs				0.8
	SM input				0.6
Operating free-air temperature, $T_A$	SN65C198	-40		85	°C
	SN75C198	0		70	



# SN65C198, SN75C198 QUAD LOW-POWER LINE DRIVERS

SLLS051A—D3472, JULY 1990—REVISED MARCH 1993

**electrical characteristics over recommended operating free-air temperature range,  $V_{CC\pm} = \pm 12\text{ V}$ , SM at 2 V (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT	
VOH	High-level output voltage	$V_{IH} = 0.8\text{ V}$ , $R_L = 3\text{ k}\Omega$	$V_{CC\pm} = \pm 5\text{ V}$	4			V	
			$V_{CC\pm} = \pm 12\text{ V}$	10				
VOL	Low-level output voltage (see Note 2)	$V_{IH} = 2\text{ V}$ , $R_L = 3\text{ k}\Omega$	$V_{CC\pm} = \pm 5\text{ V}$			-4	V	
			$V_{CC\pm} = \pm 12\text{ V}$			-10		
I <sub>IH</sub>	High-level input current	$V_I = 5\text{ V}$				10	$\mu\text{A}$	
I <sub>IL</sub>	Low-level input current	$V_I = 0\text{ V}$				-10	$\mu\text{A}$	
I <sub>OZ</sub>	High-impedance-state output current	$\overline{\text{SM}}$ at 0.6 V	$V_O = 12\text{ V}$ , $V_{CC\pm} = \pm 12\text{ V}$			100	$\mu\text{A}$	
			$V_O = -12\text{ V}$ , $V_{CC\pm} = \pm 12\text{ V}$			-100		
I <sub>OS(H)</sub>	High-level short-circuit output current‡	$V_I = 0.8\text{ V}$ , $V_O = 0$ or $V_{CC-}$			-4.5	-10	-19.5	mA
I <sub>OS(L)</sub>	Low-level short-circuit output current‡	$V_I = 2\text{ V}$ , $V_O = 0$ or $V_{CC+}$			4.5	10	19.5	mA
r <sub>o</sub>	Output resistance	$V_{CC\pm} = 0$ , $V_O = -2\text{ V}$ to $2\text{ V}$			300			$\Omega$
I <sub>CC+</sub>	Supply current from $V_{CC+}$	A and B inputs at 0.8 V or 2 V, No load	$V_{CC\pm} = \pm 5\text{ V}$	90		160	$\mu\text{A}$	
			$V_{CC\pm} = \pm 12\text{ V}$	95		160		
		A and B inputs at 0.8 V or 2 V, $R_L = 3\text{ k}\Omega$ , $\overline{\text{SM}}$ at 0.6 V	$V_{CC\pm} = \pm 5\text{ V}$	17		40		
			$V_{CC\pm} = \pm 12\text{ V}$	17		40		
I <sub>CC-</sub>	Supply current from $V_{CC-}$	A and B inputs at 0.8 V or 2 V, No load	$V_{CC\pm} = \pm 5\text{ V}$	-90		-160	$\mu\text{A}$	
			$V_{CC\pm} = \pm 12\text{ V}$	-95		-160		
		A and B inputs at 0.8 V or 2 V, $R_L = 3\text{ k}\Omega$ , $\overline{\text{SM}}$ at 0.6 V	$V_{CC\pm} = \pm 5\text{ V}$	-17		-40		
			$V_{CC\pm} = \pm 12\text{ V}$	-17		-40		

† All typical values are at  $T_A = 25^\circ\text{C}$ .

‡ Not more than one output should be shorted at a time.

NOTE 2: The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only, e.g., if  $-10\text{ V}$  is a maximum, the typical value is a more negative voltage.

**switching characteristics over recommended operating free-air temperature range,  $V_{CC\pm} = \pm 12\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output§	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$ , See Figure 1				3	$\mu\text{s}$
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output§					3.5	$\mu\text{s}$
t <sub>TLH</sub>	Transition time, low-to-high-level output¶			0.53	1	3.2	$\mu\text{s}$
t <sub>THL</sub>	Transition time, high-to-low-level output¶			0.53	1	3.2	$\mu\text{s}$
t <sub>TLH</sub>	Transition time, low-to-high-level output#	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 2500\text{ pF}$ , See Figure 2				1.5	$\mu\text{s}$
t <sub>THL</sub>	Transition time, high-to-low-level output#					1.5	$\mu\text{s}$
t <sub>PZH</sub>	Output enable time to high level	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$ , See Figure 3				50	$\mu\text{s}$
t <sub>PHZ</sub>	Output disable time from high level					10	$\mu\text{s}$
t <sub>PZL</sub>	Output enable time to low level	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$ , See Figure 4				15	$\mu\text{s}$
t <sub>PLZ</sub>	Output disable time from low level					10	$\mu\text{s}$
SR	Output slew rate#	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 15\text{ pF}$	6	15	30	V/ $\mu\text{s}$	

† All typical values are at  $T_A = 25^\circ\text{C}$ .

§ t<sub>PHL</sub> and t<sub>PLH</sub> include the additional time due to on-chip slew rate and are measured at the 50% points.

¶ Measured between 10% and 90% points of output waveform

# Measured between 3-V and  $-3\text{ V}$  points of output waveform

**TEXAS  
INSTRUMENTS**

PARAMETER MEASUREMENT INFORMATION

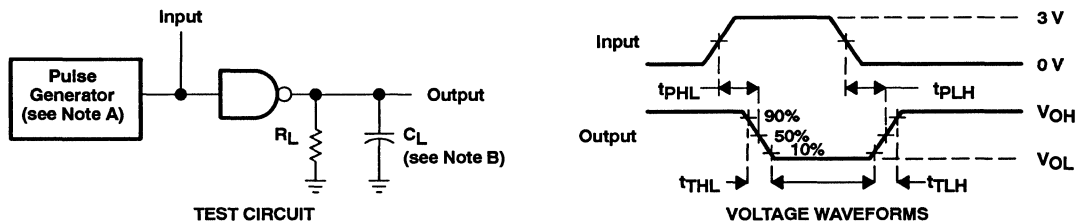


Figure 1. Test Circuit and Voltage Waveforms, Propagation and Transition Times

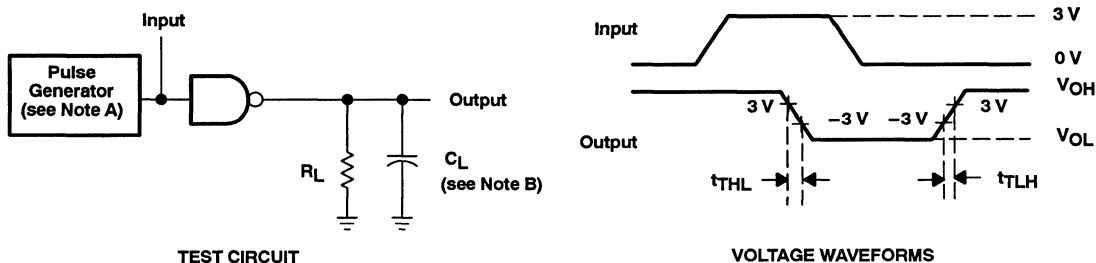


Figure 2. Test Circuit and Voltage Waveforms, Transition Times

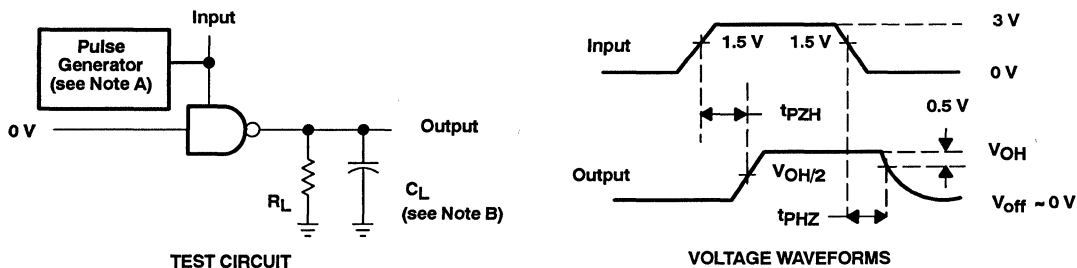


Figure 3. Driver Test Circuit and Voltage Waveforms

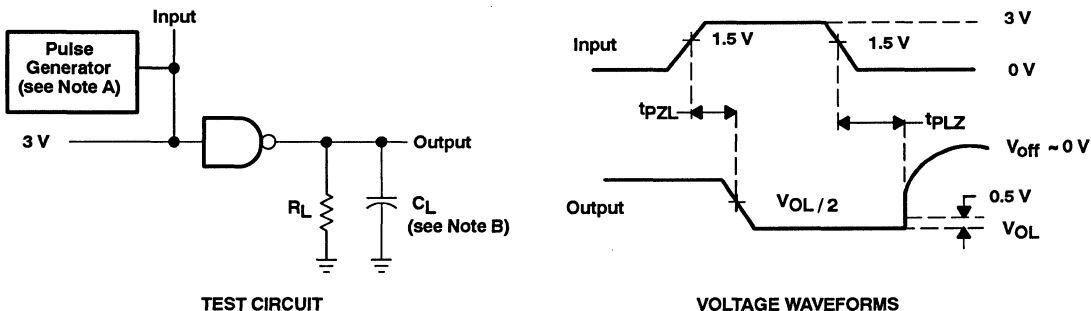


Figure 4. Driver Test Circuit and Voltage Waveforms

NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu s$ ,  $PRR = 20 \text{ kHz}$ ,  $Z_o = 50 \Omega$ ,  $t_r = t_f \leq 50 \text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.

# SN65C198, SN75C198 QUAD LOW-POWER LINE DRIVERS

SLLS051A—D3472, JULY 1990—REVISED MARCH 1993

## TYPICAL CHARACTERISTICS

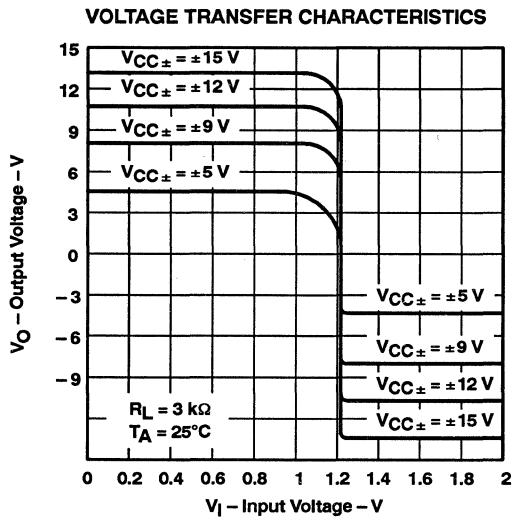


Figure 5

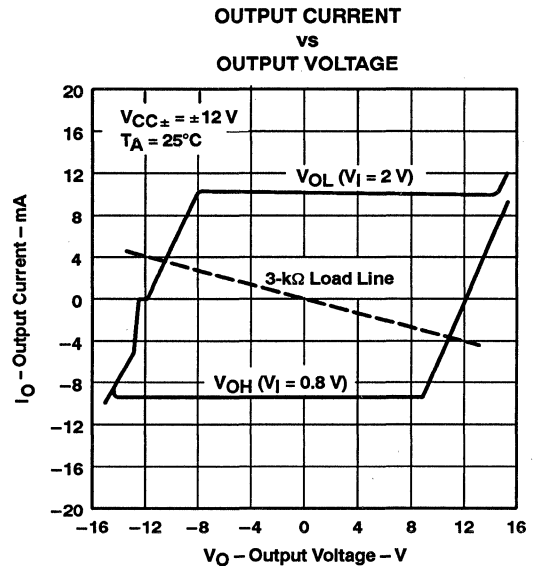


Figure 6

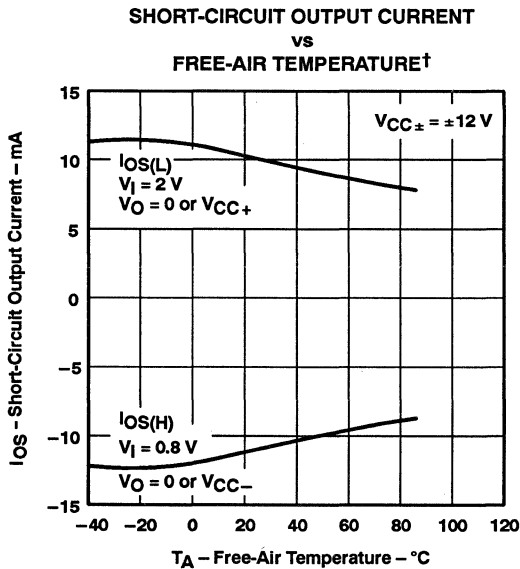


Figure 7

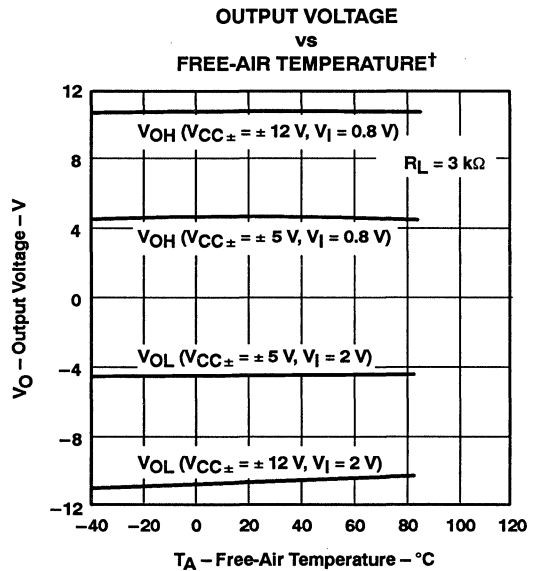


Figure 8

† Only the  $0^\circ\text{C}$  to  $70^\circ\text{C}$  portion of the curves applies to the SN75C198.

TYPICAL CHARACTERISTICS

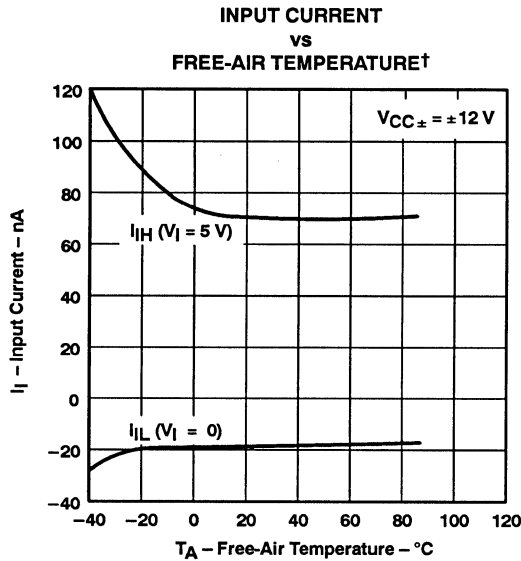


Figure 9

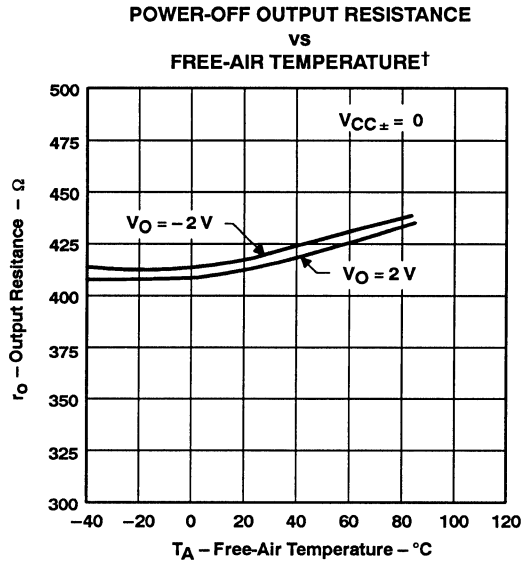


Figure 10

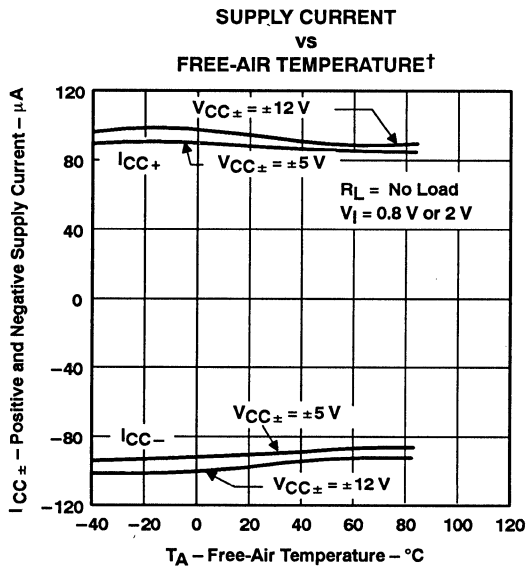


Figure 11

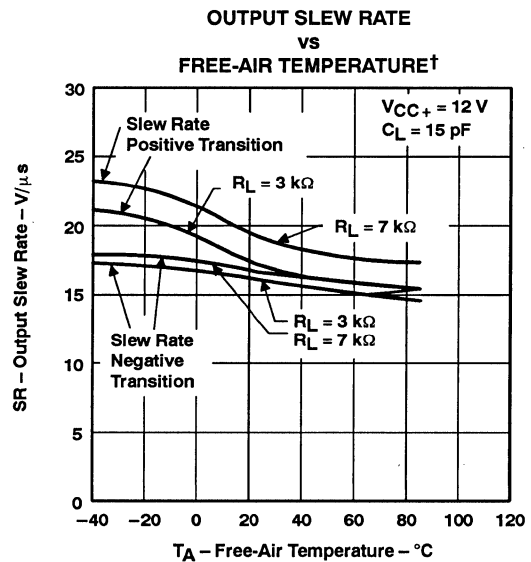


Figure 12

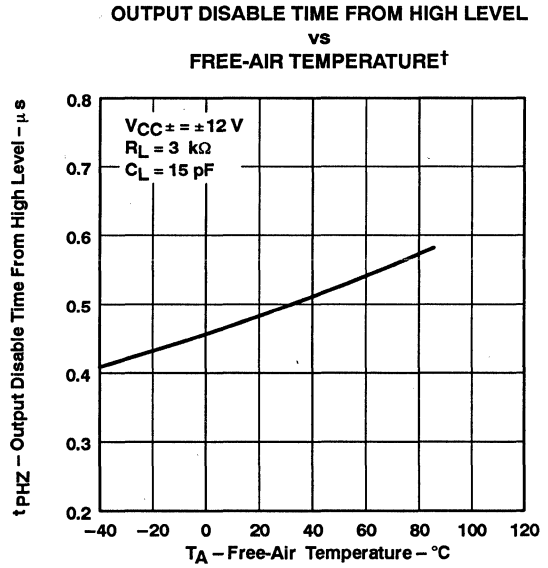
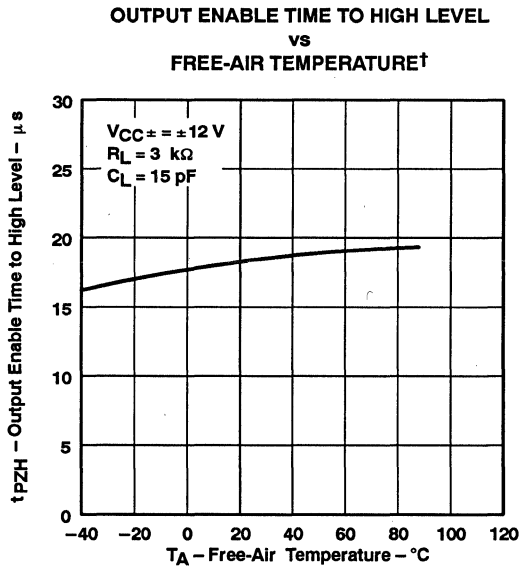
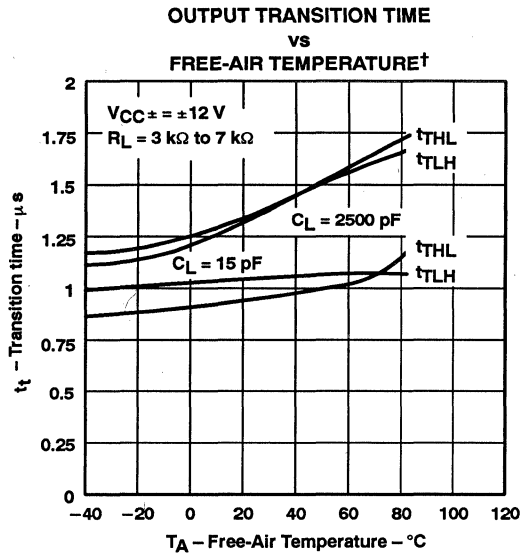
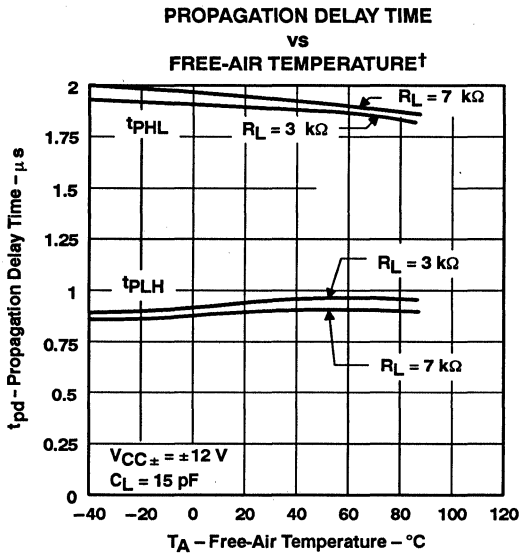
† Only the 0°C to 70°C portion of the curves applies to the SN75C198.



# SN65C198, SN75C198 QUAD LOW-POWER LINE DRIVERS

SLLS051A - D3472, JULY 1990 - REVISED MARCH 1993

## TYPICAL CHARACTERISTICS



† Only the 0°C to 70°C portion of the curves applies to the SN75C198.

TYPICAL CHARACTERISTICS

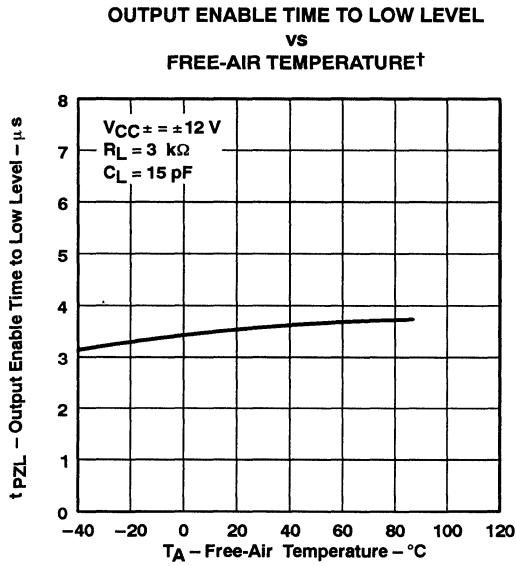


Figure 17

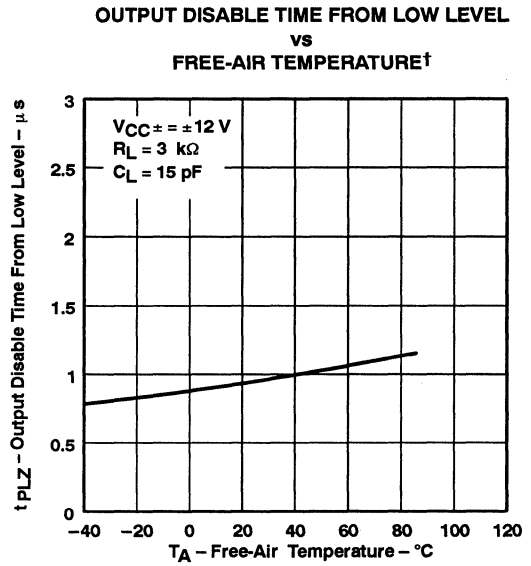


Figure 18

† Only the 0°C to 70°C portion of the curves applies to the SN75C198.



# SN75ALS199 QUAD DIFFERENTIAL LINE RECEIVER WITH 3-STATE OUTPUTS

SLLS046B – D3204, JANUARY 1989 – REVISED MARCH 1993

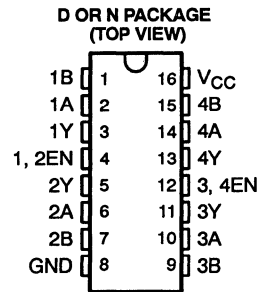
- Meets CCITT Recommendations V.10, V.11, X.26, and X.27
- –7 V to 7 V Common-Mode Input Voltage Range With 300-mV Sensitivity
- 3-State TTL-Compatible Outputs
- High Input Impedance . . . 12 k $\Omega$  Min
- Input Hysteresis . . . 120 mV Typ
- Single 5-V Supply Operation
- Low Supply Current Requirement  
35 mA Max
- Improved Speed and Power Consumption Compared to MC3486

## description

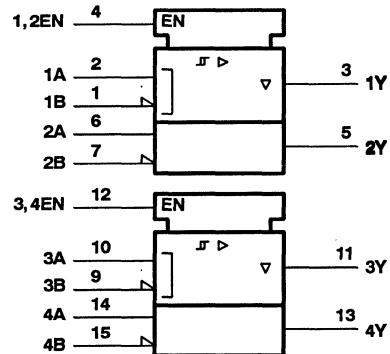
The SN75ALS199 is a monolithic quad line receiver with 3-state outputs designed using advanced low-power Schottky technology. This technology provides combined improvements in bar design, tooling production, and wafer fabrication, providing significantly less power consumption and permitting much higher data throughput than other designs. The device meets the specification of CCITT Recommendation V.10, V.11, X.26 and X.27.

The SN75ALS199 features 3-state outputs that permit direct connection to a bus-organized system with a fail-safe design that ensures the outputs will always be high if the inputs are open. The device is optimized for balanced multipoint bus transmission at rates up to 10 megabits per second. The input features high input impedance, input hysteresis for increased noise immunity, and an input sensitivity of  $\pm 300$  mV over a common-mode input voltage range of  $\pm 7$  V. It also features an active-high enable function for each of two receiver pairs. The SN75ALS199 is designed for optimum performance when used with the SN75ALS194 quadruple differential line driver.

The SN75ALS199 is characterized for operation from 0°C to 70°C.

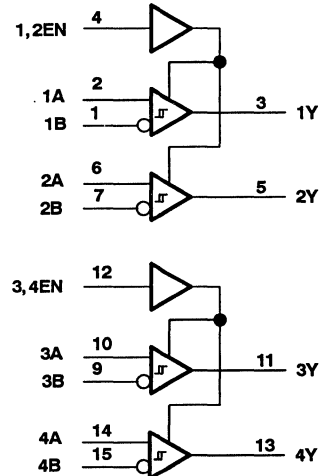


## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

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2-839

# SN75ALS199 QUAD DIFFERENTIAL LINE RECEIVER WITH 3-STATE OUTPUTS

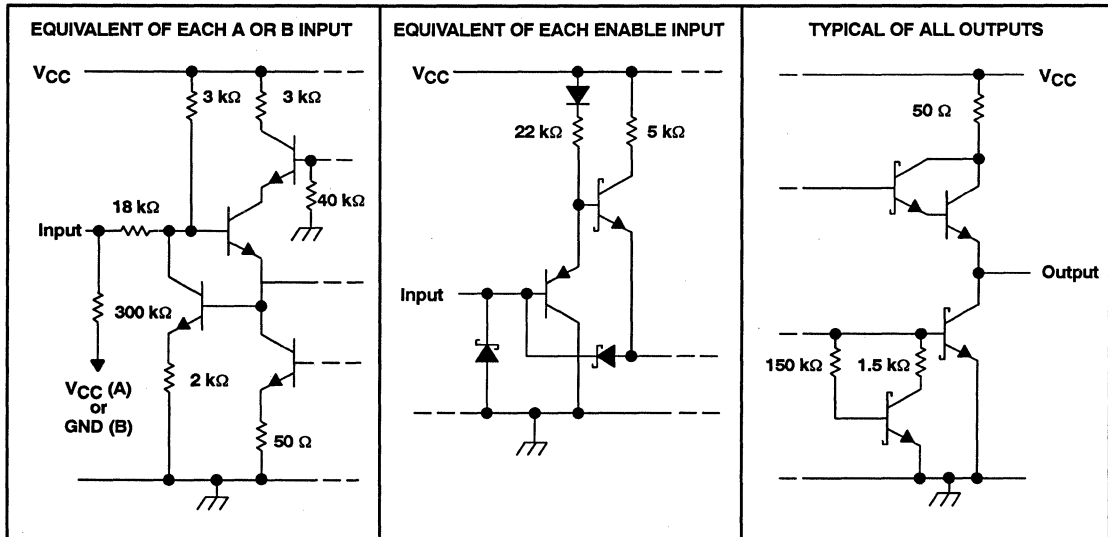
SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

FUNCTION TABLE  
(each receiver)

DIFFERENTIAL INPUTS A-B	EN	OUTPUT Y
$V_{ID} \geq 0.3 \text{ V}$	H	H
$-0.3 \text{ V} < V_{ID} < 0.3 \text{ V}$	H	?
$V_{ID} \leq -0.3 \text{ V}$	H	L
X	L	Z
Open	H	H

H = high level, L = low level, X = irrelevant,  
? = indeterminate, Z = high impedance (off)

## schematics of inputs and outputs



**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B – D3204, JANUARY 1989 – REVISED MARCH 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, A or B inputs, $V_I$	$\pm 15$ V
Differential input voltage (see Note 2)	$\pm 15$ V
Enable input voltage	7 V
Low-level output current	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	- 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to network ground terminal.  
 2. Differential-input voltage is measured at the noninverting input with respect to the corresponding inverting input.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Differential input voltage, $V_{ID}$			$\pm 12$	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
High-level output current, $I_{OH}$			- 400	$\mu\text{A}$
Low-level output current, $I_{OL}$			16	mA
Operating free-air temperature, $T_A$	0		70	°C



**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V <sub>T+</sub>	Positive-going threshold voltage					300	mV
V <sub>T-</sub>	Negative-going threshold voltage			-300‡			mV
V <sub>hys</sub>	Hysteresis (V <sub>T+</sub> - V <sub>T-</sub> )				120		mV
V <sub>IK</sub>	Enable-input clamp voltage	I <sub>I</sub> = -18 mA				-1.5	V
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 300 mV,	I <sub>OH</sub> = -400 μA	2.7	3.6		V
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -300 mV	I <sub>OL</sub> = 8 mA			0.45	V
			I <sub>OL</sub> = 16 mA			0.5	
I <sub>OZ</sub>	High-impedance-state output current	V <sub>IL</sub> = 0.8 V, V <sub>ID</sub> = -3 V, V <sub>O</sub> = 2.7 V				20	μA
		V <sub>IL</sub> = 0.8 V, V <sub>IO</sub> = 3 V, V <sub>O</sub> = 0.5 V				-20	
I <sub>I</sub>	Line input current	Other input at 0 V, See Note 3	V <sub>I</sub> = 15 V		0.7	1.2	mA
			V <sub>I</sub> = -15 V		-1	-1.7	
I <sub>IH</sub>	High-level enable-input current		V <sub>IH</sub> = 2.7 V			20	μA
			V <sub>IH</sub> = 5.25 V			100	
I <sub>IL</sub>	Low-level enable-input current	V <sub>IL</sub> = 0.4 V				-100	μA
	Input resistance			12	18		kΩ
I <sub>OS</sub>	Short-circuit output current§	V <sub>ID</sub> = 3 V,	V <sub>O</sub> = 0	-15	-78	-130	mA
I <sub>CC</sub>	Supply current	Outputs disabled			22	35	mA

† All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

‡ The algebraic convention, in which the less positive limit is designated minimum, is used in this data sheet for threshold voltage levels only.

§ Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

NOTE 3: Refer to CCITT Recommendations V.10 and V.11 for exact conditions.

**switching characteristics, V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C**

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	V <sub>ID</sub> = 0 V to 3 V,	C <sub>L</sub> = 15 pF,		15	22	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output	See Figure 2			15	22	
t <sub>PZH</sub>	Output enable time to high level	C <sub>L</sub> = 15 pF,	See Figure 3		13	25	ns
t <sub>PZL</sub>	Output enable time to low level				11	25	
t <sub>PHZ</sub>	Output disable time from high level	C <sub>L</sub> = 15 pF,	See Figure 3		13	25	ns
t <sub>PLZ</sub>	Output disable time from low level				15	22	



**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

**PARAMETER MEASUREMENT INFORMATION**

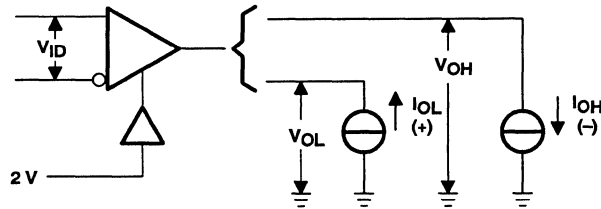
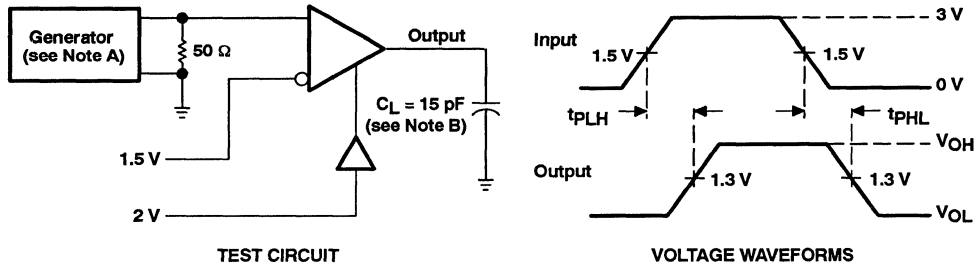


Figure 1.  $V_{OH}$ ,  $V_{OL}$  Test Circuit



TEST CIRCUIT

VOLTAGE WAVEFORMS

NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1$  MHz, duty cycle  $\leq 50\%$ ,  $Z_0 = 50 \Omega$ ,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns.

B.  $C_L$  includes probe and jig capacitance.

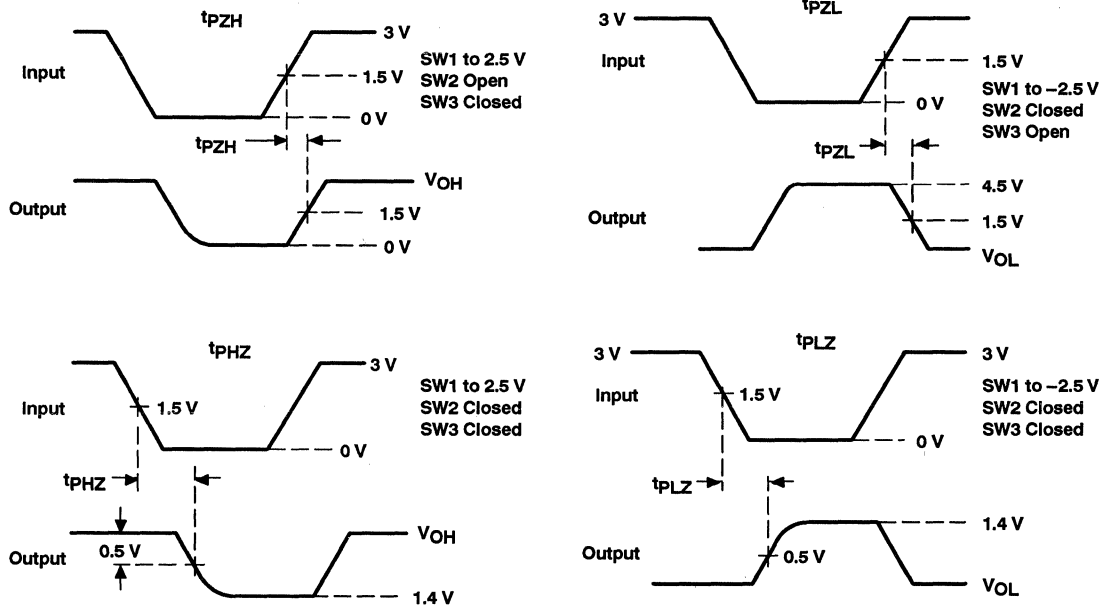
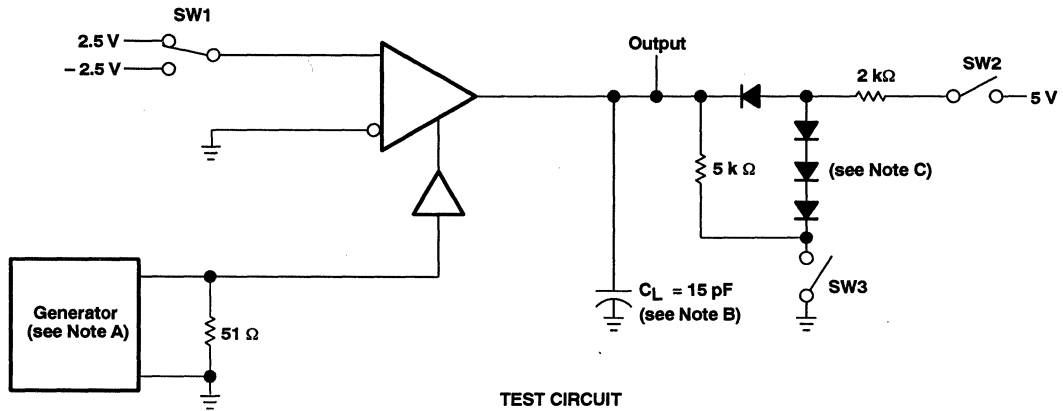
Figure 2. Test Circuit and Voltage Waveforms



**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

**PARAMETER MEASUREMENT INFORMATION**



**VOLTAGE WAVEFORMS**

- NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, duty cycle  $\leq$  50%,  $Z_O = 50$   $\Omega$ ,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns.  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.

**Figure 3. Test Circuit and Voltage Waveforms**

**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

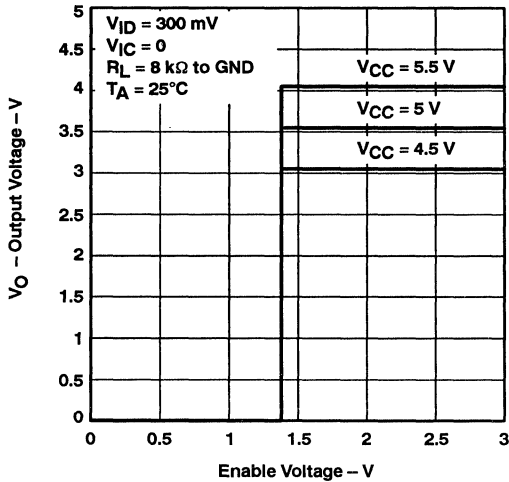


Figure 4

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

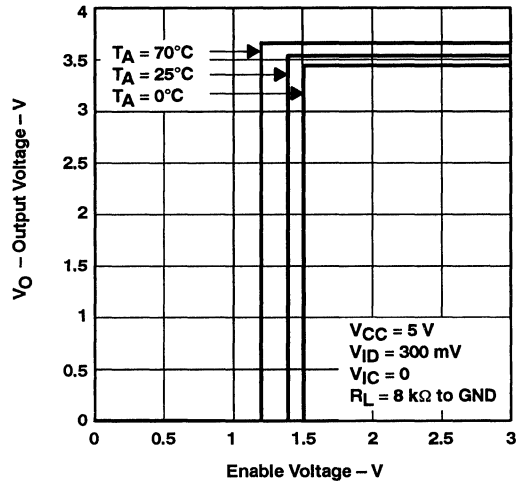


Figure 5

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

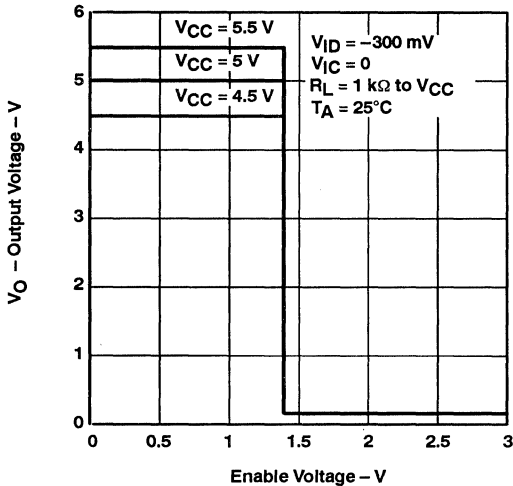


Figure 6

**OUTPUT VOLTAGE  
vs  
ENABLE VOLTAGE**

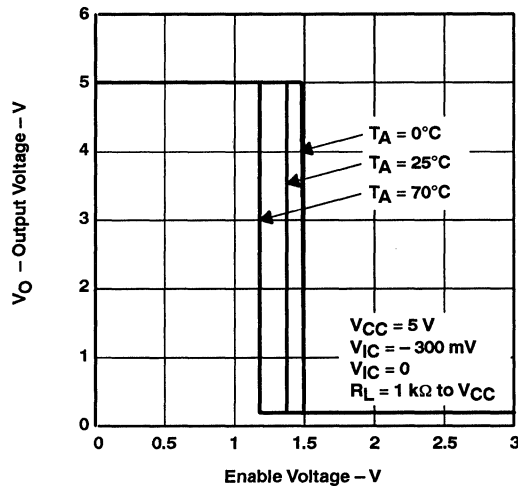


Figure 7



**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

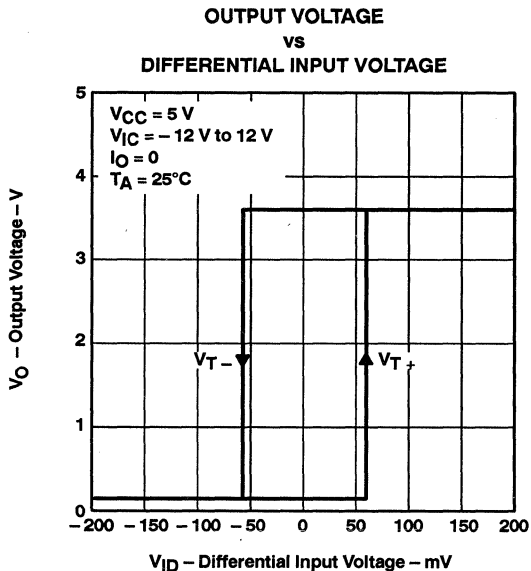


Figure 8

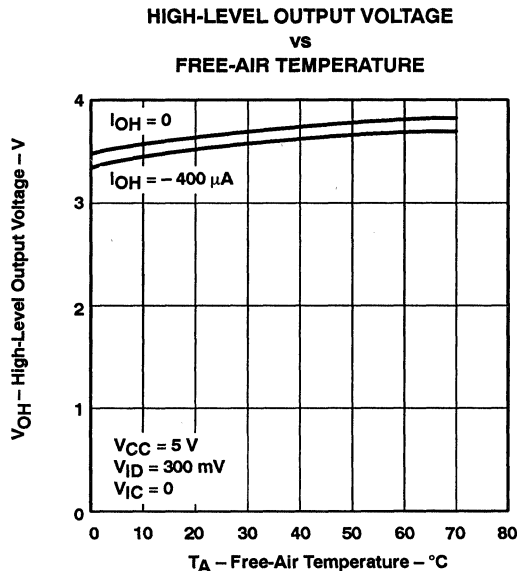


Figure 9

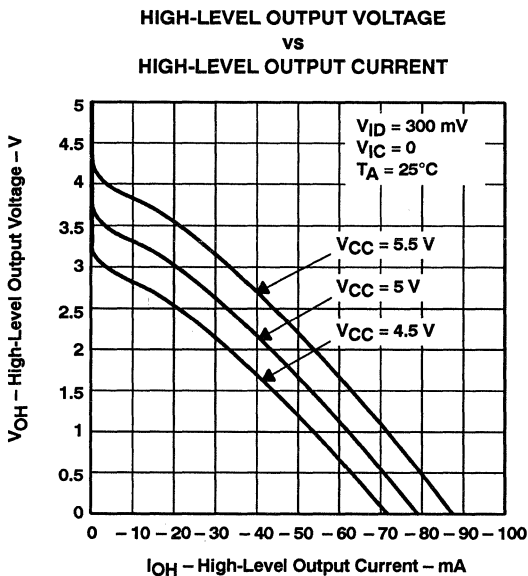


Figure 10

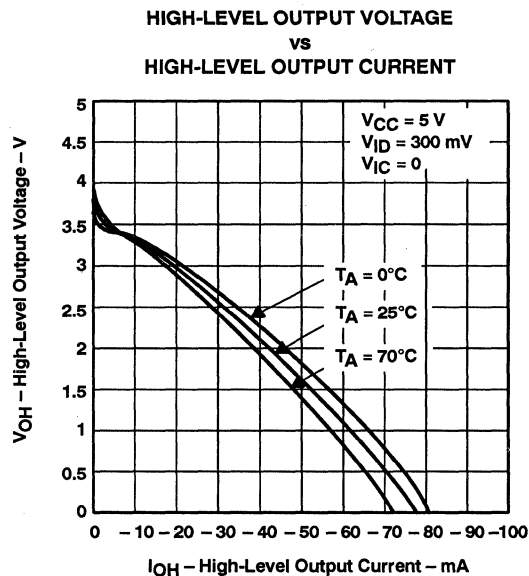


Figure 11

**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

**LOW-LEVEL OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**

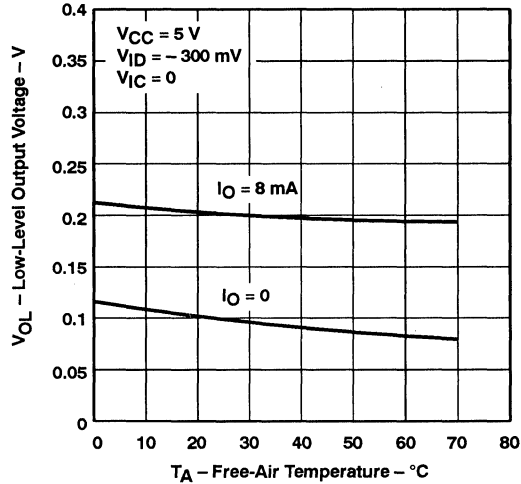


Figure 12

**LOW-LEVEL OUTPUT VOLTAGE**  
**vs**  
**LOW-LEVEL OUTPUT CURRENT**

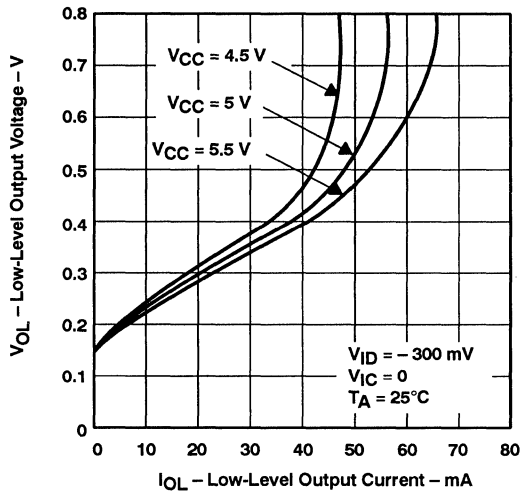


Figure 13

**LOW-LEVEL OUTPUT VOLTAGE**  
**vs**  
**LOW-LEVEL OUTPUT CURRENT**

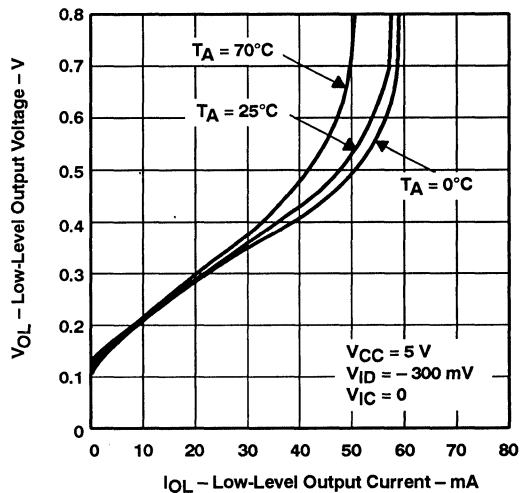


Figure 14



**SN75ALS199**  
**QUAD DIFFERENTIAL LINE RECEIVER**  
**WITH 3-STATE OUTPUTS**

SLLS046B - D3204, JANUARY 1989 - REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

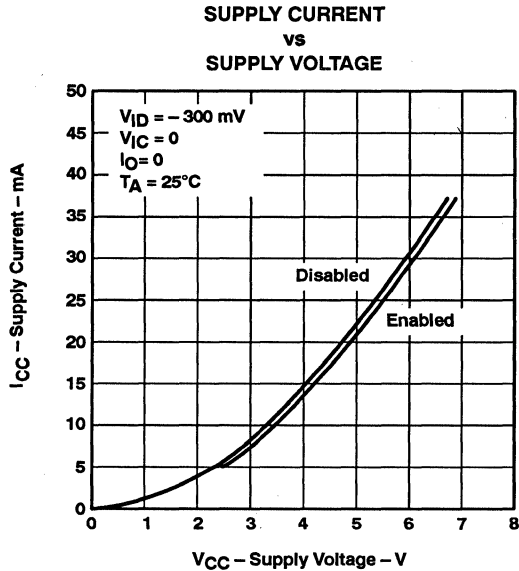


Figure 15

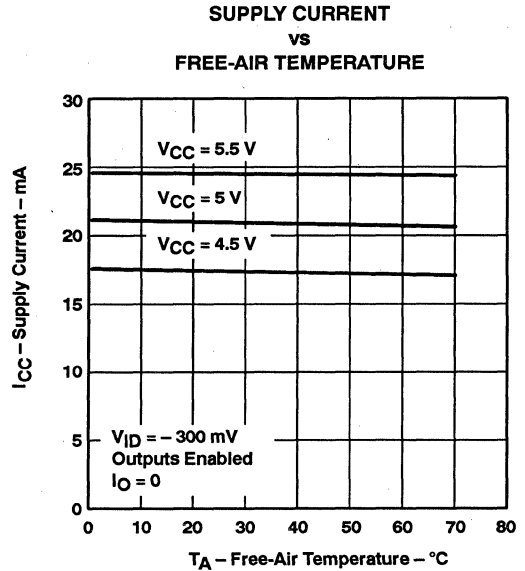


Figure 16

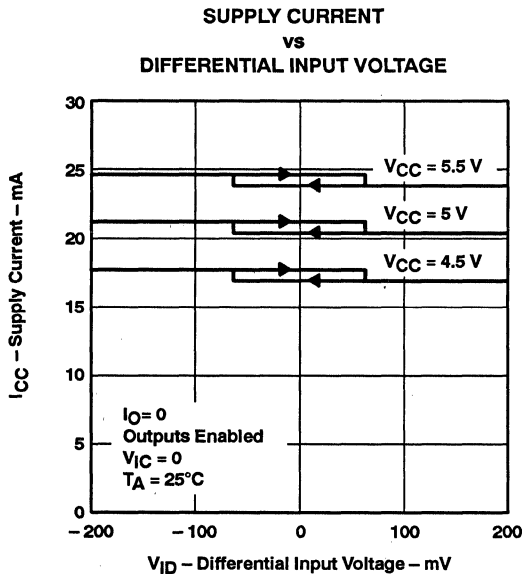


Figure 17

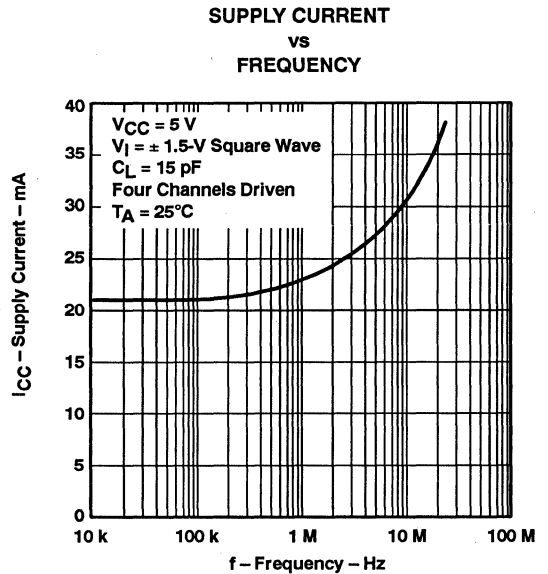


Figure 18



TYPICAL CHARACTERISTICS

INPUT RESISTANCE  
 VS  
 FREE-AIR TEMPERATURE

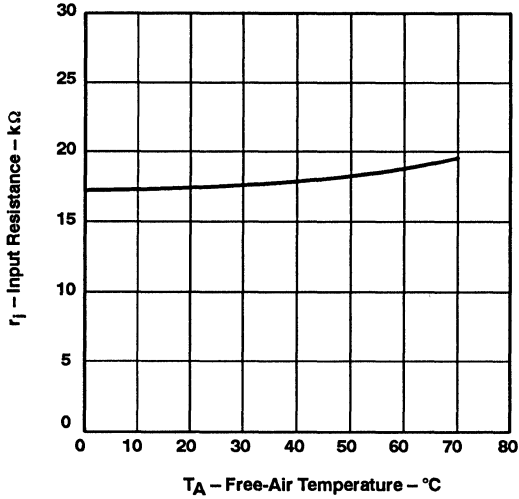


Figure 19

INPUT CURRENT  
 VS  
 INPUT VOLTAGE TO GND

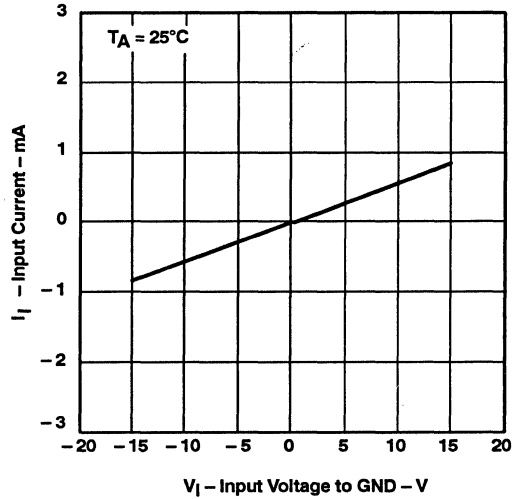


Figure 20

SWITCHING CHARACTERISTICS  
 VS  
 FREE-AIR TEMPERATURE

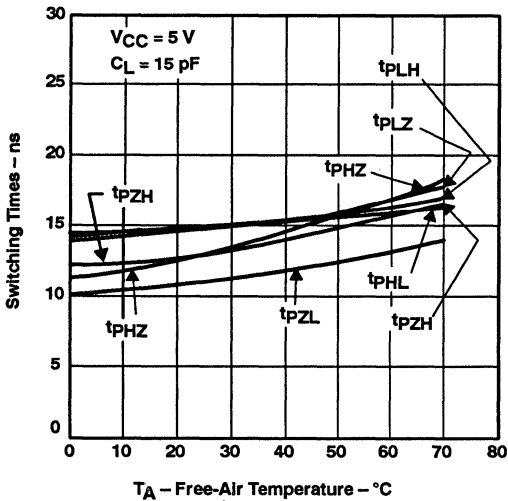


Figure 21

PROPAGATION DELAY TIME  
 VS  
 SUPPLY VOLTAGE

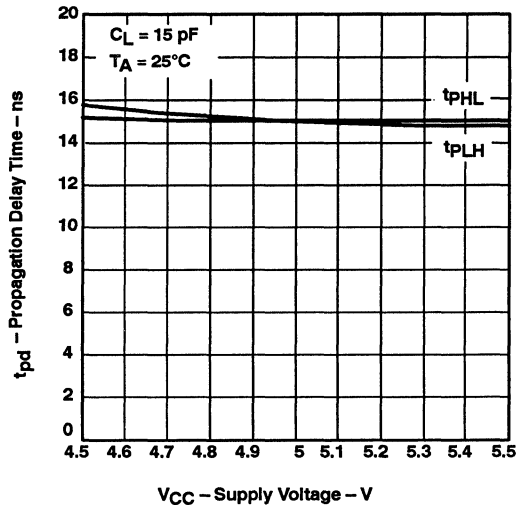


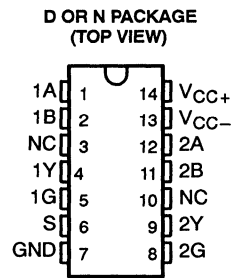
Figure 22



# SN75207, SN75207B DUAL SENSE AMPLIFIER FOR MOS MEMORIES OR DUAL HIGH-SENSITIVITY LINE RECEIVERS

SLLS069A - D1314, JULY 1973 - REVISED JANUARY 1993

- Plug-In Replacement for SN75107A and SN75107B With Improved Characteristics
- $\pm 10$ -mV Input Sensitivity
- TTL Compatible
- Standard Supply Voltages . . .  $\pm 5$  V
- Differential Input Common-Mode Voltage Range of  $\pm 3$  V
- Strobe Inputs for Channel Selection
- Totem-Pole Outputs
- SN75207B Has Diode-Protected Input Stage for Power-Off Condition
- Sense Amplifier for MOS Memories
- Dual Comparator
- High-Sensitivity Line Receiver

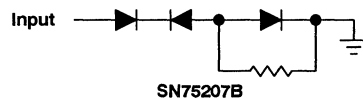
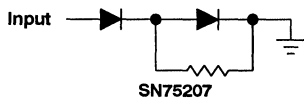


NC - No internal connection

## description

The SN75207 and SN75207B are pin-for-pin replacements for the SN75107A and SN75107B, respectively. The improved input sensitivity makes them more suitable for MOS memory sense amplifiers and can result in faster memory cycles. Improved sensitivity also makes them more useful in line receiver applications by allowing use of longer transmission line lengths. The '207 and '207B each features a TTL-compatible active-pullup output.

The essential difference between the SN75207 and SN75207B can be seen in the schematics. Input protection diodes are in series with the collectors of the differential-input transistors of the SN75207B. These diodes are useful in certain party-line systems that may have multiple  $V_{CC+}$  power supplies and may be operated with some of the  $V_{CC+}$  supplies turned off. In such a system, if a supply is turned off and allowed to go to ground, the equivalent input circuit connected to that supply would be as follows:



This would be a problem in specific systems that might have the transmission lines biased to some potential greater than 1.4 V.

These devices are characterized for operation from 0°C to 70°C.

FUNCTION TABLE

DIFFERENTIAL INPUTS A-B	STROBES		OUTPUT Y
	G	S	
$V_{ID} \geq 10$ mV	X	X	H
$-10$ mV $< V_{ID} < 10$ mV	X	L	H
	L	X	H
$V_{ID} \leq -10$ mV	H	H	Indeterminate
	X	L	H
	L	X	H
	H	H	L

H = high level, L = low level, X = irrelevant

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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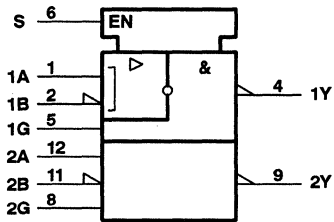
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# SN75207, SN75207B DUAL SENSE AMPLIFIER FOR MOS MEMORIES OR DUAL HIGH-SENSITIVITY LINE RECEIVERS

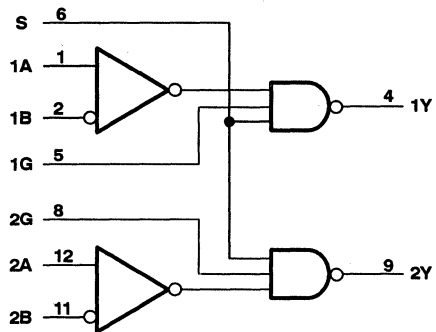
SLLS069A - D1314, JULY 1973 - REVISED JANUARY 1993

## logic symbol†

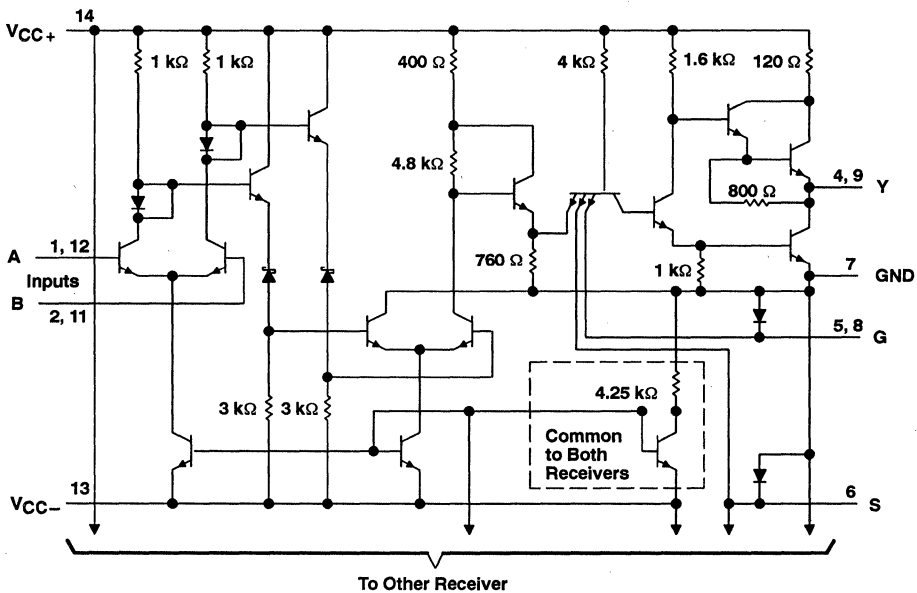


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## schematic (each receiver)



Resistor values shown are normal.

TEXAS  
INSTRUMENTS

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**SN75207, SN75207B**  
**DUAL SENSE AMPLIFIER FOR MOS MEMORIES**  
**OR DUAL HIGH-SENSITIVITY LINE RECEIVERS**

SLLS069A - D1314, JULY 1973 - REVISED JANUARY 1993

**design characteristics**

The '207 and '207B line receivers/sense amplifiers are TTL-compatible dual circuits intended for use in high-speed data-transmission systems or MOS memory systems. They are designed to detect low-level differential signals in the presence of common-mode noise and variations of temperature and supplies. The dc specifications reflect worst-case conditions of temperature, supply voltages, and input voltages.

The input common-mode voltage range is  $\pm 3$  V. This is adequate for application in most systems. In systems with requirements for greater common-mode voltage range, input attenuators may be used to decrease the noise to an acceptable level at the receiver-input terminals.

The circuits feature individual strobe inputs for each channel and a strobe input common to both channels for logic versatility. The strobe inputs are tested to ensure 400 mV of dc noise margin when interfaced with Series 54/74 TTL.

The circuits feature high input impedance and low input currents, which induce very little loading on the transmission line. This makes these devices especially useful in party-line systems. The excellent input sensitivity (3 mV typical) is particularly important when data is to be detected at the end of a long transmission line and the amplitude of the data has deteriorated due to cable losses. The circuits are designed to detect input signals of 10-mV (or greater) amplitude and convert the polarity of the signal into appropriate TTL-compatible output logic levels.

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC+}$ (see Note 1)	7 V
Supply voltage, $V_{CC-}$ (see Note 1)	-7 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 6$ V
Common-mode input voltage, $V_{IC}$ (see Note 3)	$\pm 5$ V
Strobe input voltage	5.5 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to ground terminal.  
 2. Differential input voltage values are at the noninverting (A) terminal with respect to the inverting (B) terminal.  
 3. Common-mode input voltage is the average of the voltages at the A and B inputs.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1050 mW	9.2 mW/°C	736 mW



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## recommended operating conditions (see Note 4)

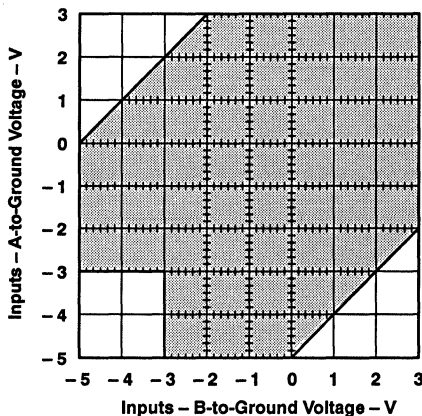
	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$	4.75	5	5.25	V
Supply voltage, $V_{CC-}$	-4.75	-5	-5.25	V
High-level differential input voltage, $V_{IDH}$ (see Note 5)	0.01		5	V
Low-level differential input voltage, $V_{IDL}$	-5†		-0.01	V
Common-mode input voltage, $V_{IC}$ (see Notes 5 and 6)	-3†		3	V
Input voltage, any differential input to ground (see Note 5)	-5†		3	V
High-level input voltage at strobe inputs, $V_{IH(S)}$	2		5.5	V
Low-level input voltage at strobe inputs, $V_{IL(S)}$	0		0.8	V
Low-level output current, $I_{OL}$			-16	mA
Operating free-air temperature, $T_A$	0		70	°C

† The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for logic voltage levels only.

NOTES: 4. When using only one channel of the line receiver, the strobe G of the unused channel should be grounded and at least one of the differential inputs of the unused receiver should be terminated at some voltage between -3 V and 3 V.

5. The recommended combinations of input voltages fall within the shaded area of the figure shown.

6. The common-mode voltage may be as low as -4 V provided that the more positive of the two inputs is not more negative than -3 V.



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**electrical characteristics over recommended free-air temperature range (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT	
I <sub>IH</sub>	High-level input current	'207	V <sub>CC±</sub> = ± 5.25 V	V <sub>ID</sub> = 5 V	30	75	μA	
		'207B		V <sub>ID</sub> = -5 V	30	75		
I <sub>IL</sub>	Low-level input current	'207	V <sub>CC±</sub> = ± 5.25 V	V <sub>ID</sub> = -5 V		-10	μA	
		'207B		V <sub>ID</sub> = 5 V		-10		
I <sub>IH</sub>	High-level input current into 1G or 2G	V <sub>CC±</sub> = ± 5.25 V, V <sub>IH(S)</sub> = 2.4 V				40	μA	
		V <sub>CC±</sub> = ± 5.25 V, V <sub>IH(S)</sub> = ± 5.25 V				1	mA	
I <sub>IL</sub>	Low-level input current into 1G or 2G	V <sub>CC±</sub> = ± 5.25 V, V <sub>IL(S)</sub> = 0.4 V				-1.6	mA	
I <sub>IH</sub>	High-level input current into S	V <sub>CC±</sub> = ± 5.25 V, V <sub>IH(S)</sub> = 2.4 V				80	μA	
		V <sub>CC±</sub> = ± 5.25 V, V <sub>IH(S)</sub> = ± 5.25 V				2	mA	
I <sub>IL</sub>	Low-level input current into S	V <sub>CC±</sub> = ± 5.25 V, V <sub>IL(S)</sub> = 0.4 V				-3.2	mA	
V <sub>OH</sub>	High-level output voltage	V <sub>CC±</sub> = ± 4.75 V, I <sub>OH</sub> = -400 μA,	V <sub>IL(S)</sub> = 0.8 V, V <sub>IC</sub> = -3 V to 3 V	V <sub>IDH</sub> = 10 mV,	2.4		V	
V <sub>OL</sub>	Low-level output voltage	V <sub>CC±</sub> = ± 4.75 V, I <sub>OL</sub> = 16 mA,	V <sub>IH(S)</sub> = 2 V, V <sub>IC</sub> = -3 V to 3 V	V <sub>IDL</sub> = -10 mV,		0.4	V	
I <sub>OH</sub>	High-level output current	V <sub>CC±</sub> = ± 4.75 V, V <sub>OH</sub> = ± 5.25 V				400	μA	
I <sub>OS</sub>	Short-circuit output current‡	V <sub>CC±</sub> = ± 5.25 V				-18	-70	mA
I <sub>CCH+</sub>	Supply current from V <sub>CC+</sub> , outputs high	V <sub>CC±</sub> = ± 5.25 V, T <sub>A</sub> = 25°C				18	30	mA
I <sub>CCH-</sub>	Supply current from V <sub>CC-</sub> , outputs high	V <sub>CC±</sub> = ± 5.25 V, T <sub>A</sub> = 25°C				-8.4	-15	mA

† All typical values are at V<sub>CC+</sub> = 5 V, V<sub>CC-</sub> = -5 V, T<sub>A</sub> = 25°C.

‡ Not more than one output should be shorted at a time.

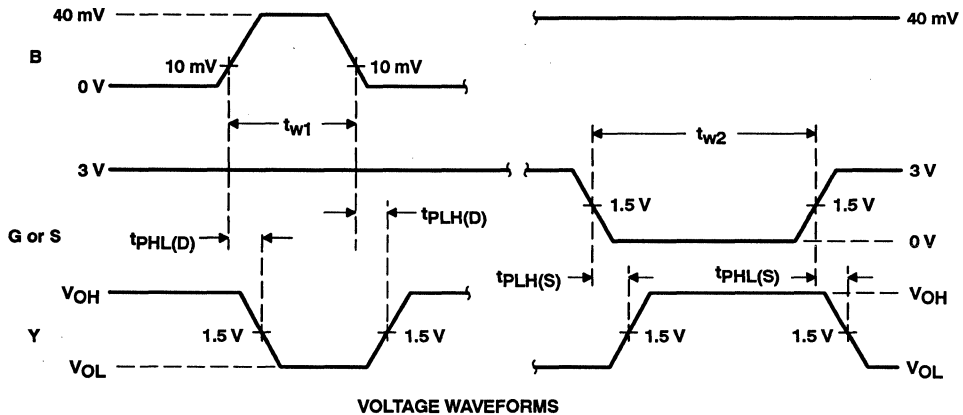
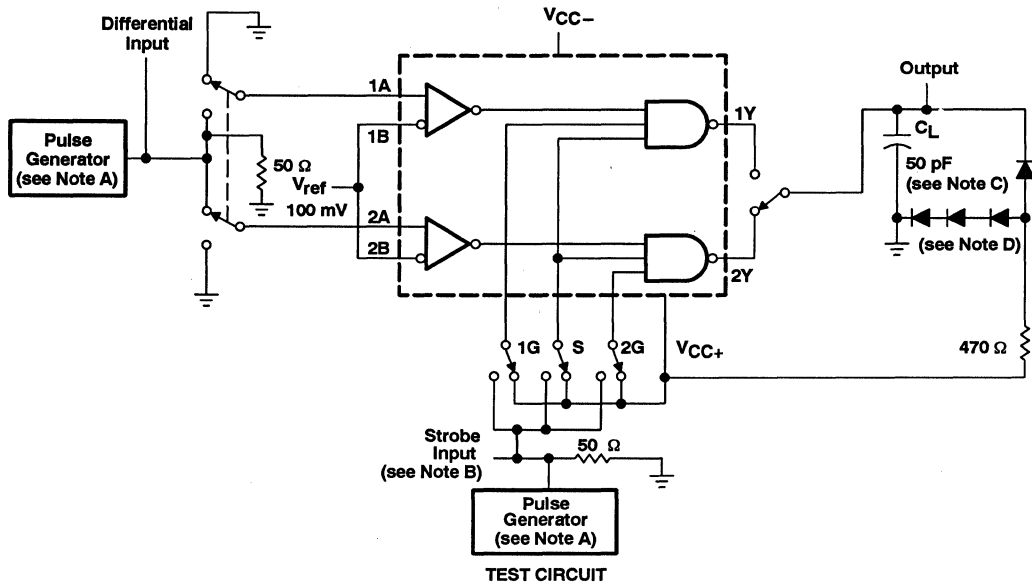
**switching characteristics, V<sub>CC+</sub> = 5 V, V<sub>CC-</sub> = -5 V, T<sub>A</sub> = 25°C**

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
t <sub>PLH(D)</sub>	Propagation delay time, low-to-high level output, from differential inputs A and B	R <sub>L</sub> = 470 Ω, C <sub>L</sub> = 50 pF, See Figure 1		35	ns
t <sub>PHL(D)</sub>	Propagation delay time, high-to-low level output, from differential inputs A and B			20	ns
t <sub>PLH(S)</sub>	Propagation delay time, low-to-high level output, from strobe input G or S			17	ns
t <sub>PHL(S)</sub>	Propagation delay time, high-to-low level output, from strobe input G or S			17	ns



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**PARAMETER MEASUREMENT INFORMATION**



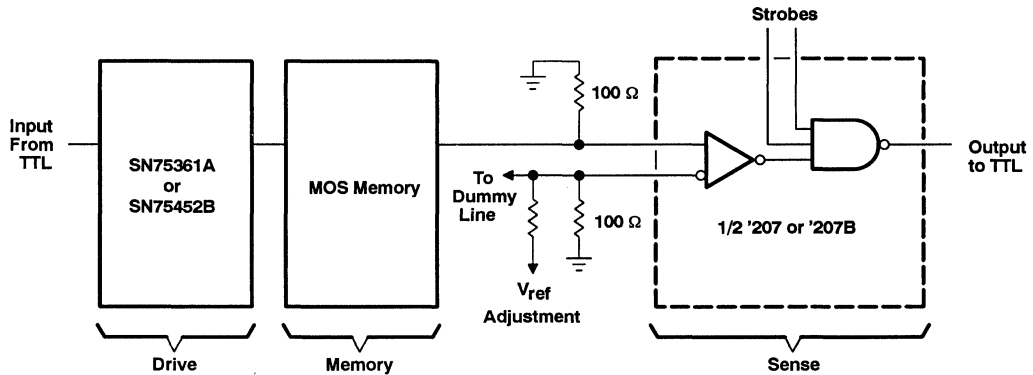
- NOTES: A. The pulse generators have the following characteristics:  $Z_0 = 50 \Omega$ ,  $t_r \leq 5 \text{ ns}$ ,  $t_f \leq 5 \text{ ns}$ ,  $t_{w1} = 500 \text{ ns}$  with PRR = 1 MHz,  $t_{w2} = 1 \mu\text{s}$  with PRR = 500 kHz.  
 B. Strobe input pulse is applied to Strobe 1G when inputs 1A-1B are being tested, to Strobe S when inputs 1A-1B or 2A-2B are being tested, and to Strobe 2G when inputs 2A-2B are being tested.  
 C.  $C_L$  includes probe and jig capacitance.  
 D. All diodes are 1N916.

**Figure 1. Test Circuit and Voltage Waveforms**

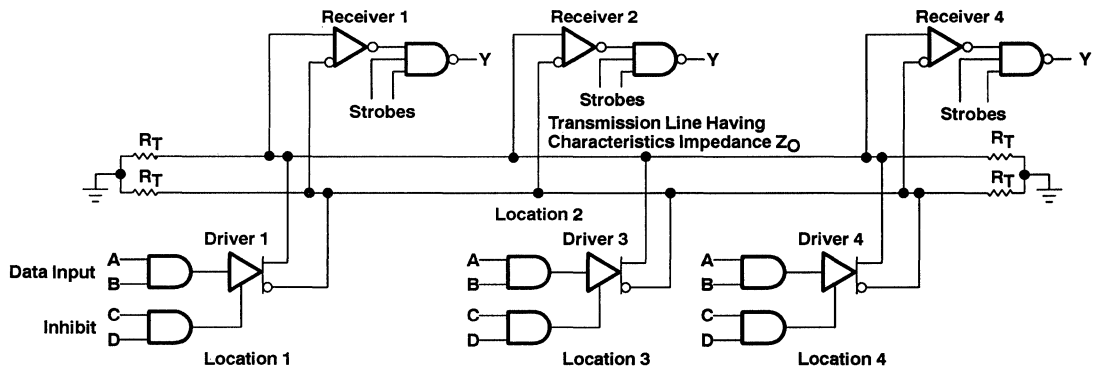
**SN75207, SN75207B**  
**DUAL SENSE AMPLIFIER FOR MOS MEMORIES**  
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SLLS069A - D1314, JULY 1973 - REVISED JANUARY 1993

**APPLICATION INFORMATION**



**Figure 2. Mos Memory Sense Amplifier**



Receivers are '207 or '207B; drivers are SN55109A, SN75109A, SN55110A, SN75110A, or SN75112.

**Figure 3. Data-Bus or Parity-Line System**

**PRECAUTIONS:** When only one receiver in a package is being used, at least one of the differential inputs of the unused receiver should be terminated at some voltage between  $-3\text{ V}$  and  $3\text{ V}$ , preferably at GND. Failure to do so will cause improper operation of the unit being used because of common bias circuitry for the current sources of the two receivers. Strobe G of the unused channel should be grounded.



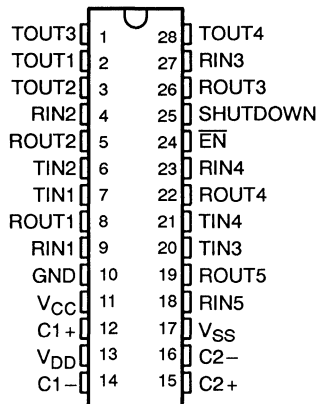
# SN75LBC241

## LOW-POWER LinBiCMOS™ MULTIPLE DRIVERS AND RECEIVERS

SLLS137C – D4027, MAY 1992 – REVISED APRIL 1993

- Operates With Single 5-V Power Supply
- Meets EIA/TIA-232-E, 1991 Specifications (Revision of EIA-232-D and CCITT V.28)
- Improved Performance Replacement for MAX241
- ESD Protection on EIA/TIA-232 Pins Exceeds 6 kV ESD Per MIL-STD-883C, Method 3015
- Operates at Data Rates Up to 100 kb/s Over a 3-Meter Cable
- Low Power Shutdown Mode:  $\leq 1 \mu\text{A}$  Typ
- LinBiCMOS™ Process Technology
- 4 Drivers and 5 Receivers
- $\pm 30\text{-V}$  Input Levels
- 3-State TTL/CMOS Receiver Outputs
- $\pm 9\text{-V}$  Output Swing With a 5-V Supply
- Applications
  - EIA-232 Interface
  - Battery-Powered Systems
  - Terminals
  - Modems
  - Computers

DW PACKAGE  
(TOP VIEW)



### description

The SN75LBC241† is a low-power LinBiCMOS™ line interface device containing four independent drivers and five receivers. It is designed to provide a plug-in replacement for the Maxim MAX241 with improved ESD protection and other key performance specifications. The SN75LBC241 provides a capacitive charge-pump voltage generator to produce EIA/TIA-232 voltage levels from a 5-V supply. The charge-pump oscillator frequency is 20 kHz. Each receiver converts EIA/TIA-232 inputs to 5-V TTL/CMOS levels. The receivers have a typical threshold of 1.2 V and a typical hysteresis of 0.5 V, and can accept  $\pm 30\text{-V}$  inputs. Each driver converts TTL/CMOS input levels into EIA/TIA-232 levels.

The SN75LBC241 includes a receiver 3-state control line and a low-power shutdown control line. Whenever, the  $\overline{\text{EN}}$  line is high, the receiver outputs are placed in a high-impedance state. When  $\overline{\text{EN}}$  is low, normal operation is enabled.

The shutdown mode reduces power dissipation to less than  $5 \mu\text{W}$  typical. In this mode, receiver outputs are high impedance, driver outputs are turned off, and the charge-pump circuit is turned off. When the SHUTDOWN line is high, the shutdown mode is enabled. When the SHUTDOWN line is low, normal operation is enabled.

This device has been designed to conform to standard EIA/TIA-232-E, 1991 and CCITT V.28 specifications.

The SN75LBC241 has been designed using LinBiCMOS™ technology and cells contained in the Texas Instruments LinASIC™ library. Use of LinBiCMOS™ circuitry increases latch-up immunity in this device over an all-CMOS design. The SN75LBC241 is characterized for operation from  $0^\circ\text{C}$  to  $70^\circ\text{C}$ .

† Patent pending

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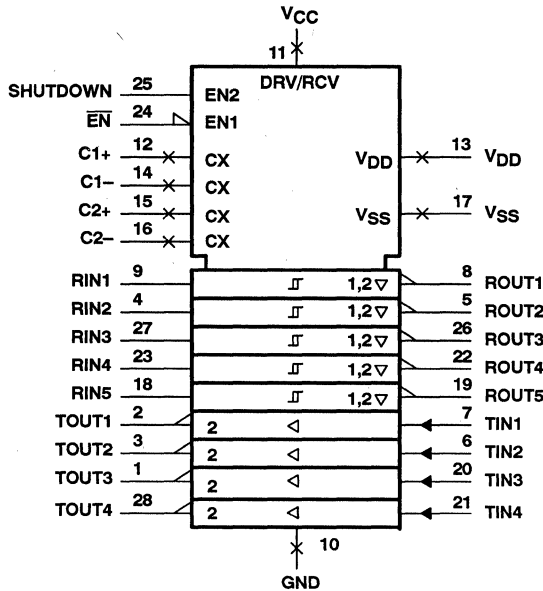




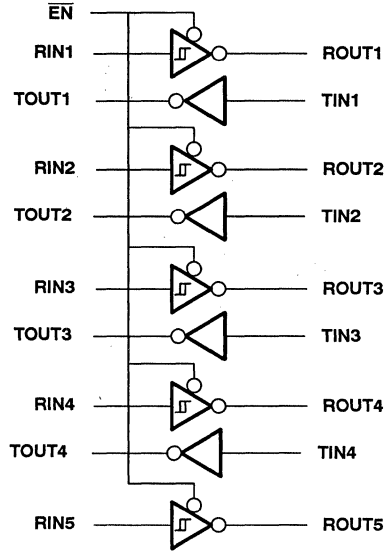
# SN75LBC241 LOW-POWER LinBiCMOS™ MULTIPLE DRIVERS AND RECEIVERS

SLLS137C – D4027, MAY 1992 – REVISED APRIL 1993

## logic symbol†



## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std. 91-1984 and IEC Publication 617-12.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Input supply voltage range, $V_{CC}$ (see Note 1)	-0.3 V to 6 V
Positive output supply voltage range, $V_{DD}$	$V_{CC} - 0.3$ V to 15 V
Negative output supply voltage range, $V_{SS}$	0.3 V to -15 V
Input voltage range: Driver	-0.3 V to $V_{CC} + 0.3$ V
Receiver	$\pm 30$ V
Output voltage range: TOUT	$V_{SS} - 0.3$ V to $V_{DD} + 0.3$ V
ROUT	-0.3 V to $V_{CC} + 0.3$ V
Short-circuit duration: TOUT	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1348 mW	10.8 mW/°C	863 mW

# SN75LBC241

## LOW-POWER LinBiCMOS™ MULTIPLE DRIVERS AND RECEIVERS

SLLS137C – D4027, MAY 1992 – REVISED APRIL 1993

### recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.5	5	5.5	V
High-level input voltage, $V_{IH}$	TIN	2			V
	$\overline{EN}$ , SHUTDOWN	2.4			
Low-level input voltage, $V_{IL}$	TIN, $\overline{EN}$ , SHUTDOWN	0.8			V
External charge-pump capacitor	C1, C2 (see Figure 1)	1			$\mu\text{F}$
	C3, C4 (see Figure 1)	1			
External charge-pump capacitor voltage rating	C1, C3 (see Figure 1)	6.3			V
	C2, C4 (see Figure 1)	16			
Receiver input voltage, RIN		$\pm 30$			V
Operating free-air temperature, $T_A$		0	70		$^{\circ}\text{C}$

### electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{OH}$ High-level output voltage	TOUT	$R_L = 3\text{ k}\Omega$ to GND,	See Note 2	5	9		V
	ROUT	$I_{OH} = -1\text{ mA}$		3.5			
$V_{OL}$ Low-level output voltage	TOUT	$R_L = 3\text{ k}\Omega$ to GND,	See Note 3		$-9^{\ddagger}$	$-5$	V
	ROUT	$I_{OL} = 3.2\text{ mA}$		0.4			
$V_{T+}$ Receiver positive-going input threshold voltage	RIN	$V_{CC} = 5\text{ V}$ ,	$T_A = 25^{\circ}\text{C}$		1.7	2.4	V
$V_{T-}$ Receiver negative-going input threshold voltage	RIN	$V_{CC} = 5\text{ V}$ ,	$T_A = 25^{\circ}\text{C}$	0.8	1.2		V
$V_{hys}$ Input hysteresis ( $V_{T+} - V_{T-}$ )	RIN	$V_{CC} = 5\text{ V}$			0.5	1	V
$r_i$ Receiver input resistance	RIN	$V_{CC} = 5\text{ V}$ ,	$T_A = 25^{\circ}\text{C}$	3	5	7	$\text{k}\Omega$
$r_o$ Output resistance	TOUT	$V_{DD} = V_{SS} = V_{CC} = 0$ , $V_O = \pm 2\text{ V}$		300			$\Omega$
$I_{OS}$ Short circuit output current <sup>§</sup>	TOUT	$V_{CC} = 5.5\text{ V}$ ,	$V_O = 0$	$\pm 10$			mA
$I_{IS}$ Short circuit input current	TIN	$V_I = 0$				200	$\mu\text{A}$
$I_{CC}$ Supply current			$V_{CC} = 5.5\text{ V}$ , All output open	$T_A = 25^{\circ}\text{C}$ ,	4	8	mA
			All outputs open, Shutdown pin high	$T_A = 25^{\circ}\text{C}$ ,	1	10	$\mu\text{A}$

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^{\circ}\text{C}$ .

‡ The algebraic convention, in which the least positive (most negative) value is designated minimum, is used in this data sheet for logic voltage levels only.

§ Not more than one output should be shorted at one time.

NOTES: 2. Total  $I_{OH}$  drawn from TOUT1, TOUT2, TOUT3, TOUT4 and  $V_{DD}$  pins should not exceed 12 mA.

3. Total  $I_{OL}$  drawn from TOUT1, TOUT2, TOUT3, TOUT4 and  $V_{SS}$  pins should not exceed  $-12\text{ mA}$ .

### switching characteristics, $V_{CC} = 5\text{ V}$ , $T_A = 25^{\circ}\text{C}$

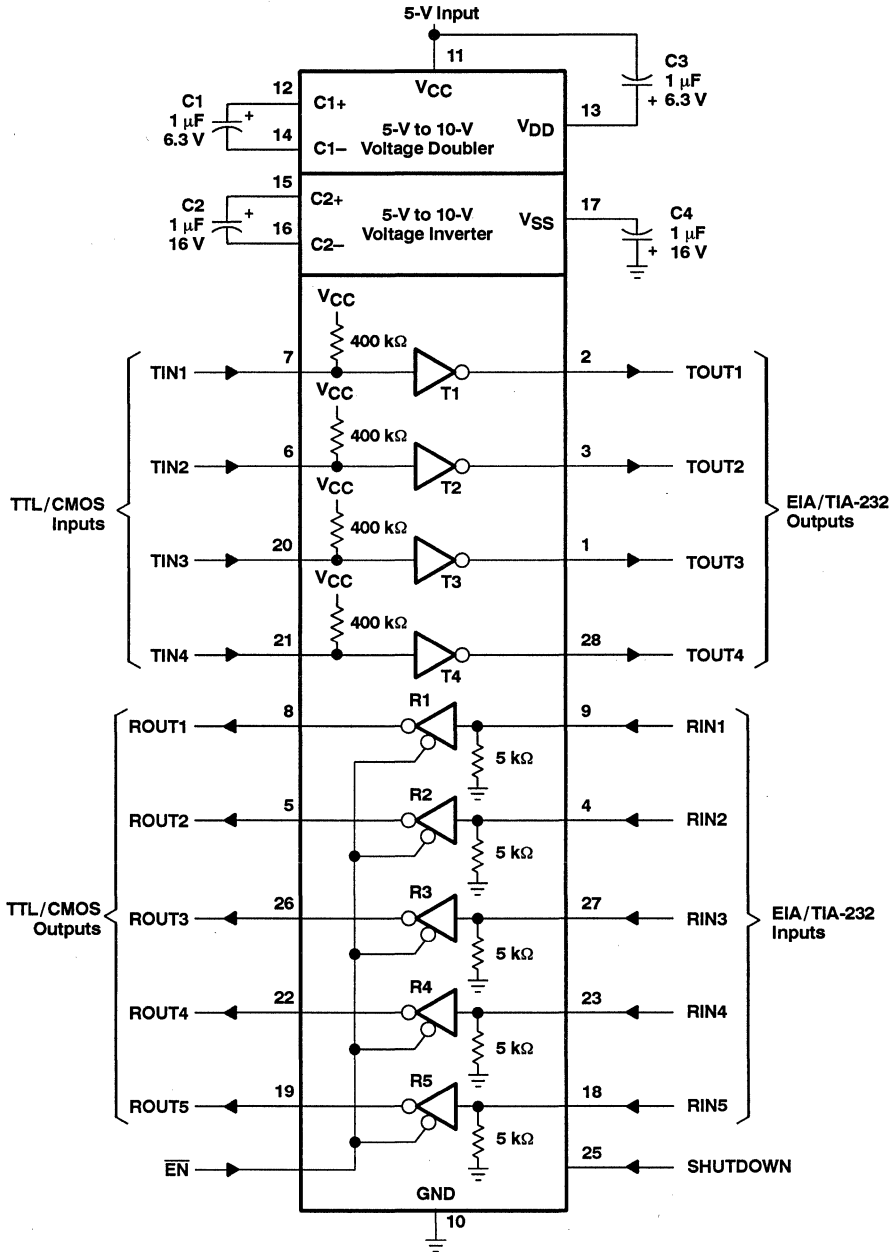
PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH(R)}$	Receiver propagation delay time, low-to-high-level output	See Figure 2		500			ns
$t_{PHL(R)}$	Receiver propagation delay time, high-to-low-level output	See Figure 2		500			ns
$t_{PZH}$	Receiver output enable time to high level	See Figure 5		100			ns
$t_{PZL}$	Receiver output enable time to low level	See Figure 5		100			ns
$t_{PHZ}$	Receiver output disable time from high level	See Figure 5		50			ns
$t_{PLZ}$	Receiver output disable time from low level	See Figure 5		50			ns
SR	Driver slew rate	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ ,	See Figure 4			30	$\text{V}/\mu\text{s}$
$SR(tr)$	Driver transition region slew rate	$C_L = 2500\text{ pF}$ ,	See Figure 4	4	6		$\text{V}/\mu\text{s}$



**SN75LBC241**  
**LOW-POWER LinBiCMOS™ MULTIPLE DRIVERS AND RECEIVERS**

SLLS137C - D4027, MAY 1992 - REVISED APRIL 1993

**APPLICATION INFORMATION**

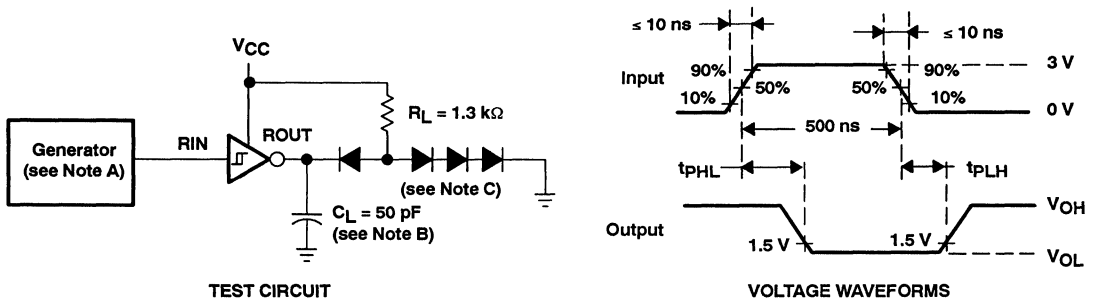


**Figure 1. Typical Operating Circuit**

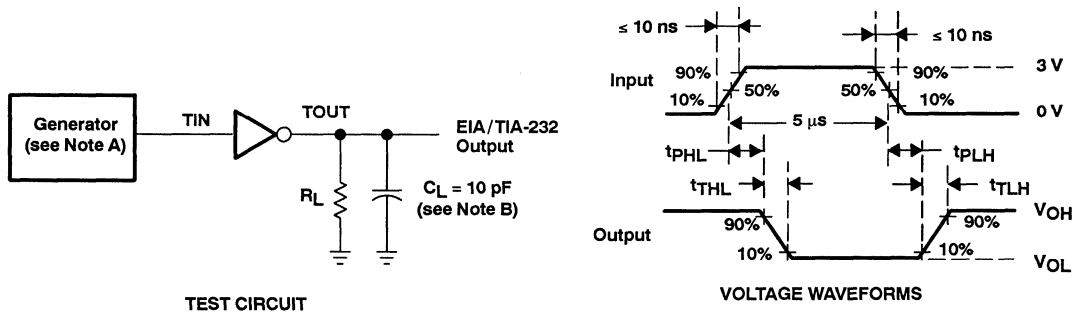


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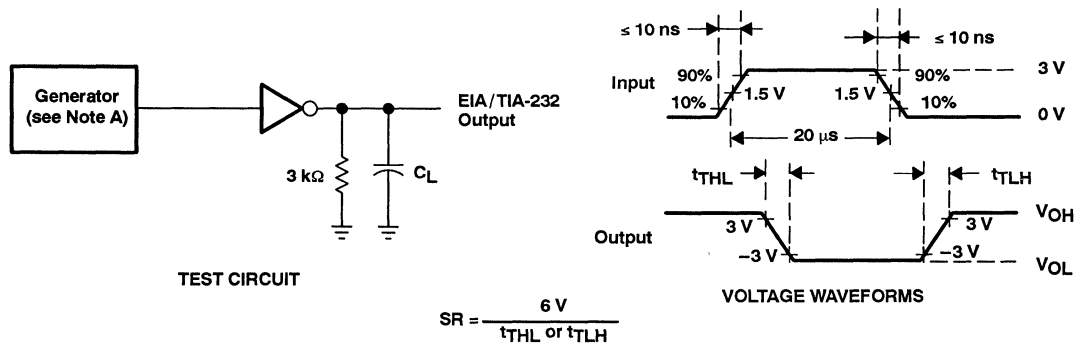
**PARAMETER MEASUREMENT INFORMATION**



**Figure 2. Receiver Test Circuit and Waveforms for  $t_{PHL}$  and  $t_{PLH}$  Measurement**



**Figure 3. Driver Test Circuit and Waveforms for  $t_{PHL}$  and  $t_{PLH}$  Measurement (5- $\mu$ s Input)**



**Figure 4. Test Circuit and Waveforms for  $t_{THL}$  and  $t_{TLH}$  Measurement (20- $\mu$ s Input)**

- NOTES: A. The pulse generator has the following characteristics:  $Z_O = 50\ \Omega$ , duty cycle  $\leq 50\%$ .  
 B.  $C_L$  includes probe and jig capacitance.  
 C. All diodes are 1N3064 or equivalent.

SN75LBC241  
LOW-POWER LinBiCMOS™ MULTIPLE DRIVERS AND RECEIVERS

SLLS137C - D4027, MAY 1992 - REVISED APRIL 1993

PARAMETER MEASUREMENT INFORMATION

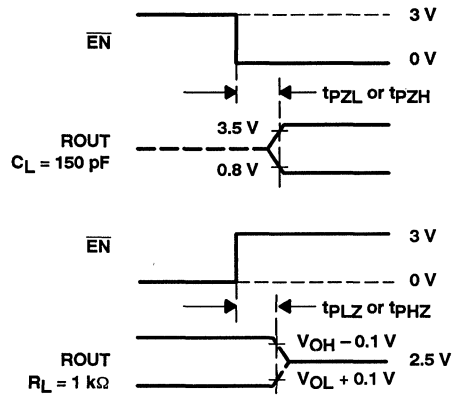


Figure 5. Receiver Output Enable and Disable Timing

# SN75LBC976 9-CHANNEL DIFFERENTIAL TRANSCEIVER

SLLS133B – D4042, AUGUST 1992 – REVISED FEBRUARY 1993

- **9 Differential Channels for the Data and Control Paths of the Differential Small Computer Systems Interface (SCSI) and Intelligent Peripheral Interface (IPI-2)**
- **Each Transceiver Meets EIA-RS-485 and ISO 8482:1987(E) Standards**
- **Packaged in Shrink Small-Outline Package With 25-mil Terminal Pitch**
- **Designed to Operate at 10 Million Transfers Per Second**
- **Low Disabled Supply Current  
1.4 mA Typical**
- **Thermal Shutdown Protection**
- **Power-Up/Power-Down Glitch Protection**
- **Positive and Negative Output Current Limiting**
- **Open-Circuit Fail-Safe Receiver Design**

## description

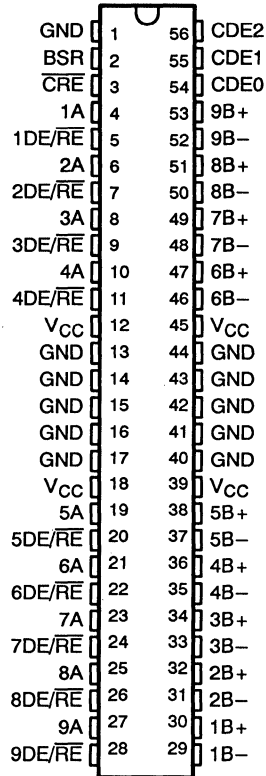
The SN75LBC976 is a nine-channel differential transceiver based on the 75LBC176 LinASIC™ cell. Use of Texas Instruments' LinBiCMOS™† process technology allows the power reduction necessary to integrate nine differential transceivers. On-chip enabling logic makes this device applicable for the data path (eight data bits plus parity) and the control path (nine bits) for both the Small Computer Systems Interface (SCSI) and the Intelligent Peripheral Interface (IPI-2) standard data interfaces. This part is ESD Class 1 (A) and Class 2 (B) per MIL-STD-883, Method 3015.

The SN75LBC976 is packaged in a shrink small-outline package (DL) with improved thermal characteristics using heat sink pins. This package is ideal for low-profile, space-restricted applications such as hard disk drives.

The switching speed and testing of the SN75LBC976 is sufficient to transfer data over the data bus at 10 million transfers per second. Each of the nine channels conforms to the requirements of the EIA-RS-485 and ISO 8482:1987(E) standards referenced by ANSI X3.129-1986 (IPI), ANSI X3.131-1986 (SCSI-1), and the proposed SCSI-2 and SCSI-3 standards.

The SN75LBC976 is characterized for operation from 0°C to 70°C.

DL PACKAGE  
(TOP VIEW)



Pins 13 through 17 and 40 through 44 are connected together to the package lead frame and signal ground.

† Patent pending  
LinASIC and LinBiCMOS are trademarks of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



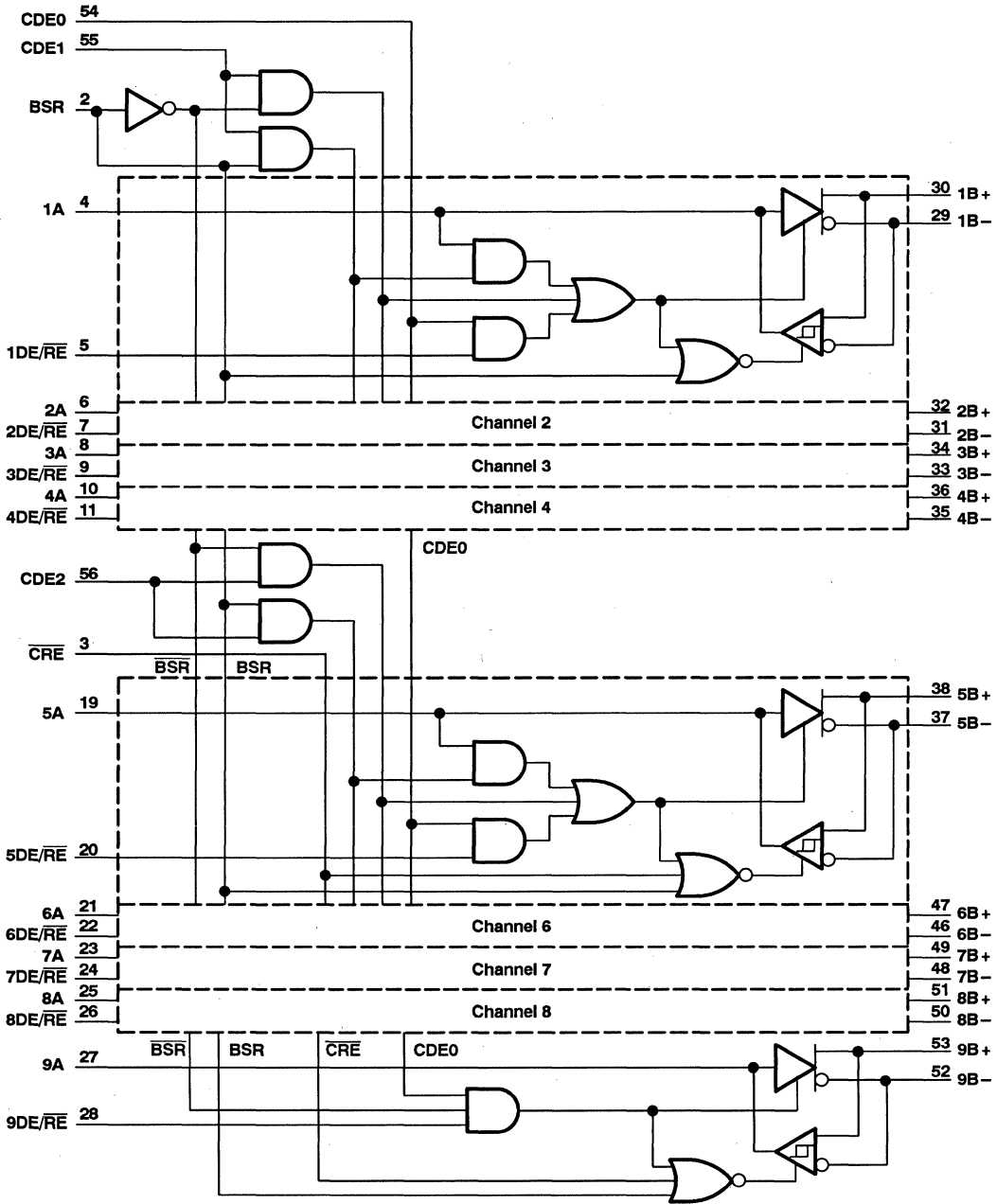
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# SN75LBC976 9-CHANNEL DIFFERENTIAL TRANSCIEVER

SLLS133B - D4042, AUGUST 1992 - REVISED FEBRUARY 1993

## logic diagram (positive logic)†

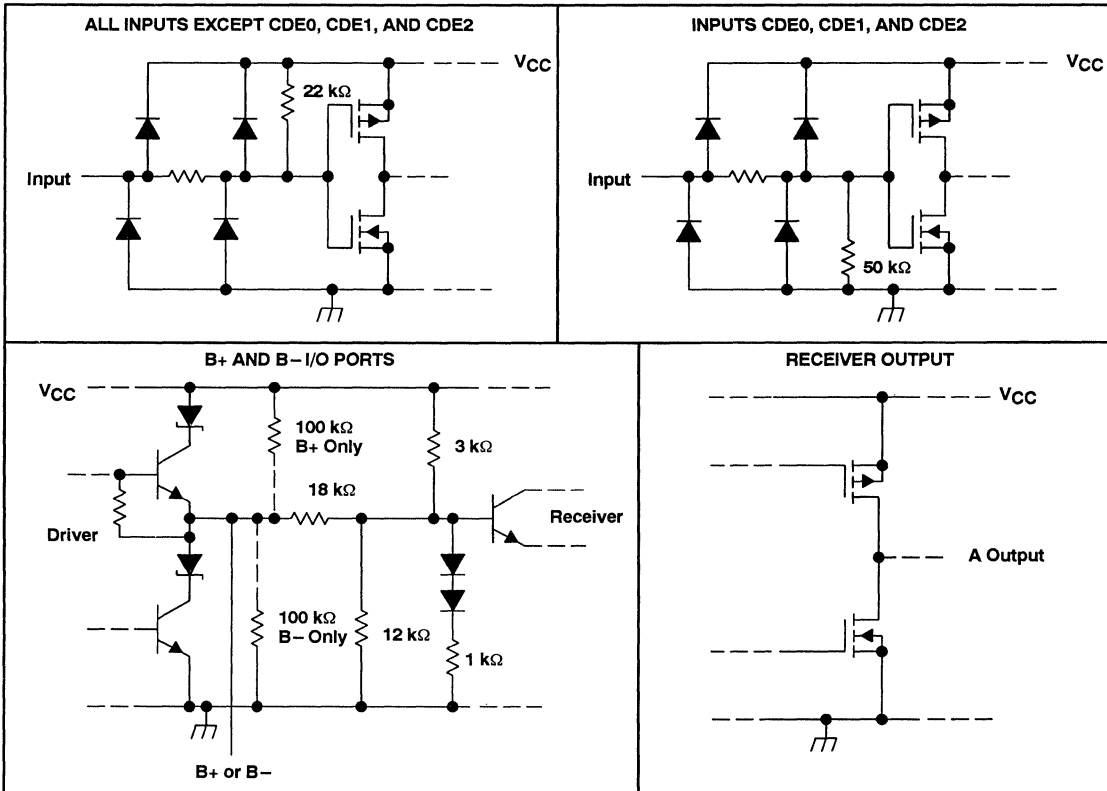


† For additional logic diagrams, see Application Information.

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## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage range, $V_{CC}$ (see Note 1)	–0.3 V to 7 V
Bus voltage range	–10 V to 15 V
Data I/O and control (A-side) voltage range	–0.3 V to 7 V
Continuous power dissipation	internally limited
Operating free-air temperature range	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to the GND terminal.



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## 9-CHANNEL DIFFERENTIAL TRANSCEIVER

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### recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Voltage at any bus terminal (separately or common-mode), $V_O$ , $V_I$ , or $V_{IC}$	B+ or B-	12			V
		-7			
High-level input voltage, $V_{IH}$	All except B+ and B-	2			V
Low-level input voltage, $V_{IL}$	All except B+ and B-	0.8			V
High-level output current, $I_{OH}$	B+ or B-	-60			mA
	A	-8			mA
Low-level output current, $I_{OL}$	B+ or B-	60			mA
	A	8			mA
Operating free-air temperature, $T_A$		0	70		°C

### device electrical characteristics over recommended ranges of operating conditions

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$I_{IH}$	High-level input current	BSR, A, DE/ $\overline{RE}$ , and CRE	See Figure 3	$V_{IH} = 5$ V	-100		$\mu$ A
		CDE0, CDE1, and CDE2			100		
$I_{IL}$	Low-level input current	BSR, A, DE/ $\overline{RE}$ , and CRE	See Figure 3	$V_{IL} = 0$ V	-200		$\mu$ A
		CDE0, CDE1, and CDE2			50		
$I_{CC}$	Supply current	All drivers and receivers disabled	BSR and CDE0 at 5 V, Other inputs at 0 V	1.4		3	mA
		All receivers enabled	No load, $V_{ID} = 5$ V, All other inputs at 0 V	29		45	mA
		All drivers enabled	BSR at 0 V, No load, All other inputs at 5 V	4.8		10	mA
$C_o$	Bus port capacitance	B+ or B-		16			pF
$C_{pd}$	Power dissipation capacitance‡	One driver		460			pF
		One receiver		50			pF

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ$ C.

‡  $C_{pd}$  determines the no-load dynamic current consumption,  $I_S = C_{pd} V_{CC} f + I_{CC}$ .

### driver electrical characteristics over recommended ranges of operating conditions (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$ V_{OD} $	Differential output voltage	$V_{test} = -7$ V to 12 V, See Figure 2		1	2		V
$I_{OS}$	Output short-circuit current	See Figure 1				$\pm 250$	mA
$I_{OZ}$	High-impedance state output current	See receiver input current					



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**receiver electrical characteristics over recommended ranges of operating conditions (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage	$V_{ID} = 200$ mV, See Figure 3 $I_{OH} = -8$ mA,	2.5			V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200$ mV, See Figure 3 $I_{OL} = 8$ mA,			0.8	V
$V_{T+}$	Positive-going threshold voltage	$I_{OH} = -8$ mA, See Figure 3			0.2	V
$V_{T-}$	Negative-going threshold voltage	$I_{OL} = 8$ mA, See Figure 3	-0.2			V
$V_{hys}$	Receiver input hysteresis ( $V_{T+} - V_{T-}$ )			45		mV
$I_I$	Receiver input current, B+ and B-	$V_I = 12$ V, $V_{CC} = 5$ V, Other input at 0 V, See Figure 3		0.7	1	mA
		$V_I = 12$ V, $V_{CC} = 0$ V, Other input at 0 V, See Figure 3		0.8	1	mA
		$V_I = -7$ V, $V_{CC} = 5$ V, Other input at 0 V, See Figure 3	-0.5		-0.8	mA
		$V_I = -7$ V, $V_{CC} = 0$ V, Other input at 0 V, See Figure 3	-0.4		-0.8	mA
$I_{OZ}$	High-impedance-state output current	See Figure 3	$V_O = GND$		-100	$\mu$ A
			$V_O = V_{CC}$		50	

**driver switching characteristics over recommended operating conditions (see Figure 4) (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{dD}$	Differential delay time, high-to-low-level output ( $t_{dDH}$ ) or low-to-high-level output ( $t_{dDL}$ )		7.6		19.6	ns
		$V_{CC} = 5$ V, $T_A = 25^\circ$ C	9.1		17.1	
		$V_{CC} = 5$ V, $T_A = 70^\circ$ C	11.5		19.5	
$t_{sk(lim)}$	Skew limit, the maximum difference in propagation delay times between any two drivers on any two devices				12	ns
		$V_{CC} = 5$ V, See Note 2			8	
$t_{sk(p)}$	Pulse skew ( $ t_{dDL} - t_{dDH} $ )			0	6	ns
$t_t$	Transistion time ( $t_r$ or $t_f$ )			10		ns

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ$ C.

NOTE 2: This specification applies to any  $5^\circ$ C band within the operating temperature range.



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receiver switching characteristics over recommended operating conditions (see Figure 5) (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{pd}$ Propagation delay time, high-to-low-level output ( $t_{PLH}$ ) or low-to-high-level output ( $t_{PHL}$ )		21.5		33	ns
	$V_{CC} = 5\text{ V}, T_A = 25^\circ\text{C}$	22.6		31.6	
	$V_{CC} = 5\text{ V}, T_A = 70^\circ\text{C}$	23.4		32.4	
$t_{sk(lim)}$ Skew limit, the maximum difference in propagation delay times between any two drivers on any two devices				12	ns
	$V_{CC} = 5\text{ V},$ See Note 2			9	
$t_{sk(p)}$ Pulse skew ( $ t_{PHL} - t_{PLH} $ )			2	6	ns
$t_t$ Transition time ( $t_r$ or $t_f$ )			3		ns

† All typical values are at  $V_{CC} = 5\text{ V}, T_A = 25^\circ\text{C}$ .

NOTE 2: This specification applies to any  $5^\circ\text{C}$  band within the operating temperature range.

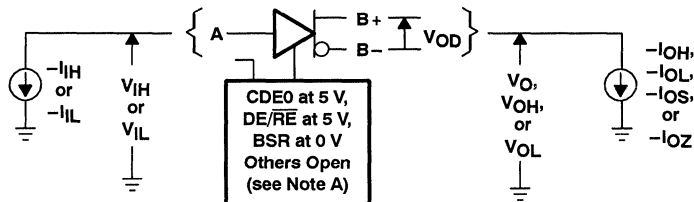
transceiver switching characteristics over recommended operating conditions

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
$t_{enRX(L)}$ Enable time, transmit-to-receive to low-level output	See Figure 6		150	ns
$t_{enRX(H)}$ Enable time, transmit-to-receive to high-level output			150	ns
$t_{enTX(L)}$ Enable time, receive-to-transmit to low-level output			80	ns
$t_{enTX(H)}$ Enable time, receive-to-transmit to high-level output			80	ns
$t_{su}$ Setup time, CDE0, CDE1, CDE2, BSR, or $\overline{CRE}$ to active input(s) or output(s)			150	ns

thermal characteristics

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$R_{\theta JA}$ Junction-to-free-air thermal resistance	Board mounted, No air flow		50		$^\circ\text{C}/\text{W}$
$R_{\theta JC}$ Junction-to-case thermal resistance			12		$^\circ\text{C}/\text{W}$

## PARAMETER MEASUREMENT INFORMATION



NOTE A: For the  $I_{OZ}$  test, the BSR input is at 5 V and all others are at 0 V.

Figure 1. Driver Test and Input Conditions

PARAMETER MEASUREMENT INFORMATION

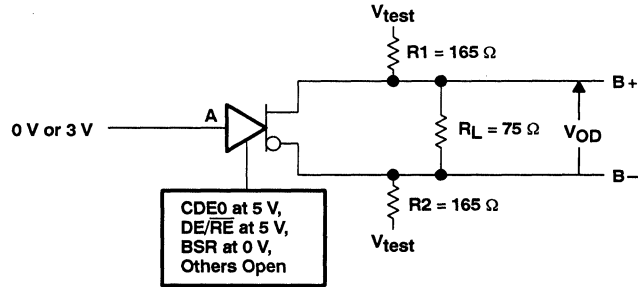
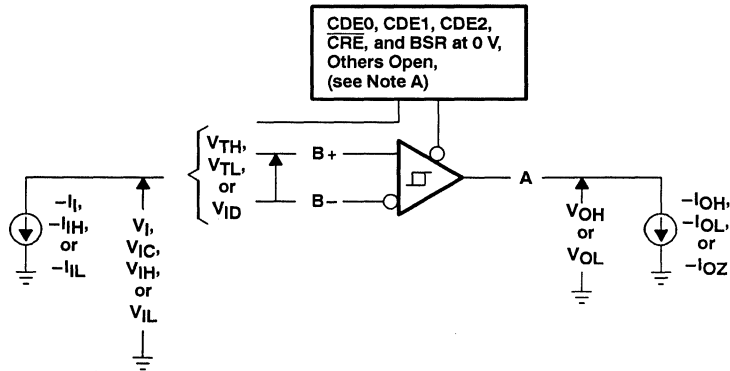


Figure 2. Driver  $V_{OD}$  Test Circuit



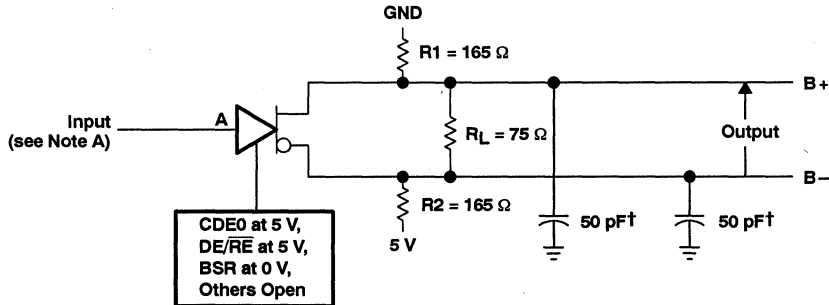
NOTE A: For the  $I_{OZ}$  measurement, BSR is at 5 V and CDE0, CDE1, and CDE2 are at 0 V.

Figure 3. Receiver Test Circuit and Input Conditions

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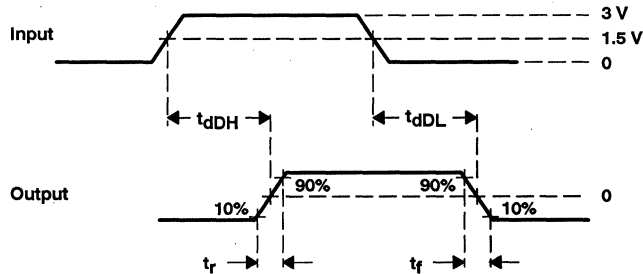
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**PARAMETER MEASUREMENT INFORMATION**



**TEST CIRCUIT**

† Includes probe and jig capacitance.



**VOLTAGE WAVEFORMS**

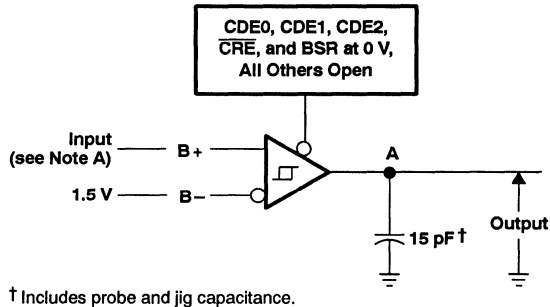
NOTE A: The input is provided by a pulse generator with an output of 0 to 3 V, PRR of 1 MHz, 50% duty cycle,  $t_r$  and  $t_f < 6$  ns, and  $Z_O = 50 \Omega$ .

**Figure 4. Driver Test Circuit and Voltage Waveforms**

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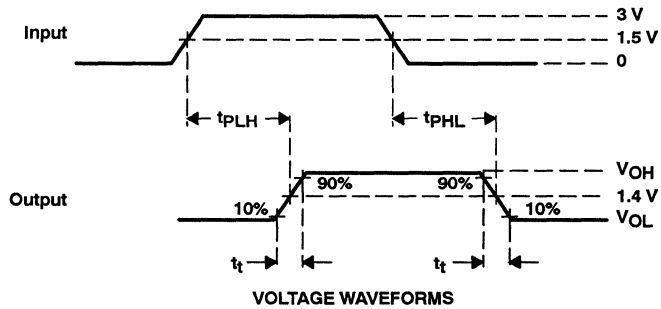
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## PARAMETER MEASUREMENT INFORMATION



† Includes probe and jig capacitance.

### TEST CIRCUIT



### VOLTAGE WAVEFORMS

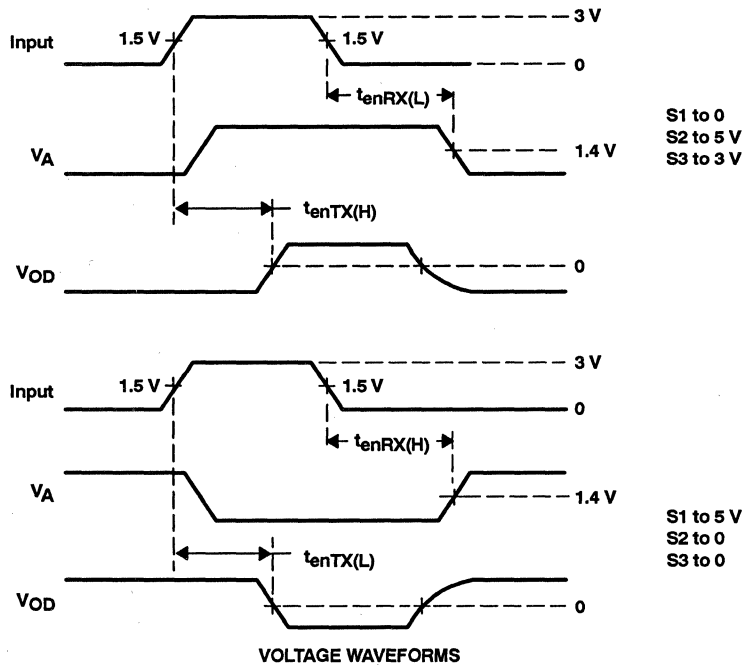
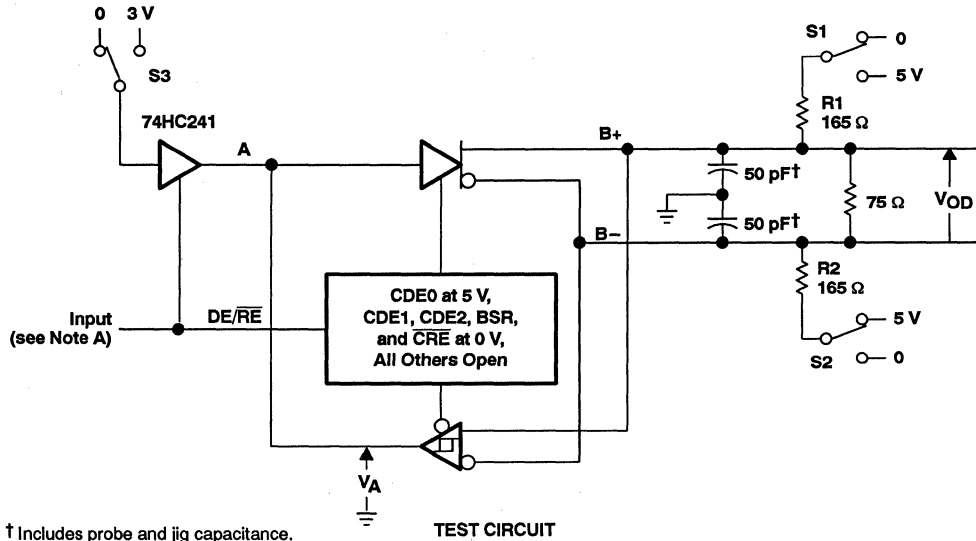
NOTE A: The input is provided by a pulse generator with an output of 0 to 3 V, PRR of 1 MHz, 50% duty cycle,  $t_r$  and  $t_f < 6$  ns, and  $Z_O = 50 \Omega$ .

**Figure 5. Receiver Test Circuit and Voltage Waveforms**

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## PARAMETER MEASUREMENT INFORMATION



NOTE A: The input is provided by a pulse generator with an output of 0 to 3 V, PRR of 1 MHz, 50% duty cycle,  $t_r$  and  $t_f < 6$  ns, and  $Z_0 = 50 \Omega$ .

Figure 6. Enable and Disable Test Circuit and Voltage Waveforms

TYPICAL CHARACTERISTICS

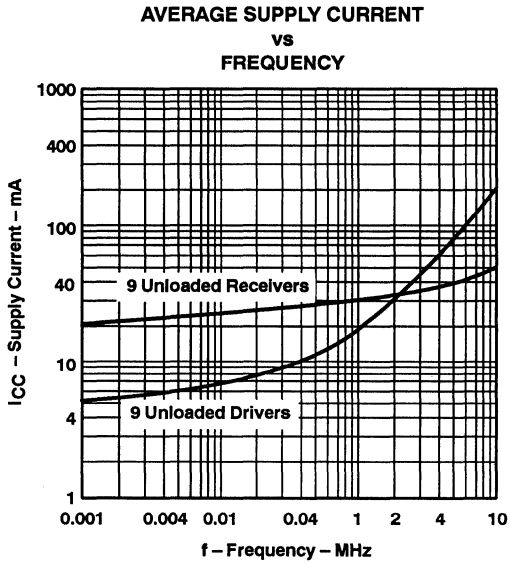


Figure 7

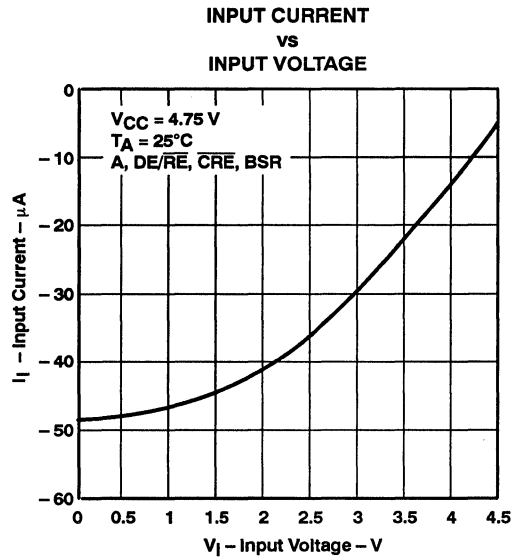


Figure 8

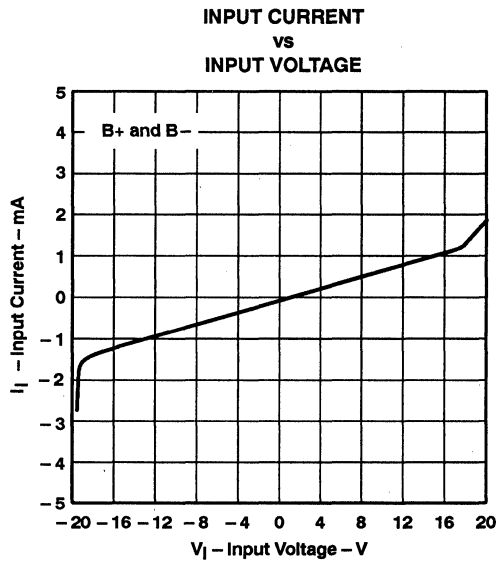


Figure 9



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**TYPICAL CHARACTERISTICS**

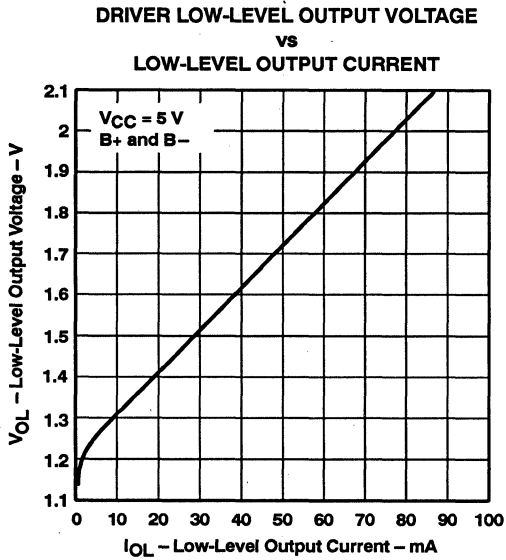


Figure 10

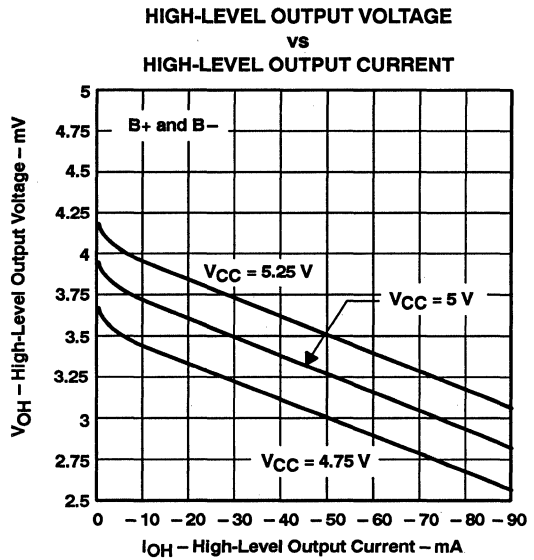


Figure 11

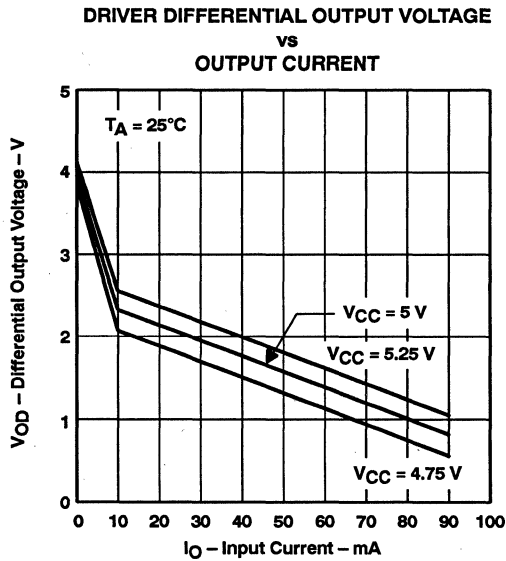


Figure 12

TYPICAL CHARACTERISTICS

DRIVER LOW-LEVEL OUTPUT CURRENT  
vs  
SUPPLY VOLTAGE

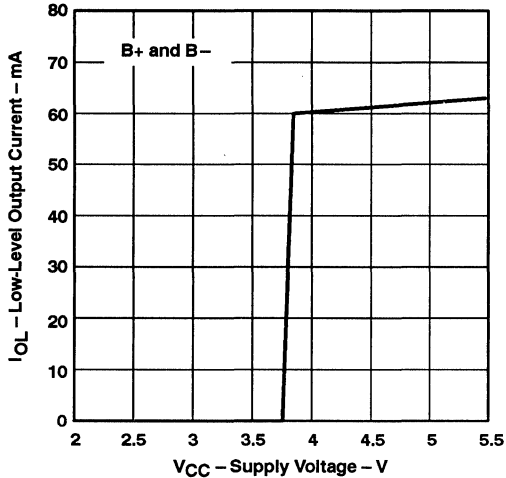


Figure 13

DRIVER HIGH-LEVEL OUTPUT CURRENT  
vs  
SUPPLY VOLTAGE

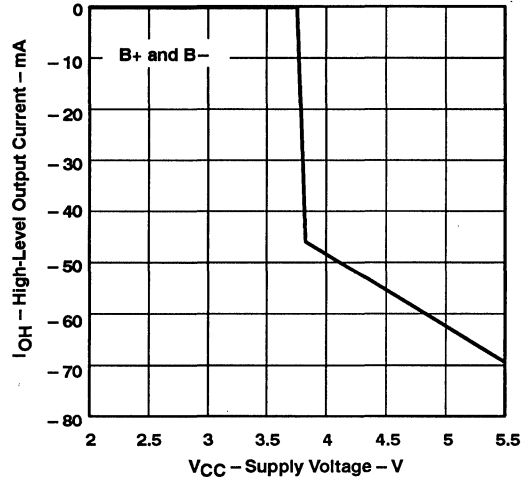


Figure 14

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

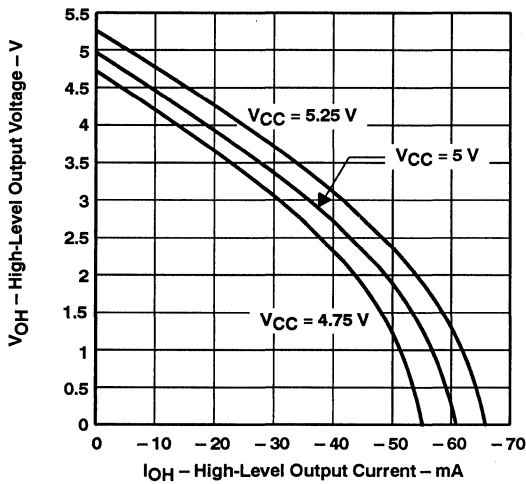


Figure 15

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

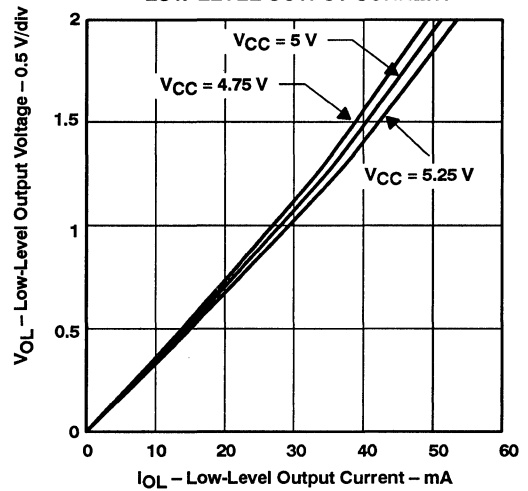
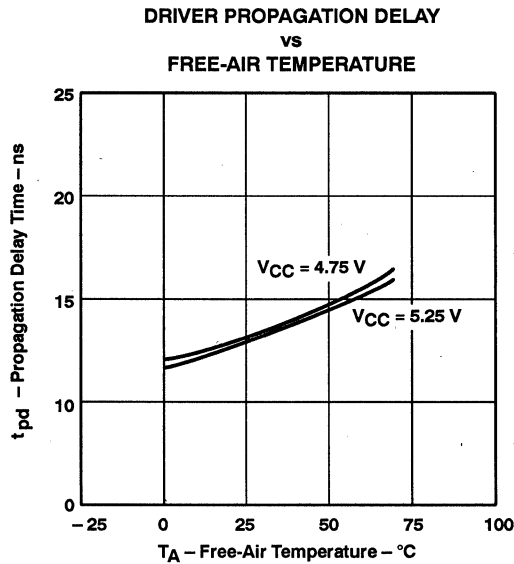
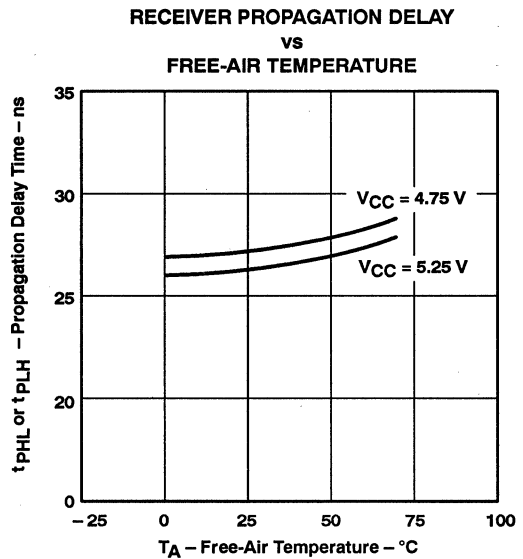
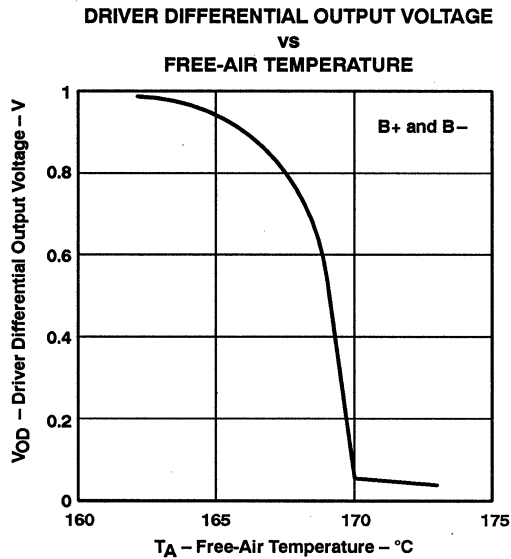


Figure 16

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**TYPICAL CHARACTERISTICS**



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**APPLICATION INFORMATION**

**Table 1. Typical Signal and Pin Assignments**

SIGNAL	PIN	SCSI DATA	SCSI CONTROL	IPI DATA	IPI CONTROL
CDE0	54	DIFFSENSE	DIFFSENSE	VCC	VCC
CDE1	55	GND	GND	XMTA, XMTB	GND
CDE2	56	GND	GND	XMTA, XMTB	SLAVE/MASTER
BSR	2	GND	GND	GND, BSR	GND
CRE	3	GND	GND	GND	VCC
1A	4	DB0, DB8	ATN	AD7, BD7	NOT USED
1DE/ $\overline{RE}$	5	DBE0, DBE8	INIT EN	GND	GND
2A	6	DB1, DB9	BSY	AD6, BD6	NOT USED
2DE/ $\overline{RE}$	7	DBE1, DBE9	BSY EN	GND	GND
3A	8	DB2, DB10	ACK	AD5, BD5	SYNC IN
3DE/ $\overline{RE}$	9	DBE2, DBE10	INIT EN	GND	GND
4A	10	DB3, DB11	RST	AD4, BD4	SLAVE IN
4DE/ $\overline{RE}$	11	DBE3, DBE11	GND	GND	GND
5A	19	DB4, DB12	MSG	AD3, BD3	NOT USED
5DE/ $\overline{RE}$	20	DBE4, DBE12	TARG EN	GND	GND
6A	21	DB5, DB13	SEL	AD2, BD2	SYNC OUT
6DE/ $\overline{RE}$	22	DBE5, DBE13	SEL EN	GND	GND
7A	23	DB6, DB14	C/D	AD1, BD1	MASTER OUT
7DE/ $\overline{RE}$	24	DBE6, DBE14	TARG EN	GND	GND
8A	25	DB7, DB15	REQ	AD0, BD0	SELECT OUT
8DE/ $\overline{RE}$	26	DBE7, DBE15	TARG EN	GND	GND
9A	27	DBP0, DBP1	I/O	AP, BP	ATTENTION IN
9DE/ $\overline{RE}$	28	DBPE0, DBPE1	TARG EN	XMTA, XMTB	VCC

**ABBREVIATIONS:**

- DBn, data bit n, where n = {0, 1, . . . , 15}
- DBEn, data bit n enable, where n = {0, 1, . . . , 15}
- DBP0, parity bit for data bits 0 through 7 or IPI bus A
- DBPE0, parity bit enable for P0
- P1, parity bit for data bits 8 through 15 or IPI bus B
- PE1, parity bit enable for P1
- ADn or BDn, IPI Bus A – Bit n (ADn) or Bus B – Bit n (BDn), where n = {0, 1, . . . , 7}
- AP or BP, IPI parity bit for bus A or bus B
- XMTA or XMTB, transmit enable for IPI bus A or B
- BSR, bit significant response
- INIT EN, common enable for SCSI initiator mode
- TARG EN, common enable for SCSI target mode

**NOTE:** Signal inputs are shown as active high. If only active-low inputs are available, logic inversion is accomplished by reversing the B+ and B– connector pin assignments.



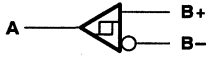
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## APPLICATION INFORMATION

### Function Tables

RECEIVER



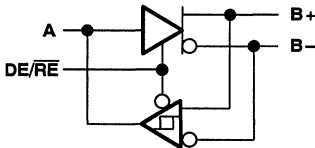
INPUTS		OUTPUT
B+†	B-†	A
L	H	L
H	L	H

DRIVER



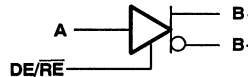
INPUT	OUTPUTS	
A	B+	B-
L	L	H
H	H	L

TRANSCIEVER



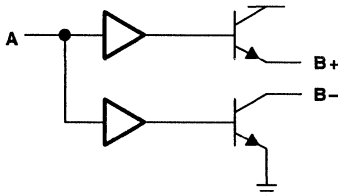
DE/RE	INPUTS			OUTPUTS		
	A	B+†	B-†	A	B+	B-
L	-	L	H	L	-	-
L	-	H	L	H	-	-
H	L	-	-	-	L	H
H	H	-	-	-	H	L

DRIVER WITH ENABLE



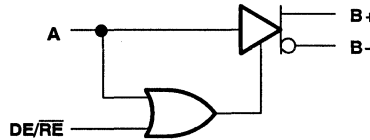
DE/RE	INPUTS	OUTPUTS	
	A	B+	B-
L	L	Z	Z
L	H	Z	Z
H	L	L	H
H	H	H	L

WIRED-OR DRIVER



INPUT	OUTPUTS	
	B+	B-
L	Z	Z
H	H	L

TWO-ENABLE INPUT DRIVER



DE/RE	INPUTS	OUTPUTS	
	A	B+	B-
L	L	Z	Z
L	H	H	L
H	L	L	H
H	H	H	L

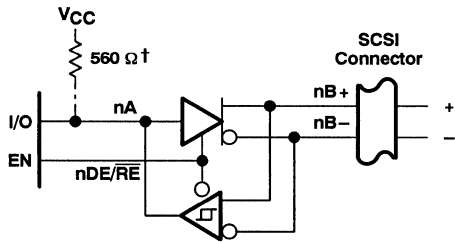
H = high level, L = low level, X = irrelevant, Z = high impedance (off)

† An H in this column represents a voltage 200 mV higher than the other bus input. An L represents a voltage 200 mV lower than the other bus input. Any voltage less than 200 mV results in an indeterminate receiver output.

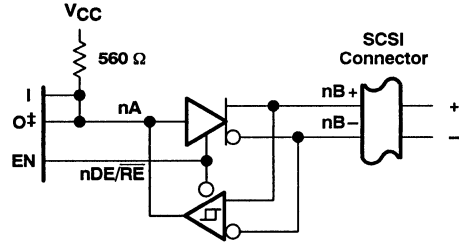
# SN75LBC976 9-CHANNEL DIFFERENTIAL TRANSCIEVER

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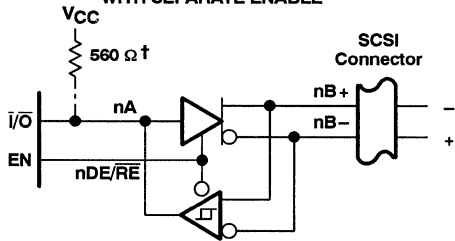
## APPLICATION INFORMATION



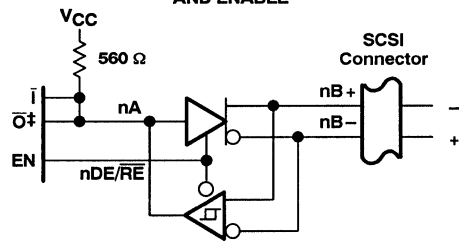
(a) ACTIVE-HIGH BIDIRECTIONAL I/O WITH SEPARATE ENABLE



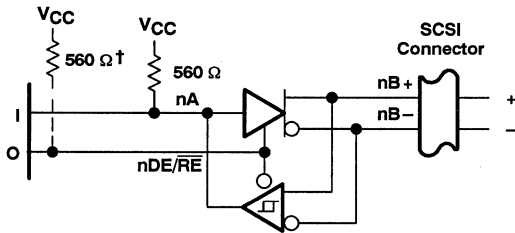
(d) SEPARATE ACTIVE-HIGH INPUT, OUTPUT, AND ENABLE



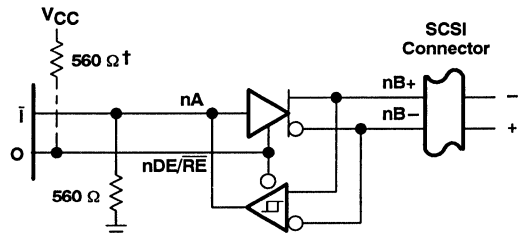
(b) ACTIVE-LOW BIDIRECTIONAL I/O WITH SEPARATE ENABLE



(e) SEPARATE ACTIVE-LOW INPUT AND OUTPUT AND ACTIVE-HIGH ENABLE



(c) WIRED-OR DRIVER AND ACTIVE-HIGH INPUT



(f) WIRED-OR DRIVER AND ACTIVE-LOW INPUT

† If O is open drain

‡ Must be open-drain or 3-state output

NOTE: The BSR,  $\overline{\text{CRE}}$ , A, and DE/RE inputs have internal pullups. CDE0, CDE1, and CDE2 have internal pulldowns.

**Figure 20. Typical SCSI Transceiver Connections**

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**APPLICATION INFORMATION**

**channel logic configurations with control input logic**

The following logic diagrams show the positive logic representation for all combinations of control inputs. The control inputs are from MSB to LSB; BSR, CDE0, CDE1, CDE2, and  $\overline{CRE}$ , and are shown below the diagrams. Channel 1 is at the top and channel 9 at the bottom of the logic diagrams.

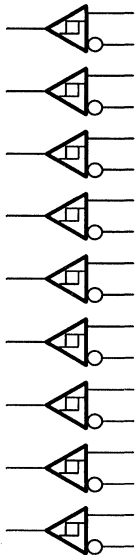


Figure 21. 00000

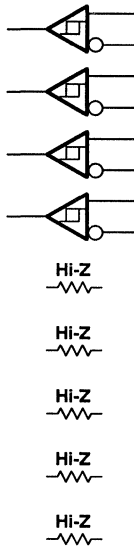


Figure 22. 00001

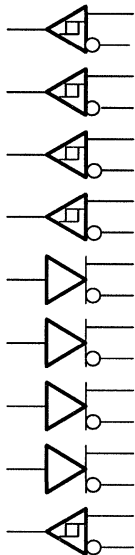


Figure 23. 00010

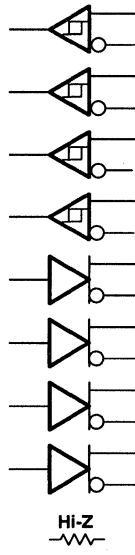


Figure 24. 00011

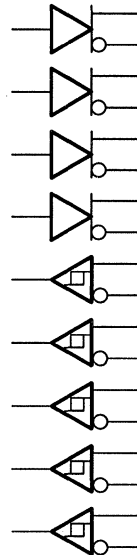


Figure 25. 00100

APPLICATION INFORMATION

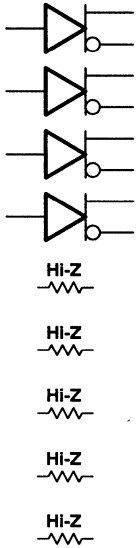


Figure 26. 00101

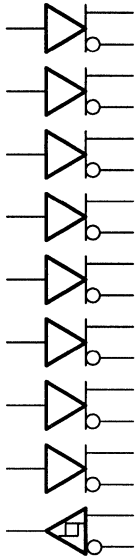


Figure 27. 00110

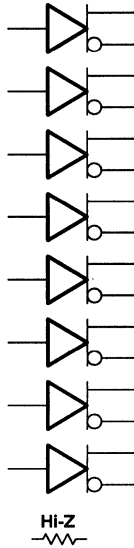


Figure 28. 00111

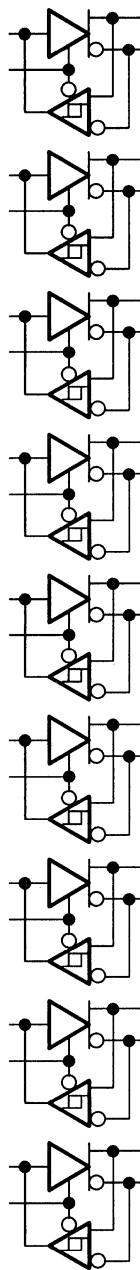


Figure 29. 01000

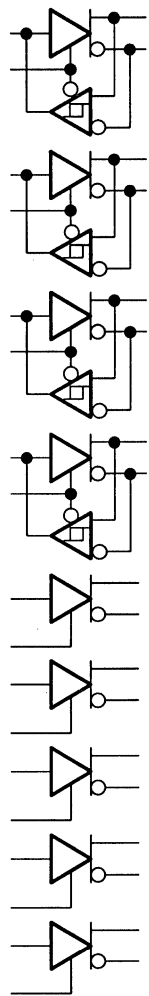


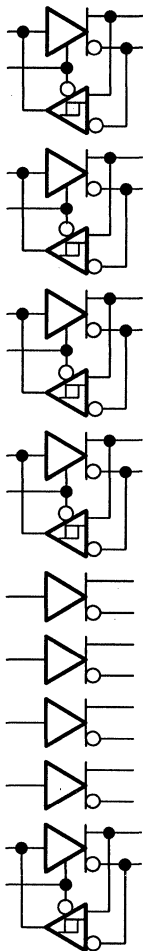
Figure 30. 01001



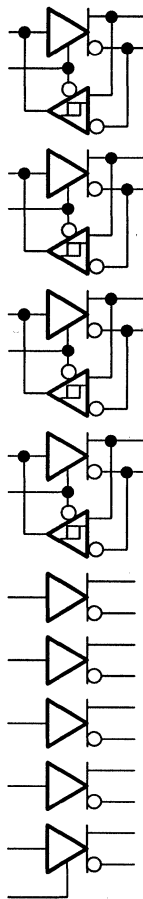
**SN75LBC976  
9-CHANNEL DIFFERENTIAL TRANSCIEVER**

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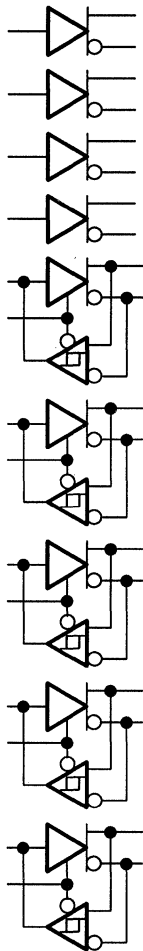
**APPLICATION INFORMATION**



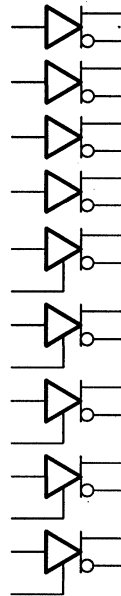
**Figure 31. 01010**



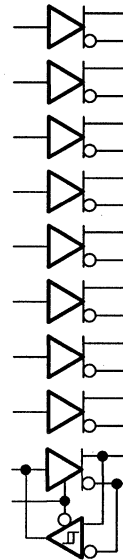
**Figure 32. 01011**



**Figure 33. 01100**



**Figure 34. 01101**



**Figure 35. 01110**

APPLICATION INFORMATION

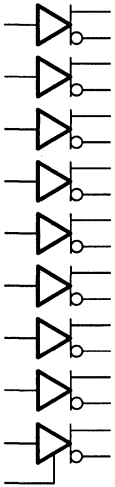


Figure 36. 01111

Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

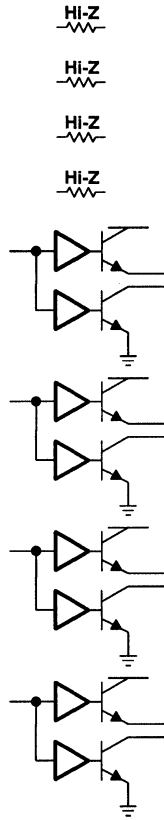
Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

Figure 37.  
10000  
and 10001



Hi-Z  
~

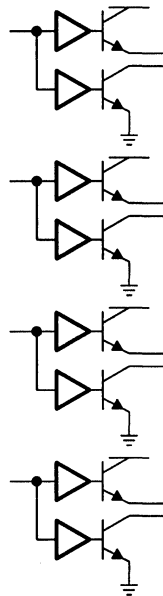
Figure 38. 10010  
and 10011

Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

Hi-Z  
~



Hi-Z  
~

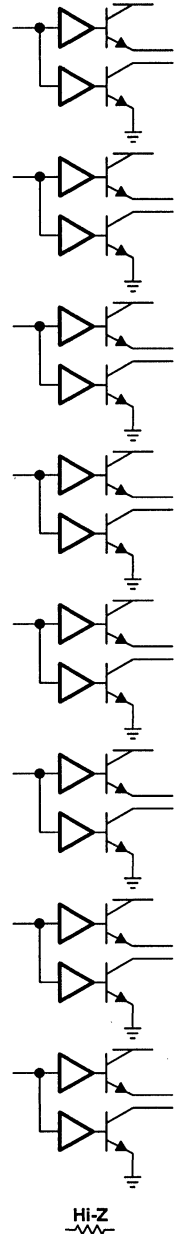
Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

Hi-Z  
~

Figure 39. 10100  
and 10101



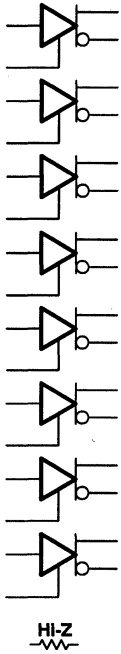
Hi-Z  
~

Figure 40. 10110  
and 10111

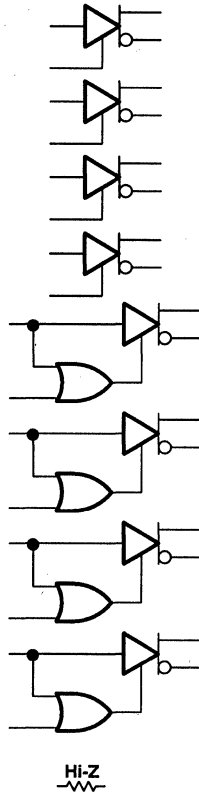
**SN75LBC976**  
**9-CHANNEL DIFFERENTIAL TRANSCIEVER**

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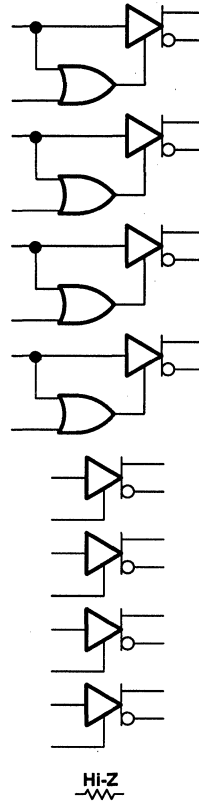
**APPLICATION INFORMATION**



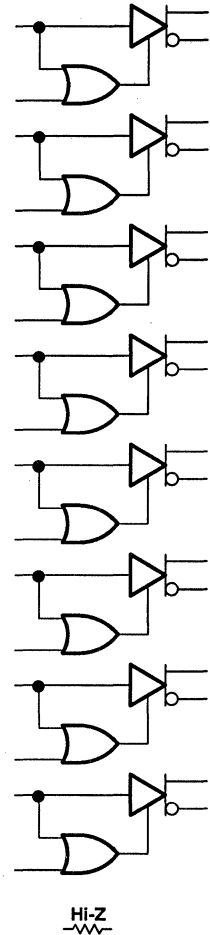
**Figure 41. 11000 and 11001**



**Figure 42. 11010 and 11011**



**Figure 43. 11100 and 11101**



**Figure 44. 11110 and 11111**

# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCEIVER

SLLS134B – D4088, APRIL 1992 – REVISED FEBRUARY 1993

- **9 Differential Channels for the Data and Control Paths of the Differential Small Computer Systems Interface (SCSI)**
- **Each Transceiver Meets EIA-RS-485 and ISO 8482:1987(E) Standards**
- **Packaged in Shrink Small-Outline Package With 25-mil Terminal Pitch**
- **Designed to Operate at 10 Million Transfers Per Second**
- **Low Disabled Supply Current  
1.4 mA Typ**
- **Thermal Shutdown Protection**
- **Power Up/Down Glitch Protection**
- **Positive and Negative Output Current Limiting**
- **Open-Circuit Fail-Safe Receiver Design**

## description

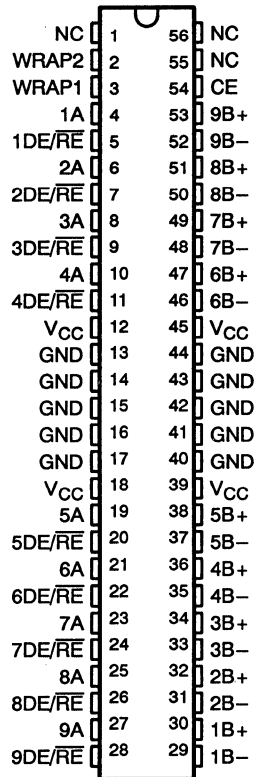
The SN75LBC978 is a nine-channel differential transceiver based on the 75LBC176 LinASIC™ cell. Use of the Texas Instruments LinBiCMOS™ process technology allows the power reduction necessary to integrate nine differential balanced transceivers†. On-chip enabling logic makes this device applicable for the data path (eight data bits plus parity) and the control path (nine bits) for the Small Computer Systems Interface (SCSI) standard. The WRAP function allows in-circuit testing and wired-OR channels for the BSY, RST, and SEL signals of the SCSI bus.

The SN75LBC978 is packaged in a shrink small-outline package (DL) with improved thermal characteristics using heat sink pins. This package is ideal for low-profile, space-restricted applications such as hard disk drives.

The switching speed of the SN75LBC978 is sufficient to transfer data over the data bus at 10 million transfers per second. Each of the nine identical channels conforms to the requirements of the EIA-RS-485 and ISO 8482:1987(E) standards referenced by ANSI X3.131-1986 (SCSI-1) and the proposed SCSI-2 standards. This device conforms to ESD Class 1 (A) and Class 2 (B) per MIL-STD-883, Method 3015.

The SN75LBC978 is characterized for operation from 0°C to 70°C.

DL PACKAGE  
(TOP VIEW)



Pins 13 through 17 and 40 through 44 are connected together to the package lead frame and signal ground.

† Patent Pending

LinASIC and LinBiCMOS are trademarks of Texas Instruments Incorporated.

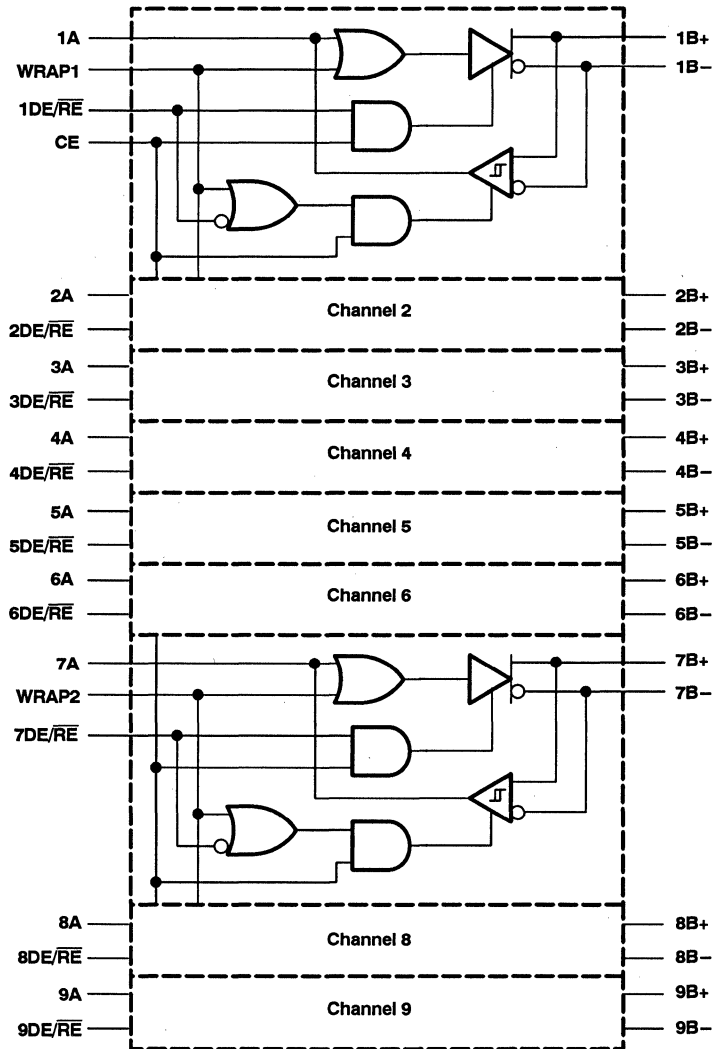
PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCIEVER

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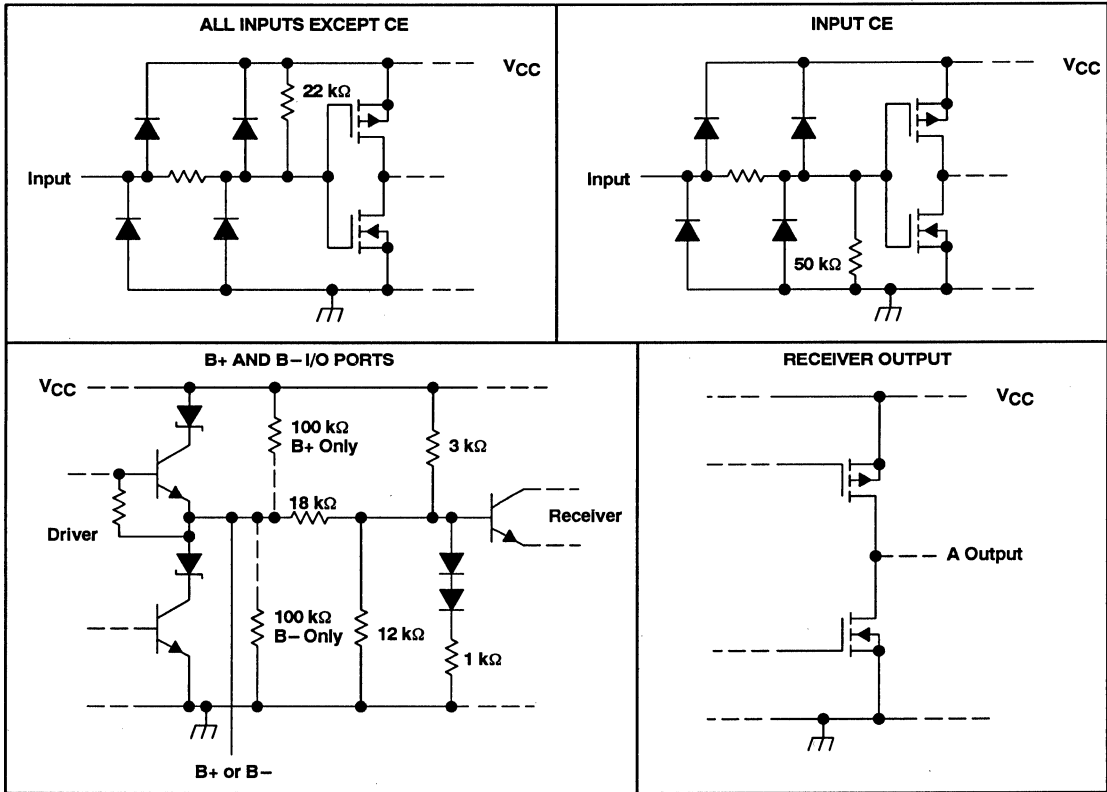
## logic diagram (positive logic)



# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCIEVER

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## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)<sup>†</sup>

Supply voltage range, $V_{CC}$ (see Note 1)	–0.3 V to 7 V
Bus voltage range	–10 V to 15 V
Data I/O and control (A-side) voltage range	–0.3 V to 7 V
Continuous power dissipation	internally limited
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are dc and with respect to the GND terminal.

# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCEIVER

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## recommended operating conditions

		MIN	NOM	MAX	UNIT	
Supply voltage, $V_{CC}$		4.75	5	5.25	V	
Voltage at any bus terminal (separately or common-mode), $V_O$ , $V_I$ , or $V_{IC}$	B+ or B-				12	V
					-7	
High-level input voltage, $V_{IH}$	All except B+ and B-	2			V	
Low-level input voltage, $V_{IL}$	All except B+ and B-				0.8	V
High-level output current, $I_{OH}$	B+ or B-				-60	mA
	A				-8	mA
Low-level output current, $I_{OL}$	B+ or B-				60	mA
	A				8	mA
Operating free-air temperature, $T_A$		0	70		°C	

## driver electrical characteristics over recommended ranges of operating conditions (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$ V_{OD} $ Differential output voltage	$V_{test} = -7\text{ V to }12\text{ V}$ , See Figure 2	1	2		V
$I_{OS}$ Output short-circuit current	See Figure 1			±250	mA
$I_{OZ}$ High-impedance state output current	See receiver input current				

## receiver electrical characteristics over recommended ranges of operating conditions (unless otherwise noted) (see Figure 3)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}$ High-level output voltage	$V_{ID} = 200\text{ mV}$ , $I_{OH} = -8\text{ mA}$	2.5			V
$V_{OL}$ Low-level output voltage	$V_{ID} = -200\text{ mV}$ , $I_{OL} = 8\text{ mA}$			0.8	V
$V_{T+}$ Differential-input high-level threshold voltage	$I_{OH} = -8\text{ mA}$			0.2	V
$V_{T-}$ Differential-input low-level threshold voltage	$I_{OL} = 8\text{ mA}$	-0.2			V
$V_{hys}$ Receiver input hysteresis ( $V_{T+} - V_{T-}$ )		45			mV
$I_I$ Receiver input current	B+ and B-	$V_I = 12\text{ V}$ , $V_{CC} = 5\text{ V}$ , Other input at 0 V	0.7	1	mA
		$V_I = 12\text{ V}$ , $V_{CC} = 0\text{ V}$ , Other input at 0 V	0.8	1	mA
		$V_I = -7\text{ V}$ , $V_{CC} = 5\text{ V}$ , Other input at 0 V	-0.5	-0.8	mA
		$V_I = -7\text{ V}$ , $V_{CC} = 0\text{ V}$ , Other input at 0 V	-0.4	-0.8	mA
$I_{OZ}$ High-impedance state output current	$V_O = \text{GND}$			-100	μA
	$V_O = V_{CC}$			50	

† All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCEIVER

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## device electrical characteristics over recommended ranges of operating conditions

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
I <sub>IH</sub>	High-level input current	A, WRAP, DE/ $\overline{RE}$	See Figure 3	V <sub>IH</sub> = 5 V		-100	μA
		CE				50	μA
I <sub>IL</sub>	Low-level input current	A, WRAP, DE/ $\overline{RE}$		V <sub>IL</sub> = 0 V		-200	μA
		CE			-50	50	μA
I <sub>CC</sub>	Supply current	All drivers and receivers disabled	CE at 0 V		1.4	3	mA
		All receivers enabled	No load, V <sub>ID</sub> = 5 V, CE at 5 V, WRAP and DE/ $\overline{RE}$ at 0 V		29	45	mA
		All drivers enabled	No load, CE and DE/ $\overline{RE}$ at 5 V, WRAP at 0 V		7	10	mA
C <sub>O</sub>	Bus port capacitance	B+ or B-			19		pF
C <sub>pd</sub>	Power dissipation capacitance‡	One driver			460		pF
		One receiver			40		pF

## driver switching characteristics over recommended operating conditions (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT	
t <sub>dD</sub>	Differential delay time, high-to-low-level output (t <sub>dDH</sub> ) or low-to-high-level output (t <sub>dDL</sub> )			11.8		26.4	ns	
		V <sub>CC</sub> = 5 V, T <sub>A</sub> = 25°C		14	18	22		
		V <sub>CC</sub> = 5 V, T <sub>A</sub> = 70°C		18	22	26		
t <sub>sk(lim)</sub>	Skew limit, the maximum difference in propagation delay times between any two drivers on any two devices					15	ns	
		V <sub>CC</sub> = 5 V, See Note 2				8		
t <sub>sk(p)</sub>	Pulse skew ((t <sub>dDL</sub> - t <sub>dDH</sub> ))					0	6	ns
t <sub>t</sub>	Transition time (t <sub>r</sub> or t <sub>f</sub> )					10		ns

† All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

‡ C<sub>pd</sub> determines the no-load dynamic current consumption, I<sub>S</sub> = C<sub>pd</sub> V<sub>CC</sub> f + I<sub>CC</sub>.

NOTE 2: This specification applies to any 5°C band within the operating temperature range.

## transceiver switching characteristics over recommended operating conditions

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
t <sub>en(TL)</sub>	Transmit-to-receive enable time to low-level output	See Figure 6		80	ns
t <sub>en(TH)</sub>	Transmit-to-receive enable time to high-level output			80	ns
t <sub>en(FL)</sub>	Receive-to-transmit enable time to low-level output			150	ns
t <sub>en(RH)</sub>	Receive-to-transmit enable time to high-level output			150	ns
t <sub>su</sub>	Setup time, WRAP1 or WRAP2 before active input(s) or output(s)			150	ns

## thermal characteristics

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
R <sub>θJA</sub>	Junction-to-free-air thermal resistance	Board mounted, No air flow		50		°C/W
R <sub>θJC</sub>	Junction-to-case thermal resistance			12		°C/W





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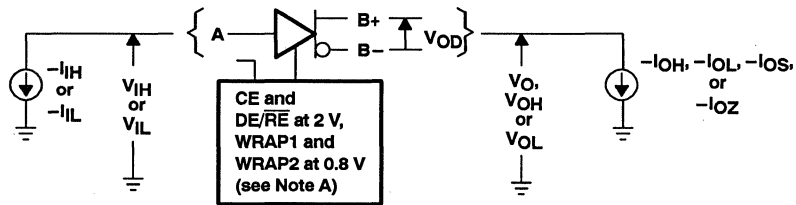
receiver switching characteristics over recommended operating conditions (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$t_{pd}$	Propagation delay time, high-to-low-level output ( $t_{pHL}$ ) or low-to-high-level output ( $t_{pLH}$ )		19.5	30.7	ns	
		$V_{CC} = 5\text{ V}, T_A = 25^\circ\text{C}$	20.2	24.7		29.2
		$V_{CC} = 5\text{ V}, T_A = 70^\circ\text{C}$	21.1	25.6		30.1
$t_{sk(lim)}$	Skew limit, the maximum difference in propagation delay times between any two drivers on any two devices			12	ns	
		$V_{CC} = 5\text{ V},$ See Note 2		9		
$t_{sk(p)}$	Pulse skew ( $t_{pHL} - t_{pLH}$ )		2	6	ns	
$t_t$	Transition time ( $t_r$ or $t_f$ )		3		ns	

† All typical values are at  $V_{CC} = 5\text{ V}, T_A = 25^\circ\text{C}$ .

NOTE 2: This specification applies to any  $5^\circ\text{C}$  band within the operating temperature range.

## PARAMETER MEASUREMENT INFORMATION



NOTE A: For the  $I_{OZ}$  test, the CE input is at 0.8 V.

Figure 1. Driver Test and Input Conditions

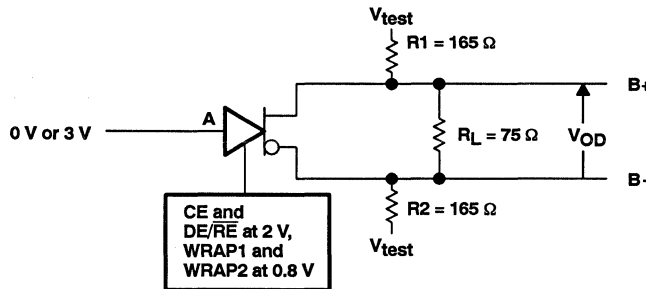
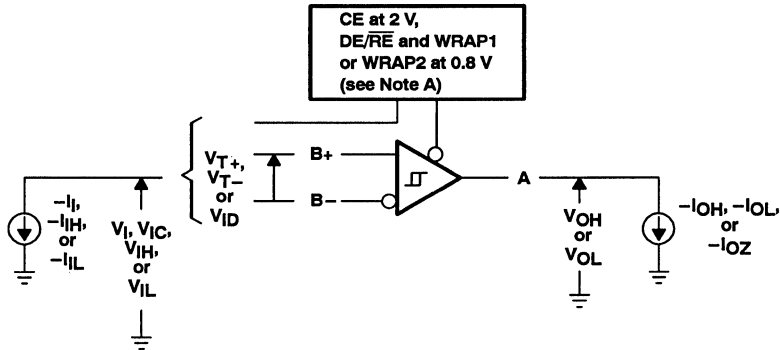


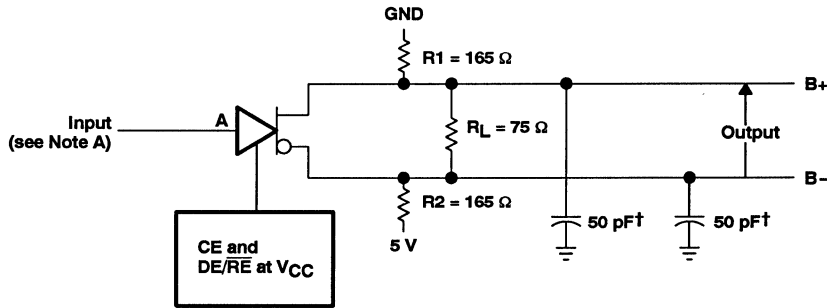
Figure 2. Driver  $V_{OD}$  Test Circuit

PARAMETER MEASUREMENT INFORMATION

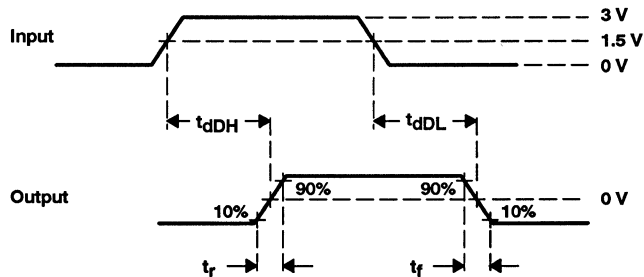


NOTE A: For the  $I_{OZ}$  measurement, CE is at 0.8 V.

Figure 3. Receiver Test and Input Conditions



† Includes probe and jig capacitance.



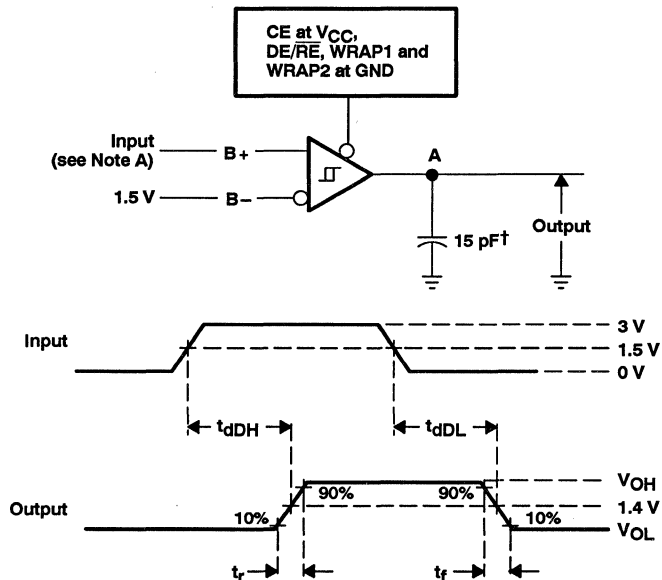
NOTE A: The input is provided by a pulse generator with an output of 0 to 3 V, PRR of 1 MHz, 50% duty cycle,  $t_r$  and  $t_f < 6$  ns, and  $Z_O = 50 \Omega$ .

Figure 4. Driver Propagation Delay Time Test Circuit and Waveforms

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## PARAMETER MEASUREMENT INFORMATION

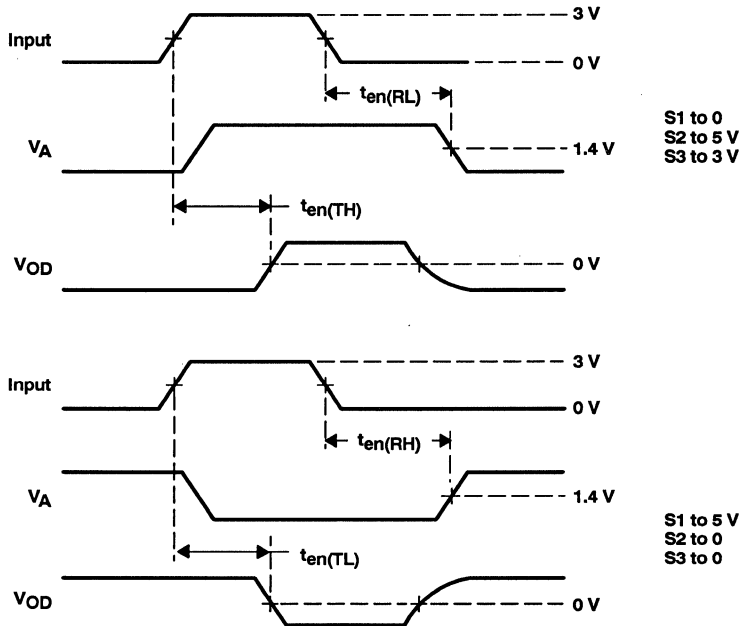
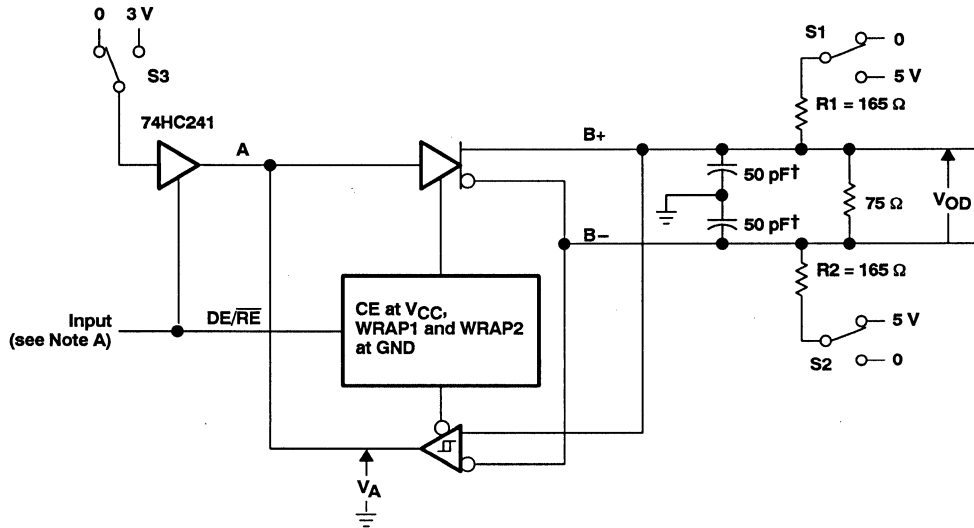


† Includes probe and jig capacitance.

NOTE A: The input is provided by a pulse generator with an output of 0 to 3 V, PRR of 1 MHz, 50% duty cycle, t<sub>r</sub> and t<sub>f</sub> < 6 ns, and Z<sub>O</sub> = 50 Ω.

**Figure 5. Receiver Propagation Delay Time Test Circuit and Waveforms**

PARAMETER MEASUREMENT INFORMATION



† Includes probe and jig capacitance.

NOTE A: The input is provided by a pulse generator with an output of 0 to 3 V, PRR of 1 MHz, 50% duty cycle,  $t_r$  and  $t_f < 6$  ns, and  $Z_O = 50 \Omega$ .

Figure 6. Enable and Disable Time Measurements

# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCIEVER

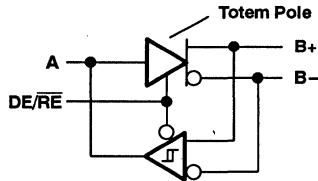
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## APPLICATION INFORMATION

### Truth Tables for Possible Channel Functions

Table 1

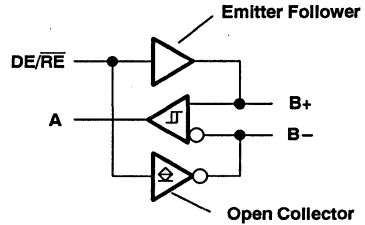
CE is high  
WRAP1 or WRAP2 is low



INPUTS				OUTPUTS		
DE/RE	A	B+†	B-†	A	B+	B-
L	X	L	H	L	Z	Z
L	X	H	L	H	Z	Z
H	L	X	X	Z	L	H
H	H	X	X	Z	H	L

Table 2

CE is high  
WRAP1 or WRAP2 is high



INPUTS			OUTPUTS		
DE/RE	B+	B-	A	B+	B-
L	L	H	L	Z	Z
L	H	L	H	Z	Z
H	X	X	H	H	L
H	X	X	H	H	L

H = high level, L = low level, X = irrelevant Z = high impedance

† An H in this column represents a voltage 200 mV higher than the other bus input. An L represents a voltage 200 mV lower than the other bus input. Any voltage less than 200 mV results in an indeterminate receiver output.

TYPICAL CHARACTERISTICS

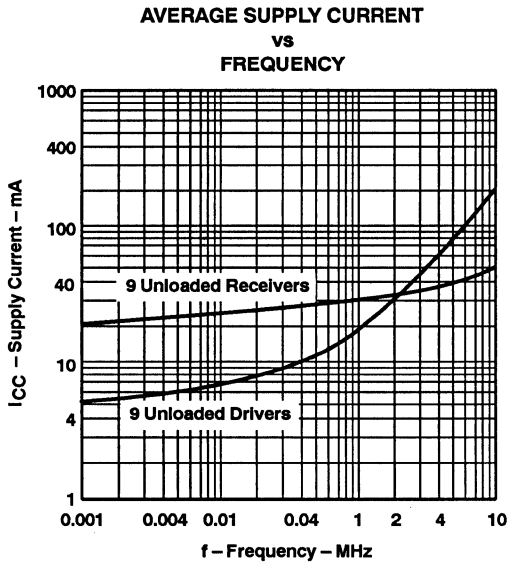


Figure 7

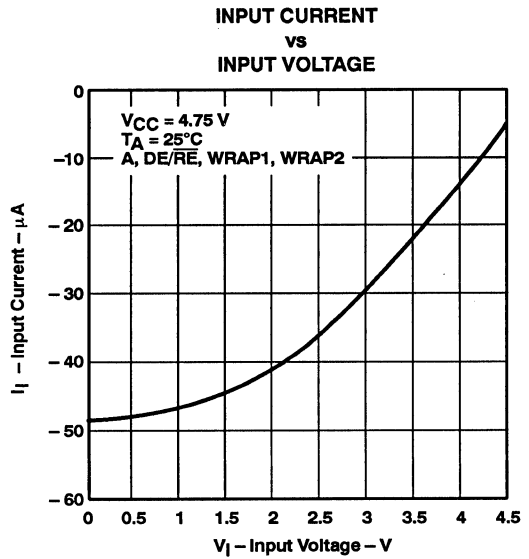


Figure 8

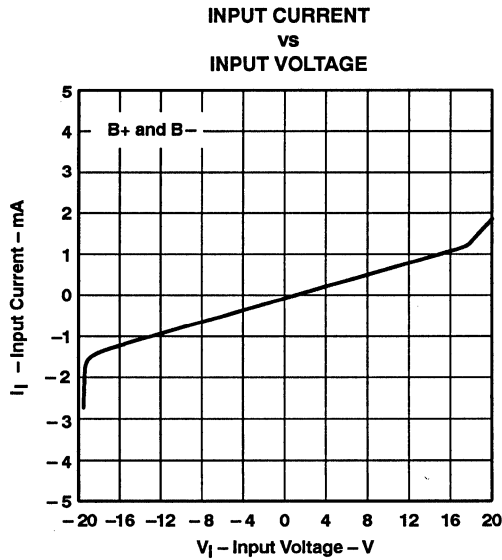
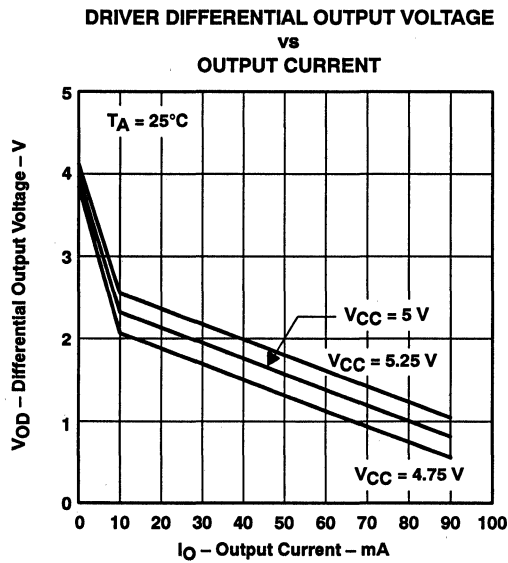
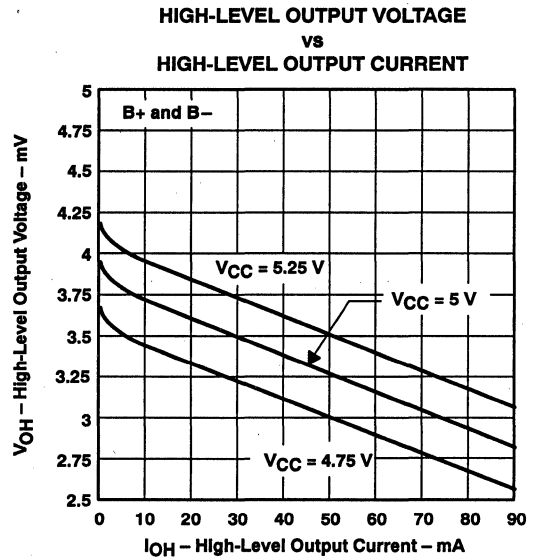
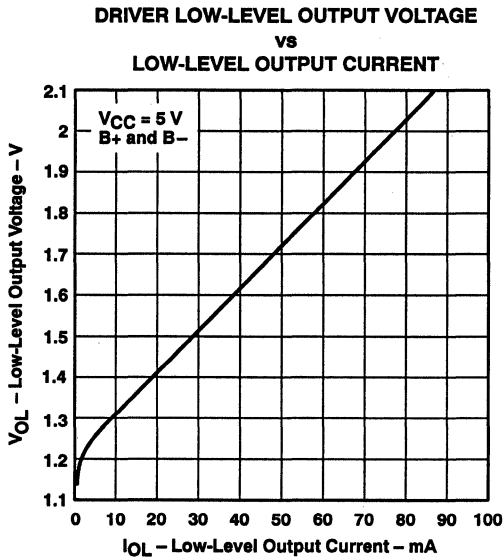


Figure 9

# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCEIVER

SLLS134B - D4088, APRIL 1992 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS



# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCIEVER

SLLS134B – D4088, APRIL 1992 – REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

**DRIVER LOW-LEVEL OUTPUT CURRENT  
vs  
SUPPLY VOLTAGE**

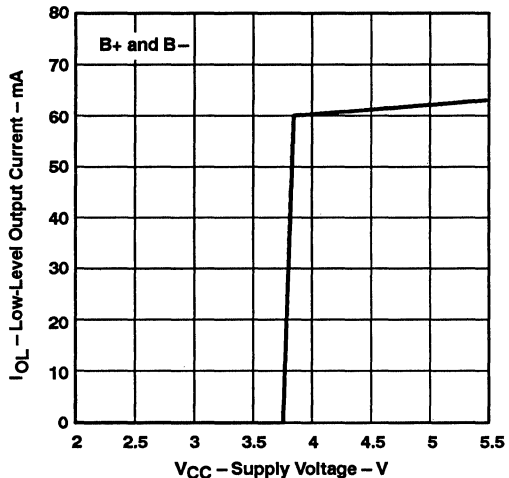


Figure 13

**DRIVER HIGH-LEVEL OUTPUT CURRENT  
vs  
SUPPLY VOLTAGE**

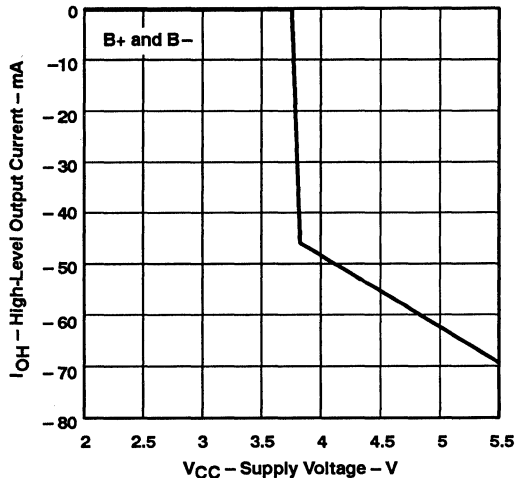


Figure 14

**RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

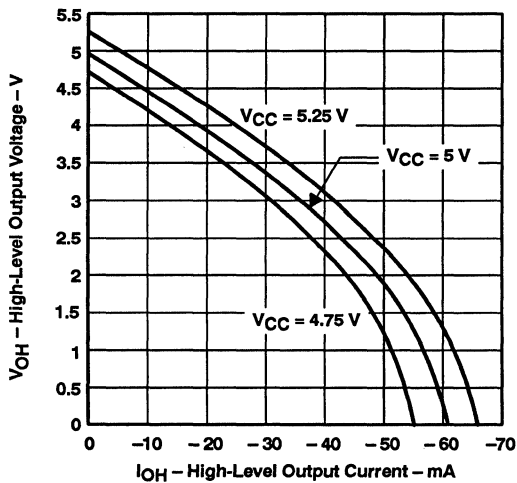


Figure 15

**RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

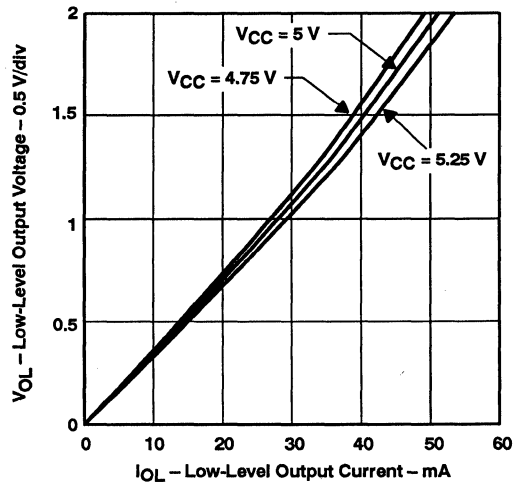


Figure 16





# SN75LBC978 9-CHANNEL DIFFERENTIAL TRANSCIEVER

SLLS134B – D4088, APRIL 1992 – REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

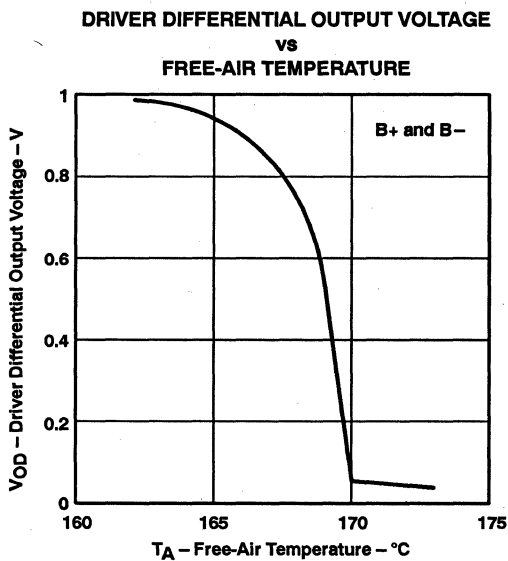


Figure 17

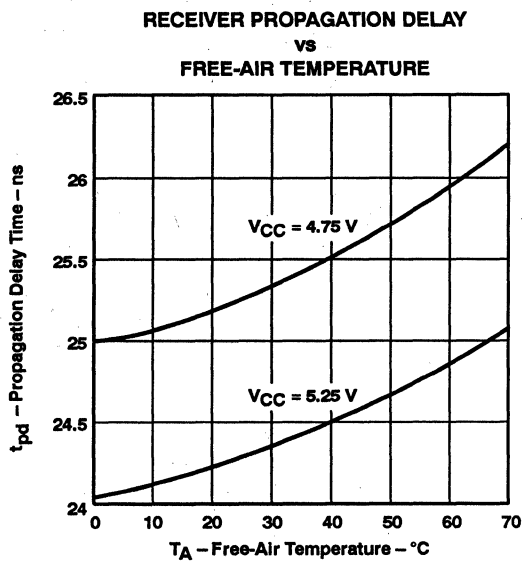


Figure 18

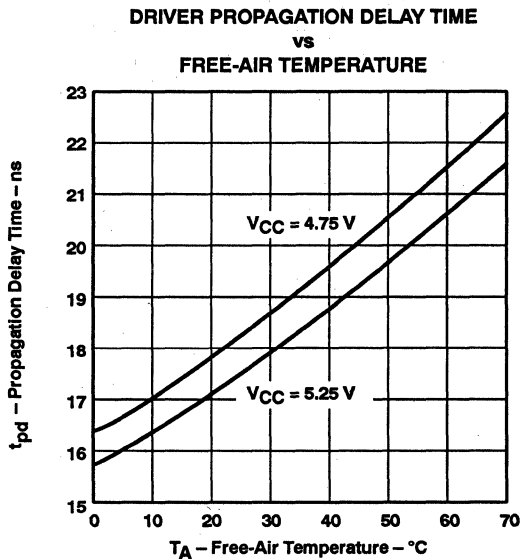


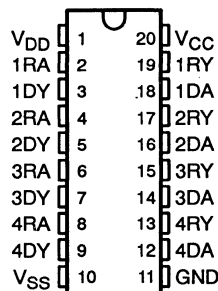
Figure 19

# SN65C1154, SN75C1154 QUAD LOW-POWER DRIVERS/RECEIVERS

SLLS151A – D3230, DECEMBER 1988 – REVISED MARCH 1993

- Meets Standard EIA-232-D (Revision of RS-232-C)
- Very Low Power Consumption  
5 mW Typ
- Wide Driver Supply Voltage . . .  $\pm 4.5$  V  
to  $\pm 15$  V
- Driver Output Slew Rate Limited to 30 V/ $\mu$ s  
Max
- Receiver Input Hysteresis . . . 1000 mV Typ
- Push-Pull Receiver Outputs
- On-Chip Receiver 1- $\mu$ s Noise Filter

DW OR N PACKAGE  
(TOP VIEW)



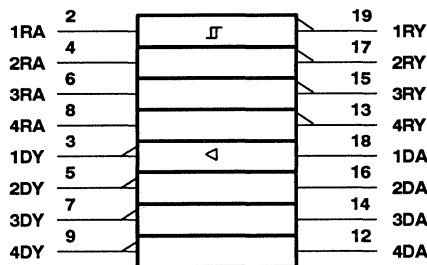
## description

The SN65C1154 and SN75C1154 are low-power BI-MOS devices containing four independent drivers and receivers that are used to interface data terminal equipment (DTE) with data circuit-terminating equipment (DCE). This device has been designed to conform to Standards ANSI/EIA-232-D-1986 (which supersedes RS-232-C). The drivers and receivers of the SN65C1154 and SN75C1154 are similar to those of the SN75C188 quadruple driver and SN75C189A quadruple receiver, respectively. The drivers have a controlled output slew rate that is limited to a maximum of 30 V/ $\mu$ s and the receivers have filters that reject input noise pulses of shorter than 1  $\mu$ s. Both these features eliminate the need for external components.

The SN65C1154 and SN75C1154 have been designed using low-power techniques in a BI-MOS technology. In most applications, the receivers contained in these devices interface to single inputs of peripheral devices such as ACEs, UARTs, or microprocessors. By using sampling, such peripheral devices are usually insensitive to the transition times of the input signals. If this is not the case or for other uses, it is recommended that the SN65C1154 and SN75C1154 receiver outputs be buffered by single Schmitt input gates or single gates of the HCMOS, ALS, or 74F logic families.

The SN65C1154 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75C1154 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

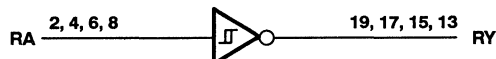
## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)

typical of each receiver



typical of each driver



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

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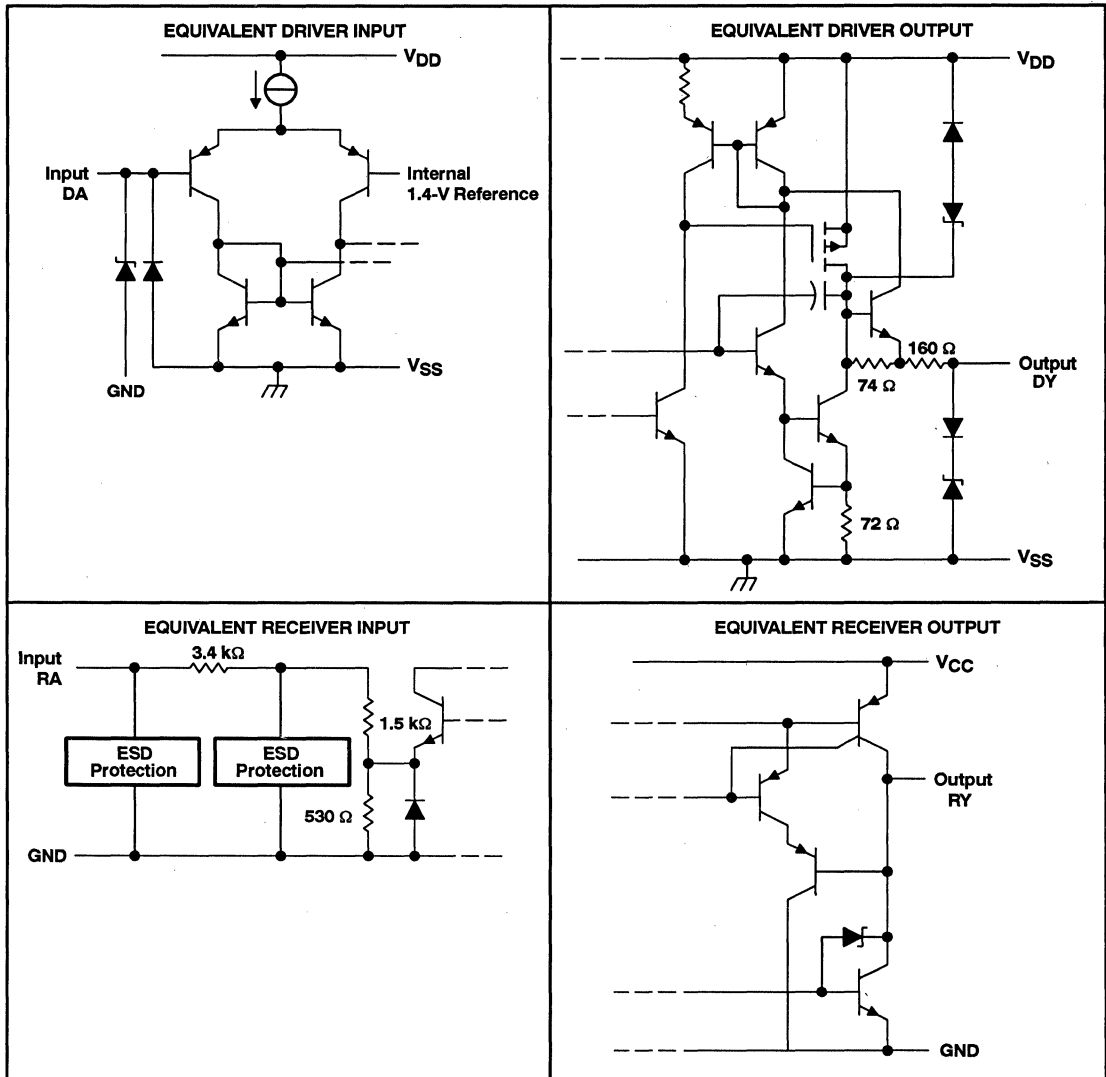
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2-901

# SN65C1154, SN75C1154 QUAD LOW-POWER DRIVERS/RECEIVERS

SLLS151A—D3230, DECEMBER 1988—REVISED MARCH 1993

## schematics of inputs and outputs



Resistor values shown are nominal.

**TEXAS**  
**INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# SN65C1154, SN75C1154 QUAD LOW-POWER DRIVERS/RECEIVERS

SLLS151A – D3230, DECEMBER 1988 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{DD}$ (see Note 1)	15 V
Supply voltage, $V_{SS}$	-15 V
Supply voltage, $V_{CC}$	7 V
Input voltage range: Driver	$V_{SS}$ to $V_{DD}$
Receiver	-30 V to 30 V
Output voltage range: Driver	$(V_{SS} - 6 V)$ to $(V_{DD} + 6 V)$
Receiver	-0.3 V to $(V_{CC} + 0.3 V)$
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65C1154	-40°C to 85°C
SN75C1154	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the network GND terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	920 mW	585 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{DD}$		4.5	12	15	V
Supply voltage, $V_{SS}$		-4.5	-12	-15	V
Supply voltage, $V_{CC}$		4.5	5	6	V
Input voltage, $V_I$	Driver	$V_{SS}+2$		$V_{DD}$	V
	Receiver			$\pm 25$	
High-level input voltage, $V_{IH}$	Driver	2		0.8	V
Low-level input voltage, $V_{IL}$					
High-level output current, $I_{OH}$	Receiver			-1	mA
High-level output current, $I_{OL}$				3.2	mA
Operating free-air temperature, $T_A$	SN65C1154	-40			°C
	SN75C1154	0	70		



# SN65C1154, SN75C1154 QUAD LOW-POWER DRIVERS/RECEIVERS

SLLS151A - D3230, DECEMBER 1988 - REVISED MARCH 1993

## DRIVER SECTION

electrical characteristics over operating free-air temperature range,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
VOH	High-level output voltage	V <sub>IL</sub> = 0.8 V, See Figure 1	R <sub>L</sub> = 3 k $\Omega$ ,	V <sub>DD</sub> = 5 V, V <sub>SS</sub> = -5 V	4	4.5	V
				V <sub>DD</sub> = 12 V, V <sub>SS</sub> = -12 V	10	10.8	
VOL	Low-level output voltage (see Note 2)	V <sub>IH</sub> = 2 V, See Figure 1	R <sub>L</sub> = 3 k $\Omega$ ,	V <sub>DD</sub> = 5 V, V <sub>SS</sub> = -5 V	-4.4	-4	V
				V <sub>DD</sub> = 12 V, V <sub>SS</sub> = -12 V	-10.7	-10	
I <sub>IH</sub>	High-level input current	V <sub>I</sub> = 5 V,	See Figure 2			1	$\mu\text{A}$
I <sub>IL</sub>	Low-level input current	V <sub>I</sub> = 0,	See Figure 2			-1	$\mu\text{A}$
I <sub>OS(H)</sub>	High-level short circuit output current‡	V <sub>I</sub> = 0.8 V,	V <sub>O</sub> = 0 or V <sub>SS</sub> , See Figure 1	-7.5	-12	-19.5	mA
I <sub>OS(L)</sub>	Low-level short circuit output current‡	V <sub>I</sub> = 2 V,	V <sub>O</sub> = 0 or V <sub>DD</sub> , See Figure 1	7.5	12	19.5	mA
I <sub>DD</sub>	Supply current from V <sub>DD</sub>	No load, All inputs at 2 V or 0.8 V		V <sub>DD</sub> = 5 V, V <sub>SS</sub> = -5 V	115	250	$\mu\text{A}$
				V <sub>DD</sub> = 12 V, V <sub>SS</sub> = -12 V	115	250	
I <sub>SS</sub>	Supply current from V <sub>SS</sub>	No load, All inputs at 2 V or 0.8 V		V <sub>DD</sub> = 5 V, V <sub>SS</sub> = -5 V	-115	-250	$\mu\text{A}$
				V <sub>DD</sub> = 12 V, V <sub>SS</sub> = -12 V	-115	-250	
r <sub>o</sub>	Output resistance	V <sub>DD</sub> = V <sub>SS</sub> = V <sub>CC</sub> = 0, V <sub>O</sub> = -2 V to 2 V, See Note 3		300	400		$\Omega$

† All typical values are at T<sub>A</sub> = 25°C.

‡ Not more than one output should be shorted at one time.

NOTES: 2. The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only.

3. Test conditions are those specified by EIA-232-D.

switching characteristics,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$ , T<sub>A</sub> = 25°C

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT	
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output§	R <sub>L</sub> = 3 to 7 k $\Omega$ , C <sub>L</sub> = 15 pF, See Figure 3			1.2	3	$\mu\text{s}$	
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output§				2.5	3.5	$\mu\text{s}$	
t <sub>TLH</sub>	Transition time, low-to-high-level output¶				0.53	2	3.2	$\mu\text{s}$
t <sub>THL</sub>	Transition time, high-to-low-level output¶				0.53	2	3.2	$\mu\text{s}$
t <sub>TLH</sub>	Transition time, low-to-high-level output#	R <sub>L</sub> = 3 to 7 k $\Omega$ , C <sub>L</sub> = 2500 pF, See Figure 3			1	2	$\mu\text{s}$	
t <sub>THL</sub>	Transition time, high-to-low level output#				1	2	$\mu\text{s}$	
SR	Output slew rate	R <sub>L</sub> = 3 to 7 k $\Omega$ , C <sub>L</sub> = 150 pF, See Figure 3		4	10	30	V/ $\mu\text{s}$	

§ t<sub>PHL</sub> and t<sub>PLH</sub> include the additional time due to on-chip slew rate and are measured at the 50% points.

¶ Measured between 10% and 90% points of output waveform.

# Measured between 3 V and -3 V points of output waveform (EIA-232-D conditions) with all unused inputs tied either high or low.



# SN65C1154, SN75C1154 QUAD LOW-POWER DRIVERS/RECEIVERS

SLLS151A – D3230, DECEMBER 1988 – REVISED MARCH 1993

## RECEIVER SECTION

**electrical characteristics over operating free-air temperature range,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
$V_{T+}$	Positive-going threshold voltage	See Figure 5	1.7	2.1	2.55	V	
$V_{T-}$	Negative-going threshold voltage	See Figure 5	0.65	1	1.25	V	
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )		600	1000		mV	
$V_{OH}$	High-level output voltage	$V_I = 0.75\text{ V}$ , $I_{OH} = -20\text{ }\mu\text{A}$ , See Figure 5 and Note 4	3.5			V	
			$V_{CC} = 4.5\text{ V}$		2.8 4.4		
		$V_I = 0.75\text{ V}$ , $I_{OH} = -1\text{ mA}$ , See Figure 5	$V_{CC} = 5\text{ V}$		3.8 4.9		
			$V_{CC} = 5.5\text{ V}$		4.3 5.4		
$V_{OL}$	Low-level output voltage	$V_I = 3\text{ V}$ , $I_{OL} = 3.2\text{ mA}$ , See Figure 5		0.17	0.4	V	
$I_{IH}$	High-level input current	$V_I = 25\text{ V}$	3.6	4.6	8.3	mA	
		$V_I = 3\text{ V}$	0.43	0.55	1		
$I_{IL}$	Low-level input current	$V_I = -25\text{ V}$	-3.6	-5	-8.3		
		$V_I = -3\text{ V}$	-0.43	-0.55	-1		
$I_{OS(H)}$	Short-circuit output at high level	$V_I = 0.75\text{ V}$ , $V_O = 0$ , See Figure 4		-8	-15	mA	
$I_{OS(L)}$	Short-circuit output at low level	$V_I = V_{CC}$ , $V_O = V_{CC}$ , See Figure 4		13	25	mA	
$I_{CC}$	Supply current from $V_{CC}$	No load, All inputs at 0 or 5 V	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$		400	600	$\mu\text{A}$
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$		400	600	

† All typical values are at  $T_A = 25^\circ\text{C}$ .

NOTE 4: If the inputs are left unconnected, the receiver interprets this as an input low and the receiver outputs will remain in the high state.

**switching characteristics,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$C_L = 50\text{ pF}$ , $R_L = 5\text{ k}\Omega$ , See Figure 6		3	4	$\mu\text{s}$
$t_{PHL}$	Propagation delay time, high-to-low-level output			3	4	$\mu\text{s}$
$t_{TLH}$	Transition time, low-to-high-level output			300	450	ns
$t_{THL}$	Transition time, high-to-low-level output			100	300	ns
$t_w(N)$	Duration of longest pulse rejected as noise‡		$C_L = 50\text{ pF}$ , $R_L = 5\text{ k}\Omega$	1		4

‡ The receiver ignores any positive- or negative-going pulse that is less than the minimum value of  $t_w(N)$  and accepts any positive- or negative-going pulse greater than the maximum of  $t_w(N)$ .



# SN65C1154, SN75C1154 QUAD LOW-POWER DRIVERS/RECEIVERS

SLLS151A - D3230, DECEMBER 1988 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

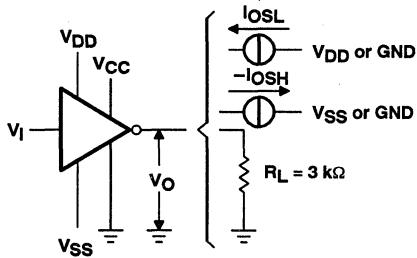


Figure 1. Driver Test Circuit  
 $V_{OH}$ ,  $V_{OL}$ ,  $I_{OS(L)}$ ,  $I_{OS(H)}$

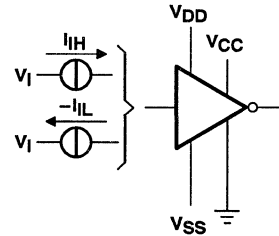
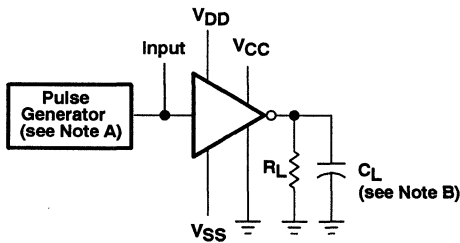
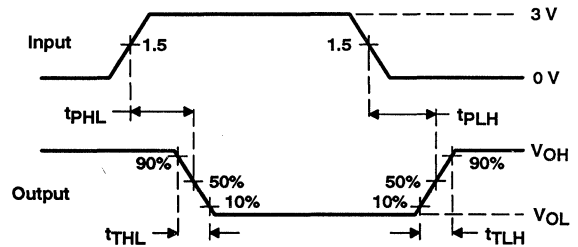


Figure 2. Driver Test Circuit,  $I_{IL}$ ,  $I_{IH}$



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu s$ , PRR = 20 kHz,  $Z_O = 50 \Omega$ ,  $t_r = t_f < 50 ns$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 3. Driver Test Circuit and Voltage Waveforms

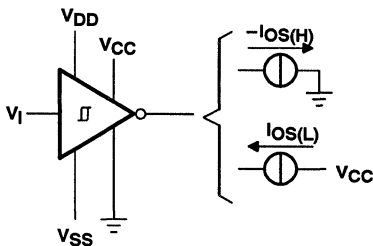


Figure 4. Receiver Test Circuit,  $I_{OS(H)}$ ,  $I_{OS(L)}$

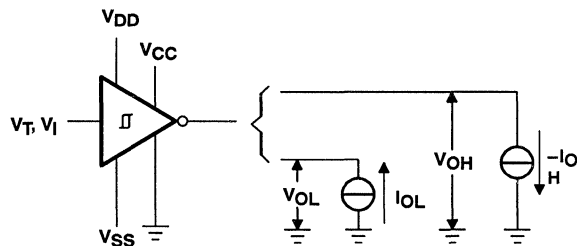
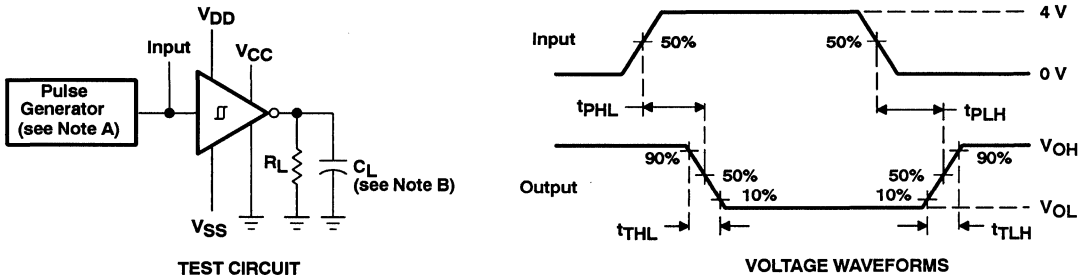


Figure 5. Receiver Test Circuit,  $V_T$ ,  $V_{OL}$ ,  $V_{OH}$

# SN65C1154, SN75C1154 QUAD LOW-POWER DRIVERS/RECEIVERS

SLLS151A – D3230, DECEMBER 1988 – REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu\text{s}$ ,  $\text{PRR} = 20 \text{ kHz}$ ,  $Z_O = 50 \Omega$ ,  $t_r = t_f < 50 \text{ ns}$ .  
B.  $C_L$  includes probe and jig capacitance.

**Figure 6. Receiver Test Circuit and Voltage Waveforms**





# SN65C1167, SN65C1168, SN75C1167, SN75C1168 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS159 – D4062, MARCH 1993

- Meets EIA Standards RS-422-A and CCITT Recommendation V.11
- BICMOS Process Technology
- Low Supply Current Requirements  
9 mA Max
- Low Pulse Skew
- Receiver Input Impedance . . . 17 k $\Omega$  Typ
- Receiver Input Sensitivity . . .  $\pm 200$  mV
- Receiver Common-Mode Input Voltage Range of  $-7$  V to 7 V
- Operates From Single 5-V Supply
- Glitch-Free Power-Up/Power-Down Protection
- Receiver 3-State Outputs Active-Low Enable for SN751167 Only
- Improved Replacement for the MC34050 and MC34051

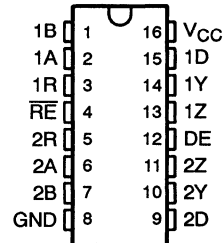
## description

The SN75C1167 and SN75C1168 dual drivers and receivers are monolithic integrated circuits designed for balanced transmission lines and meets EIA Standards RS-422-A, CCITT recommendations V.10, V.11, X.26, and X.27.

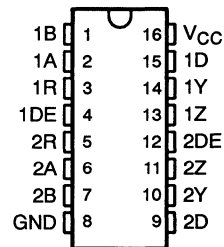
The SN75C1167 combines dual 3-state differential line driver and 3-state differential line receivers both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, which can be externally connected together to function as direction control. The SN75C1168 drivers each have individual active-high enable.

The SN65C1167 and SN65C1168 are characterized from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75C1167 and SN75C1168 are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

SN65C1167 . . . DB $\dagger$ , N, OR NS $\dagger$  PACKAGE  
(TOP VIEW)



SN75C1177 . . . DB $\dagger$ , N, OR NS $\dagger$  PACKAGE  
(TOP VIEW)



$\dagger$  The DB and NS packages are only available left-ended taped and reeled (order device SN\_5C116\_DBLE or SN\_5C116\_NSLE).

## Function Tables

EACH DRIVER

INPUT D	ENABLE DE	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	Z	Z

EACH RECEIVER

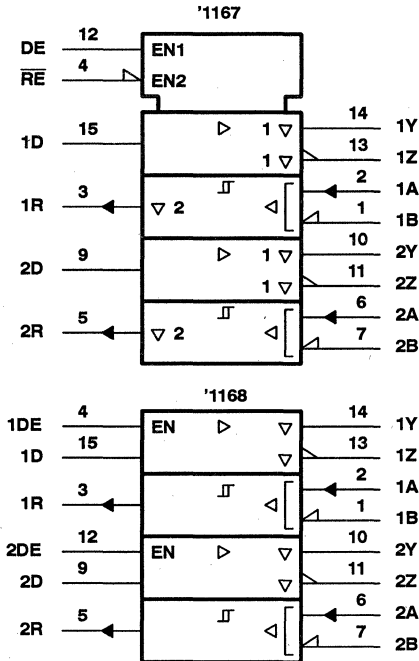
DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2$ V	L	H
$-0.2$ V $< V_{ID} < 0.2$ V	L	?
$V_{ID} \leq -0.2$ V	L	L
X	H	Z
Open	L	H

H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = high impedance (off)

# SN65C1167, SN65C1168, SN75C1167, SN75C1168 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

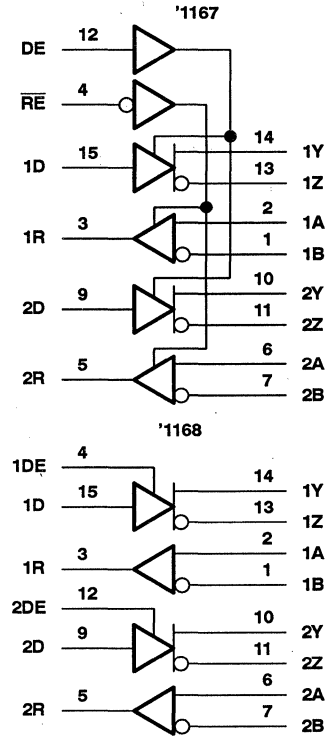
SLLS159 – D4062, MARCH 1993

## logic symbols†

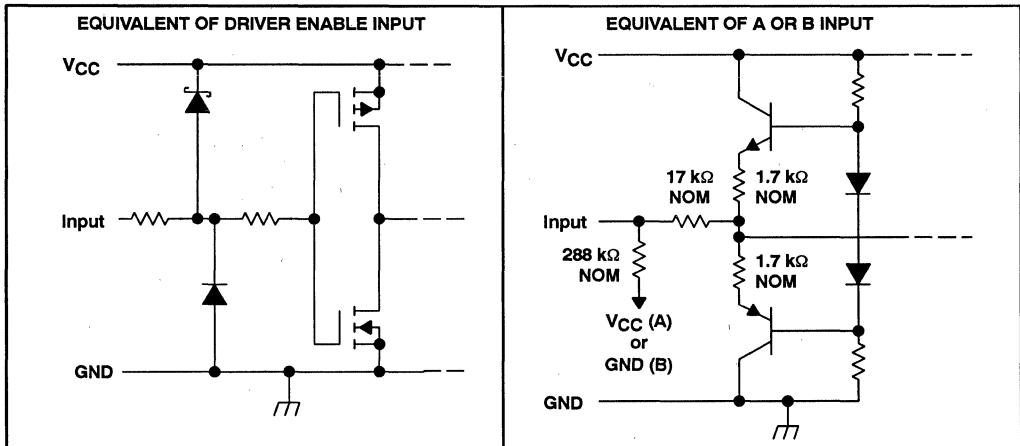


† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagrams (positive logic)



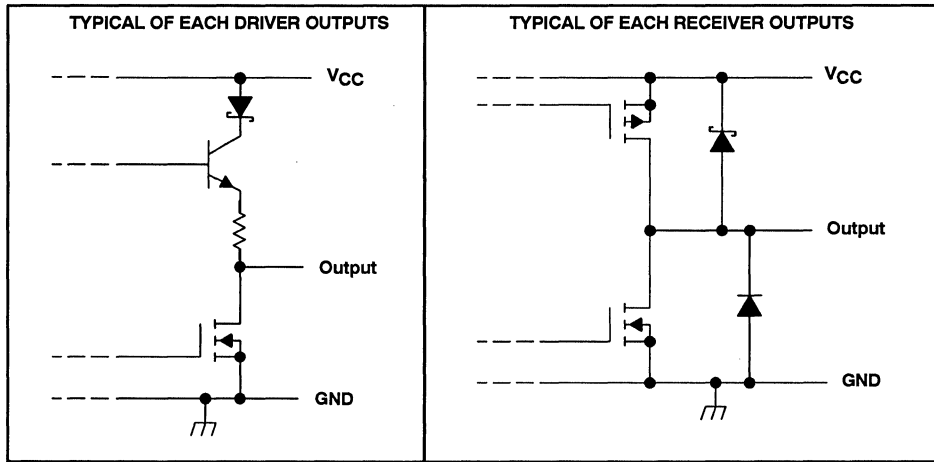
## schematics of inputs



# SN65C1167, SN65C1168, SN75C1167, SN75C1168 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS159 – D4062, MARCH 1993

## schematics of outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, V <sub>CC</sub> (see Note 1)	–0.5 V to 7 V
Input voltage range, V <sub>I</sub>	–0.5 V to V <sub>CC</sub> + 0.5 V
Input voltage range, A or B, Receiver	–11 V to 14 V
Differential input voltage range, V <sub>ID</sub> , Receiver (see Note 2)	–14 V to 14 V
Output voltage range, V <sub>O</sub> , Driver	–5 V to 7 V
Clamp current range, I <sub>IK</sub> or I <sub>OK</sub> , Driver	±20 mA
Output current range, I <sub>O</sub> , Driver	±150 mA
Supply current, I <sub>CC</sub>	200 mA
GND current	–200 mA
Output current range, I <sub>O</sub> , Receiver	±25 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T <sub>A</sub> : SN75C1167, SN75C1168	0°C to 70°C
SN65C1167, SN65C1168	–40°C to 85°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to the network ground terminal.  
 2. Differential input voltage is measured at the noninverting terminal with respect to the inverting terminal.

DISSIPATION RATING TABLE

PACKAGE	T <sub>A</sub> ≤ 25°C	OPERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C	T <sub>A</sub> = 85°C
	POWER RATING		POWER RATING	POWER RATING
DB	781 mW	6.2 mW/°C	502 mW	409 mW
N	1250 mW	10 mW/°C	736 mW	650 mW
NS	625 mW	5 mW/°C	445 mW	325 mW

  
**TEXAS  
INSTRUMENTS**

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# SN65C1167, SN65C1168, SN75C1167, SN75C1168 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

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## recommended operating conditions

PARAMETER		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.5	5	5.5	V
Common-mode input voltage, $V_{IC}$ (see Note 3)	Receiver			$\pm 7$	V
Differential input voltage, $V_{ID}$	Receiver			$\pm 7$	V
High-level input voltage, $V_{IH}$	Except A, B	2			V
Low-level input voltage, $V_{IL}$	Except A, B			0.8	V
High-level output current, $I_{OH}$	Receiver			-6	mA
	Driver			-20	
Low-level output current, $I_{OL}$	Receiver			6	mA
	Driver			20	
Operating free-air temperature, $T_A$	SN65C1167 and SN65C1168	-40		85	°C
	SN75C1167 and SN75C1168	0		70	

NOTE 3: Refer to EIA standards RS-422-A for exact conditions.

# SN65C1167, SN65C1168, SN75C1167, SN75C1168 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS159 – D4062, MARCH 1993

## DRIVER SECTION

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
V <sub>IK</sub>	Input clamp voltage	I <sub>I</sub> = -18 mA			-1.5	V	
V <sub>OH</sub>	High-level output voltage	V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V, I <sub>OH</sub> = -20 mA	2.4	3.4		V	
V <sub>OL</sub>	Low-level output voltage	V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V, I <sub>OL</sub> = 20 mA		0.2	0.4	V	
V <sub>OD1</sub>	Differential output voltage	I <sub>O</sub> = 0 mA	2		6	V	
V <sub>OD2</sub>	Differential output voltage	R <sub>L</sub> = 100 Ω, See Figure 1 and Note 3	2	3.1		V	
Δ V <sub>OD</sub>	Change in magnitude of differential output voltage				±0.4	V	
V <sub>OC</sub>	Common-mode output voltage				±3	V	
Δ V <sub>OC</sub>	Change in magnitude of common-mode output voltage				±0.4	V	
I <sub>O(OFF)</sub>	Output current with power off (see Note 3)		V <sub>CC</sub> = 0 V			100	μA
		V <sub>O</sub> = 6 V					
		V <sub>O</sub> = -0.25 V			-100	μA	
I <sub>OZ</sub>	High-impedance-state output current	V <sub>O</sub> = 2.5 V			20	μA	
		V <sub>O</sub> = 5 V			-20	μA	
I <sub>IH</sub>	High-level input current	V <sub>I</sub> = V <sub>CC</sub> or V <sub>IH</sub>			1	μA	
I <sub>IL</sub>	Low-level input current	V <sub>I</sub> = GND or V <sub>IL</sub>			-1	μA	
I <sub>OS</sub>	Short-circuit output current	V <sub>O</sub> = V <sub>CC</sub> or GND, See Note 4	-30		-150	mA	
I <sub>CC</sub>	Supply current (total package)	No load, Enabled			4	6	mA
		V <sub>I</sub> = V <sub>CC</sub> or GND			5	9	
		V <sub>I</sub> = 2.4 or 0.5 V, See Note 5					

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output	R <sub>1</sub> = R <sub>2</sub> = 50 Ω, R <sub>3</sub> = 500 Ω,		7	12	ns
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	C <sub>1</sub> = C <sub>2</sub> = C <sub>3</sub> = 40 pF, S1 is open,		7	12	ns
t <sub>sk(p)</sub>	Pulse skew	See Figure 2		0.5	4	ns
t <sub>r</sub>	Rise time	R <sub>1</sub> = R <sub>2</sub> = 50 Ω, R <sub>3</sub> = 500 Ω,		5	10	ns
t <sub>f</sub>	Fall time	C <sub>1</sub> = C <sub>2</sub> = C <sub>3</sub> = 40 pF, S1 is open,		5	10	ns
t <sub>PZH</sub>	Output enable time to high level	R <sub>1</sub> = R <sub>2</sub> = 50 Ω, R <sub>3</sub> = 500 Ω,		10	19	ns
t <sub>PZL</sub>	Output enable time to low level	C <sub>1</sub> = C <sub>2</sub> = C <sub>3</sub> = 40 pF, S1 is closed,		10	19	ns
t <sub>PHZ</sub>	Output disable time from low level	R <sub>1</sub> = R <sub>2</sub> = 50 Ω, R <sub>3</sub> = 500 Ω,		7	16	ns
t <sub>PLZ</sub>	Output disable time from high level	C <sub>1</sub> = C <sub>2</sub> = C <sub>3</sub> = 40 pF, S1 is closed,		7	16	ns
C <sub>i</sub>	Input capacitance			6		pF

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

NOTES: 3. Refer to EIA standards RS-422-A for exact conditions.

4. Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

5. Measured per input while the other inputs are at V<sub>CC</sub> or GND.



# SN65C1167, SN65C1168, SN75C1167, SN75C1168 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS159 – D4062, MARCH 1993

## RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage, differential input				0.2	V
$V_{T-}$	Negative-going threshold voltage, differential input		-0.2‡			V
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )			60		mV
$V_{IK}$	Input clamp voltage, $\overline{RE}$	'1167 $I_I = -18$ mA			-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 200$ mV, $I_{OH} = -6$ mA	3.8	4.2		V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200$ mV, $I_{OL} = 6$ mA		0.1	0.3	V
$I_{OZ}$	High-impedance-state output current	'1167 $V_O = V_{CC}$ or GND		$\pm 0.5$	$\pm 5$	$\mu$ A
$I_I$	Line input current	Other input at 0 V			1.5	mA
		$V_I = 10$ V			-2.5	
$I_I$	Enable input current, $\overline{RE}$	'1167 $V_I = V_{CC}$ or GND			$\pm 1$	$\mu$ A
$r_i$	Input resistance	$V_{IC} = -7$ V to 7 V, Other input at 0 V	4	17		k $\Omega$
$I_{CC}$	Supply current (total package)	No load, Enabled			4	mA
		$V_I = V_{CC}$ or GND			6	
		$V_{IH} = 2.4$ V or 0.5 V, See Note 5		5	9	

switching characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	See Figure 5	9	17	27	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output		9	17	27	ns
$t_{TLH}$	Transition time, low-to-high-level output	$V_{IC} = 0$ V, See Figure 5		4	9	ns
$t_{THL}$	Transition time, high-to-low-level output			4	9	ns
$t_{PZH}$	Output enable time to high level	'1167 $R_L = 1$ k $\Omega$ , See Figure 6		13	22	ns
$t_{PZL}$	Output enable time to low level	'1167		13	22	ns
$t_{PHZ}$	Output disable time from high level	'1167		13	22	ns
$t_{PLZ}$	Output disable time from low level	'1167 $R_L = 1$ k $\Omega$ , See Figure 6		13	22	ns

† All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^\circ$ C.

‡ The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 5: Measured per input while the other inputs are at  $V_{CC}$  or GND.

## PARAMETER MEASUREMENT INFORMATION

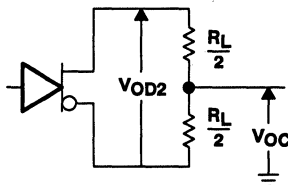


Figure 1. Driver Test Circuit,  $V_{OD}$  and  $V_{OC}$

PARAMETER MEASUREMENT INFORMATION

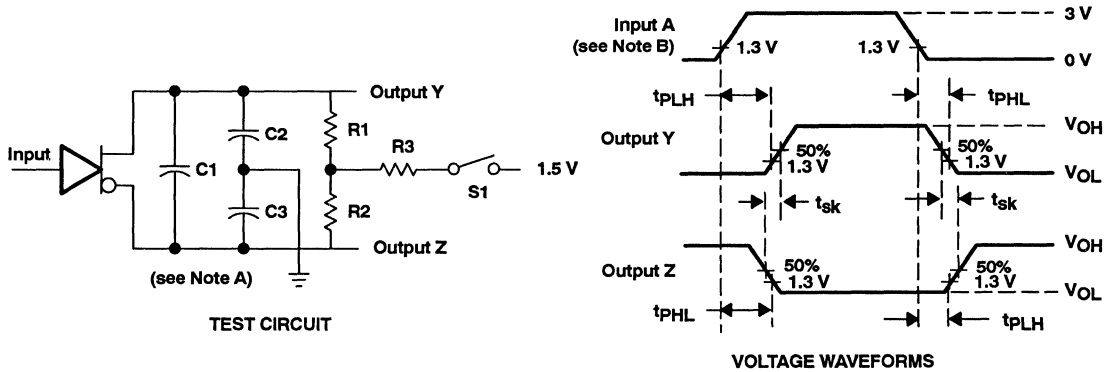


Figure 2. Driver Test Circuit and Voltage Waveforms

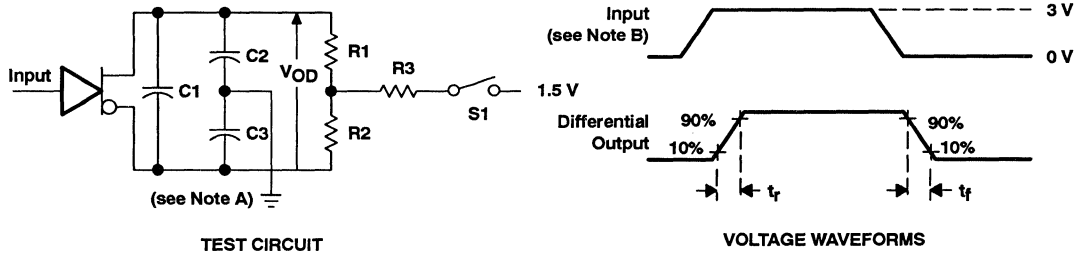


Figure 3. Driver Test Circuit and Voltage Waveforms

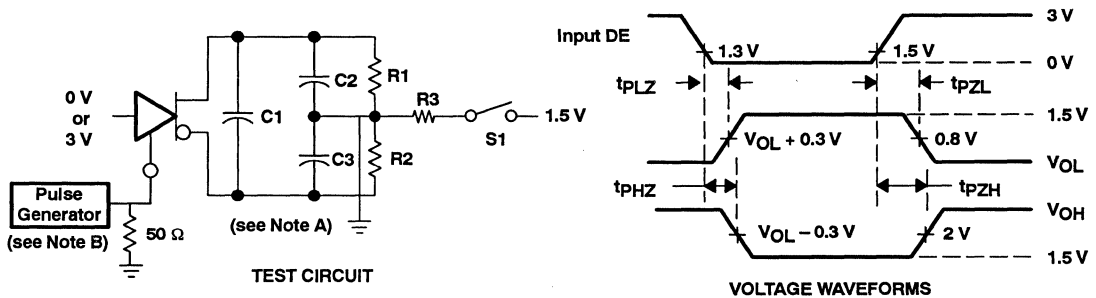


Figure 4. Driver Test Circuit and Voltage Waveforms

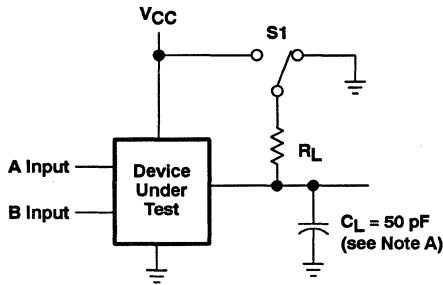
NOTES: A. C1 ~ C3 includes probe and jig capacitance.  
 B. The input pulse is supplied by a generator having the following characteristics: PRR = 1 MHz, duty cycle = 50%,  $t_r = t_f \leq 6$  ns.



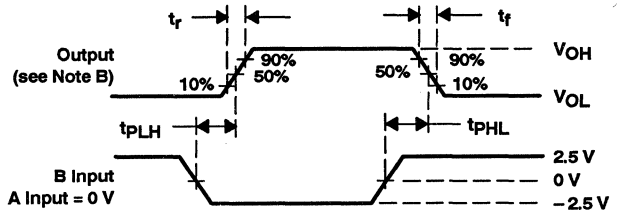
# SN65C1167, SN65C1168, SN75C1167, SN75C1168 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS159 – D4062, MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

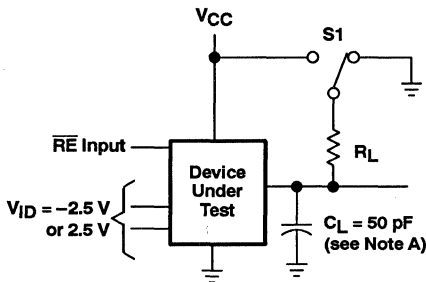


TEST CIRCUIT



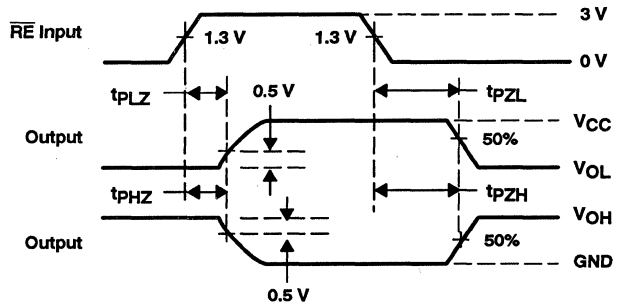
VOLTAGE WAVEFORMS

Figure 5. Receiver Test Circuit and Voltage Waveforms



$t_{pZL}$ ,  $t_{pLZ}$  Measurement: S1 to  $V_{CC}$   
 $t_{pZH}$ ,  $t_{pHZ}$  Measurement: S1 to GND

TEST CIRCUIT



VOLTAGE WAVEFORMS

Figure 6. Receiver Test Circuit and Voltage Waveforms

NOTES: A.  $C_L$  includes probe and jig capacitance.

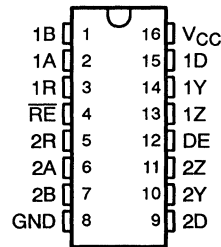
B. The pulse generator has the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r = t_f \leq$  6 ns.

# SN751177, SN751178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

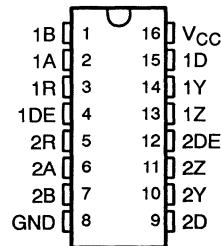
SLLS059A – D3381, FEBRUARY 1990 – REVISED JANUARY 1993

- Meets EIA Standards RS-422-A, RS485
- Meets CCITT Recommendations V.10, V.11, X.26, X.27
- Designed for Multipoint Bus Transmission on Long Bus Lines in Noise Environments
- Driver Common-Mode Output Voltage Range of  $-7\text{ V}$  to  $12\text{ V}$
- Driver Positive- and Negative-Current Limiting
- Thermal Shutdown Protection
- Driver 3-State Outputs Active-High Enable
- Receiver Common-Mode Input Voltage Range of  $-12\text{ V}$  to  $12\text{ V}$
- Receiver Input Sensitivity . . .  $\pm 200\text{ mV}$
- Receiver Hysteresis . . .  $50\text{ mV Typ}$
- Receiver High-Input-Impedance  $12\text{ k}\Omega\text{ Min}$
- Receiver 3-State Outputs Active-Low Enable for SN751177 Only
- Operates From Single 5-V Supply

SN751177 . . . N PACKAGE  
(TOP VIEW)



SN751178 . . . N OR NS† PACKAGE  
(TOP VIEW)



† The NS package is only available left-end taped and reeled (order device SN751177NSLE).

## description

The SN751177 and SN751178 dual differential drivers and receivers are monolithic integrated circuits that are designed for balanced multipoint bus transmission at rates up to 10 Mbits per second. They are designed to improve the performance of full-duplex data communications over long bus lines and meet EIA standards RS-422-A, RS-485 and several CCITT recommendations.

The SN751177 and SN751178 driver outputs provide limiting for both positive and negative currents and thermal shutdown protection from line fault conditions on the transmission bus line.

The receiver features high input impedance of  $12\text{ k}\Omega$ , an input sensitivity of  $\pm 200\text{ mV}$  over a common-mode input voltage range of  $-12\text{ V}$  to  $12\text{ V}$  and typical input hysteresis of  $50\text{ mV}$ . Fail-safe design ensures that if the receiver inputs are open, the receiver outputs will always be high.

The SN751177 and SN751178 are characterized for operation from  $-20^\circ\text{C}$  to  $85^\circ\text{C}$ .

## Function Tables

SN751177, SN751178  
EACH DRIVER

INPUT D	ENABLE DE	OUTPUTS	
		Y	Z
H	H	H	L
L	L	L	H
X	X	Z	Z

SN751177  
EACH RECEIVER

DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2\text{ V}$	H	H
$-0.2\text{ V} < V_{ID} < 0.2\text{ V}$	L	L
$V_{ID} \geq -0.2\text{ V}$	X	X
X	H	Z
Open	L	H

SN751178  
EACH RECEIVER

DIFFERENTIAL INPUTS A – B	OUTPUT R
$V_{ID} \geq 0.2\text{ V}$	H
$-0.2\text{ V} < V_{ID} < 0.2\text{ V}$	?
$V_{ID} \geq -0.2\text{ V}$	L

H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = high impedance (off)

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

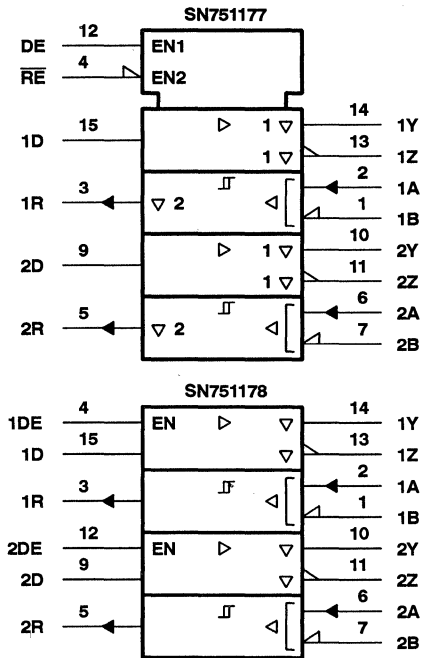
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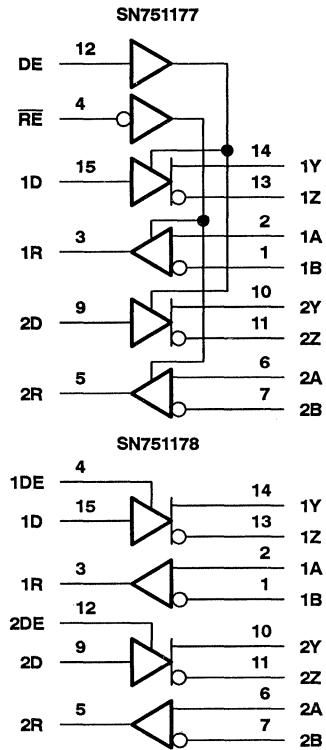
# SN751177, SN751178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS059A - D3381, FEBRUARY 1990 - REVISED JANUARY 1993

## logic symbols†

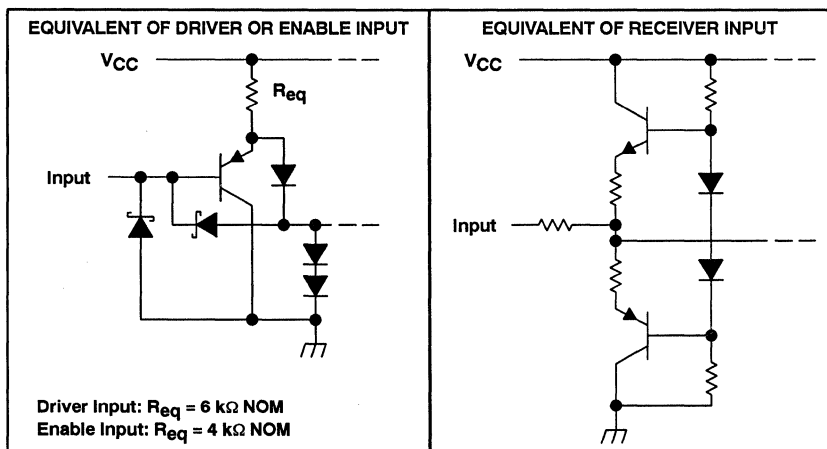


## logic diagrams (positive logic)



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematics of inputs

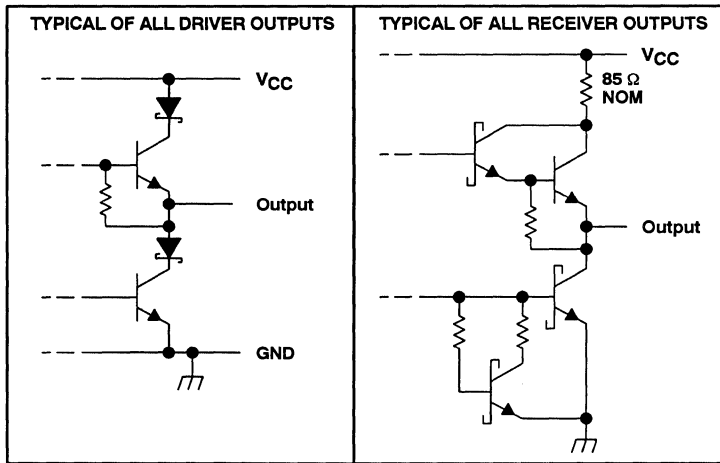


All resistors values are nominal.

# SN751177, SN751178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS059A – D3381, FEBRUARY 1990 – REVISED JANUARY 1993

## schematics of outputs



All resistors values are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage, DE, $\overline{RE}$ , and D inputs	7 V
Input voltage range, Receiver, A or B inputs	–25 V to 25 V
Receiver differential input voltage range (see Note 2)	–25 V to 25 V
Output voltage range, Driver	–10 V to 15 V
Receiver low-level output current	50 mA
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 3)	1150 mW
Operating free-air temperature range, $T_A$	–20°C to 85°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to the network ground terminal.  
 2. Differential input voltage is measured at the noninverting terminal with respect to the inverting terminal.  
 3. For operation above 25°C free-air temperature, derate to 736 mW at 70°C at the rate of 9.2 mW/°C.

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
High-level input voltage, $V_{IH}$	DE, $\overline{RE}$ , and D inputs	2			V
Low-level input voltage, $V_{IL}$				0.8	V
Common-mode output voltage, $V_{OC}$	Driver	–7†		12	V
High-level output current, $I_{OH}$				–60	mA
Low-level output current, $I_{OL}$				60	mA
Common-mode input voltage, $V_{IC}$	Receiver			±12	V
Differential input voltage, $V_{ID}$				±12	V
High-level output current, $I_{OH}$				–400	μA
Low-level output current, $I_{OL}$				16	mA
Operating free-air temperature, $T_A$		–20		85	°C

† The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode output and threshold voltage levels only.



# SN751177, SN751178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS059A – D3381, FEBRUARY 1990 – REVISED JANUARY 1993

## DRIVER SECTIONS

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{IK}$ Input clamp voltage	$I_I = -18 \text{ mA}$			-1.5	V
$V_{OH}$ High-level output voltage	$V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ , $I_{OH} = -33 \text{ mA}$		3.7		V
$V_{OL}$ Low-level output voltage	$V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$ , $I_{OH} = 33 \text{ mA}$		1.1		V
$ V_{OD1} $ Differential output voltage	$I_O = 0$	1.5		6	V
$ V_{OD2} $ Differential output voltage	$R_L = 100 \Omega$ , See Figure 1	2			V
		$1/2 V_{OD1}$			
$ V_{OD2} $ Differential output voltage	$R_L = 54 \Omega$ , See Figure 1	1.5		5	V
		See Note 4			
$V_{OD3}$ Differential output voltage	See Note 4	1.5		5	V
$\Delta V_{OD} $ Change in magnitude of differential output voltage (see Note 5)	$R_L = 54 \Omega$ or $100 \Omega$ , See Figure 1			$\pm 0.2$	V
$V_{OC}$ Common-mode output voltage			$-1\ddagger$	3	V
$\Delta V_{OC} $ Change in magnitude of common-mode output voltage (see Note 5)				$\pm 0.2$	V
$I_O$ Output current with power off	$V_{CC} = 0$ , $V_O = -7 \text{ V to } 12 \text{ V}$			$\pm 100$	$\mu\text{A}$
$I_{OZ}$ High-impedance-state output current	$V_O = -7 \text{ V to } 12 \text{ V}$			$\pm 100$	$\mu\text{A}$
$I_{IH}$ High-level input current	$V_{IH} = 2.7 \text{ V}$			20	$\mu\text{A}$
$I_{IL}$ Low-level input current	$V_{IL} = 0.4 \text{ V}$			-100	$\mu\text{A}$
$I_{OS}$ Short-circuit output current (see Note 6)	$V_O = -7 \text{ V}$			-250	mA
	$V_O = V_{CC}$			250	
	$V_O = 12 \text{ V}$			250	
$I_{CC}$ Supply current	No load	Outputs enabled		80	mA
		Outputs disabled		50	

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode output and threshold voltage levels only.

NOTES: 4. See EIA Standard RS-485 Figure 3.5, Test Termination Measurement 2.

5.  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

6. Not more than one output should be shorted at a time and the duration of the short circuit should not exceed one second.

## switching characteristics at $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$ , $C_L = 50 \text{ pF}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{dD}$ Differential output delay time	$R_L = 54 \Omega$ , See Figure 3		20	25	ns
$t_{tD}$ Differential output transition time			27	35	ns
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 27 \Omega$ , See Figure 4		20	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			20	25	ns
$t_{PZH}$ Output enable time to high level	$R_L = 110 \Omega$ , See Figure 5		80	120	ns
$t_{PZL}$ Output enable time to low level	$R_L = 110 \Omega$ , See Figure 6		40	60	ns
$t_{PHZ}$ Output disable time from high level	$R_L = 110 \Omega$ , See Figure 5		90	120	ns
$t_{PLZ}$ Output disable time from low level	$R_L = 110 \Omega$ , See Figure 6		30	45	ns

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# SN751177, SN751178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS059A – D3381, FEBRUARY 1990 – REVISED JANUARY 1993

## SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422A	RS-485
$ V_{OD1} $	$V_O$	$V_O$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$\Delta  V_{OD} $	$  V_t  -  \bar{V}_t  $	$  V_t  -  \bar{V}_t  $
$V_{OC}$	$ V_{OS} $	$ V_{OS} $
$\Delta  V_{OC} $	$ V_{OS} - \bar{V}_{OS} $	$ V_{OS} - \bar{V}_{OS} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

## RECEIVER SECTIONS

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7 V$ ,	$I_O = -0.4 mA$			0.2	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5 V$ ,	$I_O = 16 mA$	-0.2‡			V
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )				50		mV
$V_{IK}$	Enable clamp voltage	SN751177	$I_I = -18 mA$			-1.5	V
$V_{OH}$	High-level output voltage	$V_{ID} = 200 mV$ ,	$I_{OH} = -400 \mu A$	2.7			V
$V_{OL}$	Low-level output voltage	$V_{ID} = -200 mV$	$I_{OL} = 8 mA$ $I_{OL} = 16 mA$			0.45 0.5	V
$I_{OZ}$	High-impedance-state output current	SN751177	$V_O = 0.4 V$ to $2.4 V$			$\pm 20$	$\mu A$
$I_I$	Line input current (see Note 7)		Other input at 0 V	$V_I = 12 V$ $V_I = -7 V$		1 -0.8	mA
$I_{IH}$	High-level enable input current	SN751177	$V_{IH} = 2.7 V$			20	$\mu A$
$I_{IL}$	Low-level enable input current	SN751177	$V_{IL} = 0.4 V$			-100	$\mu A$
$I_{OS}$	Short-circuit output current (see Note 6)			-15		-85	$\mu A$
$I_{CC}$	Supply current		No load, Outputs enabled		80	110	mA
$r_i$	Input resistance				12		k $\Omega$

† All typical values are at  $V_{CC} = 5 V$  and  $T_A = 25^\circ C$ .

‡ The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode output and threshold voltage levels only.

NOTES: 6. Not more than one output should be shorted at a time, and the duration of the short circuit should not exceed one second.

7. Refer to EIA standards RS-422-A, RS-423-A, RS-485-A for exact conditions.

switching characteristics at  $V_{CC} = 5 V$ ,  $C_L = 15 pF$ ,  $T_A = 25^\circ C$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$V_{ID} = -1.5 V$ to $1.5 V$ ,			20	35	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output	See Figure 7			22	35	ns
$t_{PZH}$	Output enable time to high level	SN751177	See Figure 8		17	25	ns
$t_{PZL}$	Output enable time to low level				20	27	ns
$t_{PHZ}$	Output disable time from high level				25	40	ns
$t_{PLZ}$	Output disable time from low level				30	40	ns

**TEXAS**  
**INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

2-921

# SN751177, SN751178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS059A—D3381, FEBRUARY 1990—REVISED JANUARY 1993

## PARAMETER MEASUREMENT INFORMATION

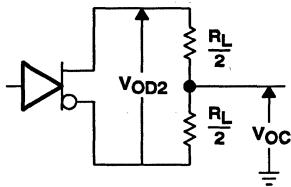


Figure 1. Driver Test Circuit,  $V_{OD}$  and  $V_{OC}$

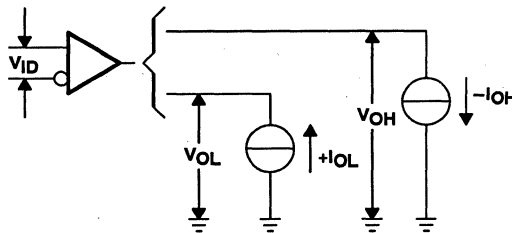
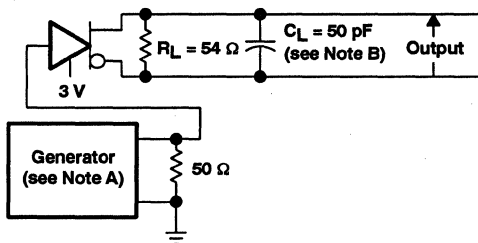
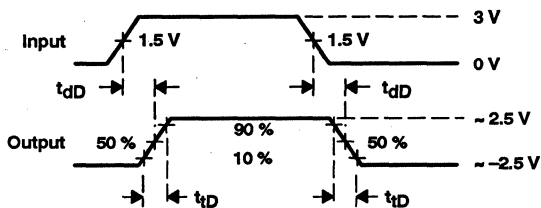


Figure 2. Receiver Test Circuit,  $V_{OH}$  and  $V_{OL}$

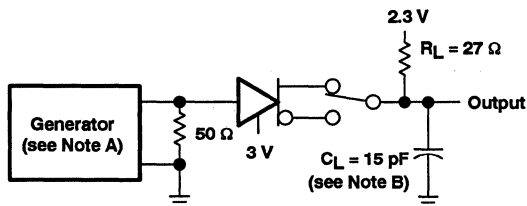


TEST CIRCUIT

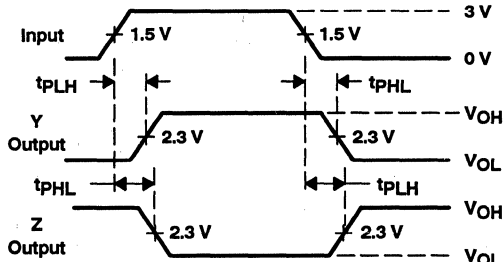


VOLTAGE WAVEFORMS

Figure 3. Driver Differential Output Delay and Transition Time Test Circuit and Voltage Waveforms



TEST CIRCUIT



VOLTAGE WAVEFORMS

Figure 4. Driver Propagation Delay Time Test Circuit and Voltage Waveforms

NOTES: A. The pulse generator has the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $Z_0 = 50 \Omega$ ,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns.  
B.  $C_L$  includes probe and jig capacitance.

PARAMETER MEASUREMENT INFORMATION

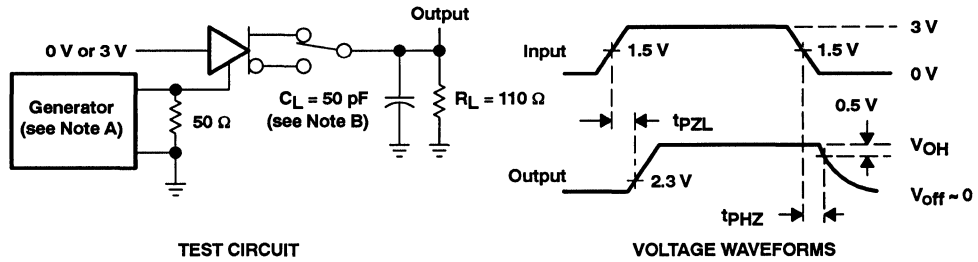


Figure 5. Driver Enable and Disable Time Test Circuit and Voltage Waveforms

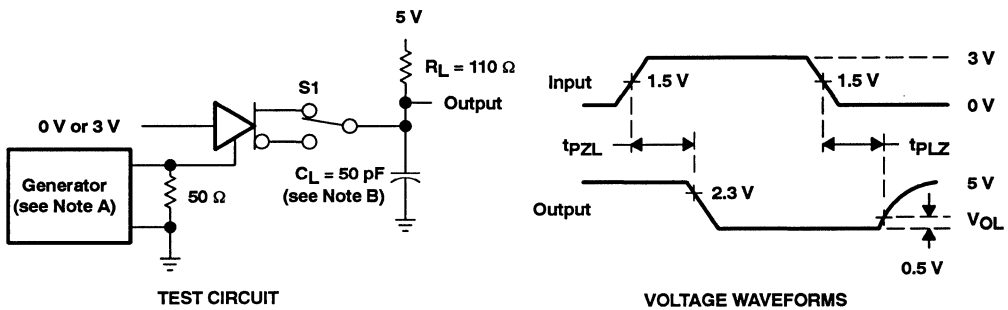


Figure 6. Driver Enable and Disable Time Test Circuit and Voltage Waveforms

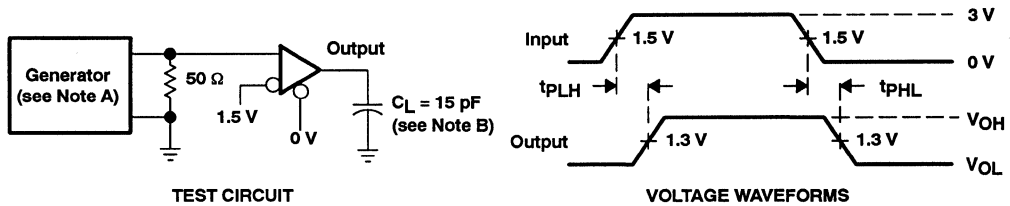


Figure 7. Receiver Propagation Delay Time Test Circuit and Voltage Waveforms

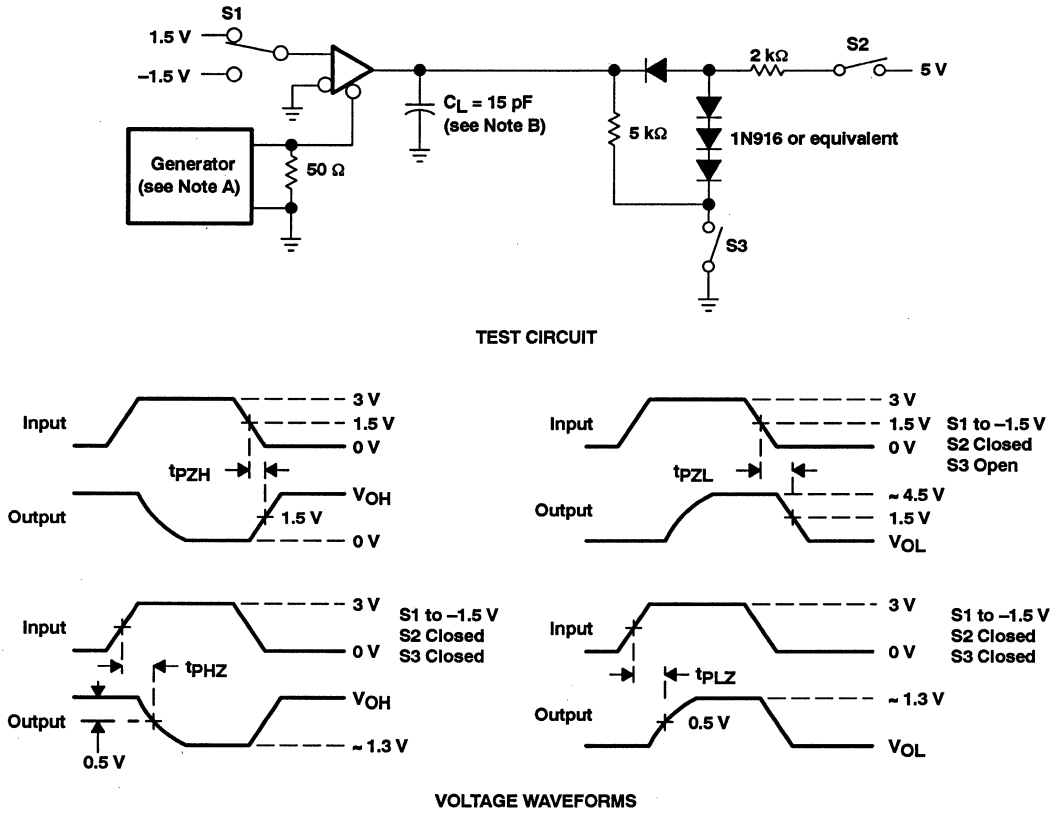
NOTES: A. The pulse generator has the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $Z_O = 50 \Omega$ ,  $t_r \leq 6$  ns,  $t_f \leq 6$  ns.  
 B.  $C_L$  includes probe and jig capacitance.



# SN751177, SN751178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS059A - D3381, FEBRUARY 1990 - REVISED JANUARY 1993

## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics:  $PRR \leq 1 \text{ MHz}$ , 50% duty cycle,  $Z_O = 50 \Omega$ ,  $t_r \leq 6 \text{ ns}$ ,  $t_f \leq 6 \text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.

**Figure 8. Receiver Output Enable and Disable Time Test Circuit and Voltage Waveforms**

# SN75ALS1177, SN75ALS1178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS154 – D4061, MARCH 1993

- Meet EIA Standards RS-422-A and RS485
- Meet CCITT Recommendations V.10, V.11, X.26, X.27
- Designed for Multipoint Bus Transmission on Long Bus Lines in Noisy Environments
- Low Supply Current Requirement  
50 mA Max
- Driver Positive- and Negative-Current Limiting
- Driver Common-Mode Output Voltage Range of  $-7\text{ V}$  to  $12\text{ V}$
- Thermal Shutdown Protection
- Driver 3-State Outputs Active-High Enable
- Receiver Common-Mode Input Voltage Range of  $-12\text{ V}$  to  $12\text{ V}$
- Receiver Input Sensitivity . . .  $\pm 200\text{ mV}$
- Receiver Hysteresis . . .  $50\text{ mV Typ}$
- Receiver High Input Impedance  
12 k $\Omega$  Min
- Receiver 3-State Outputs Active-Low Enable for SN75ALS1177 Only
- Operates From Single 5-V Supply

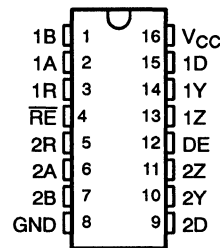
## description

The SN75ALS1177 and SN75ALS1178 dual differential drivers and receivers are monolithic integrated circuits designed for bidirectional data communication on multipoint bus transmission lines. They are designed for balanced transmission lines and meet EIA standards RS-422-A, RS-485, and CCITT recommendations V.10, V.11, X.26, X.27.

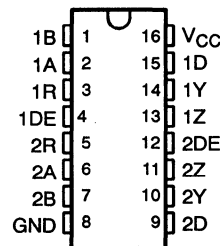
The SN75ALS1177 combines dual 3-state differential line drivers and dual 3-state differential input line receivers, both of which operate from a single 5-V power supply. The drivers and receivers have active-high and active-low enables, respectively, which can be externally connected together to function as direction control. The SN75ALS1178 drivers each have an individual active-high enable. Fail-safe design ensures that if the receiver inputs are open, the receiver outputs will always be high.

The SN75ALS1177 and SN75ALS1178 are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

SN75ALS1177 . . . N OR NS† PACKAGE  
(TOP VIEW)



SN75ALS1178 . . . N OR NS† PACKAGE  
(TOP VIEW)



† The NS package is only available in left-end taped and reeled (SN74ALS1177NSLE and SN74ALS1178SNLE).

## Function Tables

SN75ALS1177, SN75ALS1178  
EACH DRIVER

INPUT D	ENABLE DE	OUTPUTS	
		Y	Z
H	H	H	L
L	H	L	H
X	L	Z	Z

SN75ALS1177, EACH RECEIVER

DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2\text{ V}$	L	H
$-0.2\text{ V} < V_{ID} < 0.2\text{ V}$	L	?
$V_{ID} \leq -0.2\text{ V}$	L	L
X	H	Z
Open	L	H

SN75ALS1178, EACH RECEIVER

DIFFERENTIAL INPUTS A – B	OUTPUT R
$V_{ID} \geq 0.2\text{ V}$	H
$-0.2\text{ V} < V_{ID} < 0.2\text{ V}$	?
$V_{ID} \leq -0.2\text{ V}$	L
Open	H

H = high level, L = low level, ? = indeterminate,  
X = irrelevant, Z = high impedance (off)

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS**  
**INSTRUMENTS**

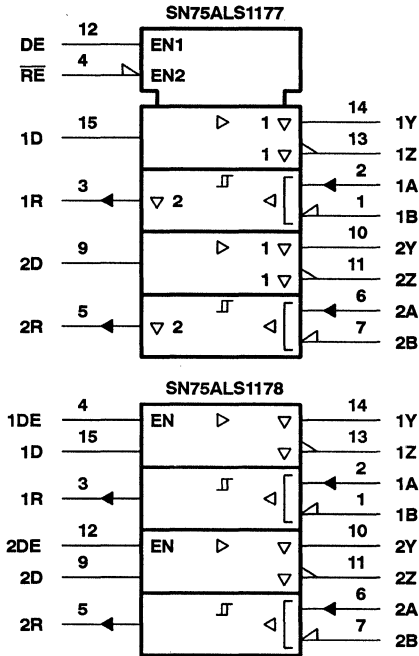
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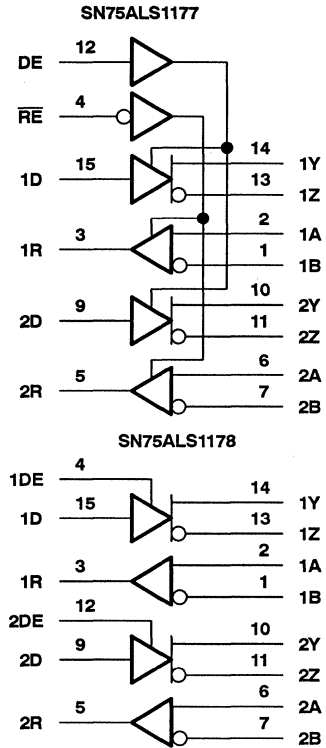
# SN75ALS1177, SN75ALS1178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS154 – D4061, MARCH 1993

## logic symbols†

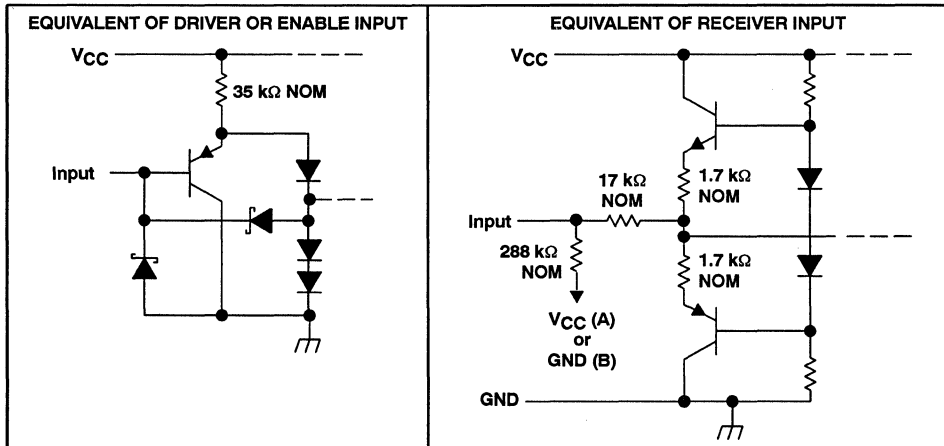


## logic diagrams (positive logic)



† These symbols are in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

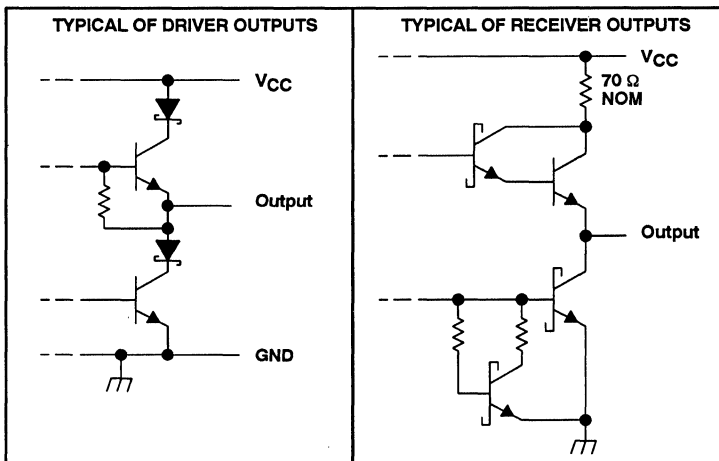
## equivalent schematics



# SN75ALS1177, SN75ALS1178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS154 – D4061, MARCH 1993

## schematics of outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC</sub> (see Note 1)	7 V
Input voltage, DE, RE, and D inputs	7 V
Output voltage range, Driver	-9 V to 14 V
Input voltage range, Receiver	-14 V to 14 V
Receiver differential input voltage range (see Note 2)	-14 V to 14 V
Receiver low-level output current	50 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T <sub>A</sub>	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to the network ground terminal.  
 2. Differential input voltage is measured at the noninverting terminal with respect to the inverting terminal.

**DISSIPATION RATING TABLE**

PACKAGE	T <sub>A</sub> = 25°C POWER RATING	OPERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
N	1150 mW	9.2 mW/°C	736 mW
NS	625 mW	4.0 mW/°C	445 mW



# SN75ALS1177, SN75ALS1178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS154 – D4061, MARCH 1983

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Differential input voltage, $V_{ID}$	Receiver			$\pm 12$	V
Common-mode output voltage, $V_{OC}$	Driver	$-7^\dagger$		12	V
Common-mode input voltage, $V_{IC}$	Receiver			$\pm 12$	V
High-level input voltage, $V_{IH}$	DE, RE, D		2		V
Low-level input voltage, $V_{IL}$	DE, RE, D			0.8	V
High-level output current, $I_{OH}$	Driver			-60	mA
	Receiver			-400	$\mu A$
Low-level output current, $I_{OL}$	Driver			60	mA
	Receiver			8	
Operating free-air temperature, $T_A$		0		70	$^\circ C$

$^\dagger$  The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode output and threshold voltage level only.

# SN75ALS1177, SN75ALS1178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS154 – D4061, MARCH 1993

## DRIVER SECTION

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V <sub>IK</sub>	Input clamp voltage	I <sub>I</sub> = -18 mA				-1.5	V
V <sub>OH</sub>	High-level output voltage	V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V, I <sub>OH</sub> = -33 mA			3.3		V
V <sub>OL</sub>	Low-level output voltage	V <sub>IH</sub> = 2 V, V <sub>IL</sub> = 0.8 V, I <sub>OL</sub> = 33 mA			1.1		V
V <sub>OD1</sub>	Differential output voltage	I <sub>O</sub> = 0			1.5	6	V
V <sub>OD2</sub>	Differential output voltage	V <sub>CC</sub> = 5 V, R <sub>L</sub> = 100 Ω	See Figure 1		1/2 V <sub>OD1</sub>		V
		R <sub>L</sub> = 54 Ω			2		
					1.5	5	
V <sub>OD3</sub>	Differential output voltage	See Note 3			1.5	5	V
Δ V <sub>OD</sub>	Change in magnitude of differential output voltage (see Note 4)					±0.2	V
V <sub>OC</sub>	Common-mode output voltage	R <sub>L</sub> = 54 Ω or 100 Ω,	See Figure 1		-1‡	3	V
Δ V <sub>OC</sub>	Change in magnitude of common-mode output voltage (see Note 4)						±0.2
I <sub>O(OFF)</sub>	Output current with power off	V <sub>CC</sub> = 0, V <sub>O</sub> = -7 V to 12 V				±100	μA
I <sub>OZ</sub>	High-impedance-state output current	V <sub>O</sub> = -7 V to 12 V				±100	μA
I <sub>IH</sub>	High-level input current	V <sub>IH</sub> = 2.7 V				100	μA
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0.4 V				-100	μA
I <sub>OS</sub>	Short-circuit output current	V <sub>O</sub> = -7 V				-250	mA
		V <sub>O</sub> = V <sub>CC</sub>				250	
		V <sub>O</sub> = 12 V				250	
		V <sub>O</sub> = 0 V				150	
I <sub>CC</sub>	Supply current (total package)	No load	Outputs enabled		35	50	mA
			Outputs disabled		20	50	

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

‡ The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode output and threshold voltage levels only.

NOTES: 3. See EIA Standard RS-485 Figure 3.5, test termination measurement 2.

4. Δ|V<sub>OD</sub>| and Δ|V<sub>OC</sub>| are the changes in magnitude of V<sub>OD</sub> and V<sub>OC</sub>, respectively, that occur when the input is changed from a high level to a low level.

**switching characteristics at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, high-to-low-level output	R <sub>L</sub> = 60 Ω, See Figure 3	C <sub>L1</sub> = C <sub>L2</sub> = 100 pF,	9	15	22	ns
t <sub>PHL</sub>	Propagation delay time, low-to-high-level output			9	15	22	ns
t <sub>sk</sub>	Output-to-output skew			0	2	8	ns
t <sub>PZH</sub>	Output enable time to high level	C <sub>L</sub> = 100 pF,	See Figure 4	30	35	50	ns
t <sub>PZL</sub>	Output enable time to low level	C <sub>L</sub> = 100 pF,	See Figure 5	5	15	25	ns
t <sub>PHZ</sub>	Output disable time from high level	C <sub>L</sub> = 15 pF,	See Figure 4	7	15	30	ns
t <sub>PLZ</sub>	Output disable time from low level	C <sub>L</sub> = 15 pF,	See Figure 5	7	15	30	ns

  
**TEXAS  
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# SN75ALS1177, SN75ALS1178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS154 – D4061, MARCH 1993

## RECEIVER SECTION

electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
V <sub>T+</sub>	Positive-going threshold voltage	V <sub>O</sub> = 2.7 V, I <sub>O</sub> = -0.4 mA			0.2	V
V <sub>T-</sub>	Negative-going threshold voltage	V <sub>O</sub> = 0.5 V, I <sub>O</sub> = 8 mA	-0.2‡			V
V <sub>hys</sub>	Input hysteresis (V <sub>T+</sub> - V <sub>T-</sub> )			50		mV
V <sub>IK</sub>	Enable input clamp voltage	SN75ALS1177 I <sub>I</sub> = -18 mA			-1.5	V
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 200 mV, See Figure 2 I <sub>OH</sub> = -400 μA,		2.7		V
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = 200 mV, See Figure 2 I <sub>OL</sub> = 8 mA,			0.45	V
I <sub>OZ</sub>	High-impedance-state output current	SN75ALS1177 V <sub>O</sub> = 0.4 V to 2.4 V			±20	μA
I <sub>I</sub>	Line input current (see Note 5)	Other input at 0 V				
		V <sub>I</sub> = 12 V			1	mA
		V <sub>I</sub> = -7 V			-0.8	
I <sub>IH</sub>	High-level input current, RE	SN75ALS1177 V <sub>IH</sub> = 2.7 V			20	μA
I <sub>IL</sub>	Low-level input current, RE	SN75ALS1177 V <sub>IL</sub> = 0.4 V			-100	μA
r <sub>i</sub>	Input resistance			12		kΩ
I <sub>OS</sub>	Short-circuit output current	V <sub>O</sub> = 0 V, See Note 6	-15		-85	mA
I <sub>CC</sub>	Supply current (total package)	No load, Outputs enabled		35	50	mA

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

‡ The algebraic convention, where the less positive (more negative) limit is designated as minimum, is used in this data sheet for common-mode output and threshold voltage levels only.

NOTES: 5. Refer to EIA standards RS-422-A, RS-423-A, RS-485-A for exact conditions.

6. Not more than one output should be shorted at a time.

switching characteristics at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	C <sub>L</sub> = 15 pF, See Figure 6	15	25	37	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output		15	25	37	ns
t <sub>PZH</sub>	Output enable time to high level	SN75ALS1177 C <sub>L</sub> = 100 pF, See Figure 7	10	20	30	ns
t <sub>PZL</sub>	Output enable time to low level	SN75ALS1177 C <sub>L</sub> = 100 pF, See Figure 7	10	20	30	ns
t <sub>PHZ</sub>	Output disable time from high level	SN75ALS1177 C <sub>L</sub> = 15 pF, See Figure 7	5	12	16	ns
t <sub>PLZ</sub>	Output disable time from low level	SN75ALS1177 C <sub>L</sub> = 15 pF, See Figure 7	5	12	16	ns

PARAMETER MEASUREMENT INFORMATION

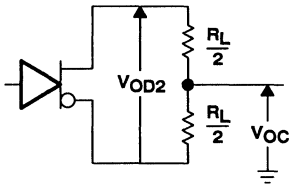


Figure 1. Driver Test Circuit,  $V_{OD}$  and  $V_{OC}$

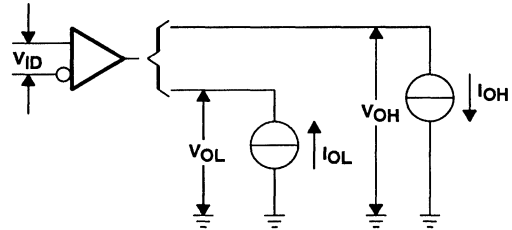
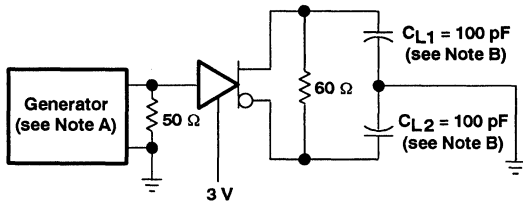
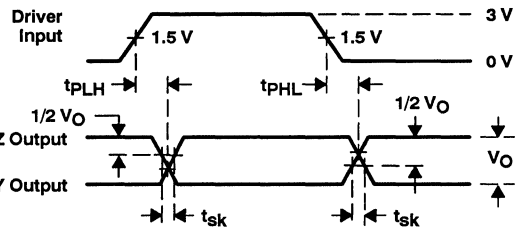


Figure 2. Receiver Test Circuit,  $V_{OH}$  and  $V_{OL}$

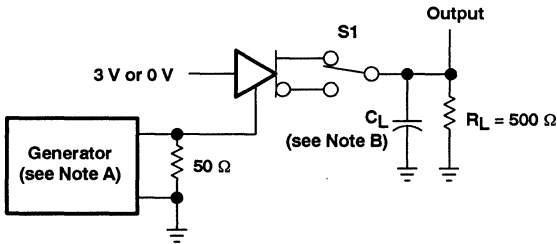


DRIVER TEST CIRCUIT

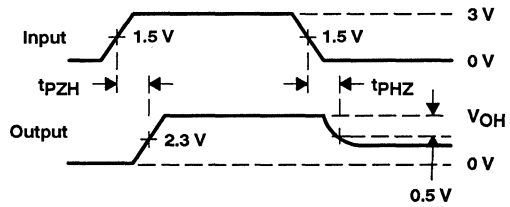


DRIVER VOLTAGE WAVEFORMS

Figure 3. Driver Propagation Delay Times



DRIVER TEST CIRCUIT



DRIVER VOLTAGE WAVEFORMS

Figure 4. Driver Enable and Disable Times

NOTES: A. The pulse generator has the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  10 ns,  $t_f \leq$  10 ns.  
 B.  $C_L$  includes probe and jig capacitance.



# SN75ALS1177, SN75ALS1178 DUAL DIFFERENTIAL DRIVERS AND RECEIVERS

SLLS154 – D4061, MARCH 1993

## PARAMETER MEASUREMENT INFORMATION

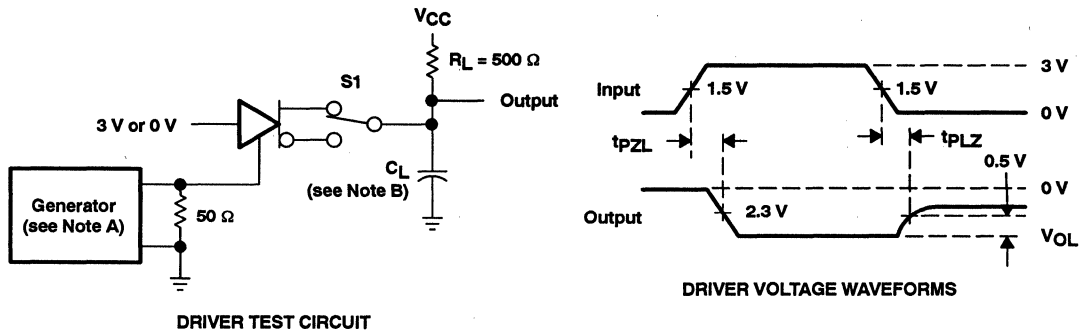


Figure 5. Driver Enable and Disable Times

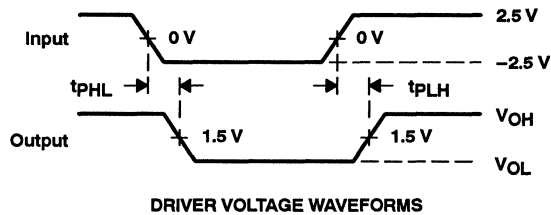
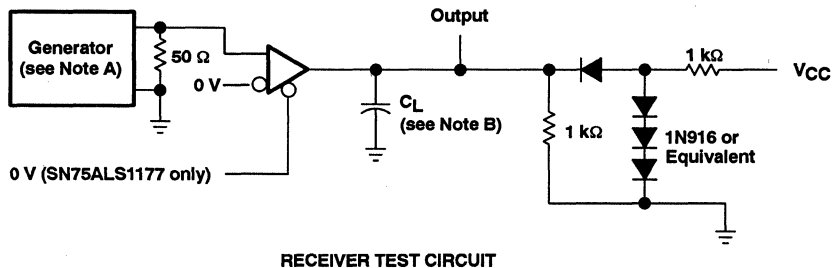
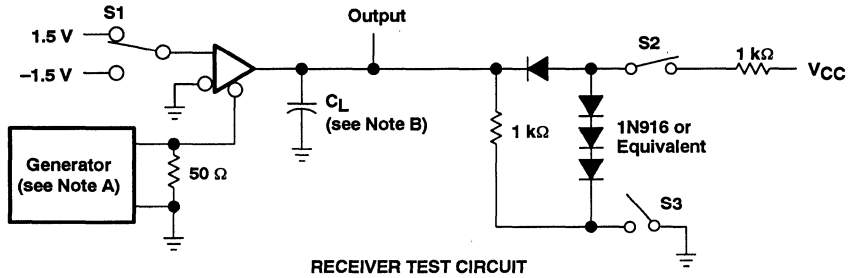


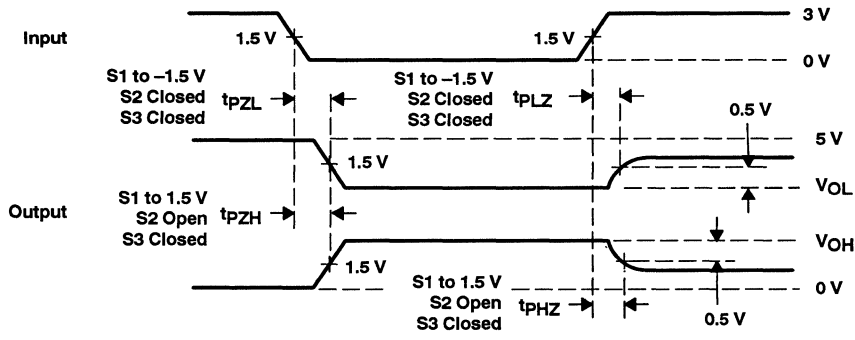
Figure 6. Receiver Propagation Delay Times

NOTES: A. The pulse generator has the following characteristics:  $PRR \leq 1$  MHz, 50% duty cycle,  $t_r \leq 10$  ns,  $t_f \leq 10$  ns.  
B.  $C_L$  includes probe and jig capacitance.

PARAMETER MEASUREMENT INFORMATION



RECEIVER TEST CIRCUIT



RECEIVER VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  10 ns,  $t_f \leq$  10 ns.  
 B.  $C_L$  includes probe and jig capacitance.

Figure 7. Receiver Output Enable and Disable Times



# SN65C1406, SN75C1406 TRIPLE LOW-POWER DRIVERS/RECEIVERS

SLLS148A – D3425, MAY 1990 – REVISED MARCH 1993

- Meets Standard EIA-232-D (Revision of RS-232-C)
- Very Low Power Consumption . . . 5 mW Typ
- Wide Driver Supply Voltage . . .  $\pm 4.5$  V to  $\pm 15$  V
- Driver Output Slew Rate Limited to 30 V/ $\mu$ s Max
- Receiver Input Hysteresis . . . 1000 mV Typ
- Push-Pull Receiver Outputs
- On-Chip Receiver 1- $\mu$ s Noise Filter
- Functionally Interchangeable With Motorola MC145406

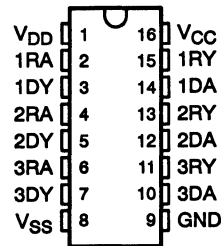
## description

The SN65C1406 and SN75C1406 are low-power BI-MOS devices containing three independent drivers and receivers that are used to interface data terminal equipment (DTE) with data circuit-terminating equipment (DCE). This device is designed to conform to Standards ANSI/EIA-232-D-1986 (which supersedes RS-232-C). The drivers and receivers of the SN65C1406 and SN75C1406 are similar to those of the SN75C188 quadruple driver and SN75C189A quadruple receiver, respectively. The drivers have a controlled output slew rate that is limited to a maximum of 30 V/ $\mu$ s and the receivers have filters that reject input noise pulses of shorter than 1  $\mu$ s. Both these features eliminate the need for external components.

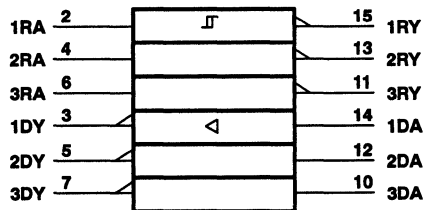
The SN65C1406 and SN75C1406 have been designed using low-power techniques in a BI-MOS technology. In most applications, the receivers contained in these devices interface to single inputs of peripheral devices such as ACES, UARTs, or microprocessors. By using sampling, such peripheral devices are usually insensitive to the transition times of the input signals. If this is not the case, or for other uses, it is recommended that the SN65C1406 and SN75C1406 receiver outputs be buffered by single Schmitt input gates or single gates of the HCMOS, ALS, or 74F logic families.

The SN65C1406 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The SN75C1406 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

D, DW, OR N PACKAGE  
(TOP VIEW)



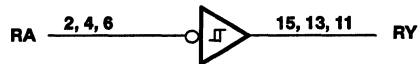
## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)

Typical of each receiver



Typical of each driver



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

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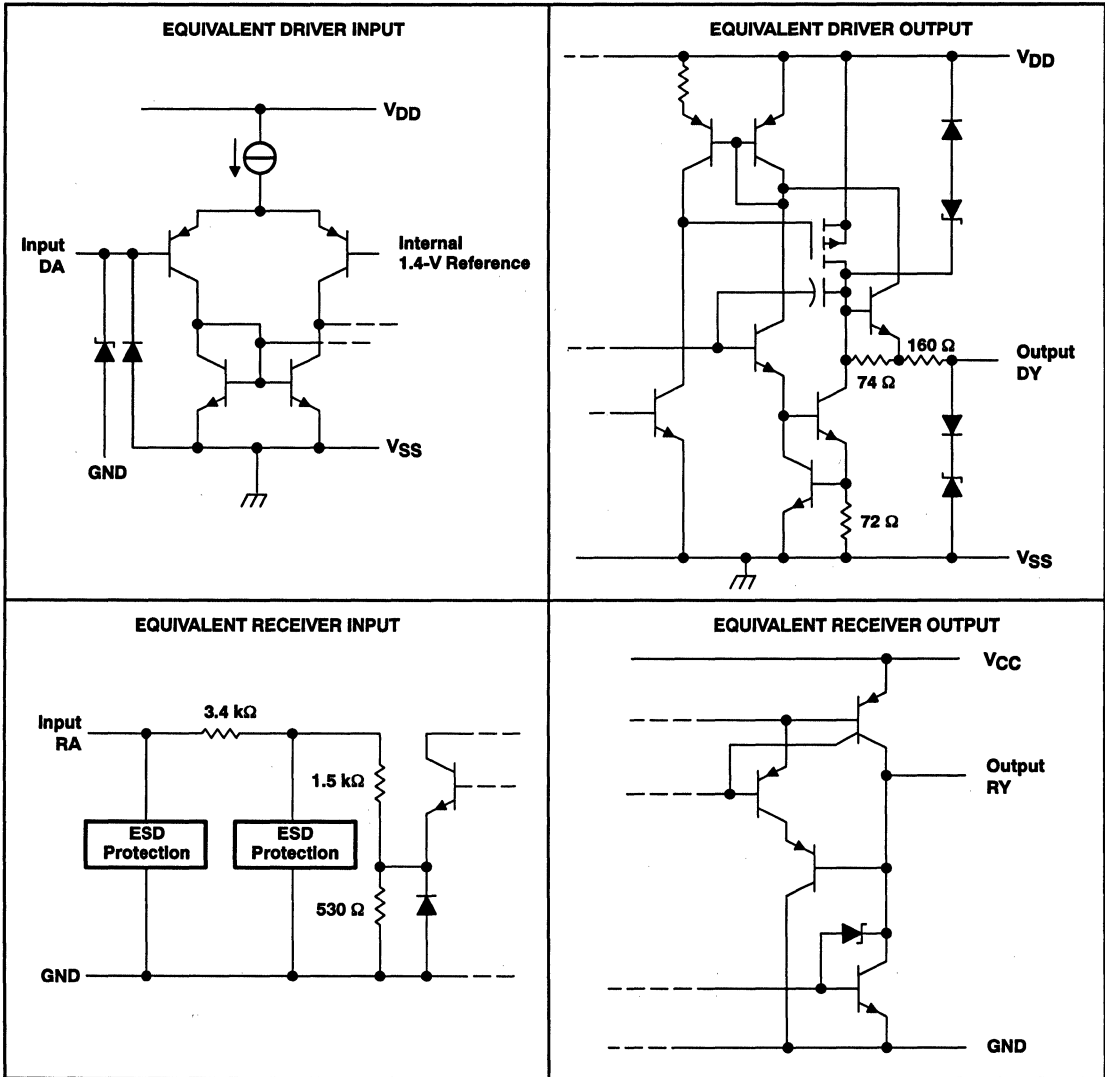
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2-935

# SN65C1406, SN75C1406 TRIPLE LOW-POWER DRIVERS/RECEIVERS

SLLS148A – D3425, MAY 1990 – REVISED MARCH 1993

## schematics of inputs and outputs



All resistor values shown are nominal.

# SN65C1406, SN75C1406 TRIPLE LOW-POWER DRIVERS/RECEIVERS

SLLS148A – D3425, MAY 1990 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{DD}$ (see Note 1)	15 V
Supply voltage, $V_{SS}$	-15 V
Supply voltage, $V_{CC}$	7 V
Input voltage range: Driver	$V_{SS}$ to $V_{DD}$
Receiver	-30 V to 30 V
Output voltage range: Driver	$(V_{SS} - 6 V)$ to $(V_{DD} + 6 V)$
Receiver	-0.3 V to $(V_{CC} + 0.3 V)$
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: SN65C1406	-40°C to 85°C
SN75C1406	0°C to 70°C
Storage temperature range	-65°C to 150 °C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
DW	1025 mW	8.2 mW/°C	656 mW	533 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{DD}$		4.5	12	15	V
Supply voltage, $V_{SS}$		-4.5	-12	-15	V
Supply voltage, $V_{CC}$		4.5	5	6	V
Input voltage, $V_I$	Driver	$V_{SS} + 2$		$V_{DD}$	V
	Receiver	$\pm 25$			
High-level input voltage, $V_{IH}$		2		0.8	V
Low-level input voltage, $V_{IL}$					
High-level output current, $I_{OH}$				-1	
Low-level output current, $I_{OL}$			3.2	mA	
Operating free-air temperature, $T_A$	SN65C1406	-40		85	°C
	SN75C1406	0		70	



# SN65C1406, SN75C1406 TRIPLE LOW-POWER DRIVERS/RECEIVERS

SLLS148A – D3425, MAY 1990 – REVISED MARCH 1993

## DRIVER SECTION

electrical characteristics over operating free-air temperature range,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage	$V_{IH} = 0.8\text{ V}$ , $R_L = 3\text{ k}\Omega$ , See Figure 1	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$	4	4.5		V
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$	10	10.8		
$V_{OL}$	Low-level output voltage (see Note 2)	$V_{IH} = 2\text{ V}$ , $R_L = 3\text{ k}\Omega$ , See Figure 1	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$		-4.4	-4	V
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$		-10.7	-10	
$I_{IH}$	High-level input current	$V_I = 5\text{ V}$ , See Figure 2				1	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0\text{ V}$ , See Figure 2				-1	$\mu\text{A}$
$I_{OS(H)}$	High-level short circuit output current‡	$V_I = 0.8\text{ V}$ , $V_O = 0$ or $V_{SS}$ , See Figure 1		-7.5	-12	-19.5	mA
$I_{OS(L)}$	Low-level short circuit output current‡	$V_I = 2\text{ V}$ , $V_O = 0$ or $V_{DD}$ , See Figure 1		7.5	12	19.5	mA
$I_{DD}$	Supply current from $V_{DD}$	No load, All inputs at 2 V or 0.8 V	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$		115	250	$\mu\text{A}$
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$		115	250	
$I_{SS}$	Supply current from $V_{SS}$	No load, All inputs at 2 V or 0.8 V	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$		-115	-250	$\mu\text{A}$
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$		-115	-250	
$r_o$	Output resistance	$V_{DD} = V_{SS} = V_{CC} = 0$ , $V_O = -2\text{ V}$ to 2 V, See Note 3		300	400		$\Omega$

† All typical values are at  $T_A = 25^\circ\text{C}$ .

‡ Not more than one output should be shorted at one time.

NOTES: 2. The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only.

3. Test conditions are those specified by EIA-232-D.

switching characteristics at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT	
$t_{PLH}$	Propagation delay time, low-to-high level output§	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Figure 3	$C_L = 15\text{ pF}$		1.2	3	$\mu\text{s}$	
$t_{PHL}$	Propagation delay time, high-to-low level output§				2.5	3.5	$\mu\text{s}$	
$t_{TLH}$	Transition time, low-to-high level output¶				0.53	2	3.2	$\mu\text{s}$
$t_{THL}$	Transition time, high-to-low level output¶				0.53	2	3.2	$\mu\text{s}$
$t_{TLH}$	Transition time, low-to-high level output#	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Figure 3	$C_L = 2500\text{ pF}$		1	2	$\mu\text{s}$	
$t_{THL}$	Transition time, high-to-low level output#				1	2	$\mu\text{s}$	
SR	Output slew rate	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , See Figure 3	$C_L = 150\text{ pF}$	4	10	30	V/ $\mu\text{s}$	

§  $t_{PHL}$  and  $t_{PLH}$  include the additional time due to on-chip slew rate and are measured at the 50% points.

¶ Measured between 10% and 90% points of output waveform.

# Measured between 3 V and -3 V points of output waveform (EIA-232-D conditions) with all unused inputs tied either high or low.



# SN65C1406, SN75C1406 TRIPLE LOW-POWER DRIVERS/RECEIVERS

SLLS148A – D3425, MAY 1990 – REVISED MARCH 1993

## RECEIVER SECTION

**electrical characteristics over operating free-air temperature range,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	See Figure 5	1.7	2	2.55	V
$V_{T-}$	Negative-going threshold voltage	See Figure 5	0.65	1	1.25	V
$V_{hys}$	Input hysteresis ( $V_{T+} - V_{T-}$ )		600	1000		mV
$V_{OH}$	High-level output voltage	$V_I = 0.75\text{ V}$ , $I_{OH} = -20\text{ }\mu\text{A}$ , See Figure 5 and Note 4	3.5			V
		$V_I = 0.75\text{ V}$ , $I_{OH} = -1\text{ mA}$ , See Figure 5	$V_{CC} = 4.5\text{ V}$	2.8	4.4	
		$V_{CC} = 5\text{ V}$	3.8	4.9		
		$V_{CC} = 5.5\text{ V}$	4.3	5.4		
$V_{OL}$	Low-level output voltage	$V_I = 3\text{ V}$ , $I_{OL} = 3.2\text{ mA}$ , See Figure 5		0.17	0.4	V
$I_{IH}$	High-level input current	$V_I = 2.5\text{ V}$	3.6	4.6	8.3	mA
		$V_I = 3\text{ V}$	0.43	0.55	1	
$I_{IL}$	Low-level input current	$V_I = -2.5\text{ V}$	-3.6	-5	-8.3	
		$V_I = -3\text{ V}$	-0.43	-0.55	-1	
$I_{OS(H)}$	Short-circuit output at high level	$V_I = 0.75\text{ V}$ , $V_O = 0$ , See Figure 4		-8	-15	mA
$I_{OS(L)}$	Short-circuit output at low level	$V_I = V_{CC}$ , $V_O = V_{CC}$ , See Figure 4		13	25	mA
$I_{CC}$	Supply current from $V_{CC}$	No load, All inputs at 0 or 5 V	$V_{DD} = 5\text{ V}$ , $V_{SS} = -5\text{ V}$	320	450	mA
			$V_{DD} = 12\text{ V}$ , $V_{SS} = -12\text{ V}$	320	450	

† All typical values are at  $T_A = 25^\circ\text{C}$ .

NOTE 4: If the inputs are left unconnected, the receiver interprets this as an input low and the receiver outputs will remain in the high state.

**switching characteristics at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = 12\text{ V}$ ,  $V_{SS} = -12\text{ V}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high level output	$C_L = 50\text{ pF}$ , $R_L = 5\text{ k}\Omega$ See Figure 6		3	4	$\mu\text{s}$
$t_{PHL}$	Propagation delay time, high-to-low level output			3	4	$\mu\text{s}$
$t_{TLH}$	Transition time, low-to-high level output‡			300	450	ns
$t_{THL}$	Transition time, high-to-low level output‡			100	300	ns
$t_{w(N)}$	Duration of longest pulse rejected as noise§		$C_L = 50\text{ pF}$ , $R_L = 5\text{ k}\Omega$	1		4

‡ Measured between 10% and 90% points of output waveform.

§ The receiver ignores any positive- or negative-going pulse that is less than the minimum value of  $t_{w(N)}$  and accepts any positive- or negative-going pulse greater than the maximum of  $t_{w(N)}$ .

**TEXAS**  
**INSTRUMENTS**

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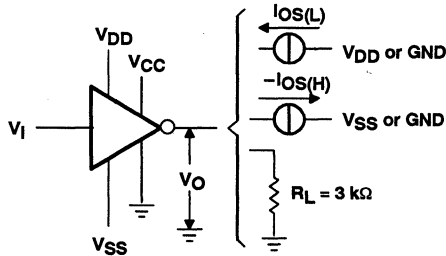
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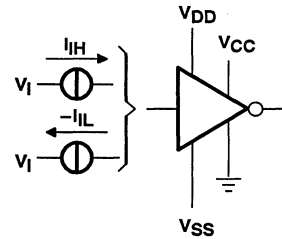
**SN65C1406, SN75C1406**  
**TRIPLE LOW-POWER DRIVERS/RECEIVERS**

SLLS148A – D3425, MAY 1990 – REVISED MARCH 1993

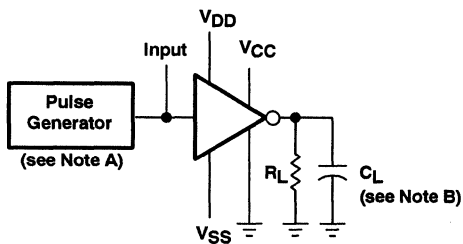
**PARAMETER MEASUREMENT INFORMATION**



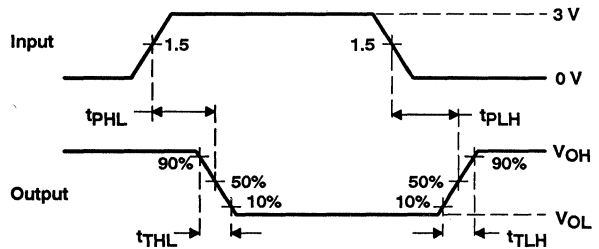
**Figure 1. Driver Test Circuit**  
 $V_{OH}$ ,  $V_{OL}$ ,  $I_{OSL}$ ,  $I_{OSH}$



**Figure 2. Driver Test Circuit,  $I_{IL}$ ,  $I_{IH}$**



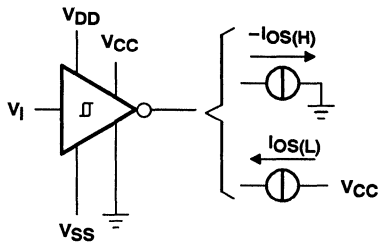
**TEST CIRCUIT**



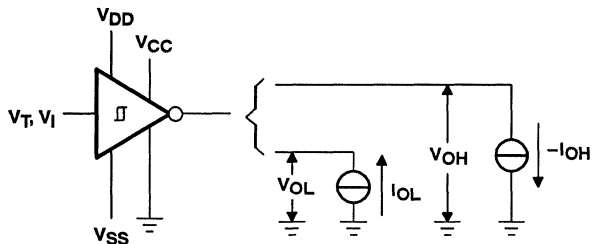
**VOLTAGE WAVEFORMS**

NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu s$ , PRR = 20 kHz,  $Z_O = 50 \Omega$ ,  $t_r = t_f < 50 ns$ .  
 B.  $C_L$  includes probe and jig capacitance.

**Figure 3. Driver Test Circuit and Voltage Waveforms**



**Figure 4. Receiver Test Circuit,  $I_{OS(H)}$ ,  $I_{OS(L)}$**

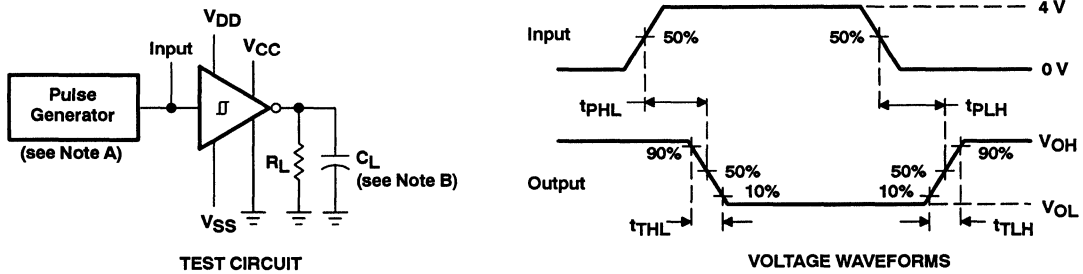


**Figure 5. Receiver Test Circuit,  $V_T$ ,  $V_{OL}$ ,  $V_{OH}$**

# SN65C1406, SN75C1406 TRIPLE LOW-POWER DRIVERS/RECEIVERS

SLLS148A – D3425, MAY 1990 – REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



NOTES: A. The pulse generator has the following characteristics:  $t_w = 25 \mu\text{s}$ ,  $\text{PRR} = 20 \text{ kHz}$ ,  $Z_0 = 50 \Omega$ ,  $t_r = t_f < 50 \text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.

**Figure 6. Receiver Test Circuit and Voltage Waveforms**

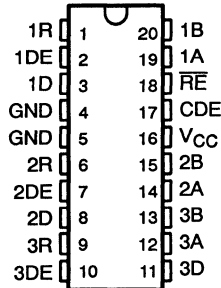


# SN75ALS1711 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS117A – D3848, APRIL 1991 – REVISED FEBRUARY 1993

- Three Bidirectional Transceivers
- Driver/Receiver Meets EIA Standard RS-485 and ANSI Standard X3.131-1986 (SCSI)
- High-Speed Advanced Low-Power Schottky Circuitry
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Wide Positive and Negative Input/Output Bus Voltages Ranges . . . -7 V to 12 V
- Driver Output Capacity . . .  $\pm 60$  mA
- Driver Positive and Negative Current Limiting
- Thermal Shutdown Protection
- Receiver Input Sensitivity . . .  $\pm 200$  mV Max
- Receiver Input Impedance . . . 12 k $\Omega$  Min
- Receiver Input Hysteresis . . . 50 mV Typ
- Operates From a Single 5-V Supply
- Low Supply-Current Requirements  
72 mA Max
- Glitch-Free Power-Up and Power-Down Protection

DW OR N PACKAGE  
(TOP VIEW)



### Function Tables

EACH DRIVER

INPUT D	ENABLES		OUTPUTS	
	DE	CDE	A	B
H	H	H	H	L
L	H	H	L	H
X	L	X	Z	Z
X	X	L	Z	Z

EACH RECEIVER

DIFFERENTIAL INPUTS A – B	ENABLE $\overline{RE}$	OUTPUT R
$V_{ID} \geq 0.2$ V	L	H
$V_{ID} = -0.2$ V to 0.2 V	L	?
$V_{ID} \leq -0.2$ V	L	L
X	H	Z
Open	L	H

H = high-level, L = low-level, ? = indeterminate,  
X = irrelevant, Z = high impedance (off)

### description

The SN75ALS1711 triple differential bus transceiver is a monolithic integrated circuit designed for bidirectional data communication on multipoint bus transmission lines. It is designed for balanced transmission lines and meets EIA Standards RS-485 and ANSI Standard X3.131-1986 (SCSI).

The SN75ALS1711 operates from a single 5-V power supply. The drivers and receivers have individual active-high and active-low enables, respectively, which can be externally connected together to function as a direction control. The driver differential output and the receiver differential input pairs are connected internally to form differential input/output (I/O) bus ports that are designed to offer minimum loading to the bus when the driver is disabled or  $V_{CC}$  is at 0. These ports feature wide positive and negative common-mode voltage ranges making the device suitable for party-line applications.

The SN75ALS1711 is characterized for operation from 0°C to 70°C.

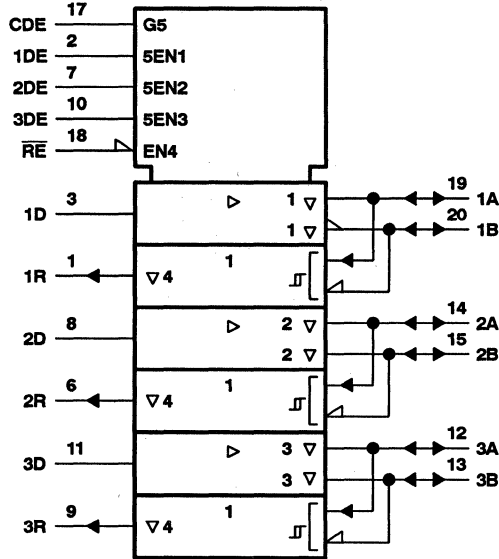
PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



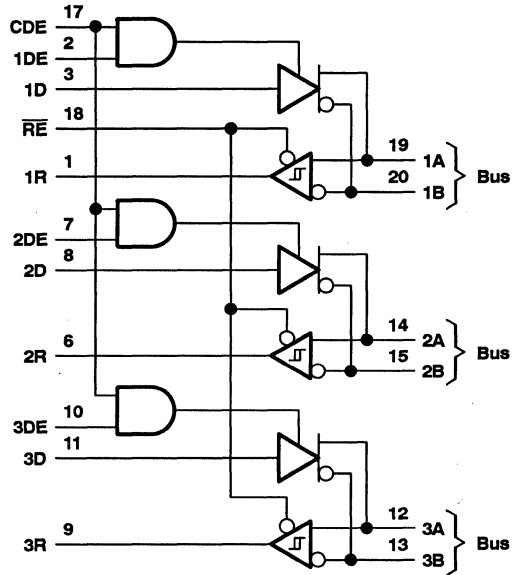
# SN75ALS1711 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS117A - D3848, APRIL 1991 - REVISED FEBRUARY 1993

## logic symbol†

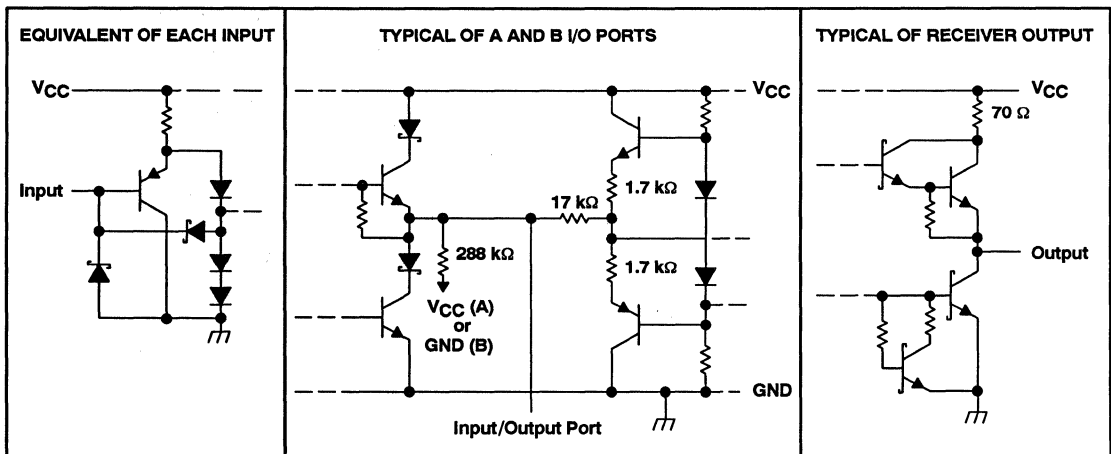


## logic diagram (positive logic)



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## schematics of inputs and outputs



All values are nominal.

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# SN75ALS1711 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS117A – D3848, APRIL 1991 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	–0.5 V to 7 V
Enable input voltage range	–0.5 V to $V_{CC} + 0.5$ V
Input voltage range, $V_I$ : Driver	–0.5 V to $V_{CC} + 0.5$ V
Receiver	–9 V to 14 V
Output voltage range, $V_O$ : Driver	–9 V to 14 V
Receiver	–0.5 V to $V_{CC} + 0.5$ V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values, except differential input/output bus voltage, are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DW	1125 mW	9.0 mW/°C	720 mW
N	1150 mW	9.2 mW/°C	736 mW

## recommended operating conditions

	MIN	TYP	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage at any bus terminal, $V_{IC}$ (see Note 2)	–7†		12	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$	0.8			V
High-level output current, $I_{OH}$	–60			mA
	–400			µA
Low-level output current, $I_{OL}$	60			mA
	8			µA
Operating free-air temperature, $T_A$	0		70	°C

† The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.



# SN75ALS1711

## TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS117A – D3848, APRIL 1991 – REVISED FEBRUARY 1993

### DRIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V <sub>IK</sub>	Input clamp voltage	I <sub>I</sub> = -18 mA				-1.5	V
V <sub>O</sub>	Output voltage	I <sub>O</sub> = 0		0		6	V
V <sub>OD1</sub>	Differential output voltage	I <sub>O</sub> = 0		1.5		5	V
V <sub>OD2</sub>	Differential output voltage	R <sub>L</sub> = 54 Ω,	See Figure 1	1.5		5	V
V <sub>OD3</sub>	Differential output voltage	See Note 3 and Figure 2		1.5		5	V
Δ V <sub>OD</sub>	Change in magnitude of differential output voltage‡	R <sub>L</sub> = 54 Ω,	See Figure 1			±0.2	V
V <sub>OC</sub>	Common-mode output voltage	R <sub>L</sub> = 54 Ω,	See Figure 1			3 -1	V
Δ V <sub>OC</sub>	Change in magnitude of common-mode output voltage‡	R <sub>L</sub> = 54 Ω,	See Figure 1			±0.2	V
I <sub>OZ</sub>	High-impedance state output current	Output disabled, V <sub>CC</sub> = 5.25 V	V <sub>O</sub> = 12 V V <sub>O</sub> = 7 V			1 -0.8	mA
I <sub>IH</sub>	High-level input current, DE, EN, CDE	V <sub>IH</sub> = 2.4 V				20	μA
I <sub>IL</sub>	Low-level input current, DE, EN, CDE	V <sub>IL</sub> = 0.4 V				-200	μA
I <sub>OS</sub>	Short-circuit output current	V <sub>O</sub> = 12 V V <sub>O</sub> = 7 V				-250 250	mA
I <sub>CC</sub>	Supply current	No load	Outputs enabled Outputs disabled			48 30 72 48	mA

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

‡ Δ|V<sub>OD</sub>| and Δ|V<sub>OC</sub>| are the changes in magnitude of V<sub>OD</sub> and V<sub>OC</sub>, respectively, that occur when the input is changed from a high level to a low level.

NOTE 3: This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions.

### switching characteristics, V<sub>CC</sub> = 5 V ± 5%, T<sub>A</sub> = 25°C

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Differential propagation delay time, low-to-high level output	R <sub>L</sub> = 54 Ω,	C <sub>L</sub> = 100 pF,	8	13	22	ns
t <sub>PHL</sub>	Differential propagation delay time, high-to-low level output	See Figure 3		8	15	22	
t <sub>PZH</sub>	Output enable time to high level	R <sub>L</sub> = 110 Ω, See Figure 4	S1 open, S2 closed	30	50	60	ns
t <sub>PHZ</sub>	Output disable time from high level			4	16	30	
t <sub>PZL</sub>	Output enable time to low level			16	26	45	
t <sub>PLZ</sub>	Output disable time from low level			4	8	20	
				S1 closed, S2 open			

# SN75ALS1711 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS117A – D3848, APRIL 1991 – REVISED FEBRUARY 1993

## RECEIVER SECTION

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{T+}$	Positive-going threshold voltage	$V_O = 2.7\text{ V}$ ,	$I_O = -0.4\text{ mA}$			0.2	V
$V_{T-}$	Negative-going threshold voltage	$V_O = 0.5\text{ V}$ ,	$I_O = 4\text{ mA}$	-0.2‡			V
$V_{hys}$	Hysteresis ( $V_{T+} - V_{T-}$ )				50		mV
$V_{IK}$	Input clamp voltage, $\overline{RE}$	$I_I = 18\text{ mA}$				-1.5	V
$V_{OH}$	High-level output voltage	$I_{OH} = -0.4\text{ mA}$		2.4			V
$V_{OL}$	Low-level output voltage	$I_{OL} = 4\text{ mA}$				0.5	V
$I_{OZ}$	High-impedance-state output current	$V_{CC} = 5.25\text{ V}$ ,	$V_O = 0.4\text{ V to } 2.4\text{ V}$			$\pm 20$	$\mu\text{A}$
$I_I$	Line input current	Other input at 0 V, See Note 3	$V_I = 12\text{ V}$			1	mA
			$V_I = 7\text{ V}$			-0.8	
$I_{IH}$	High-level input current, $\overline{RE}$	$V_{IH} = 2.4\text{ V}$				20	$\mu\text{A}$
$I_{IL}$	Low-level input current, $\overline{RE}$	$V_{IL} = 0.4\text{ V}$				-200	$\mu\text{A}$
$r_i$	Input resistance			12			k $\Omega$
$I_{OS}$	Short-circuit output current§	$V_O = 0$		-15		-130	mA
$I_{CC}$	Supply current	No load	Outputs enabled		48	72	mA
			Outputs disabled		30	48	

† All typical values are at  $V_{CC} = 5\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

§ Not more than one output should be shorted at one time.

NOTE 3: This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions.

### switching characteristics, $V_{CC} = 5\text{ V} \pm 5\%$ , $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	See Figures 5 and 6		13	20	37	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output			13	20	37	
$t_{PZH}$	Output enable time to high level	See Figures 5 and 7	$S_1$ to 1.5 V, $S_2$ open, $S_3$ closed	3	9	20	ns
$t_{PHZ}$	Output disable time from high level			8	15	22	
$t_{PZL}$	Output enable time to low level		$S_1$ to -1.5 V, $S_2$ closed, $S_3$ open	5	10	20	
$t_{PZL}$	Output enable time to low level			5	9	16	





# SN75ALS1711 TRIPLE DIFFERENTIAL BUS TRANSCEIVER

SLLS117A - D3848, APRIL 1991 - REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

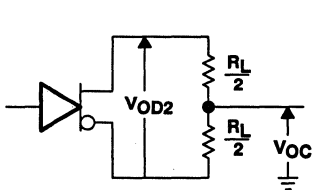


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

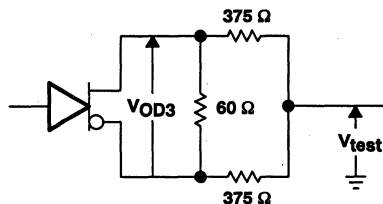


Figure 2. Driver  $V_{OD3}$

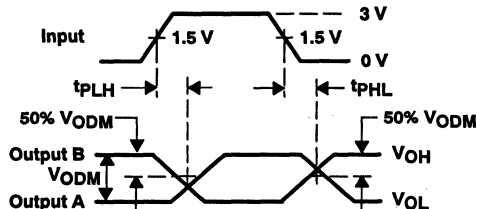
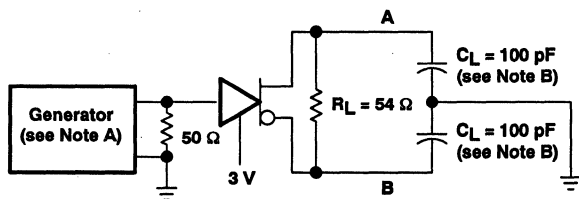


Figure 3. Driver Propagation Delay Times

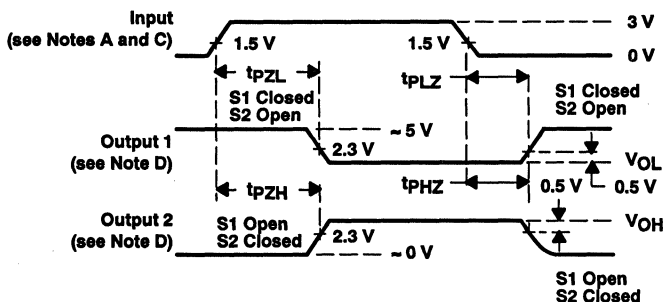
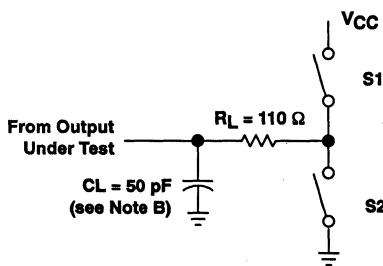


Figure 4. Driver Enable/Disable Times

- NOTES: A. All input pulses are supplied by generators having the following characteristics:  $PRR \leq 1$  MHz,  $Z_O = 50 \Omega$ ,  $t_r \leq 10$  ns,  $t_f \leq 10$  ns.  
 B.  $C_L$  includes probe and jig capacitance.  
 C. Each enable is tested separately.  
 D. Output 1 and output 2 are outputs with internal conditions such that the output is low or high except when disabled by the output control.

PARAMETER MEASUREMENT INFORMATION

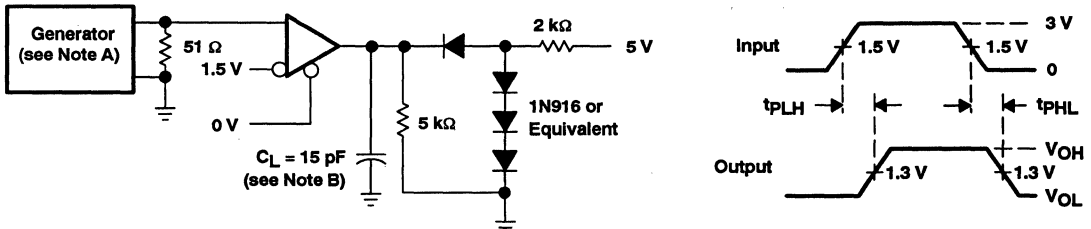


Figure 5. Receiver Propagation Delay Times

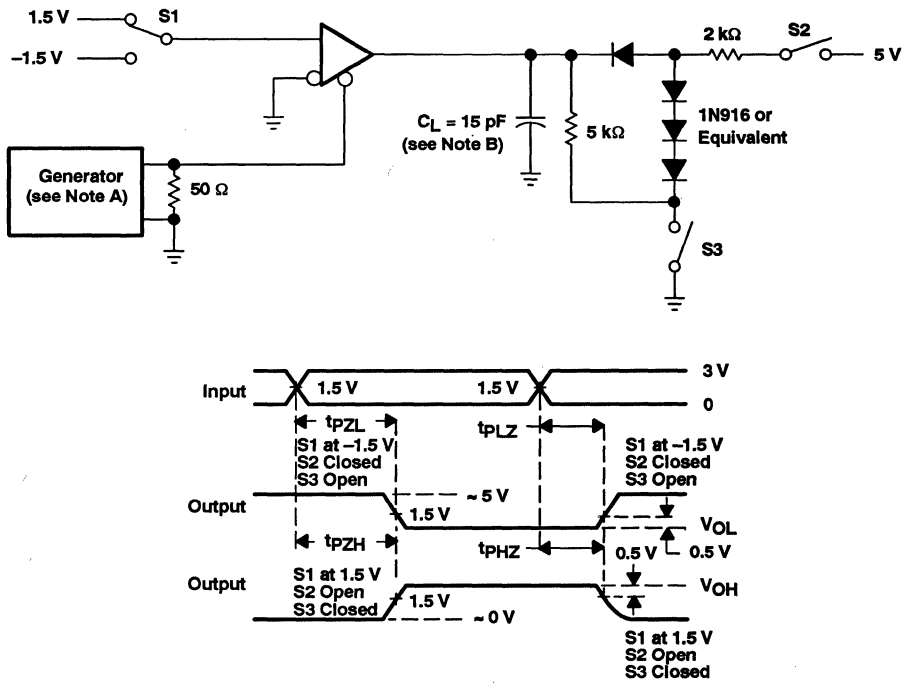


Figure 6. Receiver Enable/Disable Times

NOTES: A. All input pulses are supplied by generators having the following characteristics: PRR  $\leq$  1 MHz,  $Z_0 = 50 \Omega$ ,  $t_r \leq 10$  ns,  $t_f \leq 10$  ns.  
B.  $C_L$  includes probe and jig capacitance.

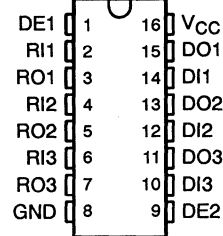


# SN751730 TRIPLE LINE DRIVER/RECEIVER

SLLS062A - D3494, MAY 1990 - REVISED JANUARY 1991

- Meets IBM 360/370 Input/Output Interface Specification for 4.5 Mb/s Operation
- Single 5-V Supply
- Uncommitted Emitter-Follower Output Structure for Party-Line Operation
- Driver Output Short-Circuit Protection
- Driver Input/Receiver Output Compatible With TTL
- Receiver Input Resistance . . . 7.4 k $\Omega$  to 20 k $\Omega$
- Ratio Specification for Propagation Delay Time, Low-to-High/High-to-Low

D OR N PACKAGE  
(TOP VIEW)



Function Tables

EACH DRIVER

INPUTS			OUTPUT
DI	DE1	DE2	DO
L	X	X	L
X	L	X	L
X	X	L	L
H	H	H	H

EACH RECEIVER

INPUT	OUTPUT
RI	RO
L	H
H	L
Open	H

H = high level, L = low level, X = irrelevant

## description

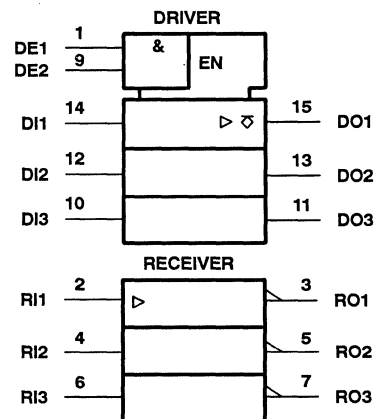
The SN751730 triple line driver/receiver is specifically designed to meet the input/output interface specifications for IBM System 360/370. It is also compatible with standard TTL logic and supply voltage levels.

The low-impedance emitter-follower driver outputs of the SN751730 will drive terminated lines such as coaxial cable or twisted pair. Having the outputs uncommitted allows wired-OR logic to be performed in party-line applications. Output short-circuit protection is provided by an internal clamping network that turns on when the output voltage drops below approximately 2.5 V.

An open line will affect the receiver input as would a low-level input voltage.

All the driver inputs and receiver outputs are in conventional TTL configuration and the gating can be used during power-up and power-down sequences to ensure that no noise is introduced to the line, by pulling either DE1 or DE2 to a low level.

## logic symbols†



† These symbols are in accordance with ANSI/IEE Std 91-1984 and IEC Publication 617-12.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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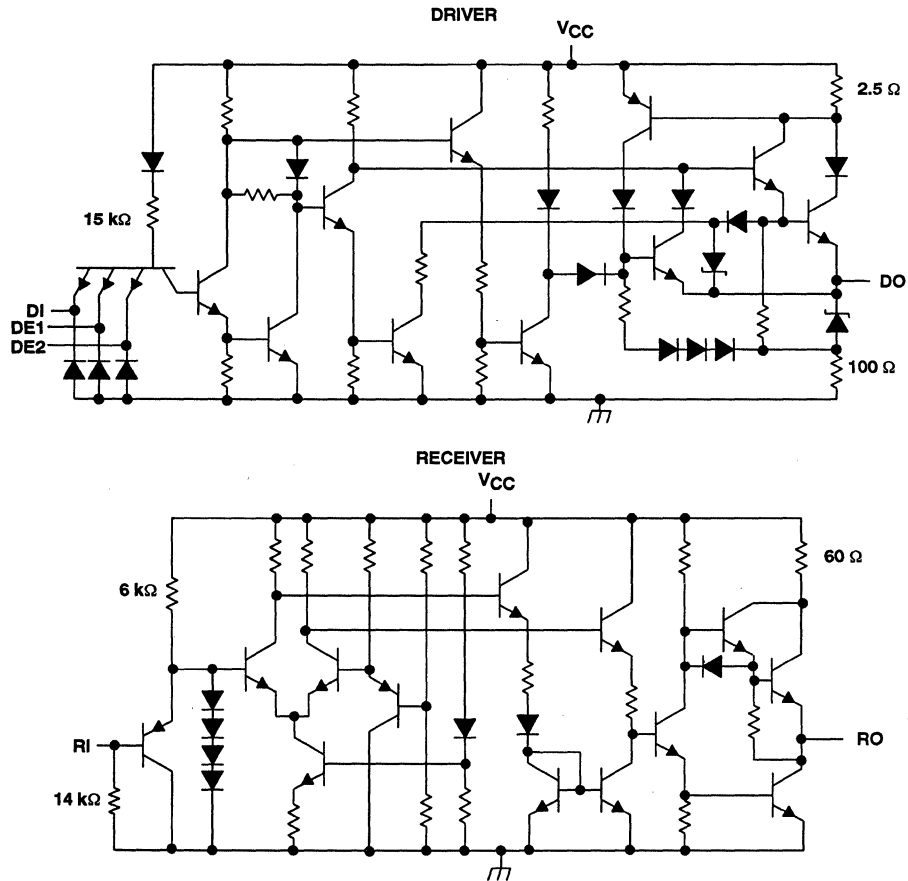
# SN751730 TRIPLE LINE DRIVER/RECEIVER

SLLS062A - D3494, MAY 1990 - REVISED JANUARY 1991

## logic diagrams (positive logic)



## equivalent schematics of driver and receiver†



† All resistor values are nominal.

**SN751730**  
**TRIPLE LINE DRIVER/RECEIVER**

SLLS062A – D3494, MAY 1990 – REVISED JANUARY 1991

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Input voltage range, $V_I$ : Driver	-0.5 V to 7 V
Receiver	-0.5 V to 7 V
Output voltage range, $V_O$ : Driver	-0.5 V to 7 V
Enable input voltage range	-0.5 V to 7 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW
N	1150 mW	9.2 mW/°C	736 mW

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	Driver, Enable			V
	Receiver			
Low-level input voltage, $V_{IL}$	Driver, Enable			V
	Receiver			
Operating free-air temperature, $T_A$	0		70	°C



**SN751730**  
**TRIPLE LINE DRIVER/RECEIVER**

SLLS062A - D3494, MAY 1990 - REVISED JANUARY 1991

**DRIVER SECTION**

**electrical characteristics over recommended ranges of supply voltage and operating free air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT	
V <sub>IK</sub>	Input clamp voltage	V <sub>CC</sub> = 4.75 V, I <sub>IL</sub> = -18 mA			-1.5	V	
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = 4.75 V, I <sub>OH</sub> = -59.3 mA, V <sub>IH</sub> = 2 V, T <sub>A</sub> = 25°C		3.11		V	
		V <sub>CC</sub> = 5.25 V, I <sub>OH</sub> = -78.1 mA, V <sub>IH</sub> = 2 V			4.10		
		V <sub>CC</sub> = 4.75 V, R <sub>L</sub> = 51.4 Ω, V <sub>IH</sub> = 2 V		3.05			
		V <sub>CC</sub> = 5.25 V, R <sub>L</sub> = 56.9 Ω, V <sub>IH</sub> = 2 V			4.20		
V <sub>ODH</sub>	Differential high-level output voltage	R <sub>L</sub> = 46.3 Ω or 56.9 Ω			0.50	V	
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = 5.25 V, V <sub>IL</sub> = 0.8 V, V <sub>IH</sub> = 4.5 V		I <sub>OL</sub> = -0.24 mA	0.15	V	
				R <sub>L</sub> = 56.9 Ω	0.15		
I <sub>IH</sub>	High-level input current	DI DE	V <sub>CC</sub> = 5.25 V, V <sub>IH</sub> = 2.7 V			20	μA
					60		
I <sub>IL</sub>	Low-level input current	DI DE	V <sub>CC</sub> = 5.25 V, V <sub>IH</sub> = 0.4 V			-400	μA
					-1200		
I <sub>OH</sub>	High-level output current	V <sub>CC</sub> = 4.75 V, V <sub>OH</sub> = 5 V		V <sub>IL</sub> = 0	100	μA	
				V <sub>IH</sub> = 4.5 V	100		
I <sub>OS</sub>	Short-circuit output current†	V <sub>CC</sub> = 5.25 V		V <sub>IH</sub> = 4.5 V	-30	mA	
I <sub>CCH</sub>	Supply current (total package)	V <sub>CC</sub> = 5.25 V, No load		V <sub>I(D)</sub> = 4.5 V, V <sub>I(R)</sub> = 0	47	mA	
I <sub>CCL</sub>				V <sub>I(D)</sub> = 0, V <sub>I(R)</sub> = 4.5 V	80		

† No more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

**switching characteristics, V<sub>CC</sub> = 5 V ± 5%, T<sub>A</sub> = 25°C**

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	R <sub>L</sub> = 47.5 Ω, See Figure 1	See Figure 1	6.5	12	18.5	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low level output			6.5	12	18.5	ns
Δt <sub>pd</sub>	Differential propagation delay time‡				10	ns	
t <sub>r</sub>	Output rise time	V <sub>CC</sub> = 5 V, R <sub>L</sub> = 47.5 Ω, V <sub>O</sub> = 0.15 V to 3.05 V, C <sub>L</sub> = 10.2 pF, See Figure 1		5	10		ns
t <sub>f</sub>	Output fall time			5	13		ns
SR	Slew rate	V <sub>O</sub> = 1 V to 3 V average, R <sub>L</sub> = 47.5 Ω, C <sub>L</sub> = 10.2 pF, See Figure 1				0.65	V/ns

‡ Δt<sub>pd</sub> = |t<sub>PLH</sub> - t<sub>PHL</sub>|



**RECEIVER SECTION**

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = 4.75 V, I <sub>OH</sub> = -400 μA	V <sub>I</sub> = 1.15 V,	2.7		V
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = 4.75 V, V <sub>IH</sub> = 1.55 V	I <sub>OL</sub> = 8 mA	0.5		V
			I <sub>OL</sub> = 4 mA	0.4		
r <sub>i</sub>	Input resistance	V <sub>CC</sub> = 0,	V <sub>I</sub> = 0.15 V to 3.9 V	7.4	20	kΩ
I <sub>IH</sub>	High-level input current	V <sub>CC</sub> = 4.75 V,	V <sub>IH</sub> = 3.11 V		0.42	mA
I <sub>IL</sub>	Low-level input current	V <sub>CC</sub> = 5.25 V,	V <sub>IL</sub> = 0.15 V	-0.24	0.04	mA
I <sub>OS</sub> <sup>†</sup>	Short-circuit output current	V <sub>CC</sub> = 5.25 V,	V <sub>IL</sub> = 0	-20	-100	mA
I <sub>CCH</sub>	Supply current (total package)	V <sub>CC</sub> = 5.25 V, No load	V <sub>I(D)</sub> = 4.5 V, V <sub>I(R)</sub> = 0		47	mA
I <sub>CCL</sub>			V <sub>I(D)</sub> = 0, V <sub>I(R)</sub> = 4.5 V		80	

<sup>†</sup> Only one output should be shorted at a time, and duration of the short circuit should not exceed one second.

**switching characteristics, V<sub>CC</sub> = 5 V ± 5%, T<sub>A</sub> = 25°C**

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high level output	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 15 pF, See Figure 2			7.5	12	19.5	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low level output				7.5	12	19.5	ns
Δt <sub>pd</sub> <sup>‡</sup>	Differential propagation delay time						10	ns

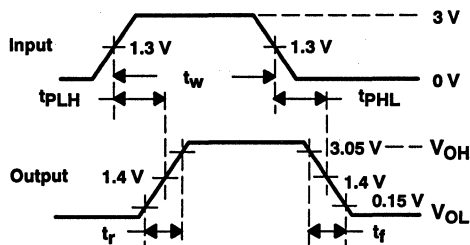
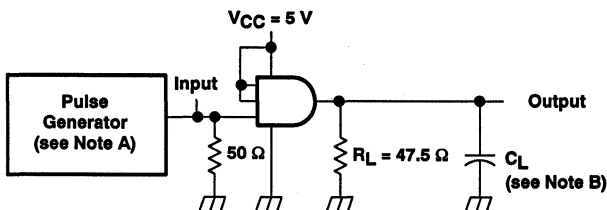
<sup>‡</sup> Δt<sub>pd</sub> = |t<sub>PLH</sub> - t<sub>PHL</sub>|



# SN751730 TRIPLE LINE DRIVER/RECEIVER

SLLS062A – D3494, MAY 1990 – REVISED JANUARY 1991

## PARAMETER MEASUREMENT INFORMATION

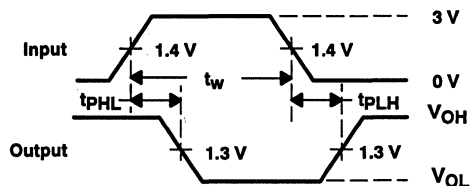
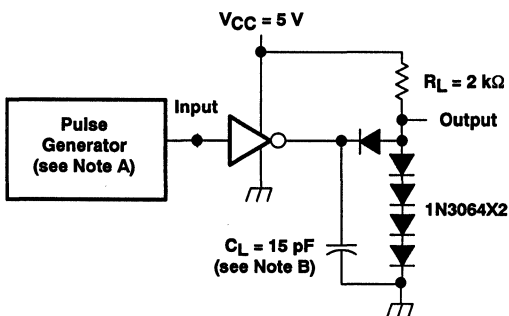


### TEST CIRCUIT

### VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O \sim 50 \Omega$ ,  $t_w \leq 500$  ns,  $PRR \leq 1$  MHz,  $t_f \leq 6$  ns,  $t_r \leq 15$  ns.  
B.  $C_L$  includes probe and jig capacitance.

Figure 1. Driver Test Circuit and Voltage Waveforms



### TEST CIRCUIT

### VOLTAGE WAVEFORMS

NOTES: A. The pulse generator has the following characteristics:  $Z_O \sim 50 \Omega$ ,  $t_w \leq 500$  ns,  $PRR \leq 1$  MHz,  $t_f \leq 10$  ns,  $t_r \leq 10$  ns.  
B.  $C_L$  includes probe and jig capacitance.

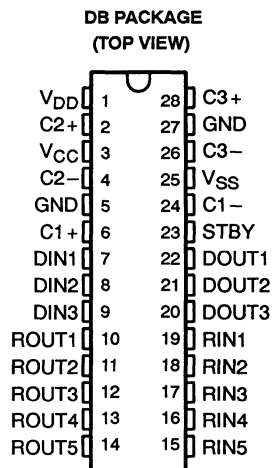
Figure 2. Receiver Test Circuit and Voltage Waveforms

# SN75LV4735

## 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

SLLS135B – D4072, FEBRUARY 1992 – REVISED APRIL 1993

- **Meets ANSI/EIA-232-D, 1986 Specifications (Revision of EIA Standard RS-232-C)**
- **Operates With Single 3.3-V Power Supply**
- **LinBiCMOS™ Process Technology**
- **Three Drivers and Five Receivers**
- **±30-V Input Levels (Receiver)**
- **ESD Protection on RS-232 Lines Exceeds 6 kV Per MIL-STD-883C, Method 3015**
- **Applications**
  - EIA-232 Interface**
  - Battery-Powered Systems**
  - Notebook PC Computers**
  - Terminals**
  - Modems**
- **Voltage Converter Operates With Low Capacitance . . . 0.47 μF Min**



### description

The SN75LV4735† is a low-power 3.3-V multichannel RS232 line driver/receiver. It includes three independent RS232 drivers and five independent RS232 receivers. It is designed to operate off of a single 3.3-V supply and has an internal switched-capacitor voltage converter to generate the RS232 output levels. The SN75LV4735 provides a single integrated circuit and single 3.3-V supply interface between the asynchronous communications element (ACE or UART) and the serial port connector of the data terminal equipment (DTE). This device has been designed to conform to standard ANSI/EIA-232-D-1986.

The switched-capacitor voltage converter of the SN75LV4735 uses four small external capacitors to generate the positive and negative voltages required by EIA-232 line drivers from a single 3.3-V input. The drivers feature output slew-rate limiting to eliminate the need for external filter capacitors. The receivers can accept ±30 V without damage.

The device also features a reduced power or standby mode that cuts the quiescent power to the integrated circuits when not transmitting data between the CPU and peripheral equipment. The STBY input is driven high for standby (reduced power) mode and driven low for normal operating mode. When in the standby mode, all driver outputs (DIN1–3) and receiver outputs (ROUT1–3) are in a high-impedance state. If the standby feature is not used in an application, STBY should be tied to GND.

The SN75LV4735 has been designed using LinBiCMOS™ technology and cells contained in the Texas Instruments LinASIC™ library. The SN75LV4735 is characterized for operation from 0°C to 70°C.

† Patent pending design

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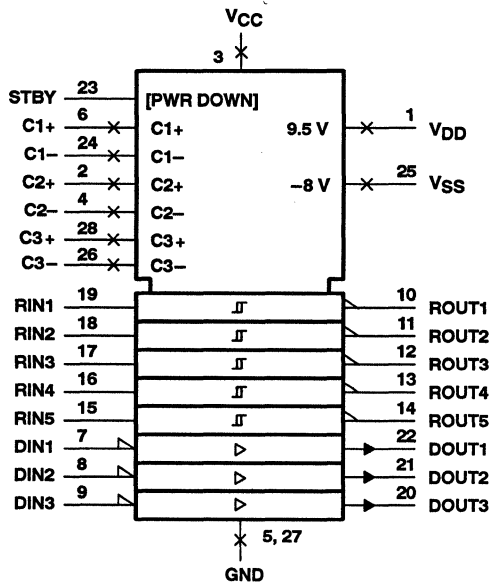
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# SN75LV4735 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

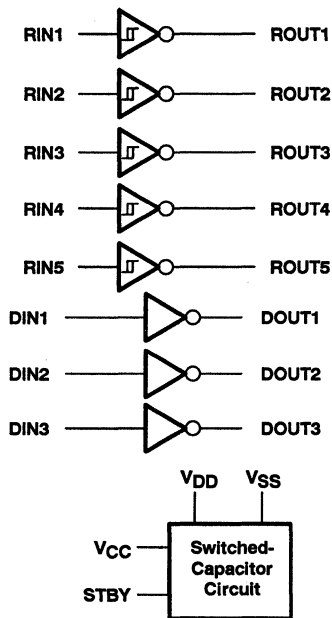
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## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std. 91-1984 and IEC Publication 617-12.

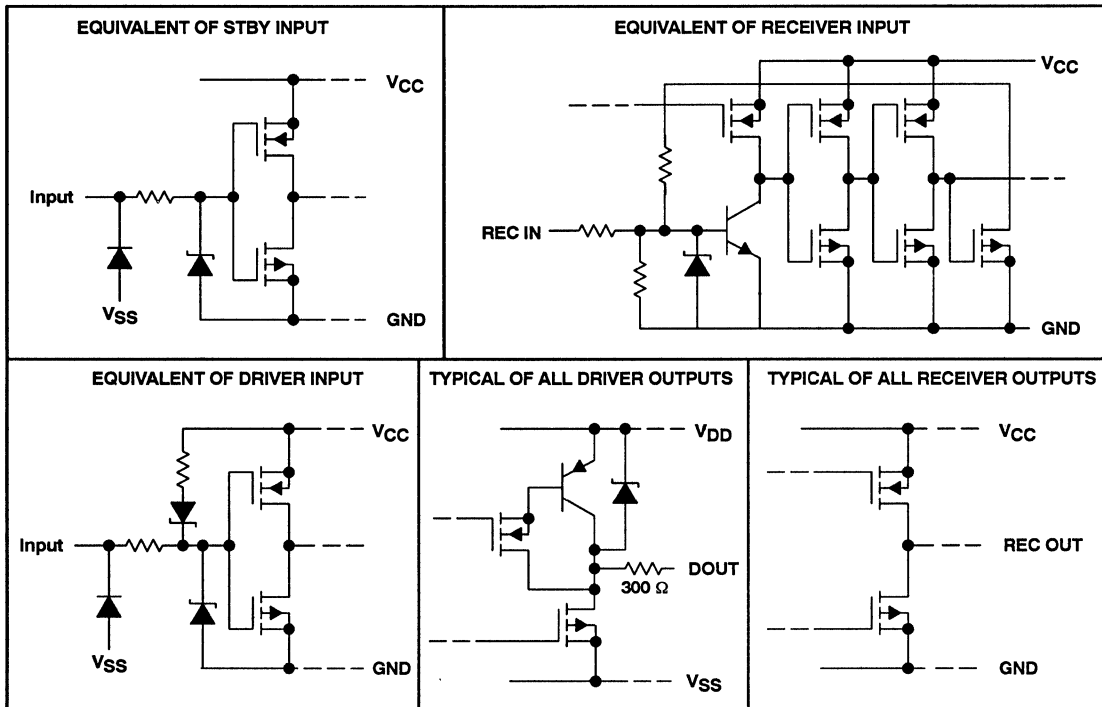
## logic diagram (positive logic)



# SN75LV4735 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

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## schematics of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC</sub>	4 V
Positive output supply voltage, V <sub>DD</sub> (see Note 1)	15 V
Negative output supply voltage, V <sub>SS</sub>	-15 V
Input voltage range, V <sub>I</sub> : DIN1–DIN3, STBY	-0.3 to 7 V
RIN1–RIN5	-30 V to 30 V
Output voltage range, V <sub>O</sub> : DOUT1–DOUT3	V <sub>SS</sub> - 0.3 V to V <sub>DD</sub> + 0.3 V
ROUT1–ROUT5	-0.3 V to 7 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T <sub>A</sub>	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	T <sub>A</sub> ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
DB	665 mW	5.3 mW/°C	430 mW

**TEXAS**  
**INSTRUMENTS**

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# SN75LV4735

## 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

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### recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		3	3.3	3.6	V
Positive output supply voltage, $V_{DD}$		8	10		V
Negative output supply voltage, $V_{SS}$		-7	-8		V
Input voltage, $V_I$ (see Note 2)	DIN1-3, STBY	0	$V_{CC}$		V
	RIN1-5		$\pm 30$		
High-level input voltage, $V_{IH}$	DIN1-3, STBY	2			V
Low-level input voltage, $V_{IL}$			0.8		
High-level output current, $I_{OH}$	RIN1-5			-1	mA
Low-level output current, $I_{OL}$				3.2	
External capacitor		0.47	1		$\mu F$
Operating free-air temperature, $T_A$		0		70	$^{\circ}C$

NOTE 2: The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only, e.g., if -10 V is a maximum, the typical value is a more negative voltage.

### supply currents

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{CC}$ Supply current from $V_{CC}$ (normal operating mode)	$V_{CC} = 3.3 V$ , No load, STBY = low-level input, All other inputs open		8.5	20	mA
$I_{CC}$ Supply current (standby mode)	$V_{CC} = 3.3 V$ , No load, STBY = high-level input			10	$\mu A$



SN75LV4735  
3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

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**DRIVER SECTION**

**electrical characteristics over operating free-air temperature range,  $V_{CC} = 3.3\text{ V} \pm 10\%$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}$	High-level output voltage	$R_L = 3\text{ k}\Omega$	5.5	7		V
$V_{OL}$	Low-level output voltage (see Note 2)	$R_L = 3\text{ k}\Omega$	-5	-5.5		V
$I_I$	Input current, STBY				$\pm 1$	$\mu\text{A}$
$I_{IH}$	High-level input current				1	$\mu\text{A}$
$I_{IL}$	Low-level input current				-1	$\mu\text{A}$
$I_{OS(H)}$	High-level short-circuit output current (see Note 3)	$V_{CC} = 3.6\text{ V}, V_O = 0$		-10	-20	$\text{mA}$
$I_{OS(L)}$	Low-level short-circuit output current (see Note 3)	$V_{CC} = 3.6\text{ V}, V_O = 0$		10	20	$\text{mA}$
$r_o$	Output resistance	$V_{CC} = V_{DD} = V_{SS} = 0,$ $V_O = -2\text{ V to } 2\text{ V},$ See Note 4	300			$\Omega$

† All typical values are at  $T_A = 25^\circ\text{C}$ .

NOTES: 2. The algebraic convention, where the more positive (less negative) limit is designated as maximum, is used in this data sheet for logic levels only.

3. Not more than one output should be shorted at one time.

4. Test conditions are those specified by EIA-232-D.

**switching characteristics,  $V_{CC} = 3.3\text{ V} \pm 10\%$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay time, low-to-high-level output	$R_L = 3\text{ k}\Omega$ to GND, $C_L = 50\text{ pF}$ , See Figure 2	200	400	600	ns
$t_{PHL}$	Propagation delay time, high-to-low-level output		100	200	350	ns
$t_{PZL}$	Output enable time to low level (see Note 5)			3	7	ms
$t_{PZH}$	Output enable time to high level (see Note 5)	$R_L = 3\text{ k}\Omega$ to GND, $C_L = 50\text{ pF}$ , See Figure 3		1	5	ms
$t_{PHZ}$	Output disable time from high level (see Note 5)			1	3	$\mu\text{s}$
$t_{PLZ}$	Output disable time from low level (see Note 5)			0.5	3	$\mu\text{s}$
SR	Output slew rate (see Note 6)	$R_L = 3\text{ k}\Omega$ to $7\text{ k}\Omega$ , $C_L = 50\text{ pF}$ , See Figure 2	3		30	$\text{V}/\mu\text{s}$
SR(tr)	Transition region slew rate	$R_L = 3\text{ k}\Omega$ to GND, $C_L = 2500\text{ pF}$ , See Figure 4		3		$\text{V}/\mu\text{s}$

NOTES: 5. Output enable occurs when STBY is driven low. Output disable occurs when STBY is driven high.

6. Measured between 3-V and -3-V points of output waveform (EIA-232-D conditions), all unused inputs tied either high or low.



# SN75LV4735

## 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

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### RECEIVER SECTION

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going threshold voltage			2.2	2.6	V
$V_{T-}$ Negative-going threshold voltage		0.6	1		V
$V_{hys}$ Input hysteresis ( $V_{T+} - V_{T-}$ )		0.5	1.2	1.8	mV
$V_{OH}$ High-level output voltage	$I_{OH} = -2$ mA, See Note 7	2.4	2.6		V
$V_{OL}$ Low-level output voltage	$I_{OL} = 2$ mA		0.2	0.4	V
$I_{IH}$ High-level input current	$V_I = 3$ V	0.43	0.55	1	mA
	$V_I = 25$ V	3.6	4.6	8.3	
$I_{IL}$ Low-level input current	$V_I = -3$ V	-0.43	-0.55	-1	mA
	$V_I = -25$ V	-3.6	-5	-8.3	
$r_i$ Input resistance	$V_I = \pm 3$ V to $\pm 25$ V	3	5	7	k $\Omega$

† All typical values are at  $T_A = 25^\circ\text{C}$ .

NOTE 7: If the inputs are left unconnected, the receiver interprets this as an input low, and the receiver outputs will remain in the high state.

### switching characteristics over recommended supply voltage range, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$R_L = 3$ k $\Omega$ to GND, $C_L = 50$ pF, See Figure 5	45	80	130	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output		70	100	170	ns
$t_{PZL}$ Output enable time to low level (see Note 5)	$R_L = 3$ k $\Omega$ to GND, $C_L = 50$ pF, See Figure 6		160	250	ns
$t_{PZH}$ Output enable time to high level (see Note 5)			4	10	$\mu\text{s}$
$t_{PHZ}$ Output disable time from high level (see Note 5)			300	500	ns
$t_{PLZ}$ Output disable time from low level (see Note 5)			140	200	ns

NOTE 5: Output enable occurs when STBY is driven low. Output disable occurs when STBY input is driven high.



# SN75LV4735 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

SLLS135B – D4072, FEBRUARY 1992 – REVISED APRIL 1993

## APPLICATION INFORMATION

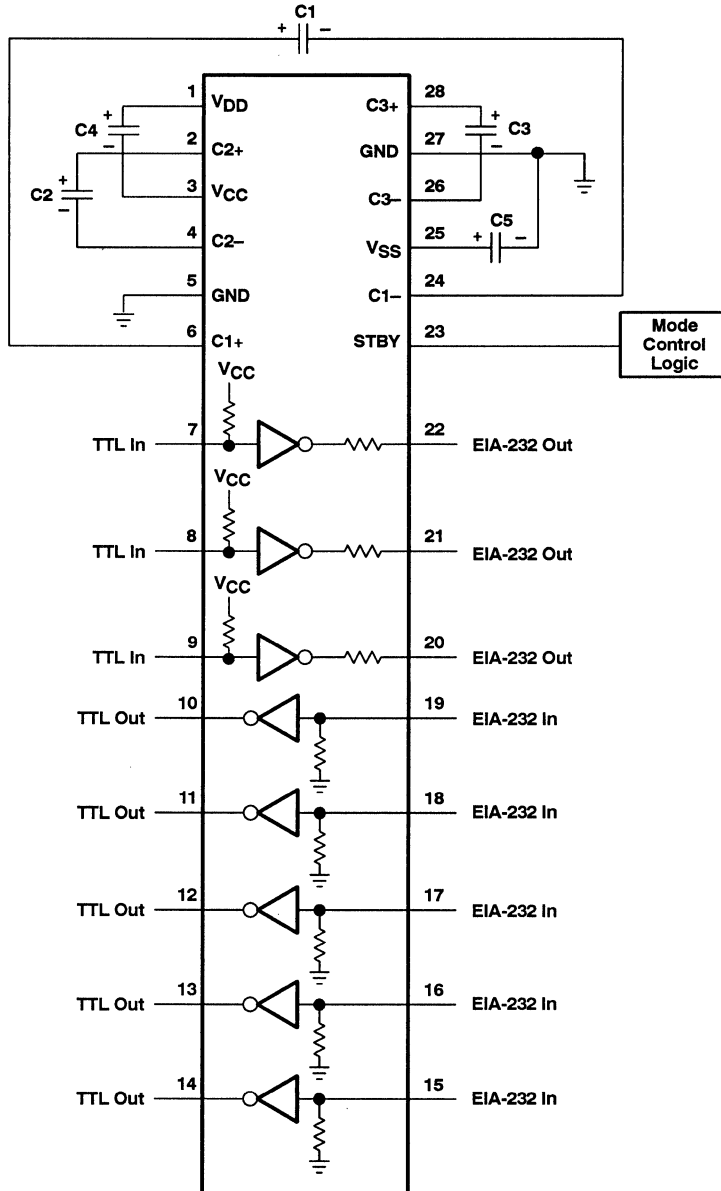


Figure 1. Typical Operating Circuit



# SN75LV4735 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

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## PARAMETER MEASUREMENT INFORMATION

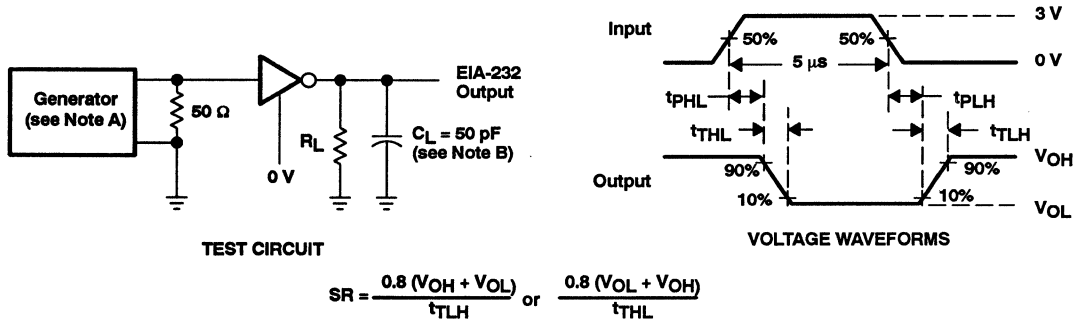


Figure 2. Driver Test Circuit and Voltage Waveforms, Slew Rate at 5-μs Input

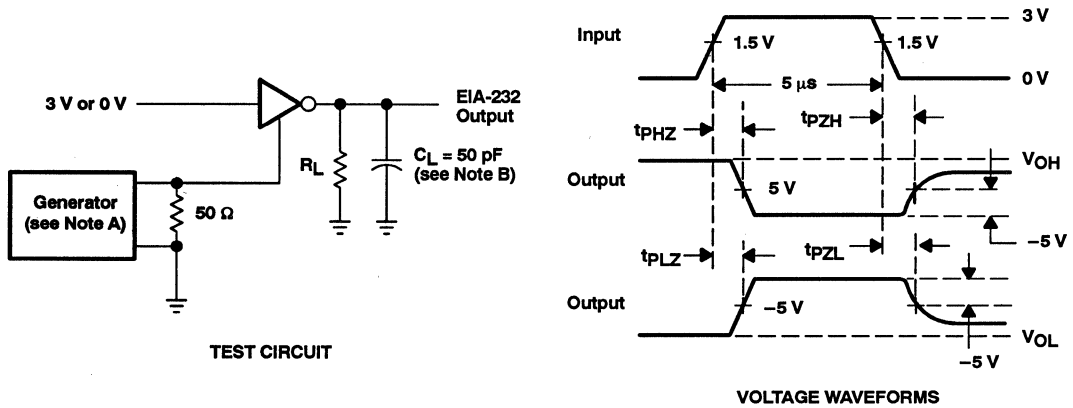


Figure 3. Driver Test Circuit and Voltage Waveforms

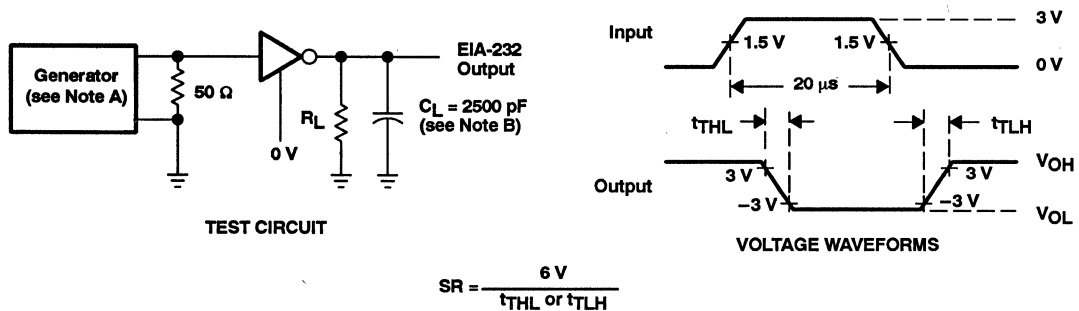


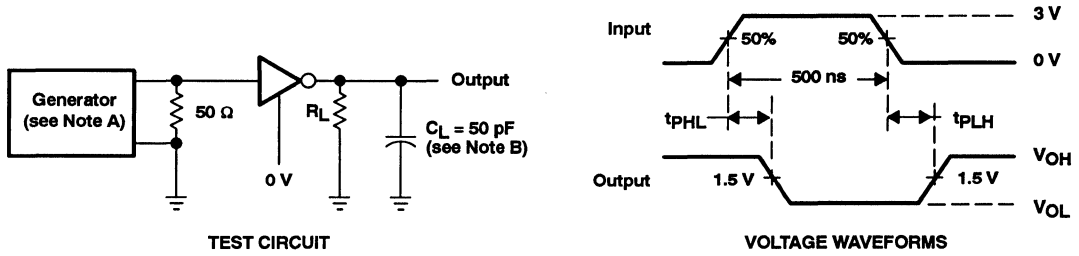
Figure 4. Driver Test Circuit and Voltage Waveforms, Slew Rate at 20-μs Input

- NOTES: A. The pulse generator has the following characteristics: 50% duty cycle,  $t_r \leq 10$  ns,  $t_f \leq 10$  ns.  
B.  $C_L$  includes probe and jig capacitance.

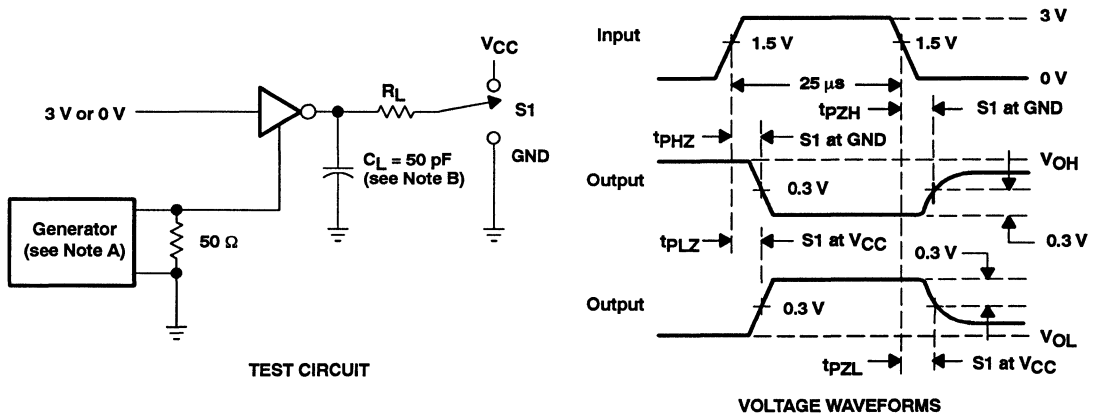
# SN75LV4735 3.3-V MULTICHANNEL RS232 LINE DRIVER/RECEIVER

SLLS135B – D4072, FEBRUARY 1992 – REVISED APRIL 1993

## PARAMETER MEASUREMENT INFORMATION



**Figure 5. Receiver Test Circuit and Voltage Waveforms**



**Figure 6. Receiver Test Circuit and Voltage Waveforms Enable and Disable Times**

NOTES: A. The pulse generator has the following characteristics: PRR = 1 MHz, 50% duty cycle,  $t_r \leq 10$  ns,  $t_f = 10$  ns.  
B.  $C_L$  includes probe and jig capacitance.



# TL3695 DIFFERENTIAL BUS TRANSCEIVER

SLLS044B – D3408, NOVEMBER 1988 – REVISED FEBRUARY 1993

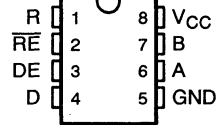
- Bidirectional Transceiver
- Meets EIA Standards RS-422-A and RS-485 and CCITT Recommendations V.11 and X.27
- High-Speed Advanced Low-Power Schottky Circuitry
- Designed for 25-MBaud Operation in Both Serial and Parallel Applications
- Low Skew . . . 6 ns Max
- Designed for Multipoint Transmission on Long Bus Lines in Noisy Environments
- Low Supply Current Requirements  
30 mA Max
- Wide Positive and Negative Input/Output Bus Voltage Ranges
- Driver Output Capacity . . .  $\pm 60$  mA
- Thermal Shutdown Protection
- Driver Positive and Negative Current Limiting
- Receiver Input Impedances . . . 12 k $\Omega$  Min
- Receiver Input Sensitivity . . .  $\pm 200$  mV Max
- Receiver Input Hysteresis . . . 120 mV Typ
- Fail Safe . . . High Receiver Output With Inputs Open
- Operates From a Single 5-V Supply
- Glitch-Free Power-Up and Power-Down Protection
- Interchangeable With National DS3695

## description

The TL3695 differential bus transceiver is a monolithic integrated circuit designed for bidirectional data communication on multipoint bus transmission lines. It is designed for balanced transmission lines and meets EIA Standards RS-422-A and RS-485 and CCITT recommendations V.11 and X.27.

The TL3695 combines a 3-state differential line driver and a differential input line receiver both of which operate from a single 5-V power supply. The driver and receiver have active-high and active-low enables, respectively, which can be externally connected together to function as a directional control. The driver differential outputs and the receiver differential inputs are connected

D OR P PACKAGE  
(TOP VIEW)



Function Tables  
DRIVER

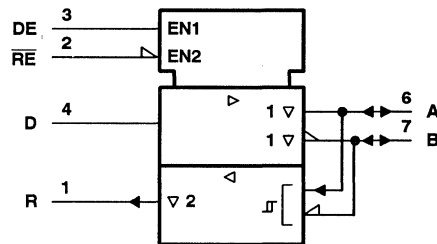
INPUT D	ENABLE DE	OUTPUTS	
		A	B
H	H	H	L
L	H	L	H
X	L	Z	Z

RECEIVER

DIFFERENTIAL INPUTS A – B	ENABLE RE	OUTPUT R
$V_{ID} \geq 0.2$ V	L	H
$-0.2$ V $< V_{ID} < 0.2$ V	L	?
$V_{ID} \leq -0.2$ V	L	L
X	H	Z
Inputs Open	L	H

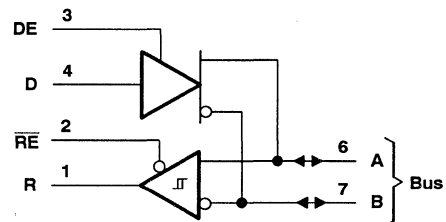
H = high level, L = low level, ? = indeterminate, X = irrelevant, Z = high impedance (off)

## logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



# TL3695 DIFFERENTIAL BUS TRANSCEIVER

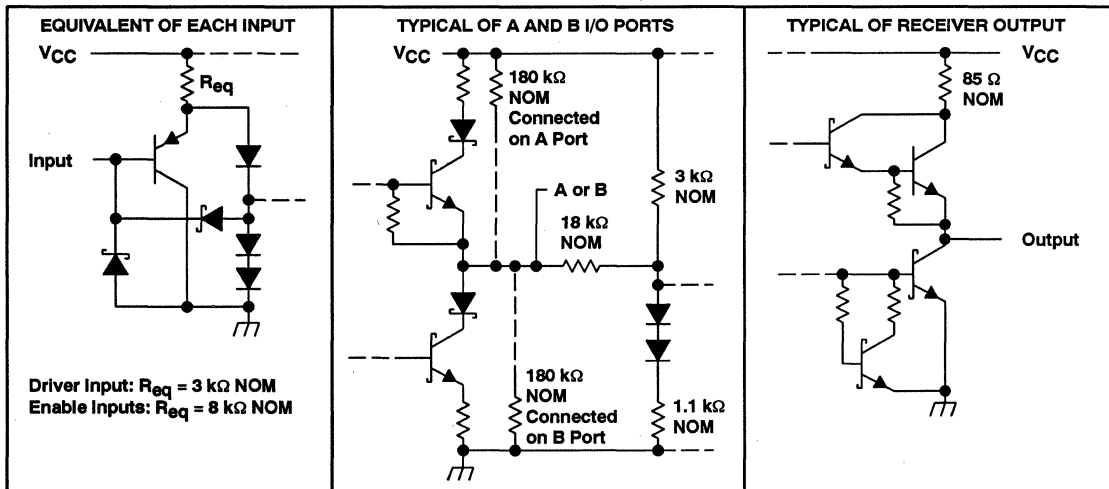
SLLS044B – D3408, NOVEMBER 1988 – REVISED FEBRUARY 1993

## description (continued)

internally to form a differential input/output (I/O) bus port that is designed to offer minimum loading to the bus whenever the driver is disabled or  $V_{CC} = 0$ . This port features wide positive and negative common-mode voltage ranges making the device suitable for party line applications.

The TL3695 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## schematic of inputs and outputs



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	7 V
Voltage range at any bus terminal	-10 V to 15 V
Enable input voltage	5.5 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	$0^{\circ}\text{C}$ to $70^{\circ}\text{C}$
Storage temperature range	$-65^{\circ}\text{C}$ to $150^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	$260^{\circ}\text{C}$

NOTE 1: All voltage values, except differential I/O bus voltage, are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	$T_A = 70^{\circ}\text{C}$ POWER RATING
D	725 mW	5.8 mW/ $^{\circ}\text{C}$	464 mW
P	1000 mW	8.0 mW/ $^{\circ}\text{C}$	640 mW

**recommended operating conditions**

		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.75	5	5.25	V
Voltage at any bus terminal (separately or common mode), $V_I$ or $V_{IC}$				12	V
				-7	V
High-level Input voltage, $V_{IH}$	D, DE, and $\overline{RE}$	2			V
Low-level Input voltage, $V_{IL}$	D, DE, and $\overline{RE}$			0.8	V
Differential input voltage, $V_{ID}$ (see Note 2)				$\pm 12$	V
High-level output current, $I_{OH}$	Driver			-60	mA
	Receiver			-400	$\mu$ A
Low-level output current, $I_{OL}$	Driver			60	mA
	Receiver			8	mA
Operating free-air temperature, $T_A$		0		70	$^{\circ}$ C

NOTE 2: Differential-input/output bus voltage is measured at the noninverting terminal A with respect to the inverting terminal B.

**DRIVER SECTION**

**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS†		MIN	TYP‡	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA				-1.5	V
$V_O$	Output voltage	$I_O = 0$		0		6	V
$ V_{OD1} $	Differential output voltage	$I_O = 0$		1.5		5	V
$ V_{OD2} $	Differential output voltage	$R_L = 100 \Omega$ ,	See Figure 1	$\frac{1}{2}V_{OD1}$ or $2^{\dagger}$			V
		$R_L = 54 \Omega$ ,	See Figure 1	1.5	2.5	5	V
$V_{OD3}$	Differential output voltage	$V_{test} = -7$ V to 12 V, See Figure 2		1.5		5	V
$\Delta  V_{OD} $	Change in magnitude of differential output voltage§	$R_L = 54 \Omega$ , See Figure 1				$\pm 0.2$	V
$V_{OC}$	Common-mode output voltage					3	V
$\Delta  V_{OC} $	Change in magnitude of common-mode output voltage§					$\pm 0.2$	V
$I_O$	Output current	Output disabled, See Note 3	$V_O = 12$ V			1	mA
			$V_O = -7$ V			-0.8	mA
$I_{IH}$	High-level input current	$V_I = 2.4$ V				20	$\mu$ A
$I_{IL}$	Low-level input current	$V_I = 0.4$ V				-200	$\mu$ A
$I_{OS}$	Short-circuit output current	$V_O = -6$ V				-250	mA
		$V_O = 0$				-150	
		$V_O = V_{CC}$				250	
		$V_O = 8$ V				250	
$I_{CC}$	Supply current	No load	Outputs enabled		23	50	mA
			Outputs disabled		19	35	

† The power-off measurement in EIA Standard RS-422-A applies to disabled outputs only and is not applied to combined inputs and outputs.

‡ All typical values are at  $V_{CC} = 5$  V and  $T_A = 25^{\circ}$ C.

§  $\Delta |V_{OD}|$  and  $\Delta |V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$  respectively, that occur when the input is changed from a high level to a low level.

† The minimum  $V_{OD2}$  with a 100- $\Omega$  load is either  $\frac{1}{2} V_{OD1}$  or 2 V whichever is greater.

NOTE 3: This applies for both power on and off; refer to EIA Standard RS-485 for exact conditions. The RS-422-A limit does not apply for a combined driver and receiver terminal.



# TL3695 DIFFERENTIAL BUS TRANSCEIVER

SLLS044B – D3408, NOVEMBER 1988 – REVISED FEBRUARY 1993

## switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$t_{dD}$ Differential-output delay time	$C_{L1} = C_{L2} = 100 \text{ pF}$ , $R_L = 60 \Omega$ , See Figure 3	8	22		ns
Skew ( $t_{dDH} - t_{dDL}$ )		1	8		ns
$t_{tD}$ Differential output transition time		8	18		ns
$t_{pZH}$ Output enable time to high level	$C_L = 100 \text{ pF}$ , $R_L = 500 \Omega$ , See Figure 4			50	ns
$t_{pZL}$ Output enable time to low level	$C_L = 100 \text{ pF}$ , $R_L = 500 \Omega$ , See Figure 5			50	ns
$t_{pHZ}$ Output disable time from high level	$C_L = 15 \text{ pF}$ , $R_L = 500 \Omega$ , See Figure 4		8	30	ns
$t_{pLZ}$ Output disable time from low level	$C_L = 15 \text{ pF}$ , $R_L = 500 \Omega$ , See Figure 5		8	30	ns

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

### SYMBOL EQUIVALENTS

DATA SHEET PARAMETER	RS-422-A	RS-485
$V_O$	$V_{Oa}, V_{Ob}$	$V_{Oa}, V_{Ob}$
$ V_{OD1} $	$V_O$	$V_O$
$ V_{OD2} $	$V_t (R_L = 100 \Omega)$	$V_t (R_L = 54 \Omega)$
$ V_{OD3} $		$V_t$ (Test Termination Measurement 2)
$V_{test}$		$V_{tst}$
$\Delta  V_{OD} $	$   V_t  -  \bar{V}_t   $	$   V_t  -  \bar{V}_t   $
$V_{OC}$	$ V_{os} $	$ V_{os} $
$\Delta  V_{OC} $	$ V_{os} - \bar{V}_{os} $	$ V_{os} - \bar{V}_{os} $
$I_{OS}$	$ I_{sa} ,  I_{sb} $	
$I_O$	$ I_{xa} ,  I_{xb} $	$I_{ia}, I_{ib}$

**RECEIVER SECTION**

**electrical characteristics over recommended ranges of common-mode input voltage, supply voltage, and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V <sub>T+</sub>	Positive-going input threshold voltage	V <sub>O</sub> = 2.7 V,	I <sub>O</sub> = -0.4 mA			0.2	V
V <sub>T-</sub>	Negative-going input threshold voltage	V <sub>O</sub> = 0.5 V,	I <sub>O</sub> = 8 mA	-0.2‡			V
V <sub>hys</sub>	Hysteresis (V <sub>T+</sub> -V <sub>T-</sub> )	V <sub>OC</sub> = 0			70		mV
V <sub>IK</sub>	Enable-input clamp voltage	I <sub>I</sub> = -18 mA				-1.5	V
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 200 mV or inputs open, I <sub>OH</sub> = -400 μA, See Figure 6		2.4			V
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -200 mV, See Figure 6	I <sub>OL</sub> = 16 mA			0.5	V
			I <sub>OL</sub> = 8 mA			0.45	
I <sub>OZ</sub>	High-impedance-state output current	V <sub>O</sub> = 0.4 V to 2.4 V				±20	μA
I <sub>I</sub>	Line input current	Other input = 0 V, See Note 4	V <sub>I</sub> = 12 V			1	mA
			V <sub>I</sub> = -7 V			-0.8	
I <sub>IH</sub>	High-level enable-input current	V <sub>IH</sub> = 2.7 V				20	μA
I <sub>IL</sub>	Low-level enable-input current	V <sub>IL</sub> = 0.4 V				-100	μA
r <sub>i</sub>	Input resistance			12			kΩ
I <sub>OS</sub>	Short-circuit output current	V <sub>O</sub> = 0		-15		-85	mA
I <sub>CC</sub>	Supply current	No load	Outputs enabled		23	50	mA
			Outputs disabled		19	35	

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated minimum, is used in this data sheet for common-mode input voltage and threshold voltage levels only.

NOTE 4: This applies for both power on and power off. Refer to EIA Standard RS-485 for exact conditions.

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature, C<sub>L</sub> = 15 pF**

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output	V <sub>ID</sub> = -1.5 V to 1.5 V, See Figure 7		14	37	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output				14	37
t <sub>PZH</sub>	Output enable time to high level	See Figure 8		7	20	ns
t <sub>PZL</sub>	Output enable time to low level				7	20
t <sub>PHZ</sub>	Output disable time from high level	See Figure 8		7	16	ns
t <sub>PLZ</sub>	Output disable time from low level				8	16

† All typical values are at V<sub>CC</sub> = 5 V and T<sub>A</sub> = 25°C.



# TL3695 DIFFERENTIAL BUS TRANSCEIVER

SLLS044B – D3408, NOVEMBER 1988 – REVISED FEBRUARY 1993

## PARAMETER MEASUREMENT INFORMATION

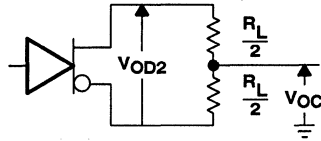


Figure 1. Driver  $V_{OD}$  and  $V_{OC}$

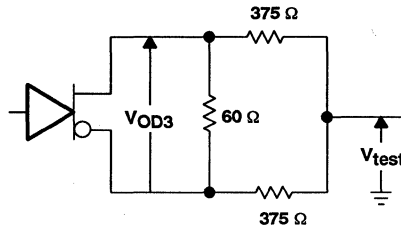
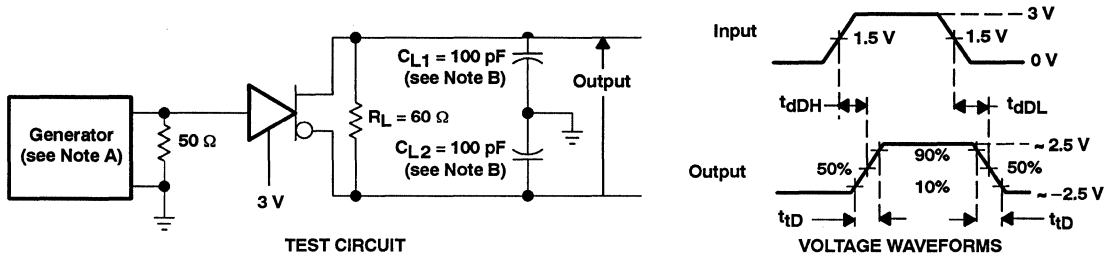
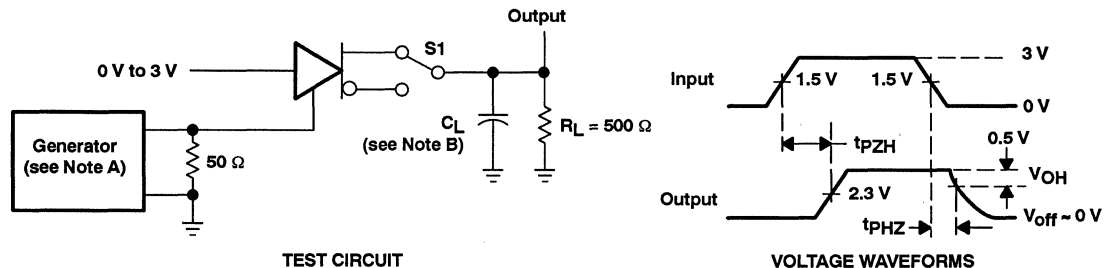


Figure 2. Driver  $V_{OD3}$



NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

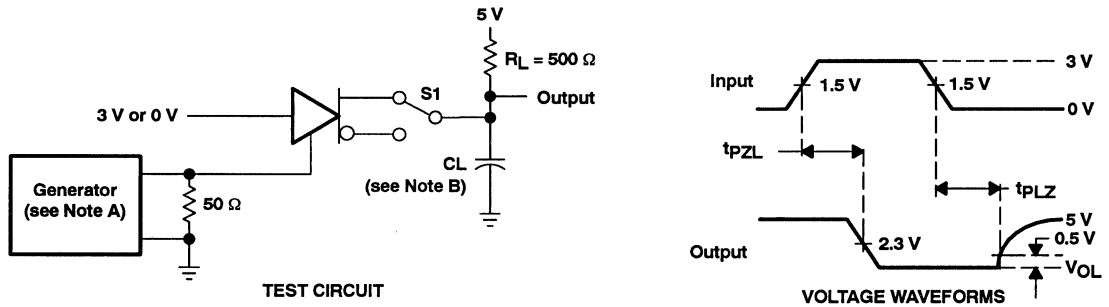
Figure 3. Driver Differential-Output Test Circuit and Voltage Waveforms



NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  10 ns,  $t_f \leq$  10 ns,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance (see switching characteristics – test conditions).

Figure 4. Driver Test Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION



NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  10 ns,  $t_f \leq$  10 ns,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance (see switching characteristics – test conditions).

Figure 5. Driver Test Circuit and Voltage Waveforms

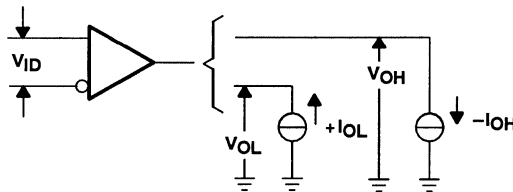
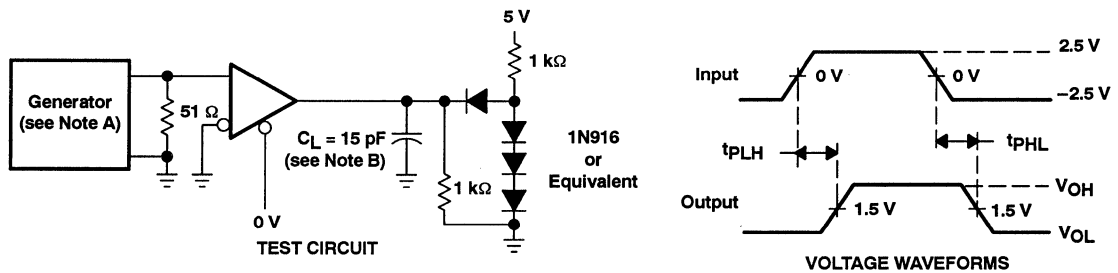


Figure 6. Receiver  $V_{OH}$  and  $V_{OL}$



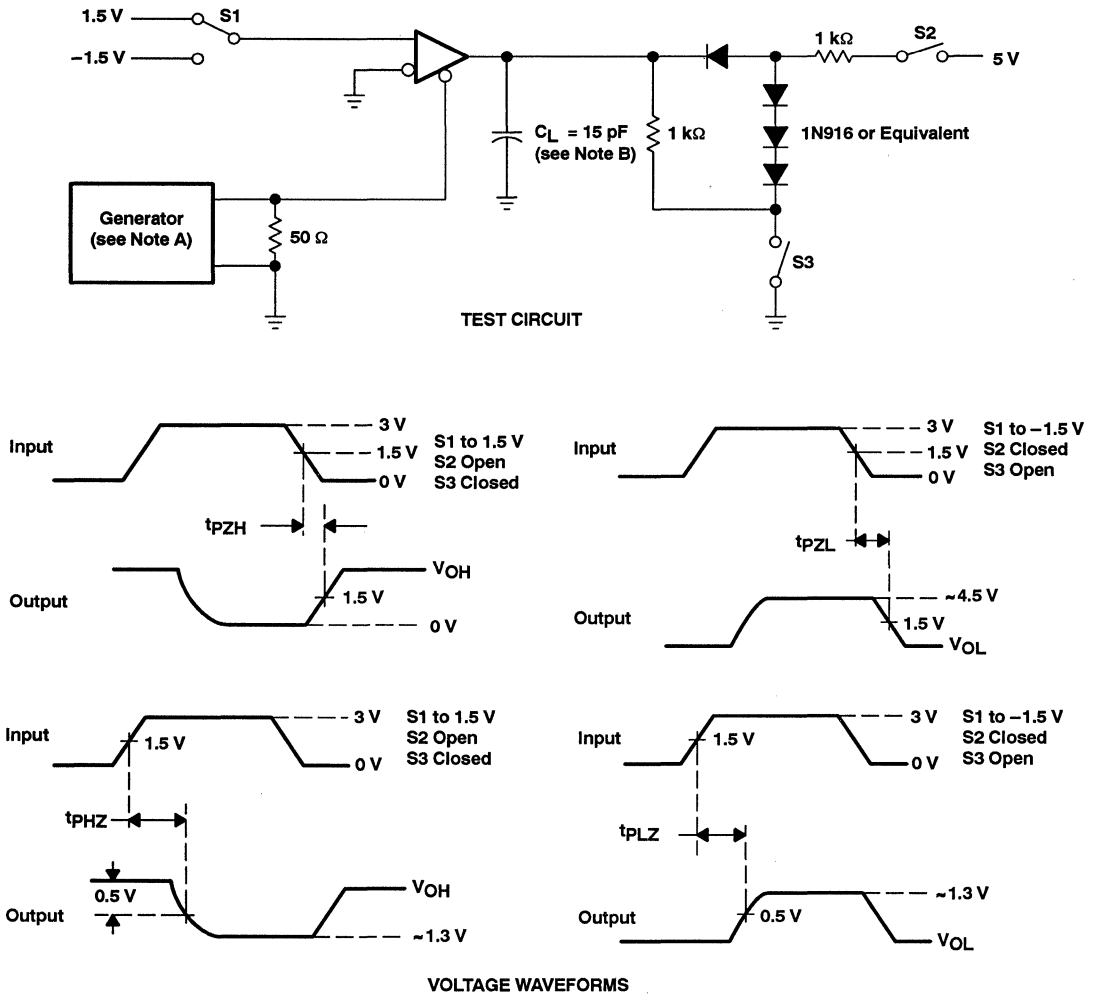
NOTES: A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  1 MHz, 50% duty cycle,  $t_r \leq$  10 ns,  $t_f \leq$  10 ns,  $Z_O = 50 \Omega$ .  
B.  $C_L$  includes probe and jig capacitance.

Figure 7. Receiver Test Circuit and Voltage Waveforms

**TL3695**  
**DIFFERENTIAL BUS TRANSCEIVER**

SLLS044B - D3408, NOVEMBER 1988 - REVISED FEBRUARY 1993

**PARAMETER MEASUREMENT INFORMATION**



NOTES: A. The input pulse is supplied by a generator having the following characteristics:  $PRR \leq 1 \text{ MHz}$ , 50% duty cycle,  $t_r \leq 10 \text{ ns}$ ,  $t_f \leq 10 \text{ ns}$ ,  $Z_O = 50 \Omega$ .  
 B.  $C_L$  includes probe and jig capacitance.

**Figure 8. Receiver Test Circuit and Voltage Waveforms**

TYPICAL CHARACTERISTICS

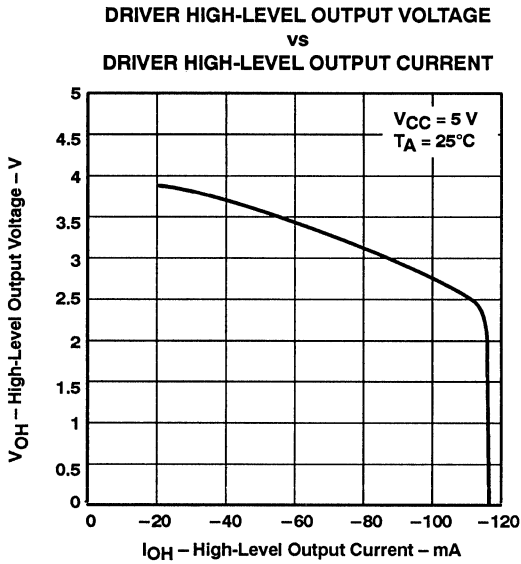


Figure 9

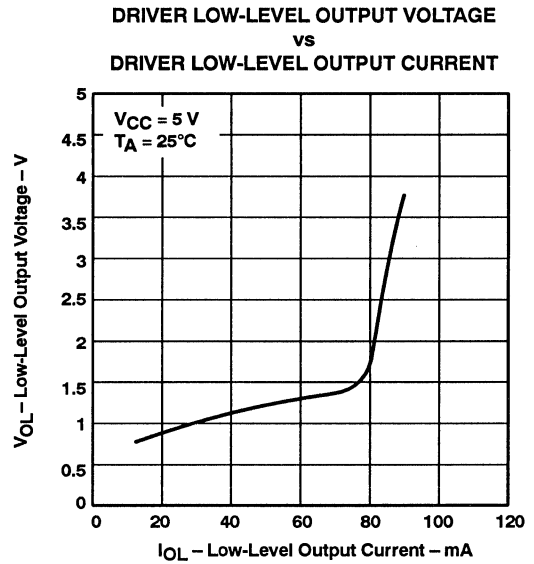


Figure 10

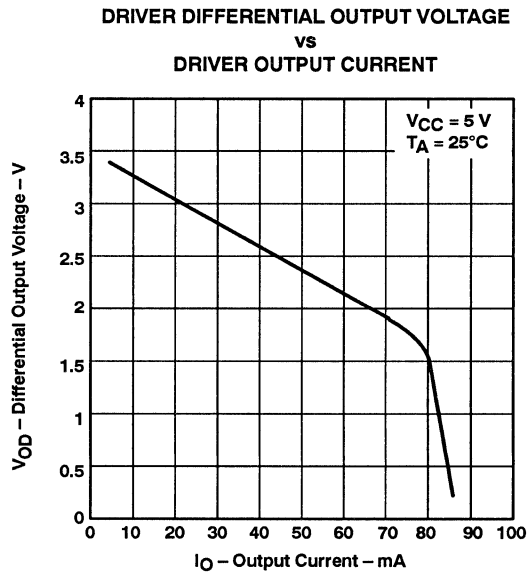


Figure 11

# TL3695 DIFFERENTIAL BUS TRANSCEIVER

SLLS044B - D3408, NOVEMBER 1988 - REVISED FEBRUARY 1993

## TYPICAL CHARACTERISTICS

RECEIVER HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

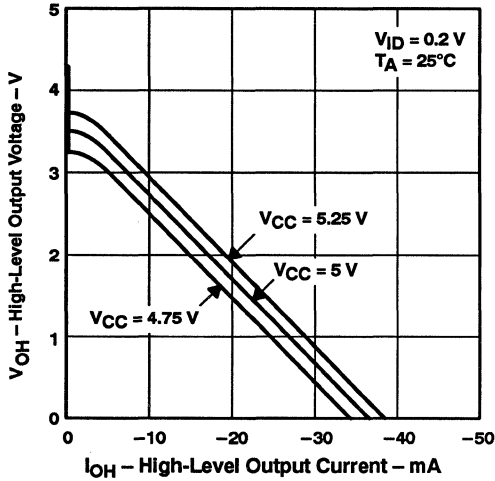


Figure 12

RECEIVER HIGH-LEVEL OUTPUT  
vs  
FREE-AIR TEMPERATURE

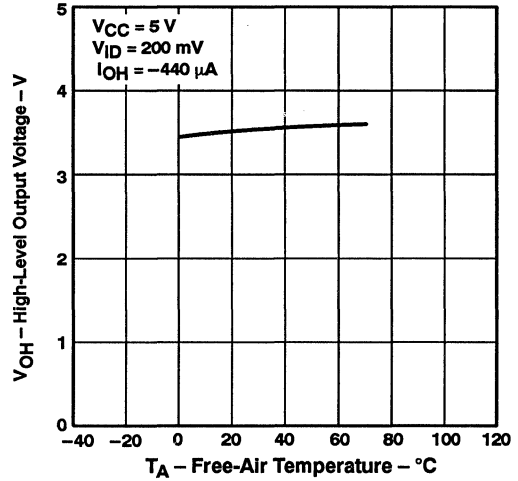


Figure 13

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
RECEIVER LOW-LEVEL OUTPUT CURRENT

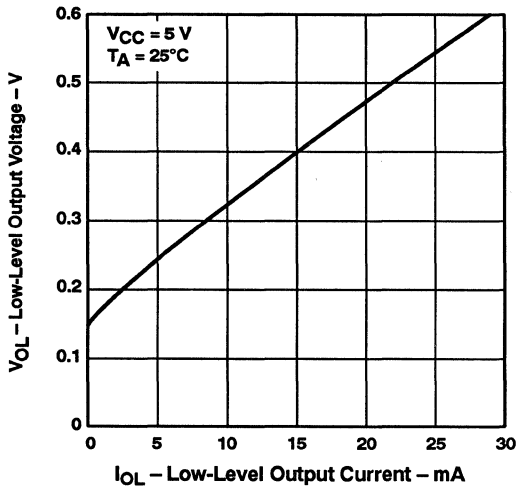


Figure 14

RECEIVER LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

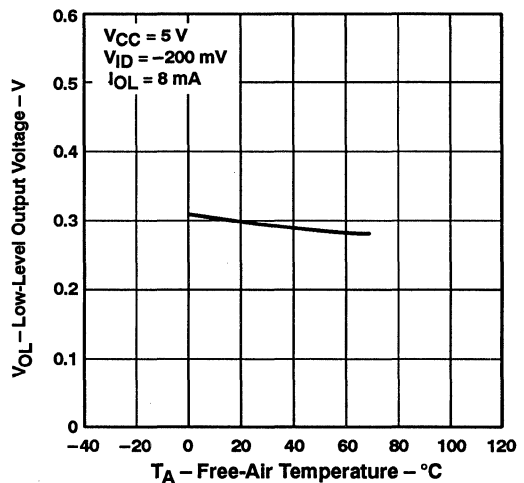


Figure 15

TYPICAL CHARACTERISTICS

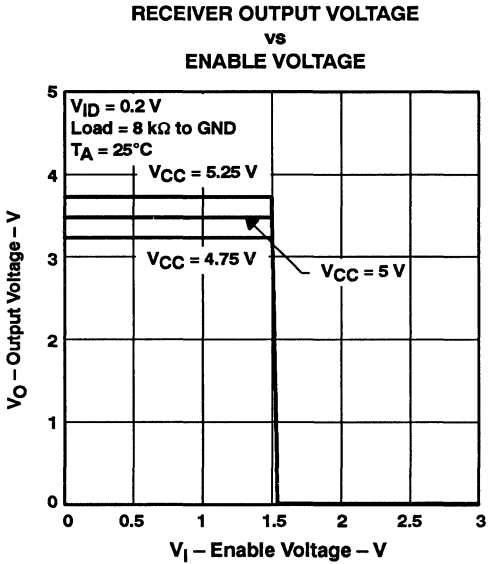


Figure 16

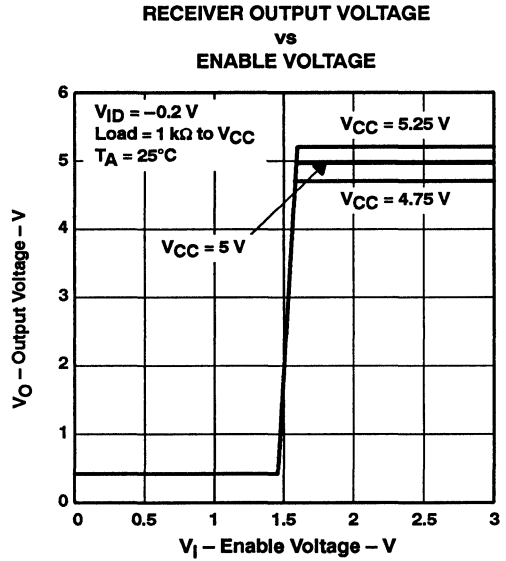
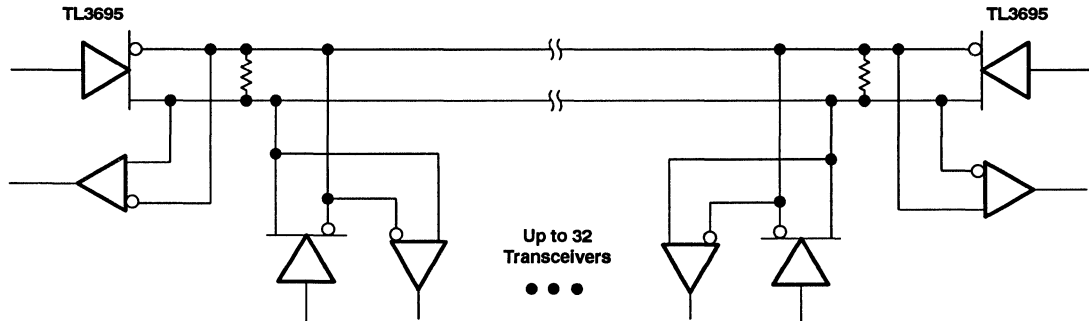


Figure 17

APPLICATION INFORMATION



NOTE: The line should be terminated at both ends in its characteristic impedance. Stub lengths off the main line should be kept as short as possible.

Figure 18. Typical Application Circuit



# uA9636AC DUAL LINE DRIVER WITH ADJUSTABLE SLEW RATE

SLLS110A – D2608, OCTOBER 1980 – REVISED MARCH 1993

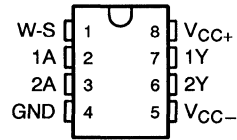
- Meets EIA Standards RS-423-A and RS-232-C and Federal Standard 1030
- Slew Rate Control
- Output Short-Circuit-Current Limiting
- Wide Supply Voltage Range
- 8-Pin Package
- Designed to Be Interchangeable With Fairchild 9636A

## description

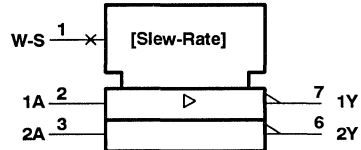
The uA9636AC is a dual single-ended line driver designed to meet EIA Standards RS-423-A and RS-232-C and Federal Standard 1030. The slew rates of both amplifiers are controlled by a single external resistor,  $R_{WS}$ , connected between the wave-shape-control (W-S) terminal and GND. Output current limiting is provided. Inputs are compatible with TTL and CMOS and are diode protected against negative transients. This device operates from  $\pm 12$  V and is supplied in an 8-pin package.

The uA9636AC is characterized for operation from 0°C to 70°C.

D OR P PACKAGE  
(TOP VIEW)

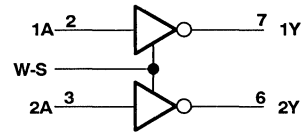


## logic symbol†

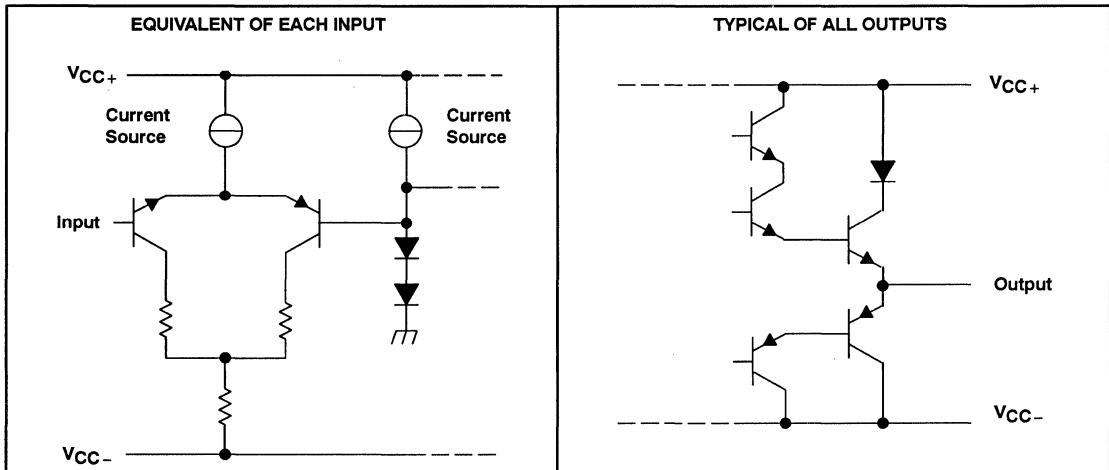


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram



## schematics of inputs and outputs



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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# uA9636AC DUAL LINE DRIVER WITH ADJUSTABLE SLEW RATE

SLLS110A - D2608, OCTOBER 1980 - REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Positive supply voltage range, $V_{CC+}$ (see Note 1)	$V_{CC-}$ to 15 V
Negative supply voltage range, $V_{CC-}$	0.5 V to -15 V
Output voltage	$\pm 15$ V
Output current	$\pm 150$ mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect to the network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Positive supply voltage, $V_{CC+}$	10.8	12	13.2	V
Negative supply voltage, $V_{CC-}$	-10.8	-12	-13.2	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$			0.8	V
Wave-shaping resistor, $R_{WS}$	10		1000	k $\Omega$
Operating free-air temperature, $T_A$	0		70	°C

# uA9636AC

## DUAL LINE DRIVER WITH ADJUSTABLE SLEW RATE

SLLS110A – D2608, OCTOBER 1980 – REVISED MARCH 1993

**electrical characteristics over recommended range of free-air temperature, supply voltage, and wave-shaping resistance (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -15 \text{ mA}$			-1.1	-1.5	V
$V_{OH}$	High-level output voltage	$V_I = 0.8 \text{ V}$	$R_L = \infty$	5	5.6	6	V
			$R_L = 3 \text{ k}\Omega$ to GND	5	5.6	6	
			$R_L = 450 \Omega$ to GND	4	5.4	6	
$V_{OL}$	Low-level output voltage	$V_I = 2 \text{ V}$	$R_L = \infty$	-6‡	-5.7	-5	V
			$R_L = 3 \text{ k}\Omega$ to GND	-6‡	-5.6	-5	
			$R_L = 450 \Omega$ to GND	-6‡	-5.4	-4	
$I_{IH}$	High-level input current	$V_I = 2.4 \text{ V}$				10	$\mu\text{A}$
		$V_I = 5.5 \text{ V}$				100	
$I_{IL}$	Low-level input current	$V_I = 0.4 \text{ V}$			-20	-80	$\mu\text{A}$
$I_O$	Output current (power off)	$V_{CC\pm} = 0$ ,	$V_O = \pm 6 \text{ V}$			$\pm 100$	$\mu\text{A}$
$I_{OS}$	Short-circuit output current§	$V_I = 2 \text{ V}$		15	25	150	mA
		$V_I = 0$		-15	-40	-150	
$r_o$	Output resistance	$R_L = 450 \Omega$			25	50	$\Omega$
$I_{CC+}$	Positive supply current	$V_{CC} = \pm 12 \text{ V}$ , $R_{WS} = 100 \text{ k}\Omega$ ,	$V_I = 0$ , Output open		13	18	mA
$I_{CC-}$	Negative supply current	$V_{CC} = \pm 12 \text{ V}$ , $R_{WS} = 100 \text{ k}\Omega$ ,	$V_I = 0$ , Output open		-13	-18	mA

† All typical values are at  $V_{CC} = \pm 12 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less-positive (more-negative) limit is designated as minimum, is used in this data sheet for logic voltage levels, e.g., when  $-5 \text{ V}$  is the maximum, the minimum is a more-negative voltage.

§ Not more than one output should be shorted to ground at a time.

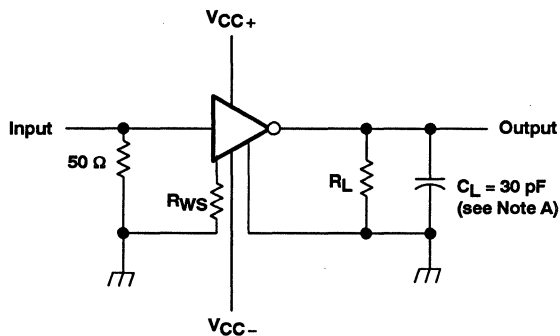
### switching characteristics, $V_{CC\pm} = \pm 12 \text{ V}$ , $T_A = 25^\circ\text{C}$ (see Figure 1)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT	
$t_{TLH}$	Transition time, low-to-high-level output	$R_L = 450 \text{ k}\Omega$ ,	$C_L = 30 \text{ pF}$	$R_{WS} = 10 \text{ k}\Omega$	0.8	1.1	1.4	$\mu\text{s}$
				$R_{WS} = 100 \text{ k}\Omega$	8	11	14	
				$R_{WS} = 500 \text{ k}\Omega$	40	55	70	
				$R_{WS} = 1 \text{ M}\Omega$	80	110	140	
$t_{THL}$	Transition time, high-to-low-level output	$R_L = 450 \text{ k}\Omega$ ,	$C_L = 30 \text{ pF}$	$R_{WS} = 10 \text{ k}\Omega$	0.8	1.1	1.4	$\mu\text{s}$
				$R_{WS} = 100 \text{ k}\Omega$	8	11	14	
				$R_{WS} = 500 \text{ k}\Omega$	40	55	70	
				$R_{WS} = 1 \text{ M}\Omega$	80	110	140	

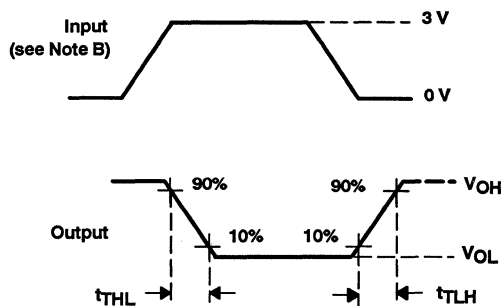
# uA9636AC DUAL LINE DRIVER WITH ADJUSTABLE SLEW RATE

SLLS110A - D2608, OCTOBER 1980 - REVISED MARCH 1993

## PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTES: A. C<sub>L</sub> includes probe and jig capacitance.

B. The input pulse is supplied by a generator having the following characteristics: t<sub>r</sub> ≤ 10 ns, t<sub>f</sub> ≤ 10 ns, Z<sub>0</sub> = 50 Ω, PRR ≤ 1 kHz, duty cycle = 50%.

Figure 1. Test Circuit and Voltage Waveforms

## TYPICAL CHARACTERISTICS

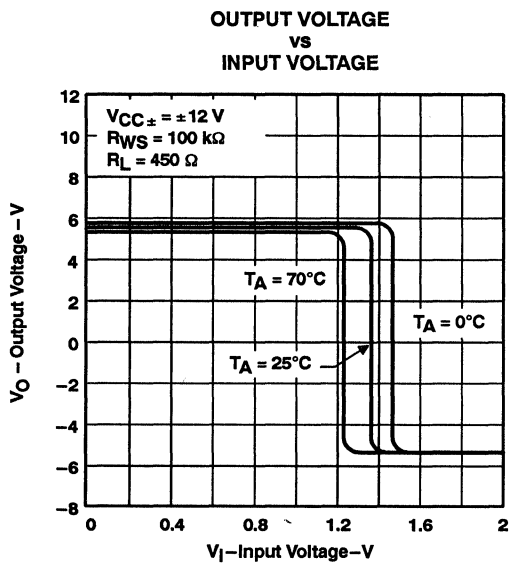


Figure 2

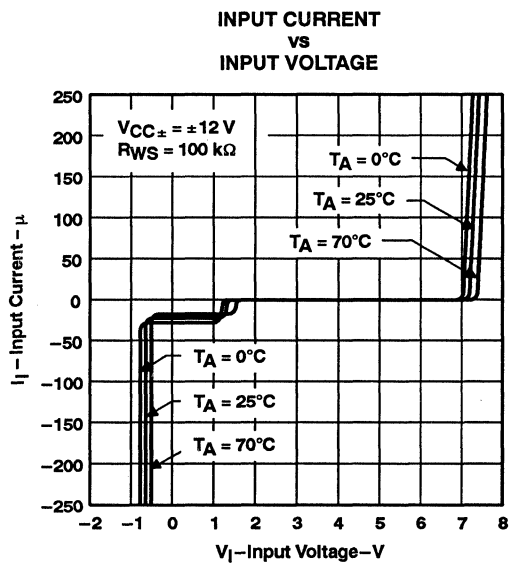


Figure 3

TYPICAL CHARACTERISTICS

OUTPUT CURRENT  
 vs  
 OUTPUT VOLTAGE  
 (POWER ON)

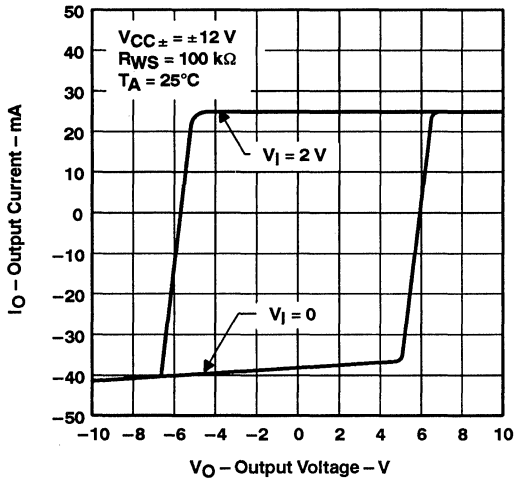


Figure 4

OUTPUT CURRENT  
 vs  
 OUTPUT VOLTAGE  
 (POWER OFF)

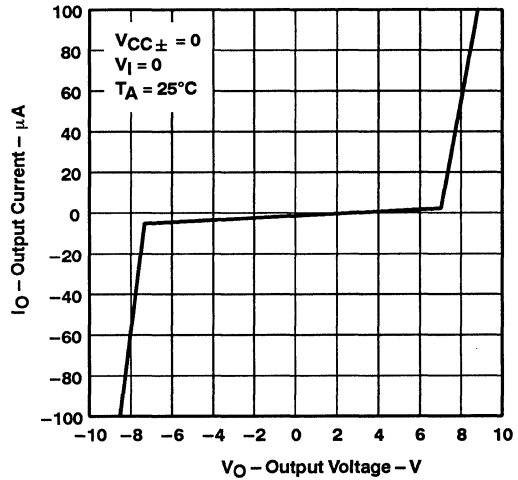


Figure 5

TRANSITION TIMES  
 vs  
 WAVE-SHAPING RESISTANCE

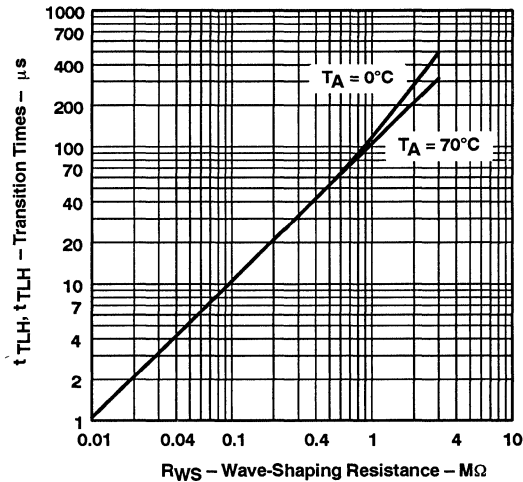


Figure 6

# uA9636AC DUAL LINE DRIVER WITH ADJUSTABLE SLEW RATE

SLLS110A - D2608, OCTOBER 1980 - REVISED MARCH 1993

## APPLICATION INFORMATION

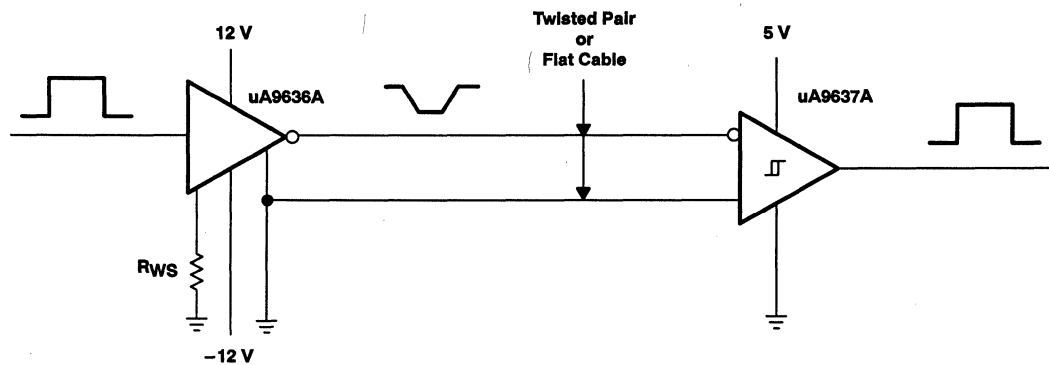


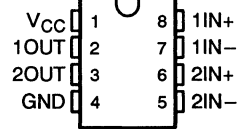
Figure 7. RS-423-A System Application

# uA9637AC DUAL DIFFERENTIAL LINE RECEIVER

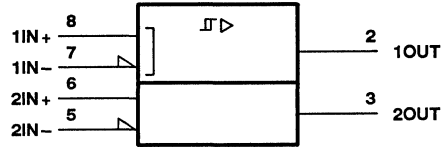
SLLS111A - D2609, SEPTEMBER 1980 - REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A and RS-423-A
- Meets Federal Standards 1020 and 1030
- Operates From Single 5-V Power Supply
- Wide Common-Mode Voltage Range
- High Input Impedance
- TTL-Compatible Outputs
- High-Speed Schottky Circuitry
- 8-Pin Dual-In-Line and Small-Outline Packages
- Similar to SN75157 Except for Corner  $V_{CC}$  and Ground Pin Positions
- Designed to Be Interchangeable With Fairchild  $\mu$ A9637A

uA9637C . . . D OR P PACKAGE  
(TOP VIEW)



### logic symbol†



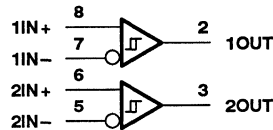
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

### description

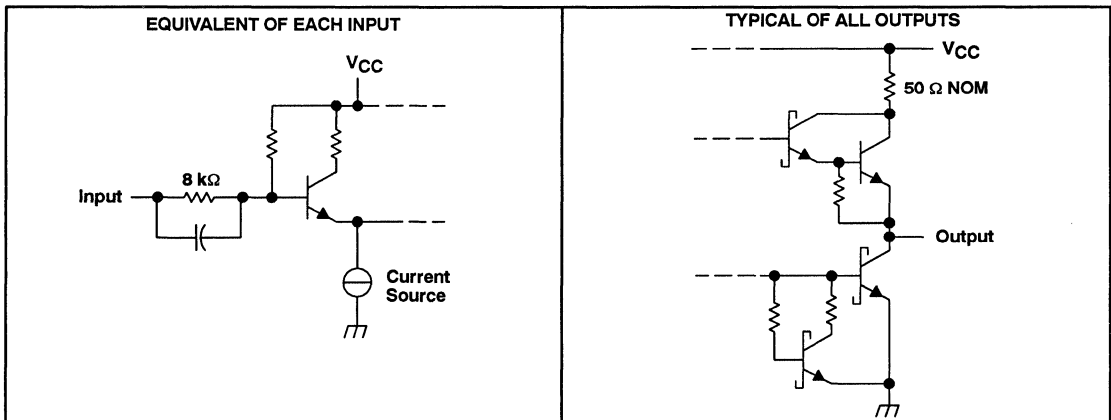
The uA9637AC is a dual differential line receiver designed to meet EIA standards RS-422-A and RS-423-A and Federal Standards 1020 and 1030. They utilize Schottky circuitry and have TTL-compatible outputs. The inputs are compatible with either a single-ended or a differential-line system. This device operates from a single 5-V power supply and is supplied in an 8-pin dual-in-line package and small-outline package.

The uA9637AC is characterized for operation from 0°C to 70°C.

### logic diagram



### schematics of inputs and outputs



PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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2-985

# uA9637AC DUAL DIFFERENTIAL LINE RECEIVER

SLLS111A – D2609, SEPTEMBER 1980 – REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1)	–0.5 V to 7 V
Input voltage, $V_I$	$\pm 15$ V
Differential input voltage (see Note 2)	$\pm 15$ V
Output voltage, $V_O$ (see Note 1)	–0.5 V to 5.5 V
Low-level output current, $I_{OL}$	50 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to the network ground terminal.  
2. Differential input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	—
P	1000 mW	8.0 mW/°C	640 mW	—

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended ranges of supply voltage, common-mode input voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going input threshold voltage	See Note 4			0.2 0.4	V
$V_{T-}$ Negative-going input threshold voltage	See Note 4	–0.2 –0.4‡			V
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )			70		mV
$V_{OH}$ High-level output voltage	$V_{ID} = 0.2$ V, $I_O = -1$ mA	2.5	3.5		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -0.2$ V, $I_O = 20$ mA		0.35	0.5	V
$I_I$ Input current	$V_{CC} = 0$ to 5.5 V, See Note 5 $V_I = 10$ V $V_I = -10$ V		1.1 –1.6	3.25 –3.25	mA
$I_{OS}$ Short-circuit output current§	$V_O = 0$ , $V_{ID} = 0.2$ V	–40	–75	–100	mA
$I_{CC}$ Supply current	$V_{ID} = -0.5$ V, No load		35	50	mA

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold levels only.

§ Only one output should be shorted at a time, and duration of the short circuit should not exceed one second.

NOTES: 3. The expanded threshold parameter is tested with a 500- $\Omega$  resistor in series with each input.

4. The input not under test is grounded.



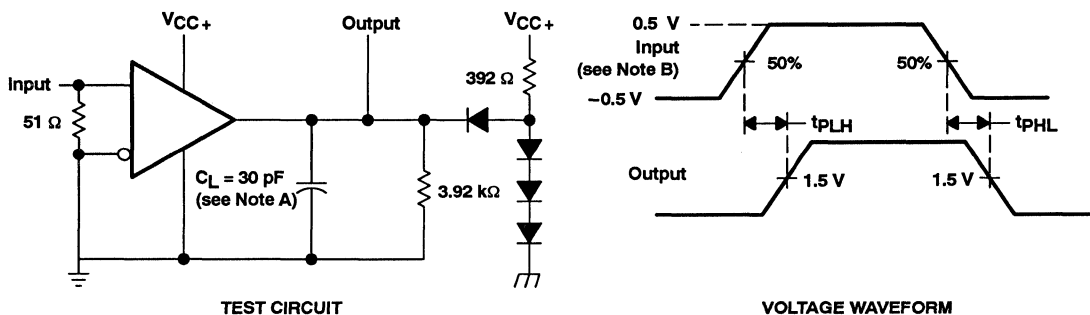
# uA9637AC DUAL DIFFERENTIAL LINE RECEIVER

SLLS111A – D2609, SEPTEMBER 1980 – REVISED FEBRUARY 1993

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 30\text{ pF}$ , See Figure 1		15	25	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			13	25	ns

## PARAMETER MEASUREMENT INFORMATION



NOTES: A.  $C_L$  includes probe and jig capacitance.  
 B. The input pulse is supplied by a generator having the following characteristics:  $t_r \leq 5\text{ ns}$ ,  $t_f \leq 5\text{ ns}$ ,  $PRR \leq 5\text{ MHz}$ , duty cycle = 50%.

Figure 1. Test Circuit and Voltage Waveform

## TYPICAL CHARACTERISTICS

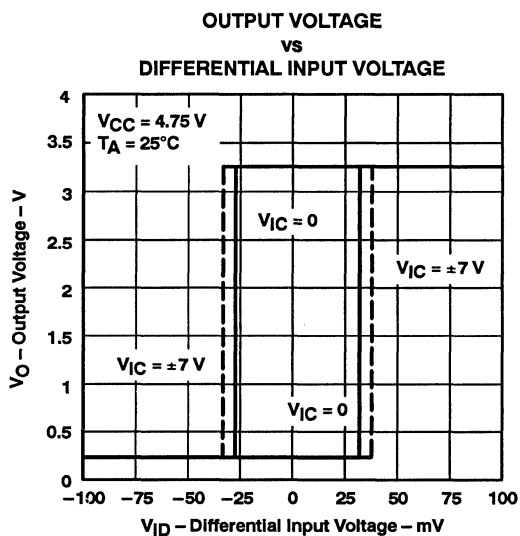


Figure 2

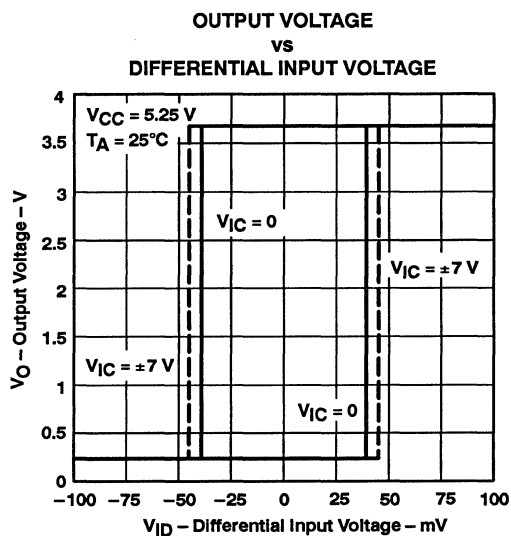


Figure 3

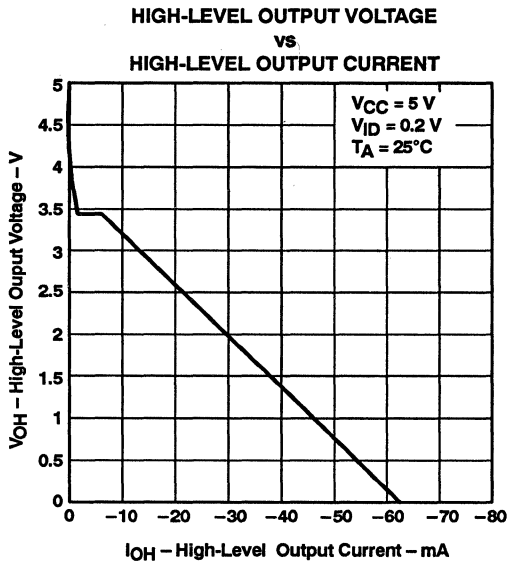




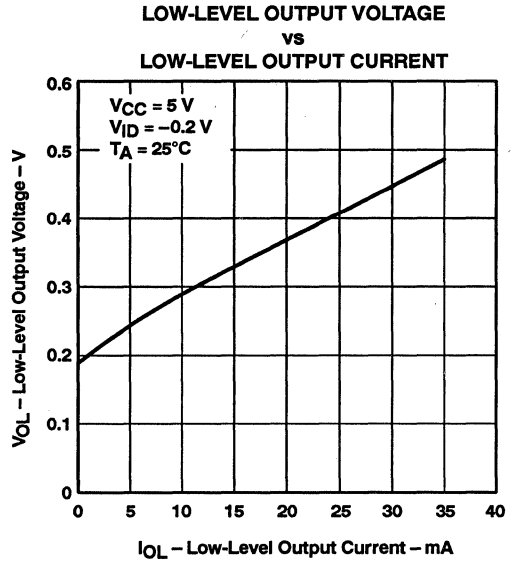
**uA9637AC**  
**DUAL DIFFERENTIAL LINE RECEIVER**

SLS111A - D2609, SEPTEMBER 1980 - REVISED FEBRUARY 1993

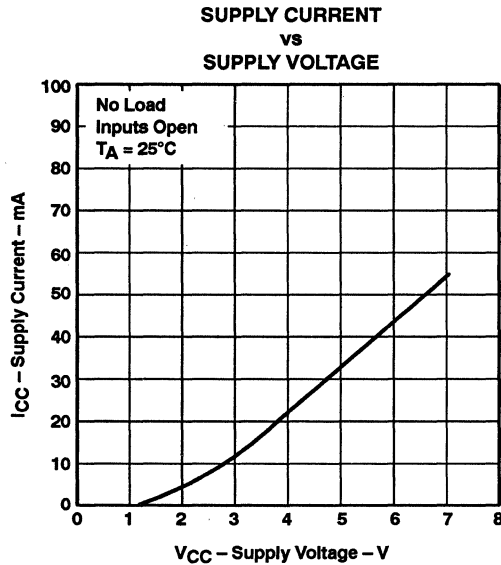
**TYPICAL CHARACTERISTICS**



**Figure 4**



**Figure 5**



**Figure 6**

APPLICATION INFORMATION

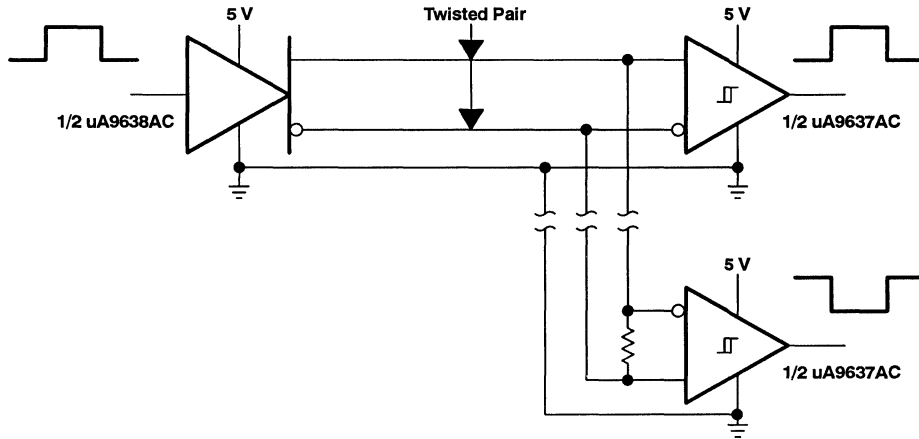


Figure 7. RS-A System Applications

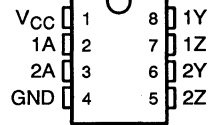


# uA9638C DUAL HIGH-SPEED DIFFERENTIAL LINE DRIVER

SLLS112A - D2612, OCTOBER 1980 - REVISED MARCH 1993

- Meets EIA Standard RS-422-A
- Operates From a Single 5-V Supply
- TTL-and CMOS-Input Compatibility
- Output Short-Circuit Protection
- Schottky Circuitry
- Designed to Be Interchangeable With Fairchild 9638

D OR P PACKAGE  
(TOP VIEW)

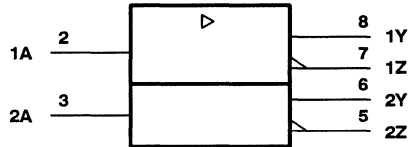


## description

The uA9638C is a dual high-speed differential line driver designed to meet EIA Standard RS-422-A. The inputs are TTL- and CMOS-compatible and have input clamp diodes. Schottky-diode-clamped transistors are used to minimize propagation delay time. This device operates from a single 5-V power supply and is supplied in an 8-pin package.

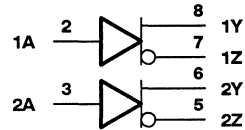
The uA9638C is characterized for operation from 0°C to 70°C.

## logic symbol†

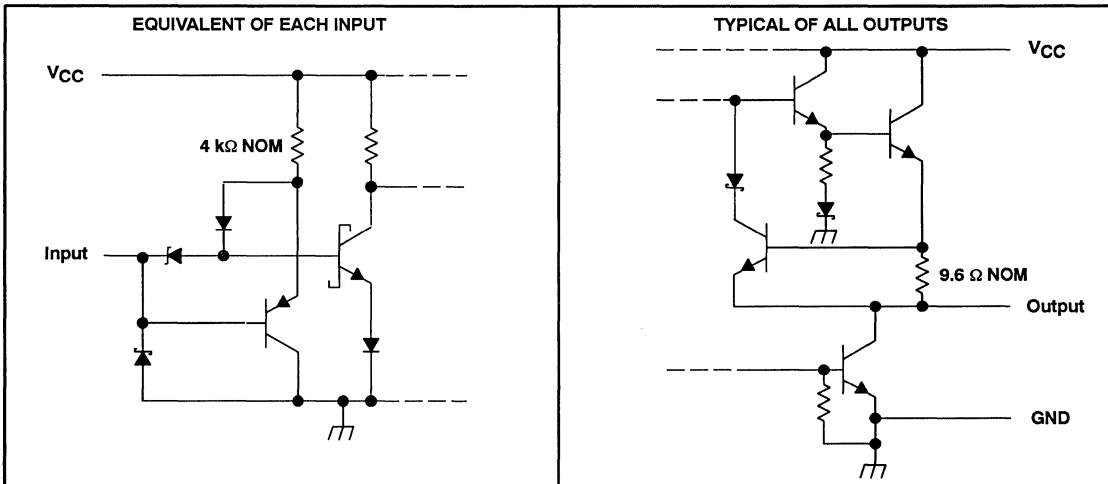


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram



## schematics of inputs and outputs



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**TEXAS  
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# uA9638C DUAL HIGH-SPEED DIFFERENTIAL LINE DRIVER

SLLS112A – D2612, OCTOBER 1980 – REVISED MARCH 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	–0.5 V to 7 V
Input voltage range, $V_I$	–0.5 V to 7 V
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from 10 seconds	260°C

NOTE 1: Voltage values except differential output voltages are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			V
Low-level input voltage, $V_{IL}$	0.8			V
High-level output current, $I_{OH}$	–50			mA
Low-level output current, $I_{OL}$	50			mA
Operating free-air temperature, $T_A$	0	70		°C

# uA9638C DUAL HIGH-SPEED DIFFERENTIAL LINE DRIVER

SLLS112A – D2612, OCTOBER 1980 – REVISED MARCH 1993

## electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT	
$V_{IK}$	Input clamp voltage	$V_{CC} = 4.75 \text{ V}$ , $I_I = -18 \text{ mA}$		-1	-1.2		V	
$V_{OH}$	High-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IH} = 2 \text{ V}$ , $V_{IL} = 0.8 \text{ V}$	$I_{OH} = -10 \text{ mA}$	2.5	3.5		V	
			$I_{OH} = -40 \text{ mA}$	2				
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.75 \text{ V}$ , $V_{IH} = 0.2 \text{ V}$ , $I_{OL} = 40 \text{ mA}$	$V_{IL} = 0.8 \text{ V}$			0.5	V	
$ V_{OD1} $	Differential output voltage	$V_{CC} = 5.25 \text{ V}$ , $I_O = 0$				$2V_{OD2}$	V	
$ V_{OD2} $	Differential output voltage	$V_{CC} = 4.75 \text{ V}$ to $5.25 \text{ V}$ , See Figure 1		2			V	
$\Delta V_{OD} $	Change in magnitude of differential output voltage‡			$R_L = 100 \Omega$		$\pm 0.4$		V
$V_{OC}$	Common-mode output voltage§					3		V
$\Delta V_{OC} $	Change in magnitude of common-mode output voltage‡					$\pm 0.4$		V
$I_O$	Output current with power off	$V_{CC} = 0$	$V_O = 6 \text{ V}$		0.1	100	$\mu\text{A}$	
			$V_O = -0.25 \text{ V}$		-0.1	-100		
			$V_O = -0.25 \text{ V}$ to $6 \text{ V}$		$\pm 100$			
$I_I$	Input current	$V_{CC} = 5.25 \text{ V}$ , $V_I = 5.5 \text{ V}$				50	$\mu\text{A}$	
$I_{IH}$	High-level input current	$V_{CC} = 5.25 \text{ V}$ , $V_I = 2.7 \text{ V}$				25	$\mu\text{A}$	
$I_{IL}$	Low-level input current	$V_{CC} = 5.25 \text{ V}$ , $V_I = 0.5 \text{ V}$				-200	$\mu\text{A}$	
$I_{OS}$	Short-circuit output current¶	$V_{CC} = 5.25 \text{ V}$ , $V_O = 0$		-50		-150	mA	
$I_{CC}$	Supply current (both drivers)	$V_{CC} = 5.25 \text{ V}$ , No load, All inputs at 0 V			45	65	mA	

† All typical values are at  $V_{CC} = 5 \text{ V}$  and  $T_A = 25^\circ\text{C}$ .

‡  $\Delta|V_{OD}|$  and  $\Delta|V_{OC}|$  are the changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively, that occur when the input is changed from a high level to a low level.

§ In EIA Standard RS-422-A,  $V_{OC}$ , which is the average of the two output voltages with respect to ground, is called output offset voltage,  $V_{OS}$ .

¶ Only one output at a time should be shorted, and duration of the short circuit should not exceed one second.

## switching characteristics, $V_{CC} = 5 \text{ V}$ , $T_A = 25^\circ\text{C}$

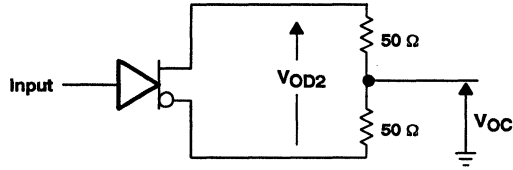
PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{dD}$	Differential-output delay time	$C_L = 15 \text{ pF}$ , $R_L = 100 \Omega$ , See Figure 2			10	15	ns
$t_{TD}$	Differential-output transition time				10	15	ns
Skew				See Figure 2		1	



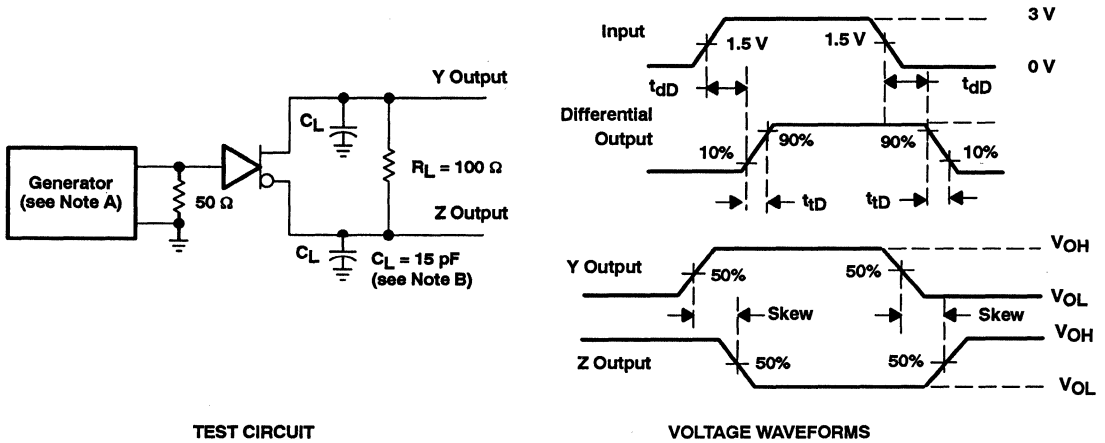
**uA9638C**  
**DUAL HIGH-SPEED DIFFERENTIAL LINE DRIVER**

SLLS112A – D2612, OCTOBER 1980 – REVISED MARCH 1993

**PARAMETER MEASUREMENT INFORMATION**



**Figure 1. Differential and Common-Mode Output Voltages**



NOTES: A. The input pulse generator has the following characteristics:  $Z_O = 50 \Omega$ ,  $PRR \leq 500 \text{ kHz}$ ,  $t_w = 100 \text{ ns}$ ,  $t_r = \leq 5 \text{ ns}$ .  
 B.  $C_L$  includes probe and jig capacitance.

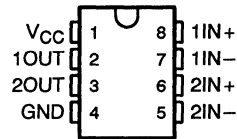
**Figure 2. Test Circuit and Voltage Waveforms**

# uA9639C DUAL DIFFERENTIAL LINE RECEIVER

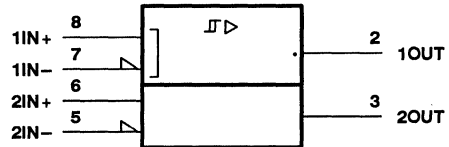
SLLS113A - D3009, OCTOBER 1986 - REVISED FEBRUARY 1993

- Meets EIA Standards RS-422-A and RS-423-A
- Meets Federal Standards 1020 and 1030
- Operates From Single 5-V Power Supply
- Wide Common-Mode Voltage Range
- High Input Impedance
- TTL-Compatible Outputs
- High-Speed Schottky Circuitry
- 8-Pin Dual-In-Line and Small-Outline Packages
- Designed to Be Interchangeable With Fairchild  $\mu$ A9639AC

D OR P PACKAGE  
(TOP VIEW)



logic symbol†



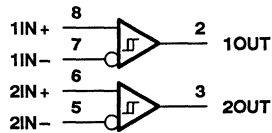
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## description

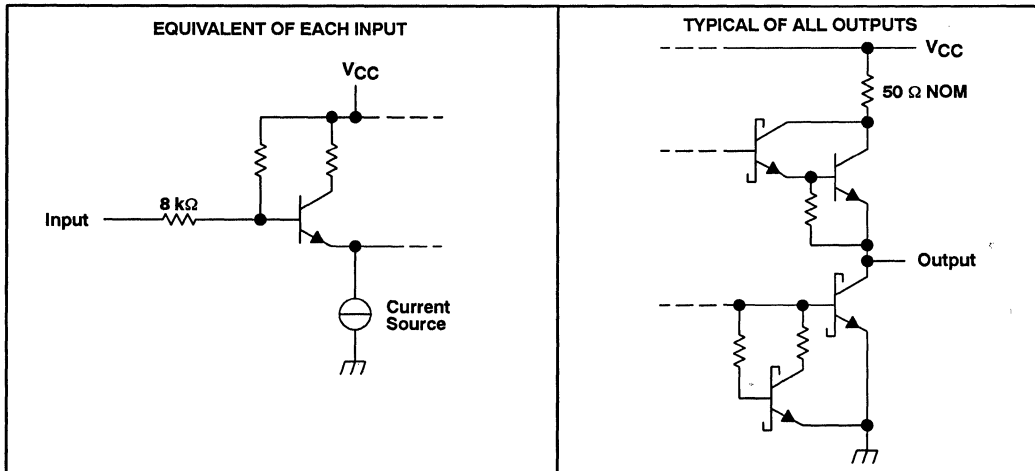
The uA9639C is a dual differential line receiver designed to meet EIA standards RS-422-A and RS-423-A and Federal Standards 1020 and 1030. It utilizes Schottky circuitry and has TTL-compatible outputs. The inputs are compatible with either a single-ended or a differential-line system. This device operates from a single 5-V power supply and is supplied in an 8-pin dual-in-line package and small-outline package.

The uA9639C is characterized for operation from 0°C to 70°C.

## logic diagram



## schematics of inputs and outputs



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2-995



# uA9639C DUAL DIFFERENTIAL LINE RECEIVER

SLLS113A - D3009, OCTOBER 1986 - REVISED FEBRUARY 1993

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	-0.5 V to 7 V
Input voltage, $V_I$	$\pm 15$ V
Differential input voltage (see Note 2)	$\pm 15$ V
Output voltage range (see Note 1)	-0.5 V to 5.5 V
Low-level output current	50 mA
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential input voltage, are with respect to the network ground terminal.  
2. Differential input voltage is measured at the noninverting input with respect to the corresponding inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
Common-mode input voltage, $V_{IC}$			$\pm 7$	V
Operating free-air temperature, $T_A$	0		70	°C

## electrical characteristics over recommended ranges of supply voltage, common-mode input voltage, and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{T+}$ Positive-going input threshold voltage	See Note 3			0.2 0.4	V
$V_{T-}$ Negative-going input threshold voltage	See Note 3	-0.2		-0.4‡	V
$V_{hys}$ Hysteresis ( $V_{T+} - V_{T-}$ )			70		mV
$V_{OH}$ High-level output voltage	$V_{ID} = 0.2$ V, $I_O = -1$ mA	2.5	3.5		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -0.2$ V, $I_O = 20$ mA	0.35	0.5		V
$I_I$ Input current	$V_{CC} = 0$ to 5.5 V, $V_I = 10$ V See Note 5 $V_I = -10$ V		1.1	3.25	mA
$I_{OS}$ Short-circuit output current§	$V_O = 0$ , $V_{ID} = 0.2$ V	-40	-75	-100	mA
$I_{CC}$ Supply current	$V_{ID} = -0.5$ V, No load		35	50	mA

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

‡ The algebraic convention, in which the less positive (more negative) limit is designated as minimum, is used in this data sheet for threshold levels only.

§ Only one output should be shorted at a time, and duration of the short circuit should not exceed one second.

NOTES: 3. The expanded threshold parameter is tested with a 500- $\Omega$  resistor in series with each input.

4. The input not under test is grounded.

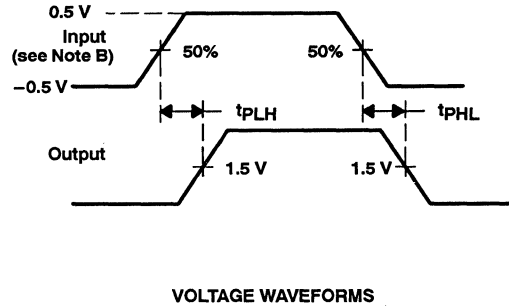
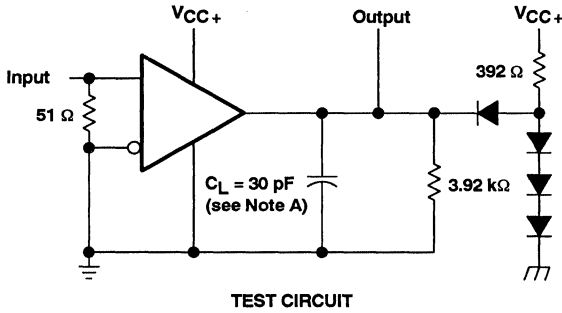
## switching characteristics, $V_{CC} = 5$ V, $T_A = 0^\circ\text{C}$ to $70^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
$t_{PLH}$ Propagation delay time, low-to-high-level output	$C_L = 50$ pF, See Figure 1		85	ns
$t_{PHL}$ Propagation delay time, high-to-low-level output			85	ns

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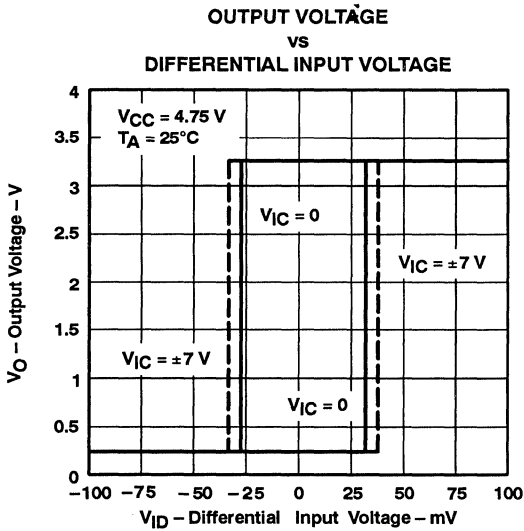
**PARAMETER MEASUREMENT INFORMATION**



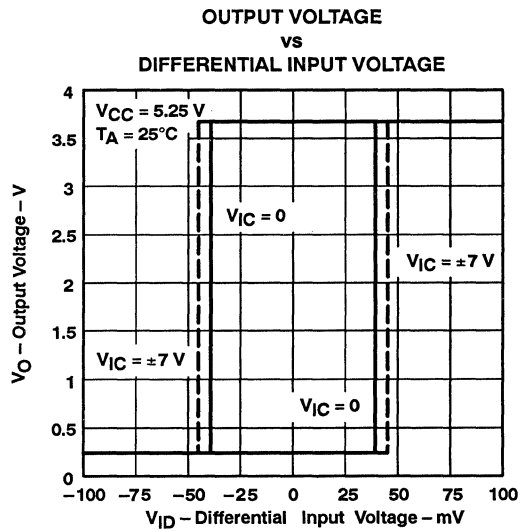
- NOTES: A. CL includes probe and jig capacitance.  
 B. The input pulse is supplied by a generator having the following characteristics: tr ≤ 5 ns, tf ≤ 5 ns, PRR ≤ 5 MHz, duty cycle = 50%.

**Figure 1. Test Circuit and Voltage Waveforms**

**TYPICAL CHARACTERISTICS**



**Figure 2**



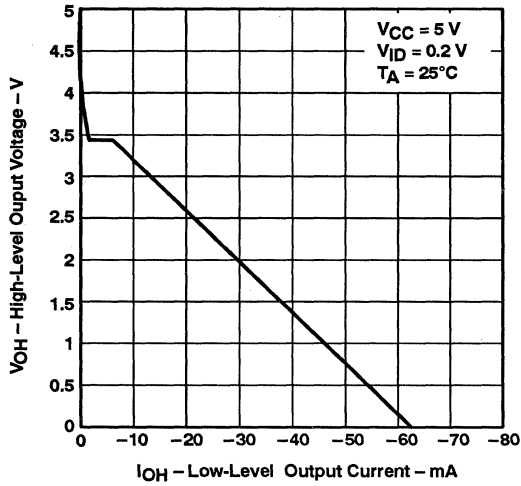
**Figure 3**

**uA9639C**  
**DUAL DIFFERENTIAL LINE RECEIVER**

SLLS113A - D3009, OCTOBER 1986 - REVISED FEBRUARY 1993

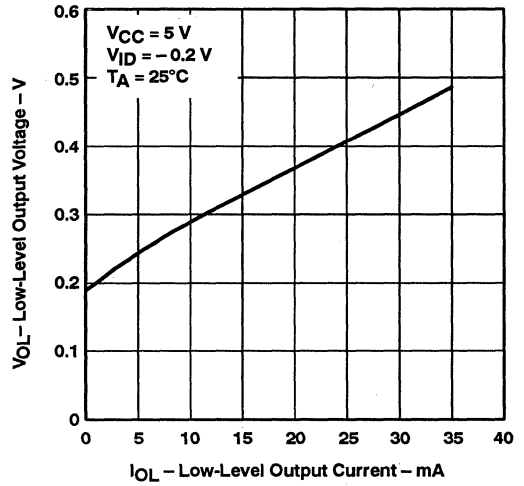
**TYPICAL CHARACTERISTICS**

**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT**



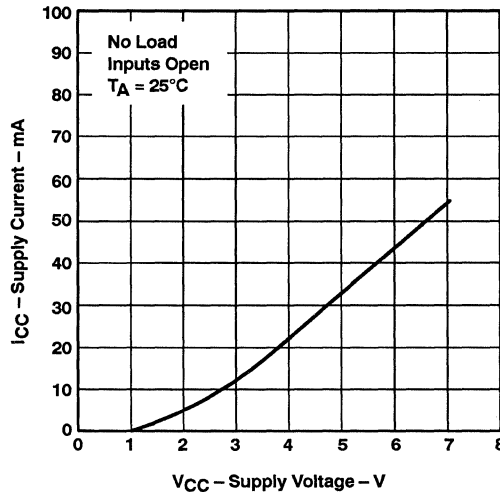
**Figure 4**

**LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT**



**Figure 5**

**SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE**



**Figure 6**

APPLICATION INFORMATION

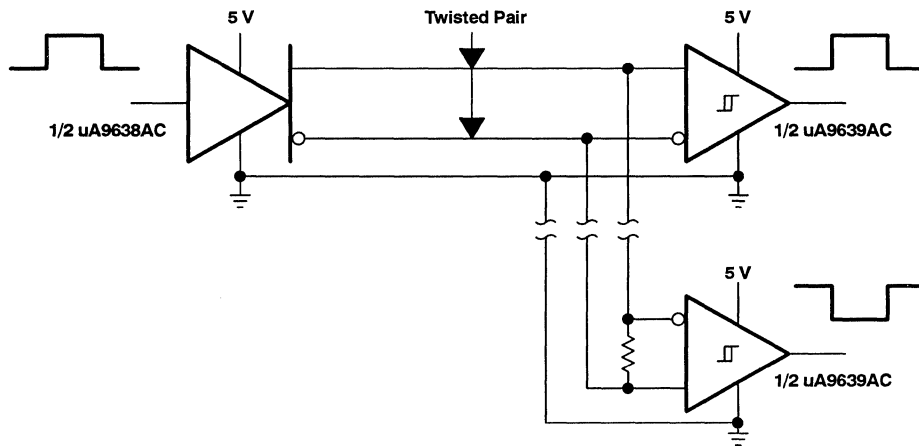


Figure 7. RS-422-A System Applications



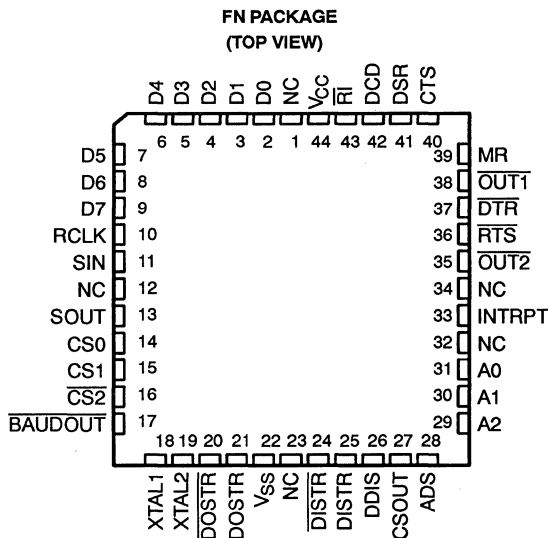
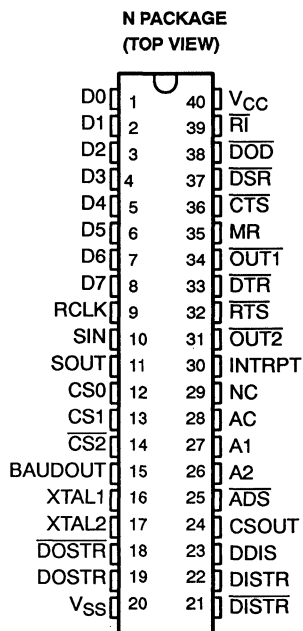
<b>General Information</b>	<b>1</b>
<b>Line Drivers, Receivers, Transceivers</b>	<b>2</b>
<b>Universal Async Receivers/Transmitters</b>	<b>3</b>
<b>Explanation of Logic Symbols</b>	<b>4</b>
<b>Applications</b>	<b>5</b>
<b>Mechanical Data</b>	<b>6</b>

## Universal Async Receivers/Transmitters

# TL16C450 ASYNCHRONOUS COMMUNICATIONS ELEMENT

SLLS037A - D3096, MARCH 1988 - REVISED APRIL 1989

- Programmable Baud Rate Generator Allows Division of Any Input Reference Clock by 1 to  $(2^{16} - 1)$  and Generates an Internal 16 X Clock
- Full Double Buffering Eliminates the Need for Precise Synchronization
- Adds or Deletes Standard Asynchronous Communication Bits (Start, Stop, and Parity) to or From the Serial Data Stream
- Independent Receiver Clock Input
- Independently Controlled Transmit, Receive, Line Status, and Data Set Interrupts
- Fully Programmable Serial Interface Characteristics:
  - 5-, 6-, 7-, or 8-Bit Characters
  - Even-, Odd-, or No-Parity Bit Generation and Detection
  - 1-, 1 1/2-, or 2-Stop Bit Generation
  - Baud Generation (dc to 256 kb/s Per Second)
- False Start Bit Detection
- Complete Status Reporting Capabilities
- 3-State TTL Drive Capabilities for Bidirectional Data Bus and Control Bus
- Line Break Generation and Detection
- Internal Diagnostic Capabilities:
  - Loopback Controls for Communications
  - Link Fault Isolation
  - Break, Parity, Overrun, Framing Error Simulation
- Fully-Prioritized Interrupt System Controls
- Modem Control Functions ( $\overline{\text{CTS}}$ ,  $\overline{\text{RTS}}$ ,  $\overline{\text{DSR}}$ ,  $\overline{\text{DTR}}$ ,  $\overline{\text{RI}}$ , and  $\overline{\text{DCD}}$ )
- Easily Interfaces to Most Popular Microprocessors
- Faster Plug-In Replacement for National Semiconductor NS16C450



NC - No internal connection

## description

The TL16C450 is a CMOS version of an asynchronous communications element (ACE). It typically functions in a microcomputer system as a serial input/output interface.

The TL16C450 performs serial-to-parallel conversion on data received from a peripheral device or modem and parallel-to-serial conversion on data received from its CPU. The CPU can read and report on the status of the

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# TL16C450 ASYNCHRONOUS COMMUNICATIONS ELEMENT

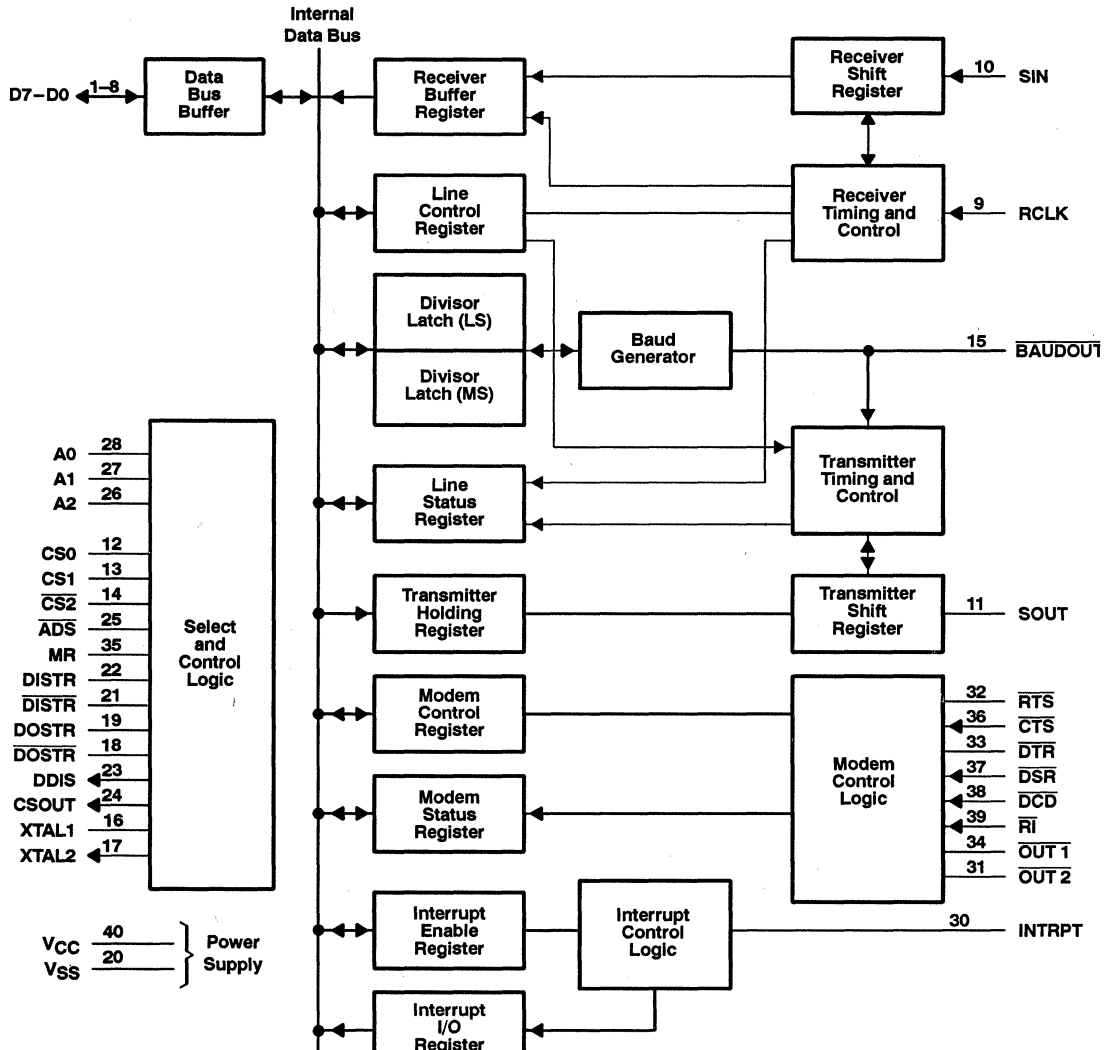
SLLS037A - D3096, MARCH 1988 - REVISED APRIL 1989

## description (continued)

ACE at any point in the ACE's operation. Reported status information includes: the type of transfer operation in progress, the status of the operation, and any error conditions encountered.

The TL16C450 ACE includes a programmable, on-board, baud rate generator. This generator is capable of dividing a reference clock input by divisors from 1 to  $(2^{16} - 1)$  and producing a 16 X clock for driving the internal transmitter logic. Provisions are also included to use this 16 X clock to drive the receiver logic. Also included in the ACE is a complete modem control capability and a processor interrupt system that may be software tailored to the user's requirements to minimize the computing required to handle the communications link.

## block diagram



Pin numbers shown are for the N package.



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## Terminal Functions

PIN		I/O	DESCRIPTION
NAME	NO.†		
A0	28	I	Register select. Three inputs used during read and write operations to select the ACE register to read from or write to. Refer to Table 1 for register addresses, also refer to the address strobe ( $\overline{ADS}$ ) signal description.
A1	27		
A2	26		
$\overline{ADS}$	25	I	Address strobe. When $\overline{ADS}$ is active (low), the register select signals (A0, A1, and A2) and chip select signals ( $\overline{CS0}$ , $\overline{CS1}$ , $\overline{CS2}$ ) drive the internal select logic directly; when high, the register select and chip select signals are held in the state they were in when the low-to-high transition of $\overline{ADS}$ occurred.
$\overline{BAUDOUT}$	15	O	Baud out. 16 X clock signal for the transmitter section of the ACE. The clock rate is established by the reference oscillator frequency divided by a divisor specified by the baud generator divisor latches. $\overline{BAUDOUT}$ may also be used for the receiver section by tying this output to the RCLK input.
$\overline{CS0}$	12	I	Chip select. When active (high and low, respectively), these three inputs select the ACE. Refer to the $\overline{ADS}$ signal description.
$\overline{CS1}$	13		
$\overline{CS2}$	14		
$\overline{CSOUT}$	24	O	Chip select out. When $\overline{CSOUT}$ is high, it indicates that the ACE has been selected by the chip select inputs ( $\overline{CS0}$ , $\overline{CS1}$ , and $\overline{CS2}$ ). $\overline{CSOUT}$ is low when the chip is deselected.
$\overline{CTS}$	36	I	Clear to send. $\overline{CTS}$ is a modem status signal whose condition can be checked by reading bit 4 (CTS) of the modem status register. Bit 0 (DCTS) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when $\overline{CTS}$ changes state, an interrupt is generated.
D0	1	I/O	Data bus. Eight 3-state data lines provide a bidirectional path for data, control, and status information between the ACE and the CPU.
D1	2		
D2	3		
D3	4		
D4	5		
D5	6		
D6	7		
D7	8		
$\overline{DCD}$	38	I	Data carrier detect. $\overline{DCD}$ is a modem status signal whose condition can be checked by reading bit 7 (DCD) of the modem status register. Bit 3 (DDCD) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when the $\overline{DCD}$ changes state, an interrupt is generated.
$\overline{DDIS}$	23	O	Driver disable. This output is active (high) when the CPU is not reading data. When active, this output can be used to disable an external transceiver.
$\overline{DISTR}$	22	I	Data input strobes. When either input is active (high or low, respectively) while the ACE is selected, the CPU is allowed to read status information or data from a selected ACE register. Only one of these inputs is required for the transfer of data during a read operation; the other input should be tied in its inactive state (i.e., $\overline{DISTR}$ tied low or $\overline{DISTR}$ tied high).
$\overline{DISTR}$	21		
$\overline{DOSTR}$	19	I	Data output strobes. When either input is active (high or low, respectively), while the ACE is selected, the CPU is allowed to write control words or data into a selected ACE register. Only one of these inputs is required to transfer data during a write operation; the other input should be tied in its inactive state (i.e., $\overline{DOSTR}$ tied low or $\overline{DOSTR}$ tied high).
$\overline{DOSTR}$	18		
$\overline{DSR}$	37	I	Data set ready. $\overline{DSR}$ is a modem status signal whose condition can be checked by reading bit 5 (DSR) of the modem status register. Bit 1 (DDSR) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when the $\overline{DSR}$ changes state, an interrupt is generated.
$\overline{DTR}$	33	O	Data terminal ready. When active (low), $\overline{DTR}$ informs a modem or data set that the ACE is ready to establish communication. $\overline{DTR}$ is placed in the active state by setting the DTR bit of the modem control register to a high level. $\overline{DTR}$ is placed in the inactive state either as a result of a master reset or during loop mode operation or resetting bit 0 (DTR) of the modem control register.

† Pin numbers shown are for the N package.



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## Terminal Functions (continued)

PIN		I/O	DESCRIPTION
NAME	NO.†		
INTRPT	30	O	Interrupt. When active (high), INTRPT informs the CPU that the ACE has an interrupt to be serviced. Four conditions that cause an interrupt to be issued are: a receiver error, received data is available, the transmitter holding register is empty, and an enabled modem status interrupt. The INTRPT output is reset (inactivated) either when the interrupt is serviced or as a result of a master reset.
MR	35	I	Master reset. When active (high), MR clears most ACE registers and sets the state of various output signals. Refer to Table 2.
OUT1 OUT2	34 31	O	Outputs 1 and 2. User-designated output pins that are set to their active states by setting their respective modem control register bits (OUT 1 and OUT 2) high. OUT 1 and OUT 2 are set to their inactive (high) states as a result of master reset or during loop mode operations or by resetting bit 2 (OUT 1) or bit 3 (OUT 2) of the MCR.
RCLK	9	I	Receiver clock. The 16 X baud rate clock for the receiver section of the ACE.
RI	39	I	Ring indicator. RI is a modem status signal whose condition can be checked by reading bit 6 (RI) of the modem status register. Bit 2 (TERI) of the modem status register indicates that the RI input has transitioned from a low to a high state since the last read from the modem status register. If the modem status interrupt is enabled when this transition occurs, an interrupt is generated.
RTS	32	O	Request to send. When active, informs the modem or data set that the ACE is ready to transmit data. RTS is set to its active state by setting the RTS modem control register bit and is set to its inactive (high) state either as a result of a master reset or during loop mode operations or by resetting bit 1 (RTS) of the MCR.
SIN	10	I	Serial input. Serial data input from a connected communications device.
SOUT	11	O	Serial output. Composite serial data output to a connected communication device. SOUT is set to the marking (logic 1) state as a result of MR.
VCC	40		5-V supply voltage
VSS	20		Supply common
XTAL1 XTAL2	16 17	I/O	External clock. Connects the ACE to the main timing reference (clock or crystal).

† Pin numbers shown are for the N package.

### absolute maximum ratings over free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	–0.5 V to 7 V
Input voltage range at any input, $V_I$	–0.5 V to 7 V
Output voltage range, $V_O$	–0.5 V to 7 V
Continuous total dissipation at (or below) 70°C free-air temperature: FN package	1100 mW
N package	800 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	–65°C to 150°C
Case temperature for 10 seconds: FN package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: N package	260°C

NOTE 1: All voltage values are with respect to  $V_{SS}$ .

### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2		$V_{CC}$	V
Low-level input voltage, $V_{IL}$	–0.5		0.8	V
Operating free-air temperature, $T_A$	0		70	°C

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**electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V <sub>OH</sub> ‡	High-level output voltage	I <sub>OH</sub> = -1 mA		2.4			V
V <sub>OL</sub> ‡	Low-level output voltage	I <sub>OL</sub> = 1.6 mA				0.4	V
I <sub>IKG</sub>	Input leakage current	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 0 to 5.25 V,	V <sub>SS</sub> = 0, All other pins floating			±10	µA
I <sub>OZ</sub>	High-impedance output current	V <sub>CC</sub> = 5.25 V, V <sub>O</sub> = 0 V to 5.25 V,	V <sub>SS</sub> = 0, Chip selected, write mode, or chip deselected			±20	µA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = 5.25 V, SIN, DSR, DCD, CTS, and All other inputs at 0.8 V, No load on outputs,	T <sub>A</sub> = 25°C, RI at 2 V, XTAL1 at 4 MHz, Baud rate = 50 kb/s			10	mA
C <sub>XTAL1</sub>	Clock input capacitance	V <sub>CC</sub> = 0, f = 1 MHz, All other pins grounded			15	20	pF
C <sub>XTAL2</sub>	Clock output capacitance				20	30	pF
C <sub>i</sub>	Input capacitance				6	10	pF
C <sub>o</sub>	Output capacitance				10	20	pF

† All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

‡ These parameters apply for all outputs except XTAL2.

**system timing requirements over recommended ranges of supply voltage and operating free-air temperature**

PARAMETER		FIGURE	MIN	MAX	UNIT
t <sub>cR</sub>	Cycle time, read (t <sub>w7</sub> + t <sub>d8</sub> + t <sub>d9</sub> )		175		ns
t <sub>cW</sub>	Cycle time, write (t <sub>w6</sub> + t <sub>d5</sub> + t <sub>d6</sub> )		175		ns
t <sub>w5</sub>	Pulse duration, address strobe low	2,3	15		ns
t <sub>w6</sub>	Pulse duration, write strobe	2	80		ns
t <sub>w7</sub>	Pulse duration, read strobe	3	80		ns
t <sub>wMR</sub>	Pulse duration, master reset		1000		ns
t <sub>su1</sub>	Setup time, address	2,3	15		ns
t <sub>su2</sub>	Setup time, chip select	2,3	15		ns
t <sub>su3</sub>	Setup time, data	2	15		ns
t <sub>h1</sub>	Hold time, address	2,3	0		ns
t <sub>h2</sub>	Hold time, chip select	2,3	0		ns
t <sub>h3</sub>	Hold time, write to chip select	2	20		ns
t <sub>h4</sub>	Hold time, write to address	2	20		ns
t <sub>h5</sub>	Hold time, data	2	15		ns
t <sub>h6</sub>	Hold time, read to chip select	3	20		ns
t <sub>h7</sub>	Hold time, read to address	3	20		ns
t <sub>d4</sub> §	Delay time, select to write	2	15		ns
t <sub>d5</sub> §	Delay time, address to write	2	15		ns
t <sub>d6</sub>	Delay time, write cycle	2	80		ns
t <sub>d7</sub> §	Delay time, chip select to read	3	15		ns
t <sub>d8</sub> §	Delay time, address to read	3	15		ns
t <sub>d9</sub>	Delay time, read cycle	3	80		ns

§ Only applies when ADS is low.



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### system switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{w1}$	Pulse duration, clock high	1	f = 9 MHz maximum	50		ns
$t_{w2}$	Pulse duration, clock low	1	f = 9 MHz maximum	50		ns
$t_{d3}$	Delay time, select to CS output	2,3	$C_L = 100$ pF		70	ns
$t_{d10}$	Delay time, read to data	3	$C_L = 100$ pF		60	ns
$t_{d11}$	Delay time, read to floating data	3	$C_L = 100$ pF	0	60	ns
$t_{dis(F)}$	Read to driver disable	3	$C_L = 100$ pF		60	ns

### baud generator switching requirements over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{w3}$	Pulse duration, $\overline{\text{BAUDOUT}}$ low	1	f = 6.25 MHz, CLK + 1, $C_L = 100$ pF	80		ns
$t_{w4}$	Pulse duration, $\overline{\text{BAUDOUT}}$ high	1	f = 6.25 MHz, CLK + 1, $C_L = 100$ pF	80		ns
$t_{d1}$	Delay time, $\overline{\text{BAUDOUT}}$ low to high	1	$C_L = 100$ pF		125	ns
$t_{d2}$	Delay time, $\overline{\text{BAUDOUT}}$ high to low	1	$C_L = 100$ pF		125	ns

### receiver switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{d12}$	Delay time, RCLK to sample	4			100	ns
$t_{d13}$	Delay time, stop to set interrupt	4		1	1	RCLK cycles
$t_{d14}$	Delay time, read RBR/LSR to reset interrupt	4	$C_L = 100$ pF		140	ns

### transmitter switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{d15}$	Delay time, initial write THR to transmit start	5		8	24	baudout cycles
$t_{d16}$	Delay time, stop to interrupt	5		8	8	baudout cycles
$t_{d17}$	Delay time, write THR to reset interrupt	5	$C_L = 100$ pF		140	ns
$t_{d18}$	Delay time, initial write to interrupt (THRE)	5		16	32	baudout cycles
$t_{d19}$	Delay time, read IIR to reset interrupt (THRE)	5	$C_L = 100$ pF		140	ns

### modem control switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{d20}$	Delay time, write MCR to output	6	$C_L = 100$ pF		100	ns
$t_{d21}$	Delay time, modem input to set interrupt	6	$C_L = 100$ pF		170	ns
$t_{d22}$	Delay time, read MSR to reset interrupt	6	$C_L = 100$ pF		140	ns

PARAMETER MEASUREMENT INFORMATION

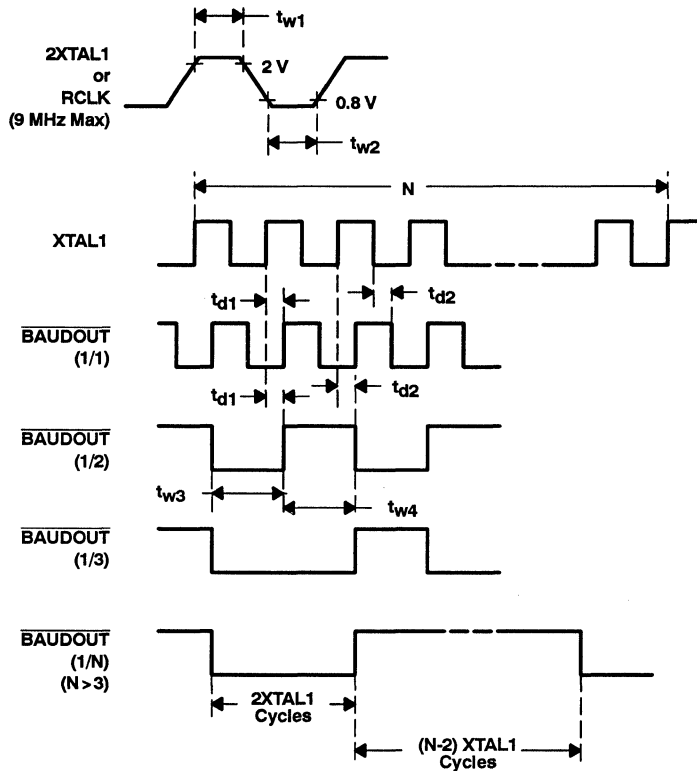


Figure 1. Baud Generator Timing

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## PARAMETER MEASUREMENT INFORMATION

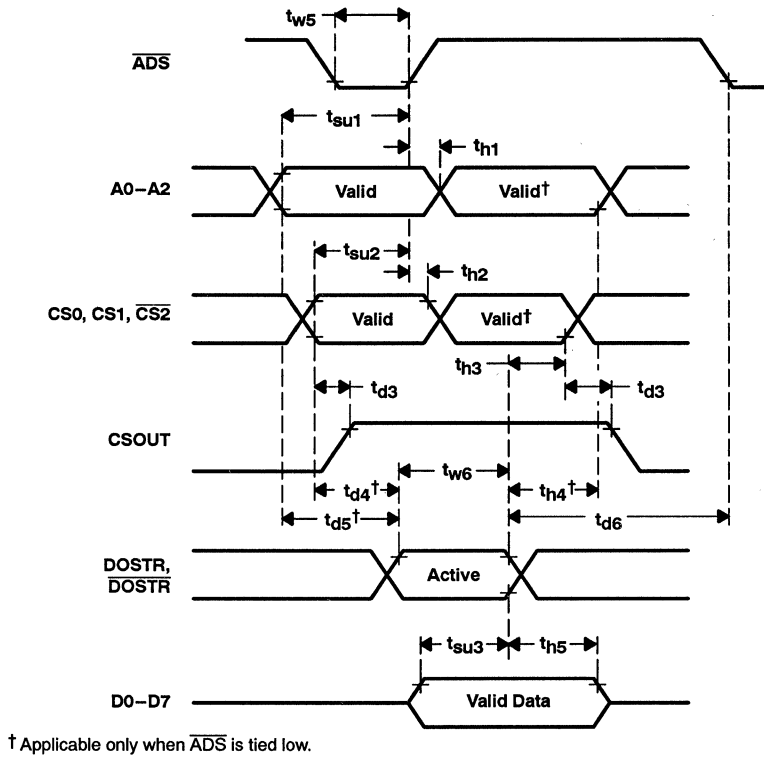
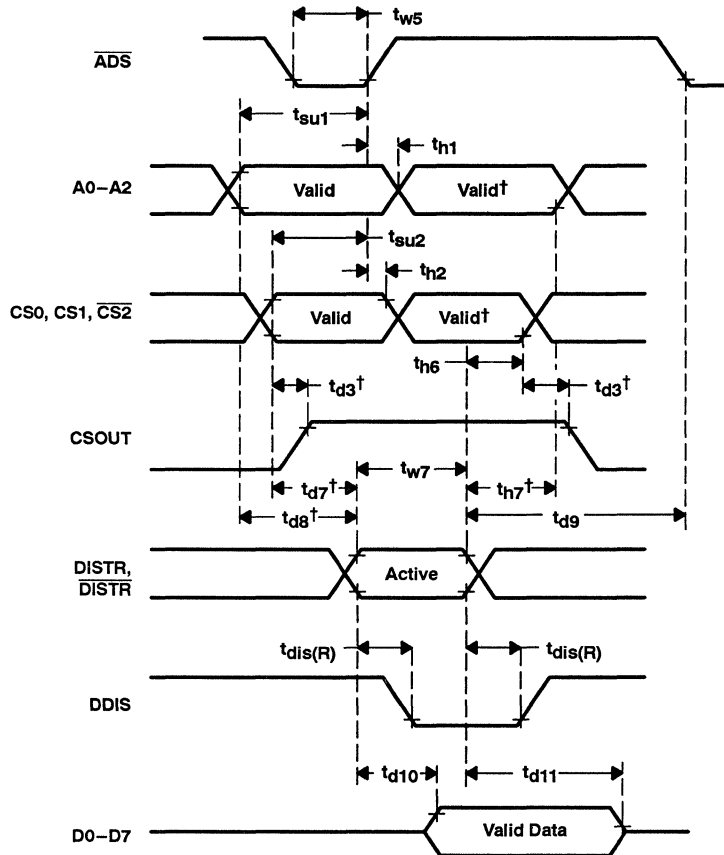


Figure 2. Write Cycle Timing

PARAMETER MEASUREMENT INFORMATION



† Applicable only when  $\overline{ADS}$  is tied low.

Figure 3. Read Cycle Timing



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## PARAMETER MEASUREMENT INFORMATION

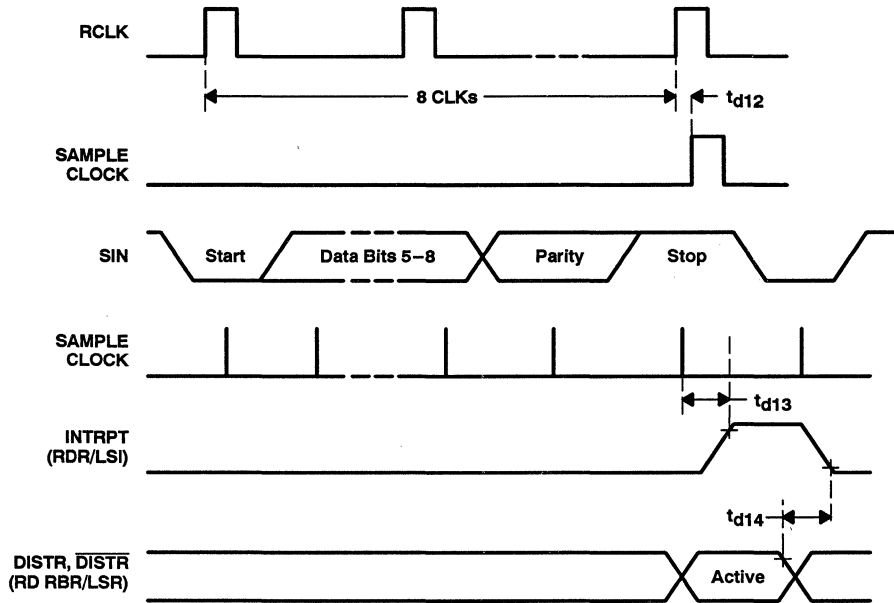


Figure 4. Receiver Timing

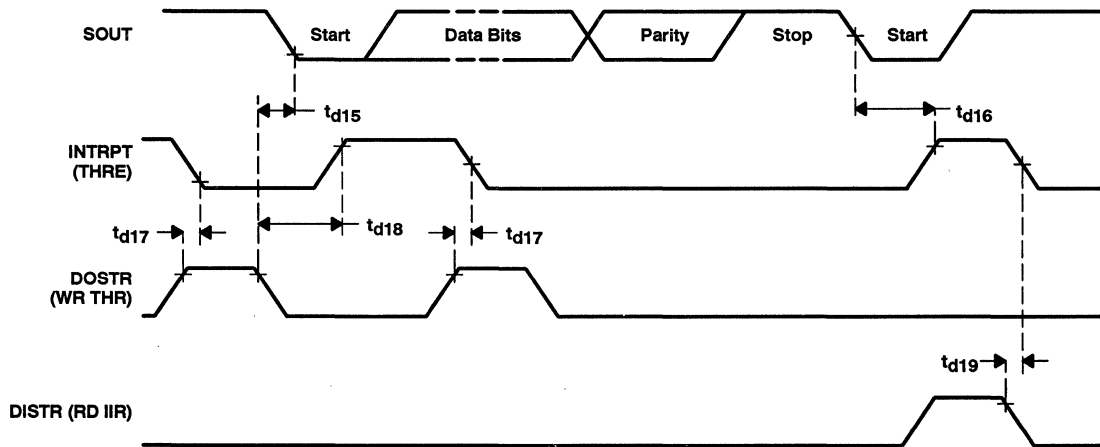


Figure 5. Transmitter Timing

PARAMETER MEASUREMENT INFORMATION

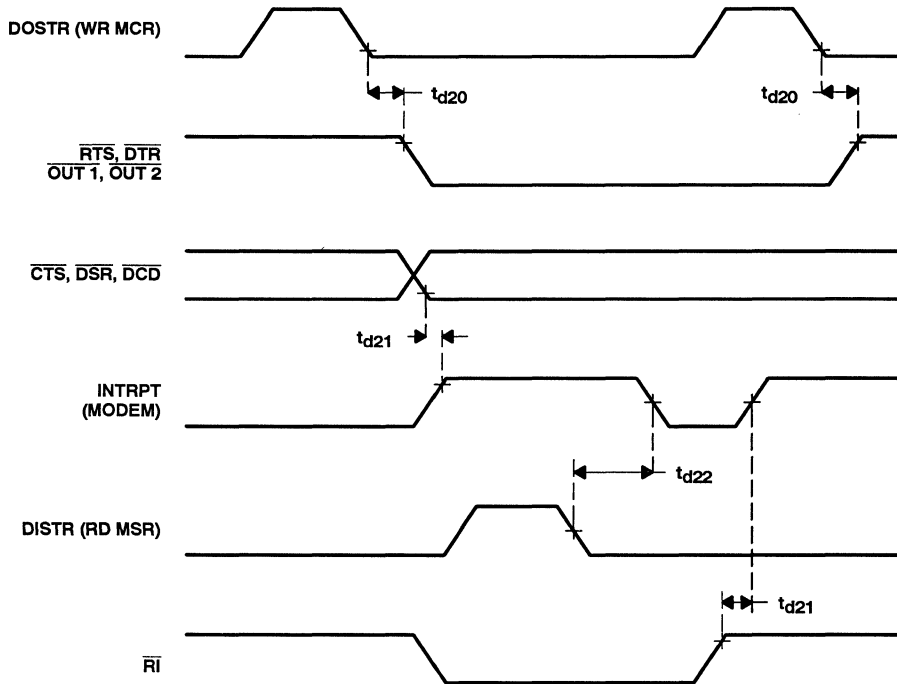


Figure 6. Modem Control Timing

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## APPLICATION INFORMATION

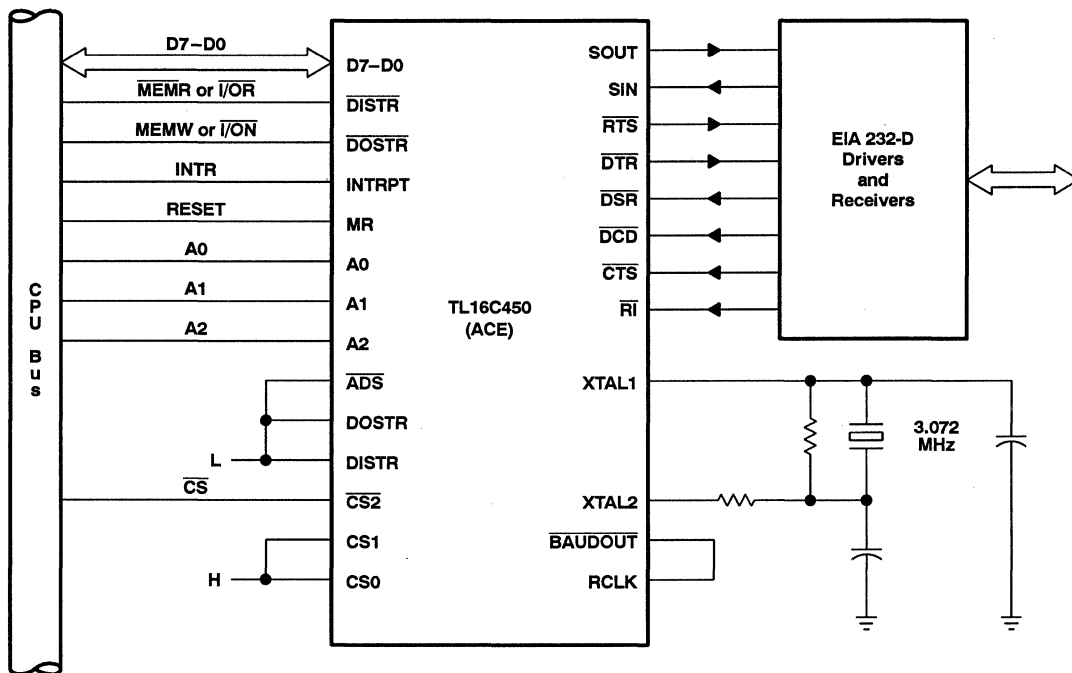


Figure 7. Basic TL16C450 Configuration

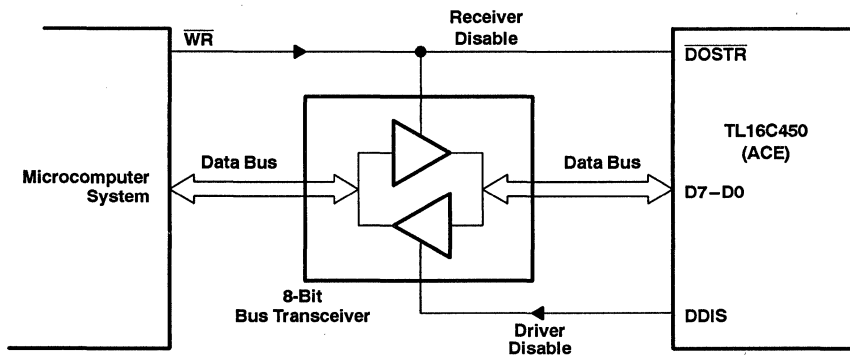


Figure 8. Typical Interface for a High-Capacity Data Bus

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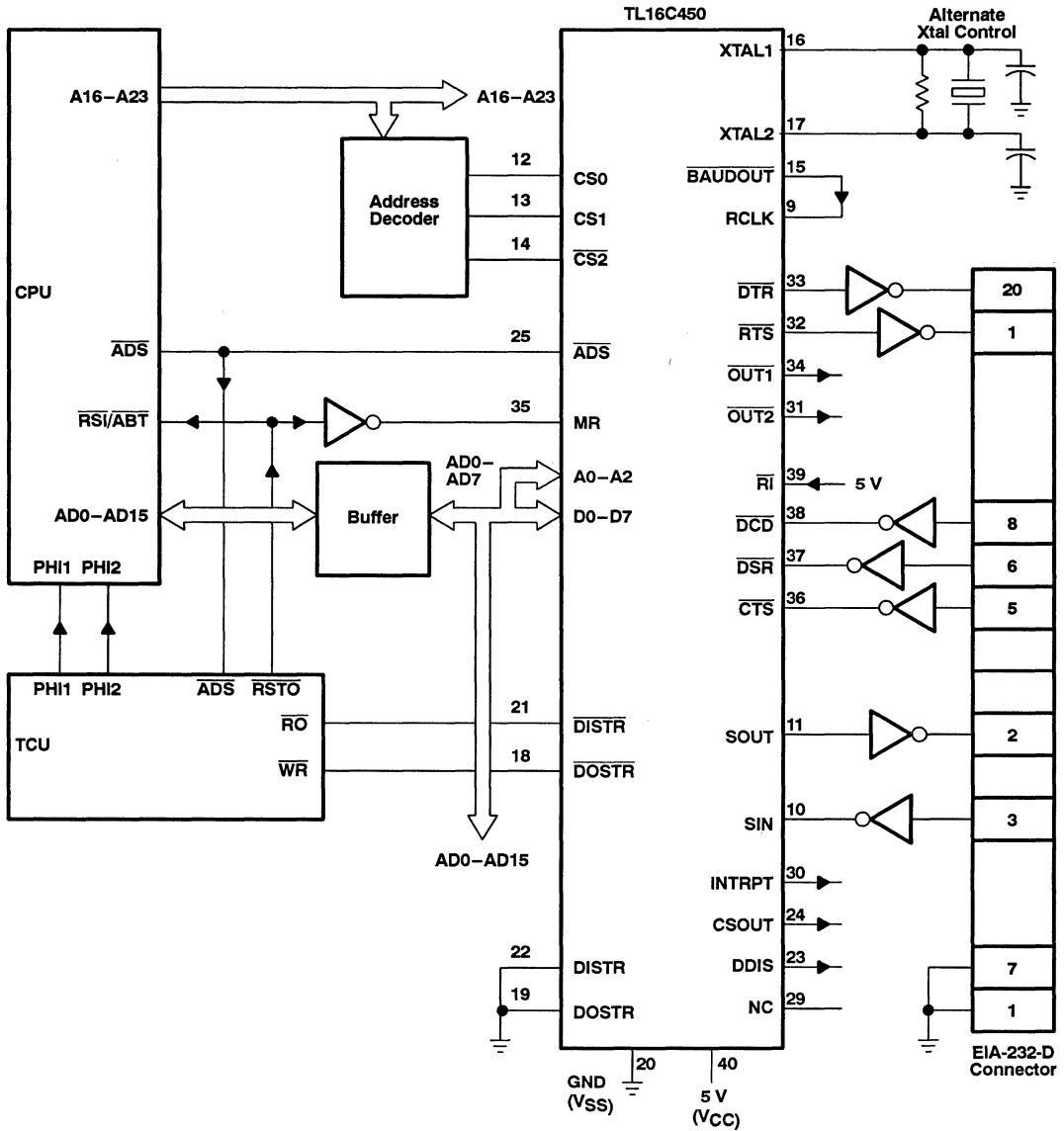


Figure 9. Typical TL16C450 Connection to a CPU

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## PRINCIPLES OF OPERATION

Table 1. Register Selection

DLAB†	A2	A1	A0	REGISTER
0	L	L	L	Receiver buffer (read), transmitter holding register (write)
0	L	L	H	Interrupt enable
X	L	H	L	Interrupt identification (read only)
X	L	H	H	Line control
X	H	L	L	Modem control
X	H	L	H	Line status
X	H	H	L	Modem status
X	H	H	H	Scratch
1	L	L	L	Divisor latch (LSB)
1	L	L	H	Divisor latch (MSB)

† The divisor latch access bit (DLAB) is the most significant bit of the line control register. The DLAB signal is controlled by writing to this bit location (see Table 3).

Table 2. ACE Reset Functions

REGISTER/SIGNAL	RESET CONTROL	RESET STATE
Interrupt Enable Register	Master Reset	All bits low (0-3 forced and 4-7 permanent)
Interrupt Identification Register	Master Reset	Bit 0 is high, bits 1 and 2 are low, and bits 3-7 are permanently low
Line Control Register		All bits low
Modem Control Register	Master Reset	All bits low
Line Status Register	Master Reset	Bits 5 and 6 are high, all other bits are low
Modem Status Register	Master Reset	Bits 0-3 are low, bits 4-7 are input signals
SOUT	Master Reset	High
INTRPT (receiver error flag)	Read LSR/MR	Low
INTRPT (received data available)	Read RBR/MR	Low
INTRPT (transmitter holding register empty)	Read IIR/Write THR/MR	Low
INTRPT (modem status changes)	Read MSR/MR	Low
OUT 2	Master Reset	High
RTS	Master Reset	High
DTR	Master Reset	High
OUT 1	Master Reset	High
Scratch Register	Master Reset	No effect
Divisor Latch (LSB and MSB) Register	Master Reset	No effect
Receiver Buffer Register	Master Reset	No effect
Transmitter Holding Register	Master Reset	No effect

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**PRINCIPLES OF OPERATION**

**accessible registers**

The system programmer, via the CPU, has access to and control over any of the ACE registers that are summarized in Table 3. These registers are used to control ACE operations, receive data, and transmit data. Descriptions of these registers follow Table 3.

**Table 3. Summary of Accessible Registers**

Bit No.	REGISTER ADDRESS										
	O DLAB = 0	O DLAB = 0	1 DLAB = 0	2	3	4	5	6	7	O DLAB = 1	1 DLAB = 0
	Receiver Buffer Register (Read Only)	Transmitter Holding Register (Write Only)	Interrupt Enable Register IER	Interrupt Ident. Register (Read Only)	Line Control Register LCR	Modem Control Register	Line Status Register	Modem Status Register	Scratch Register	Divisor Latch (LSB)	Latch (MSB)
	RBR	THR	IER	IIR	LCR	MCR	LSR	MSR	SCR	DLL	DLM
0	Data Bit 0*	Data Bit 0	Enable Received Data Available Interrupt (ERBF)	"0" if Interrupt Pending	Word Length Select Bit 0 (WLS0)	Data Terminal Ready (DTR)	Data Ready (DR)	Delta Clear to Send (DCTS)	Bit 0	Bit 0	Bit 8
1	Data Bit 1	Data Bit 1	Enable Transmitter Holding Register Empty Interrupt (ETBE)	Interrupt ID Bit (0)	Word Length Select Bit 1 (WLS1)	Request to Send (RTS)	Overrun Error (OE)	Delta Data Set Ready (DDSR)	Bit 1	Bit 1	Bit 9
2	Data Bit 2	Data Bit 2	Enable Receiver Line Status Interrupt (ELSI)	Interrupt ID Bit (1)	Number of Stop Bits (STB)	Out 1	Parity Error (PE)	Trailing Edge Ring Indicator (TERI)	Bit 2	Bit 2	Bit 10
3	Data Bit 3	Data Bit 3	Enable Modem Status Interrupt (EDSSI)	0	Parity Enable (PEN)	Out 2	Framing Error (FE)	Delta Data Carrier Detect (DDCD)	Bit 3	Bit 3	Bit 11
4	Data Bit 4	Data Bit 4	0	0	Even Parity Select (EPS)	Loop	Break Interrupt (BI)	Clear to Send (CTS)	Bit 4	Bit 4	Bit 12
5	Data Bit 5	Data Bit 5	0	0	Stick Parity	0	Transmitter Holding Register (THRE)	Data Set Ready (DSR)	Bit 5	Bit 5	Bit 13
6	Data Bit 6	Data Bit 6	0	0	Set Break	0	Transmitter Empty (TEMT)	Ring Indicator (RI)	Bit 6	Bit 6	Bit 14
7	Data Bit 7	Data Bit 7	0	0	Divisor Latch Access Bit (DLAB)	0	0	Data Carrier Detect (DCD)	Bit 7	Bit 7	Bit 15

\*Bit 0 is the least significant bit. It is the first bit serially transmitted or received.

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## PRINCIPLES OF OPERATION

### receiver buffer register (RBR)

The ACE receiver section consists of a receiver shift register and a receiver buffer register. Timing is supplied by the 16 X receiver clock (RCLK). Receiver section control is a function of the ACE line control register.

The ACE receiver shift register receives serial data from the serial input (SIN) pin. The receiver shift register then converts the data to a parallel form and loads it into the receiver buffer register. When a character is placed in the receiver buffer register and the received data available interrupt is enabled, an interrupt is generated. This interrupt is cleared when the data is read out of the receiver buffer register.

### transmitter holding register (THR)

The ACE transmitter section consists of a transmitter holding register and a transmitter shift register. Timing is supplied by the Baud Out (BAUDOUT) clock signal. Transmitter section control is a function of the ACE line control register.

The ACE transmitter holding register receives data off the internal data bus and, when the shift register is idle, moves it into the transmitter shift register. The transmitter shift register serializes the data and outputs it at the serial output (SOUT). If the transmitter holding register is empty and the transmitter holding register empty (THRE) interrupt is enabled, an interrupt is generated. This interrupt is cleared when a character is loaded into the register.

### interrupt enable register (IER)

The interrupt enable register enables each of the four types of interrupts (refer to Table 4) and the INTRPT output signal in response to an interrupt generation. The interrupt enable register can also be used to disable the interrupt system by setting bits 0 through 3 to logic 0. The contents of this register are summarized in Table 3 and are described below.

Bit 0. This bit, when set to logic 1, enables the received data available interrupt.

Bit 1. This bit, when set to logic 1, enables the transmitter holding register empty interrupt.

Bit 2. This bit, when set to logic 1, enables the receiver line status interrupt.

Bit 3. This bit, when set to logic 1, enables the modem status interrupt.

Bits 4 thru 7. Bits 4 through 7 in the interrupt enable register are not used and are always set to logic 0.



**PRINCIPLES OF OPERATION**

**interrupt identification register (IIR)**

The ACE has an on-chip interrupt generation and prioritization capability that permits a flexible interface with most microprocessors.

The ACE provides four prioritized levels of interrupts:

- Priority 1 – Receiver line status (highest priority)
- Priority 2 – Receiver data ready
- Priority 3 – Transmitter holding register empty
- Priority 4 – Modem status (lowest priority)

When an interrupt is generated, the interrupt identification register indicates that an interrupt is pending and the type of that interrupt in its three least significant bits (bits 0, 1, and 2). The contents of this register are summarized in Table 3 and described in Table 4.

Bit 0. This bit can be used either in a hardwire prioritized or polled interrupt system. If this bit is a logic 0, an interrupt is pending. When bit 0 is a logic 1, no interrupt is pending.

Bits 1 and 2. These two bits are used to identify the highest priority interrupt pending as indicated in Table 4.

Bits 3 thru 7. Bits 3 through 7 in the interrupt identification register are not used and are always set at logic 0.

**Table 4. Interrupt Control Functions**

INTERRUPT IDENTIFICATION REGISTER			PRIORITY LEVEL	INTERRUPT TYPE	INTERRUPT SOURCE	INTERRUPT RESET METHOD
BIT 2	BIT 1	BIT 0				
0	0	1	None	None	None	–
1	1	0	1	Receiver line status	Overrun error, parity error, framing error or break interrupt	Reading the line status register
1	0	0	2	Received data available	Receiver data available	Reading the receiver buffer Buffer register
0	1	0	3	Transmitter holding register empty	Transmitter holding register empty	Reading the interrupt identification register (if source of interrupt) or writing into the transmitter holding register
0	0	0	4	Modem status	Clear to send, data set ready, ring indicator, or data carrier detect	Reading the modem status register



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### PRINCIPLES OF OPERATION

#### line control register (LCR)

The system programmer controls the format of the asynchronous data communication exchange through the line control register. In addition, the programmer is able to retrieve, inspect, and modify the contents of the line control register; this eliminates the need for separate storage of the line characteristics in system memory. The contents of this register are summarized in Table 3 and are described below.

Bits 0 and 1. These two bits specify the number of bits in each transmitted or received serial character. These bits are encoded as follows:

Bit 1	Bit 0	Word Length
0	0	5 Bits
0	1	6 Bits
1	0	7 Bits
1	1	8 Bits

Bit 2. This bit specifies either one, one and one-half, or two stop bits in each transmitted character. If bit 2 is a logic 0, one stop bit is generated in the data. If bit 2 is a logic 1, the number of stop bits generated is dependent on the word length selected with bits 0 and 1. The receiver checks the first stop bit only, regardless of the number of stop bits selected. The number of stop bits generated, in relation to word length and bit 2, is shown in the following.

Bit 2	Word Length Selected by Bits 1 and 2	Number of Stop Bits Generated
0	Any word length	1
1	5 bits	1 1/2
1	6 bits	2
1	7 bits	2
1	8 bits	2

Bit 3. This bit is the parity enable bit. When bit 3 is a logic 1, a parity bit is generated in transmitted data between the last data word bit and the first stop bit. In received data, if bit 3 is a logic 1, parity is checked. When bit 3 is a logic 0, no parity is generated or checked.

Bit 4. Bit 4 is the even parity select bit. When parity is enabled by bit 3: a logic 1 in bit 4 produces even parity (an even number of logic 1s in the data and parity bits) and a logic 0 in bit 4 produces odd parity (an odd number of logic 1s).

Bit 5. This is the stick parity bit. When bits 3, 4, and 5 are logic 1s, the parity bit is transmitted and checked as a logic 0. When bits 3 and 5 are logic 1s and bit 4 is a logic 0, the parity bit is transmitted and checked as a logic 1.

Bit 6. This bit is the break control bit. Bit 6 is set to a logic 1 to force a break condition, i.e., a condition where the serial output (SOUT) pin is forced to the spacing (logic 0) state. When bit 6 is set to a logic 0, the break condition is disabled. The break condition has no effect on the transmitter logic, it only effects the serial output.

Bit 7. This bit is the divisor latch access bit (DLAB). Bit 7 must be set to a logic 1 to access the divisor latches of the baud generator during a read or write. Bit 7 must be set to a logic 0 during a read or write to access the receiver buffer, the transmitter holding register, or the interrupt enable register.

## PRINCIPLES OF OPERATION

### modem control register (MCR)

The modem control register is an 8-bit register that controls an interface with a modem, data set, or peripheral device that is emulating a modem. The contents of this register are summarized in Table 3 and are described below.

**Bit 0.** Bit 0 (DTR) controls the data terminal ready ( $\overline{\text{DTR}}$ ) output. Setting this bit to a logic 1 forces the  $\overline{\text{DTR}}$  output to its active state (low). When bit 0 is set to a logic 0,  $\overline{\text{DTR}}$  goes high.

**Bit 1.** Bit 1 (RTS) controls the request to send ( $\overline{\text{RTS}}$ ) output in a manner identical to Bit 0's control over the  $\overline{\text{DTR}}$  output.

**Bit 2.** Bit 2 (OUT 1) controls the output 1 ( $\overline{\text{OUT 1}}$ ) signal, a user designated output signal, in a manner identical to Bit 0's control over the  $\overline{\text{DTR}}$  output.

**Bit 3.** Bit 3 (OUT 2) controls the output 2 ( $\overline{\text{OUT 2}}$ ) signal, a user designated output signal, in a manner identical to Bit 0's control over the  $\overline{\text{DTR}}$  output.

**Bit 4.** Bit 4 provides a local loopback feature for diagnostic testing of the ACE. When this bit is set to a logic high, the following occurs:

1. The transmitter serial output (SOUT) is set high.
2. The receiver serial input (SIN) is disconnected.
3. The output of the transmitter shift register is looped back into the receiver shift register input.
4. The four modem control inputs ( $\overline{\text{CTS}}$ ,  $\overline{\text{DSR}}$ ,  $\overline{\text{DCD}}$ , and  $\overline{\text{RI}}$ ) are disconnected.
5. The four modem control outputs (DTR, RTS,  $\overline{\text{OUT 1}}$ , and  $\overline{\text{OUT 2}}$ ) are internally connected to the four modem control inputs.
6. The four modem control output pins are forced to their inactive states (high).

In the diagnostic mode, data that is transmitted is immediately received. This allows the processor to verify the transmit- and receive-data paths to the ACE. The receiver and transmitter interrupts are fully operational. The modem control interrupts are also operational but the modem control interrupt sources are now the lower four bits of the modem control register instead of the four modem control inputs. All interrupts are still controlled by the interrupt enable register.

Bit 5 through 7. These bits are set to logic 0.

### line status register (LSR)<sup>†</sup>

The line status register provides information to the CPU concerning the status of data transfers. The contents of this register are summarized in Table 3 and are described below.

**Bit 0.** Bit 0 is the data ready (DR) indicator for the receiver. This bit is set to a logic 1 condition whenever a complete incoming character has been received and transferred into the receiver buffer register and is reset to logic 0 by reading the receiver buffer register.

**Bit 1<sup>‡</sup>.** Bit 1 is the overrun error (OE) indicator. When this bit is set to logic 1, it indicates that before the character in the receiver buffer register was read, it was overwritten by the next character transferred into the register. The OE indicator is reset every time the CPU reads the contents of the line status register.

**Bit 2<sup>‡</sup>.** Bit 2 is the parity error (PE) indicator. When this bit is set to logic 1, it indicates that the parity of the received data character does not match the parity selected in the line control register (bit 4). The PE bit is reset every time the CPU reads the contents of the line status register.

**Bit 3<sup>‡</sup>.** Bit 3 is the framing error (FE) indicator. When this bit is set to logic 1, it indicates that the received character did not have a valid (logic 1) stop bit. The FE bit is reset every time the CPU reads the contents of the line status register.

<sup>†</sup> The line status register is intended for read operations only; writing to this register is not recommended outside of a factory testing environment.

<sup>‡</sup> Bits 1 through 4 are the error conditions that produce a receiver line status interrupt.

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### PRINCIPLES OF OPERATION

#### line status register (LSR)<sup>†</sup> (continued)

Bit 4<sup>‡</sup>. Bit 4 is the break interrupt (BI) indicator. When this bit is set to logic 1, it indicates that the received data input was held in the logic 0 state for longer than a full-word transmission time. A full-word transmission time is defined as the total time of the start, data, parity, and stop bits. The BI bit is reset every time the CPU reads the contents of the line status register.

Bit 5. Bit 5 is the transmitter holding register empty (THRE) indicator. This bit is set to a logic 1 condition when the transmitter holding register is empty, indicating that the ACE is ready to accept a new character. If the THRE interrupt is enabled when the THRE bit is a logic 1, then an interrupt is generated. THRE is set to a logic 1 when the contents of the transmitter holding register are transferred to the transmitted shift register. This bit is reset to logic 0 concurrent with the loading of the transmitter holding register by the CPU.

Bit 6. Bit 6 is the transmitter empty (TEMT) indicator. This bit is set to a logic 1 when the transmitter holding register and the transmitter shift register are both empty. When either the transmitter holding register or the transmitter shift register contains a data character, the TEMT bit is reset to logic 0.

Bit 7. This bit is always reset to logic 0.

<sup>†</sup> The line status register is intended for read operations only; writing to this register is not recommended outside of a factory testing environment.

<sup>‡</sup> Bits 1 through 4 are the error conditions that produce a receiver line status interrupt.

#### modem status register (MSR)

The modem status register is an 8-bit register that provides information about the current state of the control lines from the modem, data set, or peripheral device to the CPU. Additionally, four bits of this register provides change information; when a control input from the modem changes state the appropriate bit is set to logic 1. All four bits are reset to logic 0 when the CPU reads the modem status register. The contents of this register are summarized in Table 3 and are described below.

Bit 0. Bit 0 is the delta clear to send (DCTS) indicator. This bit indicates that the  $\overline{\text{CTS}}$  input has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 1. Bit 1 is the delta data set ready (DDSR) indicator. This bit indicates that the  $\overline{\text{DSR}}$  input has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 2. Bit 2 is the trailing edge of ring indicator (TERI) detector. This bit indicates that the  $\overline{\text{RI}}$  input to the chip has changed from a low to a high state. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 3. Bit 3 is the delta data carrier detect (DDCD) indicator. This bit indicates that the  $\overline{\text{DCD}}$  input to the chip has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 4. Bit 4 is the complement of the clear to send ( $\overline{\text{CTS}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 1 (RTS).

Bit 5. Bit 5 is the complement of the data set ready ( $\overline{\text{DSR}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 0 (DTR).

Bit 6. Bit 6 is the complement of the ring indicator ( $\overline{\text{RI}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 2 (OUT 1).

Bit 7. Bit 7 is the complement of the data carrier detect ( $\overline{\text{DCD}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 3 (OUT 2).



## PRINCIPLES OF OPERATION

### scratch register (SCR)

The scratch register is an 8-bit register that is intended for programmer use as a scratchpad, in the sense that it will temporarily hold programmer data without affecting any other ACE operation.

### programmable baud generator

The ACE contains a programmable baud generator that takes a clock input in the range between DC and 9 MHz and divides it by a divisor in the range between 1 and  $2^{16} - 1$ . The output frequency of the baud generator is sixteen times (16 X) the baud rate. The formula for the divisor is:

$$\text{divisor \#} = \text{XTAL1 frequency input} \div (\text{desired baud rate} \times 16)$$

Two 8-bit registers, called divisor latches, are used to store the divisor in a 16-bit binary format. These divisor latches must be loaded during initialization of the ACE in order to ensure desired operation of the baud generator. When either of the divisor latches is loaded, a 16-bit baud counter is also loaded to prevent long counts on initial load.

Tables 5 and 6 illustrate the use of the baud generator with crystal frequencies of 1,8432 MHz and 3.072 MHz, respectively. For baud rates of 38.4 kilobits per second and below, the error obtained is very small. The accuracy of the selected baud rate is dependent on the selected crystal frequency.

Refer to Figure 10 for examples of typical clock circuits.

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**PRINCIPLES OF OPERATION**

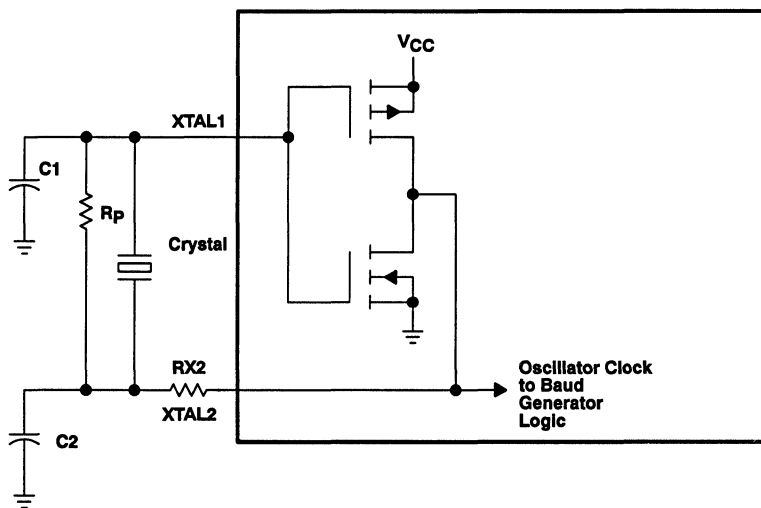
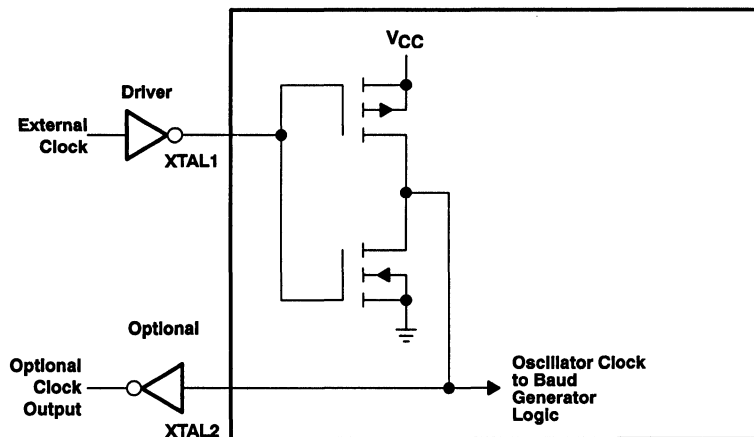
**Table 5. Baud Rates Using a 1.8432-MHz Crystal**

DESIRED BAUD RATE	DIVISOR USED TO GENERATE 16 X CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	2304	
75	1536	
110	1047	0.026
134.5	857	0.058
150	768	
300	384	
600	192	
1200	96	
1800	64	
2000	58	0.69
2400	48	
3600	32	
4800	24	
7200	16	
9600	12	
19200	6	
38400	3	
56000	2	2.86

**Table 6. Baud Rates Using a 3.072-MHz Crystal**

DESIRED BAUD RATE	DIVISOR USED TO GENERATE 16 X CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	3840	
75	2560	
110	1745	0.026
134.5	1428	0.034
150	1280	
300	640	
600	320	
1200	160	
1800	107	0.312
2000	96	
2400	80	
3600	53	0.628
4800	40	
7200	27	1.23
9600	20	
19200	10	
38400	5	

PRINCIPLES OF OPERATION



TYPICAL CRYSTAL OSCILLATOR NETWORK

CRYSTAL	Rp	RX2	C1	C2
3.1 MHz	1 MΩ	1.5 kΩ	10-30 pF	40-60 pF
1.8 MHz	1 MΩ	1.5 kΩ	10-30 pF	40-60 pF

Figure 10. Typical Clock Circuits



# TL16C451, TL16C452 ASYNCHRONOUS COMMUNICATIONS ELEMENTS

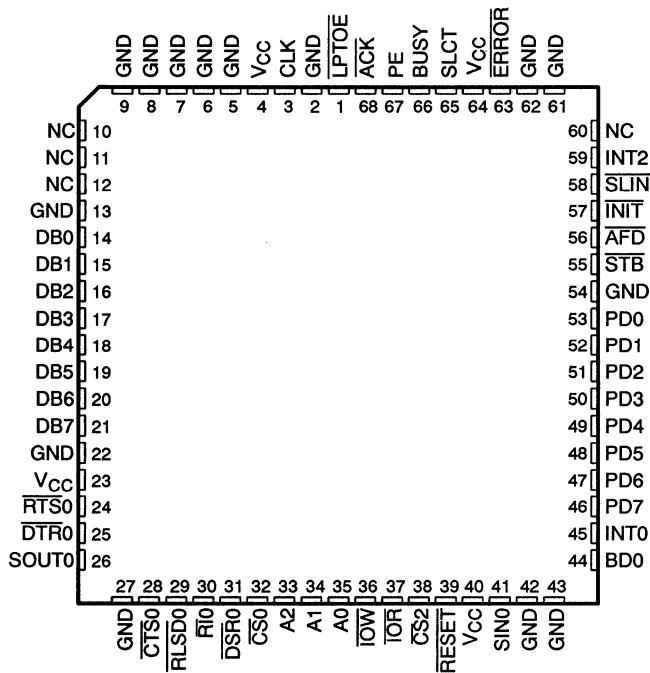
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- Integrates Most Communications Card Functions From the IBM PC/AT or Compatibles With Single- or Dual-Channel Serial Ports
- TL16C451 Consists of One TL16C450 Plus Centronix Printer Interface
- TL16C452 Consists of Two TL16C450s Plus Centronix Printer Interface
- Fully Programmable Serial Interface Characteristics:  
5-, 6-, 7-, or 8-Bit Characters  
Even-, Odd-, or No-Parity Bit  
Generation and Detection  
1-, 1 1/2-, or 2 Stop-Bit Generation  
Programmable Baud Rate  
(dc to 256 kb/s Per Second)
- Fully Double Buffered for Reliable Asynchronous Operation

## description

The TL16C451 and TL16C452 provide single- and dual-channel (respectively) serial interfaces along with a single Centronix parallel-port interface. The serial interfaces provide a serial-to-parallel conversion for data received from a peripheral device or modem and a parallel-to-serial conversion for data transmitted by a computer CPU. The parallel interface provides a bidirectional parallel data port that fully conforms to the requirements for a Centronix-type printer. A computer CPU can read the status of the asynchronous-communications-element (ACE) interfaces at any point in the operation. The status includes the state of the modem signals ( $\overline{CTS}$ ,  $\overline{DSR}$ ,  $\overline{RLSD}$ , and  $\overline{RI}$ ) and any changes to these signals that have occurred since the last time they were read, the state of the transmitter and receiver including errors detected on received data, and printer status. The TL16C451 and TL16C452 provide control for modem signals ( $\overline{RTS}$  and  $\overline{DTR}$ ), interrupt enables, baud-rate programming, and parallel-port control signals.

TL16C451 ... FN PACKAGE  
(TOP VIEW)



NC – No internal connection

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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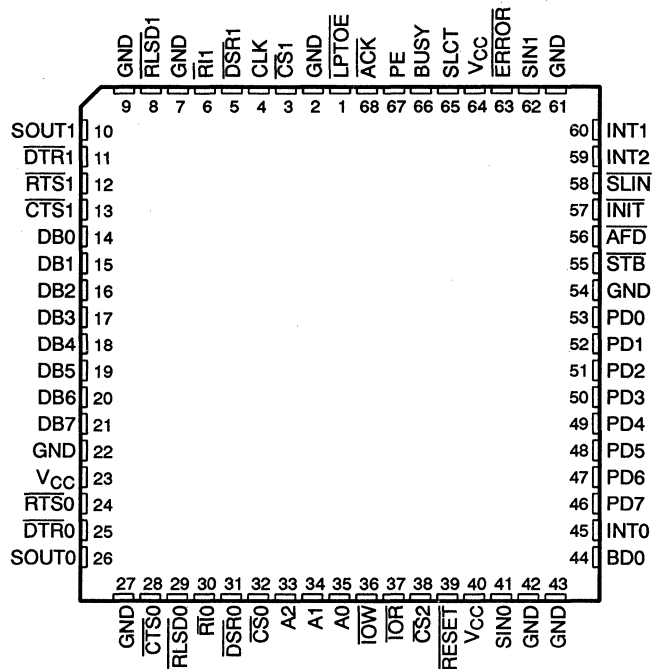
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# TL16C451, TL16C452 ASYNCHRONOUS COMMUNICATIONS ELEMENTS

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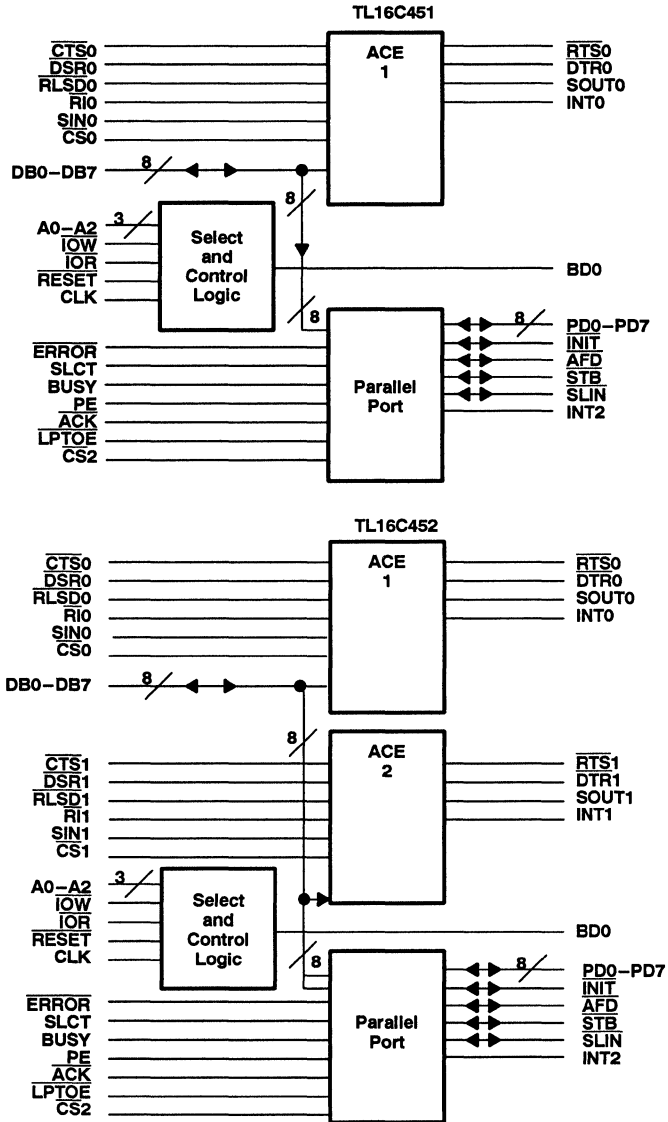
TL16C452 ... FN PACKAGE  
(TOP VIEW)



# TL16C451, TL16C452 ASYNCHRONOUS COMMUNICATIONS ELEMENTS

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## functional block diagrams



**TEXAS**   
**INSTRUMENTS**

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# TL16C451, TL16C452 ASYNCHRONOUS COMMUNICATIONS ELEMENTS

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## Terminal Functions

NAME†	PIN		I/O	DESCRIPTION
		NO.		
A0 A1 A2	35 34 33		I	Register select. Three inputs used during read and write operations to select the register to read from or write to. Refer to Table 1 for register addresses, also refer to the chip select signals ( $\overline{CS0}$ , $\overline{CS1}$ , $\overline{CS2}$ ).
ACK	68		I	Line printer acknowledge. This input goes low to indicate a successful data transfer has taken place. It generates a printer-port interrupt during its positive transition.
AFD	56		I/O	Line printer autofeed. This open-drain line provides the line printer with a low signal when continuous-form paper is to be autofed to the printer. An internal pullup is provided.
BDO	44		O	Bus buffer output. This output is active (high) when the CPU is reading data. When active, this output can be used to disable an external transceiver.
BUSY	66		I	Line printer busy. This is an input line from the line printer that goes high when the line printer is not ready to accept data.
CLK	4		I/O	External clock. Connects the ACE to the main timing reference.
$\overline{CS0}$ $\overline{CS1}$ [V <sub>CC</sub> ] $\overline{CS2}$	32 3 38		I	Chip selects. Each chip select enables read and write operations to its respective channel. $\overline{CS0}$ and $\overline{CS1}$ select serial channels 0 and 1, respectively, and $\overline{CS2}$ selects the parallel port.
CTS0 CTS1 [GND]	28 13		I	Clear to send. $\overline{CTS}$ is an active-low modem status signal whose state can be checked by reading bit 4 (CTS) of the modem status register. Bit 0 (DCTS) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when $\overline{CTS}$ changes state, an interrupt is generated.
DB0 DB1 DB2 DB3 DB4 DB5 DB6 DB7	14 15 16 17 18 19 20 21		I/O	Data bus. Eight 3-state data lines provide a bidirectional path for data, control, and status information between the TL16C451/TL16C452 and the CPU. DBO is the least significant bit (LSB).
DSR0 DSR1 [GND]	31 5		I	Data set ready. DSR is an active-low modem status signal whose state can be checked by reading bit 5 (DSR) of the modem status register. Bit 1 (DDSR) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when the DSR changes state, an interrupt is generated.
DTR0 DTR1 [NC]	25 11		O	Data terminal ready. When active (low), DTR informs a modem or data set that the ACE is ready to establish communication. DTR is placed in the active state by setting the DTR bit of the modem control register to a high level. DTR is placed in the inactive state either as a result of a reset or during loop mode operation or resetting bit 0 (DTR) of the modem control register.
ERROR	63		I	Line printer error. This is an input line from the line printer. The line printer reports an error by holding this line low during the error condition.
INIT	57		I/O	Line printer initialize. This open-drain line provides the line printer with a signal that allows the line printer initialization routine to be started. An internal pullup is provided.
INT0 INT1 [NC]	45 60		O	Interrupt. INTn is an active-high 3-state output that is enabled by bit 3 of the MCR. When active, INTn informs the CPU that the ACE has an interrupt to be serviced. Four conditions that cause an interrupt to be issued are: a receiver error, received data is available, the transmitter holding register is empty, and an enabled modem status interrupt. The INTn output is reset (low) either when the interrupt is serviced or as a result of a reset.
INT2	59		O	Printer port interrupt. This signal is an active-high 3-state output generated by the positive transition of ACK. It is enabled by bit 4 of the write control register.
IOR	37		I	Data read strobe. When IOR input is active (low) while the ACE is selected, the CPU is allowed to read status information or data from a selected ACE register.
IOW	66		I	Data write strobe. When IOW input is active (low) while the ACE is selected, the CPU is allowed to write control words or data into a selected ACE register.

† Names shown in brackets are for the TL16C451.

  
**TEXAS  
INSTRUMENTS**

# TL16C451, TL16C452 ASYNCHRONOUS COMMUNICATIONS ELEMENTS

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## Terminal Functions (continued)

PIN NAME†	NO.	I/O	DESCRIPTION
LPTOE	1	I	Parallel data output enable. When low, this signal enables the write data register to the PD0–PD7 lines. A high puts the PD0–PD7 lines in the high-impedance state allowing them to be used as inputs. LPTOE is usually tied low for line printer operation.
PD0–PD7	53–46	I/O	Parallel data bits (0–7). These eight lines provide a byte-wide input or output port to the system. The eight lines are held in a high-impedance state when LPTOE is high.
PE	67	I	Line printer paper empty. This is an input line from the line printer that goes high when the printer runs out of paper.
RESET	39	I	Reset. When active (low), RESET clears most ACE registers and sets the state of various output signals. Refer to Table 2.
RI0	30	I	Ring indicator. RI is an active-low modem status signal whose state can be checked by reading bit 6 (RI) of the modem status register. Bit 2 (TERI) of the modem status register indicates that the RI input has transitioned from a low to a high state since the last read from the modem status register. If the modem status interrupt is enabled when this transition occurs, an interrupt is generated.
RI1 [GND]	6		
RLSD0	29	I	Receive line signal detect. RLSD0 is an active-low modem status signal whose state can be checked by reading bit 7 of the modem status register. Bit 3 (DRLSD) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when RLSD0 changes state, an interrupt is generated. This bit is low when a data carrier is detected.
RLSD1 [GND]	8		
RTS0	24	O	Request to send. When active (low), this signal informs the modem or data set that the ACE is ready to transmit data. RTS is set to its active state by setting the RTS modem control register bit and is set to its inactive (high) state either as a result of a reset or during loop mode operations or by resetting bit 1 (RTS) of the modem control register.
RTS1 [NC]	12		
SIN0 SIN1 [GND]	41	I	Serial input. Serial data input from a connected communications device.
SLCT	65	I	Line printer selected. This is an input line from the line printer that goes high when the line printer has been selected.
SLIN	58	I/O	Line printer select. This open-drain line selects the printer when it is active (low). An internal pullup is provided.
SOUT0 SOUT1 [NC]	26 10	I	Serial output. Composite serial data output to a connected communication device. SOUT is set to the marking (logic 1) state as a result of reset.
STB	55	I/O	Line printer strobe. This open-drain line provides communication synchronization between the TL16C451/TL16C452 and the line printer. When it is active (low), it provides the line printer with a signal to latch the data currently on the parallel port. An internal pullup is provided.
VCC	23,40, 64		5-V supply voltage
GND	2,7,9 22,27,42, 43,54,61		Supply common

† Names shown in brackets are for the TL16C451.



# TL16C451, TL16C452 ASYNCHRONOUS COMMUNICATIONS ELEMENTS

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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	-0.5 V to 7 V
Input voltage range at any input, $V_I$	-0.5 V to 7 V
Output voltage range, $V_O$	-0.5 V to 7 V
Continuous total power dissipation	1100 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 10 seconds	260°C

NOTE 1: All voltage values are with respect to GND.

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2	$V_{CC}$	V	V
Low-level input voltage, $V_{IL}$	-0.5	0.8	V	V
Operating free-air temperature, $T_A$	0	70	°C	

## electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}$ High-level output voltage	$I_{OH} = -0.4$ mA on DB0-DB7	2.4			V
	$I_{OH} = -2$ mA to 4 mA on PD0-PD7				
	$I_{OH} = -0.2$ mA on INIT, AFD, STB, and SLIN				
	$I_{OH} = -0.2$ mA on all other outputs				
$V_{OL}$ Low-level output voltage	$I_{OL} = 4$ mA on DB0-DB7			0.4	V
	$I_{OL} = 12$ mA on PD0-PD7				
	$I_{OL} = 10$ mA on INIT, AFD, STB, and SLIN (see Note 2)				
	$I_{OL} = 2$ mA on all other outputs				
$I_{lk}$ Input leakage current	$V_{CC} = 5.25$ V, $V_{SS} = 0$ , $V_I = 0$ to 5.25 V, All other pins floating			±10	µA
High-impedance output current	$V_{CC} = 5.25$ V, $V_{SS} = 0$ , $V_O = 0$ to 5.25 V, Chip selected and write mode, or chip deselected			±20	µA
$I_{CC}$ Supply current	$V_{CC} = 5.25$ V, $V_{SS} = 0$ , SIN, DSR, RLSD, CTS, and $\overline{RI}$ at 2 V, All other inputs at 0.8 V, XTAL1 at 4 MHz, No load on outputs, Baud rate = 50 kilobits per second			10	mA
$C_{XTAL1}$ Clock input capacitance	$V_{CC} = 0$ , $V_{SS} = 0$ , $f = 1$ MHz, $T_A = 25^\circ\text{C}$ , All others pins grounded		15	20	pF
$C_{XTAL2}$ Clock output capacitance			20	30	pF
$C_i$ Input capacitance			6	10	pF
$C_o$ Output capacitance			10	20	pF

† All typical values are at  $V_{CC} = 5$  V,  $T_A = 25^\circ\text{C}$ .

NOTE 2: INIT, AFD, STB, and SLIN are open-collector output pins that each have an internal pullup to  $V_{CC}$ . This will generate a maximum of 2 mA of internal  $I_{OL}$  per pin. In addition to this internal current, each pin will sink at least 10 mA while maintaining the  $V_{OL}$  specification of 0.4 V Max.



# TL16C451, TL16C452 ASYNCHRONOUS COMMUNICATIONS ELEMENTS

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## system timing requirements over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		FIGURE	MIN	MAX	UNIT
t <sub>cR</sub>	Cycle time, read (t <sub>w7</sub> + t <sub>d8</sub> + t <sub>d9</sub> )		175		ns
t <sub>cW</sub>	Cycle time, write (t <sub>w6</sub> + t <sub>d5</sub> + t <sub>d6</sub> )		175		ns
t <sub>w1</sub>	Pulse duration, clock high	1	50		ns
t <sub>w2</sub>	Pulse duration, clock low	1	50		ns
t <sub>w5</sub>	Pulse duration, write strobe	2	80		ns
t <sub>w6</sub>	Pulse duration, read strobe	3	80		ns
t <sub>wRST</sub>	Pulse duration, reset		1000		ns
t <sub>su1</sub>	Setup time, address	2,3	15		ns
t <sub>su2</sub>	Setup time, chip select	2,3	15		ns
t <sub>su3</sub>	Setup time, data	2	15		ns
t <sub>h1</sub>	Hold time, address	2,3	20		ns
t <sub>h2</sub>	Hold time, chip select	2,3	20		ns
t <sub>h3</sub>	Hold time, data	2	15		ns
t <sub>d3</sub>	Delay time, write cycle	2	80		ns
t <sub>d4</sub>	Delay time, read cycle	3	80		ns

## system switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>d5</sub>	3	C <sub>L</sub> = 100 pF		60	ns
t <sub>d6</sub>	3	C <sub>L</sub> = 100 pF	0	60	ns
t <sub>dis(R)</sub>	3	C <sub>L</sub> = 100 pF		60	ns

## receiver switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>d7</sub>	4			100	ns
t <sub>d8</sub>	4		1	1	RCLK cycles
t <sub>d9</sub>	4	C <sub>L</sub> = 100 pF		140	ns

## transmitter switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>d10</sub>	5		8	24	baudout cycles
t <sub>d11</sub>	5		8	8	baudout cycles
t <sub>d12</sub>	5	C <sub>L</sub> = 100 pF		140	ns
t <sub>d13</sub>	5		16	32	baudout cycles
t <sub>d14</sub>	5	C <sub>L</sub> = 100 pF		140	ns



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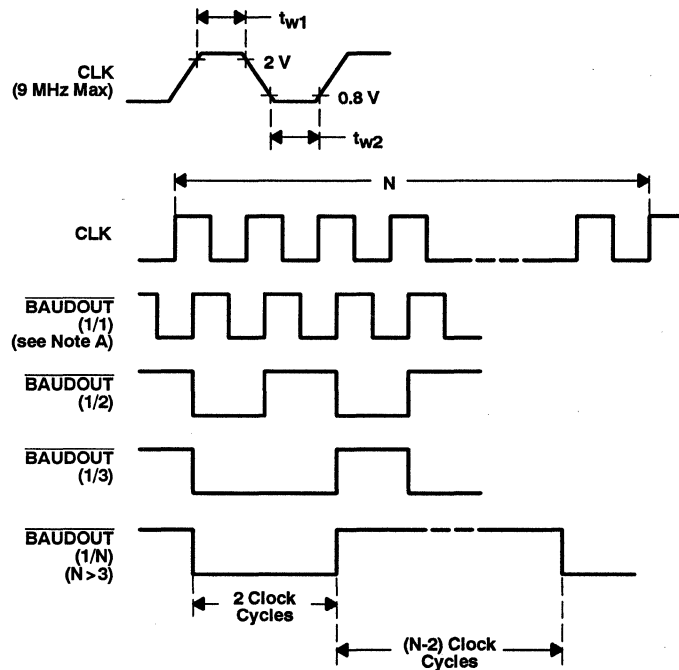
**modem control switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

PARAMETER	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{d15}$ Delay time, write MCR to output	6	$C_L = 100 \text{ pF}$		100	ns
$t_{d16}$ Delay time, modem input to set interrupt	6	$C_L = 100 \text{ pF}$		170	ns
$t_{d17}$ Delay time, read MSR to reset interrupt	6	$C_L = 100 \text{ pF}$		140	ns

**parallel port switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

PARAMETER	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{d18}$ Delay time, write parallel port control to output	7	$C_L = 100 \text{ pF}$		60	ns
$t_{d19}$ Delay time, write parallel port data to output	7	$C_L = 100 \text{ pF}$		60	ns
$t_{d20}$ Delay time, output enable to data	7	$C_L = 100 \text{ pF}$		60	ns
$t_{d21}$ Delay time, $\overline{\text{ACK}}$ to INT2	7	$C_L = 100 \text{ pF}$		100	ns

## PARAMETER MEASUREMENT INFORMATION



NOTE A:  $\overline{\text{BAUDOUT}}$  is an internally generated signal used in the receiver and transmitter circuits to synchronize data.

Figure 1. Baud Generator Timing

PARAMETER MEASUREMENT INFORMATION

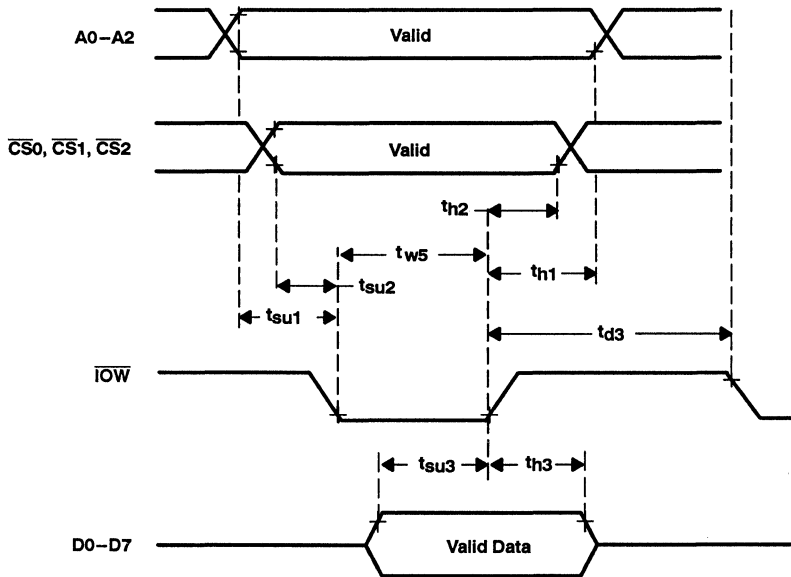


Figure 2. Write Cycle Timing



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## PARAMETER MEASUREMENT INFORMATION

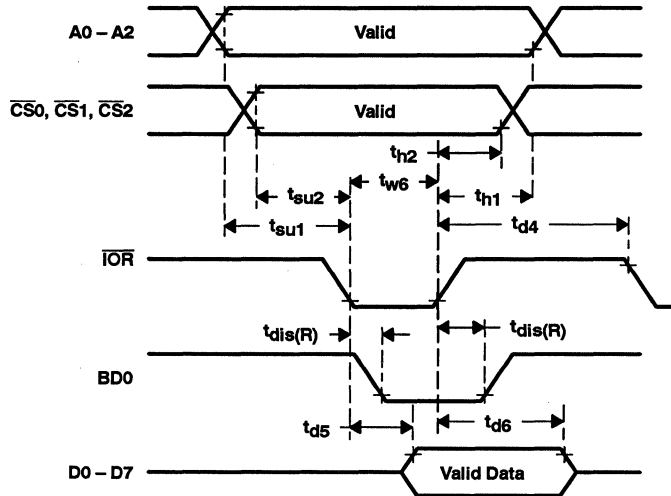


Figure 3. Read Cycle Timing

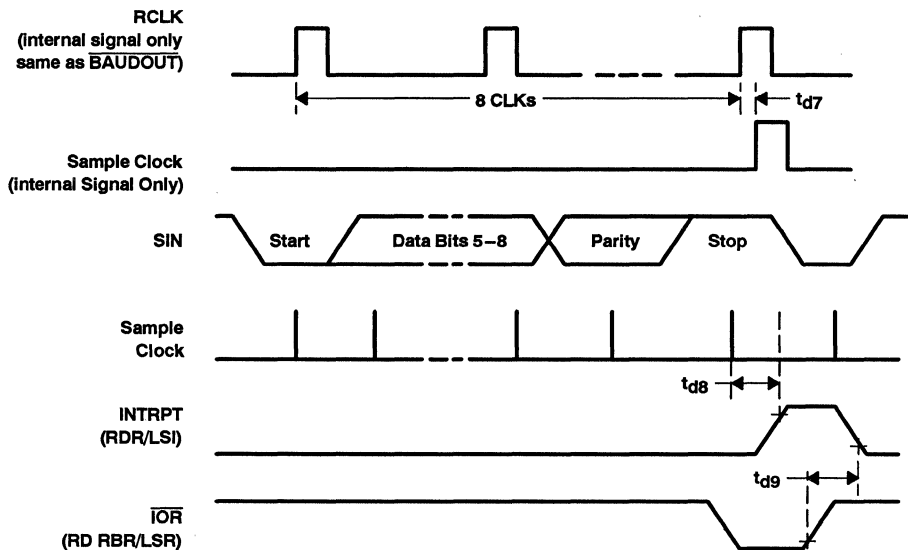


Figure 4. Receiver Timing

PARAMETER MEASUREMENT INFORMATION

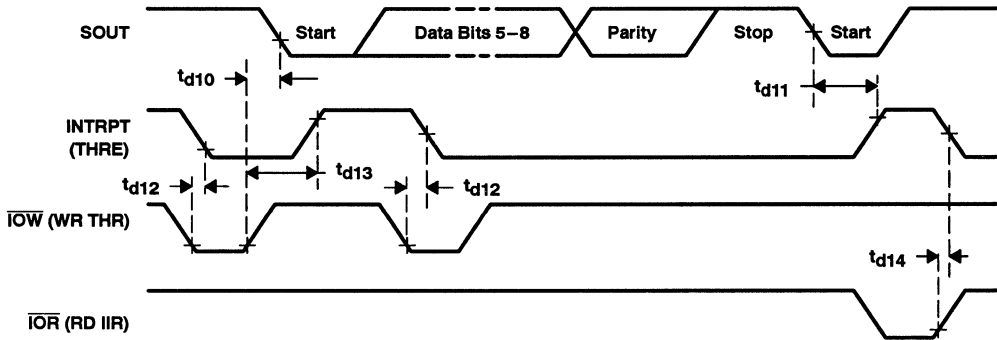


Figure 5. Transmitter Timing

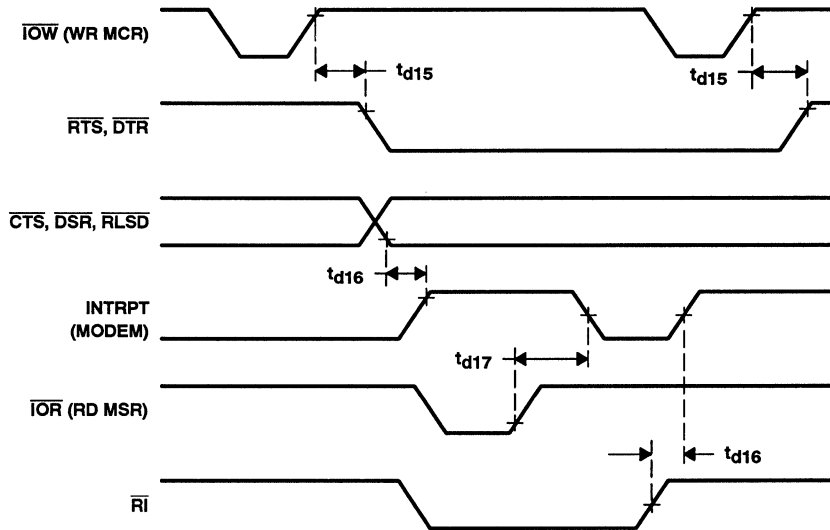


Figure 6. Modem Control Timing

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## PARAMETER MEASUREMENT INFORMATION

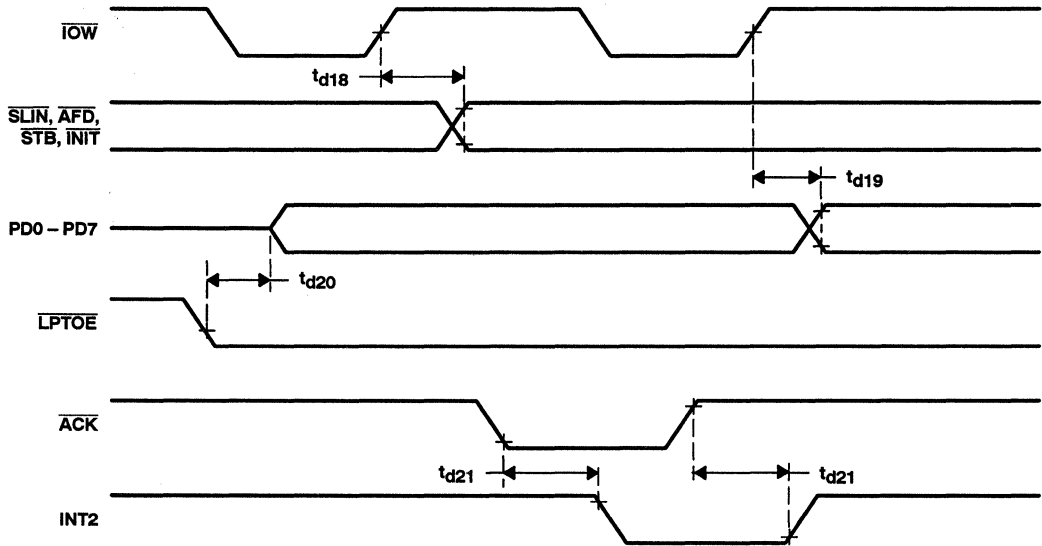


Figure 7. Parallel Port Timing

APPLICATION INFORMATION

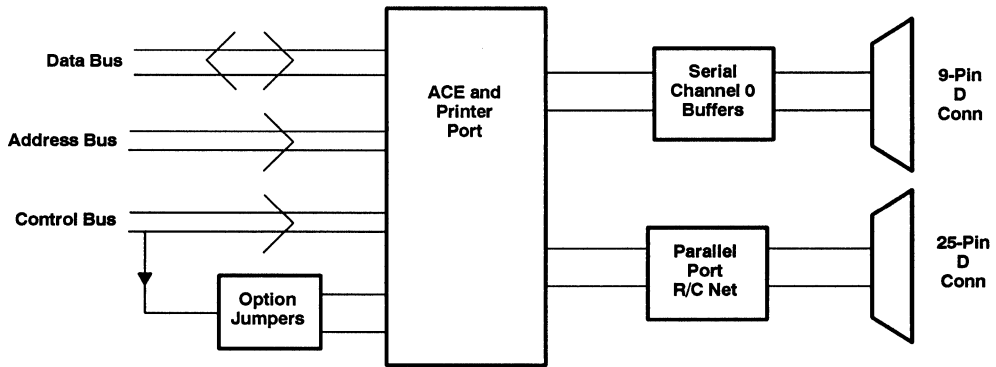


Figure 8. TL16C451

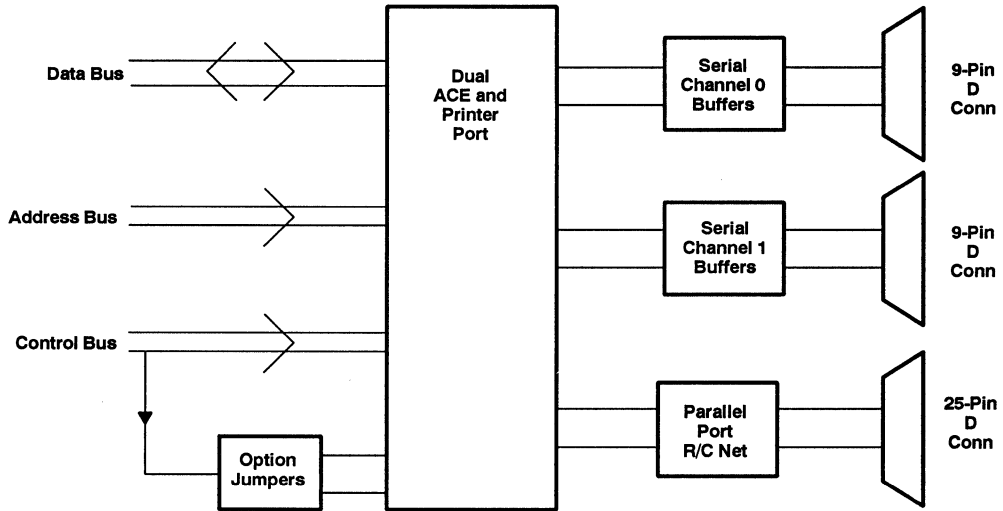


Figure 9. TL16C452

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## PRINCIPLES OF OPERATION

Table 1. Register Selection

DLAB†	A2	A1	A0	REGISTER
0	L	L	L	Receiver buffer (read), transmitter holding register (write)
0	L	L	H	Interrupt enable
X	L	H	L	Interrupt identification (read only)
X	L	H	H	Line control
X	H	L	L	Modem control
X	H	L	H	Line status
X	H	H	L	Modem status
X	H	H	H	Scratch
1	L	L	L	Divisor latch (LSB)
1	L	L	H	Divisor latch (MSB)

† The divisor latch access bit (DLAB) is the most significant bit of the line control register. The DLAB signal is controlled by writing to this bit location (see Table 3).

Table 2. ACE Reset Functions

REGISTER/SIGNAL	RESET CONTROL	RESET STATE
Interrupt Enable Register	Reset	All bits low (0–3 forced and 4–7 permanent)
Interrupt Identification Register	Reset	Bit 0 is high, bits 1 and 2 are low, and bits 3–7 are permanently low
Line Control Register	Reset	All bits low
Modem Control Register	Reset	All bits low
Line Status Register	Reset	Bits 5 and 6 are high, all other bits are low
Modem Status Register	Reset	Bits 0–3 are low, bits 4–7 are input signals
SOUT	Reset	High
INTRPT (receiver error flag)	Read LSR/Reset	Low
INTRPT (received data available)	Read RBR/Reset	Low
INTRPT (transmitter holding register empty)	Read IIR/Write THR/Reset	Low
INTRPT (modem status changes)	Read MSR/Reset	Low
OUT 2 (interrupt enable)	Reset	High
RTS	Reset	High
DTR	Reset	High
OUT 1	Reset	High
Scratch Register	Reset	No effect
Divisor Latch (LSB and MSB) Registers	Reset	No effect
Receiver Buffer Registers	Reset	No effect
Transmitter Holding Registers	Reset	No effect

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## PRINCIPLES OF OPERATION

### accessible registers

The system programmer, via the CPU, has access to and control over any of the ACE registers that are summarized in Table 3. These registers are used to control ACE operations, receive data, and transmit data. Descriptions of these registers follow Table 3.

**Table 3. Summary of Accessible Registers**

Bit No.	REGISTER ADDRESS										
	0 DLAB = 0	0 DLAB = 0	1 DLAB = 0	2	3	4	5	6	7	0 DLAB = 1	1 DLAB = 0
	Receiver Buffer Register (Read Only)	Transmitter Holding Register (Write Only)	Interrupt Enable Register	Interrupt Ident. Register (Read Only)	Line Control Register	Modem Control Register	Line Status Register	Modem Status Register	Scratch Register	Divisor Latch (LSB)	Latch (MSB)
	RBR	THR	IER	IIR	LCR	MCR	LSR	MSR	SCR	DLL	DLM
0	Data Bit 0†	Data Bit 0	Enable Received Data Available Interrupt (ERBF)	"0" If Interrupt Pending	Word Length Select Bit 0 (WLS0)	Data Terminal Ready (DTR)	Data Ready (DR)	Delta Clear to Send (DCTS)	Bit 0	Bit 0	Bit 8
1	Data Bit 1	Data Bit 1	Enable Transmitter Holding Register Empty Interrupt (ETBE)	Interrupt ID Bit (0)	Word Length Select Bit 1 (WLS1)	Request to Send (RTS)	Overrun Error (OE)	Delta Data Set Ready (DDSR)	Bit 1	Bit 1	Bit 9
2	Data Bit 2	Data Bit 2	Enable Receiver Line Status Interrupt (ELSI)	Interrupt ID Bit (1)	Number of Stop Bits (STB)	Out 1	Parity Error (PE)	Trailing Edge Ring Indicator (TERI)	Bit 2	Bit 2	Bit 10
3	Data Bit 3	Data Bit 3	Enable Modem Status Interrupt (EDSSI)	0	Parity Enable (PEN)	Out 2 (Interrupt Enable)	Framing Error (FE)	Delta Receive Line Signal Detect (DRLSD)	Bit 3	Bit 3	Bit 11
4	Data Bit 4	Data Bit 4	0	0	Even Parity Select (EPS)	Loop	Break Interrupt (BI)	Clear to Send (CTS)	Bit 4	Bit 4	Bit 12
5	Data Bit 5	Data Bit 5	0	0	Stick Parity	0	Transmitter Holding Register (THRE)	Data Set Ready (DSR)	Bit 5	Bit 5	Bit 13
6	Data Bit 6	Data Bit 6	0	0	Set Break	0	Transmitter Empty (TEMT)	Ring Indicator (RI)	Bit 6	Bit 6	Bit 14
7	Data Bit 7	Data Bit 7	0	0	Divisor Latch Access Bit (DLAB)	0	0	Receive Line Signal Detect (RLSD)	Bit 7	Bit 7	Bit 15

† Bit 0 is the least significant bit. It is the first bit serially transmitted or received.



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## PRINCIPLES OF OPERATION

### receiver buffer register (RBR)

The ACE receiver section consists of a receiver shift register and a receiver buffer register. Timing is supplied by the 16 X receiver clock (RCLK). Receiver section control is a function of the ACE line control register.

The ACE receiver shift register receives serial data from the serial input (SIN) pin. The receiver shift register then converts the data to a parallel form and loads it into the receiver buffer register. When a character is placed in the receiver buffer register and the received data available interrupt is enabled, an interrupt is generated. This interrupt is cleared when the data is read out of the receiver buffer register.

### transmitter holding register (THR)

The ACE transmitter section consists of a transmitter holding register and a transmitter shift register. Timing is supplied by the baud out (BAUDOUT) clock signal. Transmitter section control is a function of the ACE line control register.

The ACE transmitter holding register receives data off the internal data bus and, when the shift register is idle, moves it into the transmitter shift register. The transmitter shift register serializes the data and outputs it at the serial output (SOUT). If the transmitter holding register is empty and the transmitter holding register empty (THRE) interrupt is enabled, an interrupt is generated. This interrupt is cleared when a character is loaded into the register.

### interrupt enable register (IER)

The interrupt enable register enables each of the four types of interrupts (refer to Table 4) and the INTRPT output signal in response to an interrupt generation. The interrupt enable register can also be used to disable the interrupt system by setting bits 0 through 3 to logic 0. The contents of this register are summarized in Table 3 and are described below.

Bit 0. This bit, when set to logic 1, enables the received data available interrupt.

Bit 1. This bit, when set to logic 1, enables the transmitter holding register empty interrupt.

Bit 2. This bit, when set to logic 1, enables the receiver line status interrupt.

Bit 3. This bit, when set to logic 1, enables the modem status interrupt.

Bits 4 thru 7. Bits 4 through 7 in the interrupt enable register are not used and are always set to logic 0.

**PRINCIPLES OF OPERATION**

**interrupt identification register (IIR)**

The ACE has an on-chip interrupt generation and prioritization capability that permits a flexible interface with most microprocessors.

The ACE provides four prioritized levels of interrupts:

- Priority 1 – Receiver line status (highest priority)
- Priority 2 – Receiver data ready
- Priority 3 – Transmitter holding register empty
- Priority 4 – Modem status (lowest priority)

When an interrupt is generated, the interrupt identification register indicates that an interrupt is pending and the type of that interrupt in its three least significant bits (bits 0, 1, and 2). The contents of this register are summarized in Table 3 and described in Table 4.

Bit 0. This bit can be used either in a hardwire prioritized or polled interrupt system. If this bit is a logic 0, an interrupt is pending. When bit 0 is a logic 1, no interrupt is pending.

Bits 1 and 2. These two bits are used to identify the highest priority interrupt pending as indicated in Table 4.

Bits 3 thru 7. Bits 3 through 7 in the interrupt identification register are not used and are always set at logic 0.

**Table 4 . Interrupt Control Functions**

INTERRUPT IDENTIFICATION REGISTER			PRIORITY LEVEL	INTERRUPT TYPE	INTERRUPT SOURCE	INTERRUPT RESET METHOD
BIT 2	BIT 1	BIT 0				
0	0	1	None	None	None	—
1	1	0	1	Receiver line status	Overrun error, parity error, framing error or break interrupt	Reading the line status register
1	0	0	2	Received data available	Receiver data available	Reading the receiver buffer register
0	1	0	3	Transmitter holding register empty	Transmitter holding register empty	Reading the interrupt Identification register (if source of interrupt) or writing into the transmitter holding register
0	0	0	4	Modem status	Clear to send, data set ready, ring indicator, or data carrier detect	Reading the modem status register



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## PRINCIPLES OF OPERATION

### line control register (LCR)

The system programmer controls the format of the asynchronous data communication exchange through the line control register. In addition, the programmer is able to retrieve, inspect, and modify the contents of the line control register; this eliminates the need for separate storage of the line characteristics in system memory. The contents of this register are summarized in Table 3 and are described below.

Bits 0 and 1. These two bits specify the number of bits in each transmitted or received serial character. These bits are encoded as follows:

Bit 1	Bit 0	Word Length
0	0	5 Bits
0	1	6 Bits
1	0	7 Bits
1	1	8 Bits

Bit 2. This bit specifies either one, one and one-half, or two stop bits in each transmitted character. If bit 2 is a logic 0, one stop bit is generated in the data. If bit 2 is a logic 1, the number of stop bits generated is dependent on the word length selected with bits 0 and 1. The number of stop bits generated, in relation to word length and bit 2, is as follows:

Bit 2	Word Length Selected by Bits 1 and 2	Number of Stop Bits Generated
0	Any word length	1
1	5 bits	1 1/2
1	6 bits	2
1	7 bits	2
1	8 bits	2

Bit 3. This bit is the parity enable bit. When bit 3 is a logic 1, a parity bit is generated in transmitted data between the last data word bit and the first stop bit. In received data, if bit 3 is a logic 1, parity is checked. When bit 3 is a logic 0, no parity is generated or checked.

Bit 4. Bit 4 is the even parity select bit. When parity is enabled by bit 3: a logic 1 in bit 4 produces even parity (an even number of logic 1s in the data and parity bits) and a logic 0 in bit 4 produces odd parity (an odd number of logic 1s).

Bit 5. This is the stick parity bit. When bits 3, 4, and 5 are logic 1s, the parity bit is transmitted and checked as a logic 0. When bits 3 and 5 are logic 1s and bit 4 is a logic 0, the parity bit is transmitted and checked as a logic 1.

Bit 6. This bit is the break control bit. Bit 6 is set to a logic 1 to force a break condition, i.e., a condition where the serial output (SOUT) pin is forced to the spacing (logic 0) state. When bit 6 is set to a logic 0, the break condition is disabled. The break condition has no effect on the transmitter logic, it only effects the serial output.

Bit 7. This bit is the divisor latch access bit (DLAB). Bit 7 must be set to a logic 1 to access the divisor latches of the baud generator during a read or write. Bit 7 must be set to a logic 0 during a read or write to access the receiver buffer, the transmitter holding register, or the interrupt enable register.

## PRINCIPLES OF OPERATION

### modem control register (MCR)

The modem control register is an 8-bit register that controls an interface with a modem, data set, or peripheral device that is emulating a modem. The contents of this register are summarized in Table 3 and are described below.

**Bit 0.** Bit 0 (DTR) controls the data terminal ready ( $\overline{\text{DTR}}$ ) output. Setting this bit to a logic 1 forces the  $\overline{\text{DTR}}$  output to its active state (low). When bit 0 is set to a logic 0,  $\overline{\text{DTR}}$  goes high.

**Bit 1.** Bit 1 (RTS) controls the request to send ( $\overline{\text{RTS}}$ ) output in a manner identical to Bit 0's control over the  $\overline{\text{DTR}}$  output.

**Bit 2.** Bit 2 (OUT 1) is a reserved location used only in the loopback mode.

**Bit 3.** Bit 3 (OUT 2) controls the output enable for the interrupt signal. When set to a logic 1, the interrupt is enabled. When bit 3 is set to a logic 0, the interrupt is disabled.

**Bit 4.** Bit 4 provides a local loopback feature for diagnostic testing of the ACE. When this bit is set to a logic high, the following occurs:

1. The transmitter serial output (SOUT) is set high.
2. The receiver serial input (SIN) is disconnected.
3. The output of the transmitter shift register is looped back into the receiver shift register input.
4. The four modem status inputs (CTS, DSR, RLSD, and  $\overline{\text{RI}}$ ) are disconnected.
5. The modem control register bits (DTR, RTS, OUT1, and OUT2) are connected to the modem status register bits (DSR, CTS, RI, and RLSD), respectively.
6. The four modem control output pins are forced to their inactive states (high).

In the diagnostic mode, data that is transmitted is immediately received. This allows the processor to verify the transmit- and receive-data paths to the ACE. The receiver and transmitter interrupts are fully operational. The modem control interrupts are also operational but the modem control interrupt sources are now the lower four bits of the modem control register instead of the four modem control inputs. All interrupts are still controlled by the interrupt enable register.

Bit 5 through 7. These bits are set to logic 0.

### line status register (LSR)<sup>†</sup>

The line status register provides information to the CPU concerning the status of data transfers. The contents of this register are summarized in Table 3 and are described below.

**Bit 0.** Bit 0 is the data ready (DR) indicator for the receiver. This bit is set to a logic 1 condition whenever a complete incoming character has been received and transferred into the receiver buffer register and is reset to logic 0 by reading the receiver buffer register.

**Bit 1<sup>‡</sup>.** Bit 1 is the overrun error (OE) indicator. When this bit is set to logic 1, it indicates that before the character in the receiver buffer register was read, it was overwritten by the next character transferred into the register. The OE indicator is reset every time the CPU reads the contents of the line status register.

**Bit 2<sup>‡</sup>.** Bit 2 is the parity error (PE) indicator. When this bit is set to logic 1, it indicates that the parity of the received data character does not match the parity selected in the line control register (bit 4). The PE bit is reset every time the CPU reads the contents of the line status register.

**Bit 3<sup>‡</sup>.** Bit 3 is the framing error (FE) indicator. When this bit is set to logic 1, it indicates that the received character did not have a valid (logic 1) Stop bit. The FE bit is reset every time the CPU reads the contents of the line status register.

<sup>†</sup> The line status register is intended for read operations only; writing to this register is not recommended outside of a factory testing environment.

<sup>‡</sup> Bits 1 through 4 are the error conditions that produce a receiver line status interrupt.

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## PRINCIPLES OF OPERATION

### line status register (LSR)<sup>†</sup> (continued)

Bit 4<sup>‡</sup>. Bit 4 is the break interrupt (BI) indicator. When this bit is set to logic 1, it indicates that the received data input was held in the logic 0 state for longer than a full-word transmission time. A full-word transmission time is defined as the total time of the start, data, parity, and stop bits. The BI bit is reset every time the CPU reads the contents of the line status register.

Bit 5. Bit 5 is the transmitter holding register empty (THRE) indicator. This bit is set to a logic 1 condition when the transmitter holding register is empty, indicating that the ACE is ready to accept a new character. If the THRE interrupt is enabled when the THRE bit is a logic 1, then an interrupt is generated. THRE is set to a logic 1 when the contents of the transmitter holding register are transferred to the transmitted shift register. This bit is reset to logic 0 concurrent with the loading of the transmitter holding register by the CPU.

Bit 6. Bit 6 is the transmitter empty (TEMT) indicator. This bit is set to a logic 1 when the transmitter holding register and the transmitter shift register are both empty. When either the transmitter holding register or the transmitter shift register contains a data character, the TEMT bit is reset to logic 0.

Bit 7. This bit is always reset to logic 0.

<sup>†</sup> The line status register is intended for read operations only; writing to this register is not recommended outside of a factory testing environment.

<sup>‡</sup> Bits 1 through 4 are the error conditions that produce a receiver line status interrupt.

### modem status register (MSR)

The modem status register is an 8-bit register that provides information about the current state of the control lines from the modem, data set, or peripheral device to the CPU. Additionally, four bits of this register provides change information; when a control input from the modem changes state the appropriate bit is set to logic 1. All four bits are reset to logic 0 when the CPU reads the modem status register. The contents of this register are summarized in Table 3 and are described below.

Bit 0. Bit 0 is the delta clear to send (DCTS) indicator. This bit indicates that the  $\overline{\text{CTS}}$  input has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status Interrupt is enabled, a modem status interrupt is generated.

Bit 1. Bit 1 is the delta data set ready (DDSR) indicator. This bit indicates that the  $\overline{\text{DSR}}$  input has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status Interrupt is enabled, a modem status interrupt is generated.

Bit 2. Bit 2 is the trailing edge of ring indicator (TERI) detector. This bit indicates that the  $\overline{\text{RI}}$  input to the chip has changed from a low to a high state. When this bit is a logic 1 and the modem status Interrupt is enabled, a modem status interrupt is generated.

Bit 3. Bit 3 is the delta receive line signal detect (DRLSD) indicator. This bit indicates that the  $\overline{\text{RLSD}}$  input to the chip has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 4. Bit 4 is the complement of the clear to send ( $\overline{\text{CTS}}$ ) input. If Bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 1 (RTS).

Bit 5. Bit 5 is the complement of the data set ready ( $\overline{\text{DSR}}$ ) input. If Bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 0 (DTR).

Bit 6. Bit 6 is the complement of the ring indicator ( $\overline{\text{RI}}$ ) input. If Bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control registers bit 2 (OUT 1).

Bit 7. Bit 7 is the complement of the receive line signal detect ( $\overline{\text{RLSD}}$ ) input. If Bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control registers bit 3 (OUT 2).



**PRINCIPLES OF OPERATION**

**scratch register (SCR)**

The scratch register is an 8-bit register that is intended for programmer use as a scratchpad, in the sense that it will temporarily hold programmer data without affecting any other ACE operation.

**programmable baud generator**

The ACE contains a programmable baud generator that takes a clock input in the range between dc and 9 MHz and divides it by a divisor in the range between 1 and  $2^{16}-1$ . The output frequency of the baud generator is sixteen times (16 X) the baud rate. The formula for the divisor is:

$$\text{divisor \#} = \text{CLK frequency input} / (\text{desired baud rate} \times 16)$$

Two 8-bit registers, called divisor latches, are used to store the divisor in a 16-bit binary format. These divisor latches must be loaded during initialization of the ACE in order to ensure desired operation of the baud generator. When either of the divisor latches is loaded, a 16-bit baud counter is also loaded to prevent long counts on initial load. For baud rates of 38.4 kilobits per second and below, the error obtained is very small. The accuracy of the selected baud rate is dependent on the selected crystal frequency.

**interrupt control logic**

The interrupt control logic is shown in Figure 10.

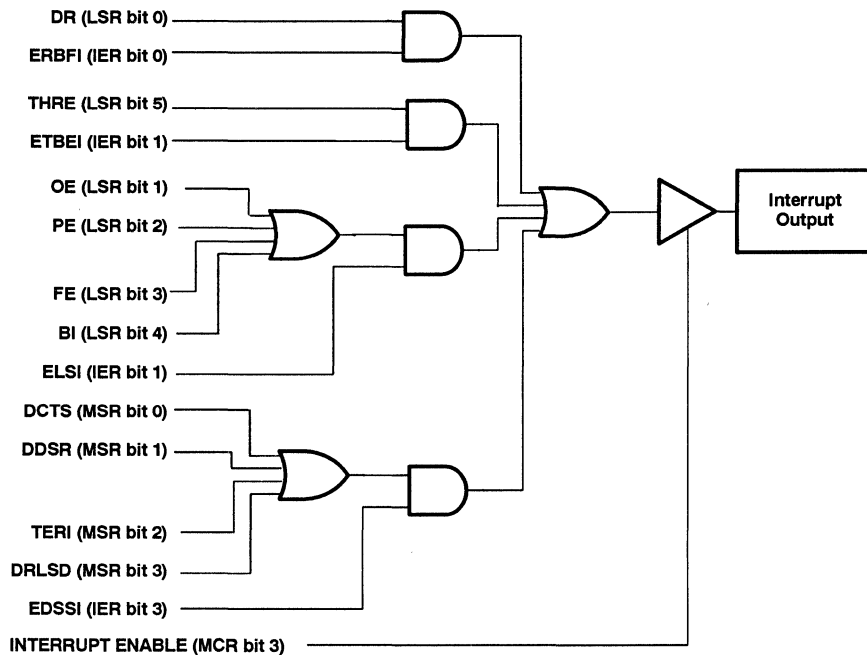


Figure 10. Interrupt Control Logic

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## PRINCIPLES OF OPERATION

### parallel port registers

The parallel port registers interface either device to a Centronix-style printer. When chip select 2 ( $\overline{CS2}$ ) is low, the parallel port is selected. Tables 5 and 6 show the registers associated with this parallel port. The read or write function of the register is controlled by the state or the read ( $\overline{IOR}$ ) and write ( $\overline{IOW}$ ) pin as shown. The read data register allows the microprocessor to read the information on the parallel bus.

The read status register allows the microprocessor to read the status of the printer in the five most significant bits. The status bits are printer busy ( $\overline{BUSY}$ ), acknowledge ( $\overline{ACK}$ ) which is a handshake function, paper empty (PE), printer selected (SLCT), and error ( $\overline{ERROR}$ ). The read control register allows the state of the control lines to be read. The write control register sets the state of the control lines, which are interrupt enable (IRQ ENB), select In (SLIN), initialize the printer ( $\overline{INIT}$ ), autofeed the paper (AFD), and strobe (STB), which informs the printer of the presence of a valid byte on the parallel bus. These signals are set to 0 when a reset occurs. The write data register allows the microprocessor to write a byte to the parallel bus. The parallel port is completely compatible with the parallel port implementation used in the IBM serial/parallel adaptor.

**Table 5. Parallel Port Registers**

REGISTER	REGISTER BITS							
	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
Read Data	PD7	PD6	PD5	PD4	PD3	PD2	PD1	PD0
Read Status	$\overline{BUSY}$	$\overline{ACK}$	PE	SLCT	$\overline{ERROR}$	1	1	1
Read Control	1	1	1	IRQ ENB	SLIN	$\overline{INIT}$	AFD	STB
Write Data	PD7	PD6	PD5	PD4	PD3	PD2	PD1	PD0
Write Control	1	1	1	IRQ ENB	SLIN	$\overline{INIT}$	AFD	STB

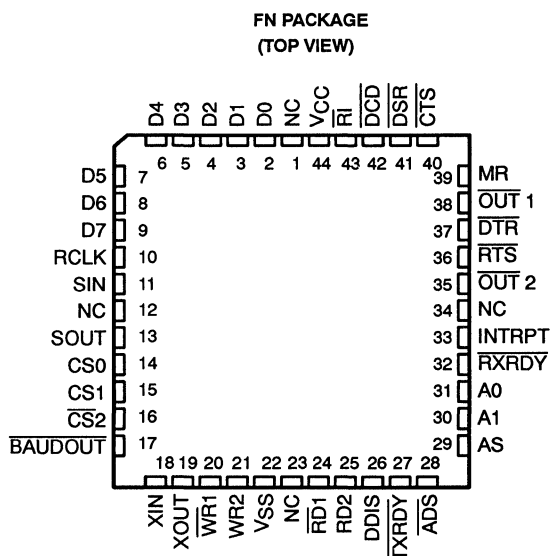
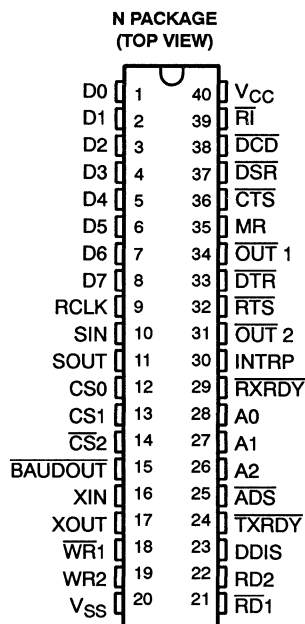
**Table 6. Parallel Port Register Select**

CONTROL PINS					REGISTER SELECTED
$\overline{IOR}$	$\overline{IOW}$	$\overline{CS2}$	A1	A0	
L	H	L	L	L	Read Data
L	H	L	L	H	Read Status
L	H	L	H	L	Read control
L	H	L	H	H	Invalid
H	L	L	L	L	Write Data
H	L	L	L	H	Invalid
H	L	L	H	L	Write Control
H	L	L	H	H	Invalid

# TL16C550A ASYNCHRONOUS COMMUNICATIONS ELEMENT

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- Capable of Running With All Existing TL16C450 Software
- After Reset, All Registers Are Identical to the TL16C450 Register Set
- In the FIFO Mode, Transmitter and Receiver Are Each Buffered With 16-Byte FIFOs to Reduce the Number of Interrupts to the CPU
- In the TL16C450 Mode, Holding and Shift Registers Eliminate the Need for Precise Synchronization Between the CPU and Serial Data
- Programmable Baud Rate Generator Allows Division of Any Input Reference Clock by 1 to  $(2^{16} - 1)$  and Generates an Internal 16 X Clock
- Adds or Deletes Standard Asynchronous Communication Bits (Start, Stop, and Parity) to or From the Serial Data Stream
- Independent Receiver Clock Input
- Independently Controlled Transmit, Receive, Line Status, and Data Set Interrupts
- Fully Programmable Serial Interface Characteristics:
  - 5-, 6-, 7-, or 8-Bit Characters
  - Even-, Odd-, or No-Parity Bit Generation and Detection
  - 1-, 1 1/2-, or 2-Stop Bit Generation
  - Baud Generation (dc to 256 Kilobits Per Second)
- False-Start Bit Detection
- Complete Status Reporting Capabilities
- 3-State TTL Drive Capabilities for Bidirectional Data Bus and Control Bus
- Line Break Generation and Detection
- Internal Diagnostic Capabilities:
  - Loopback Controls for Communications
  - Link Fault Isolation
  - Break, Parity, Overrun, Framing Error Simulation
- Full Prioritized Interrupt System Controls
- Modem Control Functions ( $\overline{\text{CTS}}$ ,  $\overline{\text{RTS}}$ ,  $\overline{\text{DSR}}$ ,  $\overline{\text{DTR}}$ ,  $\overline{\text{RI}}$ , and  $\overline{\text{DCD}}$ )
- Faster Plug-In Replacement for National Semiconductor NS16550A



NC—No internal connection

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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# TL16C550A

## ASYNCHRONOUS COMMUNICATIONS ELEMENT

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### description

The TL16C550A is a functional upgrade of the TL16C450 asynchronous communications element (ACE). Functionally identical to the TL16C450 on power up (character mode<sup>†</sup>), the TL16C550A can be placed in an alternate mode (FIFO) to relieve the CPU of excessive software overhead.

In this mode, internal FIFOs are activated allowing 16 bytes (plus 3 bits of error data per byte in the receiver FIFO) to be stored in both receive and transmit modes. To minimize system overhead and maximize system efficiency, all logic is on the chip. Two of the TL16C450 pin functions (pins 24 and 29 on the N package and pins 27 and 32 on the FN package) have been changed to allow signalling of DMA transfers.

The TL16C550A performs serial-to-parallel conversion on data received from a peripheral device or modem and parallel-to-serial conversion on data received from its CPU. The CPU can read and report on the status of the ACE at any point in the ACE's operation. Reported status information includes: the type of transfer operation in progress, the status of the operation, and any error conditions encountered.

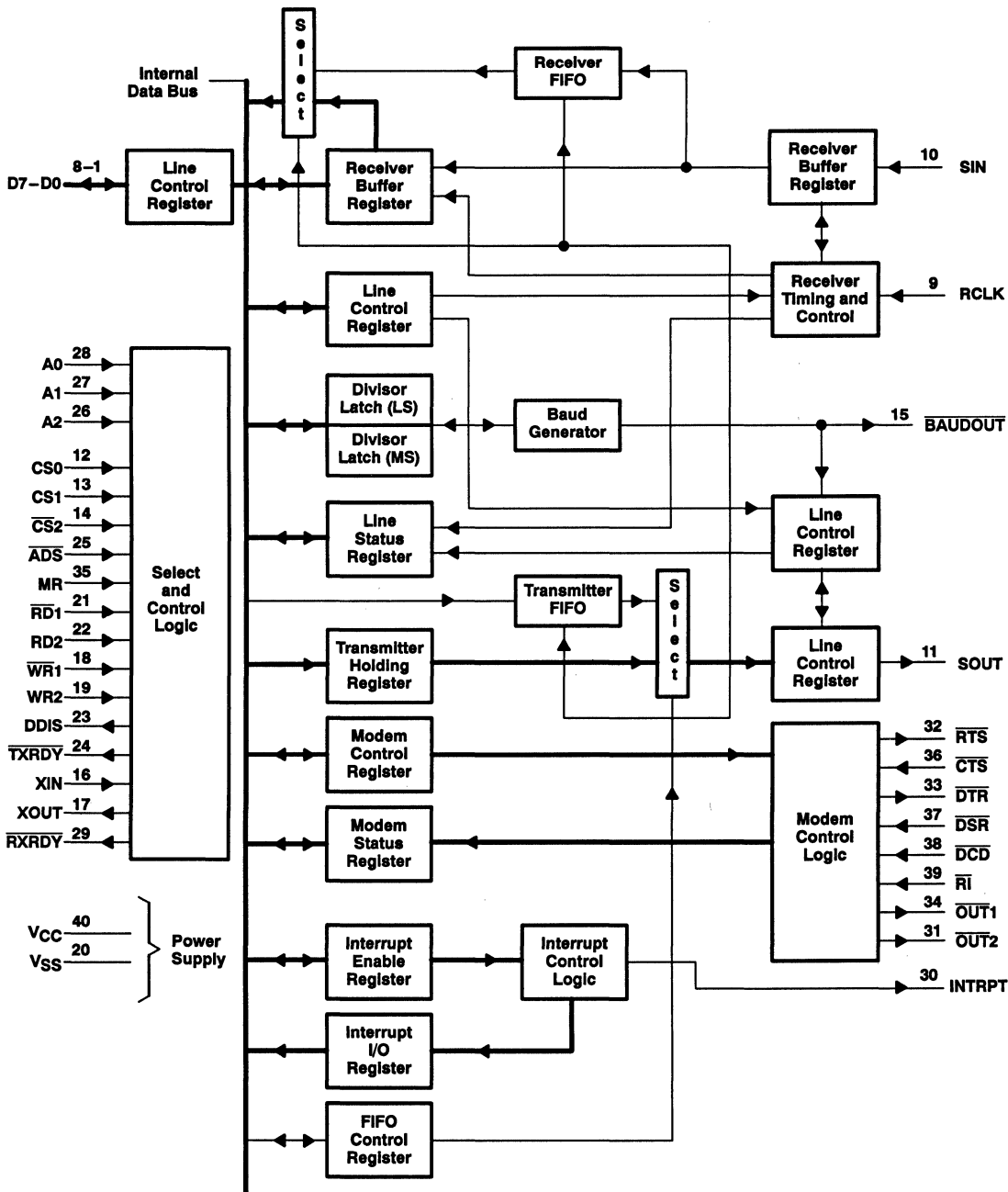
The TL16C550A ACE includes a programmable, on-board, baud-rate generator. This generator is capable of dividing a reference clock input by divisors from 1 to  $(2^{16} - 1)$  and producing a 16 X clock for driving the internal transmitter logic. Provisions are also included to use this 16 X clock to drive the receiver logic. Also included in the ACE is a complete modem control capability and a processor interrupt system that may be software tailored to the user's requirements to minimize the computing required to handle the communications link.

<sup>†</sup> The TL16C550A can also be reset to the TL16C450 mode under software control.

# TL16C550A ASYNCHRONOUS COMMUNICATIONS ELEMENT

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block diagram



Pin numbers shown are for the N package.



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# TL16C550A ASYNCHRONOUS COMMUNICATIONS ELEMENT

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## Terminal Functions

PIN		I/O	DESCRIPTION
NAME	NO.†		
A0 A1 A2	28 [31] 27 [30] 26 [29]	I	Register select. Three inputs used during read and write operations to select the ACE register to read from or write to. Refer to Table 1 for register addresses, also refer to the address strobe (ADS) signal description.
ADS	25 [28]	I	Address strobe. When ADS is active (low), the register select signals (A0, A1, and A2) and chip select signals (CS0, CS1, CS2) drive the internal select logic directly; when high, the register select and chip select signals are held in the state they were in when the low-to-high transition of ADS occurred.
BAUDOUT	15 [17]	O	Baud out. 16 X clock signal for the transmitter section of the ACE. The clock rate is established by the reference oscillator frequency divided by a divisor specified by the baud generator divisor latches. BAUDOUT may also be used for the receiver section by tying this output to the RCLK input.
CS0 CS1 CS2	12 [14] 13 [15] 14 [16]	I	Chip select. When active (high, high, and low, respectively), these three inputs select the ACE. If any of these inputs are inactive, the ACE remains inactive. Refer to the ADS (address strobe) signal description.
CTS	36 [40]	I	Clear to send. CTS is a modem status signal whose condition can be checked by reading bit 4 (CTS) of the modem status register. Bit 0 (DCTS) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when CTS changes state, an interrupt is generated.
D0 D1 D2 D3 D4 D5 D6 D7	1 [2] 2 [3] 3 [4] 4 [5] 5 [6] 6 [7] 7 [8] 8 [9]	I/O	Data bus. Eight 3-state data lines provide a bidirectional path for data, control, and status information between the ACE and the CPU.
DCD	38 [42]	I	Data carrier detect. DCD is a modem status signal whose condition can be checked by reading bit 7 (DCD) of the modem status register. Bit 3 (DDCD) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when the DCD changes state, an interrupt is generated.
DDIS	23 [26]	O	Driver disable. This output is active (high) when the CPU is not reading data. When active, this output can be used to disable an external transceiver.
DSR	37 [41]	I	Data set ready. DSR is a modem status signal whose condition can be checked by reading bit 5 (DSR) of the modem status register. Bit 1 (DDSR) of the modem status register indicates that this signal has changed state since the last read from the modem status register. If the modem status interrupt is enabled when the DSR changes state, an interrupt is generated.
DTR	33 [37]	O	Data terminal ready. When active (low), DTR informs a modem or data set that the ACE is ready to establish communication. DTR is placed in the active state by setting the DTR bit of the modem control register to a high level. DTR is placed in the inactive state either as a result of a master reset or during loop mode operation or resetting bit 0 (DTR) of the modem control register.
INTRPT	30 [33]	O	Interrupt. When active (high), INTRPT informs the CPU that the ACE has an interrupt to be serviced. Four conditions that cause an interrupt to be issued are: a receiver error, received data is available or timeout (FIFO mode only), the transmitter holding register is empty, and an enabled modem status interrupt. The INTRPT output is reset (deactivated) either when the interrupt is serviced or as a result of a master reset.
MR	35 [39]	I	Master reset. When active (high), MR clears most ACE registers and sets the state of various output signals. Refer to Table 2.
OUT1 OUT2	34 [38] 31 [35]	O	Outputs 1 and 2. User-designated output pins that are set to their active states by setting their respective modem control register bits (OUT 1 and OUT 2) high. OUT 1 and OUT 2 are set to their inactive (high) states as a result of master reset or during loop mode operations or by resetting bit 2 (OUT 1) or bit 3 (OUT 2) of the MCR.
RCLK	9 [10]	I	Receiver clock. The 16 X baud rate clock for the receiver section of the ACE.
RD1 RD2	21 [24] 22 [25]	I	Read inputs. When either input is active (high or low, respectively) while the ACE is selected, the CPU is allowed to read status information or data from a selected ACE register. Only one of these inputs is required for the transfer of data during a read operation; the other input should be tied in its inactive state (i.e., RD2 tied low or RD1 tied high).

† Pin numbers shown in brackets are for the FN package.



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## Terminal Functions (continued)

PIN NAME	NO.†	I/O	DESCRIPTION
RI	39 [43]	I	Ring indicator. RI is a modem status signal whose condition can be checked by reading bit 6 (RI) of the modem status register. Bit 2 (TERI) of the modem status register indicates that the $\bar{R}I$ input has transitioned from a low to a high state since the last read from the modem status register. If the modem status interrupt is enabled when this transition occurs, an interrupt is generated.
RTS	32 [36]	O	Request to send. When active, informs the modem or data set that the ACE is ready to transmit data. RTS is set to its active state by setting the RTS modem control register bit and is set to its inactive (high) state either as a result of a master reset or during loop-mode operations or by resetting bit 1 (RTS) of the MCR.
RXRDY	29 [32]	O	Receiver ready output. Receiver DMA signalling is available with this pin. When operating in the FIFO mode, one of two types of DMA signalling can be selected via FCR3. When operating in the TL16V450 mode, only DMA Mode 0 is allowed. Mode 0 supports single-transfer DMA in which a transfer is made between CPU bus cycles. Mode 1 supports multitransfer DMA in which multiple transfers are made continuously until the receiver FIFO has been emptied. In DMA Mode 0 (FCRO = 0 or FCRO = 1, FCR3 = 0), if there is at least 1 character in the receiver FIFO or receiver holding register, $\bar{R}XR\bar{D}Y$ will be active (low). When $\bar{R}XR\bar{D}Y$ has been active but there are no characters in the FIFO or holding register, $\bar{R}XR\bar{D}Y$ will go inactive (high). In DMA Mode 1 (FCRO = 1, FCR3 = 1), when the trigger level or the timeout has been reached, $\bar{R}XR\bar{D}Y$ will go active (low); when it has been active but there are no more characters in the FIFO or holding register, it will go inactive (high).
SIN	10 [11]	I	Serial input. Serial data input from a connected communications device.
SOUT	11 [13]	O	Serial output. Composite serial data output to a connected communication device. SOUT is set to the marking (logic 1) state as a result of master reset.
TXRDY	24 [27]	O	Transmitter ready output. Transmitter DMA signalling is available with this pin. When operating in the FIFO mode, one of two types of DMA signalling can be selected via FCR3. When operating in the TL16C450 mode, only DMA mode 0 is allowed. Mode 0 supports single-transfer DMA in which a transfer is made between CPU bus cycles. Mode 1 supports multitransfer DMA in which multiple transfers are made continuously until the transmit FIFO has been filled.
VCC	40 [44]		5-V supply voltage
VSS	20 [22]		Supply common
WR1	18 [20]	I	Write inputs. When either input is active (high or low, respectively), while the ACE is selected, the CPU is allowed to write control words or data into a selected ACE register. Only one of these inputs is required to transfer data during a write operation; the other input should be tied in its inactive state (i.e., WR2 tied low or WR1 tied high).
WR2	19 [21]	I	
XIN	16 [18]	I/O	External clock. Connects the ACE to the main timing reference (clock or crystal).
XOUT	17 [19]		

† Pin numbers shown in brackets are for the FN package.

### absolute maximum ratings over free-air temperature range (unless otherwise noted)

Supply voltage range, $V_{CC}$ (see Note 1)	–0.5 V to 7 V
Input voltage range at any input, $V_I$	–0.5 V to 7 V
Output voltage range, $V_O$	–0.5 V to 7 V
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	–65°C to 150°C
Case temperature for 10 seconds: FN package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: N package	260°C

NOTE 1: All voltage values are with respect to  $V_{SS}$ .

### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2	$V_{CC}$		V
Low-level input voltage, $V_{IL}$	–0.5		0.8	V
Operating free-air temperature, $T_A$	0		70	°C

  
**TEXAS INSTRUMENTS**

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# TL16C550A ASYNCHRONOUS COMMUNICATIONS ELEMENT

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## electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
$V_{OH}^{\ddagger}$ High-level output voltage	$I_{OH} = -1 \text{ mA}$	2.4			V
$V_{OL}^{\ddagger}$ Low-level output voltage	$I_{OL} = 1.6 \text{ mA}$			0.4	V
$I_{lkg}$ Input leakage current	$V_{CC} = 5.25 \text{ V}$ , $V_I = 0 \text{ to } 5.25 \text{ V}$ , $V_{SS} = 0$ , All other pins floating			$\pm 10$	$\mu\text{A}$
$I_{OZ}$ High-impedance output current	$V_{CC} = 5.25 \text{ V}$ , $V_O = 0 \text{ to } 5.25 \text{ V}$ , Chip selected in write mode or chip deselected			$\pm 20$	$\mu\text{A}$
$I_{CC}$ Supply current	$V_{CC} = 5.25 \text{ V}$ , SIN, DSR, DCD, CTS, and $\overline{RI}$ at 2 V, All other inputs at 0.8 V, XTAL1 at 4 MHz, No load on outputs, Baud rate = 50 kilobits per second			10	mA
$C_{XIN}$ Clock input capacitance			15	20	pF
$C_{XOUT}$ Clock output capacitance			20	30	pF
$C_i$ Input capacitance			6	10	pF
$C_o$ Output capacitance			10	20	pF

† All typical values are at  $V_{CC} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

‡ These parameters apply for all outputs except XOUT.

## system timing requirements over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	ALT. SYMBOL	FIGURE	MIN	MAX	UNIT
$t_{cR}$ Cycle time, read ( $t_{w7} + t_{d8} + t_{d9}$ )	RC		175		ns
$t_{cW}$ Cycle time, write ( $t_{w6} + t_{d5} + t_{d6}$ )	WC		175		ns
$t_{w5}$ Pulse duration, address strobe low	$t_{ADS}$	2, 3	15		ns
$t_{w6}$ Pulse duration, write strobe	$t_{WR}$	2	80		ns
$t_{w7}$ Pulse duration, read strobe	$t_{RD}$	3	80		ns
$t_{w8}$ Pulse duration, master reset	$t_{MR}$		1		ns
$t_{su1}$ Setup time, address	$t_{AS}$	2, 3	15		ns
$t_{su2}$ Setup time, chip select	$t_{CS}$	2, 3	15		ns
$t_{su3}$ Setup time, data	$t_{DS}$	2	15		ns
$t_{h1}$ Hold time, address	$t_{AH}$	2, 3	0		ns
$t_{h2}$ Hold time, chip select	$t_{CH}$	2, 3	0		ns
$t_{h3}$ Hold time, write to chip select	$t_{WCS}$	2	20		ns
$t_{h4}$ Hold time, write to address	$t_{WA}$	2	20		ns
$t_{h5}$ Hold time, data	$t_{DH}$	2	15		ns
$t_{h6}$ Hold time, read to chip select	$t_{RCS}$	3	20		ns
$t_{h7}$ Hold time, read to address	$t_{RA}$	3	20		ns
$t_{d4}^{\S}$ Delay time, select to write	$t_{CSW}$	2	15		ns
$t_{d5}^{\S}$ Delay time, address to write	$t_{AW}$	2	15		ns
$t_{d6}^{\S}$ Delay time, write cycle	$t_{WC}$	2	80		ns
$t_{d7}^{\S}$ Delay time, chip select to read	$t_{CSR}$	3	15		ns
$t_{d8}^{\S}$ Delay time, address to read	$t_{AR}$	3	15		ns
$t_{d9}$ Delay time, read cycle	$t_{RC}$	3	80		ns

§ Only applies when ADS is low.



# TL16C550A

## ASYNCHRONOUS COMMUNICATIONS ELEMENT

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**system switching characteristics over recommended ranges of supply voltage and operating free-air temperature (see Note 2)**

PARAMETER	ALT. SYMBOL	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>w1</sub> Pulse duration, clock high	t <sub>XH</sub>	1	f = 9 MHz maximum	50		ns
t <sub>w2</sub> Pulse duration, clock low	t <sub>XL</sub>	1	f = 9 MHz maximum	50		ns
t <sub>d10</sub> Delay time, read to data	t <sub>RVD</sub>	3	C <sub>L</sub> = 100 pF		60	ns
t <sub>d11</sub> Delay time, read to floating data	t <sub>HZ</sub>	3	C <sub>L</sub> = 100 pF	0	60	ns
t <sub>dis(R)</sub> Read to driver disable	t <sub>RDD</sub>	3	C <sub>L</sub> = 100 pF		60	ns

**baud generator switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

PARAMETER	ALT. SYMBOL	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>w3</sub> Pulse duration, $\overline{\text{BAUDOOUT}}$ low	t <sub>LW</sub>	1	f = 9 MHz, CLK + 2, C <sub>L</sub> = 100 pF	80		ns
t <sub>w4</sub> Pulse duration, $\overline{\text{BAUDOOUT}}$ low	t <sub>HW</sub>	1	f = 9 MHz, CLK + 2, C <sub>L</sub> = 100 pF	100		ns
t <sub>d1</sub> Delay time, $\overline{\text{BAUDOOUT}}$ low to high	t <sub>BLD</sub>	1	C <sub>L</sub> = 100 pF		125	ns
t <sub>d2</sub> Delay time, $\overline{\text{BAUDOOUT}}$ high to low	t <sub>BHD</sub>	1	C <sub>L</sub> = 100 pF		125	ns

**receiver switching characteristics over recommended ranges of supply voltage and operating free-air temperature (see Note 3)**

PARAMETER	ALT. SYMBOL	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>d12</sub> Delay time, RCLK to sample	t <sub>SCD</sub>	4			100	ns
t <sub>d13</sub> Delay time, stop to set interrupt or read RBR to LSI interrupt	t <sub>SINT</sub>	4,5,6,7,8			1	RCLK cycles
t <sub>d14</sub> Delay time, read RBR/LSR to reset interrupt	t <sub>RINT</sub>	4,5,6,7,8	C <sub>L</sub> = 100 pF		150	ns

NOTES: 2. Charge and discharge time is determined by V<sub>OL</sub>, V<sub>OH</sub>, and external loading.

3. In FIFO mode RC = 425 ns (minimum) between reads of the receiver FIFO and the status registers (interrupt identification register or line status register).

**transmitter switching characteristics over recommended ranges of supply voltage and operating free-air temperature**

PARAMETER	ALT. SYMBOL	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
t <sub>d15</sub> Delay time, initial write to transmit start	t <sub>IRS</sub>	9		8	24	baudout cycles
t <sub>d16</sub> Delay time, stop to interrupt	t <sub>STI</sub>	9		8	8	baudout cycles
t <sub>d17</sub> Delay time, write THR to reset interrupt	t <sub>HR</sub>	9	C <sub>L</sub> = 100 pF		140	ns
t <sub>d18</sub> Delay time, initial write to interrupt (THRE)	t <sub>SI</sub>	9		16	32	baudout cycles
t <sub>d19</sub> Delay time, read IIR to reset interrupt (THRE)	t <sub>IR</sub>	9	C <sub>L</sub> = 100 pF		140	ns
t <sub>d20</sub> Delay time, write to TXRDY inactive	t <sub>WXI</sub>	10,11	C <sub>L</sub> = 100 pF		195	ns
t <sub>d21</sub> Delay time, start to TXRDY active	t <sub>SXA</sub>	10,11	C <sub>L</sub> = 100 pF		8	baudout cycles



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modem control switching characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	ALT. SYMBOL	FIGURE	TEST CONDITIONS	MIN	MAX	UNIT
$t_{d22}$	Delay time, write MCR to output	$t_{MDO}$	12	$C_L = 100 \text{ pF}$	100	ns
$t_{d23}$	Delay time, modem interrupt to set interrupt	$t_{SIM}$	12	$C_L = 100 \text{ pF}$	170	ns
$t_{d24}$	Delay time, read MSR to reset interrupt	$t_{RIM}$	12	$C_L = 100 \text{ pF}$	140	ns

## PARAMETER MEASUREMENT INFORMATION

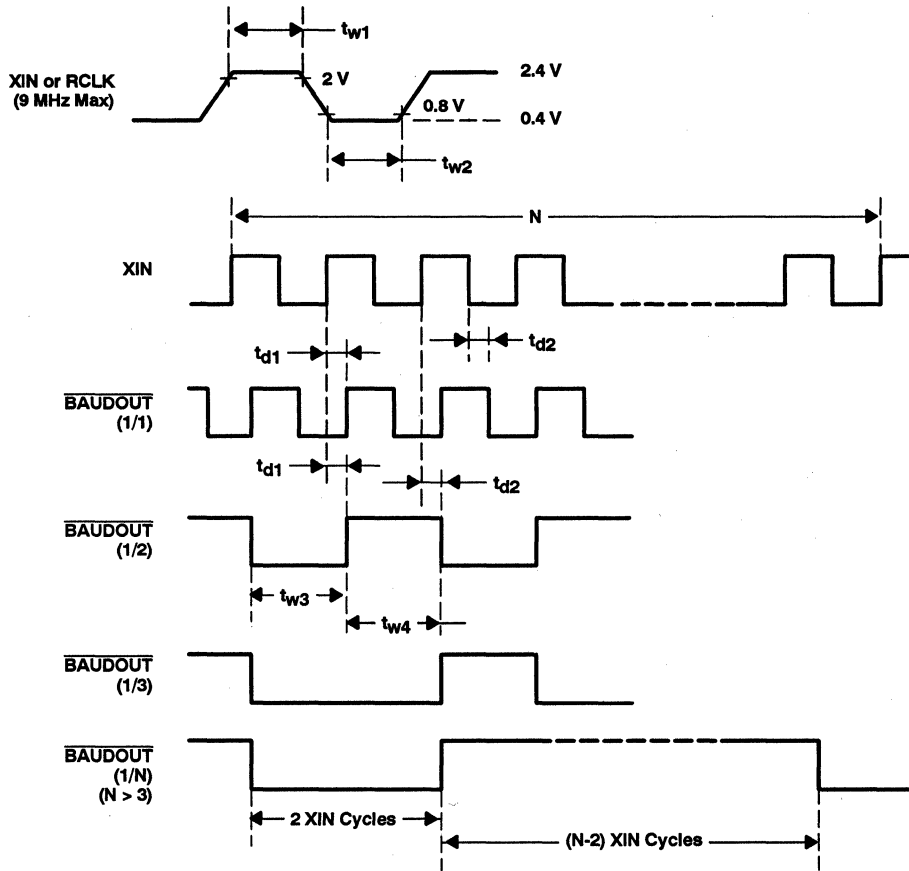
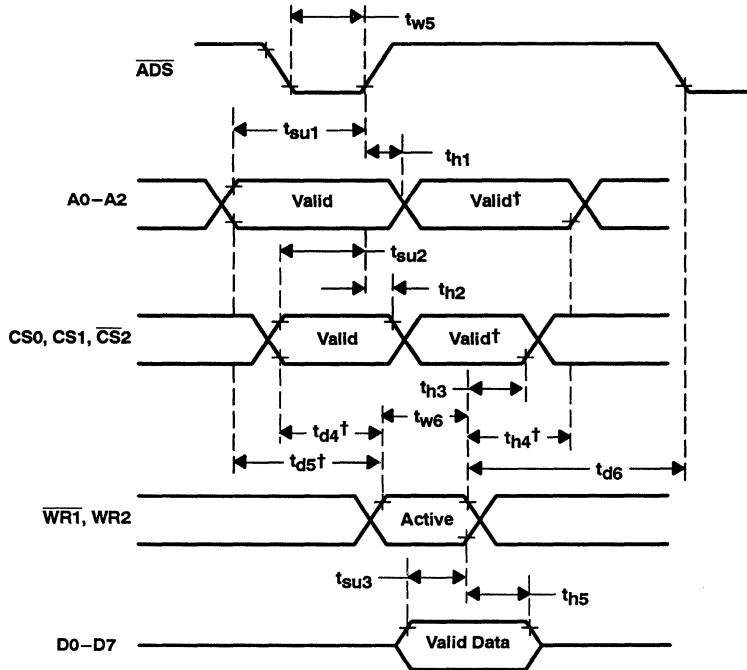


Figure 1. Baud Generator Timing

PARAMETER MEASUREMENT INFORMATION



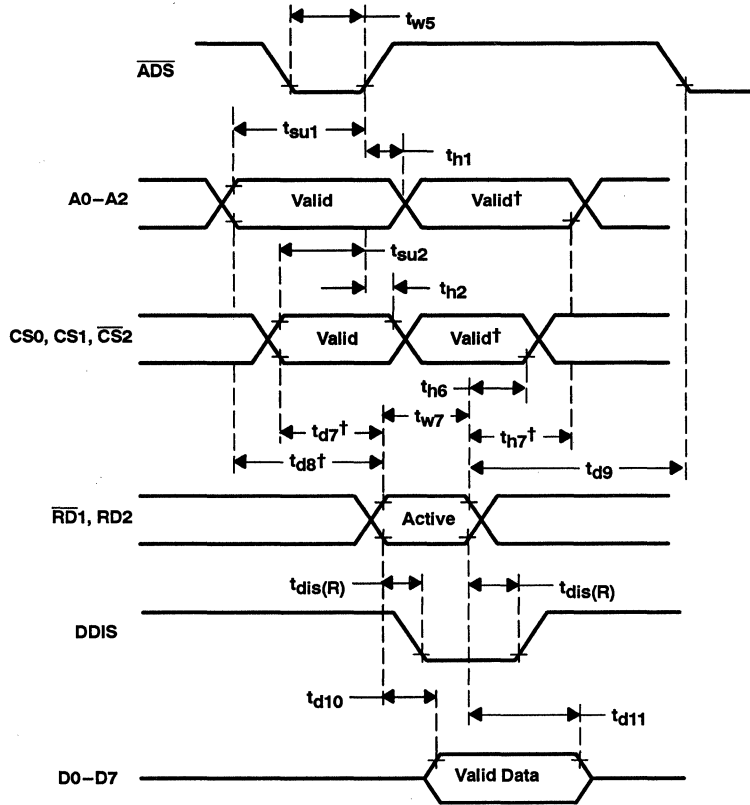
† Applicable only when  $\overline{ADS}$  is tied low.

Figure 2. Write Cycle Timing

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**PARAMETER MEASUREMENT INFORMATION**



† Applicable only when  $\overline{ADS}$  is tied low.

**Figure 3. Read Cycle Timing**

PARAMETER MEASUREMENT INFORMATION

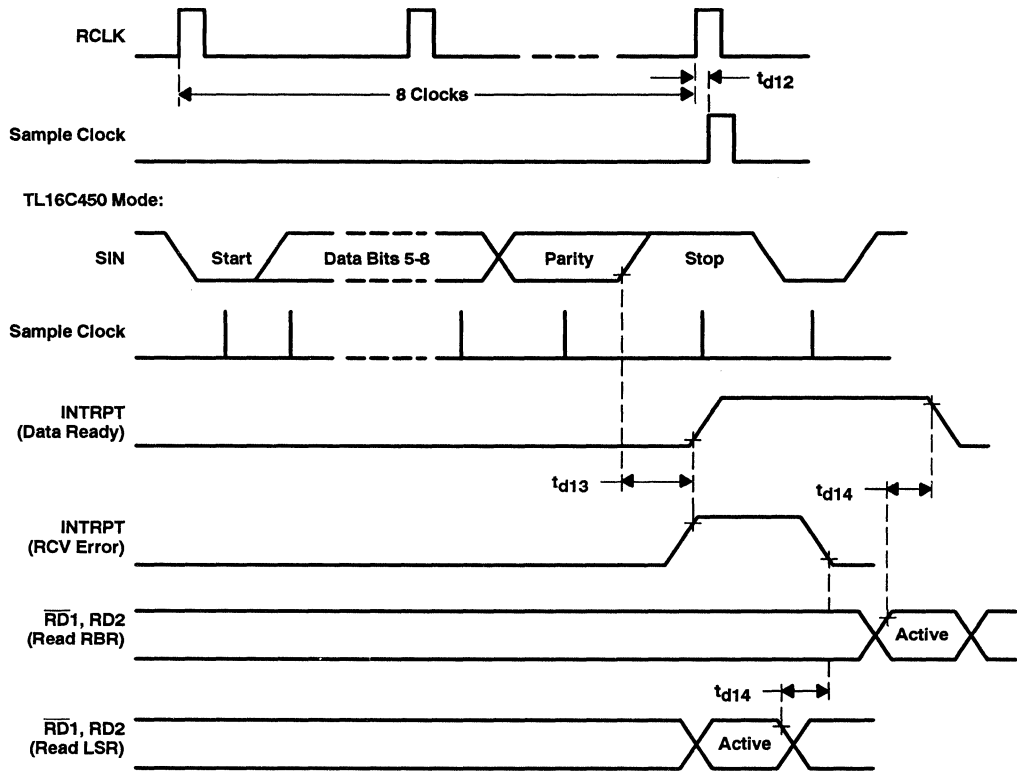


Figure 4. Receiver Timing



# TL16C550A ASYNCHRONOUS COMMUNICATIONS ELEMENT

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## PARAMETER MEASUREMENT INFORMATION

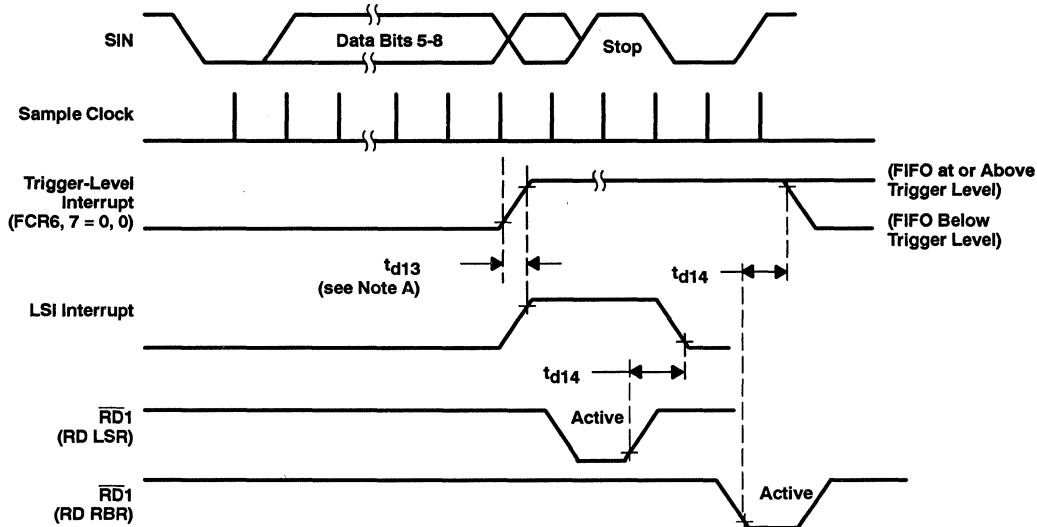


Figure 5. Receiver FIFO First Byte (Sets DR Bit)

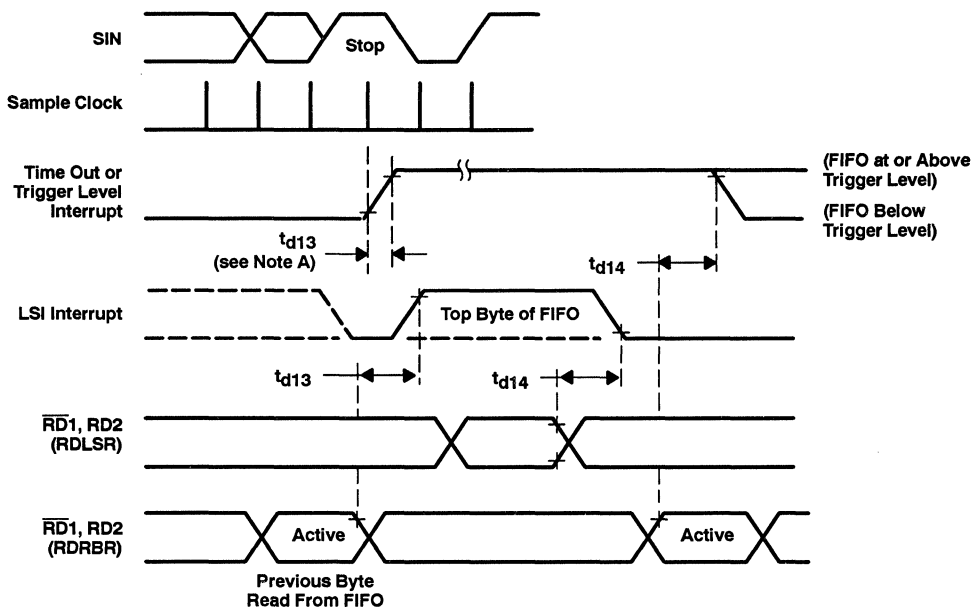


Figure 6. Receiver FIFO Bytes Other Than the First Byte (DR Internal Bit Already Set)

NOTE A: For a timeout interrupt,  $t_{d13} = 8$  RCLKs.

PARAMETER MEASUREMENT INFORMATION

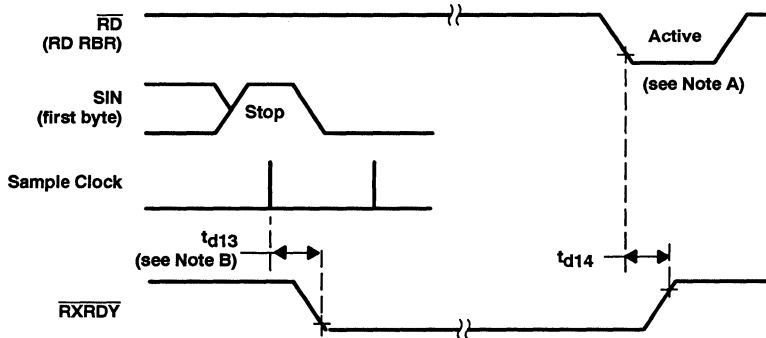


Figure 7. Receiver Ready ( $\overline{RXRDY}$ ), FCR0 = 0 or FCR0 = 1 and FCR3 = 0 (mode 0)

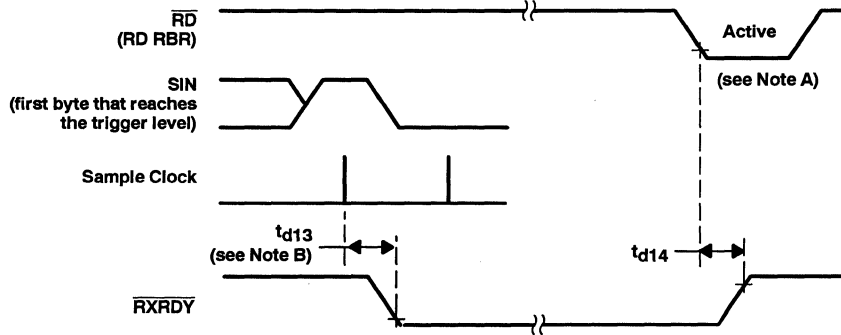


Figure 8. Receiver Ready ( $\overline{RXRDY}$ ), FCR = 1 or FCR3 = 1 (mode 1)

- NOTES: A. This is the reading of the last byte in the FIFO.  
 B. For a timeout interrupt,  $t_{d13} = 8$  RCLKs.

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## PARAMETER MEASUREMENT INFORMATION

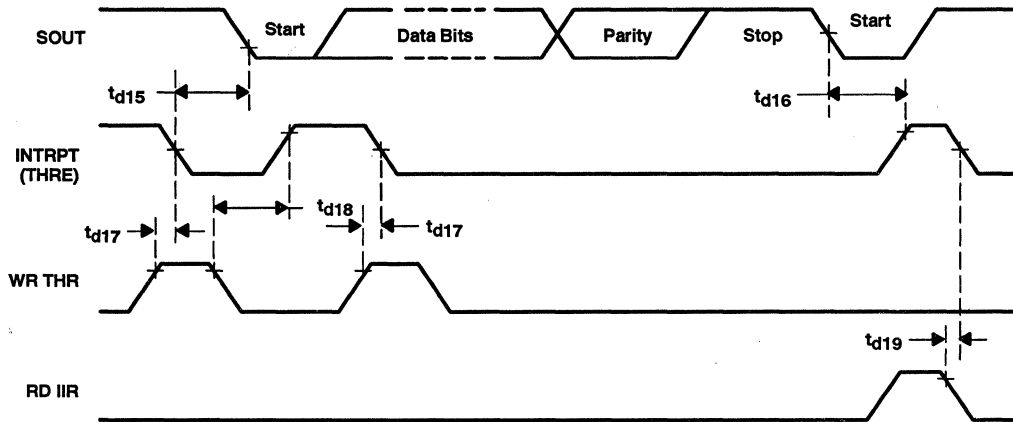


Figure 9. Transmitter Timing

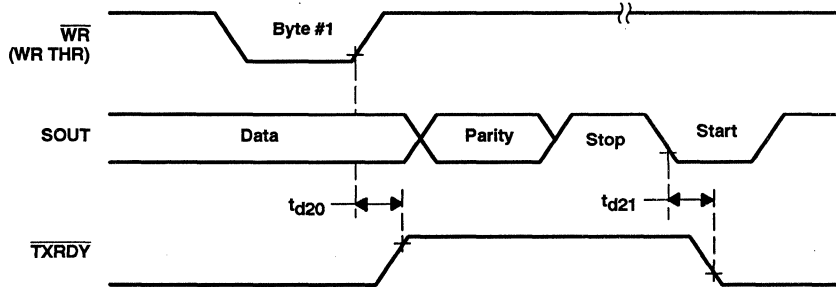


Figure 10. Transmitter Ready ( $\overline{\text{TXRDY}}$ ), FCR0 = 0 or FCR0 = 1 and FCR3 = 0 (mode 0)

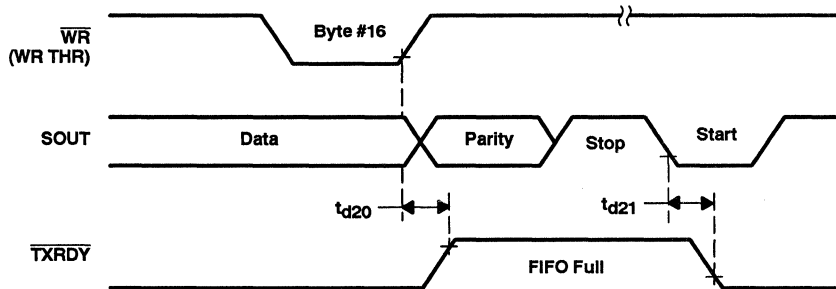


Figure 11. Transmitter Ready ( $\overline{\text{TXRDY}}$ ), FCR0 = 1 and FCR3 = 1 (mode 1)

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## PARAMETER MEASUREMENT INFORMATION

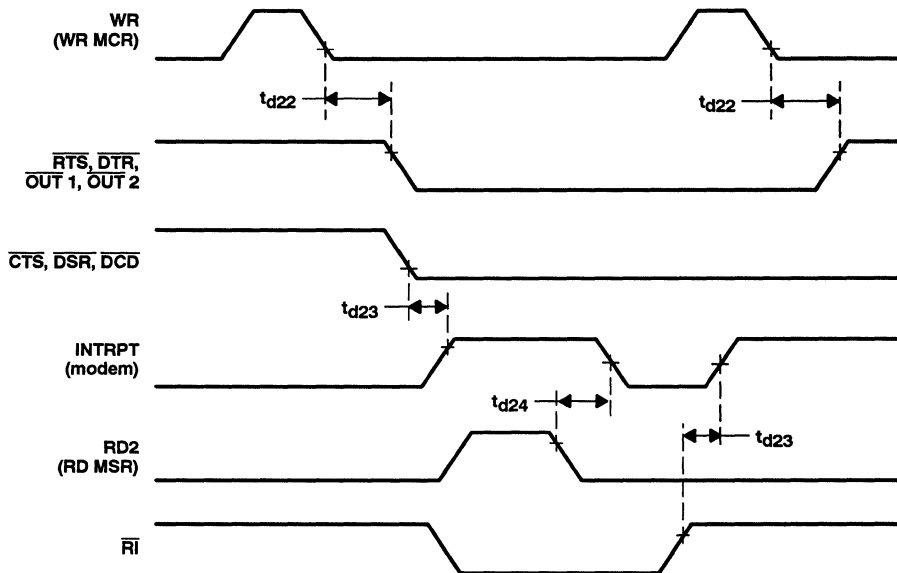


Figure 12. Modem Control Timing

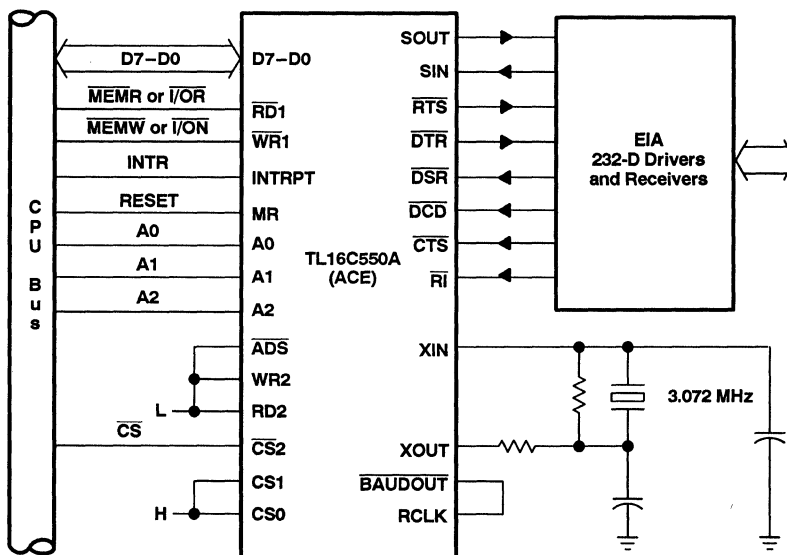


Figure 13. Basic TL16C550A Configuration

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## APPLICATION INFORMATION

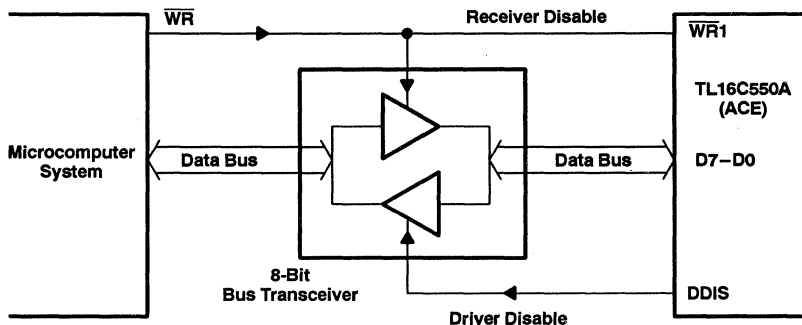


Figure 14. Typical Interface for a High-Capacity Data Bus

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## APPLICATION INFORMATION

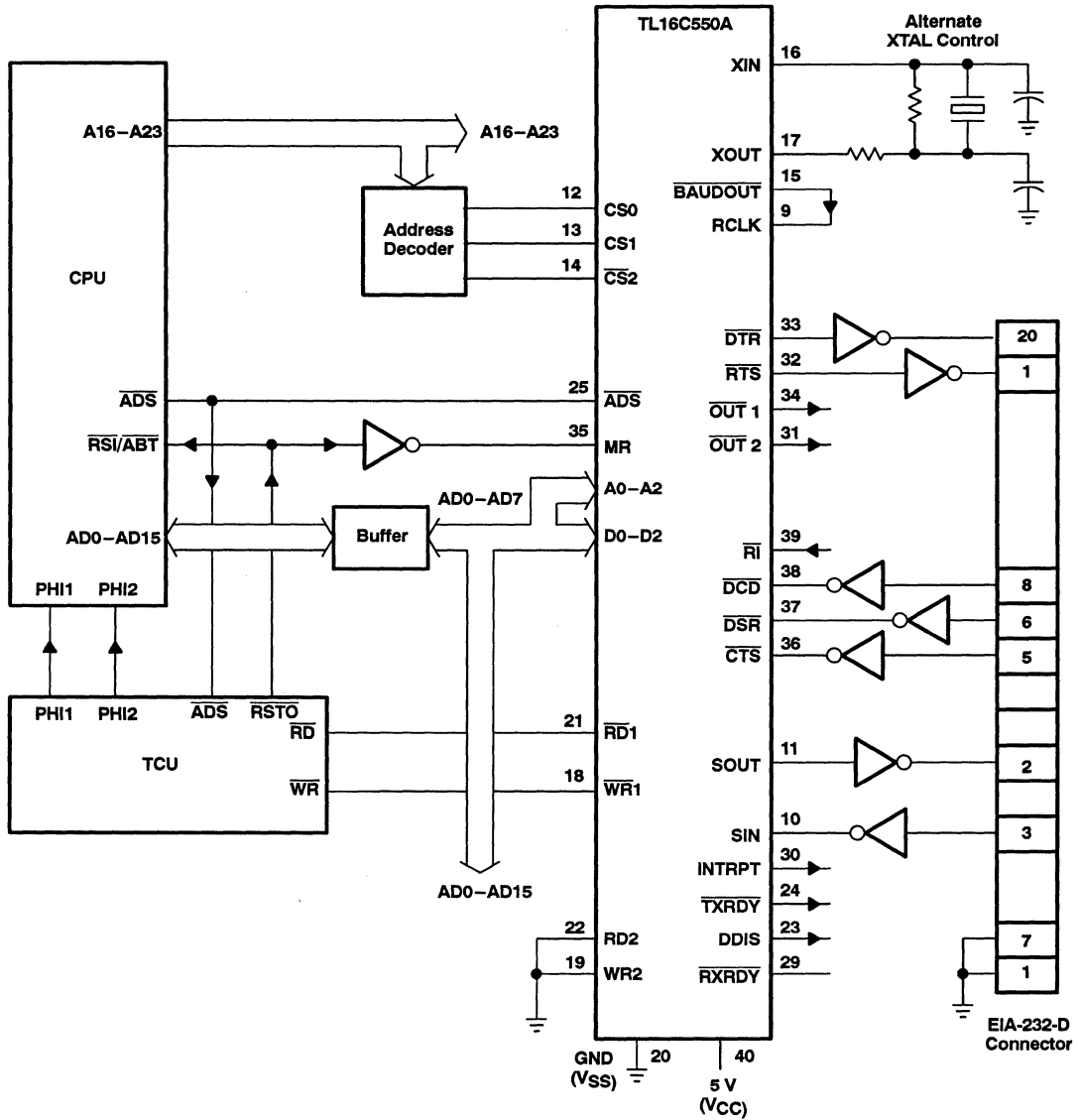


Figure 15. Typical TL16C550A Connection to a CPU

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## PRINCIPLES OF OPERATION

**Table 1. Register Selection**

DLAB†	A2	A1	A0	REGISTER
0	L	L	L	Receiver buffer (read), transmitter holding register (write)
0	L	L	H	Interrupt enable
X	L	H	L	Interrupt identification (read only)
X	L	H	L	FIFO control (write)
X	L	H	H	Line control
X	H	L	L	Modem control
X	H	L	H	Line status
X	H	H	L	Modem status
X	H	H	H	Scratch
1	L	L	L	Divisor latch (LSB)
1	L	L	H	Divisor latch (MSB)

† The divisor latch access bit (DLAB) is the most significant bit of the line control register. The DLAB signal is controlled by writing to this bit location (see Table 3).

**Table 2. Ace Reset Functions**

REGISTER/SIGNAL	RESET CONTROL	RESET STATE
Interrupt Enable Register	Master Reset	All bits low (0-3 forced and 4-7 permanent)
Interrupt Identification Register	Master Reset	Bit 0 is high, bits 1-3 are low, and bits 4-7 are permanently low
FIFO Control Register	Master Reset	All bits low
Line Control Register	Master Reset	All bits low
Modem Control Register	Master Reset	All bits low (5-7 permanent)
Line Status Register	Master Reset	Bits 5 and 6 are high, all other bits are low
Modem Status Register	Master Reset	Bits 0-3 are low, bits 4-7 are input signals
SOUT	Master Reset	High
INTRPT (receiver error flag)	Read LSR/MR	Low
INTRPT (received data available)	Read RBR/MR	Low
INTRPT (transmitter holding register empty)	Read IR/Write THR/MR	Low
INTRPT (modem status changes)	Read MSR/MR	Low
OUT 2	Master Reset	High
RTS	Master Reset	High
DTR	Master Reset	High
OUT 1	Master Reset	High
Scratch Register	Master Reset	No effect
Divisor Latch (LSB and MSB) Registers	Master Reset	No effect
Receiver Buffer Registers	Master Reset	No effect
Transmitter Holding Registers	Master Reset	No effect
RCVR FIFO	MR/FCR1-FCR0/ ΔFCR0	All bits low
XMIT FIFO	MR/FCR2-FCR0/ ΔFCR0	All bits low

**PRINCIPLES OF OPERATION**

**accessible registers**

The system programmer, via the CPU, has access to and control over any of the ACE registers that are summarized in Table 3. These registers are used to control ACE operations, receive data, and transmit data. Descriptions of these registers follow Table 3.

**Table 3. Summary of Accessible Registers**

Bit No.	REGISTER ADDRESS											
	0 DLAB = 0	0 DLAB = 0	1 DLAB = 0	2	2	3	4	5	6	7	0 DLAB = 1	1 DLAB = 1
	Receiver Buffer Register (Read Only)	Transmitter Holding Register (Write Only)	Interrupt Enable Register	Interrupt Ident. Register (Read Only)	FIFO Control Register (Write Only)	Line Control Register	Modem Control Register	Line Status Register	Modem Status Register	Scratch Register	Divisor Latch (LSB)	Latch (MSB)
RBR	THR	IER	IIR	FCR	LCR	MCR	LSR	MSR	SCR	DLL	DLM	
0	Data Bit 0†	Data Bit 0	Enable Received Data Available Interrupt (ERB)	*0† if Interrupt Pending	FIFO Enable	Word Length Select Bit 0 (WLS0)	Data Terminal Ready (DTR)	Data Ready (DR)	Delta Clear to Send (ΔCTS)	Bit 0	Bit 0	Bit 8
1	Data Bit 1	Data Bit 1	Enable Transmitter Holding Register Empty Interrupt (ETBEI)	Interrupt ID Bit 0	Receiver FIFO Reset	Word Length Select Bit 1 (WLS1)	Request to Send (RTS)	Overrun Error (OE)	Delta Data Set Ready (ΔDSR)	Bit 1	Bit 1	Bit 9
2	Data Bit 2	Data Bit 2	Enable Receiver Line Status Interrupt (ELSI)	Interrupt ID Bit (1)	Transmitter FIFO Reset	Number of Stop Bits (STB)	Out 1	Parity Error (PE)	Trailing Edge Fling Indicator (TEFI)	Bit 2	Bit 2	Bit 10
3	Data Bit 3	Data Bit 3	Enable Modem Status Interrupt (EDSSI)	Interrupt ID Bit (2) (Note 4)	DMA Mode Select	Parity Enable (PEN)	Out 2	Framing Error (FE)	Delta Data Carrier Detect (ΔDCD)	Bit 3	Bit 3	Bit 11
4	Data Bit 4	Data Bit 4	0	0	Reserved	Even Parity Select (EPS)	Loop	Break Interrupt (BI)	Clear to Send (CTS)	Bit 4	Bit 4	Bit 12
5	Data Bit 5	Data Bit 5	0	0	Reserved	Stick Parity	0	Transmitter Holding Register (THRE)	Data Set Ready (DSR)	Bit 5	Bit 5	Bit 13
6	Data Bit 6	Data Bit 6	0	FIFOs Enabled (Note 4)	Receiver Trigger (LSB)	Set Break	0	Transmitter Empty (TEMT)	Ring Indicator (RI)	Bit 6	Bit 6	Bit 14
7	Data Bit 7	Data Bit 7	0	FIFOs Enabled (Note 4)	Receiver Trigger (MSB)	Divisor Latch Access Bit (DLAB)	0	Error in RCVR FIFO (Note 4)	Data Carrier Detect (DCD)	Bit 7	Bit 7	Bit 15

† Bit 0 is the least significant bit. It is the first bit serially transmitted or received.

NOTE 4: These bits are always 0 in the TL16C450 mode.



# TL16C550A

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### PRINCIPLES OF OPERATION

#### receiver buffer register (RBR)

The ACEs receiver section consists of a receiver shift register (RSR) and a receiver buffer register (RBR). The RBR is actually a 16-byte FIFO. Timing is supplied by the 16X Receiver Clock (RCLK). Receiver section control is a function of the ACEs line control register.

The ACEs RSR receives serial data from the serial input (SIN) pin. The RSR then deserializes the data and moves it into the RBR FIFO. In the TL16C450 mode, when a character is placed in the receiver buffer register and the received data available interrupt is enabled, an interrupt is generated. This interrupt is cleared when the data is read out of the receiver buffer register. In the FIFO mode, the interrupts are generated based on the control setup in the FIFO control register.

#### transmitter holding register (THR)

The ACEs transmitter section consists of a transmitter holding register (THR) and a transmitter shift register (TSR). The THR is actually a 16-byte FIFO. Timing is supplied by the baud out (BAUDOUT) clock signal. Transmitter section control is a function of the ACE's line control register.

The ACE THR receives data off the internal data bus and, when the shift register is idle, moves it into the TSR. The TSR serializes the data and outputs it at the serial output (SOUT). In the TL16C450 mode, if the THR is empty and the transmitter holding register empty (THRE) interrupt is enabled, an interrupt is generated. This interrupt is cleared when a character is loaded into the register. In the FIFO mode, the interrupts are generated based on the control setup in the FIFO control register.

#### interrupt enable register (IER)

The Interrupt enable register enables each of the five types of interrupts (refer to Table 4) and the INTRPT output signal in response to an interrupt generation. The interrupt enable register can also be used to disable the interrupt system by setting bits 0 through 3 to logic 0. The contents of this register are summarized in Table 3 and are described below.

Bit 0. This bit, when set to logic 1, enables the received data available interrupt.

Bit 1. This bit, when set to logic 1, enables the transmitter holding register empty interrupt.

Bit 2. This bit, when set to logic 1, enables the receiver line status interrupt.

Bit 3. This bit, when set to logic 1, enables the modem status interrupt.

Bits 4 thru 7. Bits 4 through 7 in the interrupt enable register are not used and are always set to logic 0.

#### FIFO control register

The FIFO control register (FCR) is a write-only register at the same location as the IIR, which is a read-only register. The FCR is used to enable the FIFOs, clear the FIFOs, set the receiver FIFO trigger level, and select the type of DMA signalling.

Bit 0. FCR0, when set to logic 1, enables the transmit and receive FIFOs. This bit must be a 1 when other FCR bits are written to or they will not be programmed. Changing this bit clears the FIFOs.

Bit 1. FCR1, when set to logic 1, clears all bytes in the receiver FIFO and resets its counter logic to 0. The shift register is not cleared. The 1 that is written to this bit position is self clearing.

Bit 2. FCR2, when set to logic 1, clears all bytes in the transmit FIFO and resets its counter to 0. The shift register is not cleared. The 1 that is written to this bit position is self clearing.

Bit 3. If FCR0 is a 1, setting FCR3 to a 1 causes the  $\overline{RXRDY}$  and  $\overline{TXRDY}$  to change from mode 0 to mode 1.



**PRINCIPLES OF OPERATION**

Bits 4 and 5. FCR4 and FCR5 are reserved for future use.

Bits 6 and 7. FCR6 and FCR7 are used to set the trigger level for the receiver FIFO interrupt.

BIT 7	BIT 6	RECEIVER FIFO TRIGGER LEVEL (BYTES)
0	0	01
0	1	04
1	0	08
1	1	14

**interrupt identification register (IIR)**

The ACE has an on-chip interrupt generation and prioritization capability that permits a flexible interface with most popular microprocessors.

The ACE provides four prioritized levels of interrupts:

- Priority 1—Receiver line status (highest priority)
- Priority 2—Receiver data ready or receiver character timeout
- Priority 3—Transmitter holding register empty
- Priority 4—Modem status (lowest priority)

When an interrupt is generated, the interrupt identification register indicates that an interrupt is pending and the type of that interrupt in its three least significant bits (bits 0, 1, and 2). The contents of this register are summarized in Table 3 and described in Table 4. Detail on each bit are as follows:

Bit 0. This bit can be used either in a hardware-prioritized, or polled interrupt system. If this bit is a logic 0, an interrupt is pending. When bit 0 is a logic 1, no interrupt is pending.

Bits 1 and 2. These two bits are used to identify the highest priority interrupt pending, as indicated in Table 4.

Bit 3. This bit is always 0 in the TL16C450 mode. In FIFO mode, this bit is set with bit 2 to indicate that a timeout interrupt is pending.

Bits 4 thru 5. These two bits are not used and are always set at logic 0.

Bits 6 and 7. These two bits are always 0 in the TL16C450 mode. They are set when bit 0 of the FIFO control register is equal to 1.

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Table 4. Interrupt Control Functions

INTERRUPT IDENTIFICATION REGISTER				PRIORITY LEVEL	INTERRUPT TYPE	INTERRUPT SOURCE	INTERRUPT RESET METHOD
BIT 3	BIT 2	BIT 1	BIT 0				
0	0	0	1	None	None	None	-
0	1	1	0	1	Receiver line status	Overrun error, parity error, framing error or break interrupt	Reading the line status register
0	1	0	0	2	Received data available	Receiver data available in the TL16C450 mode or trigger level reached in the FIFO mode.	Reading the receiver buffer register
1	1	0	0	2	Character timeout indication	No characters have been removed from or input to the receiver FIFO during the last four character times and there is at least one character in it during this time	Reading the receiver buffer register
0	0	1	0	3	Transmitter holding register empty	Transmitter holding register empty	Reading the interrupt identification register (if source of interrupt) or writing into the transmitter holding register
0	0	0	0	4	Modem status	Clear to send, data set ready, ring indicator, or data carrier detect	Reading the modem status register

### line control register (LCR)

The system programmer controls the format of the asynchronous data communication exchange through the line control register. In addition, the programmer is able to retrieve, inspect, and modify the contents of the line control register; this eliminates the need for separate storage of the line characteristics in system memory. The contents of this register are summarized in Table 3 and described below.

Bits 0 and 1. These two bits specify the number of bits in each transmitted or received serial character. These bits are encoded as follows:

Bit 1	Bit 0	Word Length
0	0	5 Bits
0	1	6 Bits
1	0	7 Bits
1	1	8 Bits

**PRINCIPLES OF OPERATION**

Bit 2. This bit specifies either one, one and one-half, or two stop bits in each transmitted character. If bit 2 is a logic 0, one stop bit is generated in the data. If bit 2 is a logic 1, the number of stop bits generated is dependent on the word length selected with bits 0 and 1. The receive clocks the first stop bit only, regardless of the number of stop bits selected. The number of stop bits generated, in relation to word length and bit 2, is shown in the following.

Bit 2	Word Length Selected by Bits 1 and 2	Number of Stop Bits Generated
0	Any word length	1
1	5 bits	1 1/2
1	6 bits	2
1	7 bits	2
1	8 bits	2

Bit 3. This bit is the parity enable bit. When bit 3 is a logic 1, a parity bit is generated in transmitted data between the last data word bit and the first stop bit. In received data, if bit 3 is a logic 1, parity is checked. When bit 3 is a logic 0, no parity is generated or checked.

Bit 4. Bit 4 is the even parity select bit. When parity is enabled by bit 3: a logic 1 in bit 4 produces even parity (an even number of logic 1's in the data and parity bits) and a logic 0 in bit 4 produces odd parity (an odd number of logic 1's).

Bit 5. This is the stick parity bit. When bits 3, 4, and 5 are logic 1s, the parity bit is transmitted and checked as a logic 0. When bits 3 and 5 are logic 1's and bit 4 is a logic 0, the parity bit is transmitted and checked as a logic 1. If bit 5 is a logic 0, stick parity is disabled.

Bit 6. This bit is the break control bit. Bit 6 is set to a logic 1 to force a break condition, i.e., a condition where the serial output (SOUT) pin is forced to the spacing (logic 0) state. When bit 6 is set to a logic 0, the break condition is disabled. The break condition has no effect on the transmitter logic; it only effects the serial output.

Bit 7. This bit is the divisor latch access bit (DLAB). Bit 7 must be set to a logic 1 to access the divisor latches of the baud generator during a read or write. Bit 7 must be set to a logic 0 during a read or write to access the receiver buffer, the transmitter holding register, or the interrupt enable register.

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## PRINCIPLES OF OPERATION

### modem control register (MCR)

The modem control register is an 8-bit register that controls an interface with a modem, data set, or peripheral device that is emulating a modem. The contents of this register are summarized in Table 3 and are described below.

Bit 0. Bit 0 (DTR) controls the data terminal ready ( $\overline{\text{DTR}}$ ) output. Setting this bit to a logic 1 forces the  $\overline{\text{DTR}}$  output to its low state. When bit 0 is set to a logic 0,  $\overline{\text{DTR}}$  goes high.

Bit 1. Bit 1 (RTS) controls the request to send ( $\overline{\text{RTS}}$ ) output in a manner identical to Bit 0's control over the  $\overline{\text{DTR}}$  output.

Bit 2. Bit 2 (OUT 1) controls the output 1 ( $\overline{\text{OUT 1}}$ ) signal, a user-designated output signal, in a manner identical to bit 0's control over the  $\overline{\text{DTR}}$  output.

Bit 3. Bit 3 (OUT 2) controls the output 2 ( $\overline{\text{OUT 2}}$ ) signal, a user-designated output signal, in a manner identical to bit 0's control over the  $\overline{\text{DTR}}$  output.

Bit 4. Bit 4 provides a local loopback feature for diagnostic testing of the ACE. When this bit is set to a logic high, the following occurs:

1. The transmitter serial output (SOUT) is set high.
2. The receiver serial input (SIN) is disconnected.
3. The output of the transmitter shift register is looped back into the receiver shift register input.
4. The four modem control inputs ( $\overline{\text{CTS}}$ ,  $\overline{\text{DSR}}$ ,  $\overline{\text{DCD}}$ , and  $\overline{\text{RI}}$ ) are disconnected.
5. The four modem control outputs (DTR, RTS, OUT 1, and OUT 2) are internally connected to the four modem control inputs.
6. The four modem control output pins are forced to their inactive states (high).

In the diagnostic mode, data that is transmitted is immediately received. This allows the processor to verify the transmit- and receive-data paths to the ACE. The receiver and transmitter interrupts are fully operational. The modem control interrupts are also operational but the modem control interrupt's sources are now the lower four bits of the modem control register instead of the four modem control inputs. All interrupts are still controlled by the interrupt enable register.

Bit 5 through 7. These bits are permanently set to logic 0.

### line status register (LSR)<sup>†</sup>

The line status register provides information to the CPU concerning the status of data transfers. The contents of this register are described below and summarized in Table 3.

Bit 0. Bit 0 is the data ready (DR) indicator for the receiver. This bit is set to a logic 1 condition whenever a complete incoming character has been received and transferred into the receiver buffer register or the FIFO and is reset to logic 0 by reading all of the data in the receiver buffer register or the FIFO.

Bit 1<sup>‡</sup>. Bit 1 is the overrun error (OE) indicator. When this bit is set to logic 1, it indicates that before the character in the receiver buffer register was read, it was overwritten by the next character transferred into the register. The OE indicator is reset every time the CPU reads the contents of the line status register. If the FIFO mode data continues to fill the FIFO beyond the trigger level, an overrun error will occur only after the FIFO is full and the next character has been completely received in the shift register. OE is indicated to the CPU as soon as it happens. The character in the shift register is overwritten but is not transferred to the FIFO.

<sup>†</sup> The line status register is intended for read operations only; writing to this register is not recommended outside of a factory testing environment.  
<sup>‡</sup> Bits 1 through 4 are the error conditions that produce a receiver line status interrupt.

## PRINCIPLES OF OPERATION

### line status register (LSR) (continued)

Bit 2<sup>‡</sup>. Bit 2 is the parity error (PE) indicator. When this bit is set to logic 1, it indicates that the parity of the received data character does not match the parity selected in the line control register (bit 4). The PE bit is reset every time the CPU reads the contents of the line status register. In the FIFO mode, this error is associated with the particular character in the FIFO to which it applies. This error is revealed to the CPU when its associated character is at the top of the FIFO.

Bit 3<sup>‡</sup>. Bit 3 is the framing error (FE) indicator. When this bit is set to logic 1, it indicates that the received character did not have a valid (logic 1) stop bit. The FE bit is reset every time the CPU reads the contents of the line status register. In the FIFO mode, this error is associated with the particular character in the FIFO to which it applies. This error is revealed to the CPU when its associated character is at the top of the FIFO. The ACE will try to resynchronize after a framing error. To accomplish this, it is assumed that the framing error is due to the next start bit. The ACE then samples this start bit twice and then accepts the input data.

Bit 4<sup>‡</sup>. Bit 4 is the break interrupt (BI) indicator. When this bit is set to logic 1, it indicates that the received data input was held in the logic 0 state for longer than a full-word transmission time. A full-word transmission time is defined as the total time of the start, data, parity, and Stop bits. The B1 bit is reset every time the CPU reads the contents of the line status register. In the FIFO mode, this error is associated with the particular character in the FIFO to which it applies. This error is revealed to the CPU when its associated character is at the top of the FIFO. When break occurs, only one character is loaded into the FIFO. The next character transfer is enabled after SIN goes to the marking state and receives the next valid start bit.

Bit 5. Bit 5 is the transmitter holding register empty (THRE) indicator. This bit is set to logic 1 when the transmitter holding register is empty, indicating that the ACE is ready to accept a new character. If the THRE interrupt is enabled when the THRE bit is a logic 1, then an interrupt is generated. THRE is set to a logic 1 when the contents of the transmitter holding register are transferred to the transmitted shift register. This bit is reset to logic 0 concurrent with the loading of the transmitter holding register by the CPU. In the FIFO mode, this bit is set when the transmit FIFO is empty; it is cleared when at least 1 byte is written to the transmit FIFO.

Bit 6. Bit 6 is the transmitter empty (TEMT) indicator. This bit is set to a logic 1 when the transmitter holding register and the transmitter shift register are both empty. When either the transmitter holding register or the transmitter shift register contains a data character, the TEMT bit is reset to logic 0. In the FIFO mode, this bit is set to a 1 when the transmitter FIFO and shift register are both empty.

Bit 7. In the TL16C550A, this bit is always reset to logic 0. In the TL16C450 mode, this bit is always a 0. In the FIFO mode, LSR7 is set when there is at least one parity, framing, or break error in the FIFO. It is cleared when the microprocessor reads the LSR and there are no subsequent errors in the FIFO.

### modem status register (MSR)

The modem status register is an 8-bit register that provides information about the current state of the control lines from the modem, data set, or peripheral device to the CPU. Additionally, four bits of this register provides change information; when a control input from the modem changes state, the appropriate bit is set to logic 1. All four bits are reset to logic 0 when the CPU reads the modem status register. The contents of this register are summarized in Table 3 and are described below.

Bit 0. Bit 0 is the change in clear to send (DCTS) indicator. This bit indicates that the  $\overline{\text{CTS}}$  input has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

<sup>‡</sup> Bits 1 through 4 are the error conditions that produce a receiver line status interrupt.

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## PRINCIPLES OF OPERATION

### modem status register (MSR) (continued)

Bit 1. Bit 1 is the change in data set ready (DDSR) indicator. This bit indicates that the  $\overline{\text{DSR}}$  input has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 2. Bit 2 is the trailing edge of ring indicator (TERI) detector. This bit indicates that the  $\overline{\text{RI}}$  input to the chip has changed from a low to a high state. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 3. Bit 3 is the change in data carrier detect (DDCD) indicator. This bit indicates that the  $\overline{\text{DCD}}$  input to the chip has changed state since the last time it was read by the CPU. When this bit is a logic 1 and the modem status interrupt is enabled, a modem status interrupt is generated.

Bit 4. Bit 4 is the compliment of the clear to send ( $\overline{\text{CTS}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 1 (RTS).

Bit 5. Bit 5 is the compliment of the data set ready ( $\overline{\text{DSR}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control register bit 1 (DTR).

Bit 6. Bit 6 is the compliment of the ring indicator ( $\overline{\text{RI}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control registers bit 2 (OUT 1).

Bit 7. Bit 7 is the compliment of the data carrier detect ( $\overline{\text{DCD}}$ ) input. If bit 4 (loop) of the modem control register is set to a logic 1, this bit is equivalent to the modem control registers bit 3 (OUT 2).

### scratch register (SCR)

The scratch register is an 8-bit register that is intended for the programmer's use as a scratchpad in the sense that it will temporarily hold the programmer's data without affecting any other ACE operation.

### programmable baud generator

The ACE contains a programmable baud generator that takes a clock input in the range between dc and 8 MHz and divides it by a divisor in the range between 1 and  $2^{16}-1$ . The output frequency of the baud generator is sixteen times (16 X) the baud rate. The formula for the divisor is:

$$\text{divisor \#} = \text{XIN frequency input} + (\text{desired baud rate} \times 16)$$

Two 8-bit registers, called divisor latches, are used to store the divisor in a 16-bit binary format. These divisor latches must be loaded during initialization of the ACE in order to ensure desired operation of the baud generator. When either of the divisor latches is loaded, a 16-bit baud counter is also loaded to prevent long counts on initial load.

Tables 5 and 6, which follow, illustrate the use of the baud generator with crystal frequencies of 1.8432 MHz and 3.072 MHz, respectively. For baud rates of 38.4 kilobits per second and below, the error obtained is very small. The accuracy of the selected baud rate is dependent on the selected crystal frequency.

Refer to Figure 10 for examples of typical clock circuits.

## PRINCIPLES OF OPERATION

### FIFO interrupt-mode operation

When the receiver FIFO and receiver interrupts are enabled ( $FCR0 = 1$ ,  $IER0 = 1$ ) receiver interrupts will occur as follows:

1. The receive data available interrupt will be issued to the microprocessor when the FIFO has reached its programmed trigger level. It will be cleared as soon as the FIFO drops below its programmed trigger level.
2. The IIR receive data available indication also occurs when the FIFO trigger level is reached, and, like the interrupt, it is cleared when the FIFO drops below the trigger level.
3. The receiver line status interrupt ( $IIR = 06$ ), as before, has higher priority than the received data available ( $IIR = 04$ ) interrupt.
4. The data ready bit ( $LSR0$ ) is set as soon as a character is transferred from the shift register to the receiver FIFO. It is reset when the FIFO is empty.

When the receiver FIFO and receiver interrupts are enabled, receiver FIFO timeout interrupts will occur as follows:

1. FIFO timeout interrupt will occur if the following conditions exist:
  - a. At least one character is in the FIFO.
  - b. The most recent serial character received was longer than 4 continuous character times ago (if 2 stop bits are programmed, the second one is included in this time delay).
  - c. The most recent microprocessor read of the FIFO was longer than 4 continuous character times ago. This will cause a maximum character received to interrupt issued delay of 160 ms at 300 baud with 12-bit character.
2. Character times are calculated by using the RCLK input for a clock signal (this makes the delay proportional to the baud rate).
3. When a timeout interrupt has occurred, it is cleared and the timer reset when the microprocessor reads one character from the receiver FIFO.
4. When a timeout interrupt has not occurred, the timeout timer is reset after a new character is received or after the microprocessor reads the receiver FIFO.

When the transmit FIFO and transmitter interrupts are enabled ( $FCR0 = 1$ ,  $IER1 = 1$ ), transmit interrupts will occur as follows:

1. The transmitter holding register interrupt (02) occurs when the transmit FIFO is empty. It is cleared as soon as the transmitter holding register is written to (1 to 16 characters may be written to the transmit FIFO while servicing this interrupt) or the IIR is read.
2. The transmit FIFO empty indications will be delayed 1 character time minus the last stop bit time when the following occurs:  $THRE = 1$  and there have not been at least two bytes at the same time in the transmit FIFO since the last  $THRE = 1$ . The first transmitter interrupt after changing  $FCR0$  will be immediate, if it is enabled.

Character timeout and receiver FIFO trigger level interrupts have the same priority as the current received data available interrupt; transmit FIFO empty has the same priority as the current transmitter holding register empty interrupt.

### FIFO polled-mode operation

With  $FCR0 = 1$ , resetting  $IER0$ ,  $IER1$ ,  $IER2$ ,  $IER3$ , or all four to 0 puts the ACE in the FIFO polled mode of operation. Since the receiver and transmitter are controlled separately, either one or both can be in the polled mode of operation.



# TL16C550A ASYNCHRONOUS COMMUNICATIONS ELEMENT

SLLS057B – D3128, AUGUST 1989 – REVISED MARCH 1993

## PRINCIPLES OF OPERATION

### FIFO polled-mode operation (continued)

In this mode, the user program will check receiver and transmitter status via the LSR. As stated previously:

1. LSR0 will be set as long as there is one byte in the receiver FIFO.
2. LSR1 through LSR4 will specify which error(s) have occurred. Character error status is handled the same way as when in the interrupt mode, the IIR is not affected since IER2 = 0.
3. LSR5 will indicate when the transmit FIFO is empty.
4. LSR6 will indicate that both the transmit FIFO and shift registers are empty.
5. LSR7 will indicate whether there are any errors in the receiver FIFO.

There is no trigger level reached or timeout conditions indicated in the FIFO polled mode. However, the receiver and transmit FIFO,s are still fully capable of holding characters.

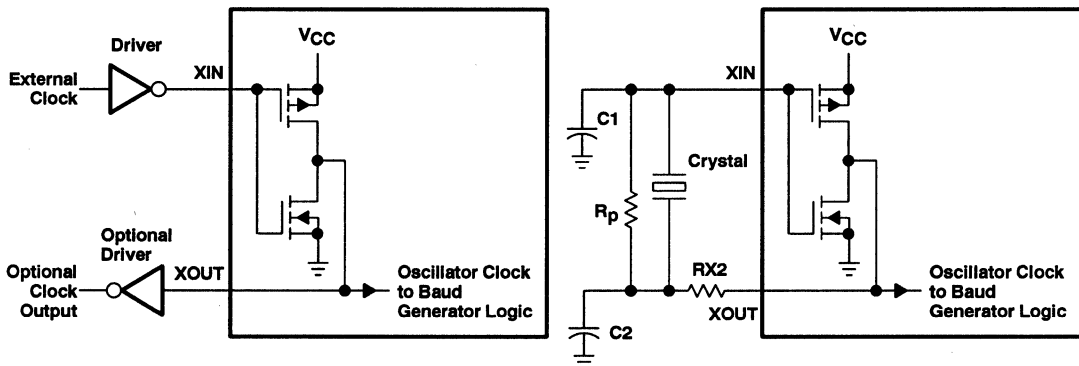
**Table 5. Baud Rates Using A 1.8432-MHz Crystal**

DESIRED BAUD RATE	DIVISOR USED TO GENERATE 16 X CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	2304	
75	1536	
110	1047	0.026
134.5	857	0.058
150	768	
300	384	
600	192	
1200	96	
1800	64	
2000	58	0.69
2400	48	
3600	32	
4800	24	
7200	16	
9600	12	
19200	6	
38400	3	
56000	2	2.86

**PRINCIPLES OF OPERATION**

**Table 6. Baud Rates Using A 3.072-MHz Crystal**

DESIRED BAUD RATE	DIVISOR USED TO GENERATE 16 X CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	3840	
75	2560	
110	1745	0.026
134.5	1428	0.034
150	1280	
300	640	
600	320	
1200	160	
1800	107	0.312
2000	96	
2400	80	
3600	53	0.628
4800	40	
7200	27	1.23
9600	20	
19200	10	
38400	5	



**TYPICAL CRYSTAL OSCILLATOR NETWORK**

CRYSTAL	R <sub>p</sub>	R <sub>X2</sub>	C <sub>1</sub>	C <sub>2</sub>
3.1 MHz	1 MΩ	1.5 kΩ	10-30 pF	40-60 pF
1.8 MHz	1 MΩ	1.5 kΩ	10-30 pF	40-60 pF

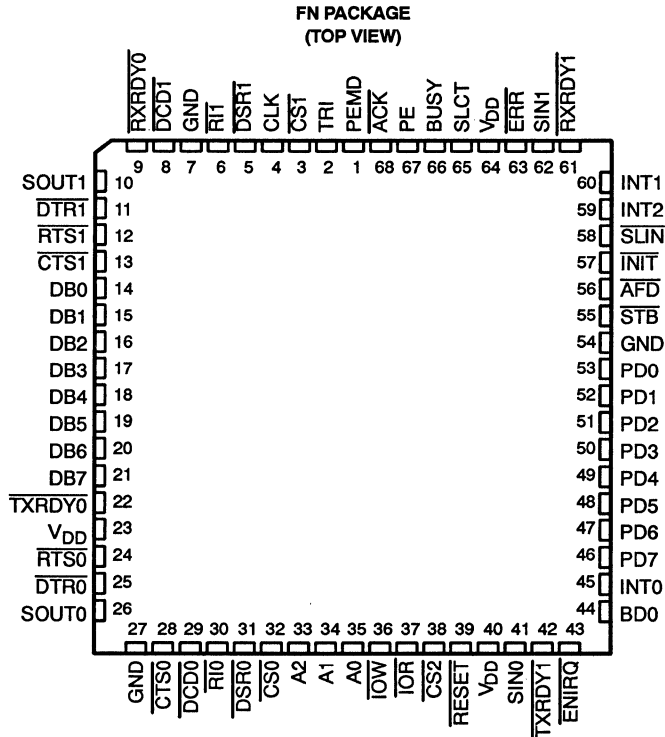
**Figure 16. Typical Clock Circuits**



# TL16C552 DUAL ASYNCHRONOUS COMMUNICATIONS ELEMENT WITH FIFO

SLLS102A – D3647, DECEMBER 1990 – REVISED MARCH 1993

- IBM PC/AT Compatible
- Two TL16C550 ACEs
- Enhanced Bidirectional Line Printer Port
- 16-Byte FIFOs Reduce CPU Interrupts
- Independent Control of Transmit, Receive, Line Status, and Data Set Interrupts on Each Channel
- Individual Modem Control Signals for Each Channel
- Programmable Serial Interface
- Characteristics for Each Channel:  
5-, 6-, 7-, or 8-bit Characters  
Even-, Odd-, or No-Parity Bit Generation and Detection  
1-, 1 1/2-, or 2-Stop Bit Generation
- 3-State TTL Drive for the Data and Control Bus on Each Channel
- Hardware and Software Compatible With TL16C452



## description

The TL16C552 is an enhanced dual-channel version of the popular TL16C550 asynchronous communications element (ACE). The device serves two serial input/output interfaces simultaneously in microcomputer or microprocessor-based systems. Each channel performs serial-to-parallel conversion on data characters received from peripheral devices or modems and parallel-to-serial conversion on data characters transmitted by the CPU. The complete status of each channel of the dual ACE can be read at any time during functional operation by the CPU. The information obtained includes the type and condition of the transfer operations being performed and the error conditions.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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# TL16C552 DUAL ASYNCHRONOUS COMMUNICATIONS ELEMENT WITH FIFO

SLLS102A - D3647, DECEMBER 1990 - REVISED MARCH 1993

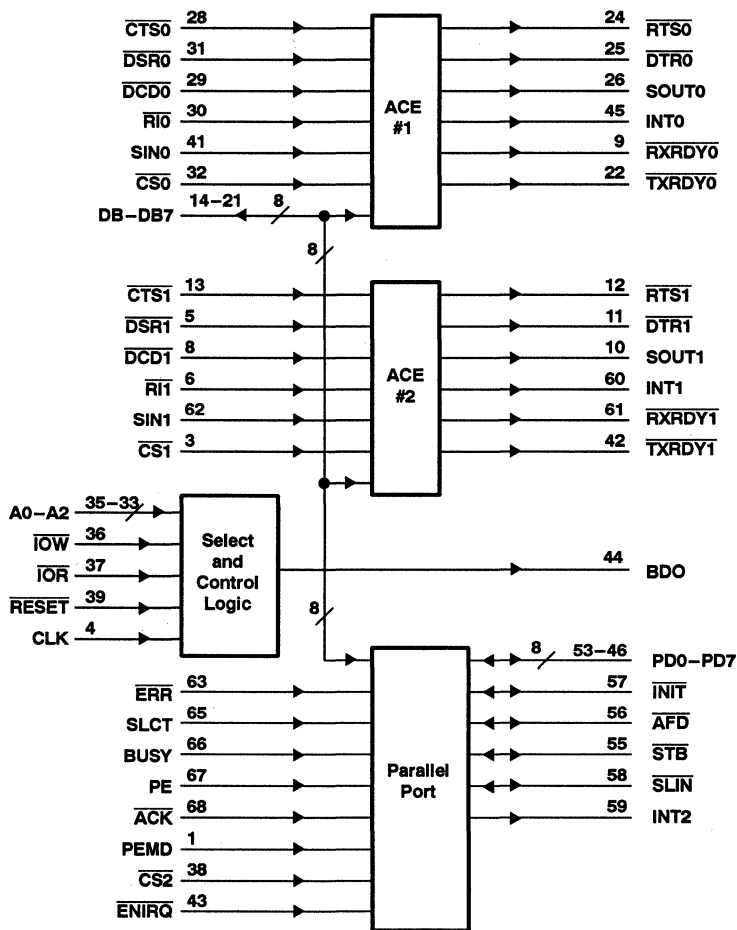
## description (continued)

In addition to its dual communications interface capabilities, the TL16C552 provides the user with a fully bidirectional parallel data port that fully supports the parallel Centronics-type printer. The parallel port and the two serial ports provide IBM PC/AT-compatible computers with a single device to serve the three system ports.

A programmable baud rate generator is included that can divide the timing reference clock input by a divisor between 1 and  $(2^{16} - 1)$ .

The TL16C552 is housed in a 68-pin plastic leaded chip carrier.

## functional block diagram



# TL16C552 DUAL ASYNCHRONOUS COMMUNICATIONS ELEMENT WITH FIFO

SLLS102A - D3647, DECEMBER 1990 - REVISED MARCH 1993

## Terminal Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
ACK	68	I	Line printer acknowledge. This input goes low to indicate a successful data transfer has taken place. It generates a printer port interrupt during its positive transition.
AFD	56	I/O	Line printer autofeed. This open-drain line provides the line printer with an active-low signal when continuous form paper is to be autofeared to the printer. This pin has an internal pullup resistor to $V_{DD}$ of approximately 10 k $\Omega$ .
A0, A1, A2	35, 34, 33	I	Address lines A0-A2. The address lines select the internal registers during CPU bus operations. See Table 1 for the decode of the serial channels and Table 11 for the decode of the parallel line printer port.
BDO	44	O	Bus buffer output. This active-high output is asserted when either serial channel or the parallel port is read. This output can be used to control the system bus driver (74LS245).
BUSY	66	I	Line printer busy. This is an input line from the line printer that goes high when the line printer is not ready to accept data.
CLK	4	I	Clock input. The external clock input to the baud rate divisor of each ACE.
CS0, CS1, CS2	32, 3, 38	I	Chip selects. Each input acts as an enable for the write and read signals for the serial channels 1 (CS0) and 2 (CS1). CS2 enables the signals to the line printer port.
CTS0, CTS1	28, 13	I	Clear to send inputs. The logical state of each CTS pin is reflected in the CTS bit of the (MSR) modem status register (CTS is bit 4 of the MSR, written MSR(4)) of each ACE. A change of state in either CTS pin since the previous reading of the associated MSR causes the setting of DCTS (MSR(0)) of each modem status register.
DB0-DB7	14-21	I/O	Data bits DB0-DB7. The data bus provides eight three-state I/O lines for the transfer of data, control, and status information between the TL16C552 and the CPU. These lines are normally in a high-impedance state except during read operations. D0 is the least significant bit (LSB) and is the first serial data bit to be received or transmitted.
DCD0, DCD1	29, 8	I	Data carrier detect. DCD is a modem input whose condition can be tested by the CPU by reading MSR(7) (DCD) of the modem status registers. MSR(3) (DDCD) of the modem status register indicates whether the DCD input has changed since the previous reading of the MSR. DCD has no effect on the receiver.
DSR0, DSR1	31, 5	I	Data set ready inputs. The logical state of the DSR pins is reflected in MSR(5) of its associated modem status register. DDSR (MSR(1)) indicates whether the associated DSR pin has changed state since the previous reading of the MSR.
DTR0, DTR1	25, 11	O	Data terminal ready lines. Each DTR pin can be set (low) by writing a logic 1 to MCR(0), modem control register bit 0 of its associated ACE. This signal is cleared (high) by writing a logic 0 to the DTR bit (MCR(0)) or whenever a reset occurs. When active (low), the DTR pin indicates that its ACE is ready to receive data.
ENIRQ	43	I	Parallel port interrupt source mode selection. When low, the AT mode of interrupts is enabled. In this mode, the INT2 output is internally connected to the ACK input. If the ENIRQ input is tied high, the INT2 output is internally tied to the PRINT signal in the line printer status register. INT2 is latched high on rising edge of ACK.
ERR	63	I	Line printer error. This is an input line from the line printer. The line printer reports an error by holding this line low during the error condition.
GND	7, 27, 54		Ground (0 V). All pins must be tied to ground for proper operation.
INIT	57	I/O	Line printer initialize. This open-drain line provides the line printer with an active-low signal that allows the line printer initialization routine to be started. This pin has an internal pullup resistor to $V_{DD}$ of approximately 10 k $\Omega$ .
IOR	37	I	Input/output read strobe. This is an active-low input that enables the selected channel to output data to the data bus (DB0-DB7). The data output depends upon the register selected by the address inputs A0, A1, A2, and chip select. Chip select 0 (CS0) selects ACE #1, chip select 1 (CS1) selects ACE #2, and chip select 2 (CS2) selects the line printer port.
IOW	36	I	Input/output write strobe. This is an active-low input causing data from the data bus to be input to either ACE or to the parallel port. The destination depends upon the register selected by the address inputs A0, A1, A2, and chip selects CS0, CS1, and CS2.

**TL16C552  
DUAL ASYNCHRONOUS COMMUNICATIONS ELEMENT  
WITH FIFO**

SLLS102A—D3647, DECEMBER 1990—REVISED MARCH 1993

**Terminal Functions (continued)**

PIN		I/O	DESCRIPTION
NAME	NO.		
INT0, INT1	45, 60	O	Serial channel interrupts. Each 3-state serial channel interrupt output (enabled by bit 3 of the MCR) goes active (high) when one of the following interrupts has an active (high) condition and is enabled by the interrupt enable register of its associated channel: receiver error flag, received data available, transmitter holding register empty, and modem status. The interrupt is reset low upon appropriate service. Upon reset, the interrupt output will be in the high-impedance state.
INT2	59	O	Printer port interrupt. This signal is an active-high, three-state output, generated by the positive transition of $\overline{ACK}$ . It is enabled by bit 4 of the write control register. Upon a reset, the interrupt output will be in the high-impedance state. Its mode is also controlled by $\overline{ENIRQ}$ .
PD0—PD7	53—46	I/O	Parallel data bits (0—7). These eight lines provide a byte-wide input or output port to the system.
PE	67	I	Line printer paper empty. This is an input line from the line printer that goes high when the printer runs out of paper.
PEMD	1	I	Printer enhancement mode. When low, this signal enables the write data register to the PD0—PD7 lines. A high on this signal allows direction control of the PD0—7 port by the DIR bit in the control register. PEMD is usually tied low for the printer operation.
RESET	39	I	Reset. When low, the reset input forces the TL16C552 into an idle mode in which all serial data activities are suspended. The modem control register (MCR) along with its associated outputs are cleared. The line status register (LSR) is cleared except for the THRE and TEMT bits, which are set. All functions of the device remain in an idle state until programmed to resume serial data activities. This input has a hysteresis level of typically 400 mV.
RTS0, RTS1	24, 12	O	Request to send outputs. An RTS pin is set low by writing a logic 1 to MCR(1) bit 1 of its UART's modem control register. Both RTS pins are reset high by RESET. A low on the RTS pin indicates that its ACE has data ready to transmit. In half-duplex operations, RTS is used to control the direction of the line.
RXRDY0, RXRDY1	9, 61	O	Receiver ready. Receiver DMA signaling is also available through this pin. One of two types of DMA signaling can be selected via FCR3 when operating in the FIFO mode. Only DMA mode 0 is allowed when operating in the TL16C450 mode. For signal transfer DMA (a transfer is made between CPU bus cycles), mode 0 is used. Multiple transfers that are made continuously until the RCVR FIFO has been emptied are supported by mode 1.  Mode 0. $\overline{RXRDY}$ will be active (low) when in the FIFO mode (FCR0=1, FCR3=0) or when in the TL16C450 mode (FCR0=0) and the RCVR FIFO or RCVR holding register contain at least one character. When there are no more characters in the FIFO or holding register, the $\overline{RXRDY}$ pin will go inactive (high).  Mode 1. $\overline{RXRDY}$ will go active (low) in the FIFO mode (FCR0=1) when FCR3=1 and the timeout or trigger levels have been reached. It will go inactive (high) when the FIFO or holding register is empty.
RI0, RI1	30, 6	I	Ring indicator inputs. The RI signal is a modem control input whose condition is tested by reading MSR(6) (RI) of each ACE. The modem status register output TERI (MSR(2)) indicates whether the $\overline{RI}$ input has changed from high to low since the previous reading of the MSR.
SIN0, SIN1	41, 62	I	Serial data inputs. The serial data inputs move information from the communication line or modem to the TL16C552 receiver circuits. A mark (1) is high and a space (0) is low. Data on serial data inputs is disabled when operating in the loop mode.
SLCT	65	I	Line printer selected. This is an input line from the line printer that goes high when the line printer has been selected.
$\overline{SLIN}$	58	I/O	Line printer select. This open-drain selects the printer when it is active (low). This pin has an internal pullup resistor to $V_{DD}$ of approximately 10 k $\Omega$ .
SOUT0, SOUT1	26, 10	O	Serial data outputs. These lines are the serial data outputs from the ACEs transmitter circuitry. A mark is a logic 1 (high) and a space is a logic 0 (low). Each SOUT is held in the mark condition when the transmitter is disabled, RESET is true (low), the transmitter register is empty, or when in the loop mode.
STB	55	I/O	Line printer strobe. This open-drain line provides communication between the TL16C552 and the line printer. When it is active (low), it provides the line printer with a signal to latch the data currently on the parallel port. This pin has an internal pullup resistor to $V_{DD}$ of approximately 10 k $\Omega$ .



# TL16C552 DUAL ASYNCHRONOUS COMMUNICATIONS ELEMENT WITH FIFO

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## Terminal Functions (continued)

PIN NAME	NO.	I/O	DESCRIPTION
TRI	2	I	3-state control. This pin is used to control the 3-state control of all I/O and output pins. When this pin is asserted, all I/O and outputs become high impedance, allowing board level testers to drive the outputs without overdriving internal buffers. This pin is level sensitive. This pin is pulled down with an internal resistor that is approximately 5 kΩ and is a CMOS input.
TXRDY0, TXRDY1	22, 42	O	<p>Transmitter ready. Two types of DMA signaling are available. Either can be selected via FCR3 when operating in the FIFO mode. Only DMA mode 0 is allowed when operating in the TL16C450 mode. Single-transfer DMA (a transfer is made between CPU bus cycles) is supported by mode 0. Multiply transfers that are made continuously until the XMIT FIFO has been filled are supported by mode 1.</p> <p>Mode 0. When in the FIFO mode (FCR0=1, FCR3=0) or in the TL16C450 mode (FCR0=0) and there are no characters in the XMIT holding register or XMIT FIFO, TXRDY will be active (low). Once TXRDY is activated (low), it will go inactive after the first character is loaded into the holding register of XMIT FIFO.</p> <p>Mode 1. TXRDY will go active (low) if in the FIFO mode (FCR0=1) when FCR3=1 and there are no characters in the XMIT FIFO. When the XMIT FIFO is completely full, TXRDY will go inactive (high).</p>
VDD	23, 40, 64		Power supply. The power supply requirement is 5 V ± 5%.

### detailed description

Three types of information are stored in the internal registers used in the ACE: control, status, and data. Mnemonic abbreviations are shown in the table below for the registers.

CONTROL	MNEMONIC	STATUS	MNEMONIC	DATA	MNEMONIC
Line control register	LCR	Line status register	LSR	Receiver buffer register	RBR
FIFO control register	FCR	Modem status register	MSR	Transmitter holding register	THR
Modem control register	MCR				
Divisor latch LSB	DLL				
Divisor latch MSB	DLM				
Interrupt enable register	IER				

The address, read, and write inputs are used with the divisor latch access bit (DLAB) in the line-control register (bit 7) to select the register to be written or read (see Table 1).

**Table 1. Serial Channel Internal Registers**

DLAB	A2	A1	A0	MNEMONIC	REGISTER
L	L	L	L	RBR	Receiver buffer register (read only)
L	L	L	L	THR	Transmitter holding register (write only)
L	L	L	H	IER	Interrupt enable register
X	L	H	L	IIR	Interrupt identification register (read only)
X	L	H	L	FCR	FIFO control register (write only)
X	L	H	H	LCR	Line control register
X	H	L	L	MCR	Modem control register
X	H	L	H	LSR	Line status register
X	H	H	L	MSR	Modem status register
X	H	H	H	SCR	Scratch register
H	L	L	L	DLL	Divisor latch (LSB)
H	L	L	H	DLM	Divisor latch (MSB)

X = irrelevant, L = low level, H = high level

NOTE: The serial channel is accessed when either  $\overline{CS0}$  or  $\overline{CS1}$  is low.



# TL16C552 DUAL ASYNCHRONOUS COMMUNICATIONS ELEMENT WITH FIFO

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## detailed description (continued)

Individual bits within the registers are referred to by the register mnemonic and the bit number in parenthesis. As an example, LCR (7) refers to line control register bit 7.

The transmitter buffer register and receiver buffer register are data registers that hold from five to eight bits of data. If less than eight data bits are transmitted, data is right justified to the LSB. Bit 0 of a data word is always the first serial data bit received and transmitted. The ACE data registers are double buffered so that read and write operations may be performed when the ACE is performing the parallel-to-serial or serial-to-parallel conversion.

## line-control register

The format of the data character is controlled by the line-control register. The LCR may be read. Its contents are described below and shown in Figure 1.

LCR(0) and LCR(1) word length select bit 1:

The number of bits in each serial character is programmed as shown in Figure 1.

LCR(2) stop bit select bit 2:

LCR(2) specifies the number of stop bits in each transmitted character as shown in Figure 1. The receiver always checks for one stop bit.

LCR(3) parity enable bit 3:

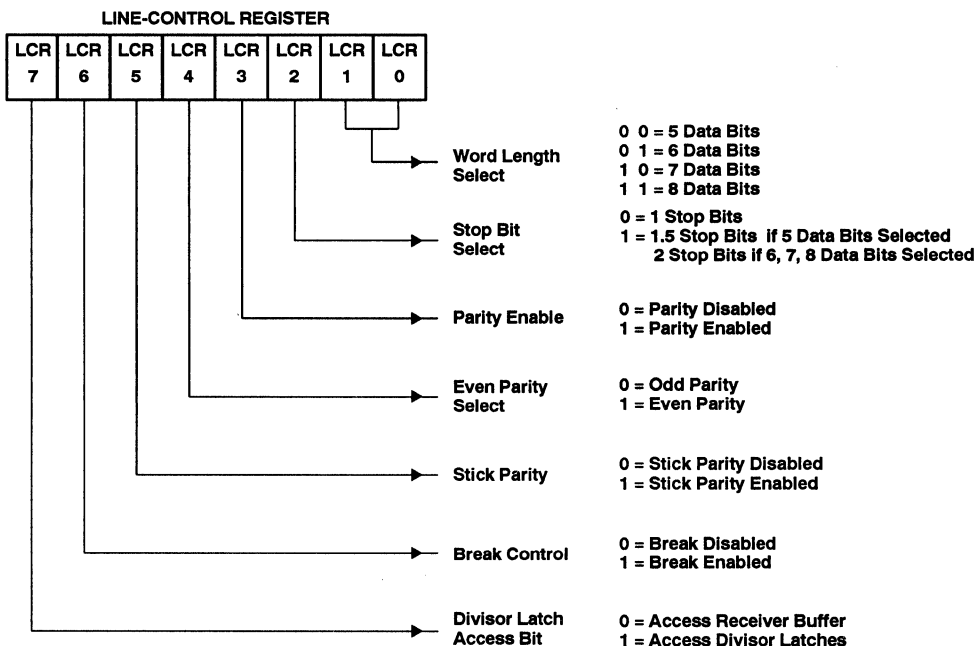
When LCR(3) is high, a parity bit between the last data word bit and stop bit is generated and checked.

LCR(4) even parity select bit 4:

When enabled, a logic one selects even parity.

LCR(5) stick parity bit 5:

When parity is enabled (LCR(3)=1), LCR(5)=1 causes the transmission and reception of a parity bit to be in the opposite state from the value of LCR(4). This forces parity to a known state and allows the receiver to check the parity bit in a known state.



**Figure 1. Line Control Register Contents**

**line-control register (continued)**

LCR(6) break control bit 6:

When LCR(6) is set to a logic 1, the serial output (SOUT1/SOUT0) is forced to the spacing state (low). The break control bit acts only on the serial output and does not affect the transmitter logic. If the following sequence is used, no invalid characters will be transmitted because of the break:

- Step 1 Load a zero byte in response to the transmitter holding register empty (THRE) status indication.
- Step 2 Set the break in response to the next THRE status indication.
- Step 3 Wait for the transmitter to be idle when transmitter empty status signal is set high (TEMT=1). Then clear the break when the normal transmission has to be restored.

LCR(7) divisor latch access bit (DLAB) bit 7:

Bit 7 must be set high (logic 1) to access the divisor latches DLL and DLM of the baud rate generator during a read or write operation. LCR(7) must be input low (logic 0) to access the receiver buffer register, the transmitter holding register or the interrupt enable register.

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## line status register

The line status register (LSR) is a single register that provides status indications. The line status register shown in Table 2 is described below:

### LSR(0) data ready (DR) bit 0:

Data ready is set high when an incoming character has been received and transferred into the receiver buffer register or the FIFO. LSR(0) is reset low by a CPU read of the data in the receiver buffer register or the FIFO.

### LSR(1) overrun error (OE) bit 1:

Overrun error indicates that data in the receiver buffer register was not read by the CPU before the next character was transferred into the receiver buffer register overwriting the previous character. The OE indicator is reset whenever the CPU reads the contents of the line status register. An overrun error will occur in the FIFO mode after the FIFO is full and the next character is completely received. The overrun error is detected by the CPU on the first LSR read after it happens. The character in the shift register is not transferred to the FIFO but it is overwritten.

### LSR(2) parity error (PE) bit 2:

Parity error indicates that the received data character does not have the correct parity as selected by LCR(3) and LCR(4). The PE bit is set high upon detection of a parity error and is reset low when the CPU reads the contents of the LSR. In the FIFO mode, the parity error is associated with a particular character in the FIFO. LSR(2) reflects the error when the character is at the top of the FIFO.

### LSR(3) framing error (FE) bit 3:

Framing error indicates that the received character did not have a valid stop bit. LSR(3) is set high when the stop bit following the last data bit or parity bit is detected as a zero bit (spacing level). The FE indicator is reset low when the CPU reads the contents of the LSR. In the FIFO mode, the framing error is associated with a particular character in the FIFO. LSR(3) reflects the error when the character is at the top of the FIFO.

Table 2. Line Status Register Bits

LSR BITS	1	0
LSR(0) data ready (DR)	Ready	Not ready
LSR(1) overrun error (OE)	Error	No error
LSR(2) parity error (PE)	Error	No error
LSR(3) framing error (FE)	Error	No error
LSR(4) break interrupt (BI)	Break	No break
LSR(5) transmitter holding register empty (THRE)	Empty	Not empty
LSR(6) transmitter empty (TEMT)	Empty	Not empty
LSR(7) RCVR FIFO error	Error in FIFO	No error in FIFO

### LSR(4) break interrupt (BI) bit 4:

Break interrupt is set high when the received data input is held in the spacing (logic 0) state for longer than a full word transmission time (start bit + data bits + parity + stop bits). The BI indicator is reset when the CPU reads the contents of the line status register. In the FIFO mode, this is associated with a particular character in the FIFO. LSR(4) reflects the BI when the break character is at the top of the FIFO. The error is detected by the CPU when its associated character is at the top of the FIFO during the first LSR read. Only one zero character is loaded into the FIFO when BI occurs.

LSR(1) - LSR(4) are the error conditions that produce a receiver line status interrupt (priority 1 interrupt in the interrupt identification register (IIR)) when any of the conditions are detected. This interrupt is enabled by setting IER(2)=1 in the interrupt enable register.



**line status register (continued)**

LSR(5) transmitter holding register empty (THRE) bit 5:

THRE indicates that the ACE is ready to accept a new character for transmission. The THRE bit is set high when a character is transferred from the transmitter holding register into the transmitter shift register. LSR(5) is reset low by the loading of the transmitter holding register by the CPU. LSR(5) is not reset by a CPU read of the LSR. In the FIFO mode when the XMIT FIFO is empty, this bit is set. It is cleared when one byte is written to the XMIT FIFO. When the THRE interrupt is enabled by IER(1), THRE causes a priority 3 interrupt in the IIR. If THRE is the interrupt source indicated in IIR, INTRPT is cleared by a read of the IIR.

LSR(6) transmitter empty (TEMT) bit 6:

TEMT is set high when the transmitter holding register (THR) and the transmitter shift register (TSR) are both empty. LSR(6) is reset low when a character is loaded into the THR and remains low until the character is transferred out of SOUT. TEMT is not reset low by a CPU read of the LSR. In the FIFO mode, when both the transmitter FIFO and shift register are empty, this bit is set to one.

LSR(7) RCVR FIFO error bit 7:

The LSR(7) bit is always 0 in the TL16C450 mode. In FIFO mode, it is set when at least one of the following data errors is in the FIFO: parity error, framing error, or break interrupt indication. It is cleared when the CPU reads the LSR if there are no subsequent errors in the FIFO.

NOTE: The line status register may be written. However, this function is intended only for factory test. It should be considered as read only by applications software.

**FIFO-control register**

This write-only register is at the same location as the IIR. It is used to enable and clear the FIFOs, set the trigger level of the RCVR FIFO, and select the type of DMA signaling.

FCR(0) enables both the XMIT and RCVR FIFOs. All bytes in both FIFOs can be cleared by resetting FCR(0). Data is cleared automatically from the FIFOs when changing from the FIFO mode to the TL16C450 mode and vice versa. Programming of other FCR bits is enabled by setting FCR(0)=1.

FCR(1)=1 clears all bytes in the RCVR FIFO and resets the counter logic to 0. This does not clear the shift register.

FCR(2)=1 clears all bytes in the XMIT FIFO and resets the counter logic to 0. This does not clear the shift register.

FCR(3)=1 will change the  $\overline{RXRDY}$  and  $\overline{TXRDY}$  pins from mode 0 to mode 1 if FCR(0)=1.

FCR(4) – FCR(5): These two bits are reserved for future use.

FCR(6) – FCR(7): These two bits are used for setting the trigger level for the RCVR FIFO interrupt as follows:

BIT		RCVR FIFO TRIGGER LEVEL (BYTES)
7	6	
0	0	01
0	1	04
1	0	08
1	1	14

# TL16C552 DUAL ASYNCHRONOUS COMMUNICATIONS ELEMENT WITH FIFO

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## modem-control register

The modem-control register (MCR) controls the interface with the modem or data set as described in Figure 2. MCR can be written and read. The  $\overline{RTS}$  and  $\overline{DTR}$  outputs are directly controlled by their control bits in this register. A high input asserts a low (true) at the output pins. MCR bits 0, 1, 2, 3, and 4 are shown as follows:

**MCR(0):** When  $\overline{MCR(0)}$  is set high, the  $\overline{DTR}$  output is forced low. When  $\overline{MCR(0)}$  is reset low, the  $\overline{DTR}$  output is forced high. The  $\overline{DTR}$  output of the serial channel may be input into an inverting line driver in order to obtain the proper polarity input at the modem or data set.

**MCR(1):** When  $\overline{MCR(1)}$  is set high, the  $\overline{RTS}$  output is forced low. When  $\overline{MCR(1)}$  is reset low, the  $\overline{RTS}$  output is forced high. The  $\overline{RTS}$  output of the serial channel may be input into an inverting line driver to obtain the proper polarity input at the modem or data set.

**MCR(2):** When  $\overline{MCR(2)}$  is set high,  $\overline{OUT1}$  is forced low.

**MCR(3):** When  $\overline{MCR(3)}$  is set high, the  $\overline{OUT2}$  output is forced low.

**MCR(4):**  $\overline{MCR(4)}$  provides a local loopback feature for diagnostic testing of the channel. When  $\overline{MCR(4)}$  is set high, serial output (SOUT) is set to the marking (logic 1) state, and the receiver data input serial input (SIN) is disconnected. The output of the transmitter shift register is looped back into the receiver shift register input. The four modem control inputs ( $\overline{CTS}$ ,  $\overline{DSR}$ ,  $\overline{DCD}$ , and  $\overline{RI}$ ) are disconnected. The modem-control outputs ( $\overline{DTR}$ ,  $\overline{RTS}$ ,  $\overline{OUT1}$ , and  $\overline{OUT2}$ ) are internally connected to the four modem control inputs. The modem-control output pins are forced to their inactive state (high) on the TL16C552. In the diagnostic mode, data transmitted is immediately received. This allows the processor to verify the transmit and receive data paths of the selected serial channel. Interrupt control is fully operational. However, interrupts are generated by controlling the lower four MCR bits internally. Interrupts are not generated by activity on the external pins represented by those four bits.

MCR(5) – MCR(7) are permanently set to logic 0.

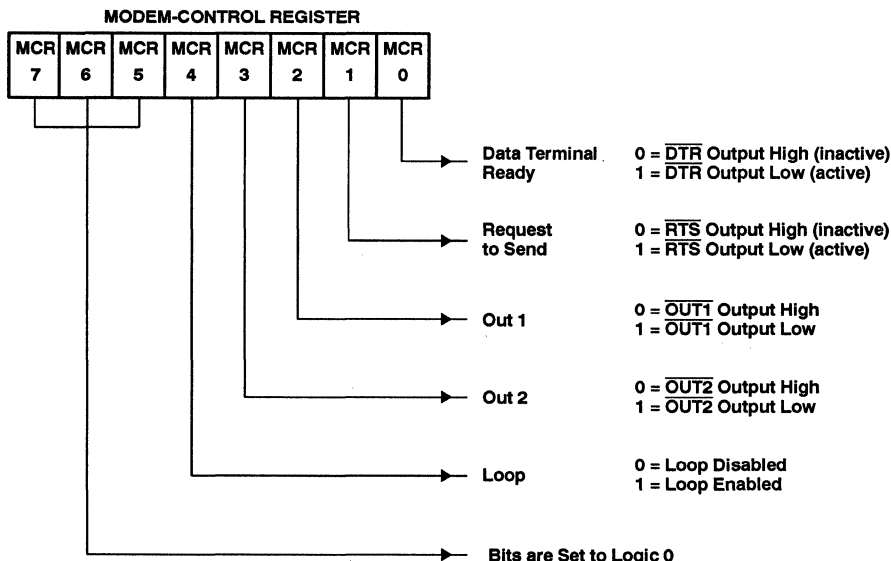


Figure 2. Modem-Control Register Contents

**modem-status register**

The MSR provides the CPU with status of the modem input lines from the modem or peripheral devices. The MSR allows the CPU to read the serial channel modem signal inputs by accessing the data bus interface of the ACE in addition to the current status of four bits of the MSR that indicate whether the modem inputs have changed since the last reading of the MSR. The delta status bits are set high when a control input from the modem changes state and reset low when the CPU reads the MSR.

The modem input lines are  $\overline{CTS}$ ,  $\overline{DSR}$ ,  $\overline{RI}$ , and  $\overline{DCD}$ . MSR(4) – MSR(7) are status indications of these lines. A status bit = 1 indicates the input is a low. A status bit = 0 indicates the input is high. If the modem-status interrupt in the interrupt enable register is enabled (IER(3)), an interrupt is generated whenever MSR(0) – MSR(3) is set to a one. The MSR is a priority 4 interrupt. The contents of the modem-status register are described in Table 3.

MSR(0) delta clear to send (DCTS) bit 0:

DCTS displays that the  $\overline{CTS}$  input to the serial channel has changed state since it was last read by the CPU.

MSR(1) delta data set ready (DDSR) bit 1:

DDSR indicates that the  $\overline{DSR}$  input to the serial channel has changed state since the last time it was read by the CPU.

MSR(2) trailing edge of ring indicator (TERI) bit 2:

TERI indicates that the  $\overline{RI}$  input to the serial channel has changed state from low to high since the last time it was read by the CPU. High-to-low transitions on RI do not activate TERI.

MSR(3) delta data carrier detect (DDCD) bit 3:

DDCD indicates that the  $\overline{DCD}$  input to the serial channel has changed state since the last time it was read by the CPU.

MSR(4) clear to send (CTS) bit 4:

CTS is the complement of the  $\overline{CTS}$  input from the modem indicating to the serial channel that the modem is ready to receive data from the serial channel's transmitter output (SOUT). If the serial channel is in the loop mode ((MCR(4)=1), MSR(4) reflects the value of RTS in the MCR.

MSR(5) data set ready (DSR) bit 5:

DSR is the complement of the  $\overline{DSR}$  input from the modem to the serial channel that indicates that the modem is ready to provide received data to the serial channel receiver circuitry. If the channel is in the loop mode (MCR(4)=1), MSR(5) reflects the value of DTR in the MCR.

**Table 3. Modem-Status Register Bits**

MSR BIT	MNEMONIC	DESCRIPTION
MSR(0)	DCTS	Delta clear to send
MSR(1)	DDSR	Delta data set ready
MSR(2)	TERI	Trailing edge of ring indicator
MSR(3)	DDCD	Delta data-carrier detect
MSR(4)	$\overline{CTS}$	Clear to send
MSR(5)	$\overline{DSR}$	Data set ready
MSR(6)	$\overline{RI}$	Ring indicator
MSR(7)	$\overline{DCD}$	Data-carrier detect

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## modem-status register (continued)

MSR(6) ring indicator (RI) bit 6:

RI is the complement of the  $\overline{\text{RI}}$  input. If the channel is in the loop mode (MCR(4)=1), MSR(6) reflects the value of  $\overline{\text{OUT1}}$  in the MCR.

MSR(7) data carrier detect (DCD) bit 7:

Data carrier detect indicates the status of the data carrier detect ( $\overline{\text{DCD}}$ ) input. If the channel is in the loop mode (MCR(4)=1), MSR(7) reflects the value of  $\overline{\text{OUT2}}$  in the MCR.

Reading the MSR register clears the delta modem status indications but has no effect on the other status bits. For LSR and MSR, the setting of status bits is inhibited during status register read operations. If a status condition is generated during a read  $\overline{\text{IOR}}$  operation, the status bit is not set until the trailing edge of the read. If a status bit is set during a read operation, and the same status condition occurs, that status bit will be cleared at the trailing edge of the read instead of being set again. In the loop-back mode, when modem status interrupts are enabled, the  $\overline{\text{CTS}}$ ,  $\overline{\text{DSR}}$ ,  $\overline{\text{RI}}$  and  $\overline{\text{DCD}}$  input pins are ignored. However, a modem-status interrupt may still be generated by writing to MCR3–MCR0. Applications software should not write to the modem-status register.

## divisor latches

The ACE serial channel contains a programmable baud-rate generator (BRG) that divides the clock (dc to 8 MHz) by any divisor from 1 to  $2^{16}-1$  (see also BRG description). The output frequency of the baud generator is 16X the data rate (divisor # = clock / (baud rate x 16)) referred to in this document as RCLK. Two 8-bit divisor latch registers store the divisor in a 16-bit binary format. These divisor latch registers must be loaded during initialization. Upon loading either of the divisor latches, a 16-bit baud counter is immediately loaded. This prevents long counts on initial load. The BRG can use any of three different popular frequencies to provide standard baud rates. These frequencies are 1.8432 MHz, 3.072 MHz, and 8 MHz. With these frequencies, standard bit rates from 50 to 512K bps are available. Tables 5, 6, and 7 illustrate the divisors needed to obtain standard rates using these three frequencies.

## scratchpad register

The scratchpad register is an 8-bit read/write register that has no effect on either channel in the ACE. It is intended to be used by the programmer to hold data temporarily.

## interrupt identification register

In order to minimize software overhead during data character transfers, the serial channel prioritizes interrupts into four levels. The four levels of interrupt conditions are as follows:

1. Receiver line status (priority 1)
2. Received data ready (priority 2) or character timeout
3. Transmitter holding register empty (priority 3)
4. Modem status (priority 4)

Information indicating that a prioritized interrupt is pending and the type of interrupt is stored in the interrupt identification register (IIR). The IIR indicates the highest priority interrupt pending. The contents of the IIR are indicated in Table 4.

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**Table 4. Interrupt Identification Register**

FIFO MODE ONLY	INTERRUPT IDENTIFICATION REGISTER			INTERRUPT SET AND RESET FUNCTIONS			
BIT 3	BIT 2	BIT 1	BIT 0	PRIORITY LEVEL	INTERRUPT TYPE	INTERRUPT SOURCE	INTERRUPT RESET CONTROL
0	0	0	1	–	None	None	–
0	1	1	0	First	Receiver line status	OE, PE, FE, or BI	LSR read
0	1	0	0	Second	Received data available	Receiver data available or trigger level reached	RBR read until FIFO drops below the trigger level
1	1	0	0	Second	Character timeout indication	No characters have been removed from or input to the receiver FIFO during the last four character times and there is at least one character in it during this time.	RBR read
0	0	1	0	Third	THRE	THRE	IIR read if THRE is the interrupt source or THR write
0	0	0	0	Fourth	Modem status	CTS, DSR, RI, or DCD	MSR read

**interrupt identification register (continued)**

IIR(0) can be used to indicate whether an interrupt is pending. When IIR(0) is low, an interrupt is pending.

IIR(1) and IIR(2) are used to identify the highest priority interrupt pending as indicated in Table 4.

IIR(3): This bit is always logic 0 when in the TL16C450 mode. This bit is set along with bit 2 when in the FIFO mode and a trigger change level interrupt is pending.

IIR(4) – IIR(5): These two bits are always set to logic 0.

IIR(6) – IIR(7): FCR(0)=1 sets these two bits.

**interrupt enable register**

The interrupt enable register (IER) is used to independently enable the four serial channel interrupt sources that activate the interrupt (INT0 or INT1) output. All interrupts are disabled by resetting IER(0) – IER(3) of the interrupt enable register. Interrupts are enabled by setting the appropriate bits of the IER high. Disabling the interrupt system inhibits the interrupt identification register and the active (high) interrupt output. All other system functions operate in their normal manner, including the setting of the line status and modem status registers. The contents of the interrupt enable register are described in Table 9 and below.

IER(0). When set to one, IER(0) enables the received data available interrupt and the timeout interrupts in the FIFO mode.

IER(1). When set to one, IER(1) enables the transmitter holding register empty interrupt.

IER(2). When set to one IER(2) enables the receiver line status interrupt.

IER(3). When set to one, IER(3) enables the modem status interrupt.

IER(4) – IER(7). These four bits of the IER are logic 0.





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## receiver

Serial asynchronous data is input into the SIN pin. The ACE continually searches for a high-to-low transition from the idle state. When the transition is detected, a counter is reset, and counts the 16X clock to 7 1/2, which is the center of the start bit. The start bit is valid if the SIN is still low. Verifying the start bits prevents the receiver from assembling a false data character due to a low-going noise spike on the SIN input.

The line-control register determines the number of data bits in a character [LCR(0), LCR(1)]. If parity is used LCR(3) and the polarity of parity LCR(4) are needed. Status for the receiver is provided in the line status register. When a full character is received, including parity and stop bits, the data received indication in LSR(0) is set high. The CPU reads the receiver buffer register, which resets LSR(0). If the character is not read prior to a new character transfer from the RSR to the RBR, the overrun error status indication is set in LSR(1). If there is a parity error, the parity error is set in LSR(2). If a stop bit is not detected, a framing error indication is set in LSR(3).

If the data into SIN is a symmetrical square wave, the center of the data cells will occur within  $\pm 3.125\%$  of the actual center, providing an error margin of 46.875%. The start bit can begin as much as one 16X clock cycle prior to being detected.

## master reset

After power up, the ACE  $\overline{\text{RESET}}$  input should be held low for one microsecond to reset the ACE circuits to an idle mode until initialization. A low on  $\overline{\text{RESET}}$  causes the following:

1. Initializes the transmitter and receiver clock counters.
2. Clears the line status register (LSR), except for transmitter shift register empty (TEMT) and transmit holding register empty (THRE), which are set. The modem control register (MCR) is also cleared. All of the discrete lines, memory elements, and miscellaneous logic associated with these register bits are also cleared or turned off. The line control register (LCR), divisor latches, receiver buffer register, and transmitter buffer register are not affected.

Following the removal of the reset condition ( $\overline{\text{RESET}}$  high), the ACE remains in the idle mode until programmed.

A hardware reset of the ACE sets the THRE and TEMT status bit in the LSR. When interrupts are subsequently enabled, an interrupt occurs due to THRE.

A summary of the effect of a reset on the ACE is given in Table 8.

## programming

The serial channel of the ACE is programmed by the control registers: LCR, IER, DLL, DLM, MCR, and FCR. These control words define the character length, number of stop bits, parity, baud rate, and modem interface.

While the control registers can be written in any order, the IER should be written last because it controls the interrupt enables. Once the serial channel is programmed and operational, these registers can be updated any time the ACE serial channel is not transmitting or receiving data.

### FIFO-interrupt-mode operation

The following RCVR status occurs when the RCVR FIFO and receiver interrupts are enabled:

1. LSR(0) is set when a character is transferred from the shift register to the RCVR FIFO. When the FIFO is empty, it is reset.
2. IIR = 06 receiver line status interrupt has higher priority than the received data available interrupt IIR = 04.
3. Receive data available interrupt will be issued to the CPU when the programmed trigger level is reached by the FIFO. As soon as the FIFO drops below its programmed trigger level, it will be cleared.
4. IIR = 04 (receive data available indication) also occurs when the FIFO reaches its trigger level. It is cleared when the FIFO drops below the programmed trigger level.

The following RCVR FIFO character timeout status occurs when RCVR FIFO and receiver interrupts are enabled.

1. If the following conditions exist, a FIFO character timeout interrupt occurs.

Minimum of one character in FIFO

Last received serial character was longer than 4 continuous previous character times ago (if two stop bits are programmed, the second one is included in the time delay).

The last CPU read of the FIFO was more than 4 continuous character times earlier. At 300 baud and 12-bit characters, the FIFO timeout interrupt causes a latency of 160 ms maximum from received character to interrupt issued.

2. By using the RCLK input for a clock signal, the character times can be calculated. (The delay is proportional to the baud rate.)
3. The timeout timer is reset after the CPU reads the RCVR FIFO or after a new character is received, when there has been no timeout interrupt.
4. A timeout interrupt is cleared and the timer is reset when the CPU reads a character from the RCVR FIFO.

XMIT interrupts occur as follows when the transmitter and XMIT FIFO interrupts are enabled (FCRO = 1, IER = 1).

1. When the transmitter FIFO is empty, the transmitter holding register interrupt (IIR = 02) occurs. The interrupt is cleared as soon as the transmitter holding register is written to or the IIR is read. One to sixteen characters may be written to the transmit FIFO when servicing this interrupt.
2. The XMIT FIFO empty indications will be delayed one character time minus the last stop bit time whenever the following occurs:

THRE = 1 and there has not been a minimum of two bytes at the same time in XMIT FIFO, since the last THRE = 1. The first transmitter interrupt after changing FCRO will be immediate, however, assuming it is enabled.

RCVR FIFO trigger level and character timeout interrupts have the same priority as the received data available interrupt. The transmitter holding register empty interrupt has the same priority as the transmitter FIFO empty interrupt.

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**FIFO polled mode operation**

Resetting IER0, IER1, IER2, IER3, or all to zero, with FCR0 = 1, puts the ACE into the FIFO polled mode. RCVR and XMITER are controlled separately. Therefore, either or both can be in the polled mode.

In the FIFO polled mode, there is no timeout condition indicated or trigger level reached. However, the RCVR and XMIT FIFOs still have the capability of holding characters. The LSR must be read to determine the ACE status.

**Table 5. Baud Rates (1.8432-MHz clock)**

BAUD RATE DESIRED	DIVISOR (N) USED TO GENERATE 16 X CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	2304	-
75	1536	-
110	1047	0.026
134.5	857	0.058
150	768	-
300	384	-
600	192	-
1200	96	-
1800	64	-
2000	58	0.690
2400	48	-
3600	32	-
4800	24	-
7200	16	-
9600	12	-
19200	6	-
38400	3	-
56000	2	2.860

**Table 6. Baud Rates (3.072-MHz clock)**

BAUD RATE DESIRED	DIVISOR (N) USED TO GENERATE 16 X CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	3840	-
75	2560	-
110	1745	0.026
134.5	1428	0.034
150	1280	-
300	640	-
600	320	-
1200	160	-
1800	107	0.312
2000	96	-
2400	80	-
3600	53	0.628
4800	40	-
7200	27	1.230
9600	20	-
19200	10	-
38400	5	-

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**Table 7. Baud Rates (8.192-MHz clock)**

BAUD RATE DESIRED	DIVISOR (N) USED TO GENERATE 16 X CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	1000	-
75	6667	0.005
110	4545	0.010
134.5	3717	0.013
150	3333	0.010
300	1667	0.020
600	833	0.040
1200	417	0.080
1800	277	0.080
2000	250	-
2400	208	0.160
3600	139	0.080
4800	104	0.160
7200	69	0.644
9600	52	0.160
19200	26	0.160
38400	13	0.160
56000	9	0.790
128000	4	2.344
256000	2	2.344
512000	1	2.400

**Table 8. RESET**

REGISTER/SIGNAL	RESET CONTROL	RESET
Interrupt enable register	Reset	All bits low (0-3 forced and 4-7 permanent)
Interrupt identification register	Reset	Bit 0 is high, bits 1, 2, 3, 6, and 7 low Bits 4-5 are permanently low
Line control register	Reset	All bits low
Modem control register	Reset	All bits low
FIFO control register	Reset	All bits low
Line status register	Reset	All bits low, except bits 5 and 6 are high
Modem status register	Reset	Bits 0-3 low, bits 4-7 input signal
SOUT	Reset	High
Interrupt (RCVR errs)	Read LSR/Reset	Low
Interrupt (RCVR data ready)	Read RBR/Reset	Low
Interrupt (THRE)	Read IIR/Write THR/Reset	Low
Interrupt (modem status changes)	Read MSR/Reset	Low
OUT2	Reset	High
RTS	Reset	High
DTR	Reset	High
OUT1	Reset	High

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**Table 9. Serial Channel Accessible Registers**

ADDRESS	REGISTER MNEMONIC	REGISTER BIT NUMBER							
		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
0	RBR (read only)	Data Bit 7 (MSB)	Data Bit 6	Data Bit 5	Data Bit 4	Data Bit 3	Data Bit 2	Data Bit 1	Data Bit 0 (LSB)
0	THR (write only)	Data Bit 7	Data Bit 6	Data Bit 5	Data Bit 4	Data Bit 3	Data Bit 2	Data Bit 1	Data Bit 0
0†	DLL	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
1†	DLM	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8
1	IER	0	0	0	0	(EDSSI) Enable modem status interrupt	(ERLSI) Enable receiver line status interrupt	(ETBEI) Enable transmitter holding register empty interrupt	(ERBFI) Enable received data available interrupt
2	FCR (write only)	RCVR Trigger (MSB)	RCVR Trigger (LSB)	Reserved	Reserved	DMA mode select	XMIT FIFO reset	RCVR FIFO reset	FIFO Enable
2	IIR (read only)	FIFOs Enabled‡	FIFOs Enabled‡	0	0	Interrupt ID Bit (2)‡	Interrupt ID Bit (1)	Interrupt ID Bit (0)	0 if interrupt pending
3	LCR	(DLAB) Divisor latch access bit	Set break	Stick parity	(EPS) Even parity select	(PEN) Parity enable	(STB) Number of stop bits	(WLSB1) Word length select bit 1	(WLSB0) Word length select bit 0
4	MCR	0	0	0	Loop	Enable external interrupt (INT0 or INT1)	OUT1 (an unused internal signal)	(RTS) Request to send	(DTR) Data terminal ready
5	LSR	Error in RCVR FIFO‡	(TEMT) Transmitter empty	(THRE) Transmitter holding register empty	(BI) Break interrupt	(FE) Framing error	(PE) Parity error	(OE) Overrun error	(DR) Data ready
6	MSR	(DCD) Data carrier detect	(RI) Ring indicator	(DSR) Data set ready	(CTS) Clear to send	(DDCD) Delta data carrier detect	(TERI) Trailing edge ring indicator	(DDSR) Delta data set ready	(DCTS) Delta clear to send
7	SCR	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0

† DLAB = 1

‡ These bits are always 0 when FIFOs are disabled.

## parallel port registers

The TL16C552's parallel port can be used to interface the device to a Centronics-style printer. When chip select 2 ( $\overline{CS2}$ ) is low, the parallel port is selected. Table 11 shows the registers associated with this parallel port. The read or write function of the register is controlled by the state of the read ( $\overline{IOR}$ ) and write ( $\overline{IOW}$ ) pin as shown. The read data register allows the microprocessor to read the information on the parallel bus.

The read status register allows the microprocessor to read the status of the printer in the six most significant bits. The status bits are printer busy ( $\overline{BSY}$ ), acknowledge ( $\overline{ACK}$ ) which is a handshake function, paper empty (PE), printer selected ( $\overline{SLCT}$ ), error ( $\overline{ERR}$ ) and printer interrupt ( $\overline{PRINT}$ ). The read control register allows the state of the control lines to be read. The write control register sets the state of the control lines. They are direction ( $\overline{DIR}$ ), interrupt enable ( $\overline{INT2 EN}$ ), select in ( $\overline{SLIN}$ ), initialize the printer ( $\overline{INIT}$ ), autofeed the paper ( $\overline{AFD}$ ), and strobe ( $\overline{STB}$ ), which informs the printer of the presence of a valid byte on the parallel bus. The write data register allows the microprocessor to write a byte to the parallel bus. The parallel port is completely compatible with the parallel port implementation used in the IBM serial parallel adaptor.



**Table 10. Parallel Port Registers**

REGISTER	REGISTER BITS							
	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
Read Data	PD7	PD6	PD5	PD4	PD3	PD2	PD1	PD0
Read Status	BSY	ACK	PE	SLCT	ERR	PRINT	1	1
Read Control	0	0	DIR	INT2 EN	SLIN	INIT	AFD	STB
Write Data	PD7	PD6	PD5	PD4	PD3	PD2	PD1	PD0
Write Control	0	0	DIR	INT2 EN	SLIN	INIT	AFD	STB

**Table 11. Parallel Port Register Select**

CONTROL PINS					REGISTER SELECTED
IOR	IOW	CS2	A1	A0	
L	H	L	L	L	Read data
L	H	L	L	H	Read status
L	H	L	H	L	Read control
L	H	L	H	H	Invalid
H	L	L	L	L	Write data
H	L	L	L	H	Invalid
H	L	L	H	L	Write control
H	L	L	H	H	Invalid

### line printer port

The line printer port contains the functionality of the port included in the TL16C452, but offers a hardware programmable extended mode controlled by the printer enhancement mode (PEMD) pin. This enhancement is the addition of a direction control bit, and an interrupt status bit.

#### register 0 line printer data register

The line printer (LPD) port is either output only or bidirectional, depending on the state of the extended mode pin and data direction control bits.

Compatibility mode (PEMD pin = 0)

Reads to the LPD register return the last data that was written to the port. Write operations immediately output data to the PD0–PD7 pins.

Extended mode (PEMD pin = 1)

Read operations return either the data last written to the LPT data register if the direction bit is set to write (low), or the data that is present on PD0–PD7 if the direction is set to read (high). Writes to the LPD register latch data into the output register, but only drive the LPT port when the direction bit is set to write (low).

The table below summarizes the possible combinations of extended mode and the direction-control bit. In either case, the bits of the LPD register are defined as follows:

PEMD	DIR	PD0–PD7 FUNCTION
L	X	PC/AT mode – output
H	0	PS/2 mode – output
H	1	PS/2 mode – input

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## register 1 read line printer status register

The line printer status (LPS) register is a read-only register that contains interrupt and printer status of the LPT connector pins. In the table below (in the default column), are the values of each bit after reset in the case of the printer being disconnected from the port. The bits are described as follows:

BIT	DESCRIPTION	DEFAULT
0	Reserved	1
1	Reserved	1
2	PRINT	1
3	ERR	†
4	SLCT	†
5	PE	†
6	ACK	†
7	BSY	†

† Outputs are dependent upon device inputs.

Bits 0 and 1 – Reserved. Read as ones.

Bit 2 – Printer interrupt ( $\overline{\text{PRINT}}$ , active low) status bit. When set (low) indicates that the printer has acknowledged the previous transfer with an ACK handshake (bit if 4 of the control register is set to 1). The bit is set to zero on the active to inactive transition of the ACK signal. This bit is set to a one after a read of the status port.

Bit 3 – Error ( $\overline{\text{ERR}}$ , active low) status bit corresponds to  $\overline{\text{ERR}}$  input.

Bit 4 – Select (SLCT) status bit corresponds to SLCT input.

Bit 5 – Paper empty (PE) status bit corresponds to PE input.

Bit 6 – Acknowledge ( $\overline{\text{ACK}}$ , active low) status bit corresponds to  $\overline{\text{ACK}}$  input.

Bit 7 – Busy ( $\overline{\text{BSY}}$ , active low) status bit corresponds to BUSY input (active high).

## register 2 line-printer-control register

The line-printer-control (LPC) register is read/write port that is used to control the PD0–PD7 direction and drive the printer-control lines. Write operations set or reset these bits, while read operations return the state of the last write operation to this register. The bits in this register are defined as follows:

BIT	DESCRIPTION
0	STB
1	AFD
2	INIT
3	SLIN
4	INT2 EN
5	DIR
6	Reserved (0)
7	Reserved (0)

Bit 0 – Printer strobe (STB) control bit; when 1, the  $\overline{\text{STB}}$  signal is asserted on the LPT interface; when 0, the signal is negated.

Bit 1 – Auto feed (AFD) control bit; when 1, the  $\overline{\text{AFD}}$  signal will be asserted on the LPT interface; when 0, the signal is negated.

Bit 2 – Initialize printer ( $\overline{\text{INIT}}$ ) control bit; when 1, the  $\overline{\text{INIT}}$  signal is negated; when 0, the  $\overline{\text{INIT}}$  signal is asserted on the LPT interface.

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**register 2 line-printer-control register (continued)**

- Bit 3 – Select input (SLIN) control bit; when 1, the SLCT signal is asserted on the LPT interface; when 0, the signal is negated.
- Bit 4 – Interrupt request enable (INT2 EN) control bit; when 1, enables interrupts from the LPT port whenever the  $\overline{ACK}$  signal is released; when 0, disables interrupts and places INT2 signal in 3-state.
- Bit 5 – Direction (DIR) control bit (only used when PEMD is high); when 1, the output buffers in the LPD port are disabled allowing data driven from external sources to be read from the LPD port. When DIR is low, the LPD port is in output mode.

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage range, $V_{DD}$ (see Note 1)	–0.5 V to $V_{DD} + 0.3$ V
Input voltage range, $V_I$	–0.5 V to 7 V
Output voltage range, $V_O$	–0.5 V to $V_{DD} + 0.3$ V
Continuous total power dissipation	500 mW
Operating free-air temperature range	–10°C to 70°C
Storage temperature range	–65°C to 150°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage levels are with respect to ground ( $V_{SS}$ ).

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{DD}$	4.75	5	5.25	V
Clock high-level input voltage, $V_{IH}(CLK)$	2		$V_{DD}$	V
Clock low-level input voltage, $V_{IL}(CLK)$	–0.5		0.8	V
High-level input voltage, $V_{IH}$	2		$V_{DD}$	V
Low-level input voltage, $V_{IL}$	–0.5		0.8	V
Clock frequency, $f_{clock}$			8	MHz
Operating free-air temperature range, $T_A$	0		70	°C





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**electrical characteristics over recommended ranges of operating free-air temperature and supply voltage**

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -0.4 mA for DB0-DB7, I <sub>OH</sub> = -2 mA for PD0-PD7, I <sub>OH</sub> = -0.4 mA for $\overline{\text{INIT}}$ , $\overline{\text{AFD}}$ , $\overline{\text{STB}}$ , and $\overline{\text{SLIN}}$ (see Note 2), I <sub>OH</sub> = -0.4 mA for all other outputs	2.4		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 4 mA for DB0-DB7, I <sub>OL</sub> = 12 mA for PD0-PD7, I <sub>OL</sub> = 10 mA for $\overline{\text{INIT}}$ , $\overline{\text{AFD}}$ , $\overline{\text{STB}}$ , and $\overline{\text{SLIN}}$ (see Note 2), I <sub>OL</sub> = 2 mA for all other outputs		0.4	V
I <sub>I</sub>	Input current	V <sub>DD</sub> = 5.25 V, All other pins are floating		±10	µA
I <sub>I</sub> (CLK)	Clock input current	V <sub>I</sub> = 0 to 5.25 V		±10	µA
I <sub>OZ</sub>	Off-state output current	V <sub>DD</sub> = 5.25 V, V <sub>O</sub> = 0 with chip deselected, or V <sub>O</sub> = 5.25 V with chip and write mode selected		±20	µA
I <sub>DD</sub>	Supply current	V <sub>DD</sub> = 5.25 V, No loads on outputs, SIN0, SIN1, DSR0, DSR1, DCD0, DCD1, CTS0, CTS1, $\overline{\text{R}}\overline{\text{I}}\overline{0}$ and $\overline{\text{R}}\overline{\text{I}}\overline{1}$ at 2 V, Other inputs at 0.8 V, Baud rate generator f <sub>clock</sub> = 8 MHz, Baud rate = 56 kb/s		50	mA

**clock timing requirements over recommended ranges of operating free-air temperature and supply voltage**

		MIN	MAX	UNIT
t <sub>w1</sub>	Pulse duration, CLK high (external clock, 8 MHz max) (see Figure 3)	55		ns
t <sub>w2</sub>	Pulse duration, CLK low (external clock, 8 MHz max) (see Figure 3)	55		ns
t <sub>w3</sub>	Pulse duration, master reset (see Figure 18)	1000		ns

**read cycle timing requirements over recommended ranges of operating free-air temperature and supply voltage (see Figure 6)**

		MIN	MAX	UNIT
t <sub>w4</sub>	Pulse duration, $\overline{\text{IOR}}$ low	80		ns
t <sub>su1</sub>	Setup time, chip select valid before $\overline{\text{IOR}}$ low (see Note 3)	15		ns
t <sub>su2</sub>	Setup time, A2-A0 valid before $\overline{\text{IOR}}$ low (see Note 3)	15		ns
t <sub>h1</sub>	Hold time, A2-A0 valid after $\overline{\text{IOR}}$ high (see Note 3)	20		ns
t <sub>h2</sub>	Hold time, chip select valid after $\overline{\text{IOR}}$ high (see Note 3)	20		ns
t <sub>d1</sub>	Delay time, t <sub>su2</sub> + t <sub>w4</sub> + t <sub>d2</sub> (see Note 4)	175		ns
t <sub>d2</sub>	Delay time, $\overline{\text{IOR}}$ high to $\overline{\text{IOR}}$ or $\overline{\text{IOW}}$ low	80		ns

NOTES: 2. These four pins contain an internal pullup resistor to V<sub>DD</sub> of approximately 10 kΩ.

3. The internal address strobe is always active.

4. In the FIFO mode, t<sub>d1</sub> = 425 ns (min) between reads of the receiver FIFO and the status registers (IIR and LSR).



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**write cycle timing requirements over recommended ranges of operating free-air temperature and supply voltage (see Figure 7)**

	MIN	MAX	UNIT
$t_{w5}$ Pulse duration, $\overline{IOW}$ low	80		ns
$t_{su4}$ Setup time, chip select valid before $\overline{IOW}$ low (see Note 3)	15		ns
$t_{su5}$ Setup time, A2–A0 valid before $\overline{IOW}$ low (see Note 3)	15		ns
$t_{su6}$ Setup time, D0–D7 valid before $\overline{IOW}$ high	15		ns
$t_{h3}$ Hold time, A2–A0 valid after $\overline{IOW}$ high (see Note 3)	20		ns
$t_{h4}$ Hold time, chip select valid after $\overline{IOW}$ high (see Note 3)	20		ns
$t_{h5}$ Hold time, D0–D7 valid after $\overline{IOW}$ high	15		ns
$t_{d3}$ Delay time, $t_{su5} + t_{w5} + t_{d4}$	175		ns
$t_{d4}$ Delay time, $\overline{IOW}$ high to $\overline{IOW}$ or $\overline{IOR}$ low	80		ns

NOTE 3: The internal address strobe is always active.

**read cycle switching characteristics over recommended ranges of operating free-air temperature and supply voltage (see Figure 6)**

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
$t_{pd1}$ Propagation delay time from $\overline{IOR}$ high to BD0 high or from $\overline{IOR}$ low to BD0 low	$C_L = 100$ pF, See Note 5		60	ns
$t_{en}$ Enable time from $\overline{IOR}$ low to D0–D7 valid	$C_L = 100$ pF, See Note 5		60	ns
$t_{dis}$ Disable time from $\overline{IOR}$ high to D0–D7 released	$C_L = 100$ pF, See Note 5	0	60	ns

NOTE 5:  $V_{OL}$  and  $V_{OH}$  (and the external loading) determine the charge and discharge time.

**transmitter switching characteristics over recommended ranges of operating free-air temperature and supply voltage (see Figures 8, 9, and 10)**

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
$t_{d5}$ Delay time, interrupt THRE low to SOUT low at start		8	24	RCLK cycles
$t_{d6}$ Delay time, SOUT low at start to interrupt THRE high	See Note 6	8	8	RCLK cycles
$t_{d7}$ Delay time, $\overline{IOW}$ (WR THR) high to interrupt THRE high	See Note 6	16	32	RCLK cycles
$t_{d8}$ Delay time, SOUT low at start to $\overline{TXRDY}$ low	$C_L = 100$ pF		8	RCLK cycles
$t_{pd2}$ Propagation delay time from $\overline{IOW}$ (WR THR) low to interrupt THRE low	$C_L = 100$ pF		140	ns
$t_{pd4}$ Propagation delay time from $\overline{IOR}$ (RD IIR) high to interrupt THRE low	$C_L = 100$ pF		140	ns
$t_{pd5}$ Propagation delay time from $\overline{IOW}$ (WR THR) high to $\overline{TXRDY}$ high	$C_L = 100$ pF		195	ns

NOTE 6: If the transmitter interrupt delay is active, this delay will be lengthened by one character time minus the last stop bit time.



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## receiver switching characteristics over recommended ranges of operating free-air temperature and supply voltage (see Figures 11, 12, 13, 14 and 15)

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
$t_{d9}$	Delay time from stop to INT high	See Note 7		1	RCLK cycle
$t_{pd6}$	Propagation delay time from RCLK high to sample CLK high			100	ns
$t_{pd7}$	Propagation delay time from $\overline{IOR}$ (RD RBR/RD LSR) high to reset interrupt low	$C_L = 100$ pF		150	ns
$t_{pd8}$	Propagation delay time from $\overline{IOR}$ (RD RBR) low to $\overline{RXRDY}$ high			150	ns

NOTE 7: The receiver data available indication, the overrun error indication, the trigger level interrupts and the active  $\overline{RXRDY}$  indication will be delayed three RCLK cycles in the FIFO mode ( $FCR0 = 1$ ). After the first byte has been received, status indicators (PE, FE, BI) will be delayed three RCLK cycles. These indicators will be updated immediately for any further bytes received after RDRBR goes active. There are eight RCLK cycle delays for trigger change level interrupts.

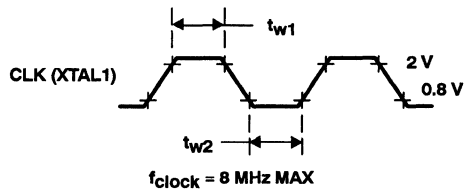
## modem control switching characteristics over recommended ranges of operating free-air temperature and supply voltage (see Figure 16)

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
$t_{pd9}$	Propagation delay time from $\overline{IOW}$ (WR MCR) high to $\overline{RTS}$ (DTR) low/high	$C_L = 100$ pF		100	ns
$t_{pd10}$	Propagation delay time from modem input ( $\overline{CTS}$ , $\overline{DSR}$ ) low/high to interrupt high	$C_L = 100$ pF		170	ns
$t_{pd11}$	Propagation delay time from $\overline{IOR}$ (RD MSR) high to interrupt low	$C_L = 100$ pF		140	ns
$t_{pd12}$	Propagation delay time from $\overline{RI}$ high to interrupt high	$C_L = 100$ pF		170	ns

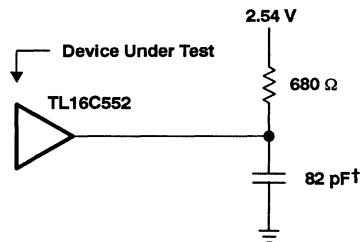
## parallel port timing requirements over recommended ranges of supply voltage and operating free-air temperature (see Figure 17)

		MIN	MAX	UNIT
$t_{s07}$	Setup time, data valid before $\overline{STB}$ low	1		$\mu$ s
$t_{h6}$	Hold time, data valid after $\overline{STB}$ high	1		$\mu$ s
$t_{w6}$	Pulse duration, $\overline{STB}$ low	1	500	$\mu$ s
$t_{d10}$	Delay time, BUSY high to $\overline{ACK}$ low	Defined by printer		
$t_{d11}$	Delay time, BUSY low to $\overline{ACK}$ low	Defined by printer		
$t_{w6}$	Pulse duration, $\overline{ACK}$ low	Defined by printer		
$t_{w7}$	Pulse duration, BUSY high	Defined by printer		
$t_{d12}$	Delay time, BUSY high after $\overline{STB}$ high	Defined by printer		

**PARAMETER MEASUREMENT INFORMATION**

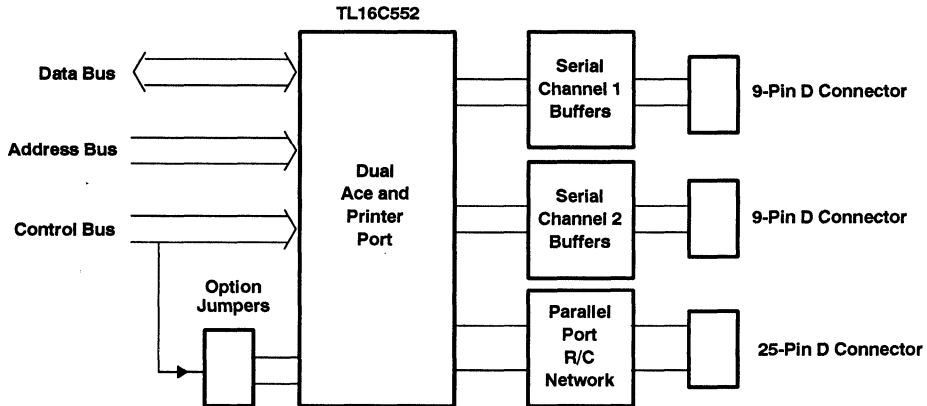


**Figure 3. Clock Input (CLK) Voltage Waveform**



†Includes scope and jig capacitance

**Figure 4. Output Load Circuit**

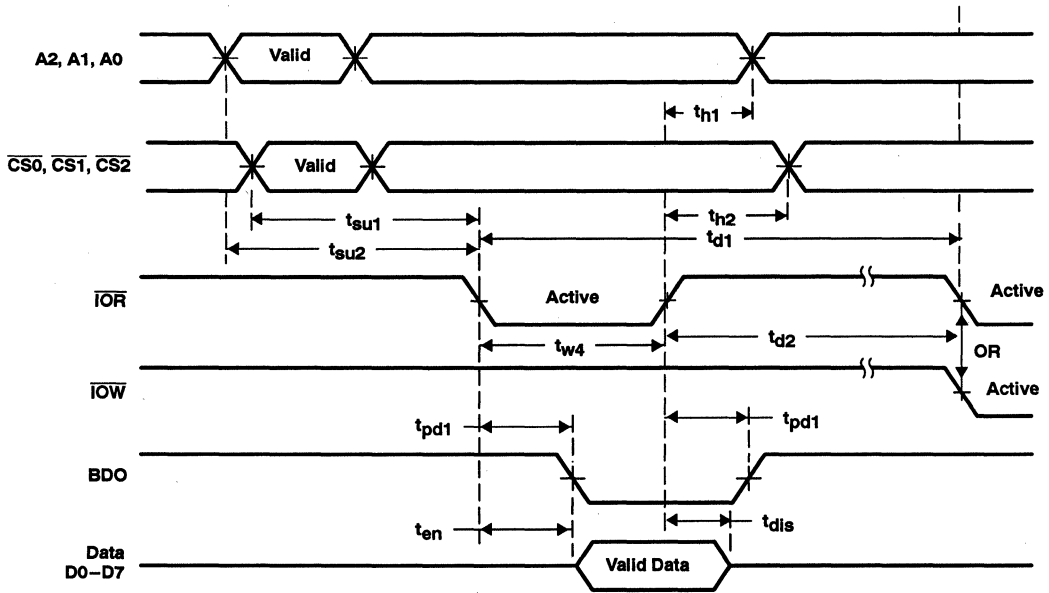


**Figure 5. Basic Test Configuration**

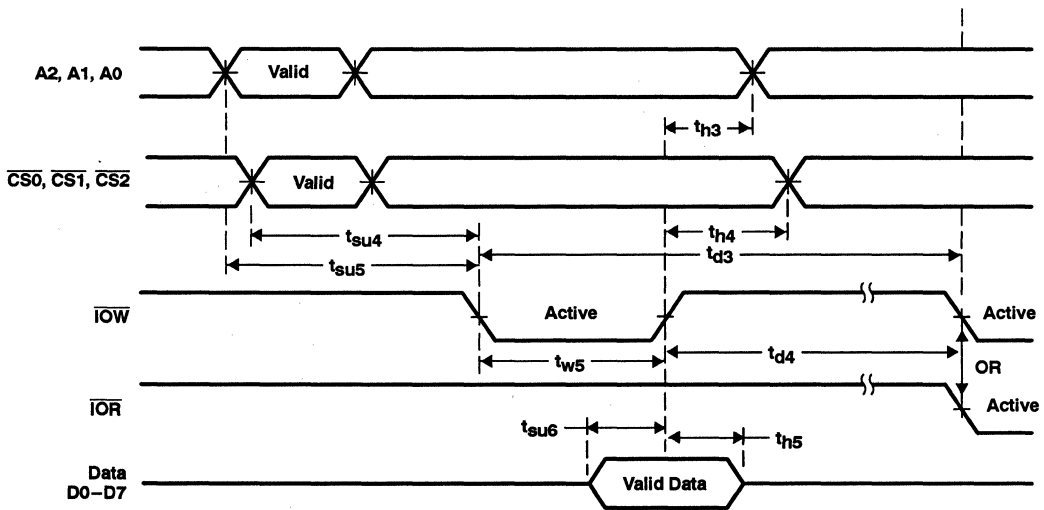
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**PARAMETER MEASUREMENT INFORMATION**



**Figure 6. Read Cycle Timing Waveforms**



**Figure 7. Write Cycle Timing Waveforms**

**PARAMETER MEASUREMENT INFORMATION**

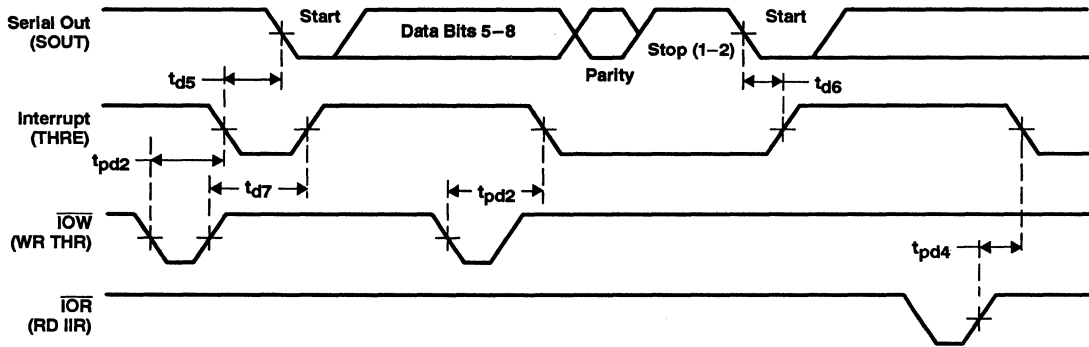


Figure 8. Transmitter Timing Waveforms

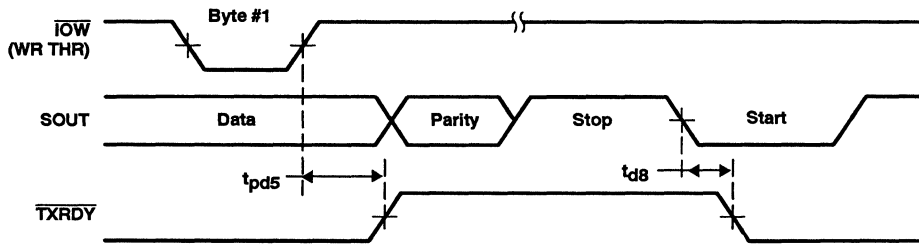


Figure 9. Transmitter Ready Mode 0 Timing Waveforms

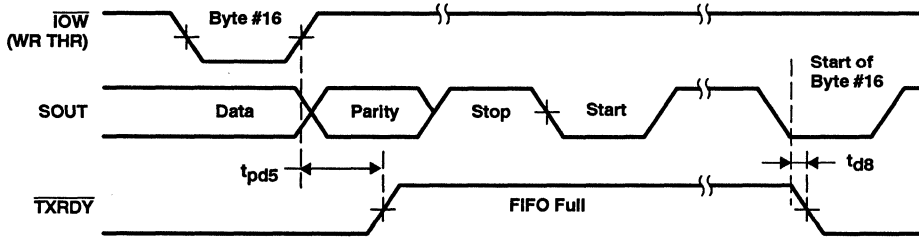
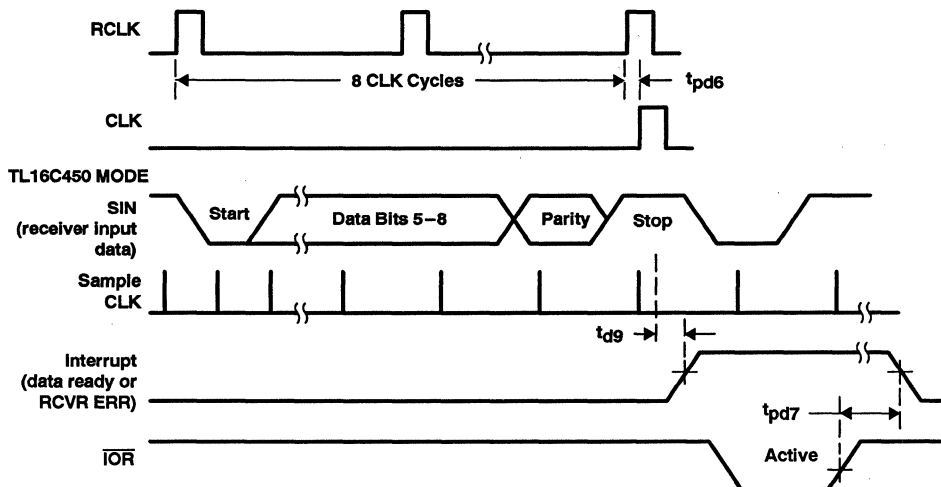


Figure 10. Transmitter Ready Mode 1 Timing Waveforms

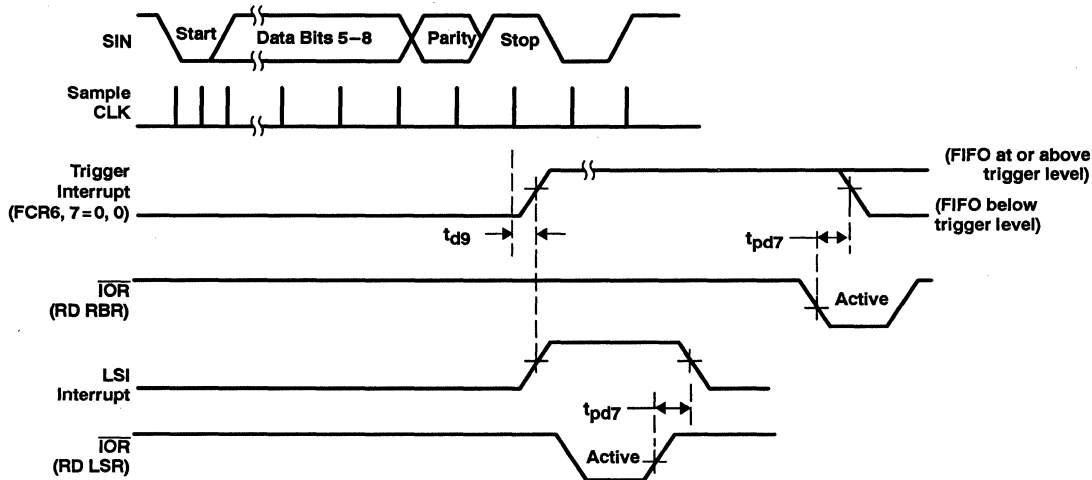
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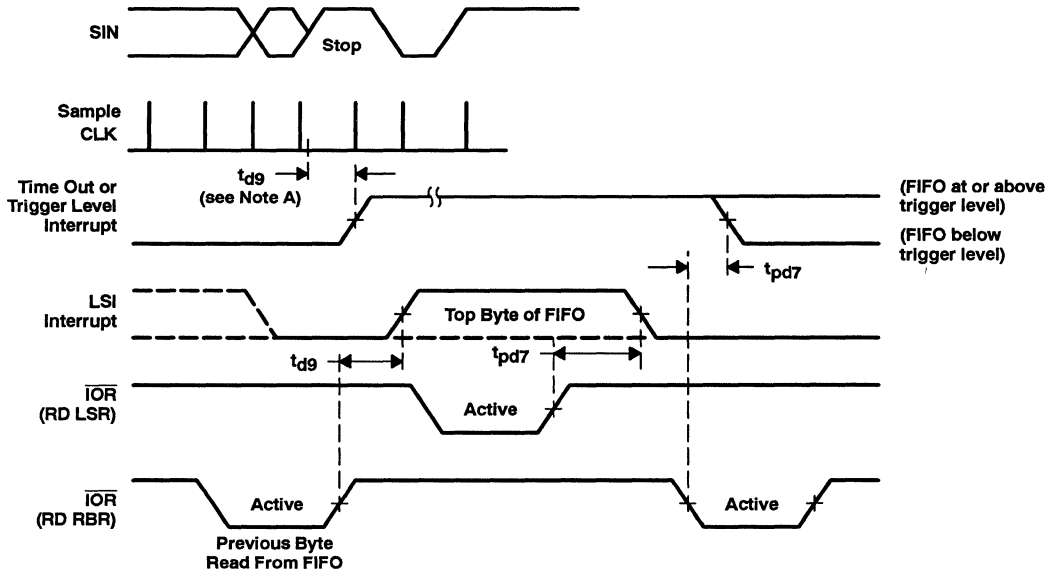


**Figure 11. Receiver Timing**

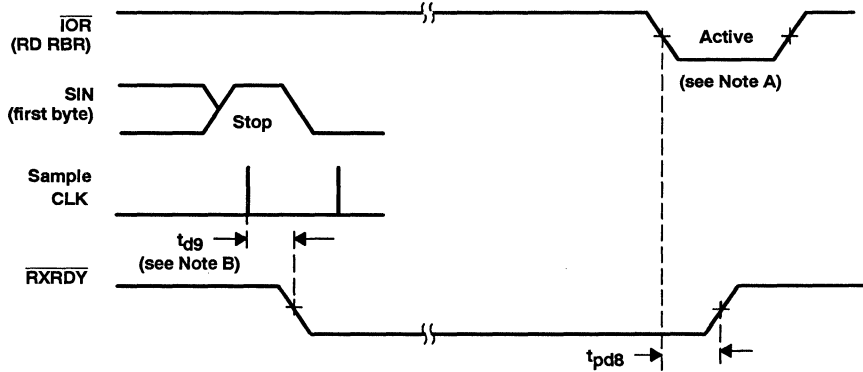


**Figure 12. Receiver FIFO First Byte (Sets RDR)**

**PARAMETER MEASUREMENT INFORMATION**



**Figure 13. Receiver FIFO After First Byte (After RDR Set)**



**Figure 14. Receiver Ready-Mode 0**

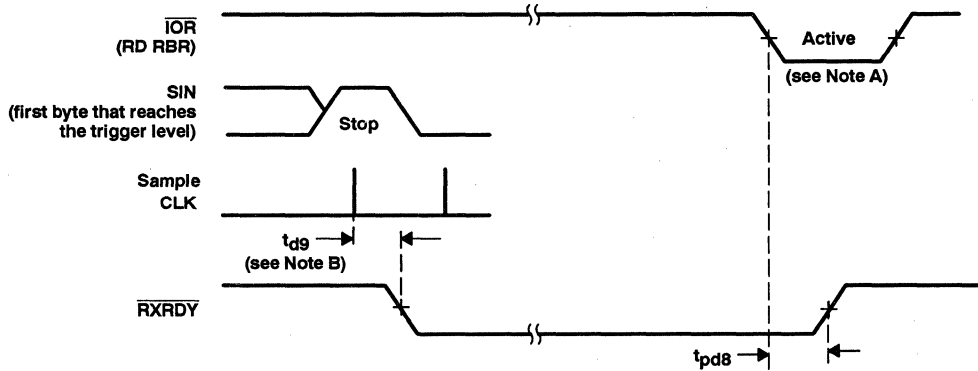
- NOTES: A. This is the reading of the last byte in the FIFO.  
 B. If FCR0=1, then  $t_{d9}$  = 3 RCLK cycles. For a timeout interrupt,  $t_{d9}$  = 8 RCLK cycles.



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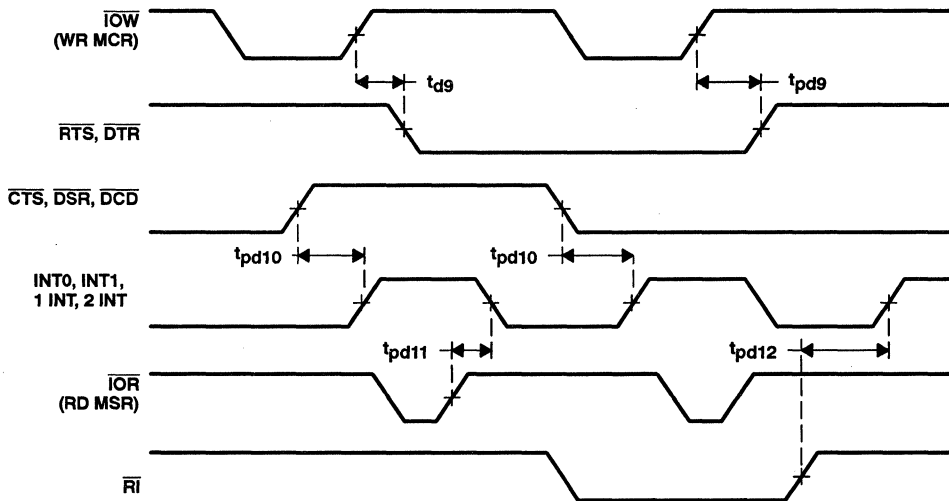
**PARAMETER MEASUREMENT INFORMATION**



NOTES: A. This is the reading of the last byte in the FIFO.

B. If FCR0=1, then  $t_{d9} = 3$  RCLK cycles. For a trigger change level interrupt,  $t_{d9} = 8$  RCLK

**Figure 15. Receiver Ready-Mode 1**



**Figure 16. Modem Timing**

PARAMETER MEASUREMENT INFORMATION

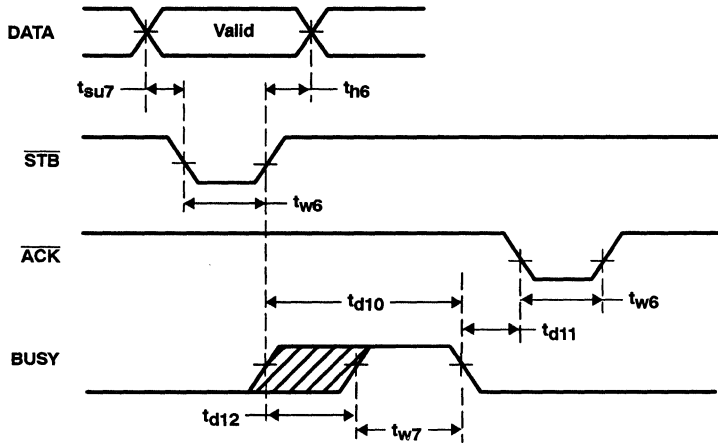


Figure 17. Parallel Port Timing

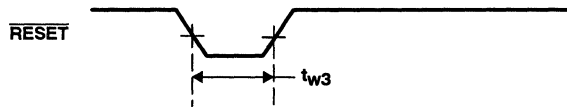


Figure 18. RESET Voltage Waveform



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<b>Universal Async Receivers/Transmitters</b>	<b>3</b>
<b>Explanation of Logic Symbols</b>	<b>4</b>
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# 4

## Explanation of Logic Symbols

# Explanation of Logic Symbols

by F.A. Mann

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Institute of Electrical and Electronics Engineers, Inc.  
IEEE Standards Office  
345 East 47th Street  
New York, N.Y. 10017

International Electrotechnical Commission (IEC) publications may be purchased from:  
American National Standards Institute, Inc.  
1430 Broadway  
New York, N.Y. 10018

## 1 Introduction

The International Electrotechnical Commission (IEC) has been developing a very powerful symbolic language that can show the relationship of each input of a digital logic circuit to each output without showing explicitly the internal logic. At the heart of the system is dependency notation, which will be explained in Section 4.

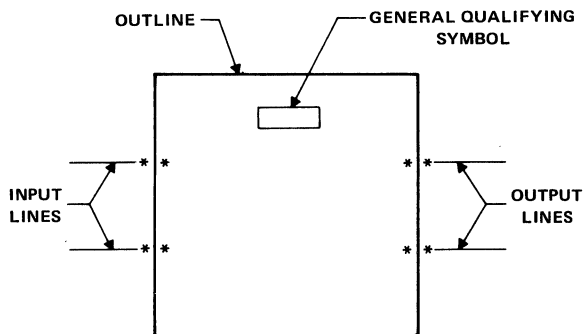
The system was introduced in the USA in a rudimentary form in IEEE/ANSI Standard Y32.14-1973. Lacking at that time a complete development of dependency notation, it offered little more than a substitution of rectangular shapes for the familiar distinctive shapes for representing the basic functions of AND, OR, negation, etc. This is no longer the case.

Internationally, Working Group 2 of IEC Technical Committee TC-3 has prepared a new document (Publication 617-12) that consolidates the original work started in the mid 1960's and published in 1972 (Publication 117-15) and the amendments and supplements that have followed. Similarly for the USA, IEEE Committee SCC 11.9 has revised the publication IEEE Std 91/ANSI Y32.14. Now numbered simply ANSI/IEEE Std 91-1984, the IEEE standard contains all of the IEC work that has been approved, and also a small amount of material still under international consideration. Texas Instruments is participating in the work of both organizations and this document introduces new logic symbols in accordance with the new standards. When changes are made as the standards develop, future editions will take those changes into account.

The following explanation of the new symbolic language is necessarily brief and greatly condensed from what the standards publications now contain. This is not intended to be sufficient for those people who will be developing symbols for new devices. It is primarily intended to make possible the understanding of the symbols used in this data book and is somewhat briefer than the explanation that appears in several of TI's data books on digital logic. However, it includes a new section (6.0) that explains several symbols for actual devices in detail. This has proven to be a powerful learning aid.

## 2 Symbol Composition

A symbol comprises an outline or a combination of outlines together with one or more qualifying symbols. The shape of the symbols is not significant. As shown in Figure 1, general qualifying symbols are used to tell exactly what logical operation is performed by the elements. Table 1 shows general qualifying symbols defined in the new standards. Input lines are placed on the left and output lines are placed on the right. When an exception is made to that convention, the direction of signal flow is indicated by an arrow as shown in Figure 9.



\*Possible positions for qualifying symbols relating to inputs and outputs

Figure 1. Symbol Composition



### 3 Qualifying Symbols

#### 3.1 General Qualifying Symbols

Table 1 shows general qualifying symbols defined by ANSI/IEEE Standard 91. These characters are placed near the top center or the geometric center of a symbol or symbol element to define the basic function of the device represented by the symbol or of the element.

X/Y is the general qualifying symbol for identifying coders, code converters, and level converters. X and Y may be used in their own right to stand for some code or either or both may be replaced by some other indication of the code or level such as BCD or TTL. As might be expected, interface circuits make frequent use of this set of qualifying symbols.

Table 1. General Qualifying Symbols

SYMBOL	DESCRIPTION
&	AND gate or function
≥ 1	OR gate or function. The symbol was chosen to indicate that at least one active input is needed to activate the output.
= 1	Exclusive OR. One and only one input must be active to activate the output.
1	A simple 1-input gate or element
▷ or ◁	A buffer or element with more than usual output capability (symbol is oriented in the direction of signal flow).
⏏	Schmitt trigger; element with hysteresis
X/Y	Coder, code converter, level converter
	The following are examples of subsets of this general class of qualifying symbol used in this book.
BCD/7-SEG	BCD to 7-segment display driver
TTL/MOS	TTL to MOS level converter
CMOS/PLASMA DISP	Plasma-display driver with CMOS-compatible inputs
MOS/LED	Light-emitting-diode driver with MOS-compatible inputs
CMOS/VAC FLUOR DISP	Vacuum-fluorescent display driver with CMOS-compatible inputs
CMOS/EL DISP	Electroluminescent display driver with CMOS-compatible inputs
TTL/GAS DISCH DISPLAY	Gas-discharge display driver with TTL-compatible inputs
SRGm	Shift register. m = number of bits.

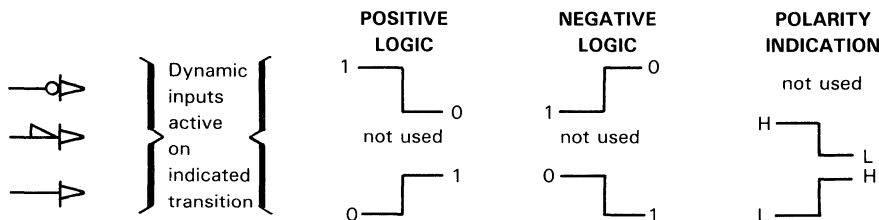
#### 3.2 Qualifying Symbols for Inputs and Outputs

Qualifying symbols for inputs and outputs are shown in Table 2 and many will be familiar to most users, a likely exception being the logic polarity symbol for directly indicating active-low inputs and outputs. The older logic negation indicator means that the external 0 state produces the internal 1 state. The internal 1 state means the active state. Logic negation may be used in pure logic diagrams; in order to tie the external 1 and 0 logic states to the levels H (high) and L (low), a statement of whether positive logic (1 = H, 0 = L) or negative logic (1 = L, 0 = H) is being used is required or must be assumed. Logic polarity indicators eliminate the need for calling out the logic convention and are used in this data book in the symbology for actual devices. The presence of the triangle polarity indicator indicates that the L logic level will produce the internal 1 state (the active state) or that, in the case of an output, the internal 1 state will produce the external L level. Note how the active direction of transition for a dynamic input is indicated in positive logic, negative logic, and with polarity indication.

When nonstandardized information is shown inside an outline, it is usually enclosed in square brackets [like these]. The square brackets are omitted when associated with a nonlogic input, which is indicated by an X superimposed on the connection line outside the symbol.

**Table 2. Qualifying Symbols for Inputs and Outputs**

	Logic negation at input. External 0 produces internal 1.
	Logic negation at output. Internal 1 produces external 0.
	Active-low input. Equivalent to  in positive logic
	Active-low output. Equivalent to  in positive logic
	Active-low input in the case of right-to-left signal flow
	Active-low output in the case of right-to-left signal flow
	Signal flow from right to left. If not otherwise indicated, signal flow is from left to right.
	Bidirectional signal flow



	Nonlogic connection. A label inside the symbol will usually define the nature of this pin.
	Input for analog signals (on a digital symbol) (see Figure 11)
	Input for digital signals (on an analog symbol) (see Figure 11)

### 3.3 Symbols Inside the Outline



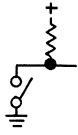


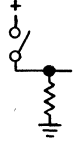






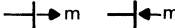


Table 3 shows some symbols used inside the outline. Note particularly that open-collector (open-drain), open-emitter (open-source), and three-state outputs have distinctive symbols. Also note that an EN input affects all of the outputs of the element and has no effect on inputs. An EN input affects all the external outputs of the element in which it is placed, plus the external outputs of any elements shown to be influenced by that element. It has no effect on inputs. When an enable input affects only certain outputs, affects outputs located outside the indicated influence of the element in which the enable input is placed, and/or affects one or more inputs, a form of dependency notation will indicate this (see 4.9). The effects of the EN input on the various types of outputs are shown.

It is particularly important to note that a D input is always the data input of a storage element. At its internal 1 state, the D input sets the storage element to its 1 state, and at its internal 0 state it resets the storage element to its 0 state.

The binary grouping symbol will be explained more fully in Section 6.11. Binary-weighted inputs are arranged in order and the binary weights of the least significant and the most significant lines are indicated by numbers. In this document weights of input and output lines will be represented by powers of two usually only when the binary grouping symbol is used, otherwise decimal numbers will be used. The grouped inputs generate an internal number on which a mathematical function can be performed or that can be an identifying number for dependency notation. This number is the sum of the weights (1, 2, 4, . . . 2<sup>n</sup>) of those input standing at their 1 states. A frequent use is in addresses for memories.

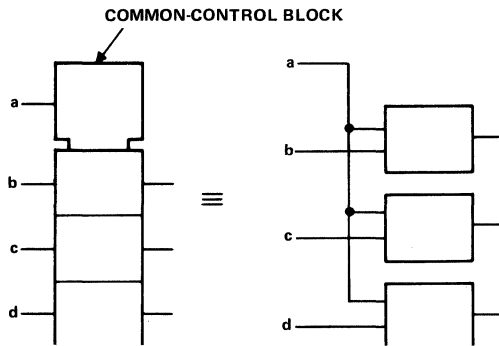
Reversed in direction, the binary grouping symbol can be used with outputs. The concept is analogous to that for the inputs and the weighted outputs will indicate the internal number assumed to be developed within the circuit.

**Table 3. Symbols Inside the Outline**

	Bithreshold input (input with hysteresis)	
	N-P-N open-collector or similar output that can supply a relatively low-impedance L level when not turned off. Requires external pull-up. Capable of positive-logic wired-AND connection.	
	Passive-pull-up output is similar to N-P-N open-collector output but is supplemented with a built-in passive pull-up.	
	N-P-N open-emitter or similar output that can supply a relatively low-impedance H level when not turned off. Requires external pull-down. Capable of positive-logic wired-OR connection.	
	Passive-pull-down output is similar to N-P-N open-emitter output but is supplemented with a built-in passive pull-down.	
	3-state output	
	Output with more than usual output capability (symbol is oriented in the direction of signal flow).	
	Enable input When at its internal 1-state, all outputs are enabled. When at its internal 0-state, open-collector, open-emitter outputs, and three-state outputs at external high-impedance state, and all other outputs (i.e., totem-poles) are at the internal 0-state.	
J, K, R, S, T	Usual meanings associated with flip-flops (e.g., R = reset, T = toggle)	
	Data input to a storage element equivalent to:	
	Shift right (left) inputs, m = 1, 2, 3, etc. If m = 1, it is usually not shown.	
	Binary grouping. m is highest power of 2. Produces a number equal to the sum of the weights of the active inputs	
	Input line grouping . . . indicates two or more terminals used to implement a single logic input. e.g., differential inputs.	

### 3.4 Combinations of Outlines and Internal Connections

When a circuit has one or more inputs that are common to more than one element of the circuit, the common-control block may be used. This is the only distinctively shaped outline used in the IEC system. Figure 2 shows that unless otherwise qualified by dependency notation, an input to the common-control block is an input to each of the elements below the common-control block.



**Figure 2. Common-Control Block**

The outlines of elements may be embedded within one another or abutted to form complex elements, in which case the following rules apply. There is no logic connection between elements when the line common to their outlines is in the direction of signal flow. There is at least one logic connection when the line common to two outlines is perpendicular to the direction of signal flow. If no indications are shown on either side of the common line, it is assumed that there is only one logic connection. If more than one internal connection exists between adjacent elements, the number of connections will be clarified by the use of one or more of the internal connection symbols from Table 4 and/or appropriate qualifying symbols or dependency notation.

**Table 4. Symbols for Internal Connections**





	Internal connection. 1 state on left produces 1 state on right.
	Negated internal connection. 1 state on left produces 0 state on right.
	Dynamic internal connection. Transition from 0 to 1 on left produces transitory 1 state on right.
	Dynamic internal connection. Transition from 1 to 0 on left produces transitory 1 state on right.

Table 4 shows symbols that are used to represent internal connection with specific characteristics. The first is a simple noninverting connection, the second is inverting, the third is dynamic. As with this symbol and an external input line, the transition from 0 to 1 on the left produces a momentary 1-state on the right. The fourth symbol is similar except that the active transition on the left is from 1 to 0.

Only logic states, not levels, exist inside symbols. The negation symbol ( ) is used internally even when direct polarity indication ( ) is used externally.

In an array of elements, if the same general qualifying symbol and the same qualifying symbols associated with inputs and outputs would appear inside each of the elements of the array, these qualifying symbols are usually shown only in the first element. This is done to reduce clutter and to save time in recognition. Similarly, large identical elements that are subdivided into smaller elements may each be represented by an unsubdivided outline. The SN75163B symbol (see 6.5) illustrates this principle.

## 4 Dependency Notation

Some readers will find it more to their liking to skip this section and proceed to the explanation of the symbols for a few actual devices in 6.0. Reference will be made there to various parts of this section as it is needed. If this procedure is followed, it is recommended that 5.0 be read after 6.0 and then all of 4.0 be reread.

### 4.1 General Explanation

Dependency notation is the powerful tool that sets the IEC symbols apart from previous systems and makes compact, meaningful, symbols possible. It provides the means of denoting the relationship between inputs, outputs, or inputs and outputs without actually showing all the elements and interconnections involved. The information provided by dependency notation supplements that provided by the qualifying symbols for an element's function.

In the convention for the dependency notation, use will be made of the terms "affecting" and "affected." In cases where it is not evident which inputs must be considered as being the affecting or the affected ones (e.g., if they stand in an AND relationship), the choice may be made in any convenient way.

So far, eleven types of dependency have been defined but only the eight used in this book are explained. They are listed below in the order in which they are presented and are summarized in Table 5 following 4.10.2.

Section	Dependency Type or Other Subject
4.2	G, AND
4.3	General Rules for Dependency Notation
4.4	V, OR
4.5	N, Negate (Exclusive-OR)
4.6	Z, Interconnection
4.7	X, Transmission
4.8	C, Control
4.9	EN, Enable
4.10	M, Mode

## 4.2 G (AND) Dependency

A common relationship between two signals is to have them ANDed together. This has traditionally been shown by explicitly drawing an AND gate with the signals connected to the inputs of the gate. The 1972 IEC publication and the 1973 IEEE/ANSI standard showed several ways to show this AND relationship using dependency notation. While ten other forms of dependency have since been defined, the ways to invoke AND dependency are now reduced to one.

In Figure 3 input *b* is ANDed with input *a* and the complement of *b* is ANDed with *c*. The letter G has been chosen to indicate AND relationships and is placed at input *b*, inside the symbol. A number considered appropriate by the symbol designer (1 has been used here) is placed after the letter G and also at each affected input. Note the bar over the 1 at input *c*.

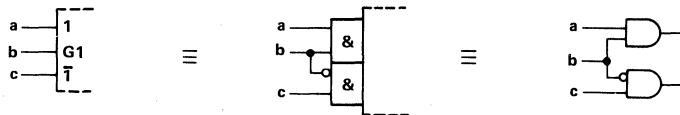


Figure 3. G Dependency Between Inputs

In Figure 4, output *b* affects input *a* with an AND relationship. The lower example shows that it is the internal logic state of *b*, unaffected by the negation sign, that is ANDed. Figure 5 shows input *a* to be ANDed with a dynamic input *b*.

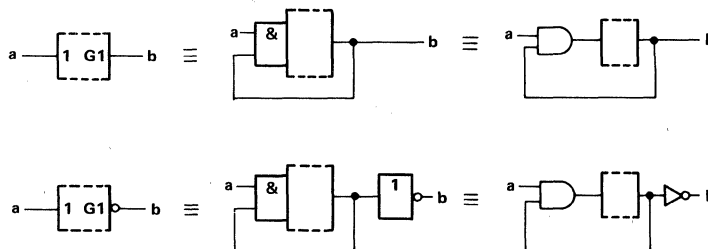


Figure 4. G Dependency Between Outputs and Inputs

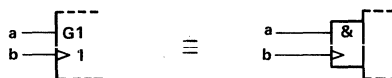


Figure 5. G Dependency with a Dynamic Input

The rules for G dependency can be summarized thus:

When a  $Gm$  input or output ( $m$  is a number) stands at its internal 1 state, all inputs and outputs affected by  $Gm$  stand at their normally defined internal logic states. When the  $Gm$  input or output stands at its 0 state, all inputs and outputs affected by  $Gm$  stand at their internal 0 states.

### 4.3 Conventions for the Application of Dependency Notation in General

The rules for applying dependency relationships in general follow the same pattern as was illustrated for G dependency.

Application of dependency notation is accomplished by:

1. Labeling the input (or output) *affecting* other inputs or outputs with the letter symbol indicating the relationship involved (e.g., G for AND) followed by an identifying number, appropriately chosen, and
2. Labeling each input or output *affected* by that affecting input (or output) with that same number.

If it is the complement of the internal logic state of the affecting input or output that does the affecting, then a bar is placed over the identifying numbers at the affected inputs or outputs (Figure 3).

If two affecting inputs or outputs have the same letter and same identifying number, they stand in an OR relationship to each other (Figure 6).

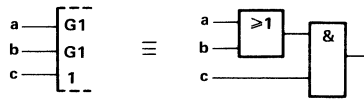


Figure 6. ORed Affecting Inputs

If the affected input or output requires a label to denote its function (e.g., "D"), this label will be *prefixed* by the identifying number of the affecting input (Figure 12).

If an input or output is affected by more than one affecting input, the identifying numbers of each of the affecting inputs will appear in the label of the affected one, separated by commas. The normal reading order of these numbers is the same as the sequence of the affecting relationships (Figure 12).

### 4.4 V (OR) Dependency

The symbol denoting OR dependency is the letter V (Figure 7).

When a  $Vm$  input or output stands at its internal 1 state, all inputs and outputs affected by  $Vm$  stand at their internal 1 states. When the  $Vm$  input or output stands at its internal 0 state, all inputs and outputs affected by  $Vm$  stand at their normally defined internal logic states.

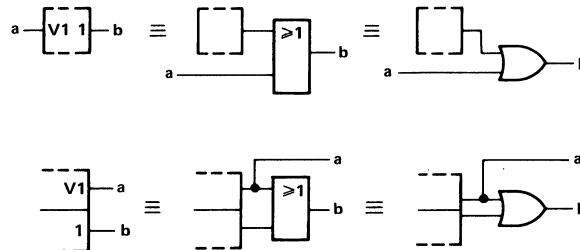


Figure 7. V (OR) Dependency

#### 4.5 N (Negate) (Exclusive-OR) Dependency

The symbol denoting negate dependency is the letter N (Figure 8). Each input or output affected by an  $Nm$  input or output stands in an Exclusive-OR relationship with the  $Nm$  input or output.

When an  $Nm$  input or output stands at its internal 1 state, the internal logic state of each input and each output affected by  $Nm$  is the complement of what it would otherwise be. When an  $Nm$  input or output stands at its internal 0 state, all inputs and outputs affected by  $Nm$  stand at their normally defined internal logic states.

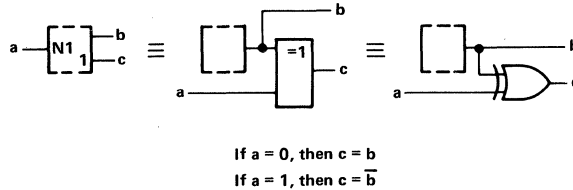


Figure 8. N (Negate) (Exclusive-OR) Dependency

#### 4.6 Z (Interconnection) Dependency

The symbol denoting interconnection dependency is the letter Z.

Interconnection dependency is used to indicate the existence of internal logic connections between inputs, outputs, internal inputs, and/or internal outputs.

The internal logic state of an input or output affected by a  $Zm$  input or output will be the same as the internal logic state of the  $Zm$  input or output, unless modified by additional dependency notation (Figure 9).

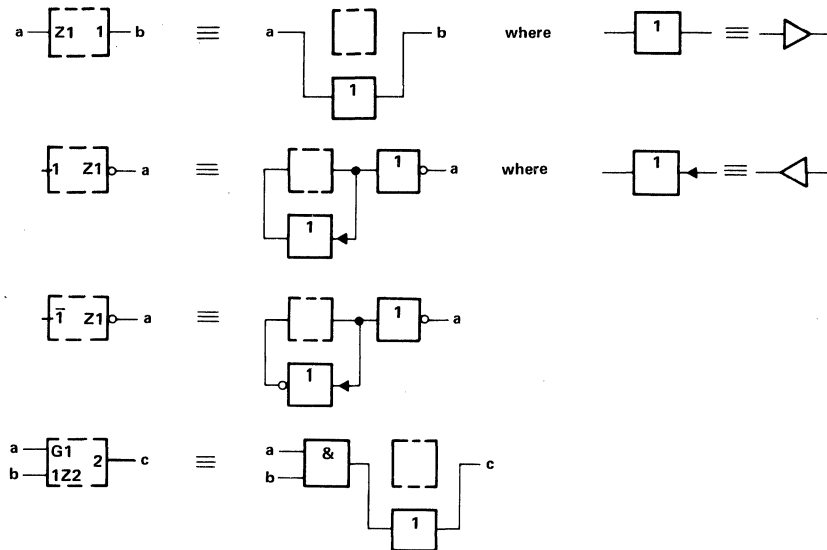


Figure 9. Z (Interconnection) Dependency

#### 4.7 X (Transmission) Dependency

The symbol denoting transmission dependency is the letter X.

Transmission dependency is used to indicate controlled bidirectional connections between affected input/output ports (Figure 10).

When an  $X_m$  input or output stands at its internal 1 state, all input-output ports affected by this  $X_m$  input or output are bidirectionally connected together and stand at the same internal logic state or analog signal level. When an  $X_m$  input or output stands at its internal 0 state, the connection associated with this set of dependency notation does not exist.

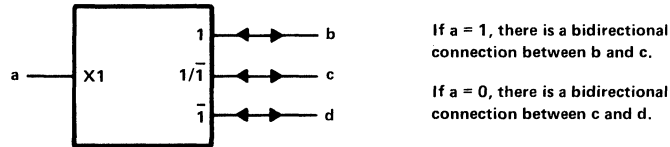


Figure 10. X (Transmission) Dependency

Although the transmission paths represented by X dependency are inherently bidirectional, use is not always made of this property. This is analogous to a piece of wire, which may be constrained to carry current in only one direction. If this is the case in a particular application, then the directional arrows shown in Figures 10 and 11 would be omitted.

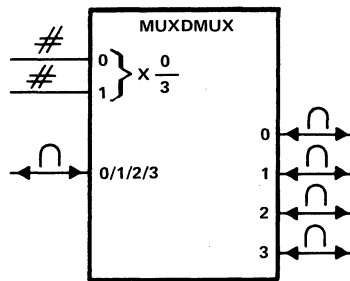


Figure 11. Analog Data Selector (Multiplexer/Demultiplexer)

#### 4.8 C (Control) Dependency

The symbol denoting control dependency is the letter C.

Control inputs are usually used to enable or disable the data (D, J, K, R, or S) inputs of storage elements. They may take on their internal 1 states (be active) either statically or dynamically. In the latter case the dynamic input symbol is used as shown in the second example of Figure 12.

When a  $C_m$  input or output stands at its internal 1 state, the inputs affected by  $C_m$  have their normally defined effect on the function of the element, i.e., these inputs are enabled. When a  $C_m$  input or output stands at its internal 0 state, the inputs affected by  $C_m$  are disabled and have no effect on the function of the element.



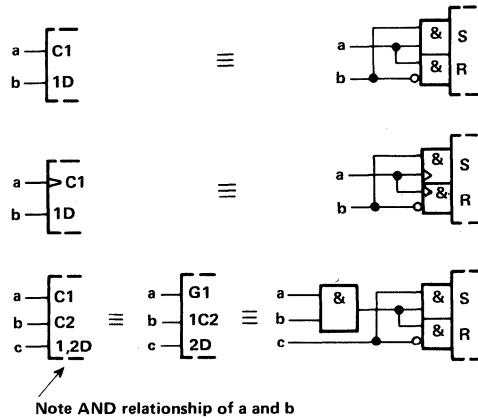


Figure 12. C (Control) Dependency

#### 4.9 EN (Enable) Dependency

The symbol denoting enable dependency is the combination of letters EN.

An  $EN_m$  input has the same effect on outputs as an EN input, see 3.3, but it affects only those outputs labeled with the identifying number  $m$ . It also affects those inputs labeled with the identifying number  $m$ . By contrast, an EN input affects all outputs and no inputs. The effect of an  $EN_m$  input on an affected input is identical to that of a  $C_m$  input (Figure 13).

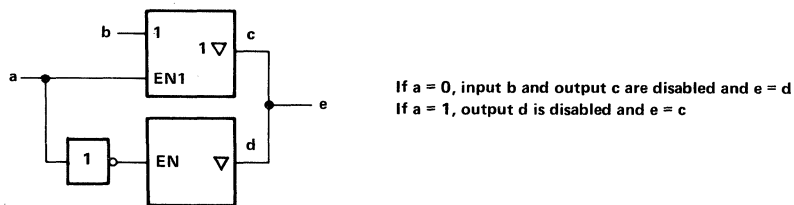


Figure 13. EN (Enable) Dependency

When an  $EN_m$  input stands at its internal 1 state, the inputs affected by  $EN_m$  have their normally defined effect on the function of the element and the outputs affected by this input stand at their normally defined internal logic states, i.e., these inputs and outputs are enabled.

When an  $EN_m$  input stands at its internal 0 state, the inputs affected by  $EN_m$  are disabled and have no effect on the function of the element, and the outputs affected by  $EN_m$  are also disabled. Open-collector outputs are turned off, three-state outputs stand at their high-impedance state, and all other outputs (e.g., totem-pole outputs) stand at their internal 0 states.

#### 4.10 M (MODE) Dependency

The symbol denoting mode dependency is the letter M.

Mode dependency is used to indicate that the effects of particular inputs and outputs of an element depend on the mode in which the element is operating.

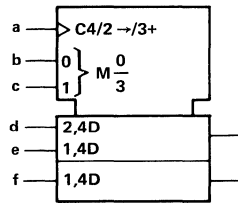
If an input or output has the same effect in different modes of operation, the identifying numbers of the relevant affecting  $M_m$  inputs will appear in the label of that affected input or output between parentheses and separated by solidi, e.g.,  $(1/2)CT=0 \equiv 1CT=0/2CT=0$  where 1 and 2 refer to M1 and M2.

#### 4.10.1 M Dependency Affecting Inputs

M dependency affects inputs the same as C dependency. When an *Mm* input or *Mm* output stands at its internal 1 state, the inputs affected by this *Mm* input or *Mm* output have their normally defined effect on the function of the element, i.e., the inputs are enabled.

When an *Mm* input or *Mm* output stands at its internal 0 state, the inputs affected by this *Mm* input or *Mm* output have no effect on the function of the element. When an affected input has several sets of labels separated by solidi (e.g., C4/2→/3+), any set in which the identifying number of the *Mm* input or *Mm* output appears has no effect and is to be ignored. This represents disabling of some of the functions of a multifunction input.

The circuit in Figure 14 has two inputs, **b** and **c**, that control which one of four modes (0, 1, 2, or 3) will exist at any time. Inputs **d**, **e**, and **f** are D inputs subject to dynamic control (clocking) by the **a** input. The numbers 1 and 2 are in the series chosen to indicate the modes so inputs **e** and **f** are only enabled in mode 1 (for parallel loading) and input **d** is only enabled in mode 2 (for serial loading). Note that input **a** has three functions. It is the clock for entering data. In mode 2, it causes right shifting of data, which means a shift away from the control block. In mode 3, it causes the contents of the register to be incremented by one count.



Note that all operations are synchronous.

In MODE 0 ( $b = 0, c = 0$ ), the outputs remain at their existing states as none of the inputs has an effect.

In MODE 1 ( $b = 1, c = 0$ ), parallel loading takes place thru inputs **e** and **f**.

In MODE 2 ( $b = 0, c = 1$ ), shifting down and serial loading thru input **d** take place.

In MODE 3 ( $b = c = 1$ ), counting up by increment of 1 per clock pulse takes place.

Figure 14. M (Mode) Dependency Affecting Inputs

#### 4.10.2 M Dependency Affecting Outputs

When an *Mm* input or *Mm* output stands at its internal 1 state, the affected outputs stand at their normally defined internal logic states, i.e., the outputs are enabled.

When an *Mm* input or *Mm* output stands at its internal 0 state, at each affected output any set of labels containing the identifying number of that *Mm* input or *Mm* output has no effect and is to be ignored. When an output has several different sets of labels separated by solidi (e.g., 2,4/3,5), only those sets in which the identifying number of this *Mm* input or *Mm* output appears are to be ignored.

Figure 15 shows a symbol for a device whose output can behave like either a 3-state output or an open-collector output depending on the signal applied to input **a**. Mode 1 exists when input **a** stands at its internal 1 state and, in that case, the three-state symbol applies and the open-element symbol has no effect. When  $a = 0$ , mode 1 does not exist so the three-state symbol has no effect and the open-element symbol applies.

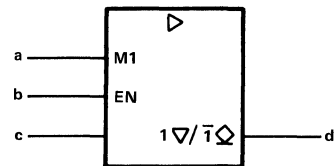


Figure 15. Type of Output Determined by Mode

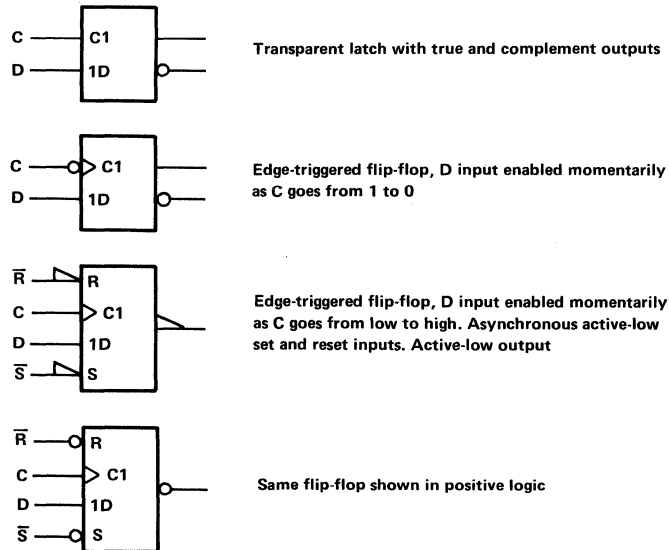
**Table 5. Summary of Dependency Notation**

TYPE OF DEPENDENCY	LETTER SYMBOL*	AFFECTING INPUT AT ITS 1-STATE	AFFECTING INPUT AT ITS 0-STATE
Control	C	Permits action	Prevents action
Enable	EN	Permits action	Prevents action of inputs ◇ outputs turned off ▽ outputs at external high impedance Other outputs at internal 0 state
AND	G	Permits action	Imposes 0 state
Mode	M	Permits action (mode selected)	Prevents action (mode not selected)
Negate (Ex-NOR)	N	Complements state	No effect
OR	V	Imposes 1 state	Permits action
Transmission	X	Bidirectional connection exists	Bidirectional connection does not exist
Interconnection	Z	Imposes 1 state	Imposes 0 state

\* These letter symbols appear at the AFFECTING input (or output) and are followed by a number. Each input (or output) AFFECTED by that input is labeled with that same number.

## 5 Bistable Elements

The dynamic input symbol and dependency notation provide the tools to identify different types of bistable elements and make synchronous and asynchronous inputs easily recognizable (Figure 16).



**Figure 16. Latches and Flip-Flops**

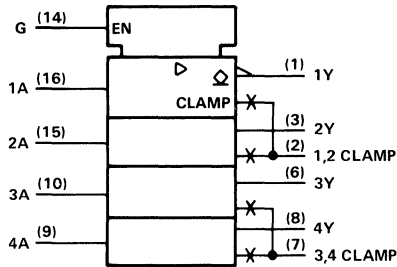
Transparent latches have a level-operated control input. The D input is active as long as the C input is at its internal 1 state. The outputs respond immediately. Edge-triggered elements accept data from D, J, K, R, or S inputs on the active transition of C.

Notice that synchronous inputs can be readily recognized by their dependency labels (a number preceding the functional label, 1D in these examples) compared to the asynchronous inputs (S and R), which are not dependent on the C inputs. Of course if the set and reset inputs were dependent on the C inputs, their labels would be similarly modified (e.g., 1S, 1R).

## 6 Examples of Actual Device Symbols

The symbols explained in this section include some of the most complex in this book. These were chosen, not to discourage the reader, but to illustrate the amount of information that can be conveyed. It is likely that if one reads these explanations and follows them reasonably well, most of the other symbols will seem simple indeed. The explanations are intended to be independent of each other so they may seem somewhat repetitious. However each illustrates new principles. They are arranged more or less in the order of complexity.

### 6.1 SN75437A Quadruple Peripheral Driver

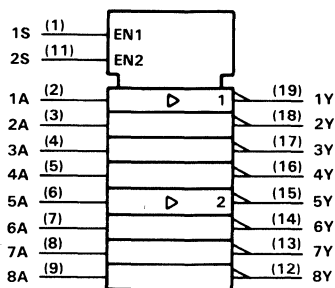


There are four identical sections. The symbology is complete for the first element; the absence of any symbology for the other elements indicates they are identical. The top two elements share a common output clamp, pin 2. This is shown to be a nonlogic connection by the superimposed X on the line. The function for this type of connection is indicated briefly and not necessarily exactly by a small amount of text within the symbol. The bottom two elements likewise share a common clamp.

Each element is shown to be an inverter with amplification (indicated by  $\triangleright$ ). Taking TTL as a reference, this means that either the input is sensitive to lower level signals, or the output has greater drive capability than usual. The latter applies in this case. The output is shown by  $\square$  to be open collector.

All the outputs share a common EN input, pin 14. See Figure 2 for an explanation of the common control block. When EN = 0 (pin 14 is low), the outputs, being open-collector types, are turned off and would be pulled high by an external pullup resistor.

### 6.2 SN75128 8-Channel Line Receiver

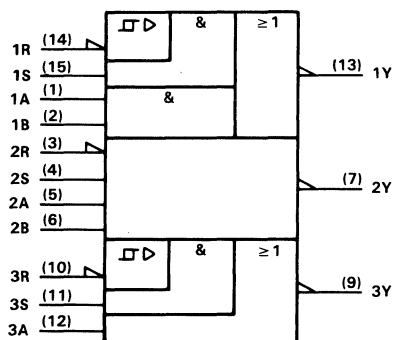


There are eight identical sections. The symbology is complete for the first element; the absence of any symbology for the next three elements indicates they are identical. Likewise the symbology is complete for the fifth element; the absence of any symbology for the next three elements indicates they are identical to the fifth.

Each element is shown to be an inverter with amplification (indicated by  $\triangleright$ ). Taking TTL as a reference, this means that either the input is sensitive to lower level signals, or the output has greater drive capability than usual. The former applies in this case. Since neither the symbol for open-collector ( $\square$ ) or 3-state ( $\nabla$ ) outputs is shown, the outputs are of the totem-pole type.

The top four outputs are shown to be affected by affecting input number 1, which is EN1, meaning they will be enabled if EN1 = 1 (pin 1 is high). See 4.9 for an explanation of EN dependency. If pin 1 is low, EN1 = 0 and the affected outputs will go to their inactive (high) levels. Similarly, the lower four outputs are controlled by pin 11.

### 6.3 SN75122 Triple Line Receivers

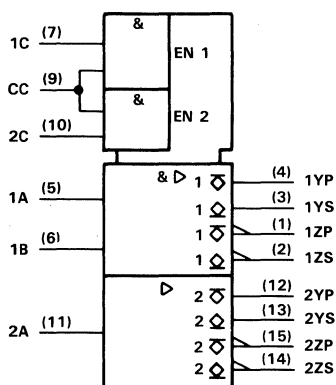


There are two identical sections. The symbology is complete for the first section; the absence of any symbology for the next section indicates it is identical. Likewise the symbology is complete for the third section, which is similar, but not identical, to the first and second.

The top section may be considered to be an OR element ( $\geq 1$ ) with two embedded ANDs (&), one of which has an active-low amplified input ( $\triangleright$ ) with hysteresis ( $\square$ ), pin 14. This is ANDed with pin 15 and the result is ORed with the AND of pins 1 and 2. The output of the OR, pin 13, is active-low.

The third section is identical to the first except that pin 12 has no input ANDed with it. Since neither the symbol for open-collector ( $\diamond$ ) or 3-state ( $\nabla$ ) outputs is shown, the outputs are of the totem-pole type.

### 6.4 SN75113 Differential Line Drivers with Split 3-State Outputs



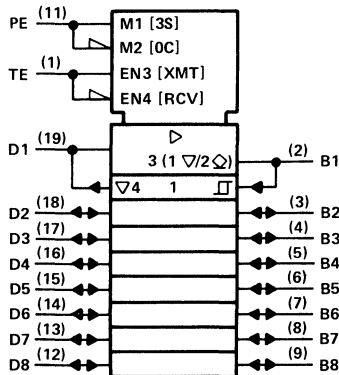
There are two similar elements in the array. The first is a 2-input AND element (indicated by &); the second has only a single input. Both elements are shown to have special amplification (indicated by  $\triangleright$ ). Taking TTL as a reference, this means that either the input is sensitive to lower level signals, or the output has greater drive capability than usual. The latter applies in this case.

Each element has four outputs. Pins 4 and 3 are a pair consisting of one open-emitter output ( $\nabla$ ) and one open-collector output ( $\diamond$ ). Relative to the AND function, both are active high. Pins 1 and 2 are a similar pair but relative to the AND function, both are active low. All outputs of a single, unsubdivided element always have identical internal logic states determined by the function of the element except when otherwise indicated by an associated symbol or label inside the element. Here there is no such contrary indication. All four outputs are shown to be affected by affecting input number 1, which is EN1,

meaning they will all be enabled if  $EN1 = 1$ . See 4.9 for an explanation of EN dependency. If  $EN1 = 0$ , all the affected outputs will be turned off. EN1 is the output of an AND gate (indicated by &) whose active-high inputs are pins 7 and 9. Both pins 7 and 9 must be high to enable the outputs of the top element. Assuming they are enabled and that pins 5 and 6 are both high, the internal state of all four outputs will be a 1. Pins 4 and 3 will both be high, pins 1 and 2 will both be low. The part is designed so that pins 3 and 4 may be connected together creating an active-high 3-state output. Likewise pins 1 and 2 may be connected together to create an active-low 3-state output.

All that has been said about the first element regarding its outputs and their enable inputs also applies to the second element. Pins 9 and 10 are the enable inputs in this case.

## 6.5 SN75163B Octal General-Purpose Interface Bus Transceiver



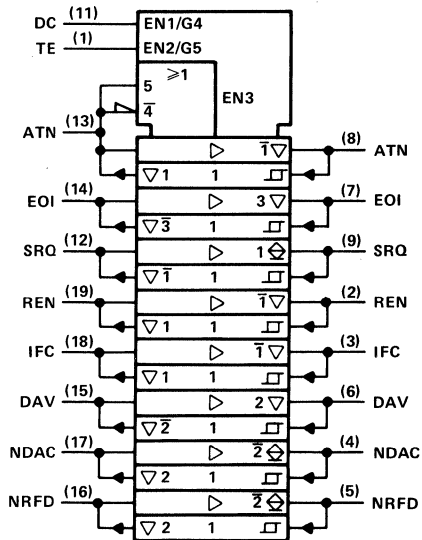
There are eight I/O ports on each side, pins 2 through 9 and 12 through 19. There are eight identical channels. The symbology is complete for the first channel; the absence of any symbology for the other channels indicates they are identical. The eight bidirectional channels each have amplification from left to right, that is, the outputs on the right have increased drive capability (indicated by  $\triangleright$ ), and the inputs on the right all have hysteresis (indicated by  $\square$ ).

The outputs on the left are shown to be 3-state outputs by the  $\nabla$ . They are also shown to be affected by affecting input number 4, which is EN4, meaning they will be enabled if EN4 = 1 (pin 1 is low). See 4.9 for an explanation of EN dependency. If EN4 = 0 (pin 1 is high), the affected outputs will go to their high-impedance (off) states.

The labeling at pin 2, which applies to all the outputs on the right, is unusual because the outputs themselves have an unusual feature. The label includes both the symbol for a 3-state output ( $\nabla$ ) and for an open-collector output ( $\diamond$ ), separated by a slash indicating that these are alternatives.

The symbol for the 3-state output is shown to be affected by affecting input number 1, which is M1, meaning the  $\nabla$  label is valid when M1 = 1 (pin 11 is high), but is to be ignored when M1 = 0 (pin 11 is low). See 4.10 for an explanation of M (mode) dependency. Likewise the symbol for the open-collector output is shown to be affected by affecting input number 2, which is M2, meaning the  $\diamond$  label is valid when M2 = 1 (pin 11 is low), but is to be ignored when M2 = 0 (pin 11 is high). These labels are enclosed in parentheses (used as in algebra); the numeral 3 indicates that in either case the output is affected by EN3. Thus the right-hand outputs will be off if pin 1 is low. It can now be seen that pin 1 is the direction control and pin 11 is used to determine whether the outputs are of the 3-state or open-collector variety.

## 6.6 SN75161B Octal IEEE Std 488 Interface Bus Transceiver



There are eight I/O ports on each side, pins 2 through 9 and 12 through 19. Pin 13 is not only an I/O port; the line running into the common-control block (see Figure 2) indicates that it also has control functions. Pins 1 and 11 are also controls. The eight bidirectional channels each have amplification from left to right, that is, the outputs on the right have increased drive capability (indicated by  $\blacktriangleright$ ), and the inputs on the right all have hysteresis (indicated by  $\blacktriangledown$ ). All of the outputs are shown to be of the 3-state type by the  $\blacktriangledown$  symbol except for the outputs at pins 9, 4, and 5, which are shown to have passive pullups by the  $\odot$  symbol.

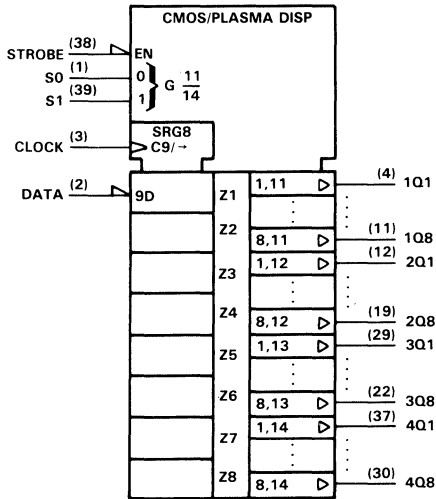
Starting with a typical I/O port, pin 18, the output portion is identified by an arrow indicating right-to-left signal flow and the three-state output symbol ( $\blacktriangledown$ ). This output is shown to be affected by affecting input number 1, which is EN1, meaning it will be enabled as an output if EN1 = 1 (pin 11 is high). See 4.9 for an explanation of EN dependency.

If pin 11 is low, EN1 = 0 and the output at pin 18 will be in its high-impedance (off) state. This also applies to the 3-state outputs at pins 13 and 19 and to the passive-pullup output at pin 9. On the other hand, the outputs at pins 8, 2, 3, and 12 all are affected by the complement of EN1. This is indicated by the bar over the 1 at each of those outputs. They are enabled only when pin 11 is low. Thus one function of pin 11 is to serve as direction control for the first, third, fourth, and fifth channels.

Similarly it can be seen that pin 1 serves as direction control for the sixth, seventh, and eighth channels. If pin 1 is high, transmission will be from left to right in the sixth channel, right to left in the seventh and eighth. These transmissions are reversed if pin 1 is low.

The direction control for the second channel, EN3, is more complex. EN3 is the output of an OR ( $\geq 1$ ) function. One of the inputs to this OR is the active-high signal on pin 13. This signal is shown to be affected at the input to the OR gate by affecting input number 5, which is G5, meaning that pin 13 is ANDed with pin 1 before entering the OR gate. See 4.2 for an explanation of G (AND) dependency. The other input to the OR is the active-low signal on pin 13. This signal is ANDed with the complement of pin 11 before entering the OR gate. This is indicated by the G4 at pin 1 and the 4 with a bar over it at pin 13. Thus for EN3 to stand at the 1 state, which would enable transmission from pin 14 to pin 7, both pins 13 and 1 must be high or both pins 13 and 11 must be low.

## 6.7 SN75500E AC Plasma Display Driver with CMOS-Compatible Inputs



The heart of this device and its symbol is an 8-bit shift register. It has a single D input, pin 2, which is shown to be affected by affecting input number 9, which is C9, meaning it will be enabled if C9 = 1. See 4.8 for an explanation of C dependency and 5.0 for a discussion of bistable elements. Since the C input is dynamic, the storage elements are edge-triggered flip-flops. While C9 = 1, which in this case will occur on the transition of pin 3 from low to high, the state of the D input will be stored. Pin 2 is shown to be active low so to store a 1, pin 2 must be low.

In addition to controlling the D input, pin 3 is shown by  $\rightarrow$  to have an additional function. As pin 3 goes from low to high, data stored in the shift register is shifted one position. The right-pointing arrow means that the data is shifted away from the control block (down).

On the right side of the symbol an abbreviation technique has been used that is practical only when

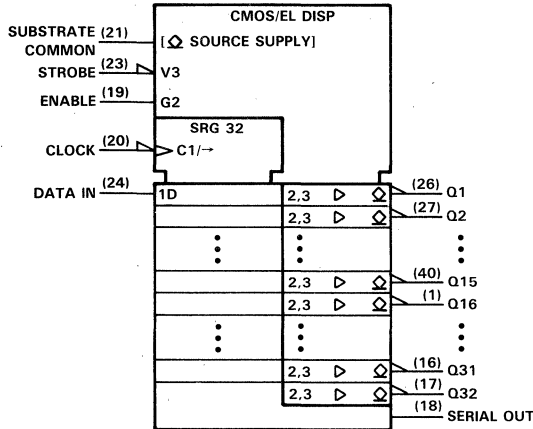
the internal labels and the pin numbers are both consecutive. Thus it should be clear that the input of the element whose output is pin 5 is affected by affecting input number 2, just as the input of the element whose output is pin 4 is affected by affecting input number 1. Affecting inputs 1 through 8 are Z inputs (Z1 through Z8), which means their signals are transferred directly to the output elements. See 4.6 for an explanation of Z dependency.

The inputs of the 32 implicitly shown output elements are also shown to be affected by affecting inputs numbers 11, 12, 13, and 14 in four blocks of eight each. These inputs will be found in the common control block preceded by a letter G and a brace. The brace is called the binary grouping symbol. It is equivalent to a decoder with outputs in this case driving four G inputs (G11, G12, G13, and G14). The weights of the inputs to the coder are shown to be  $2^0$  and  $2^1$  for pins 1 and 39, respectively. The decoder has four outputs corresponding to the four possible sums of the weights of the activated decoder inputs. If pins 1 and 39 are both low, the sum of the weights = 0 and G11 = 1. If pin 1 is low while pin 39 is high, the sum = 2 and G13 = 1 and so forth. G indicates AND dependency, see 4.2. Only one of the four affecting G inputs at a time can take on the 1 state. The block of eight output elements affected by that G input are enabled; the 0 state is imposed on the other 24 output elements and externally those output pins are low.

Because of their high-current, high-voltage characteristics, the outputs are labeled with the amplification symbol  $\triangleright$ . All the outputs share a common EN input, pin 38. See Figure 2 for an explanation of the common control block. When EN = 0 (pin 38 is high), the outputs take on their internal 0 states. Being active high, that means they are forced low.



## 6.8 SN75551 Electroluminescent Row Driver with CMOS-Compatible Inputs



The heart of this device and its symbol is a 32-bit shift register. It has a single D input, pin 24, which is shown to be affected by affecting input number 1, which is C1, meaning it will be enabled if  $C1 = 1$ . See 4.8 for an explanation of C dependency and 5.0 for a discussion of bistable elements. Since the C input is dynamic, the storage elements are edge-triggered flip-flops. While  $C1 = 1$ , which in this case will occur on the transition of pin 20 from high to low, the state of the D input will be stored. Pin 24 is shown to be active high so to store a 1, pin 24 must be high.

In addition to controlling the D input, pin 20 is shown by  $\rightarrow$  to have an additional function. As pin 20 goes from high to low, data stored in the shift register is shifted one position. The

right-pointing arrow means that the data is shifted away from the control block (down). The internal inputs of the output buffers are all shown to be affected by affecting inputs 2 and 3. Affecting input 2 is G2, meaning that pin 19 is ANDed with each of the internal register outputs, which are the buffer inputs. If pin 19 is high, the affected buffer inputs are enabled. If pin 19 is low, the 0 state is imposed on the affected buffer inputs. See 4.2 for an explanation of G (AND) dependency. Affecting input 3 is V3, meaning that pin 23 (active low) is ORed with each of the internal register outputs. If pin 23 is high,  $V3 = 0$  and the affected buffer inputs are enabled. If pin 23 is low,  $V3 = 1$  and the 1 state is imposed on the affected buffer inputs. See 4.4 for an explanation of V (OR) dependency. The effect of V3 is taken into account after that of G2 because of the order in which the labels appear. This means that the imposition of the 1 state on the internal buffer inputs by pin 23 would take precedence over the imposition of the 0 state by pin 19 in case both inputs were active. Pin 18 is shown to be an output directly from the thirty-second stage of the shift register. Pins 19 and 23 do not affect this output.

An abbreviation technique has been used for the shift register elements and associated the output lines. This technique is practical only when the pin numbers and pin names are both consecutive.

The symbol  $\square$  designates an n-p-n open-collector or similar output. In this device, the outputs are actually open-drain n-channel field-effect transistors. Instead of being grounded, the sources of these transistors are all connected to pin 21. This pin is used as an input to control the output voltage.

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# 5 Applications

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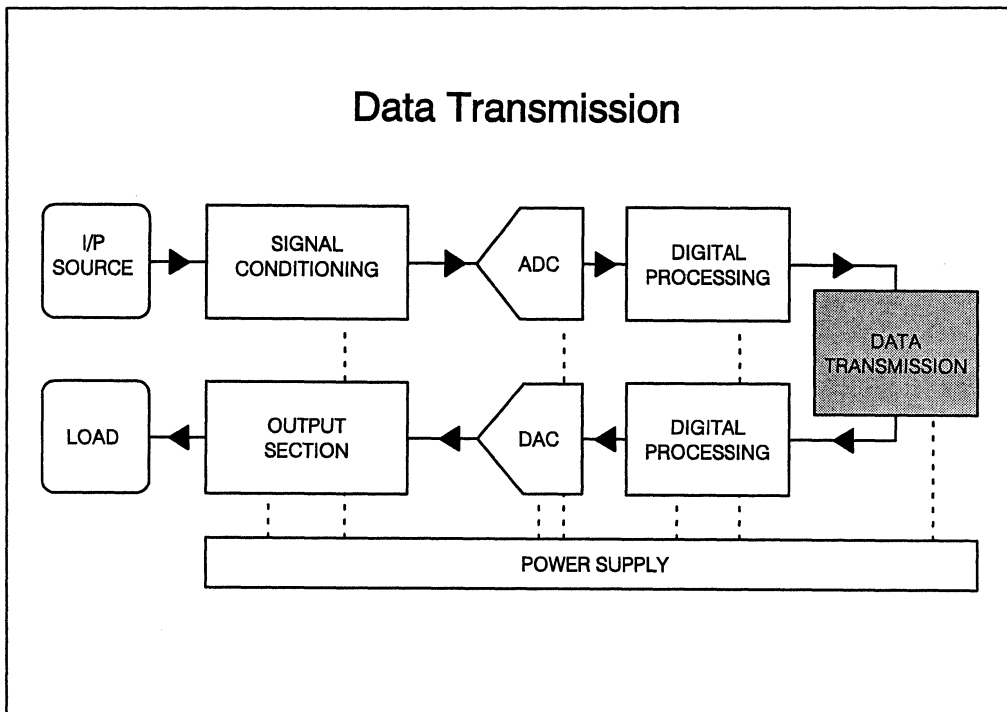
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# 1. Introduction

## 1.1. Data Transmission

It may seem strange that TI's 'Data transmission' products are included within the Linear seminar. Data Transmission as part of TI's Linear Products portfolio is concerned with the standards involving transmitting data at relatively high speeds down long line lengths, the considerations for which are



*Figure 5.1 - Data Transmission*



primarily of an analog more than a digital nature. Likewise the design of data transmission ICs requires experienced analog engineers to implement functions such as slew rate limiting, receiver filtering and common-mode protection.

In this year's seminar we will concentrate on two very popular transmission standards, RS-232 or as it is now known EIA/TIA-232-E, and the multi-point, half duplex RS-485 standard. The last section covers the physical layer of the increasingly popular Small Computer Systems Interface Standard (SCSI).

### 1.1.1. The Need for Transmission Standards

Data transmission standards evolved for two main reasons: From the need to transmit data reliably over long distances, and to provide a standard interface to facilitate communication between equipment from different suppliers. Although TTL/Logic signal levels and products can be used, they generally lack the power handling capabilities, robustness and noise margins required for reliable transmission. Indeed for backplane equipment, TTL is no longer specified for the newer high speed standards, such as Futurebus+ which uses BTL transceivers. In general the standards concerned with transmitting data over long distances incorporate wider voltage swings, increased robustness and higher power outputs than can be delivered using conventional 'Logic' products. Similarly the sub-micron technologies used in the fabrication of today's logic devices cannot provide the power handling and robustness necessary for successful long distance transmission.

### 1.1.2. Specialist Technologies

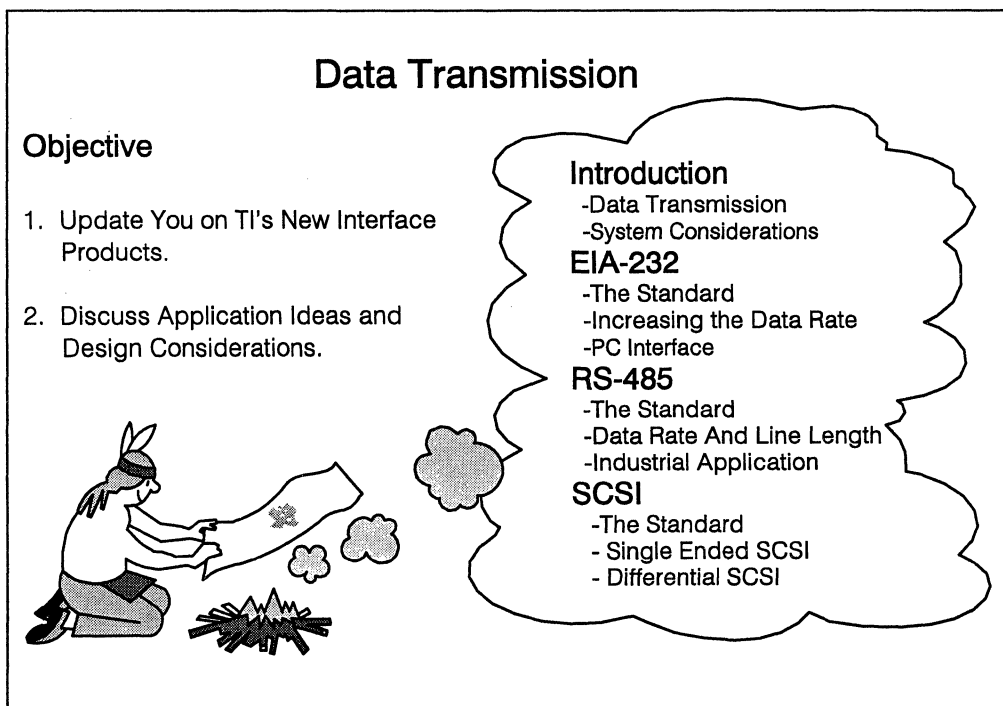
This leads to the need for specialist ICs, and technologies, to meet the exacting requirements of these transmission standards. The traditional technological answer has been to utilise the inherent robustness afforded by bipolar technologies, however the additional need for low power consumption and high levels of integration no longer makes this attractive. SC manufacturers are now having to develop their technologies to accommodate these requirements. TI has introduced its proprietary LinBiCMOS™ technology combining the robustness of bipolar together with the power consumption and integration afforded by CMOS. Other manufacturers are using pure CMOS and integrating shottky diodes to the same end. The result is very specialised and reliable products that are able to withstand the harsh environment unique to data transmission products.

Texas Instruments has been a leading supplier of data transmission products for many years, and is continually innovating new fields. Although the following sections are limited to the more common interface standards, TI is actively involved in many new emerging standards and markets, for example Futurebus+, a backplane standard with virtually no ceiling on data rate, the high speed serial data link evolving from the P1394 committee and multiplex wiring systems such as ABUS, CAN and VAN. The reader is advised to contact a TI representative for information on these product areas.

With the considerable expertise in design, product definition and range of technologies Texas Instruments is the ideal choice for supplying your data transmission product requirements.

### 1.1.3. About This Section

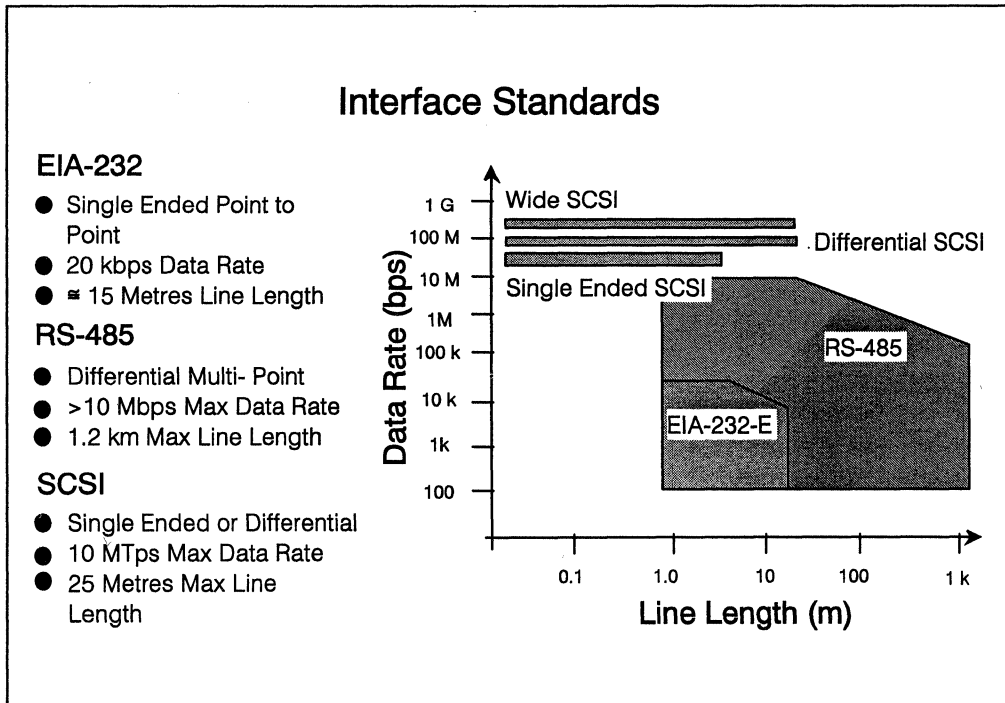
This Section is split into four distinct sections each of which provides a practical rather than theoretical approach to in an attempt to give the reader an insight into three popular data transmission standards, EIA/TIA-232, RS-485 and the SCSI standard. The Section is split as follows:



*Figure 5.2 - Data Transmission Agenda*

1. **Introduction:** An overview of the various factors that affect any data transmission system. Under discussion is the line length versus data rate trade-off, noise sources, correct line termination and network topology in addition to explaining the use of eye patterns as a tool to measure transmission quality.
2. **EIA-232:** A discussion of the standard with particular attention paid to the changes made in the 'E' revision. Also covered is the use of '232' at higher data rates, up to 116 kbps (kilo bits per second) and an application focus on the popular DB9 PC interface. Particular attention is paid to TI's new products throughout the section. The generic '232' standard will be referred to in this book as EIA-232, where a parameter is unique to a specific revision the EIA-232 reference will be used.
3. **RS-485:** An overview of the RS-485 specification followed by a design example. We use an industrial control application to understand the factors that need to be taken into account when

designing an RS-485 system. Highlighted throughout the section will be TI's new products compliant with this standard.



*Figure 5.3 - Interface Standards*

4. SCSI: We will consider the physical layer of this standard that concern both single ended and differential transmission. For single ended transmission we will look specifically at optimising the line termination to achieve maximum transmission rate over the 6 metre distance as specified in the standard. The differential SCSI system increases the line length to 25 metres and uses the RS-485 standard to achieve this. We will look at TI's new nine channel RS-485 transceiver which minimises the problems caused by the 18 line wide bus as defined by the standard.

## 1.2. Overview of the Interface Standards

Referring to Figure 5.3 we can see the relationship of each transmission standard when comparing data rate and line length.

### 1.2.1. EIA/TIA-232

EIA-232 or 'Recommended Standard' 232 is defined in the ANSI (American National Standard Institution) specification as "The Interface Between Data Terminal Equipment and Data Circuit-

Terminating Equipment Employing Serial Binary Data Interchange". The standard employs a single ended serial transmission scheme and outlines the set of rules for exchanging data between computer equipment, originally this being a Computer Terminal (DTE) and a modem (DCE). The standard has evolved over the years with the latest 'E' revision released in July 1991. The standard is now known as EIA/TIA-232-E, with EIA standing for the Electronic Industries Association and TIA for the Telecommunications Industry Association.

As with previous revisions of the standard the maximum data rate is defined as 20 k bits per second (kbps) although there are now a number of software applications that now push this data rate up to 116 kbps, well outside the standard. The 'C' revision defined the maximum line length as 15 metres however this failed to comprehend the type of cable used and consequently the load capacitance on the line driver. Both the 'D' and 'E' revisions addressed this by more correctly defining the line length in terms of load capacitance. The maximum load capacitance is specified as 2500 pF that translates using standard cables to between 15 and 20 metres. Line length and data rate are limited as the standard employs single ended communication which is prone to external factors. For longer line lengths and higher data rates a differential balanced line communication link is essential.

### 1.2.2. RS-485

RS-485 was primarily an upgrade to the EIA RS-422-A standard utilising the same signal levels but facilitating half duplex multi-point communication. The standard is less complex than the EIA-232 standard as it only specifies the physical layer of the transmission scheme. Hardware such as the connector is left to the user to define. The standard specifies a balanced transmission line whose maximum line length is undefined but is nominally 1.2 km for 24 AWG cable based on 6 dB signal attenuation. The maximum data rate is also undefined but is specified by the relationship of signal rise time to bit time which is influenced both by the line driver and the line length and the line loading. In the majority of applications it is the line length that is the limiting factor on data rate due to signal dispersion. This is discussed in later sections.

### 1.2.3. Small Computer Systems Interface (SCSI)

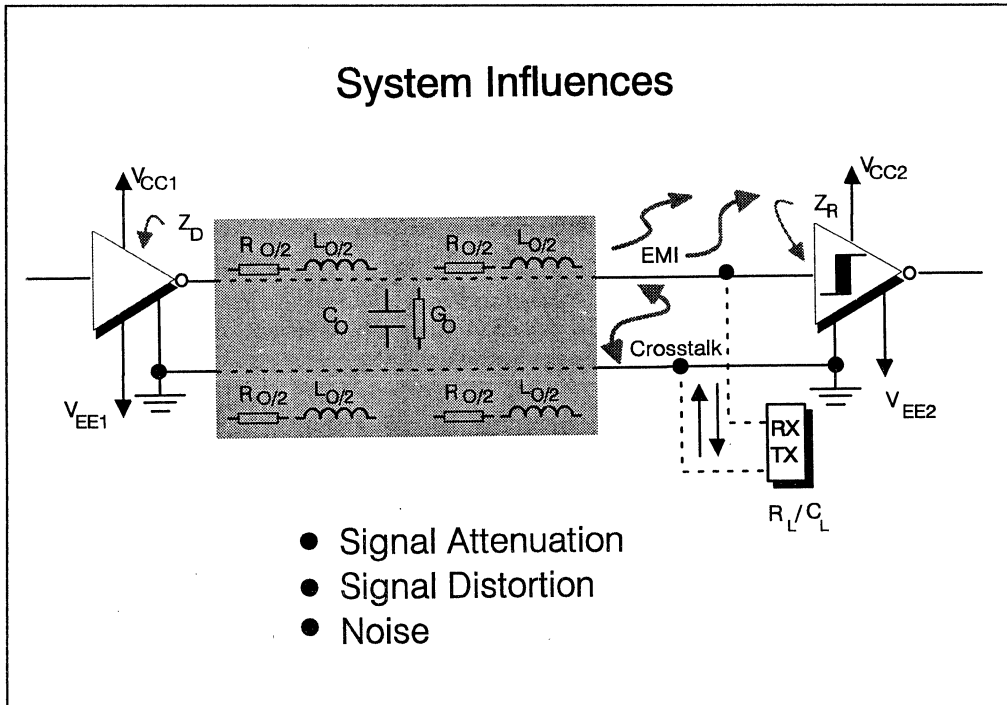
SCSI is an industry-standard interface, defined by the ANSI, for the interchange of data between computer and computer peripherals. Standard SCSI is a byte wide parallel interface for high speed data transfer over relatively short distances. The SCSI bus is bi-directional and is terminated at both ends of the cable to reduce reflections. For the single ended interface the standard specifies a maximum line length of 6 metres. The maximum data rate is not specified but at present 5 Million Transfers per second (MTps) is achievable using active termination. This can be increased up to 10 MTps using innovative termination as we will discuss later. For longer line length applications, up to 25 metres, the SCSI standard defines the interface using the RS-485 standard as the physical layer. This pushes the data rate to 10 MTps over the full 25 metres which equates to 80 Mbps. A further development of SCSI is 'Wide' SCSI which increases the data bus to 16 bits wide. Using the 10 MTps differential interface this increases the bit rate to 160 Mbps.

### 1.2.4. Summary of EIA Interface Standards

Parameter		EIA-232	RS-422-A	RS-422-A	RS-485
Mode of Operation		Single-Ended	Single-Ended	Differential	Differential
Number of Drivers and Receivers		1 Driver 1 Receiver	1 Driver 10 Receivers	1 Driver 10 Receivers	32 Drivers 32 Receivers
Maximum Cable Length (m)		15	1200	1200	1200
Maximum Data Rate (bps)		20 k	100 k	10 M	10 M
Maximum Common-Mode Voltage (V)		$\pm 25$	$\pm 6$	6 to $-0.25$	12 to $-7$
Driver Output Levels (V)	Unloaded	$\pm 5$	$\pm 3.6$	$\pm 2$	$\pm 1.5$
	Loaded	$\pm 15$	$\pm 6$	$\pm 5$	$\pm 5$
Driver Load ( $\Omega$ )		3 k to 7 k	450 (Min)	100 (Min)	60 (Min)
Driver Slew Rate		30 V/ $\mu$ s (Max.)	External Control	NA	NA
Driver Output Short Circuit Current Limit (mA)		500 to $V_{CC}$	150 to GND	150 to GND	150 to GND 250 to $-7$ or 12 V
Driver Output Resistance -	Power on	NA	NA	NA	12 k
High Z state ( $\Omega$ )	Power off	300	60 k	60 k	12 k
Receiver Input Resistance ( $\Omega$ )		3 to 7	4	4	12
Receiver Sensitivity		$\pm 3$ V	$\pm 200$ mV	$\pm 200$ mV	$\pm 200$ mV

### 1.3. System Influences

Noise, distortion and attenuation are always present in data transmission systems and strictly limit performance. We will consider each one of these in turn although there is some overlap i.e. noise can cause distortion.



*Figure 5.4 - System Influences*

#### 1.3.1. Signal Attenuation

Any data transmission over wire experiences losses and distortion due to distributed constants present along the cable: distributed series inductance, distributed shunt capacitance, distributed series resistance and distributed shunt conductance. Attenuation of the signal in a cable is affected by each of these components. The series resistance,  $R$ , is frequency dependent and is a result of the DC resistance of the cable and the skin effect. Skin effect is a term which refers to the tendency of electrons to travel to the surface of a conductor at higher frequencies, thereby reducing the overall cross sectional area and increasing the resistance. The series inductance,  $L$ , represents the opposition to change in current levels caused by the collapsing and expanding magnetic fields created due to fluctuating current levels. The shunt capacitance,  $C$ , is created by the two conductors in close proximity and separated by a dielectric. As the signal frequency increases the capacitive reactance

decreases, consequently reducing the opposition to current flow. The final component, shunt transconductance or  $G$ , is a function of the dielectric loss of the insulation around each conductor which allows some leakage current to pass between conductors. In modern dielectrics this is often assumed to be negligible.

The overall effect of these distributed constants is called the characteristic impedance of the line,  $Z_o$ , and is expressed as:

$$Z_o = \sqrt{\frac{R + j2\pi fL}{G + j2\pi fC}}$$

Where:

$L$  is in henries/unit length

$R$  is in ohms/unit length

$C$  is in farads/unit length

$G$  is in siemens/unit length

The current/voltage relationship of an incident wave travelling down a transmission line in the direction of the load will be determined by this equation. Equally a reflected wave travelling from the direction of the load will also be dependent on this relationship. We will revisit this equation when we discuss transmission line termination in section 1.5. The signal velocity along the transmission line and the attenuation depends upon the propagation constant  $\gamma$  of the line. The propagation constant, when separated into its real and imaginary parts, is symbolised by  $\alpha + j\beta$  where  $\alpha$  is known as the attenuation constant and  $\beta$  as the phase constant.  $\alpha$  determines the rate of attenuation and has units of nepers per unit length, and  $\beta$  determines the phase velocity, where:

Phase velocity,

$$V_p = \frac{\omega}{\beta}$$

Where  $\omega$  is the angular velocity.

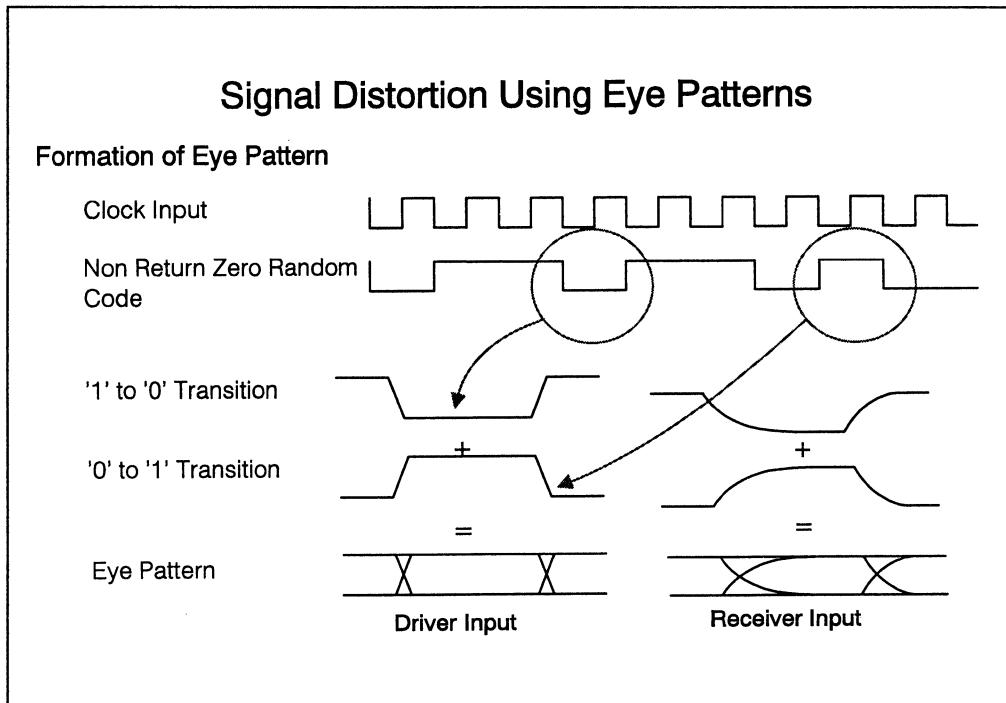
Additionally, the propagation constant,

$$\gamma = \alpha + j\beta = \sqrt{(R + j\omega L)(G + j\omega C)}$$

In practice the attenuation of a particular cable can be determined from manufacturers data where usually a curve of bit rate or frequency is plotted against dB, usually quoted per 100 ft or 30 metres. The attenuation constant,  $\beta$ , can be converted to dBs by multiplying by 8.686.

The maximum attenuation allowable will be dependent on the system configuration but a figure of 6 dBV maximum is a good guide. Actual curves are discussed later in the RS-485 section.

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*Figure 5.5 - Signal Distortion Using Eye Patterns*

### 1.3.2. Signal Distortion

One of the primary causes of signal distortion is the effect known as frequency dispersion. As discussed in 1.3.1, phase velocity and attenuation are both frequency dependent and whose effect is to distort and delay the signal pulse. The high frequency components contained in the leading and lagging edges of a pulse experience minimum delay but experience maximum attenuation. The pulse top and low frequency components are subjected to increased delays. The result is that various parts of the pulse arrive at the receiving end at different times and at differing levels causing distortion of the original signal. It follows the longer the line length the more the bit rate must be reduced. In many transmission systems it is this factor alone which determines the maximum signalling rate.

Once again cable manufacturers sometimes specify a bit rate versus line length curve but a better way to check signal distortion of your system is by the use of eye patterns or eye diagrams. Indeed cable manufacturers generate their bit rate/distance curves using eye pattern measurements. Eye patterns allow you to visibly see and measure signal distortion as a function of data rate. See later sections on how to implement Eye Patterns.



### 1.3.3. Noise

Noise is generated from a variety of sources and can strongly influence how you implement your data transmission system. All extraneous signals appearing at the receiving end of the transmission circuit that are not due to the input signal are considered as noise. The two most likely sources of noise that will affecting data transmission systems in the context of this Section are common-mode voltages and cross talk. We will discuss both these types of noise and how they relate to the type of transmission system in section 1.6.

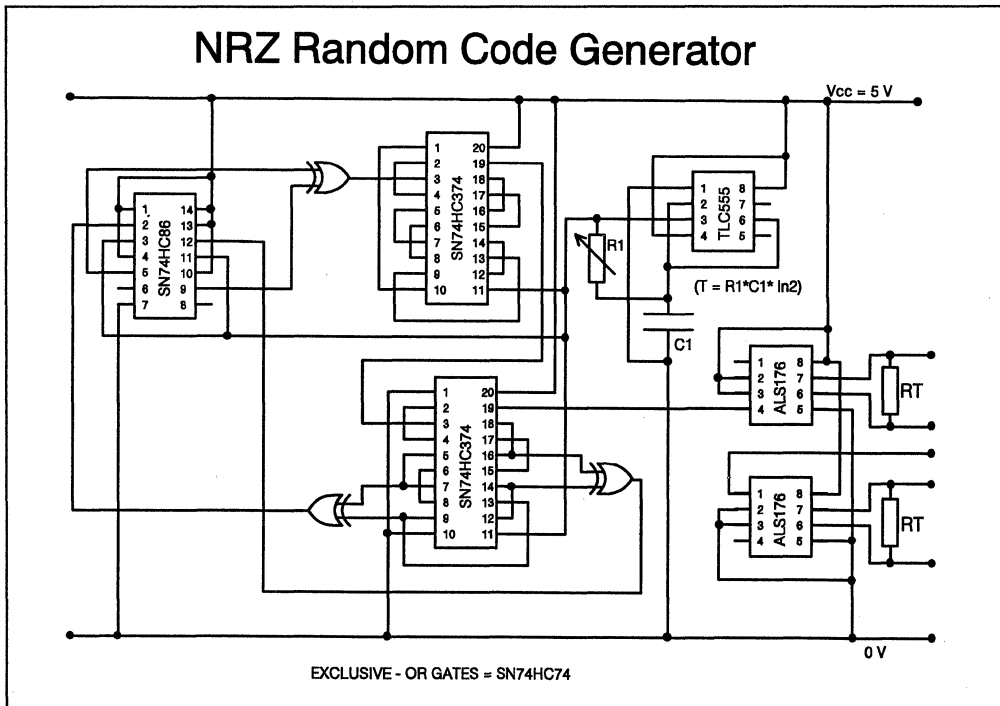


Figure 5.5.1 - NRZ Random Code Generator

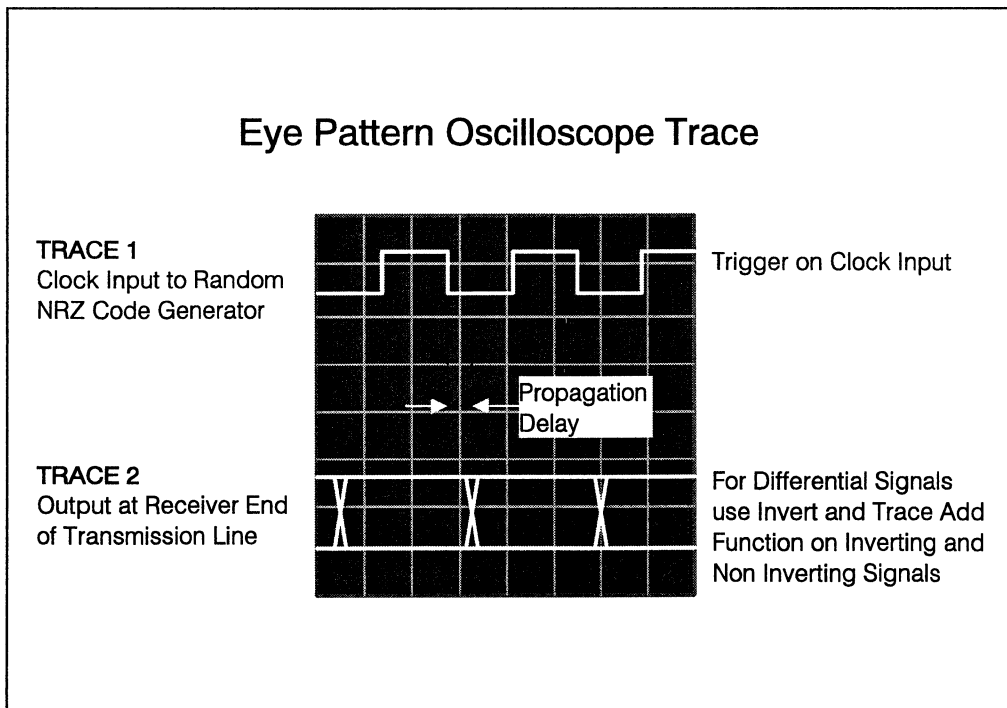
### 1.4. Eye patterns

To determine the effects of signal distortion, noise etc. on Intersymbol interference (ISI) in a data transmission system the eye pattern is used. ISI is the effect of neighbouring pulses in a pulse train spilling over into adjacent pulses and forces a reduction in the allowable permitted pulse rate for a given line length in order to maintain adequate distinction between adjacent pulses. The eye pattern is displayed on an oscilloscope with the term 'Eye' coming from the appearance of the trace on the CRT.

### 1.4.1. Setting up the eye pattern

The eye pattern is obtained by applying a random non return zero (NRZ) code down the transmission line under test. This represents all possible pulse combinations. The signal at the receiving end of the line is connected to the vertical amplifier of an oscilloscope, with the 'scope triggered using the synchronisation clock to the NRZ code generator on a separate trace. See Figure 5.5 Over any one unit interval the random code generator should produce a combination of signals. The resulting signals can then be viewed on the oscilloscope over one unit interval, each unit interval should resemble an eye, similar to Figure 5.6. For differential transmission both signals at the end of the transmission line should be applied to separate amplifiers on the oscilloscope and then summed using the summation facility on the oscilloscope.

Figure 5.5.1 shows a circuit to generate the NRZ code. In this case we have used it to test the RS-485 SN75176 type transceiver.

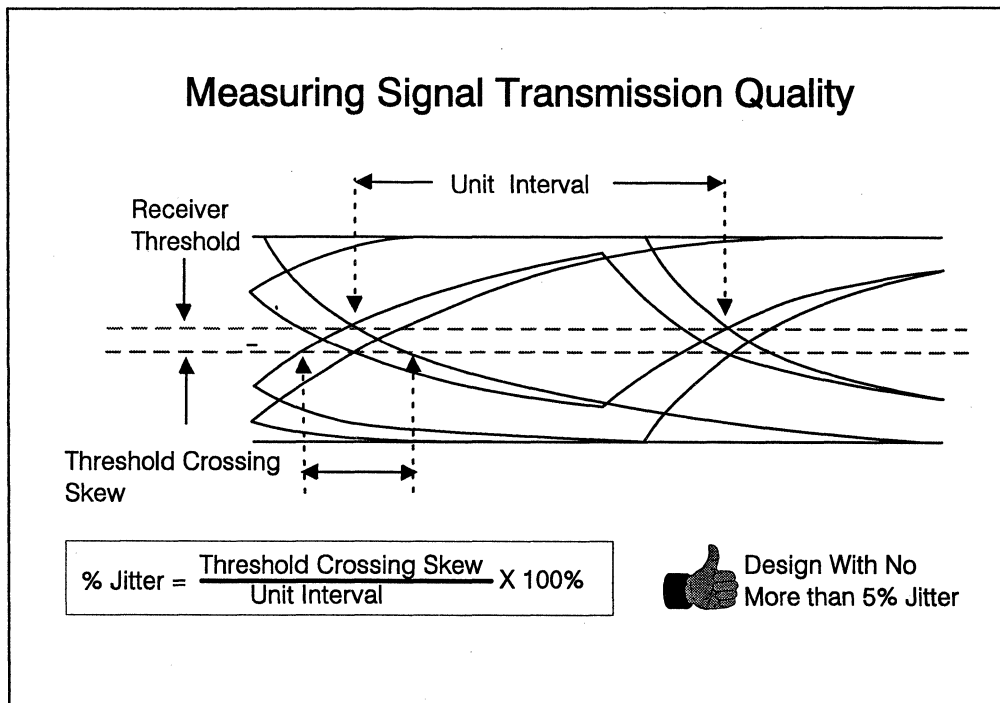


*Figure 5.6 - Eye Pattern Oscilloscope Trace*

### 1.4.2. Taking Measurements from Eye Patterns

Before considering actual measurements the first key indicator on the performance of the transmission system can be seen by simply looking at the eye pattern. The 'openness' of the eye is an

indication of the 'quality' of the transmitted signal and is an indication of the noise and distortion tolerance of the system.



**Figure 5.7 - Measuring Signal Transmission Quality**

For actual measurements the decision points of the transceiver should be superimposed upon the eye pattern. The vertical distance between the decision points and the signal trace is an approximate indication of the noise margin of the system. The horizontal appearance of the eye can be used to determine the maximum jitter tolerance of the system. A good guide, and one that is used by cable manufacturers to determine data rate versus line length curves, is to design with no more than 5% jitter. Where % jitter is defined as the ratio of Threshold crossing Skew to unit interval as shown in Figure 5.7. Jitter is caused by a number of factors including , signal frequency, noise and cross talk. (Noise frequency can modulate the transmitted signal, for example 50 Hz hum or from other low frequency sources). It should also be noted at this point the effect of threshold misalignment which can cause severe problems with the received signal, reducing the detected pulse width considerably.

## 1.5. Line Termination

The behaviour of the transmitted signal and the integrity of the data at the receiving end depends upon the data rate and line length of the cable. There are two behavioural models; of a transmission cable:

- i. **Lumped parameter model (Short wire) .**
- ii. **Distributed parameter model (Transmission line)**

As discussed in section 1.3.1 the distributed parameter model models the connecting circuit in terms of distributed parameters (inductance, capacitance, resistance, conductance), rather than as an equivalent lumped load on the line. The transmission line can be considered in terms of an infinite number of small filter sections and as a result the transmission line is said to have a characteristic impedance,  $Z_0$ .  $Z_0$  is independent of distance along the line and represents the voltage and current relationship for an incident wave at any point as it travels along the line.

### 1.5.1. Transmission Line Test

#### Classifying as a Lumped or Distributed Parameter Model

All cables can be thought of as transmission lines; but the term, transmission line, is used with differing meanings.

Consider a signal propagating down a simple data link comprising two wires. When the signal starts to change at the transmitter output the effect of this change will eventually be seen at the other end of the line. A reflection of the signal will occur, which will eventually return back to the transmitter terminals.

If this happens before the original transmitted signal has risen to its peak value then the line will normally be treated as a lumped parameter system rather than as a true transmission line. This is because the line itself does not greatly influence the performance of the system.

A general rule of thumb for determining if a system should be treated as a true transmission line can be formulated; If the rise time,  $t_r$ , of the signal is much less than the round trip propagation delay,  $2t_{pd}$ , of the signal from transmitter to receiver and back to transmitter, then the cable can be treated as a transmission line and not as a lumped parameter model. A better model is given in Figure 5.8 where a safety margin is built in to the propagation delay/rise time relationship.

### 1.5.2. Transmission Line Considerations & Effects

When the cable is operating like a transmission line, extra loads in the form of transmitters and receivers can be added, providing that they do not cause too great a shunting effect on the line. These extra loads, if they are evenly distributed along the line, can be treated as an extra distributed capacitance along the line adding to the effect of the line capacitance and inductance. This extra load decreases the line impedance and reduces the speed of the signal along the line.

In the case of the lumped parameter model the line represents a pure fixed load to the transmitter device. For example, the capacitance of the line will be modelled as a fixed value which effectively

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limits the output voltage slew rate of the transmitter (assuming it can supply a finite amount of current to the line).

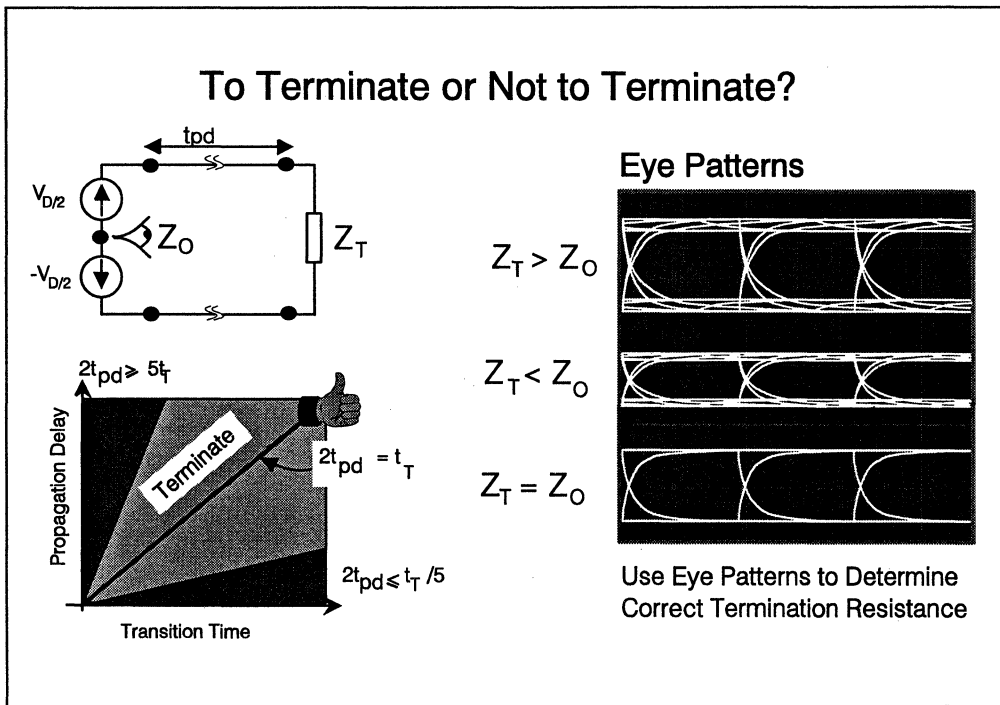


Figure 5.8 - To Terminate or Not to Terminate?

### 1.5.3. Transmission Line Reflections

Consider a driver circuit driving the line. When the driver output voltage changes state, the driver appears to see the effective characteristic impedance of the line,  $Z_O$ . This will cause the voltage at the output of the driver circuit to reduce as a result of the potential divider action formed by  $Z_O$  and the driver circuit output impedance,  $Z_D$ .

At any point along the line the ideal source impedance will appear as  $Z_O$  and the ideal load impedance will also appear as  $Z_O$ . This gives the impression that the line is being driven by a voltage source of twice the magnitude of the line voltage.

When the signal reaches the receiving end of the line it sees a terminating impedance equal to the impedance ( $Z_O$ ) of the line that it is already travelling on. It interprets this as a continuation of the line. The voltage on the line will not alter and the current flowing along the line will flow through the termination resistor and back to the driver via either ground or the other line in the system. Operation of the circuit as just described would result in optimum data transmission efficiency, with little or no signal reflections. However, circuit operation in the real world is not always so perfect.

If the termination impedance is dis-similar to the characteristic impedance of the line itself, the voltage at the termination point will alter. The voltage at the termination point is dependent on the relative size of the termination impedance to the line impedance. If the termination impedance is higher than the line impedance, the line voltage will increase causing a positive voltage reflection of the signal. When the termination impedance is lower than the line impedance, the line voltage will decrease leading to a negative reflection. The same effect will occur at the driver output terminals due to impedance mismatches between driver and line.

Reflections at each end of the line will eventually settle and leave a constant dc voltage on the line. The value of this voltage is equal to the ideal open circuit output voltage multiplied by the termination impedance divided by the sum of the driver output impedance and termination impedance.

Reflections as described can cause problems when driving lines at high frequencies. False receiver triggering can occur and repeated signal reflections will cause signal wave distortion.

#### **1.5.4. Using Eye Patterns to Determine $Z_0$**

Referring back to eye patterns, these can also be used to find the characteristic impedance of a transmission line. Figure 5.8 shows three sets of eye patterns,  $Z_T > Z_0$ ,  $Z_T < Z_0$ , and  $Z_T = Z_0$ , where  $Z_T$  is 200  $\Omega$ , 50  $\Omega$ , and 100  $\Omega$ , respectively. Where  $Z_T > Z_0$  the signal is larger with multiple traces, while with  $Z_T < Z_0$  the signal is similar but much reduced in amplitude and could cause signal to noise ratio problems at the receiver. With  $Z_T = Z_0$ , the signal is very clear with a near perfect eye pattern. In practice it is possible to use a variable resistance and the eye pattern to determine the correct termination impedance for zero reflections.

### **1.6.Noise Influences**

There are two main classification of transmission scheme, single ended or differential. Each are affected by noise influences in differing ways - the next two section describe each transmission scheme paying particular attention to the affects of noise. Figure 5.9 details both types of transmission scheme.

#### **1.6.1. Single Ended Line Considerations**

Single ended data transmission systems consist of a signal line on which data is sent down, and a ground line through which the current returns. A direct result of this is that the ground line forms part of the transmission line, which can be of benefit in some circumstances but not in others.

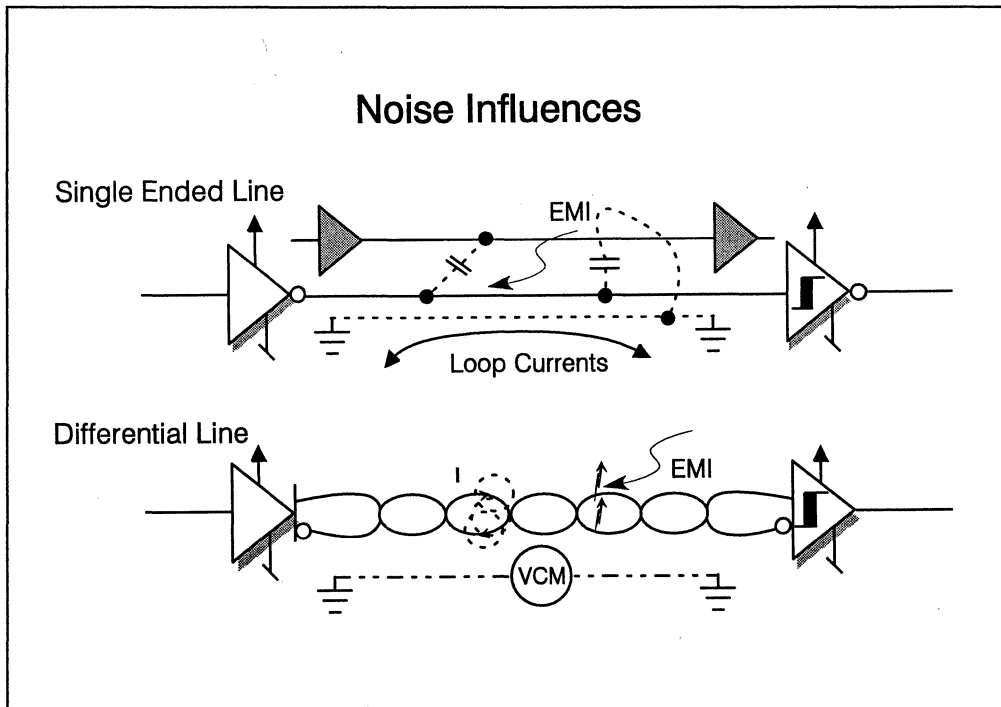
One of the major benefits, and most obvious, is that a single ended system is the lowest cost solution in terms of cabling costs. In general terms it requires only half the cable of a differential system. It is also relatively simple to install and operate.

The main disadvantage of the single ended solution is its poor noise immunity. Because the ground wire forms part of the system, any transient voltage or shifts in voltage potential may be induced (from nearby high frequency logic or high current power circuits), leading to signal degradation ultimately leading to false receiver triggering. For example, a shift in the ground potential at the

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receiver end of the system can lead to an apparent change in the input switching threshold of the receiver device, thus increasing susceptibility to noise.

Cross talk is also a major concern especially at high frequencies. Cross talk is generated from both capacitive and inductive coupling. Capacitive coupling tends to be more severe at higher signal frequencies as capacitive reactance decreases. The impedance and termination of the coupled line



*Figure 5.9 - Noise Influences*

determines whether the electric or the magnetic coupling is dominant. If the impedance of the line is high the capacitive pickup is large. Alternatively, if the line impedance is low, the series impedance as seen by the induced voltage is low, allowing large induced currents to flow.

These problems will normally limit the distance and speed of reliable operation for a single ended link.

Cross talk can be reduced by;

- i. **Limiting the slew-rate of signals so that they do not cause cross talk to be induced onto other lines**
- ii. **Limiting the line length.**

**iii. Shielding the signal conductor.**

While the common-mode noise could be reduced by:-

- i. Isolating the signal ground from power conductors (e.g. keep signal grounds separated as far as possible from power grounds).**
- ii. Ground wires should be as low as impedance as possible.**
- iii. Using star ground system configurations.**

Some of these techniques are used in systems such as EIA-232 e.g. Maximum slew rate of EIA-232 is defined as 30 V/ $\mu$ s while Futurebus+, an emerging high speed backplane standard, uses trapezoidal waveforms to limit cross talk

## **1.6.2. Differential Line Considerations**

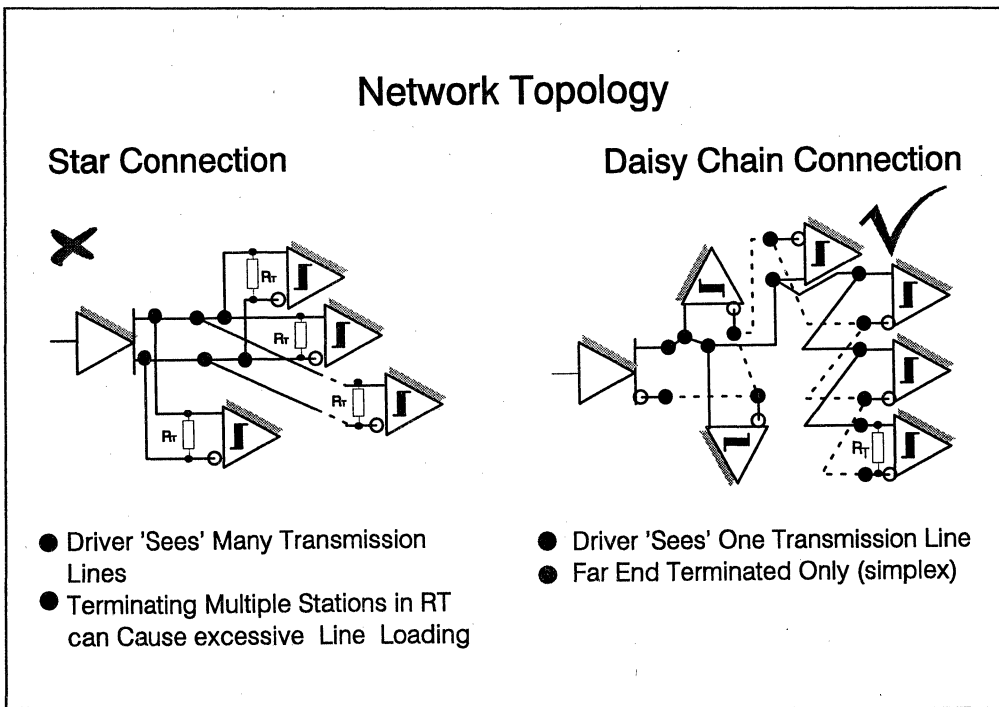
A differential communication system involves the use of two signal carrying wires between transmitter and receiver, such that the signal current flows in opposite directions in each wire. The net effect of this is the receiver is only concerned with the *difference* in voltage between the two wires. The absolute value of the dc common mode voltage of the two wires is not important. In practice, transmitters and receivers have a finite common mode voltage range in which they can operate.

The use of a differential communications interface allows transmission at higher data rates over longer distances to be accomplished. This is because the effects of external noise sources and cross talk are much less pronounced on the data signal. Any external noise source coupling onto the differential lines will appear as an extra common mode voltage which the receiver is insensitive to. The difference between the signal levels on the two lines will therefore remain the same. By the same argument, a change in the local ground potential at one end of the line will appear as just another change in the common mode voltage level of the signals. The differential output to the line will also provide a doubling of the driver's single-ended output signal. Twisted pair cable is commonly used for differential communications since its twisted nature tends to cause cancellation of the magnetic fields generated by the current flowing through each wire, thus reducing the effective inductance of the pair.

The main disadvantage of a differential system lies in the fact that two cables are required for each communication link. This increases system cost but provides superior performance when data is transmitted at high rates over long distances.

The RS-485 and RS-422-A standards both use differential type transmission.





*Figure 5.10 - Network Topology*

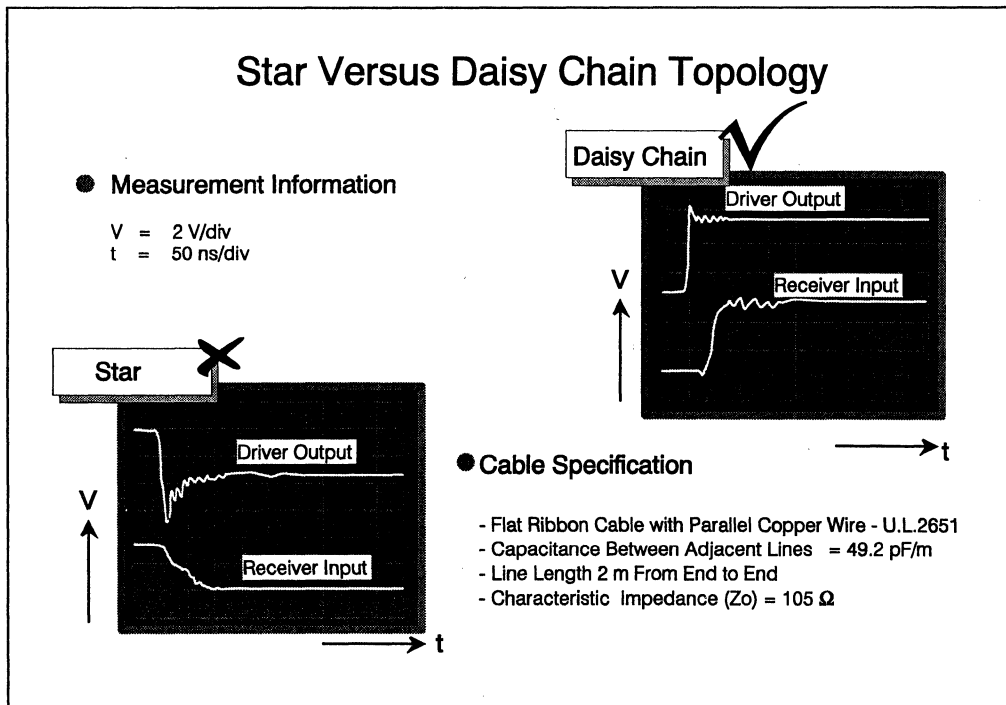
## 1.7. Network Topology

In addition to considering signal attenuation, the effects of noise, signal distortion and correct line termination, we must also consider the way in which stations are connected to the line. Furthermore the position of the line termination resistor and device positioning must be considered. There are two basic methods of connection, see Figure 5.10;

- i. The star connection
- ii. The daisy chain connection

Considering the star connection, the transition edge from the driver will be loaded by a group of separate transmission lines, rather than one. Each transmission line boundary will cause a change in impedance resulting in reflections.

Another situation to avoid is the termination of multiple stations, since this could excessively load the driver. Termination at the extreme ends for RS-485 (half duplex) and far end only for RS-422 is recommended and is accounted for in each standard. Normally stubs (taps of the main line ) should be kept as short as possible so not to appear as transmission lines themselves.



**Figure 5.11 - Star Versus Daisy Chain Topology**

The recommended method is to use the daisy chain, a configuration where the transmission line continues from one receiver to the next and only the last receiver on the chain is terminated. This means that the transmission line and hence the driver will see one continuous transmission line with only one termination resistor. Each tap-off will in effect be a stub, but in this case they will not be all grouped together and will be kept very short to reduce their effect.

The Figure 5.11 shown further confirms the need to keep stub lengths short and the use of correct termination techniques by comparing the effect on signal quality for the daisy chain and star method of connection.

In both instances exactly the same application scenario was used as was the same cable specification. The cable used was a flat ribbon cable with parallel copper wire conforming to U.L. specification 2651. Connections were made as shown in the previous figure and the total cable length from source to destination was 2 m.

#### 1.7.1. How Short is Short ?

It has been described earlier that a pair of cables will act as a transmission line if the round trip propagation delay,  $t_{pd}$ , is more than 5 times the transition times of the driver,  $t_T$ . The converse is true

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if the line is not to operate as a transmission line but as a lumped parameter model. This forms the basis of the stub length calculation given below.

The rule of thumb states that the transition time of the pulse sent down the line should take ten times the time taken for the pulse to propagate to the end of the stub. As a result, any reflections will be incorporated into the transition edge.

From this basis, the length of a stub can be calculated using the cable and driver parameters.

The pulse speed down the line,  $U$ , equals the reciprocal of the product of the line impedance and line capacitance, both of which are normally specified for the cables used. The propagation delay down the stub should be at the most one tenth of the transition time of the pulse. These facts can be brought together to give the length of the stub,  $L_s$ , as;

$$L_s = \frac{t_{\text{tr}}}{10}$$

Using the SN75ALS180 and its transition time of 13 ns, a cable with a characteristic impedance of  $78\Omega$  and line capacitance of  $65\text{pF}$ :

Using:  $Z_0 = \sqrt{\frac{L_0}{C_0}}$ , as an approximation of the equation shown in section 1.3.1. (In practical situations  $j\omega L \gg R$  and  $j\omega C \gg G$ , therefore  $R$  and  $G$  can be assumed to be negligible although the  $R$  component must be considered for long line lengths.)

And:  $V_p = \frac{1}{\sqrt{L_0 \times C_0}}$  as an approximation of the phase velocity equation in 1.3.1,

Substitution gives:  $V_p = \frac{1}{Z_0 \times C_0}$

Using the values given earlier:  $V_p = \frac{1}{78 \times 65 \times 10^{-12}} = 198 \times 10^6 \text{ ms}^{-1}$

Now, using our rule of thumb described earlier:  $t_{\text{pd}} = \frac{t_{\text{tr}}}{10}$  and  $L_s = t_{\text{pd}} \times V_p$

Gives  $t_{\text{pd}} = \frac{13 \times 10^{-9}}{10}$  and therefore  $L_s = 1.3 \times 10^{-9} \times 198 \times 10^6 = 257 \text{ mm} = 10 \text{ inches}$

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This means the length of each stub should be no more than 257 mm. Under this length the stub can be considered as a lumped load and will not cause any unwanted reflections. The main effect of each stub in this case will be a slight increase in the capacitance loading of the line.



## 2. Interface Circuits for EIA-232

### 2.1. General Information

This section on EIA/TIA-232, or RS-232 as it has been known in the past, will discuss the electrical aspects of the standard, i.e. the physical layer. Initially we will discuss the latest developments of the 'E' revision upgrade and then cover TI's latest products conforming to this standard. However the reader should note the products under discussion in this section are application specific to the 9-pin DB9 Personal Computer DTE serial interface which is effectively a sub-set of the full EIA-232 standard. As a semiconductor manufacturer we find the majority of EIA-232 applications are moving to this interface. Due to the nature of the signals i.e. 5 receive and 3 transmit lines the older established EIA-232 products no longer provide an optimum solution. This interface is now driving the need for single chip EIA-232 solutions. Additional features such as single supply operation, increased ESD protection, power down modes have moved from the desirable features to the essential features of today's interface. In the later half of this section we will discuss the DB9 interface and TI's products designed specifically for this application.

Looking at the DB9 interface one step back into the digital system, there is in most cases a UART or ACE (asynchronous communication element). The ACE provides the parallel to serial conversion and the necessary start/stop bits, parity bit generation and checking for error free data transmission. TI manufactures a number of ACEs, the most advanced being the TL16C552. This integrates two serial ports with FIFO buffers together with a PC parallel port. Although not specifically covered in this section a selection guide on ACEs is included towards the rear of this section.

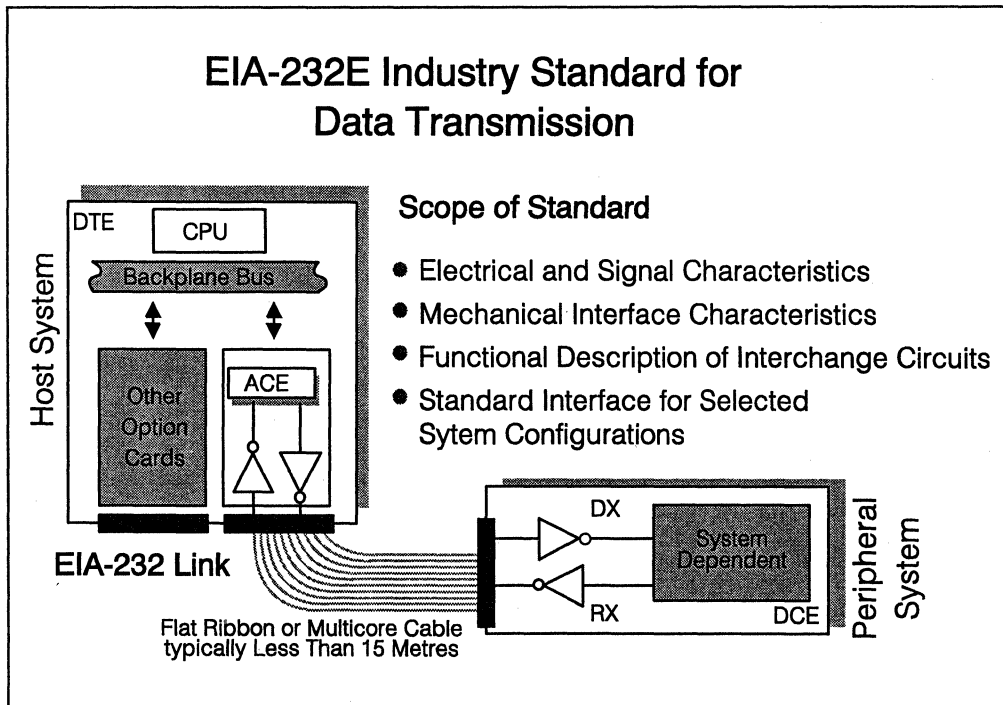
#### 2.1.1. Reliability Data

System designers have long been aware the mean-time-between-failure (MTBF) for most systems is limited by the reliability of the line interface circuitry. This is mainly due to the shear power dissipation of such line circuits. The older devices such as the SN75188 quad driver ran at quite high temperatures with obvious degradation on reliability. For today's products the use of low power bipolar and more recently BiCMOS technologies significantly reduces operating temperatures while maintaining the robustness associated with bipolar designs providing for a more reliable interface. This is show by reliability data collected on TI's products.

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Life test data collected on a range of EIA-232 devices yielded a failure rate of 1.65 FITS (failures per  $10^9$  device hours). This was at an ambient temperature of  $55^{\circ}\text{C}$  (to an upper confidence level of 60% and assumes an activation energy of 0.96 eV).

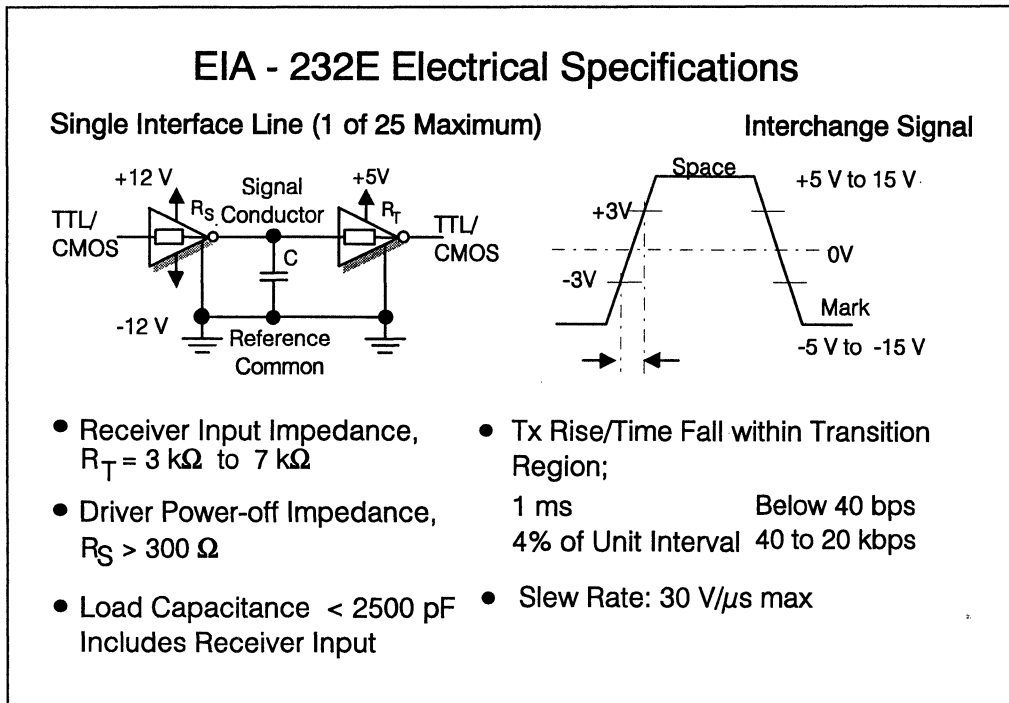
## 2.2.EIA/TIA-232-E Industry Standard for Data Transmission



**Figure 5.12 - EIA-232-E Industry Standard for Data Transmission**

The Electronic Industries Association (EIA) introduced the RS-232 standard in 1962 in an attempt to standardise the interface between Data Terminal Equipment (DTE) and Data Communication Equipment (DCE). The DTE comprises the data source, data sink or both. The DCE provides the functions to establish, maintain and terminate a connection, and to code/decode the signals between the DTE and the data channel. Although emphasis was then placed on interfacing between a modem unit and data terminal equipment, other applications were quick to adopt the EIA-232 standard. The growing use of the PC (personal computer) quickly ensured that EIA-232 became the industry standard for all low-cost serial interfaces between the DTE and peripheral. The mouse, plotter, printer, scanner, digitiser, and tracker-ball, in addition to the external modem unit, are all examples

of peripherals that connect to an EIA-232 port. Using a common standard allows widespread compatibility plus a reliable method for interconnecting a PC to peripheral functions.



**Figure 5.13 - EIA-232-E Electrical Specifications**

The EIA RS-232-C standard, revised in 1969, was superseded by EIA-232-D (1986), and recently has been once again superseded by EIA/TIA-232-E which brings it in-line with CCITT V24, V.28 and ISO IS2110. (TIA refers to the Telecommunication Industry Association). The latest revision includes an update on the rise time to unit interval ratio and reverses the changes made by the 'D' revision, see Figure 5.14. Although an older standard, with problems like high-noise susceptibility, low data rates and very limited transmission length, EIA-232 fulfils a vital need as a low cost communication system. Consequently new products are being developed at a faster rate than ever.

## 2.3.EIA-232 Specification

The standard sets out to ensure:

- i. Compatible voltage and signal levels
- ii. Common pin wiring configurations
- iii. A minimum amount of control information between the DTE and DCE.



It accomplishes this by incorporating the following areas in the standard:

### **Electrical and Signal Characteristics**

Electrical and signal characteristics of the transmitted data in terms of signal voltage levels, impedance's, and rates of change.

### **Mechanical Interface Characteristics**

Mechanical interface characteristics defined as a 25-way "D" connector, with dimensions and pin assignments specified in the standard. Although the standard only specifies a 25-pin D-type connector, most laptop and desktop PCs, today use a 9-pin "DB9S" connector shown in Figure 5.17. The reader should note the DCE equipment connector is male for the connector housing and female for the connection pins. Like wise the DTE connector is a female housing with male connection pins.

### **Handshake Information**

A functional description of the interchange circuit enables a fully interlocked handshake exchange of data between equipment's at opposite ends of the communication channel. However, V24 defines many more signal functions than RS-232, but those that are common are compatible. Twenty two of the twenty five connector pins have designated functions, although few, if any, practical implementations use all of them. The most commonly used signals are also shown in Figure 5.17.

It is worth noting that for applications which use the 25-pin D-type connector there is often a problem in communication due to different handshaking signals employed by each system.

## **2.3.1. EIA-232-E Electrical Specifications**

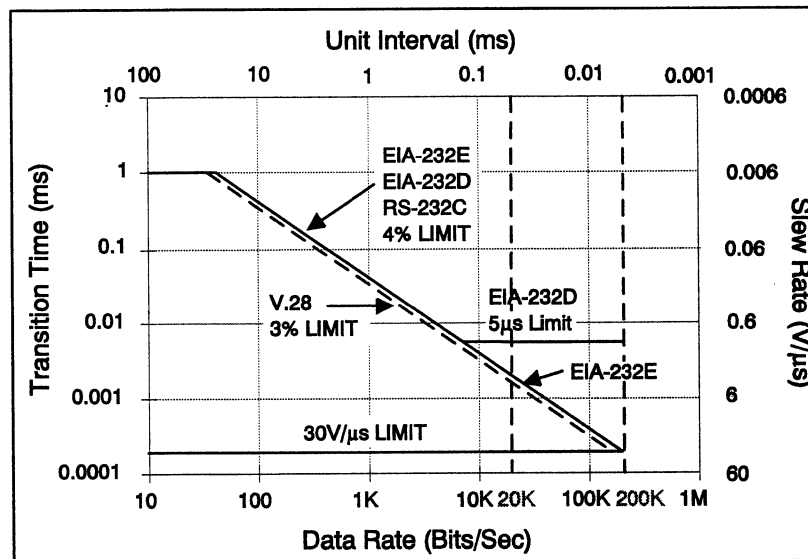
All EIA-232 circuits carry bipolar voltage signals with the voltage at the connector pins not to exceed  $\pm 25$  V. Any pin must be able to withstand short circuit to any other pin without sustaining permanent damage. Each line should have a minimum load of 3 k $\Omega$  and maximum load of 7 k $\Omega$  which is usually part of the receiver circuit. A logic '0' is represent by a driven voltage of between +5 V and +15 V and a logic '1' of between -5 V and -15 V. At the receiving end a voltage of between +3 V and +15 V represents a '0' and a voltage of between -3 V and -15 V represents a '1'. Voltages between  $\pm 3$  V are undefined and lie in the transition region. This effectively gives a 2 volt minimum noise margin at the receiver.

The maximum cable length was originally defined in RS-232C as 15 metres, however this has been revised in EIA-232-D and EIA/TIA-232-E and is now more correctly specified as a maximum capacitive load of 2500 pF. This equates to around 15 to 20 metres line length depending on cable capacitance.

As mentioned in an earlier section, EIA-232 specifies a maximum slew rate of the signal at the output of the driver to be 30 V/ $\mu$ s. This limitation is concerned with the problem of cross talk between conductors in a multiconductor cable. The faster the transition edge the greater the cross talk. This restriction together with the fact of the driver and receiver using a common signal ground and the associated noise introduced by the ground current severely limits the maximum data throughput.

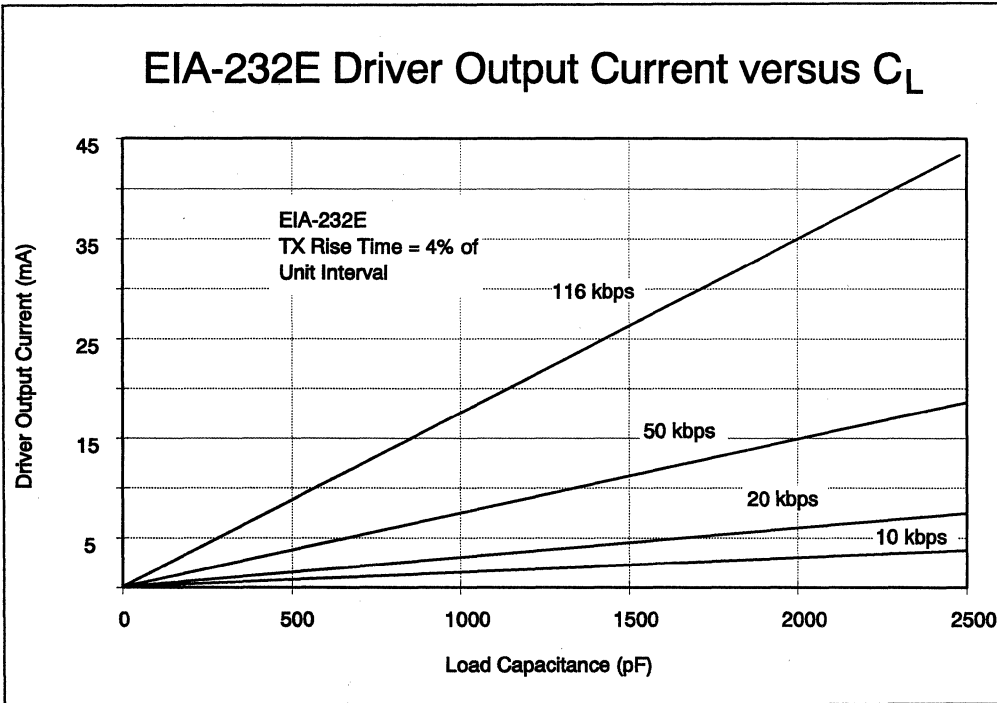
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## 'RS-232' Transition Time versus Data Rate



**Figure 5.14 - 'RS-232' Transition Time versus Data Rate**

For this reason the EIA-232 standard specifies a maximum data rate of 20 kbps. The standard also specifies the relationship between unit interval and rise time through the transition region (+3 V to -3 V) or  $t_T$ . This is the main difference between the 'E' and the 'D' revision. This is shown more clearly in Figure 5.14. EIA-232-D up to 8 kbps specified the relationship between transition time and unit interval or bit time  $t_b$  to be 4% maximum. Above 8 kbps this was relaxed to 5  $\mu$ s maximum independent of the data rate. Both the 'C' and the 'E' revision specify the ratio of  $t_T/t_b$  to be 4% all the way up to 20 kbps. One can extrapolate this further, using the 4% figure and with the maximum slew rate of 30 V/ $\mu$ s, the maximum achievable data rate is 200 kbps however practically this is limited to around 120 kbps. A number of software programs operate at transfer rates of 116 kbps. Furthermore over longer line lengths the maximum drive current or short circuit current of the line driver becomes the dominant feature on data rate as against the 30 V/ $\mu$ s slew rate. As the line length increases the load capacitance also increases requiring more current to maintain the same transition time. The curves shown in Figure 5.15 indicate the drive current required to maintain the 4% relationship at different data rates. In today's low power systems, this level of output current is not sustainable at above say 20 kbps. In practice the line length is usually limited to around 4 metres for the higher data rates. Most drivers can handle the higher transmission rates over this line length without seriously compromising supply current.



**Figure 5.15 - EIA-232-E Driver Output Current versus  $C_L$**

The curves shown in Figure 5.15 were generated using the following equation which is an approximate equation relating transition time  $t_T$ , line capacitance  $C_L$ , receiver input impedance  $R_i$ , driver short circuit current  $I_o$ , and the initial and final line voltage (-3 V and +3 V) of the transition region,  $V_i$  and  $V_f$  respectively,

$$t_T = R_i \times C_L \times \ln \left[ \frac{|R_i \times I_o| + |V_f|}{|R_i \times I_o| - |V_i|} \right]$$

Turning this equation around with respect to  $C_L$ , and cancelling  $R_i$ ,  $V_i$  and  $V_f$  we get:

$$C_L = \frac{1}{3} \times \frac{t_T}{\ln \left[ \frac{I_o + 1}{I_o - 1} \right]} \text{ nF}$$

The voltage levels,  $V_f$  and  $V_i$ , used in this equation are the extremes of the transition region. Assuming a typical driver short circuit current of 20 mA and a receiver input resistance of 5 k $\Omega$ , the typical time taken to pass through the transition region would be :-

$$t_T = 300 \times C_1 \quad \text{seconds.}$$

This equation can be manipulated further to gain a relationship of unit interval with line length in terms of load capacitance and short circuit driver current. The equation in Figure 5.16 assumes conformance to the 4% rule.

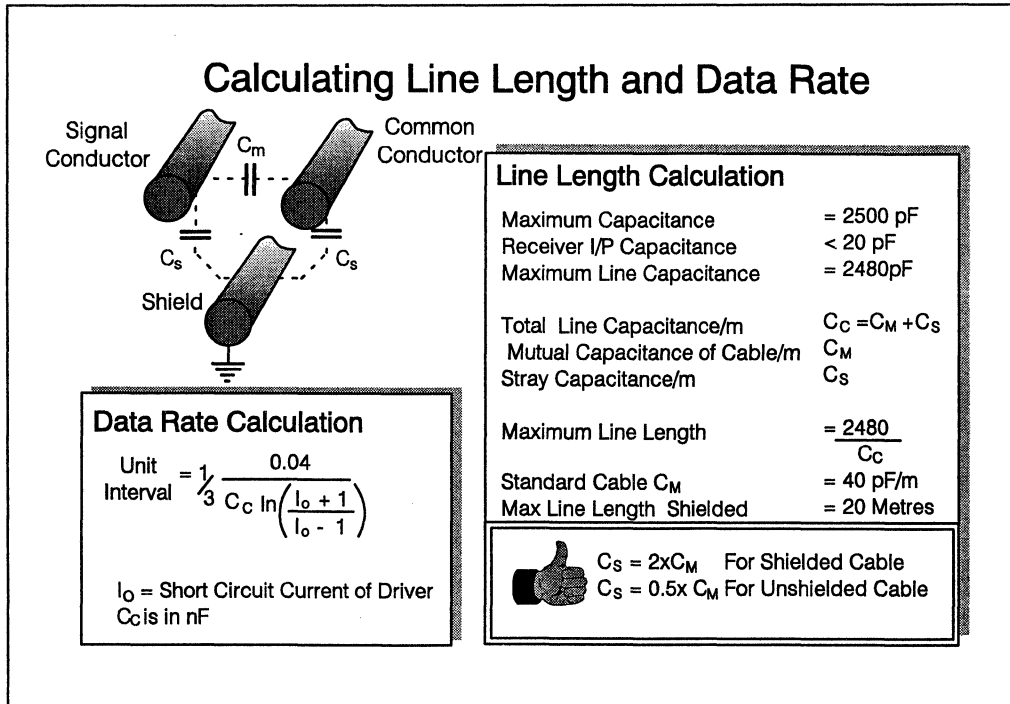


Figure 5.16 - Calculating Line Length and Data Rate

### 2.3.2. Calculating maximum line length

So far we have discussed line length in terms of load capacitance. For practical purposes we must now consider turning this value for load capacitance into true line length. The standard states a maximum line capacitance of 2500 pF. The input capacitance of a receiver is say 20 pF which leaves 2480 pF as the maximum line capacitance.

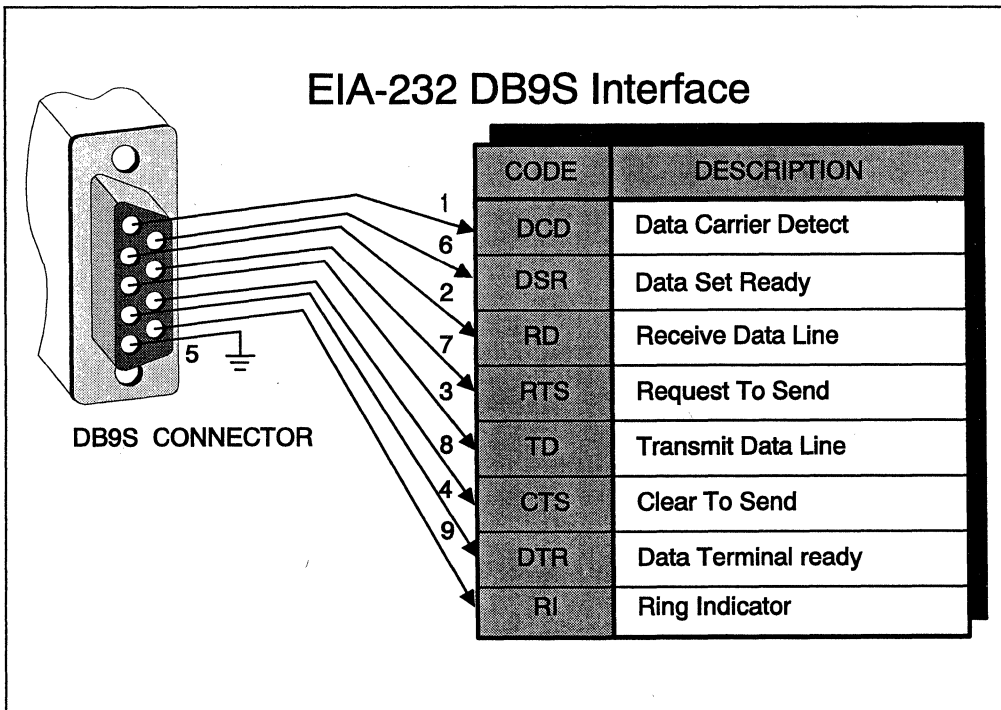
We must now consider the type of cable to be used. Standard EIA-232 cable as supplied by a number of manufacturers has a mutual capacitance of approximately 40 pF per metre. In addition to this we must add the stray capacitance. Stray capacitance varies considerably on whether the line is shielded.

For non shielded cable the stray capacitance is approximately half the mutual capacitance, for shielded cable it is double the mutual capacitance. As can be seen from Figure 5.16, for shielded cable the maximum line length is 20 metres, with unshielded cable it is over 40 metres.

### 2.3.3. The DB9S Connector

As mentioned earlier today's notebook and laptop PCs, with their quest for reduced size, no longer use the standard 25-way D-type connector detailed in the standard but have substituted it for a 9-way D-type. This is commonly known as the DB9S connector. Like the 25-way, the DCE equipment connector is a male outer casing with female connection pins, and the DTE is a female outer casing with male connecting pins.

As the interface is now made up of only nine pins the handshaking lines have been reduced accordingly but still are sufficient for most applications. Figure 5.17 shows the pins assignments for the interconnect cable into the DTE connector. An explanation of the function of each signal is given below:



*Figure 5.17 - EIA-232 DB9S Interface*

**Data Carrier Detect (DCD) - Received Line Signal Detector**

The ON condition on this signal line as sent by the DCE informs the DTE that it is receiving a carrier signal which meets its suitability criteria from the remote DCE. In modems, this circuit is held on as long as it is receiving a signal that can be recognised as a carrier. On half duplex channels, DCD is held off when RTS is in the on condition.

**Data Set Ready (DSR)**

This is a signal turned on by the DCE to indicate to the DTE that it is connected to the line.

**Receive Data Line (RD)**

The signals on the RD line are in serial form . When the DCD signal is in the off condition the RD line must be held in the Mark state.

**Request to Send (RTS)**

This signal is turned on by the DTE to indicate it is now ready to transmit data. The DCE must then prepare to receive data. In half duplex operation, it also inhibits the receive mode. After some delay the DCE will turn the CTS line on to inform the DTE it is ready to receive data. Once communication is over and no more data is transmitted by the DTE, RTS is then turned from on to off by the DTE. After a brief time delay to ensure all data has been received that was transmitted, the DCE turns CTS off.

**Transmit Data Line (TD)**

The signals on this circuit are transmitted serially from DTE to DCE. When no data is being transmitted the signal line is held in the Mark state. For data to be transmitted, DSR, DTR, RTS and CTS must all be in the on state.

**Clear to Send (CTS)**

This signal is turned on by the DCE to indicate to the DTE that it is ready to receive data. CTS is turned on in response to simultaneous on condition of the RTS, DSR and DTR signals.

**Data terminal Ready (DTR)**

This in conjunction with DSR indicate equipment readiness. DTR is turned on by the DTE to indicate to the DCE it is ready to receive or transmit data. DTE must be in the on condition before the DCE can turn on DSR. When DTR is turned off by the DTE, the DCE is removed from the communication channel following the completion of transmission.

**Ring Indicator (RI)**

The ring indicator is turned on by the DCE while ringing is being received and is a term left over from the use of the standard in telephone line modem applications. Primarily used in auto-answer systems.

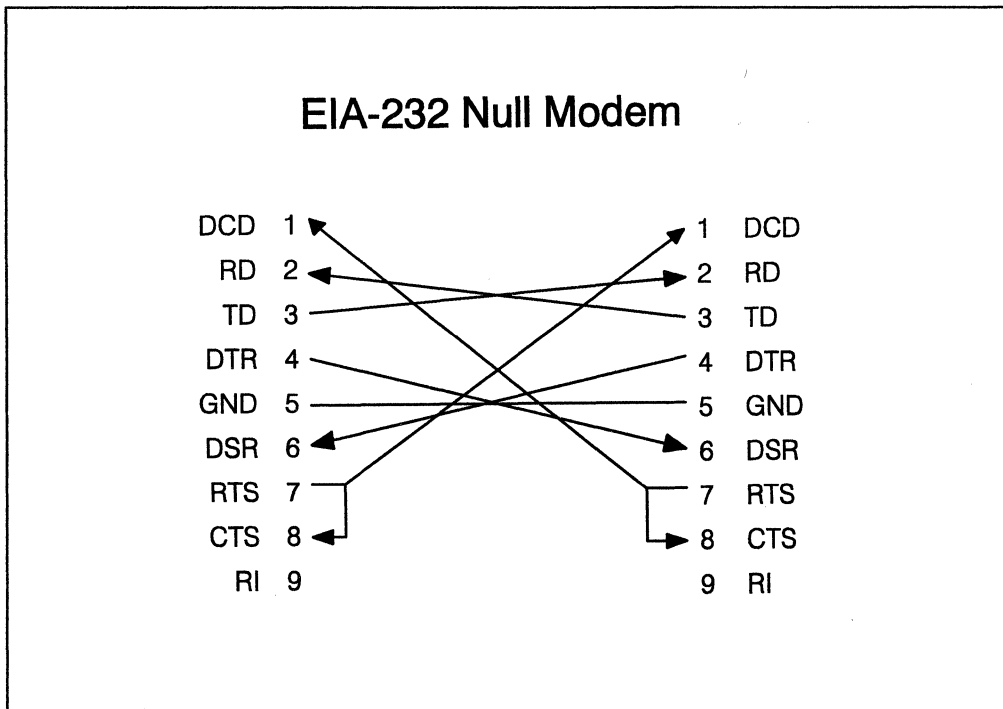
**Signal Ground (pin 5)**

This is the ground which provides the common ground reference for all the interchange circuits and is separate from the protective ground. The protective ground is electrically bonded to the equipment frame and is usually directly connected to the external ground. Any static discharges are then routed directly to ground without affecting the signal lines.

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While all these pins are assigned, once again not all equipment uses every pin. Consider the mouse which can use as few as 4 lines, Signal ground, RI, TD and RD. Most equipment does however utilise a minimum of RTS, DTR, TD, RD, CTS and DSR.

Also of note is the usage of the DTE interface. The majority of equipment uses this interface and makes use of the null modem as a means of communication between DTEs. The null modem makes use of feeding back the RTS signal to the CTS line on each interface, Figure 5.17.1 details the connections for implementing a full null modem for the DB9S connector.



*Figure 5.17.1 - EIA-232 Null Modem*

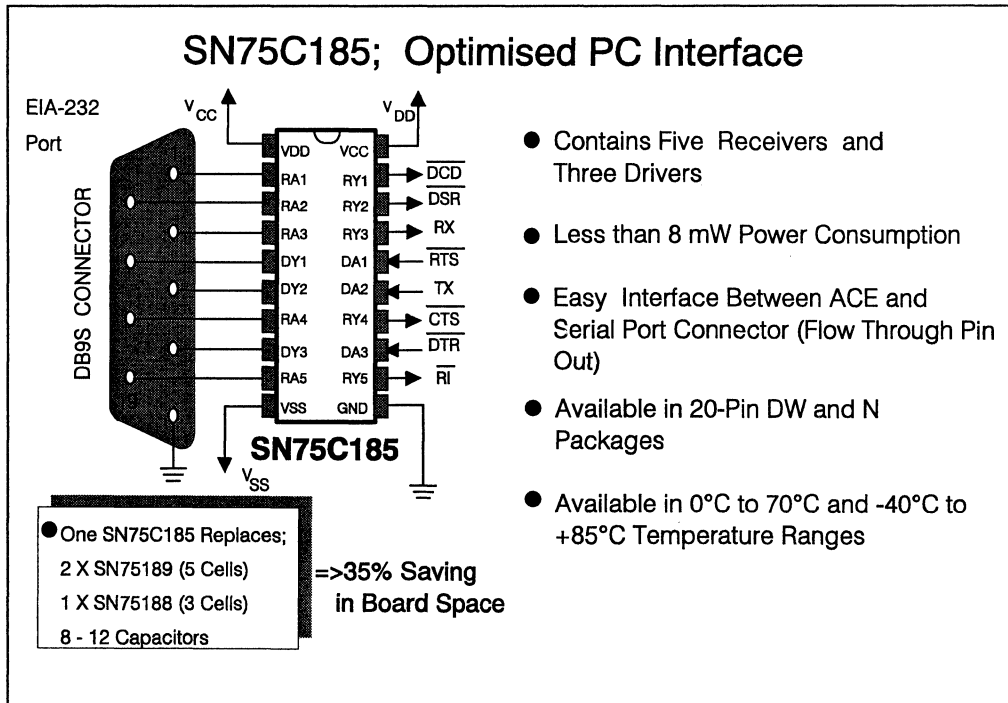
## 2.4.SN75C185: Optimised PC Interface

If we study the DB9S DTE interface further we see there are 3 transmit lines and 5 receive lines. This is an awkward combination for the standard EIA-232 IC configurations in use today. Consider the ubiquitous SN75188 and SN75189 quad drivers and receivers. To implement this interface would require 3 ICs, one '188 and two '189s. Equal combinations of drivers such as the triple driver/receiver of the SN75C1406 still requires two chips to implement the interface.

For this reason TI has developed the SN75C185. By providing the exact combinations of driving and receiving elements, along with the necessary passive components, a highly optimised solution can be

provided – the SN75C185 is just that. The SN75C185 integrates three drivers and five receivers and includes the necessary capacitors for driver slew-rate limit (30 V/ $\mu$ s) and receiver filter implementation, all in a single 20-pin package

The designer's dilemma is eased further by the use of a flow-through pin out architecture, see Figure 5.18. By aligning one side of the SN75C185 with the pins of the DB9S connector and the other to industry standard ACEs or UARTs, printed circuit board (PCB) layout can be greatly simplified.



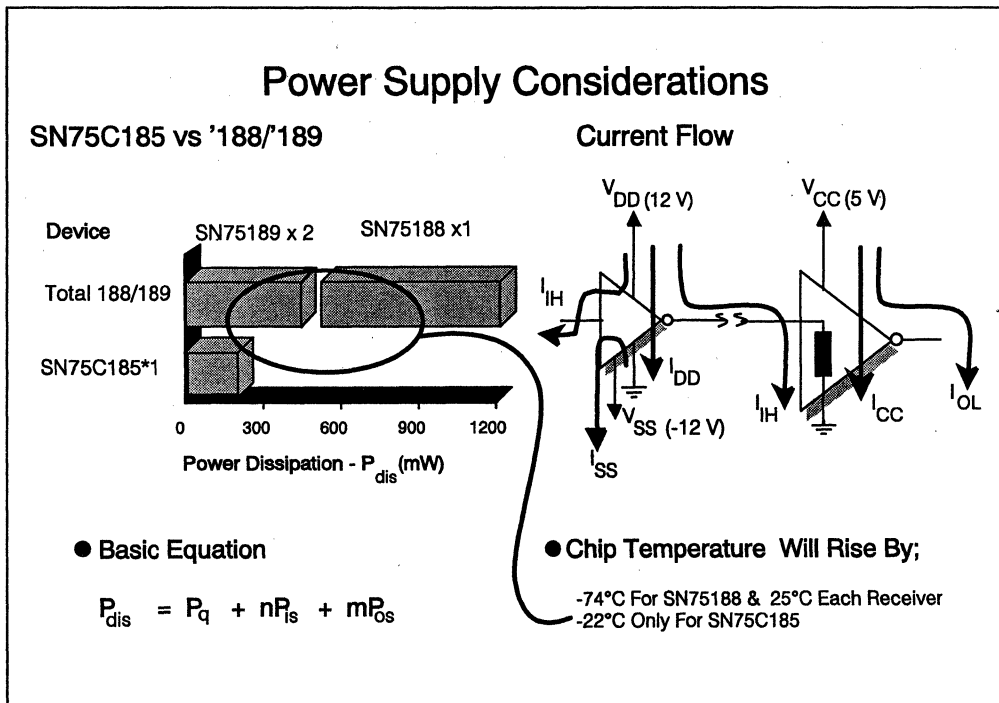
*Figure 5.18 - SN75C185; Optimised PC Interface*

### 2.4.1. Low Power as Well

In common with all of Texas Instruments BiMOS products, these devices combine the benefits of Bipolar's drive capability and robustness along with the low-power consumption of CMOS. This power saving, when compared to the alternatives is calculated in the following pages and is illustrated graphically in Figure 5.19.

Available in either a single 20-pin, wide-bodied SO pack or DIP pack, the SN75C185 offers designers greater than 25% saving in board space, compared to alternate solutions.





*Figure 5.19 - Power Supply Considerations*

### 2.4.2. SN75C185; Power Considerations

System power consumption is often considered very late in the design cycle. Of even more concern is that the power consumption of the interface circuitry, being the least attractive circuit to design, is often totally overlooked. The consequences of this can be catastrophic especially when using devices in confined spaces. These areas will normally have very poor air circulation, causing the ambient temperature of the whole system to increase.

These types of problems are particularly difficult to diagnose as failure can often be intermittent as devices pass into and out of thermal shutdown.

For these reasons, low quiescent-power devices are becoming a necessity for modern applications. As digital technologies advance, their power consumption decreases, making the interface circuits the limiting factor as far as system power consumption is concerned.

### 2.4.3. Interface Power Consumption Calculations

Before the availability of the SN75C185 common implementations of EIA-232 require one quad-driver package and two quad-receiver packages; in the driver chip, one device is redundant while in

the receiver chips, three devices are redundant. These devices would, however, still be taking their quiescent current and hence wasting power. In order to provide the interface signals, three integrated circuits were required while only two-thirds of the capability was being used. The calculations below demonstrate this difference.

When comparing the 'C185 solution to that provided by the SN75188 and SN75189 devices, the power saving is enormous.

Both implementations require three supply voltages; a 5 V and  $\pm 12$  V supplies. The power dissipated,  $P_{dis}$ , within each device is the quiescent power of the device,  $P_q$ , plus the power dissipated in the input stage,  $P_{is}$ , and the power dissipated in the output stage,  $P_{os}$ , (when it is driving the line).

Hence,

$$P_{dis} = P_q + nP_{is} + mP_{os}$$

Where  $n$  is the number of active input stages and  $m$  is the number of active output stages.

### **SN75188/SN75189 Combination**

Using an SN75188 for the driver, the quiescent power consumption would be 576 mW. In addition to this the power dissipated in the input stage,  $P_{isd}$ :-

$$\begin{aligned} P_{isd} &= V_{CC} * I_{IL} \\ &= 12 * 1.6 \text{ mW} \\ &= 19.2 \text{ mW}. \end{aligned}$$

This is multiplied by four to take into account all four drivers, putting the fourth driver into a defined state so as to reduce any noise problems that could be introduced by leaving the input floating.

The power dissipated in the output stage,  $P_{osd}$ , is:

$$\begin{aligned} P_{osd} &= (V_{CC} - V_{OH}) * \frac{V_{OH}}{R_L} \\ &= (12 - 9) * \frac{9}{3} \text{ mW} \\ &= 9 \text{ mW}. \end{aligned}$$

This figure will be multiplied by three to take into account the active three drivers driving the interface line. These sum up to give a total power dissipation of

$$\begin{aligned} P_{dis} &= 576 + 4 * 19.2 + 3 * 9 \text{ mW} \\ &= 680 \text{ mW}. \end{aligned}$$

The junction temperature of a DIP device would have risen by 74°C.

Using the SN75189 receivers, a quiescent power of 130 mW would be dissipated by each package. This would be multiplied by two to take into account both chips.

The power dissipated in the output stage has a similar equation to that of the driver.

$$\begin{aligned}P_{\text{osr}} &= V_{\text{OL}} * I_{\text{OL}} \\ &= 0.45 \times 10 \text{ mW} \\ &= 4.5 \text{ mW}\end{aligned}$$

This power dissipated is multiplied by five to take into account the five receivers being used. The input stage can also dissipate some power, but this power is not supplied by this part of the interface system. The power dissipated within the IC will however cause the junction temperature to rise.

$$\begin{aligned}P_{\text{isr}} &= \frac{V_{\text{OH(d)}}^2}{R_{\text{L}}} \\ &= \frac{9^2}{3} \text{ mW} \\ &= 27 \text{ mW}\end{aligned}$$

This power dissipation is then multiplied by five. The remaining receivers will require tying to a state where they will not be susceptible to noise. Tying them to the 5 V supply increases the power dissipation by a further 8.3 mW per receiver.

Assuming three receivers in one SN75189 are being used and two receivers in the other, the power dissipated for the first receiver is:

$$\begin{aligned}P_{\text{dis}} &= 130 + 4 \times 27 + 3 \times 4.5 \text{ mW} \\ &= 233 \text{ mW}.\end{aligned}$$

The power dissipated in the second receiver is:-

$$\begin{aligned}P_{\text{dis}} &= 130 + 4 \times 27 + 2 \times 4.5 \text{ mW} \\ &= 210 \text{ mW}.\end{aligned}$$

This raises the temperature of the first and second receiver by 25°C and 23°C, respectively.

The total power dissipated by the SN75188/189 combination is the sum of these three powers, equalling **1.12 W**.

### Using the SN75C185

The power dissipation of the SN75C185 can be calculated in a similar manner. The quiescent-power consumption of the SN75C185 is equal to:-

$$\begin{aligned}P_{\text{q}} &= V_{\text{DD}} * I_{\text{DD}} + V_{\text{SS}} * I_{\text{SS}} + V_{\text{CC}} * I_{\text{CC}} \\ &= 12 * 200 + -12 * -200 + 5 * 750 \quad \mu\text{W} \\ &= 8.55 \text{ mW}\end{aligned}$$

The power dissipated in the input stage of the driver is:-

$$\begin{aligned}P_{\text{isd}} &= V_{\text{DD}} * I_{\text{IL}} \\ &= 12 \times 1 \quad \mu\text{W}\end{aligned}$$

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$$= 12 \mu\text{W}.$$

This is multiplied by three to take into account all of the drivers.

The power dissipated in the output stage of the driver,  $P_{\text{osd}}$ , is:

$$\begin{aligned} P_{\text{osd}} &= (V_{\text{DD}} - V_{\text{OH}}) \times \frac{V_{\text{OH}}}{R_{\text{L}}} \\ &= (12 - 10) \times \frac{10}{3} \text{ mW} \\ &= 6.67 \text{ mW}. \end{aligned}$$

This is multiplied by three to take into account the three drivers driving the interface line, giving a power dissipation of 20 mW.

The power dissipated in the output stage of the receiver has a similar equation to that of the driver, so:

$$\begin{aligned} P_{\text{osr}} &= V_{\text{OL}} \times I_{\text{OL}} \\ &= 0.4 \times 3.2 \text{ mW} \\ &= 1.28 \text{ mW} \end{aligned}$$

This value is multiplied by five giving a total of 6.4 mW of power dissipated in the receiver's output stages. The input stage will also dissipate some power, but this power will not be supplied by this part of the interface system. The power dissipated within the chip will however cause the junction temperature to rise.

The power dissipated in the input stage,  $P_{\text{isr}}$ , equals:

$$\begin{aligned} P_{\text{isr}} &= \frac{V_{\text{OH}(d)}^2}{R_{\text{L}}} \\ &= \frac{10^2}{3} \text{ mW} \\ &= 33.3 \text{ mW} \end{aligned}$$

This power dissipation will also require multiplying by five. Giving a total input power dissipation of 167 mW.

Summing all the power contributors the total power dissipation is given by;

$$\begin{aligned} P_{\text{dis}} &= P_{\text{q}} + 3P_{\text{isd}} + 3P_{\text{osd}} + 5P_{\text{isr}} + 5P_{\text{osr}} \\ &= 8.55 + 3 \times 12 \times 10^{-3} + 3 \times 6.67 + 5 \times 33.3 + 5 \times 1.28 \text{ mW} \\ &= 201 \text{ mW}. \end{aligned}$$

The total power dissipated by the SN75C185 is **201 mW**

This represents a tremendous power saving, especially when considering that the line is still being driven. The temperature rise within the SN75C185 would only be 22°C, enabling it to operate more reliably and with higher ambient temperatures.

#### 2.4.4. On Chip Slew Rate Limiting

The EIA-232-E standard specifies a maximum slew rate through the transition region of 30 V/μs. Relating this to capacitance and current only 100 μA of output current into 30 pF load capacitance is needed to exceed the slew-rate limit. All devices are capable of supplying more than 5 mA. Therefore if the slew rate limit is not to be exceeded, the switching speed of the driver's output stage needs to be reduced. An established solution is to place loading capacitors on the output of the driver. The value of the loading capacitor required will depend upon the line length, but it is generally in the order of 330 pF. The effect of this capacitor is to cause the output transistors to saturate, causing it to short circuit current limit, thus preventing fast switching edges.

There are some major problems with this established process; one being the variance in current at which the output short-circuit current limit operates, especially when taking temperature changes into consideration. Again the value of capacitance placed on the line will depend upon the driver's output short-circuit capability as well as line length. For example a device capable of sourcing 10 mA will need a total capacitance of 330 pF placed on its output to meet the 30 V/μs slew rate limit, while placing this value across a device capable of sourcing 4 mA will have its slew rate limited to less than 12 V/μs.

Another problem encountered is the increase in power dissipation through the output stage. The output voltage of the driver will normally be close to one supply rail, so when it tries to switch to the other, the active transistor will have almost all of the supply voltages across it. The extra external capacitor will clamp the driver's voltage close to the supply voltage causing the output transistor to source large amounts of current. The combination of a large source current and large voltage cause it to dissipate large amounts of power. Operating at these prolonged bursts of high current will ultimately increase the chip temperature which in turn can affect the long-term life of the device. Bipolar technologies are normally much better able to withstand such effects

A better solution, and that employed by the drivers in the SN75C185, is to place the slew-rate limiting within the chip itself. Using similar techniques to those employed for slew-rate-limited operational amplifiers, the slew rate of line drivers can also be limited. Using the Miller capacitance multiplying effect, the slew rate of the driver can be slowed down. The on-chip capacitors are normally in the order of 5 pF, while the currents driving the on-chip capacitor are the order of micro amperes, thus reducing power consumption within the device. The biasing current to the output transistors is unaffected by this technique and will be more than sufficient to drive the 3 kΩ load as offered by the receiver.

#### 2.4.5. Internal Noise Filtering

The standard states a maximum line cable capacitance of 2500 pF, which corresponds to an approximate line length of 20 metres. As the interface line gets longer, it becomes more susceptible to noise pick-up from the surrounding environment. This pick-up is due in part to the inductive nature of the line. As the signal switches, a rapidly changing magnetic field induces noise currents

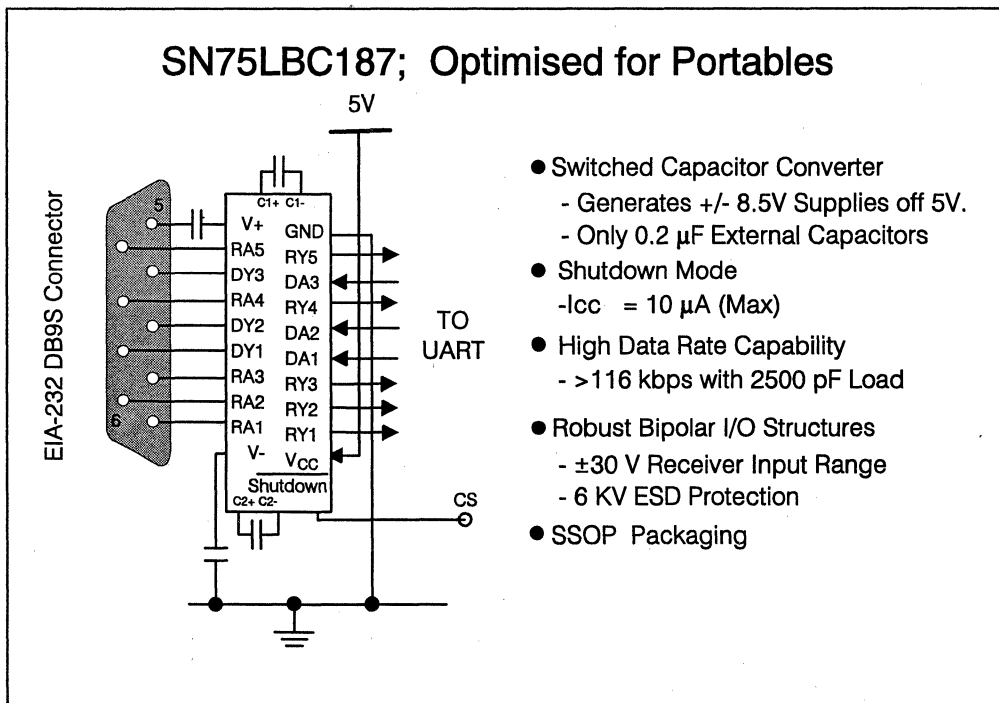
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into the line, thereby corrupting signal data. The level and cause of this noise will dictate the nature of solutions or precautions that should be taken.

For operation at high data rates, the use of a differential line might be the best solution. If however, a low cost and simple single-ended solution is required then standard EIA-232 devices can be modified to give noise protection. This is achieved by slowing down the response of the receiver's input stage, making them too slow to respond to fast switching noise pulses. Even small levels of input noise can falsely trigger the receiver. The maximum data rate specified in the standard is 20 kbps, corresponding to a minimum pulse period of 100  $\mu$ s. Therefore in normal applications, most devices are far faster than the specification requires.

To slow down older bipolar receivers such as SN75189s, a capacitor,  $C_c$ , needed to be placed on each of its response control pins. This means an additional four capacitors per device, which can be awkward and costly. The effect of this response control capacitor is to set up a low-pass filter on the receiver's input. In order to provide large pulse rejection, the capacitor needs to be quite large. Furthermore, the filter response is asymmetric, affording protection against positive noise voltage spikes only, negative spikes are unaffected, and will tend to attenuate rather than reject short noise pulses.

Receivers in the SN75C185 integrate on-chip filtering which reject fast transient noise pulses. The on-chip filters are more precise than filters implemented using external passives. Consequently the receiver response is unaffected. These filters are totally symmetrical, offering protection against both positive and negative noise pulses and with the ability to reject rather than attenuate short noise pulses. To approach the level of filtering offered by the 'C185 receivers the standard '188 type receivers require much larger capacitors and even then fall well short of filtering requirements.



*Figure 5.20 - SN75LBC187; Optimised for portables*

## 2.5. SN75LBC187; Optimised for Portables

The SN75C185 is the ideal choice for computer applications where the bipolar supplies required by EIA-232 are available within a computer system. Most desk top computers generate  $\pm 12$  volt supplies for powering the internal disk drive. However for portable equipment, e.g. laptops, notebooks, hand held measuring equipment, the EIA-232 interface may be the sole user of a negative supply. The cost of implementing a switch mode supply, using inductive switching regulators, to generate the negative supply can make this option unattractive. Switch mode supplies also have the drawback of increasing the EMI emissions, a factor becoming an increasingly important design constraint. Integrating a switch mode power supply on silicon would reduce the emissions, and has been the dream of semiconductor manufacturers, but thus far no one has yet managed to integrate the inductor.

An alternative way, and the basis of modern technology charge pumps, is to make switching regulators using capacitors. In essence they operate by applying charge to a capacitor via an input voltage and then adding, subtracting or inverting the voltage on the positive or negative voltage terminals. This charge is transferred into a holding reservoir capacitor that is then used to supply the

output voltages. Furthermore such a scheme can be integrated into silicon. Using a network of capacitors both voltage doublers and invertors can be made.

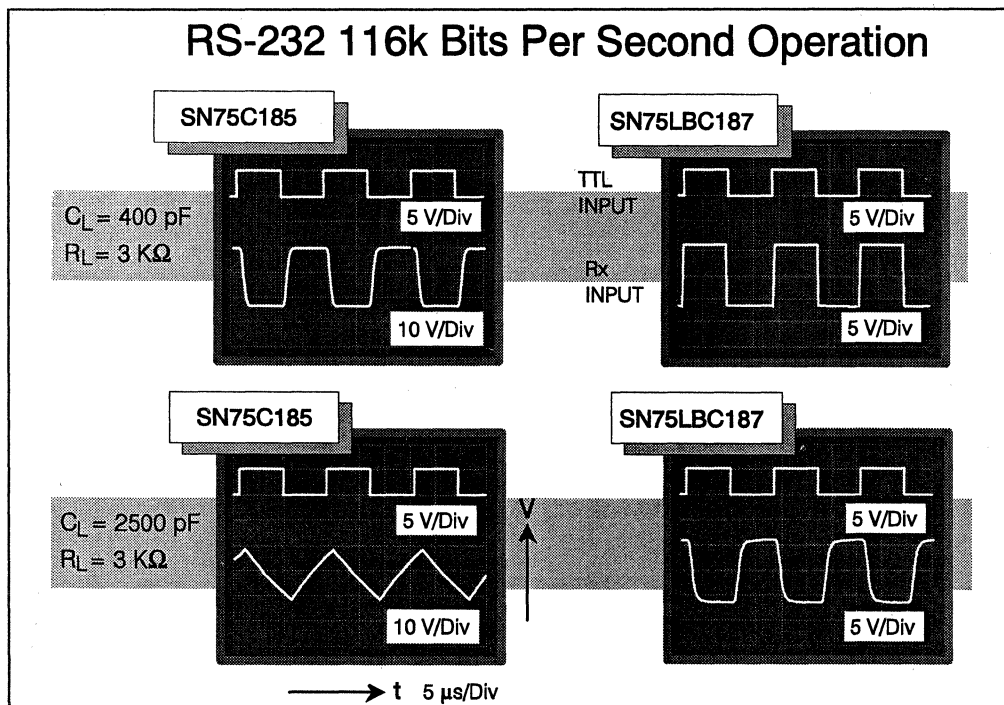
The SN75LBC187 integrates the charge pump on the same IC as the EIA-232 drivers and receivers. It is fabricated in TI's proprietary LinBiCMOS technology and contains three independent drivers and 5 independent receivers together with the switched-capacitor voltage converter. The SN75LBC187 provides a single 5 V supply interface between the asynchronous communications element (ACE or UART) and the serial port connector of the data terminal equipment (DTE). This device has been designed to conform to standards EIA/TIA-232-E-1986 and EIA/TIA-562 and CCITT recommendation V.28.

The switched-capacitor voltage converter of the SN75LBC187 uses four small (0.2  $\mu$ F) external capacitors to generate the positive and negative voltages required by EIA-232 line drivers from a single 5 V logic supply input. Like the SN75C185 the drivers feature output slew-rate limiting to eliminate the need for external filter capacitors. The receivers can accept  $\pm 30$  V without sustaining damage. Furthermore the LBC187 is guaranteed to withstand up to 6KV ESD on any of its pins making it TI's most rugged EIA-232 product.

The device also features a reduced power or shutdown mode that virtually eliminates the quiescent power supply when the IC is not active.

The primary application for the LBC187 is for battery operated, portable equipment where power consumption is a key factor. A separate consideration, and one that usually goes hand in hand with these factors, is that of sheer physical size. With the LBC187, TI has used the latest SSOP packaging to reduce board area to an absolute minimum. The new SSOP package reduces board space to 43% of the standard 28-pin SOIC package. Couple this with the small 0.2  $\mu$ F and you have the ideal single supply solution for space restricted applications.





**Figure 5.21 - RS-232 116k Bits Per Second Operation**

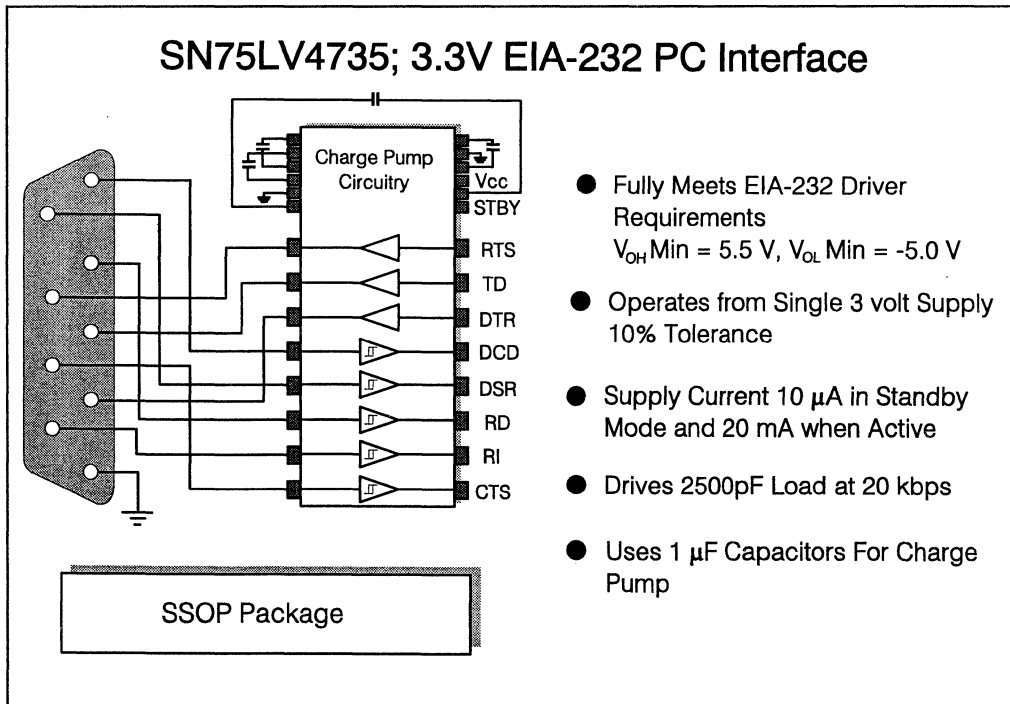
### 2.5.1. SN75LBC187; 116 kbps operation

As discussed in section 2.3.1 the limitation on data rate is one of short circuit output current and the actual load capacitance. With the 'LBC187 the driver short circuit current,  $I_{OS}$ , is higher than say the SN75C185 and is therefore able to drive longer line lengths at higher data rates. Figure 5.21 illustrates this. The 400 pF load in the top half of the figure represents a cable approximately 3 metres long. As the 'scope traces show, the 'C185 produces a perfectly acceptable output trace at 116 kbps. Similarly with the 'LBC187 trace.

If the line length is now upped to 20 metres or 2500 pF load, we can see how the short circuit current limit now limits the slew rate. With the 'LBC187 the trace is still acceptable and will provide reliable data transmission. With the 'C185, the data will still be transmitted but the probability of error is now increased. Most software programs that operate 116 kbps, e.g. Laplink™ (Laplink is a trademark of Travelling Software Inc.) provide the interconnect cable as part of the system. In most cases this cable is less than 3 metres in line length so either the 'C185 or 'LBC187 would be able to transmit data reliably. It is interesting to note that both devices would meet EIA-232-D if the rise time to unit interval relationship was extrapolated, however both would fail EIA-232-E. Of course conformance to EIA-232 is not relevant above 20 kbps.

## 2.5.2. Conformance to EIA-562

A new standard has recently been introduced in an attempt to provide a low power standard for 5 volt systems and also to increase the data rate over EIA-232. Known as EIA-562, the standard increase the maximum data rate from 20 kbps to 64 kbps and facilitates lower driver voltages. The downside is the reduced noise margin at the receiver. The specification also details the rise time and ripple conditions of the driver. The SN75LBC187 is fully conformant to this standard.



*Figure 5.22 - SN75LV4735; 3.3V EIA-232-E PC Interface*

## 2.6. SN75LV4735; 3 Volt EIA-232 PC Interface

Continuing the move to lower power systems the obvious choice is to reduce the supply voltage of the system. Assuming supply current remains constant power dissipation is instantly reduced. The driving force behind this reduction is once again the notebook type PC equipment. To facilitate the move to 3 Volts, TI has introduced the SN75LV4735. From 3 volts the device is still capable of producing the required  $V_{OH}$  and  $V_{OL}$  for conformance with EIA-232.

Once again the device is designed specifically for the DB9S PC DTE interface containing 3 drivers and 5 receivers for a single package solution.

The device is packaged in the TSSOP package with a board area of only 22 mm<sup>2</sup> and a maximum package height of 1 mm.

## 2.7.ACEs (UARTs) From Texas Instruments

Most EIA-232 systems use dedicated communication controllers. Termed ACEs (Asynchronous Communication Elements) or UARTs (Universal Asynchronous Receiver Transmitter), these devices are responsible for controlling the exchange of information over the EIA-232 interface.

### The ACE

The ACE is a dedicated asynchronous communications controller designed to off load most of the communication activities from the CPU, thus freeing the CPU for other activities. It has the ability to add or delete start and stop bits and provide odd/even parity code generation and detection. Industry standard devices such as the TL16C450 family contain many extra features as listed below:

- Programmable bps-rate generator
- Adds and deletes standard asynchronous communication bit
- Fully programmable serial interface characteristics
- Data communication diagnostic capability
- Modem-control functions
- Simple interface to microprocessors
- Maximum data rate of 256 k bits per second

All devices are designed using Texas instruments EPIC™ CMOS process and operate from a single 5 V supply. The TL16C450 is the most common choice for standard PC applications as well as many other asynchronous serial applications. The TL16C450, housed in a 40-pin package, contains all the necessary facilities for implementing a single asynchronous serial port. The CPU within the system can read and report on the status of the ACE at any point in the ACEs operation. Reported status information includes the type of transfer operation in progress, the status of the operation, and any error conditions encountered, parity, overrun etc.

The TL16C450 ACE includes a programmable, on-board, bps-rate generator. This generator is capable of dividing a reference clock input by divisors from 1 to (2<sup>16</sup> -1) and producing a 16 x clock for dividing the internal transmitter logic. Provisions are included to use this 16 x clock to drive the receiver logic. Also included in the ACE is a complete modem control capability and a processor interrupt system that may be software tailored to the user's requirements to minimise the computing required to handle the communications link. The TL16C451 is similar to TL16C450 with the single serial port, but also contains a Centronix parallel printer port. The IBM PC AT/XT sets the standard for this parallel printer interface that all "compatible" manufactures have to follow. TTL-level signals are presented on a 25-pin D-type socket. Apart from the choice of connector, this parallel printer port is directly compatible with the "Centronix" standard printer interface. The TL16C452 has two serial ports plus a parallel Centronix printer port. Using this ACE together with two SN75C185s provides a simple three chip complete solution for the two EIA-232 ports plus a printer port that is common on basic PC configurations.

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EPIC is a trademark of Texas Instruments Incorporated

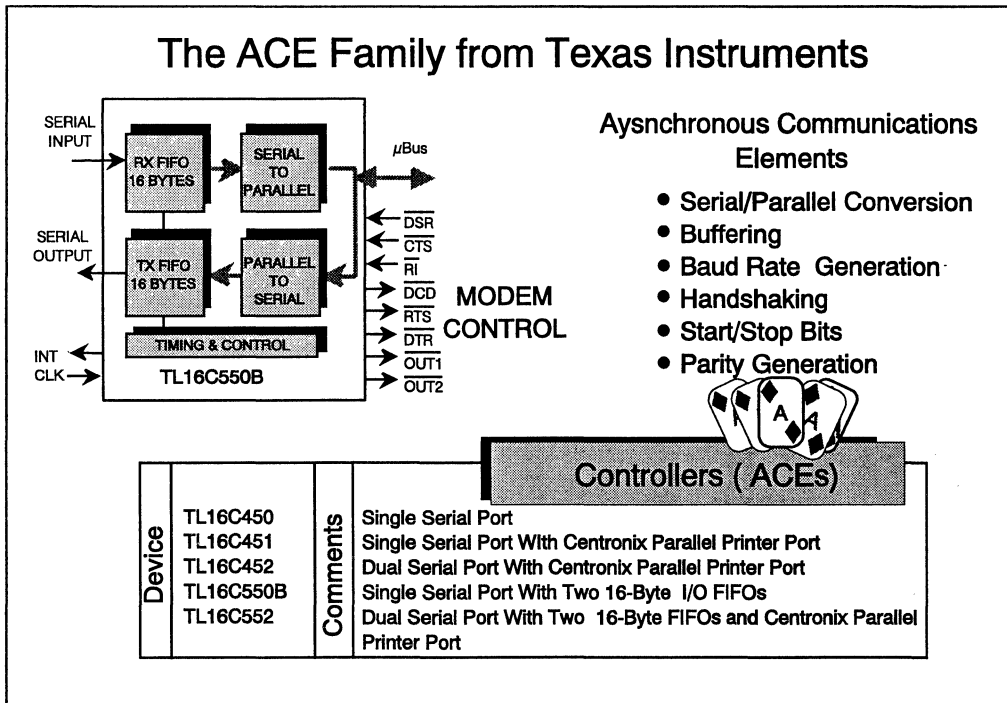


Figure 5.22.1 - The ACE Family from Texas Instruments

### 2.7.1. The FIFO (First-In-First Out)

The CPU can send data at much faster rates than a normal ACE can handle. This is particularly true for today's multitasking applications that demand high performance microprocessors. This can be expensive in CPU overheads as the CPU will be tied to the speed of the serial interface, i.e., data will be transferred over the interface through the ACE and onto the CPU bus. This is true also when the data is exchanged from the CPU to interface via the ACE.

Devices like the TL16C550B and TL16C552 alleviate this problem by including buffer registers and FIFOs in series with the ACE's transmitter and receiver. These are quick access registers that hold data until the CPU can be freed. The CPU can then execute a block read or write.

The ACE is, in effect, isolated from the slow communications channel.

The TL16C550B is similar to the TL16C450, but two 16-byte FIFOs are included to buffer the transceiver and receiver data stream, further reducing the number of interrupts from the microprocessor.

### **2.7.2. Forward-Looking Performance With Backward Compatibility**

By allowing two modes of operation, the TL16C550B allows users to maintain software compatibility with earlier industry standard ACEs such as the TL16C450. In addition to the TL16C450 mode, the TL16C550B can operate in the FIFO mode. In FIFO mode, two 16-byte FIFOs (First-In-First-Out) are enabled to relieve the CPU of excessive software overheads. The independent receive and transmit FIFOs act as buffers, vastly reducing the number of interrupts required. Furthermore two dedicated pins serve as handshaking lines to a DMA (Direct Memory Access) controller, thus allowing the FIFOs to load and unload data without direct intervention from the CPU.

The flagship of the range is the TL16C552, which is similar to the TL16C452 in structure but with the added advantage of input/output FIFOs as in the TL16C550B

This device serves two serial input/output interfaces simultaneously in either microcomputer or microprocessor-based systems. In addition to its dual asynchronous serial communication capabilities, the TL16C552 provides a fully bi-directional parallel data port that fully supports the parallel Centronix-type printer. The parallel port and the two serial ports provide IBM PC/AT compatible computers with a single low-power device to serve the three-port system. Like the TL16C550B, the TL16C552 contains 16-byte receive and transmit FIFOs that act as buffers to reduce the number of interrupts on the CPU. Also in common with the TL16C550B, the device contains two pins for each ACE that serve as handshaking lines for DMA control. The TL16C552 is housed in a 68-pin plastic-leaded chip carrier, PLCC.

**Integration of FIFO and DMA signalling circuitry onto a single chip makes the TL16C550B and TL16C552 one of the most efficient solutions for higher performance multitasking systems.**



rate is established by the reference oscillator clock frequency (xin) and divided by a driver specified by the bps generator divisor latches.

**TXRDY :**

Transmitter Ready Output. This pin is used during DMA signalling.

**RXRDY :**

Receiver Ready Output. This pin is also used during DMA signalling.

**D0 to D7:**

Databus. Eight 3-state data lines provide the bi-directional path for data, control, and status information between the ACE and CPU bus.

**RD1 , RD2:**

Read inputs. When either input is active (high or low respectively) during ACE selection, the CPU is allowed to read status information from the selected ACE register. Since only one of these inputs is required for the transfer of data during the read operation, RD2 is tied to its inactive state, i.e., low.

**DCD , DSR , SIN , RTS , SOUT , CTS , DTR , RI:**

These signals are the EIA-232 compatible modem control lines. Devices such as the SN75C185 are employed to convert the TTL/CMOS level signals from the ACE to EIA-232 compatible bipolar voltages of between  $\pm 5$  V to  $\pm 15$  V. The signal can then be transmitted over distances of up to 15 m.

The advantages of the SN75C185 can be clearly seen by the simplicity of the interface connections. For example, driving/receiving combinations precisely match the interface requirement, plus the pin-out is aligned directly to the DB9S connector.

**WR1 , WR2:**

Write inputs. A logic applied to WRI , during ACE selection allows the CPU to write either control words or data into a selected ACE registers. WR2 is tied in active, i.e.: logic low.

**INTERRUPT:**

When active (high) the interrupt pin informs the CPU that the ACE has an interrupt to be serviced. This interrupt could occur for one of four reasons;

- Receiver error
- Received data available or time-out (FIFO mode only)
- Transmitter holding register empty
- Enable modem status interrupt

The interrupt is reset (deactivated) either when the interrupt has been serviced or by a master reset (MR).

**MR:**

Master reset. When active (high), MR clears most ACE registers and sets the states of various outputs (i.e. interrupt).

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**CS0, CS1,  $\overline{\text{CS2}}$  :**

Chip Select. An active low on the  $\overline{\text{CS2}}$  pin selects the ACE. CS0 and CS1 must be tied active (high) to ensure proper functioning of the  $\overline{\text{CS2}}$  chip select. A logic high on  $\overline{\text{CS2}}$  will de-select the ACE.

**A0 to A2:**

Register Select. These three inputs are used during read or write operations to select the appropriate ACE registers. For example, providing the correct write/read operation had taken place at logic 0 at A2, A1, and A0 would cause the receiver buffer (read) or the transmitter buffer to write.

**$\overline{\text{ADS}}$  :**

Address strobe. An active low on  $\overline{\text{ADS}}$ , the register select signals (A0 TO A2) and chip-select signal ( $\overline{\text{CS2}}$ ) drive the internal logic directly.

## 2.8.EIA-232 Products Summary

In this section we have discussed devices which are concerned primarily with the DB9S connector. TI also has a wide range of other EIA-232 ICs which offer differing combinations of drivers and receivers which can offer the optimum solution for your system. The reader is advised to consult the current edition of Interface Circuits Data Book (Reference SLYD006) which contains a complete selection guide of EIA-232 products.



## 2.9.EIA-232 Selection Guide

### Data Transmission Circuits

Function	Per Package	Type	Features
	2	SN75150	Industry Standard
		UA9636AC	Industry Standard
Line Driver	4	LT1030	Robust bipolar design, with 3-state driver outputs
		SN55188	-55°C to 125°C temperature range
		SN75188	Industry standard
		SN65C188	-40°C to 85°C temperature range
		SN75C188	Low-power BiMOS
		SN65C198	-40°C to 85°C temperature range
		SN75C198	Low-power BiMOS with sleep-mode

Line Receiver	4	SN75154	Industry standard
		SN55189	-55°C to 125°C temperature range
		SN75189	Industry standard
		SN55189A	-55°C to 125°C temperature range
		SN75189A	-55°C to 125°C temperature range
		SN65C189	-40°C to 85°C temperature range
		SN65C189A	-40°C to 85°C temperature range
		SN75C189	Low-power BiMOS
		SN75C189A	Low-power BiMOS

Continued Over.....

## Data Transmission Circuits (Continued)

Function	Per Package	Type	Features
Line Driver / Receiver	1/1	SN75155	On-chip 5-v regulator
	2/2	MAX232	On-chip charge pump
	2/2	LT1080	On-chip charge pump and 3-state outputs
	2/2	LT1080	On-chip charge pump
	3/3	LT1039	Robust bipolar design, with 3-state outputs
	3/3	SN65C1406	-40°C to 85°C temperature range
	3/3	SN75C1406	Low-power BiMOS
	4/4	SN75186	Robust bipolar design, with loopback
	4/4	SN65C1154	-40°C to 85°C temperature range
	4/4	SN75C1154	Low-power BiMOS
	3/5	SN65C185	-40°C to 85°C temperature range
	3/5	SN75C185	Optimised for DB9S (9-pin) connector
	3/5	SN75LV4735	3 volt operation
	3/5	SN75LBC187	Optimised for Laptop Applications <sup>¥</sup>

## Control Circuits

Function	Type	Features
ACE <sup>+</sup>	TL16C450	Single ACE
	TL16C451	Single ACE with parallel port
	TL16C452	Dual ACE with parallel port
	TL16C550B	Single ACE with FIFO <sup>§</sup>
	TL16C552	Dual ACE with parallel port and FIFO

## Notes

<sup>+</sup> ACE: Asynchronous Communications Element.

<sup>§</sup> FIFO: First In First Out.

<sup>¥</sup> Product currently under development, contact TI representative for further details.

## 3. Interface Circuits for RS-485

### 3.1. The Need for Balanced Transmission Line Standards

This section focuses on industry's most widely used balanced transmission line standard, the EIA RS-485. After reviewing key aspects of the standard, the reader will be introduced to the practicalities of implementing a differential transmission scheme based on a factory automation example. Finally, new additions to Texas Instruments EIA product range will be discussed along with their application, where appropriate.

Data transmission between computer system components and peripherals over long distances and under high noise conditions, usually proves to be very difficult if not impossible with single-ended drivers and receivers. Recommended EIA standards for balanced digital voltage interfacing provide the design engineer with a universal solution for long line system requirements.

RS-485 is a balanced (differential) digital transmission line interface developed to incorporate and improve upon the advantages of the current-loop interface and improve on the EIA-232 limitations. The advantages are;

- Data rate - to 10 Mbps and beyond
- Longer line length - up to 1200 metres
- Differential transmission - less noise sensitive

#### 3.1.1. Application Areas

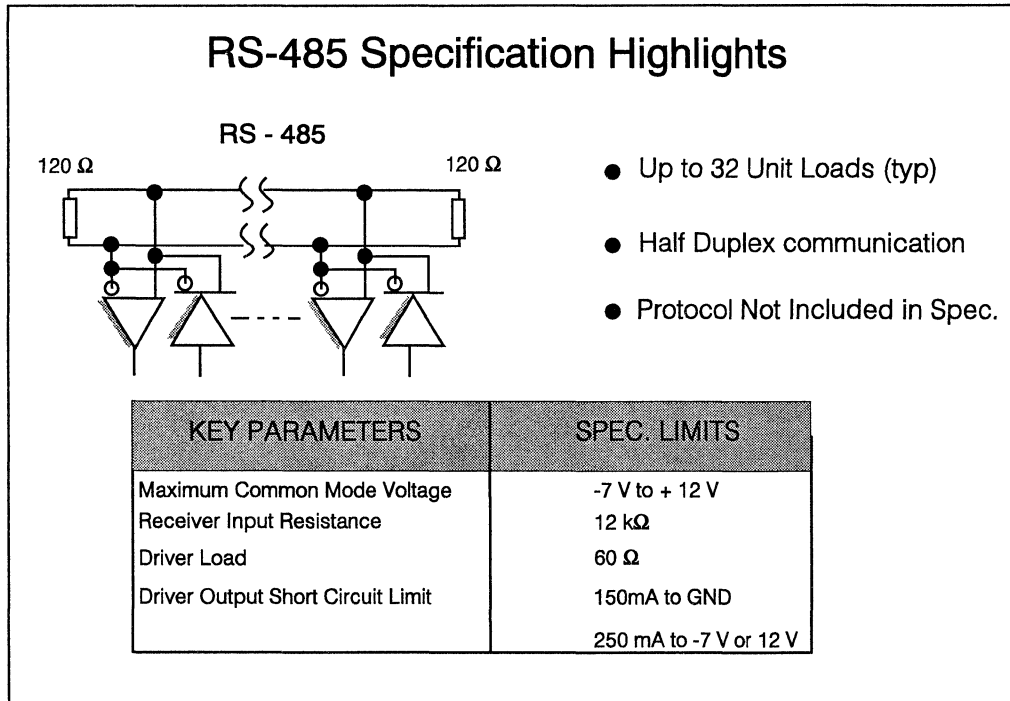
RS-485 is an upgraded version of RS-422-A extending the number of peripherals and terminals that a computer can interface to, particularly where longer line length or increased data rates are called for. Additionally, RS-485 allows for bi-directional multi-point party line communication and can effectively be used for "mini-LAN" applications, such as data transmission between a central computer and remote intelligent stations. For example, between point of sales terminals and a central computer for automatic stock debiting.

As a result of its versatility an increasing number of standard's committees are embracing the RS-485 as the physical layer specification of their standard. Examples include the ANSI (American Nationals Standards Institute) Small Computer Systems Interface (SCSI) which we will discuss in section 4, the Profibus standard, the DIN Measurement Bus.

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### 3.1.2. EIA RS-485

The balanced transmission line standard EIA RS-485 was developed in 1983 to interface a host computer's data, timing or control lines to its peripherals. The standard specifies the physical layer only. Protocols, timing, serial or parallel data, connector choice are all left to be defined by the user



**Figure 5.23 - RS-485 Specification Highlights**

RS-485 was originally defined as an upgrade and more flexible version of RS-422-A. Where RS-422 facilitates simplex communication only, RS-485 allows for multiple drivers and receivers on a single line facilitating half-duplex communication. Like RS-422 the maximum line length is not specified but, based on 24 AWG cable, is nominally around 1.2 km. Maximum data rate is unlimited and is set by the ratio of rise time to bit time, similar to EIA-232. In many cases it is the line length of the cable which limits the data rate more than the drivers due to transmission line effects, see section 1.

The differences between the RS-485 standard and the RS-422 standard lie primarily in the features that allow reliable multi-point communications.

### 3.1.3. RS-485 Driver features

- i. One driver can drive as many as 32 unit loads (one unit load is typically one passive driver and one receiver).
- ii. The driver output, off-state, leakage current should be 100  $\mu$ A or less with any line voltage from -7 V to +12 V.
- iii. The driver should be capable of providing a differential output voltage of 1.5 V to 5 V with common-mode line voltages from -7 V to 12 V.
- iv. Drivers must have self protection against contention (multiple drivers contending for the transmission line at the same time).

### 3.1.4. RS-485 Receiver features

- i. High receiver input resistance, 12 k $\Omega$  minimum.
- ii. A receiver input common-mode range of -7 V to 12 V.
- iii. Differential input sensitivity of  $\pm$ 200 mV over a common-mode range of -7 V to 12 V.

## 3.2. Process Control Design Example

To fully understand the considerations of designing an RS-485 system it is advantageous to take a specific design example. In this case we will consider a factory automation system with a host controller and several out-stations. Each out-station is capable of transmitting as well as receiving data.

The general system specification is shown in Figure 5.24 and comprises:

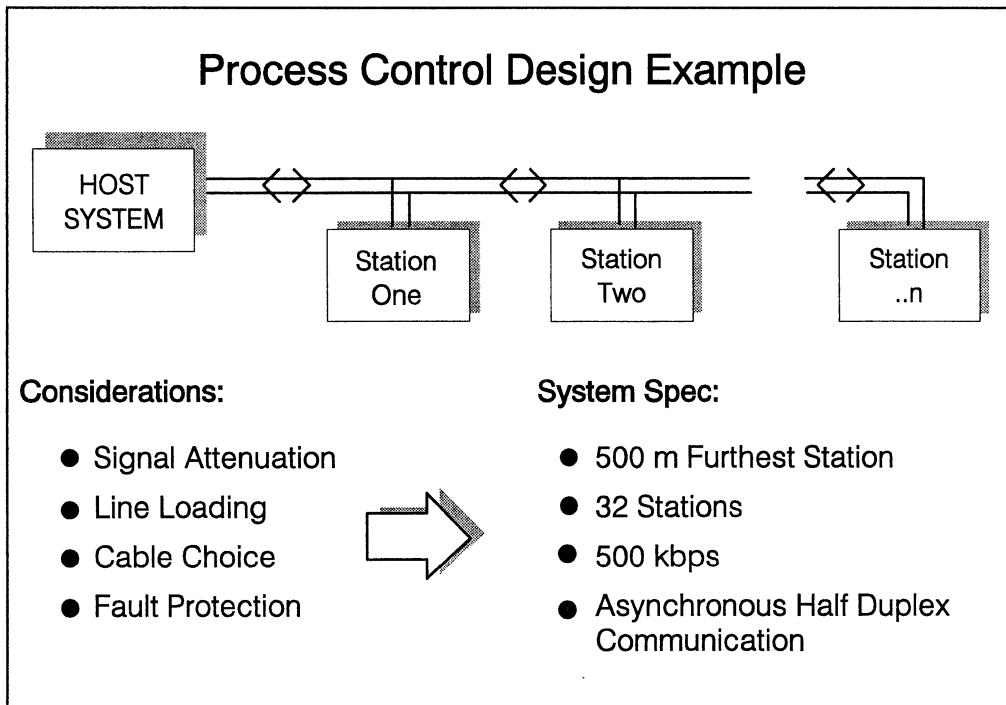
- i. Furthest out-station is 500 m from the host controller.
- ii. We require up to 31 out-stations on the line. With the host controller this totals 32 stations in total.
- iii. System data rate will be 500 kilobits per second.
- iv. Only one cable will be used for data transmission operating in half duplex mode.

With this system specification the main design consideration are:

- i. Line Loading including termination.
- ii. Cable choice
- iii. Signal Attenuation and distortion
- iv. Fault Protection including fail safe operation

Consider each one of these points:

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*Figure 5.24 - Process Control Design Example*

### 3.3.Line Loading

The RS-485 standard takes into account the need for line termination and the subsequent loading on the transmission line. The decision on whether to terminate or not will be system dependent and will be affected by the choice of line driver and the maximum line length.

#### Line Termination

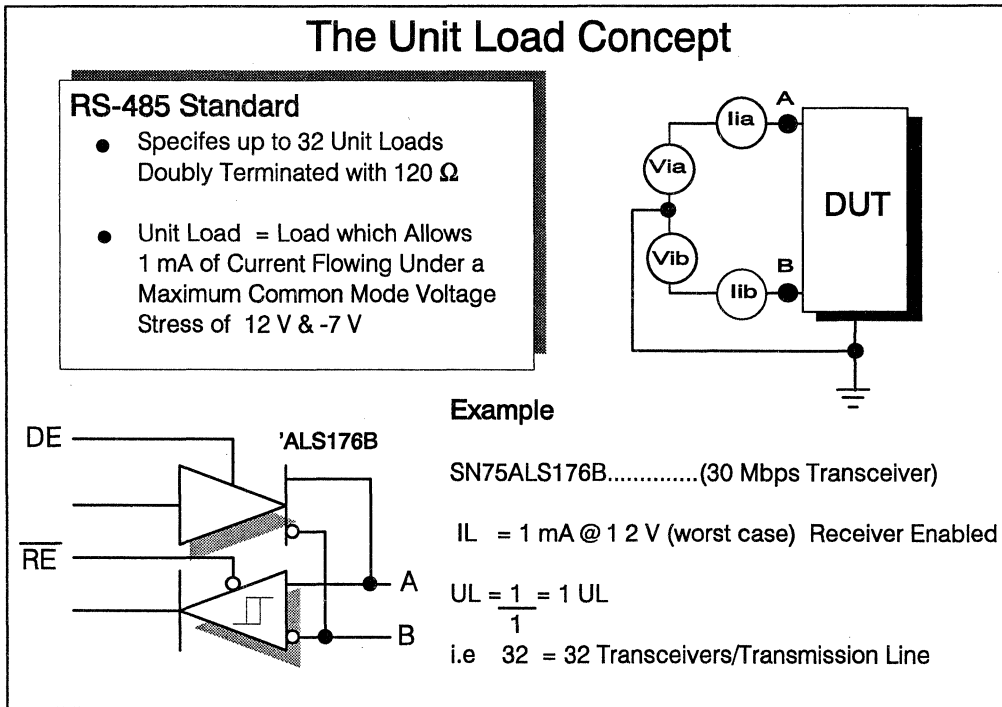
As we discussed in section 1 the test for whether a transmission line is to be considered as a distributed parameter model or a lumped parameter model is dependent upon the relationship of signal rise time,  $t_T$ , at the receiving end and the propagation time of the signal down the cable. The threshold between the two types of transmission line is given by the following equation:

$$2t_{pd} = t_T$$

If we build a margin of error into this equation a better test is to determine the relationship of twice the rise time to 5 times the propagation delay:

If the relationship  $2t_{pd} \geq 5t_T$  is true then the transmission line must be treated as a distributed

parameter model and terminated accordingly. If the converse i.e.  $2t_{pd} \geq \frac{t_T}{5}$  is true, the transmission line can be treated as a lumped parameter model and termination is not necessary.



**Figure 5.25 - The Unit Load Concept**

To determine  $t_T$  tests must be carried out on the transmission cable. For the purposes of this example we chose a low cost non shielded, twisted pair cable - 500 m of a Belden type 8205 cable as supplied by RS Components Ltd of the UK, reference number 360-964. On the driving and receiving ends we connected a SN75ALS176 single channel transceiver. The rise time at the receiver end measured :

$t_T = 0.9 \mu s$  to the 10% and 90% points.

Assuming a propagation delay down the line of 5 ns/m, the time  $t_{pd} = 500 \times 5 = 2500$  ns or 2.5  $\mu s$ .

so in this case:

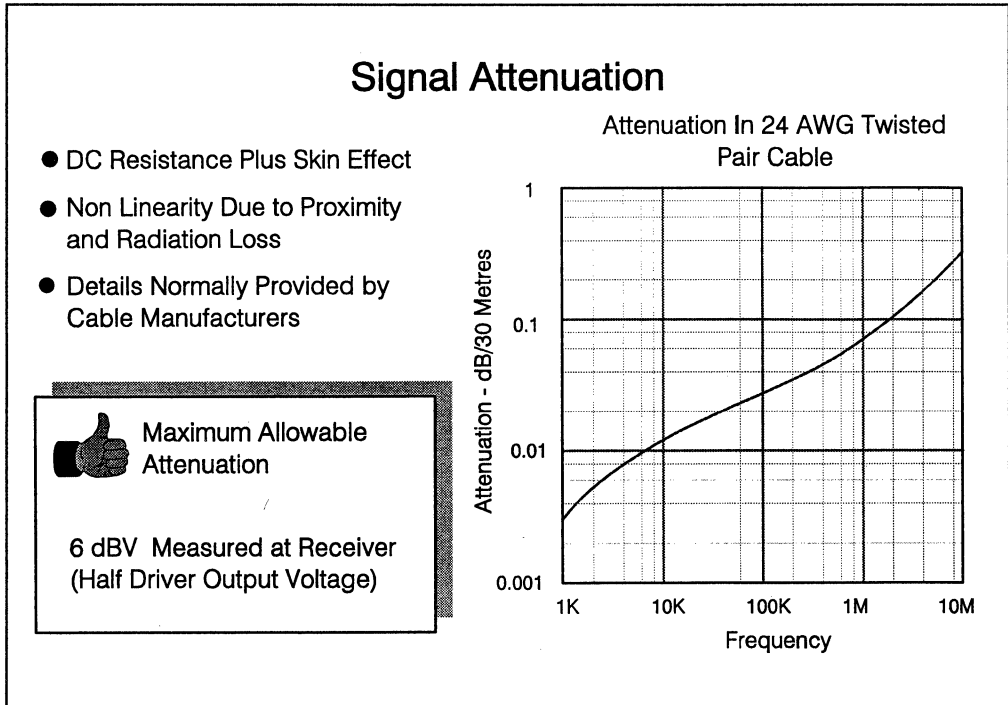
$5 t_T = 4.5 \mu s$  and  $2t_{pd} = 5 \mu s$ , so  $2 t_{pd} > 5 t_T$  and therefore the transmission line should be considered as having a distributed parameter model and consequently must be terminated in its

characteristic impedance. In this case as we are using half duplex transmission the line must be terminated at the furthest ends.

To determine the characteristic impedance of the cable we used the technique described in section 1.5.4.  $Z_0$  in this case measured at  $100 \Omega$ .

### The Unit Load Concept

The maximum number of drivers and receivers that can be placed on a single RS-485 communication bus depends upon their loading characteristics relative to the definition of a unit load (U.L). RS-485 recommends a maximum of 32 unit loads per line.



**Figure 5.26 - Signal Attenuation**

One U.L. (at worst case ) is defined as a load that allows 1 mA of current under a maximum common-mode voltage stress of 12 V. The loads may consist of drivers and/or receivers but does not include the termination resistors, which may present additional loads as low as  $50\Omega$  total for doubly terminated lines.

The example in Figure 5.25 shows a unit load calculation for the SN75ALS176B. Since this device is internally connected as a transceiver, i.e. driver output and receiver input connected to the same bus, it is difficult to obtain separate driver leakage and receiver input currents. For this calculation

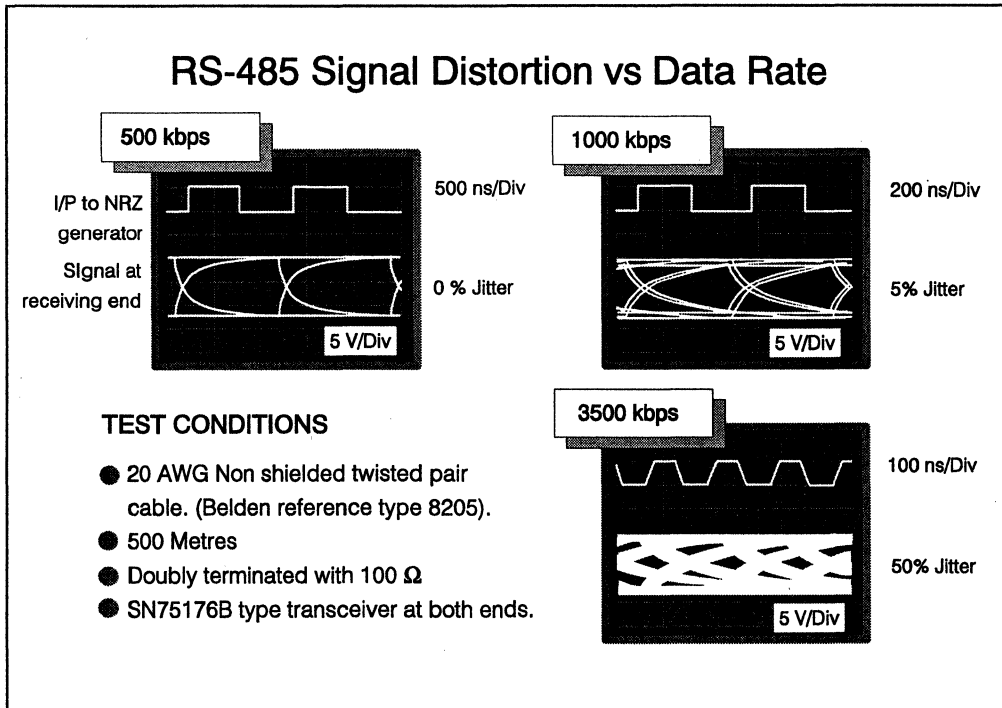


reference is made to the receiver input resistance, 12 k $\Omega$ , giving a transceiver current of 1 mA. This can be taken to represent 1 U.L. which will allow up to 32 devices to be connected to the line.

Obviously it may be possible to connect more devices than the RS-485 recommendation, but this is at the designer's risk.

### 3.3.1. Signal Attenuation

Section 1.3 discusses attenuation in more detail but a sufficient rule of thumb is where the attenuation of the line reduces the driven signal by no more than 6 dBV. Attenuation figures are usually supplied by cable manufacturers. The curve in Figure 5.26 shows the attenuation curve versus frequency for 24 AWG cable. For 500 metres of cable and using the 6 dBV figure, the maximum attenuation we can tolerate is 0.35 dBV/30 metres. In this case the 500 kbps data rate attenuation is well within this limit. The attenuation of the fundamental frequency and higher frequency components of the signal up to 10 Mbps will still be detectable at the receiver. This effect coupled with the the variation of signal velocity with frequency (termed dispersion) results in distortion of the pulse at the receiving end of the line.



*Figure 5.27 - RS-485 Signal Distortion vs Data Rate*

### 3.3.2. Signal Distortion Vs Data Rate

The simplest way to determine the effects of random noise, jitter, attenuation, dispersion, on the inter symbol interference is by the use of eye patterns. For information on how to set up eye patterns, refer to section 1.4 of this Section. Figure 5.27 shows the distortion of the signal at the receiving end of 500 metres of 20 AWG twisted pair cable at different data rates. Using the system constraint of 500 kbps, we see the distortion is limited to the rounding of the signal pulse. If the data rate is increased further, the effects of jitter then become noticeable. In this case at 1 Mbps we begin to observe 5% jitter. At 3.5 Mbps we start to lose the signal completely and the quality of transmission is severely degraded. The maximum allowable jitter in a system should be limited to 5%. The causes of jitter are discussed in more detail in section 1.4.2.

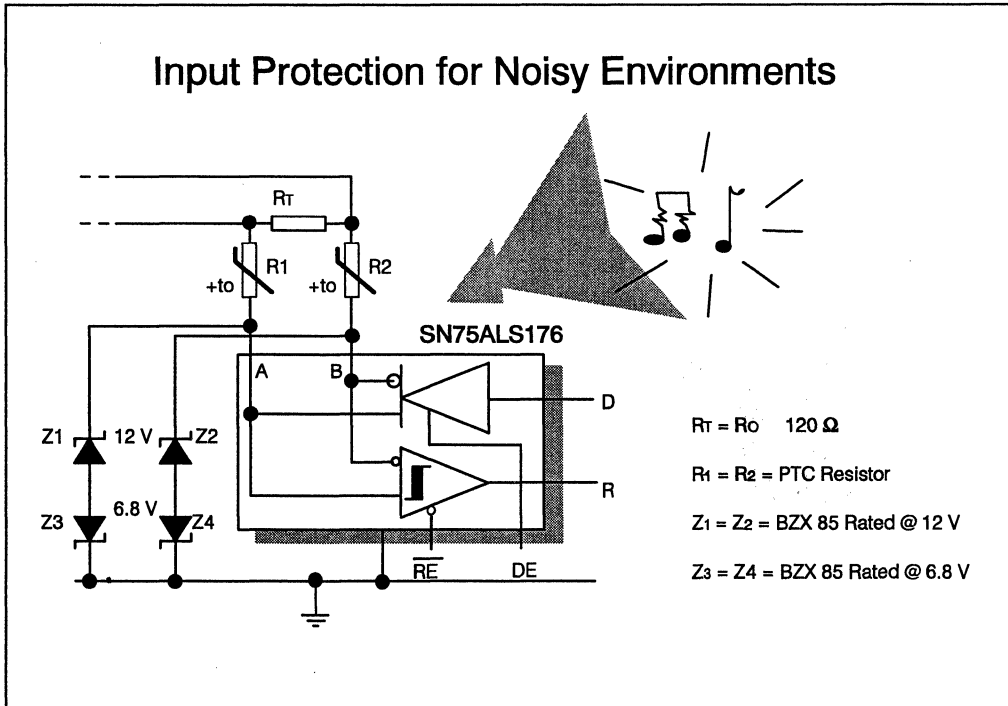


Figure 5.28 - Input Protection for Noisy Environments

### 3.3.3. Fault Protection and Fail Safe Operation

#### Fault Protection

Factory control applications generally require protection against excessive noise voltages. The noise immunity afforded by the differential transmission scheme, and in particular the wide common mode voltage range of RS-485 can be insufficient. Protection can be accomplished in a number of ways,

the most effective being through galvanic isolation which we will discuss later. Galvanic isolation provides system level protection but does not necessarily limit the voltages induced on the transmission lines with respect to the RS-485 driver/receiver grounds. This can be accomplished by the use of protection diodes.

Figure 5.28 shows how external diodes offer transient spike protection for the SN75ALS176 RS-485 transceiver.

$R_T$  is the usual termination resistance and is equivalent in value to the characteristic impedance of the line. Positive Temperature Coefficient resistors,  $R_1$  and  $R_2$ , provide current limiters for the diode chain. Provided their ambient temperature resistance is kept below  $50\Omega$  they will be transparent during normal usage and will not alter the termination value or attenuate the driver output voltage.

$Z_1$  and  $Z_2$  are chosen to protect the input from positive spikes greater than 12 V whilst  $Z_3$  and  $Z_4$  protect the device from negative going spikes greater than -6.8 V.

### Fail Safe Operation

The feature of fail safe protection is also a requirement in many RS-485 applications, however its usefulness needs to be considered and understood at an application level.

### The Need For Fail Safe Protection

In any party line interface system, with multiple driver/receivers, there will be long periods of time when the driving devices are in-active. This state known as line idle and occurs when the drivers place their outputs into a high impedance state. During line idle, the voltage along the line is left floating, i.e. indeterminate - neither logic high or logic low. As a result the receiver could be falsely triggered into either a logic high or logic low state, depending upon the presence of noise and the polarity of the floating lines. This is obviously undesirable as the circuitry following the receiver could interpret this as valid information. The receiver should be able to detect such a situation and place its outputs into a known, and pre-determined state. The name given to methods which ensure this condition is called fail safe. An Additional feature which a fail safe should provide is to protect the receiver from shorted line conditions which can again cause erroneous processing of data and/or receiver damage.

There are several ways implement a fail safe, including a hard-wired fail safe using line bias resistors or protocols. Protocols, although complicated to implement, are the preferred method. However since most system designers, hardware designers in this case, prefer to implement such functions in hardware a hard-wired fail safe is often implemented.

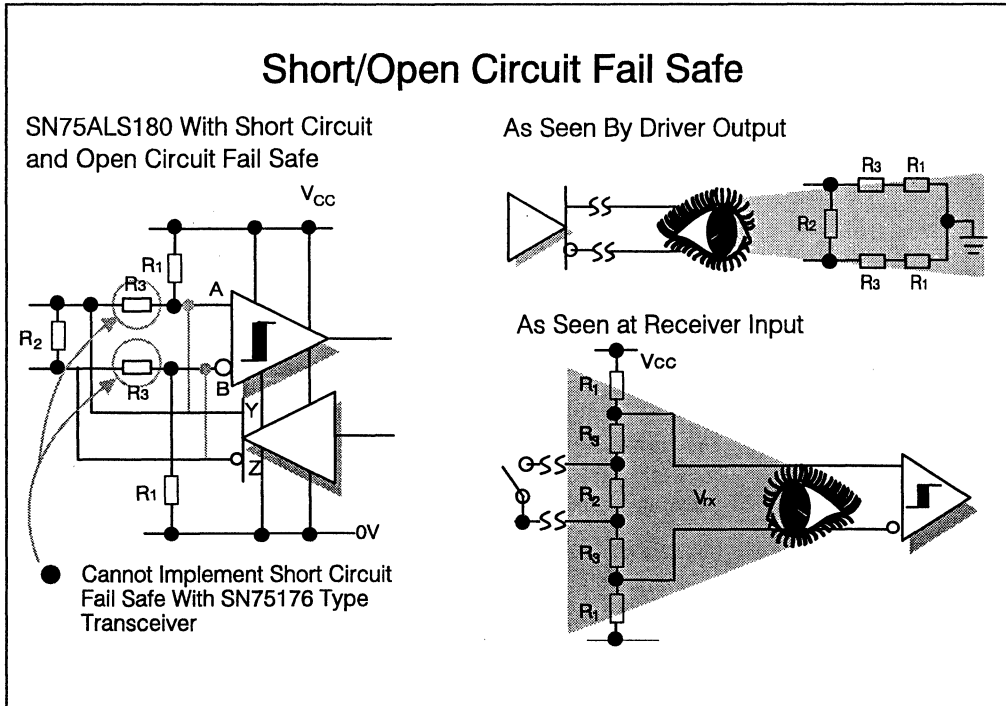
A hard wired fail safe should provide a defined voltage across the receiver's input regardless of whether the line is shorted to either supply rail or is left open circuited. The fail safe should also be incorporated into the line termination if present when at the extremes of the line.

### Internal Fail safe

Manufacturers have gone part way to facilitating fail safe design by including some form of open line fail safe circuitry within the integrated circuits. Unfortunately, due to power consumption constraints, the extra circuitry has proved of little use. The extra circuitry is quite often just a large pull-up resistor on the non-inverting receiver input, and a large pull-down resistor on the inverting

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input of the receiver. These resistors are normally in the range of 100 kΩ, and so when used in conjunction with line termination resistors to form a potential divider, only a few millivolts are generated. As a result this voltage (receiver threshold voltage) is insufficient to switch the receiver. To use these internal resistors effectively means no line termination resistors can be used, which reduces the allowed reliable data rate enormously.



*Figure 5.29 - Short/Open Circuit Fail Safe*

### External Fail safe-Open Line Conditions

A more reliable way of offering open line fail safe is to use external pull-up and pull-down resistors. There two basic ways of doing this; one way is to polarise the line with the pull-up/pull-down resistors and use these resistors to match the line impedance. Another way is to use larger polarising resistors while using an extra resistor to terminate the line. The first idea has one advantage in that it provides a low impedance path to an ac. ground, so that any currents induced on to the line have a low impedance path to ground. However a problem is encountered with this method because the driver output now has to drive very much lower impedance's. If the driver output current capability is poor the device could easily go into output short circuit current limit. The second way, although requiring an extra resistor will not load the driver's output to such an excess.

Placing external pull-up and pull-down resistors  $R_1$  on the non-inverting and inverting inputs of the receiver will produce open circuit fail safe. Terminating the transmission line with its characteristic

impedance,  $Z_0$ , produces a potential divider between  $2R_1$  and  $Z_0$ . The voltage formed across the line,  $V_{oc}$ , equals

$$V_{oc} = V_{cc} \times \frac{Z_0}{2R_1 + Z_0}$$

Devices meeting the RS-485 receiver threshold voltage specifications require  $V_{oc}$  to be greater than 200 mV. From this the relationship of  $R_1$  to  $Z_0$  can be derived:-

$$R_1 = Z_0 \times \frac{1}{2} \times \frac{V_{cc} - V_{oc}}{V_{oc}}$$

With  $V_{cc} = 5V$ ,  $V_{oc} = 200$  mV and  $Z_0 = 100 \Omega$ , yields  $R_1 = 1.2$  k $\Omega$ .

Biasing the receiver in this way will only provide open line fail safe, it will not provide shorted line fail safe. However, when using transceivers, like the SN75ALS176, it is not possible to provide shorted line fail safe configurations, since the driver and receiver share the same IC. pins. Hence for devices like the SN75ALS176 this open line configuration is the optimum fail safe available.

### External Fail safe-Shorted Line Conditions

To implement protection from the shorted line condition, further resistors are required. When the line is shorted the transmission line's impedance goes to zero and the termination resistors will also be shorted. Putting extra resistors in series with the input to the receiver can provide shorted line fail safe protection.

The extra resistors,  $R_3$  in Figure 5.29, can only be added when using devices with separate driver outputs and receiver inputs. So internally wired transceivers cannot be used to offer shorted line fail safe. If this form of protection is required then a device such as the SN75ALS180, with its separate driver outputs and receiver inputs, should be used. If a transceiver type device was used then the extra resistors  $R_3$  would cause extra attenuation of the output signal. The 'ALS180 will have its driver outputs fed directly to the line, bypassing resistors  $R_3$ .

### Calculating the Resistor Values

If the line became shorted then  $R_2$  would be removed leaving a voltage across the receiver inputs of:-

$$V_{rx} = V_{cc} \times R_3 / (R_1 + R_3) \quad (a).$$

For RS-485 applications the standard specifies  $V_{rx}$  to be greater than 200 mV. So

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$$V_{RX} = V_{th} = 200 \text{ mV.}$$

Using this figure, along with the minimum permissible supply voltage for the devices gives a relationship between  $R_1$  and  $R_3$ .

When the line goes into a high impedance state the receiver will see the two  $R_3$  in series with  $R_2$  plus the two  $R_1$ 's pulling up and down on either input. The receiver input voltage will now be:

$$V_{RX} = V_{CC} \times (R_2 + 2R_3)/(2R_1 + R_2 + 2R_3) \quad \text{(b).}$$

Relating this new  $V_{RX}$  to the minimum specified in the standard,  $V_{th}$ , gives:

$$\begin{aligned} R_1 &= \frac{1}{2} R_2 \times \left[ \frac{(V_{CC} - \alpha V_{th})(V_{CC} - V_{th})}{(\alpha - 1) V_{th} V_{CC}} \right] \\ R_3 &= R_2 \times (V_{CC} - V_{th})/V_{th} \end{aligned}$$

Where  $\alpha$  is the ratio of  $V_{CC}/V_{RX}$

The transmission line will see an effective line termination resistance of  $R_2$  in parallel with twice the sum of  $R_1$  and  $R_3$ . This should match the transmission line's characteristic impedance,  $Z_0$ , therefore

$$Z_0 = 2R_2 \times \frac{R_1 + R_3}{2R_1 + R_2 + 2R_3} \quad \text{(c)}$$

Combining equations (a), (b) and (c) yields the following equations for  $R_1$ ,  $R_2$  and  $R_3$ :-

$$R_1 = \frac{1}{2} Z_0 \times \frac{(V_{CC} - V_{th})^2}{(\alpha - 1) V_{th} V_{CC}}$$

$$R_2 = Z_0 \times \frac{V_{CC} - V_{th}}{V_{CC} - \alpha V_{th}}$$

$$R_3 = \frac{1}{2} Z_0 \times \frac{V_{CC} - V_{th}}{(\alpha - 1) V_{CC}}$$

In this application assuming the supply voltage is 4.5 V and  $V_{th} = 200$  mV with an a value  $\alpha$  of 1.5 and driving a line with characteristic impedance of 120  $\Omega$  yields the following values:-

$$R_1 = 2.2k \Omega$$

$$R_2 = 120 \Omega$$

$$R_3 = 110 \Omega$$

The values of  $R_1$ ,  $R_2$ , and  $R_3$  only apply for receivers at the extreme of the line; if there are more receivers on the line then fail safe can be accomplished by multiplying the values of  $R_1$  and  $R_3$  by half of the number of receivers on the line. This is done by assuming the input stages of all the receivers are the same, all  $R_1$  resistors are the same, and that all  $R_3$  resistors are the same. Since all of  $R_1$  and all of  $R_3$  resistors will be in parallel, their overall resistance will be divided by half the number of receivers. If there is a large number of receivers on the line there is a danger of  $R_3$  becoming too large and forming a large potential divider with the input resistance of the receiver, normally around 12 k $\Omega$ .

### 3.3.4. Galvanic Isolation

In the previous sections the need for line termination, receiver fail safe and noise protection was highlighted. All these elements can be found in an industrial process control and data collection application, which is shown in Figure 5.30.

The capability of meeting toughened noise legislation is a key requirement for many new end products and applications. Computer and industrial serial interfacing are areas where noise can seriously affect the integrity of data transfer, and a proven route to improved noise performance for any interface system is galvanic isolation.

Such isolation in data communication systems is achieved without direct galvanic connection or wires between drivers and receivers. Magnetic linkage from transformers provide the power for the system, and optical linkage provides the data connection. Galvanic isolation removes the ground loop currents from data lines and hence the impressed noise voltages which affect the signal are also eliminated. Common mode noise effects can be completely removed and many forms of radiated noise can be reduced to negligible limits using this technique.

For example consider the case in a process control system where the interface node, shown in Figure 5.30, connects between a data logger and host computer via the RS-485 link. When an adjacent electric motor is started up, a momentary difference in ground potentials at the data logger and the computer may occur due to a surge in current. If no isolation scheme is employed for the data communication path, data may be lost during the surge interval and in the worst case damage to the computer could occur.

#### Circuit Description

The schematic shown forms an interface, one node, for a "distributed controlling, regulation and supervision (DCRS) system". Such a scheme could be used in a process control type application. Transmission takes place via a 2-wire bus, formed by a twisted-pair, shielded cable connected in a ring circuit.

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capability. Low power is crucial in this type of application since many remote outstations will either be battery operated or require battery back-up capability.

Transceiver protection circuitry is formed by  $Z_1, Z_2, Z_3$  and  $Z_4$  along with current limiters  $PTC_1$  and  $PTC_2$  (see previous example). Line termination is formed by a combination of  $R_T, R_1$  and  $R_2$ . The values of which can be calculated as follows;

$$R_1 = R_2 < 0.5 \times Z_0 \times [1 + V_{CC}/V_{TH}]$$

and

$$R_T = Z_0 [1 + V_{TH}/V_{CC}]$$

The bus driver used is the **SN75LBC176**, chosen for its low power consumption and high data rate. Using a cable with a characteristic impedance of  $Z_0 = 120 \Omega$  and a desired  $V_{TH}$  of 200 mV, requires  $R_1 = R_2$  to be around 1.6 k $\Omega$  in value. The terminating resistor,  $R_T$  would be in the order of 124  $\Omega$ .

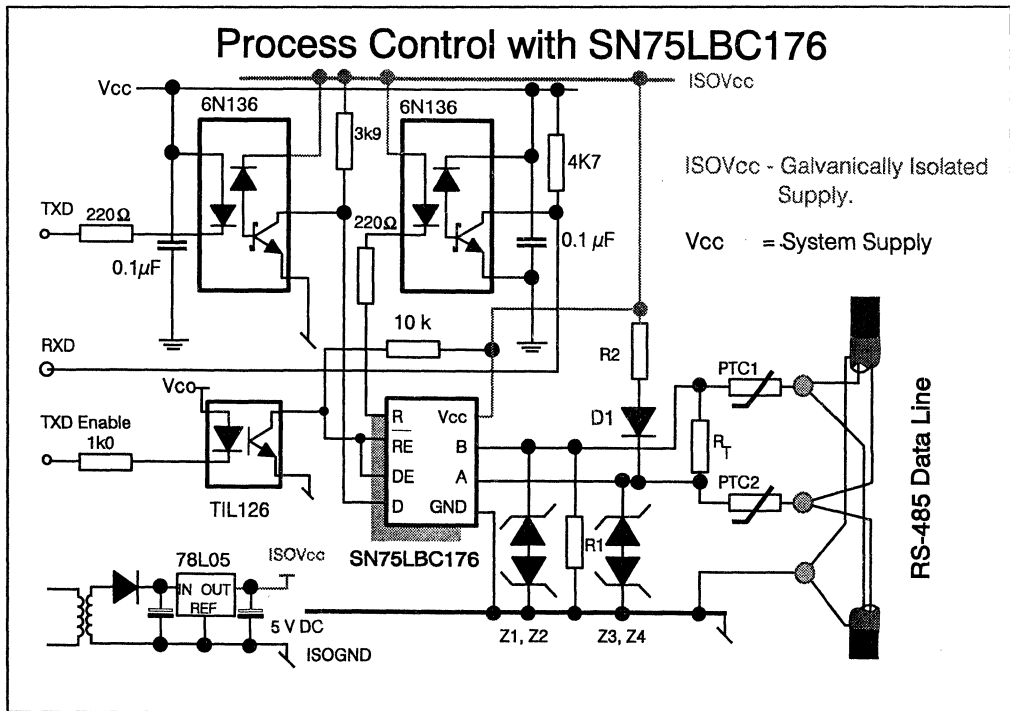


Figure 5.30 Process Control SN75LBC176

The inclusion of  $R_1 = R_2$ , provides a receiver fail safe to open line conditions by biasing the polarity of the line to a logic '1' under line idle conditions. The values of  $R_1 = R_2$  are best kept as low as



possible to increase the noise rejection when the line is left floating, but they will place some loading onto the driver.

Galvanic isolation is afforded by means of three optocouplers/opto isolators. The 6N136 is chosen for its high data rate capability,  $t_p = 75$  ns (max), and its high voltage isolation.

The 6N136 is designed for use in high speed digital interfacing applications that require high voltage isolation between the input and output. Its use is highly recommended in extremely high ground noise and induced noise environments.

The 6N136 consists of a GaAsP light emitting diode and integrated light detector, composed of a photo diode, a high gain amplifier and a Shottky clamped open collector output transistor. An input diode forward current of 5mA will switch the output transistor low, providing an on state drive current of 13 mA (eight 1.6 mA TTL loads). A TTL input is provided for applications that require output transistor gating.

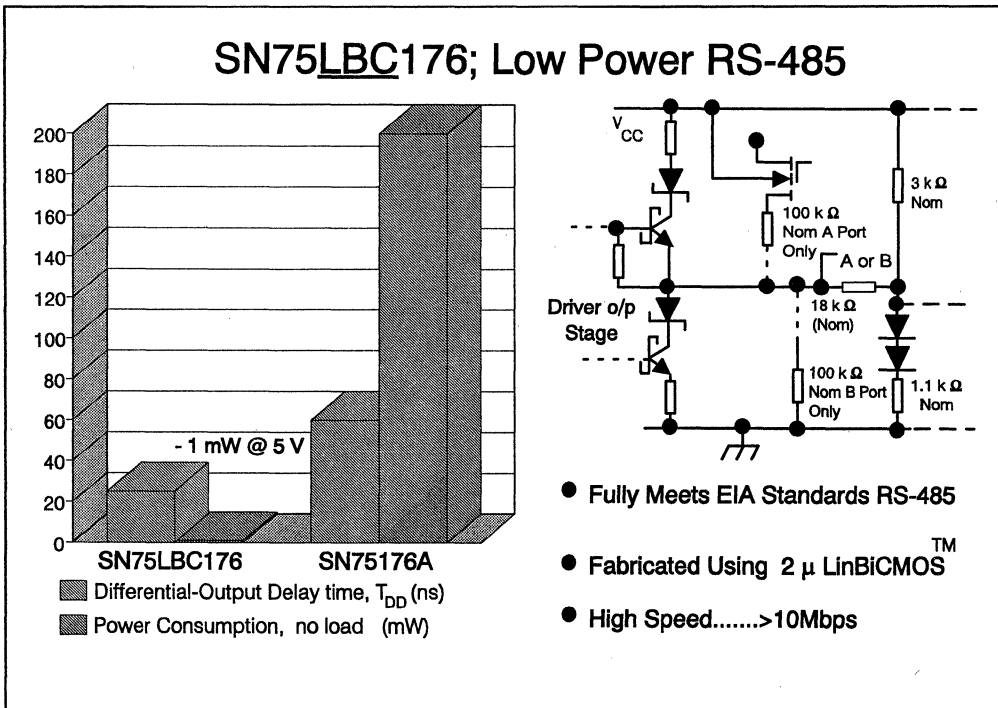


Figure 5.31 - SN75LBC176; Ultra Low Power

Housed in a single 8-pin dual-in-line plastic package the 6N136 is characterised for operation over the temperature range of 0°C to 70°C. The internal Faraday shield provides a guaranteed common mode transient immunity of 1000 V $\mu$ s.

A 0.1  $\mu$ F capacitor has been connected between  $V_{CC}$  and ground to improve switching performance.

### 3.4.SN75LBC176; Ultra Low Power

The key feature of the SN75LBC176 is the very low power consumption, 1 mW. Compare this to the consumption of the older generation SN75176A, where the quiescent power is as high as 200 mW. Normally low power consumption signifies a reduction in ac performance. In the case of the LBC176 the ac performance is improved over the SN75176A. Data rates of greater than 10 Mbps are achievable while still being conformant to RS-485. The low power consumption has further benefits than simply reducing supply current:

#### 3.4.1. Improve MTBF

Although not representing the most glamorous end of the semiconductor design spectrum, reliable line interface circuits are crucial if the overall system mean time between failure (MTBF) is to be minimised. System designers have long been aware that often the weak link in ensuing system reliability has been line interface circuits. This vulnerability is due in part to the circuits close proximity to the outside world via the edge connector. Consequently interface circuits are particularly susceptible to failure from high external voltages caused by noise, ESD or incorrect insertion of cables. For this reason the technology of choice for many Texas Instruments emerging interface products is LinBiCMOS.

#### LinBiCMOS, the Technology of Choice

LinBiCMOS is based on TI's highly successful LinCMOS process. LinCMOS is a 3  $\mu\text{m}$  pure CMOS technology with 16 V capability, making it ideal for the design of low power analog products such as op-amps and analog-to-digital converters (many examples of which are discussed elsewhere in this book). By shrinking the geometry to 2  $\mu\text{m}$  and adding a high performance 30 V bipolar structure, a new analog merged bipolar/CMOS technology has been produced.

Probably LinBiCMOS's greatest attribute is its modularity. When generating a new technology it is difficult to achieve a balance between performance and cost, as many "nice to have" features can make a process too expensive to address a wide range of opportunities.

By making LinBiCMOS modular, only the process modules needed to address a particular application need be used, making it very cost effective. Modules available for LinBiCMOS

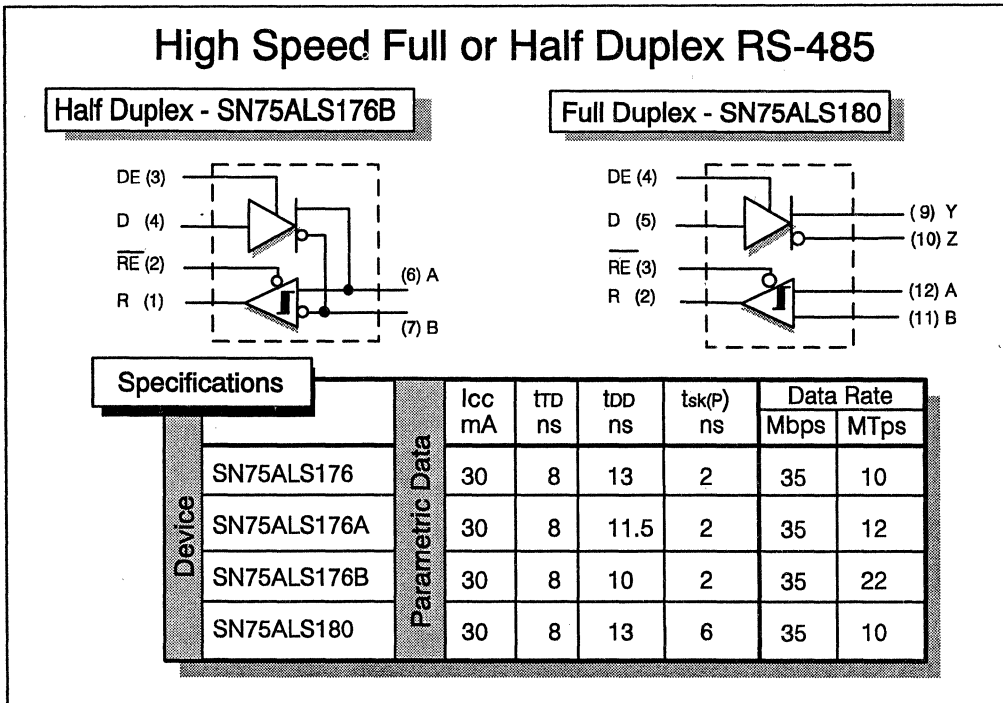
include high speed NPNs (with an  $f_T$  of 3GHz compared with 500MHz for the standard transistor), double level metal for better logic integration and current handling, isolated high value polysilicon resistors and shottky diodes for clamping.

#### The Applications

With its high voltage capability and excellent switching speed LinBiCMOS is ideal to address standards such as EIA-232 and RS-485. For example the RS-485 standard, demands that driver outputs can be shorted to +12V and -7V without damage. This is particularly difficult to implement as RS-485 devices are designed to operate from a single 5 V supply, meaning that parts of the chip must be designed to operate well outside its supply rails. Further more the "party line" nature of the standard requires devices that must be able to withstand contention (multiple drivers accessing the bus simultaneous) without failure. For this reason short circuit protection and thermal shutdown are built into the chip.

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The standard SN75LBC176 is characterised for commercial temperature range applications. For more extreme conditions the SN65LBC176 extends the operating range to -40°C to +85°C.



*Figure 5.32 - High Speed Full or Half Duplex RS-485*

### 3.4.2. Full Duplex and High Speed RS-485

In the process control example we discussed earlier, the key requirement has been for reliable data transmission over 500 metres of cable. The data rate under consideration was limited to 500 kbps however in many systems the data rate requirements are in excess of 10 Mbps. Cable length must be reduced accordingly as transmissions at this rate are only possible over short line lengths, <50 metres. We have discussed the SN75LBC176, however for speeds above 10 Mbps we must look towards a technology which is more well suited for high speed applications.

Advanced Low Power Shottky technology provides the key to the quest for speed. The SN75ALS176 utilises the true benefits of ALS pushing the data rate to over 35 Mbps. For synchronous systems where skew limits are critical, the SN75ALS176B facilitates 22 MTps (Million Transfers per second). This is the highest speed RS-485 transceiver on the market today.

The SN75ALS176, like the SN75176, DS3695, TL3695, is of a transceiver configuration. As shown in figure. 4.32, the transmit and receive pins are accessible through pins 6 and 7 only. With this configuration communication is limited to half duplex only. For quasi-full duplex operation, using

two separate RS-485 lines, one must use a device like the SN75ALS180. this device allows separate lines to be connected to both the receiver and driver. The 'ALS180 has similar performance to the 'ALS176 facilitating up to 35 Mbps.

### **3.4.3. RS-485 Selection Guide**

In this section we have concentrated on the single transceiver type function. TI does however support a wide range of parts compliant with the RS-485 standard in various configurations and multiples. The below selection guide is provided for convenience. The reader is advised to consult the Interface Circuits Data Book (Reference SLYD006) for the latest technical specifications.

**RS-485 Data Transmission Circuits**

Device	Devices	Device	Key
<b>Line Drivers</b>	4	SN75172	Industry Standard
		SN75174	Industry Standard
		SN75ALS172	High Speed
		SN75ALS174	High Speed
<b>Line Receivers</b>	4	SN75173	Industry Standard
		SN75175	Industry Standard
		SN75ALS173	High Speed
		SN75ALS175	High Speed

<b>Line Transceivers (Drivers / Receivers)</b>	1	SN75176A	Reduced Slew-rate
		SN75176B	Industry Standard
		SN75177B	Industry Standard Repeater
		SN75178B	Industry Standard Repeater
		SN75ALS176	High Speed
		SN75ALS176A	Very High Speed
		SN75ALS176B	Ultra High Speed
		SN75LBC176	Ultra- Low Power
	1/1	SN75179B	Industry Standard
		SN75ALS180	Full Duplex Communication
		SN75ALS181	Full Duplex Communication
	2/2	SN751177	High Speed
		SN751178	High Speed
		SN75ALS1177	High Speed
		SN75ALS1178	High Speed
	3	SN75ALS170	High Speed
		SN75ALS171	High Speed
		SN75ALS1711	High Speed
	9	SN75LBC976	Low Power

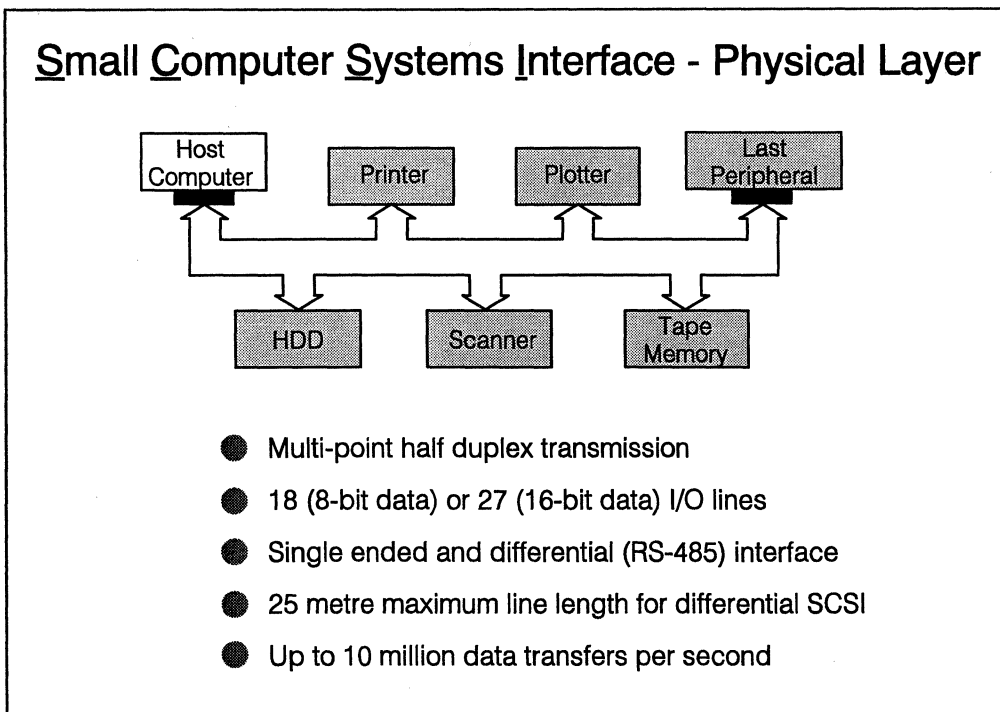
**Notes**

¥ Product currently under development, contact TI representative for further details.



## 4. Interface Circuits for SCSI

### 4.1.SCSI Overview



*Figure 5.33*

SCSI is the acronym for Small Computer Systems Interface and details the ANSI specification for a peripheral bus and command set. The specification defines a high performance peripheral interface that distributes data independently of its host, helping to free up the host for more user oriented

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commands. The objective of the interface is to provide high speed data transfer between computer peripherals independently of the host. Already there are a large number of disk drives, notebook PCs, and CD ROM drives incorporating a SCSI port.

Basic SCSI is an eight bit parallel I/O bus, with parity, and nine control/handshaking lines, making 18 lines in total. More recently, to increase the data throughput, the data bus has been increased to 16 bits with two parity bits while maintaining the nine control lines, making for 27 lines in total. This is referred to as wide SCSI.

### **4.1.1. SCSI Physical Layer**

There are two electrical specifications referred to in the SCSI standard, single ended and differential:

It is beyond the scope of this seminar hand book to discuss the complete SCSI standard, we shall concern ourselves with the physical layer only. For more information on the standard the reader is encouraged to refer to the numerous publications on SCSI.

#### **Single Ended Interface**

The single-ended driver and receiver configuration utilises TTL logic levels and is primarily intended for applications with a cabinet, the maximum line length being limited to 6 metres and the data rate is normally limited to 5 Million Transfers per second (MTps), although innovative termination techniques enable the maximum transfer rate to approach 10 MTps.

#### **Differential Interface**

The differential driver and receiver configuration uses the EIA RS-485 standard and is primarily concerned with transmitting data between cabinets. Maximum line length is 25 metres with the maximum data rate currently 10 MTps.

## **4.2. Single Ended SCSI**

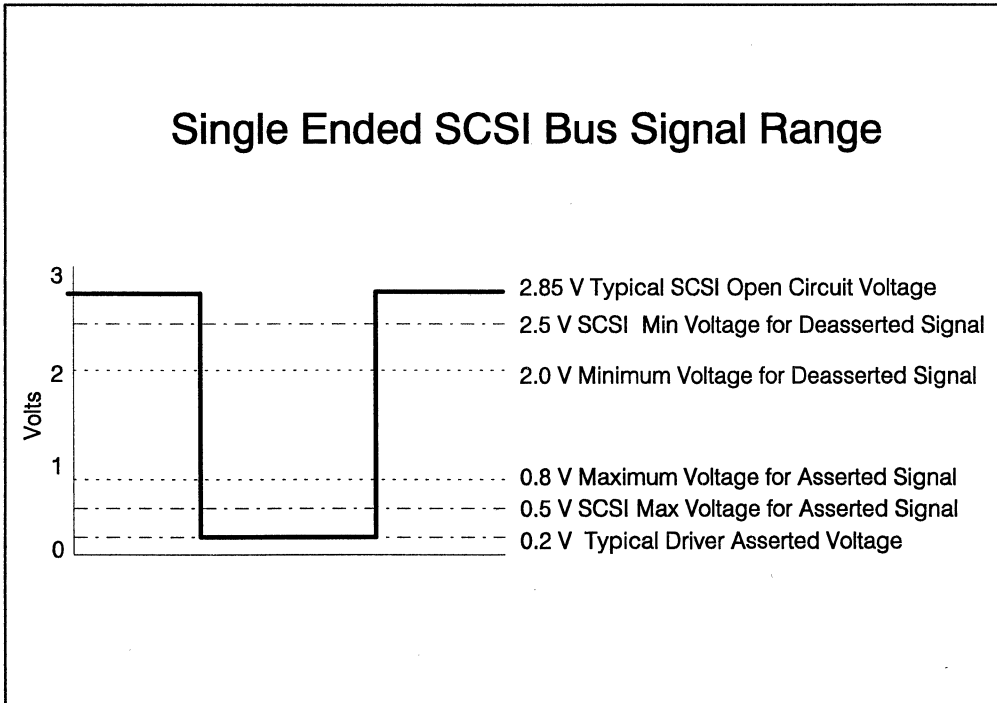
### **4.2.1. Termination of Single Ended Bus**

Termination of the single ended SCSI bus is becoming increasingly important as designers strive for faster system speeds. If error-free data rates of 10 Mbps over the 6 metre bus are to be achieved then signal integrity must be preserved. Termination will reduce unfavourable transmission line effects such as reflections and distortion which can degrade system performance as signal speeds increase. Proper termination of a bi-directional bus such as SCSI requires terminators at each end of the cable.

### **4.2.2. Signal Transitions**

The potential for high data rates depends upon quick and clean transitions between low and high signal levels. The range of SCSI signals is shown below in fig 4.33.1. A low to high transition, or de-assertion, of a signal is initiated by an open collector driver switching off and causing an instantaneous voltage step to travel down the line.





**Figure 5.33.1 - SCSI Bus Signal Range**

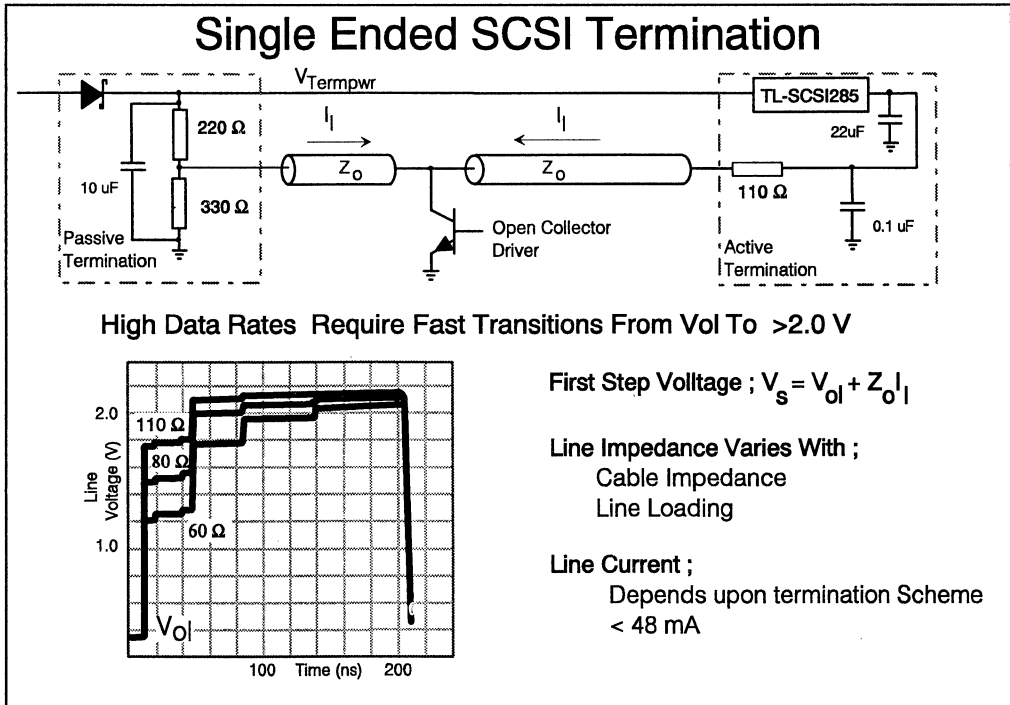
The size of the first step on de-assertion depends upon the amount of current in the line,  $I_L$ , the characteristic line impedance seen by the signal,  $Z_O$ , and the driver output-low voltage ( $V_{OL}$ ), and can be calculated as follows ;

$$V_S = V_{OL} + Z_O I_L$$

To achieve the maximum data rate the first step needs to exceed the receiver threshold voltage in a single transition. Although the  $110 \Omega$  impedance of a typical SCSI ribbon cable suggests that this will require only limited line current capability, the impedance seen by a signal is always less than the specified cable value. Extra capacitance due to peripheral connections to the bus, transmission line effects, and the position of the signal source, can all combine to reduce the effective impedance by 50% or more.

To make up for this lack of 'real' impedance, it is the terminators job to source as much current as possible during de-assertion. This role is restricted by the SCSI specification, however, which limits each terminator to supplying a maximum of 24 mA to prevent the line current from exceeding the 48 mA current sink limit of the open collector drivers.

The role of the SCSI terminator is not confined to low-to-high signal transitions. Once a signal has been de-asserted the terminator is required to bias the bus lines to the correct open circuit voltage level and thereby provide maximum noise margins.



*Figure 5.34 - Single Ended SCSI Termination*

### 4.2.3. Passive and Active SCSI Termination

SCSI termination has traditionally been carried out using passive termination networks. As illustrated in Figure 5.34 these consist of 2 resistors per signal line ; a 220  $\Omega$  pull up resistor connected to the termination power source (Tempwr), and a 330  $\Omega$  pull down resistor connected to ground. The Schottky diode is needed by all termination schemes to protect the power source from reverse currents.

This type of termination typically results in a maximum line current of around 17 mA. Assuming the terminator is on a heavily loaded bus, signified by an impedance of approximately 75  $\Omega$ , then the above equation gives a first step value of 1.76 V - well short of the desired 2.0 V level.

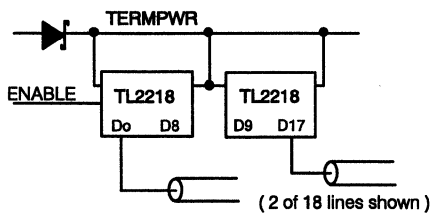
In addition to this limited current capability and the power consumption penalty imposed by the resistor dividers, passive terminators also suffer from an unregulated line bias voltage. As a result the

line voltage will fluctuate with variations in the load current and Term<sub>pwr</sub>, leading to smaller noise margins, lower line currents, and reduced data rates.

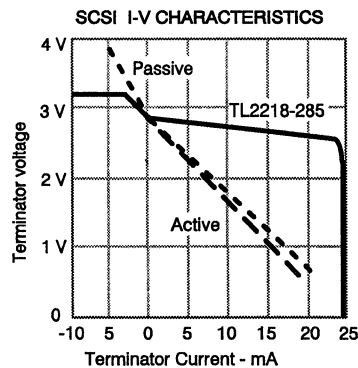
The most common alternative to passive termination replaces the resistive network with a voltage regulator in series with a single 110 Ω resistor per line (fig 4.34). This method, known as Active, Boulay, or Alternative 2 termination, was developed to overcome two of the main shortcomings of passive termination.

## TL2218-285 - Current Source Terminator

TL2218-285 Enables Single Ended Data Rates of 10 MHz



- 23 mA Applied at First High-Level Step
- Very Low Output Capacitance, Typically 6 pF
- No External Components Required
- Compatible with Active Negation
- Thin Shrink Small Outline Package (TSSOP)



### 4.35 TL2218-285 Current Source Termination

The 110 Ω resistors increase the typical line current available on de-assertion to 21 mA, which, from a transmission line viewpoint, is equivalent to a 35% increase in line impedance. The line current and the high-level noise margins are also more stable since Term<sub>pwr</sub> is no longer used to set the bias voltage directly. Instead it is used to form the input to the voltage regulator, which then provides a regulated bias voltage.

The TL-SCSI285 and TL2217-285 low dropout regulators from Texas Instruments are specifically designed for active SCSI termination. With an overall accuracy of 2 % and a maximum dropout voltage of 0.6V the TL-SCSI285 is the highest performance dropout regulator available for SCSI active termination.

#### 4.2.4. Current Source Termination Using the TL2218

Although active termination brings a number of advantages to SCSI termination these can be further improved upon. The TL2218-285 from Texas Instruments is a completely new type of SCSI terminator which does just that.

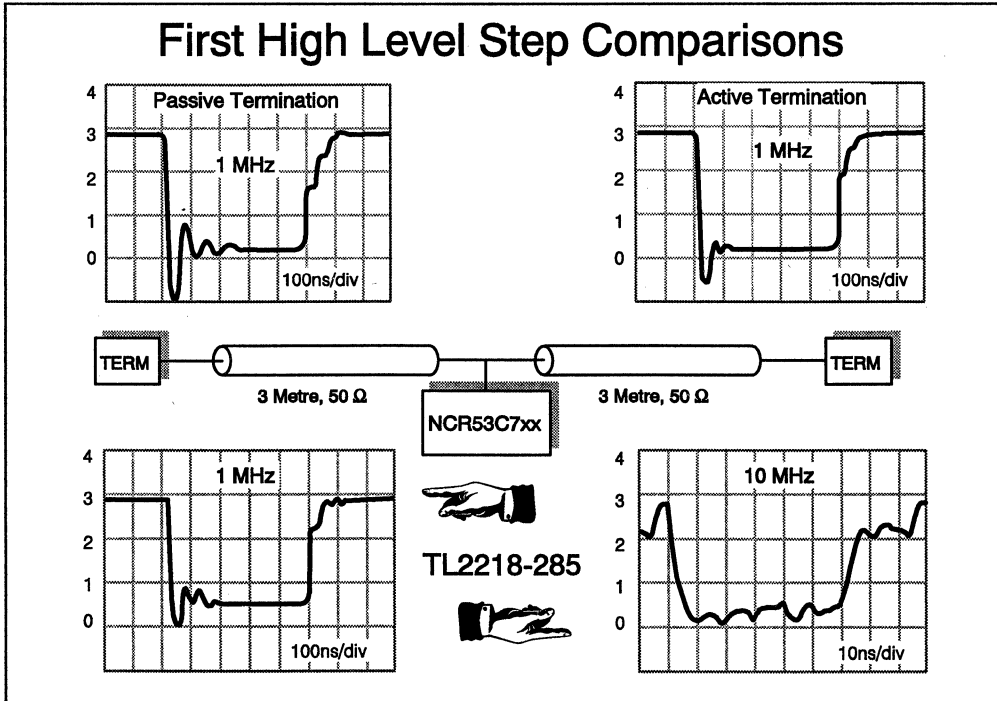
During de-assertion the TL2218-285 operates as a 23.5 mA current source which is able to maintain this current level until the signal reaches the correct SCSI open circuit voltage. At this point the TL2218-285 becomes a voltage source of 2.85 V.

The additional current supplied by the TL2218-285 reduces the low-to-high transition time by ensuring that each voltage step is consistently the largest possible. The effect of this can be seen in Figure 5.36, which shows the signal wave forms obtained after using a TL2218-285, a commercially available active terminator, and a passive termination network to terminate a 50  $\Omega$  cable (the equivalent of a heavily loaded bus). Even at 10 MHz the first step voltage of the TL2218-285 terminated system still exceeds the desired 2.0 V level.

Another feature of the TL2218-285 is the inclusion of a disable function which allows the terminator output to be shut down. This is particularly useful for a peripheral which finds itself somewhere other than the physical end of the bus, and needs some way of easily 'removing' its terminator. If disabled the TL2218-285 consumes just 500  $\mu$ A of current, and maintains an output capacitance of 6 pF. Allowing for the 15 - 16 pF typical output capacitance of a peripherals transceivers, this will give a total node capacitance well within the 25 pF SCSI limit. An active terminator, on the other hand, will normally maintain a disabled output capacitance of 10 pF, often leaving the system to operate outside of the SCSI specification.

Use of the TL2218-285 also removes two possible causes of system failure or driver damage associated with active termination.. Firstly the 2% output tolerance of the TL2218-285 ensures it does not supply more current than allowed by the driver protecting SCSI limit. The tolerances of the voltage regulator and the 110  $\Omega$  resistors used in active termination, however, can result in the terminator supplying in excess of the maximum 24 mA.

The other potentially damaging situation arises when active negation drivers are being used. These devices sense bus voltages and source sufficient additional current to ensure that first step voltages reaches the minimum SCSI level. Despite this attractive feature their relatively high cost has limited their use to ultra fast changing control lines such as ACK and REQ. To be compatible with active negation drivers it is clear that any terminator connected to the bus must be able to sink current. Again this is a problem with active termination but not with either the TL2218-285 or a passive terminator. The voltage regulator of an active terminator will shutdown when any driver voltage exceeds 2.85 V. This allows the line voltage to rise and any driver which then pulls low may sink more than their 48 mA limit.



*Figure 5.36 - First High Level Step Comparisons*

#### 4.2.5. Power Considerations

As well as enabling increased data rates, termination will also increase the power consumption of a SCSI system. As SCSI has found increased usage in portable or battery powered systems this has become more important. Exactly how much depends upon the method of termination, but not quite as obviously as might at first be thought.

During data on periods, the power dissipation of each of the SCSI termination methods is very similar. For an 8 bit bus with all the data lines asserted the power dissipation in each case will be around 1 W.

During data off periods the position is significantly changed. The resistor dividers of a passive terminator will still draw around 750 mW of power. Both the TL2218-285 and active terminators, however, require a total quiescent current of less than 10 mA, providing a 30x saving in power consumption.

For a single TL2218-285 it is possible to calculate a worst case dynamic power dissipation of 493 mW. This assumes that all nine lines in the package are asserted simultaneously and experience a 50% duty cycle. In some systems it may be desirable to avoid any single device dissipating this much power. This can be achieved by partitioning the SCSI bus lines in an appropriate manner. One

such method is to split the data lines between the two devices (for an 8 bit system) and also assign the REQ and ACK lines to separate packages.

Battery powered systems will also benefit from the extended Term<sub>pwr</sub> range of the TL2218-285. Compared to competing solutions which require a minimum Term<sub>pwr</sub> of 4 volts, the 3.5 V to 5.5 V range of the TL2218-285 greatly increases the potential for prolonged operation.

## 4.3. Differential SCSI

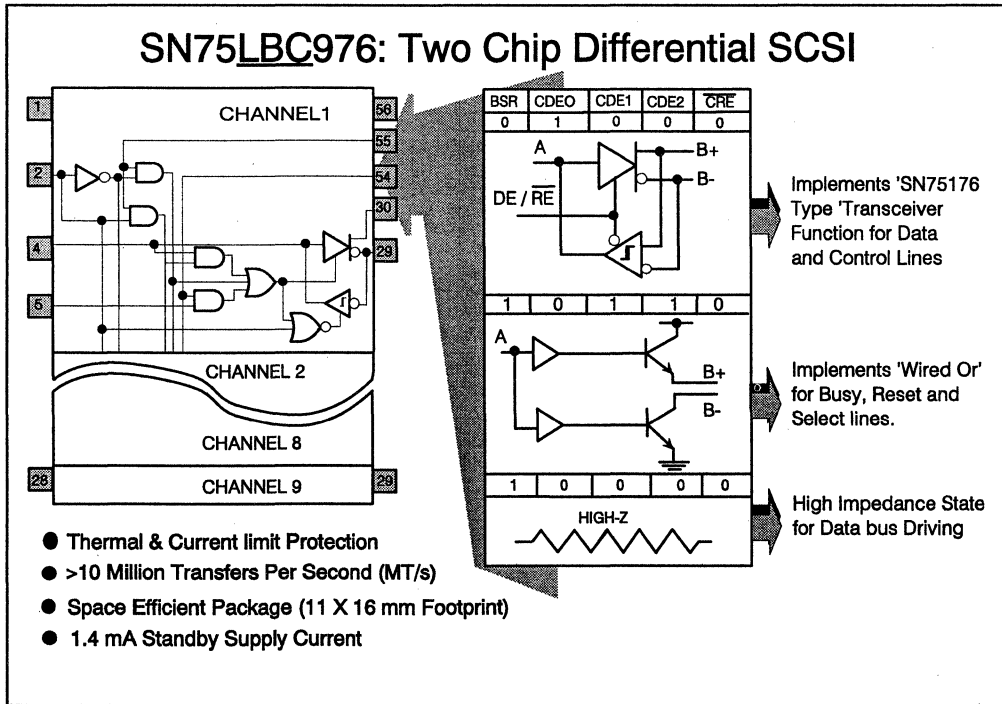
### 4.3.1. SN75LBC976DL; Two Chip Differential SCSI

Much debate has taken place on differential versus single ended SCSI for data rates above 5 million transfers per second (MTps). It is clear however, that for data rates approaching 10 MTps and at line lengths in excess of 6 metres, differential SCSI is essential.

As we discussed earlier, the standard 8-bit interface is made up of 8 data lines, one parity bit, and 9 control lines, making 18 channels in total. The only differential transceivers capable of transmitting at 10 MTps data rate have utilised the LS and ALS technologies. Using these technologies and considering the 18 transceivers per interface the power consumption is quite considerable, 2.4 W with all drivers disabled. Turn the drivers on and the power consumption rises to nearly 4 W.

From a designers viewpoint, 2.4 watts is a considerable amount of heat to remove from a system. This is evident in the case of compact hard disk drives where sheer equipment size is the limiting element. A further factor is board area, using one discrete transceiver per channel, i.e. 18 8-pin SO packages, is unacceptable for many applications.

From a semiconductor designers viewpoint integrating a number of transceivers is of course possible however the limiting factor once again is power dissipation. The SN75LBC976 is designed to overcome both the problems of power dissipation and integration. The device incorporates on a single IC, nine RS-485 configurable transceivers each capable of transmitting at 10 MTps. This is made possible using LinBiCMOS technology. With all drivers disabled the quiescent power consumption of the LBC976 is a mere 1.5 mW, with all drivers enabled the quiescent consumption rises to 45 mW, a considerable saving over LS and ALS parts. The package size has also been reduced to a minimum using the 0.635 mm pitch 56 pin SSOP package which reduces board area significantly compared with alternate packages such as PLCC. The reader should note that irrespective of the device power, there is still the relatively high line current. The SSOP package has been thermally enhanced to handle this level of power dissipation. We will cover this point later as we look at the thermal characteristics of the package.

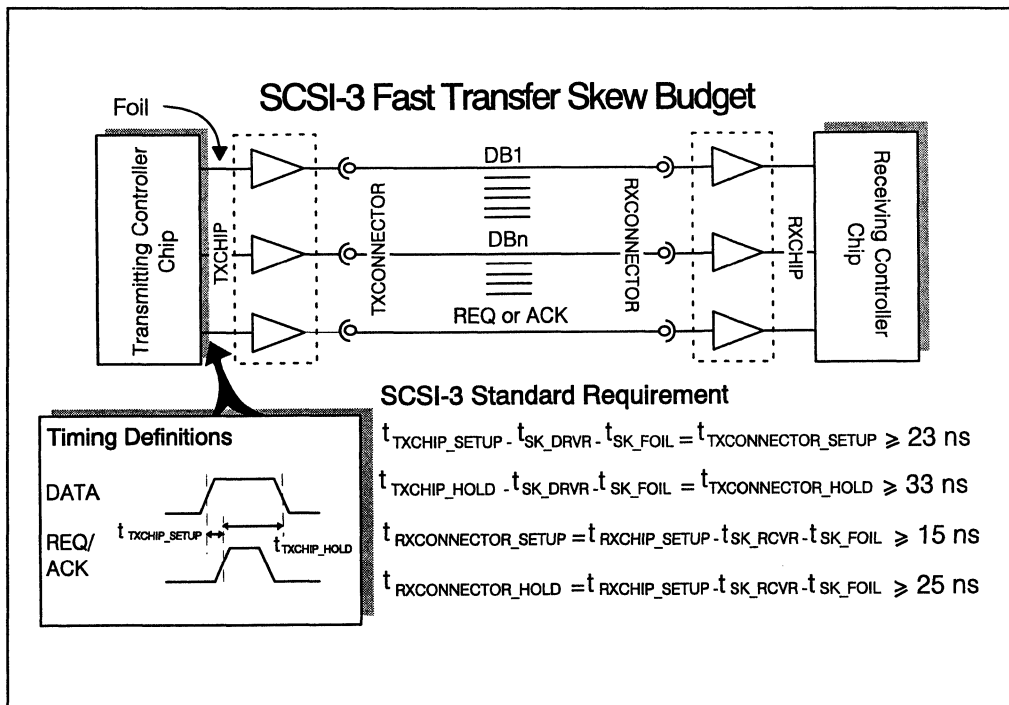


**Figure 5.37 - SN75LBC976: Two Chip Differential SCSI**

The SN75LBC976 is fully configurable to facilitate connection to any type of SCSI system arrangement. The 9 channels can be arranged into seven possible channel functions using the BSR, CDE0, CDE1, CDE2, CRE control pins.

**The 7 channel configurations are:**

1. Transparent, permanently enabled Receiver.
2. Transparent, permanently enabled Driver.
3. Bi-directional transceiver with direction control.
4. Driver with enable control.
5. Open ended driver for Wired OR control lines.
6. Driver with ORed data and enable lines.
7. Permanent high impedance state.



**Figure 5.38 - SCSI-3 Fast Transfer Skew Budget**

### 4.3.2. SCSI and IPI Skew Considerations

SCSI as we have discussed is a parallel data bus. This is also the case with IPI. IPI, an acronym for Intelligent Peripheral Interface, is similar to SCSI in that it is a high speed peripheral bus with the same high speed differential interface requirements. By the parallel nature of the interface, both standards transfer data over the cable more than one bit at a time. SCSI and IPI allow 8-bit (one Byte) or a 16-bit (one word) data width and transfers as often as once every 100 ns or 10 million transfers per second.

Since the logical state of any one bit can change every 100 ns, this defines a period during which the logical state should be valid across the bus. This is the unit Interval (UI). The voltage transitions which define the start and end of the UI can propagate along the bus at different velocities due to the physical differences along each electrical path. So the original UI at the start will be different at the destination.

Time variation of the defining voltage transitions is typically called skew. The limit for skew, designated  $t_{sk(lim)}$ , is the fastest minus the slowest propagation delays along any part of the bus. This, in effect, will reduce the UI by  $t_{sk(lim)}$  establishing a minimum unit interval,  $UI_{min}$ , that can be transmitted with a particular data bus.



The proposed SCSI-3 standard for fast transfers (10 MTps), defines  $UI_{min}$  in terms of set-up and hold times at the SCSI connector for inter-operability with any other SCSI device. At the time this document is being written the requirements are as shown in Figure 5.38.

The budget behind the connector is left to the designer and depends upon the SCSI controller, transceivers, and layout being used. The table shows some skew budget examples with various controller chips that would comply with the requirements at the SCSI connector. The column under 'Rec' (for recommended) is data for the worst case number for SCSI controllers surveyed by the SCSI SPI Working Group and budgets 8 ns for the external driver and 9 ns for the external receiver. This is the origin of the  $t_{sk(lim)}$  specifications in the SN75LBC976 data sheet.

Parameter	Rec	Vendor			Units
		A	B	C	
min Tx_controller_setup =	32	30	35	35	ns
min Tx_controller_hold =	42	42	45	45	ns
min Rx_controller_setup =	5	0	5	0	ns
min Rx_controller_hold =	15	20	15	10	ns
$t_{sk\_etch}$ =	1	1	1	1	ns
max $t_{sk\_dvr}$ =	8	6	11	11	ns
max $t_{sk\_rvc}$ =	9	4	9	14	ns

### *Transceiver Skew Budgets for Various SCSI Controllers*

The time it takes one transceiver of the LBC976 to change logic states is called the propagation delay time. For a driver this is designated as  $t_{dD}$  and for a receiver  $t_{pd}$  and does not differentiate whether the logical transition is from high-to-low or low-to-high level. The  $t_{sk(lim)}$  parameter for the transceiver is nothing more than the maximum difference of the propagation delay times between any two drivers or any two receivers on any two devices. Compliance to the  $t_{sk(lim)}$  specifications of the data sheet and the recommendation of the SCSI standard is assured by measuring the propagation delay time of each channel of each LBC976 and accepting only those devices within the  $t_{sk(lim)}$  band. To keep the production costs of the LBC976 reasonable, this testing is done at 25°C and at 70°C ambient temperatures at a  $V_{CC}$  of 5 volts.

Admittedly, the die temperatures and supply voltage are all the same (or nearly the same) during TI's production testing and not necessarily the same would be seen in actual use. However the sensitivity of the propagation delay times to these factors is the same and repeatable from device to device. In other words, as long as the operating environment of all of the channels of the SCSI interface is similar, the change in propagation delay times from the data sheet conditions will be the same. This will maintain the  $t_{sk(lim)}$  even though the actual propagation delay times may change.

It is nearly impossible to predict the instantaneous die temperatures of these devices in actual use. Due to the non-deterministic nature of the state of any one channel and the averaging affect of nine channels and of the package thermal time constant, the die temperature must be considered using the mean power dissipation. It is also reasonable to assume the mean power dissipation of separate

devices on the same printed circuit board to be close to each other and the temperature of the air around them will not have a large (<5°C) gradient between the two. Even if there was an air temperature gradient of 45°C, there would be only about a 2 ns difference in the driver propagation delay times and, from recent data, little or no difference in the receiver propagation delay times. If such a temperature gradient actually existed across a board, it' is likely that skew budgets are not going to be the problem with the equipment!

In summary:

1. The designer should determine the transceiver skew requirements based upon his controller and board design.
2. The current specification and testing of  $t_{sk(lim)}$  of the LBC976 is, the best compromise for a cost effective solution.
3. We have application information from our experience gained through numerous users with no inter-operability problems detected at the time of writing.

### 4.3.3. SN75LBC976 Channel Power Dissipation Considerations

#### Channel Power Dissipation

To understand the SN75LBC976 power dissipation when connected to a SCSI bus and the subsequent heat sinking requirements, we must develop a realistic model for the power consumption under working conditions. We must consider the power dissipation within the silicon. There are three primary sources, the dc quiescent power, the ac or switching power, and the dc or resistive losses in the output drivers.

The current necessary to bias the circuits of a single enabled LBC976 differential driver is typically 0.53 mA and a maximum of 1.1 mA. A single enabled receiver circuit requires 3.22 mA typically and 5.00 mA maximum. The typical values have been measured on 94 SN75LBC976DLs from three different wafer lots. The maximums have been verified over temperature on the same samples.

It follows the driver quiescent power consumption,  $P_{DCC}$  is:

$$\begin{aligned}P_{DCC} &= I_{CC} \times V_{CC} \\ &= 0.53 \text{ mA} \times 5.00 \text{ V} = 2.65 \text{ mW/Channel average} \\ &= 1.11 \text{ mA} \times 5.25 \text{ V} = 5.83 \text{ mW/Channel maximum}\end{aligned}$$

And the receiver quiescent power,  $P_{RCC}$  is:

$$\begin{aligned}P_{RCC} &= I_{CC} \times V_{CC} \\ &= 3.22 \text{ mA} \times 5.00 \text{ V} = 16.10 \text{ mW/Channel average} \\ &= 5.00 \text{ mA} \times 5.25 \text{ V} = 26.25 \text{ mW/Channel maximum}\end{aligned}$$

The average  $I_{CC}$  of a representative sampling of the SN75LBC976 has been measured to be 9.77 mA for nine unloaded drivers switching at 5 MHz (10 Mbps), a 50% duty cycle, and at a  $V_{CC}$  of 5 V. Nine receivers at the same frequency and duty cycle and unloaded outputs consumed 36.0 mA average. Since both measurements include  $P_{DCC}$  or  $P_{RCC}$ , they are subtracted below:

---

Driver switching losses,  $P_{DAC}$ , at 5 MHz:

$$\begin{aligned}P_{DAC} + P_{DCC(average)} &= (I_{CC(average)}/9) \times V_{CC} \\P_{DAC} &= (I_{CC(average)}/9) \times V_{CC} - P_{DCC(average)} \\&= (97.7/9) \times 5.0 - 2.65 \\&= 51.6 \text{ mW/Channel}\end{aligned}$$

Receiver switching losses,  $P_{RAC}$ , at 5 MHz:

$$\begin{aligned}P_{RAC} + P_{RCC(average)} &= (I_{CC(average)}/9) \times V_{CC} \\P_{RAC} &= (I_{CC(average)}/9) \times V_{CC} - P_{RCC(average)} \\&= (36.0 \text{ mA}/9) \times 5.0 \text{ V} - 16.10 \\&= 3.9 \text{ mW/Channel}\end{aligned}$$

The output stage losses vary with the magnitude of the output voltages or the output transistor saturation voltages and with the load conditions. The following is based upon the solution of the equivalent circuit of a differential SCSI bus and no further proof is included in this analysis.

For the differential driver, the worst case condition is with a driver asserting the line with SCSI bus termination and a differential output voltage of about 2 V. Under these conditions there would be 144 mW dissipated in the output transistors when asserted and 71 mW when negated. The average output voltage of the SN75LBC976 driver is 2 V. As such, the average power dissipated in the output transistors is also a worst case condition.

Driver output dc losses,  $P_{DOH}$  is given by:

$$P_{DOH} = 144.0 \text{ mW/Channel}$$

The same circuit but with the line negated:

$$P_{DOL} = 71.0 \text{ mW/Channel}$$

The receiver output stage is rated for sinking 8.0 mA at a maximum low-level output voltage of 0.8 V.

Receiver output dc losses,  $P_{RO}$ , is given by:

$$P_{RO} = 8.0 \text{ mA} \times 0.8 \text{ V} = 6.4 \text{ mW/Channel maximum}$$

Since  $P_{DCC} + P_{DAC} = P_{DO} \gg P_{RCC} + P_{RAC} + P_{RO}$ , the worst case power dissipation in a data channel occurs when the driver is enabled and transmitting data. This case will be used to analyse the device power dissipation.

### Device Power Dissipation

Assuming the probability that any one bit on the bus is asserted is equal to the probability of being negated, the state of the output is non-deterministic, and the thermal time constant of the device is long with respect to the data transfer period, the mean die temperature will be determined by the mean power dissipation. In the driver output this is the mean of the asserted and negated values:

Mean device output dc losses:

---

$$\begin{aligned}
 P_{DO(DEV)} &= (P_{DOH} + P_{DOL})/2 \times 9 \text{ Channels} \\
 &= (144 \text{ mW/Ch} + 71 \text{ mW/Ch})/2 \times 9 \text{ Channels} \\
 &= 967.5 \text{ mW}
 \end{aligned}$$

From the assumptions above and the probability that the driver output will change state on the next cycle is equal to the probability that it will not, the mean power dissipated due to driver switching is one-half  $P_{DAC}$  which was measured with the switching loss occurring every cycle.

Mean device switching losses:

$$\begin{aligned}
 P_{DAC(DEV)} &= P_{DAC}/2 \times 9 \text{ Channels} \\
 &= (51.6 \text{ mW/Ch})/2 \times 9 \text{ Channels} \\
 &= 232.2 \text{ mW}
 \end{aligned}$$

Total Device quiescent power:

$$\begin{aligned}
 P_{DCC(DEV)} &= P_{DCC(\text{average})} \times 9 \text{ Channels} \\
 &= 2.65 \text{ mW/Ch} \times 9 \text{ Channels} \\
 &= 23.85 \text{ mW}
 \end{aligned}$$

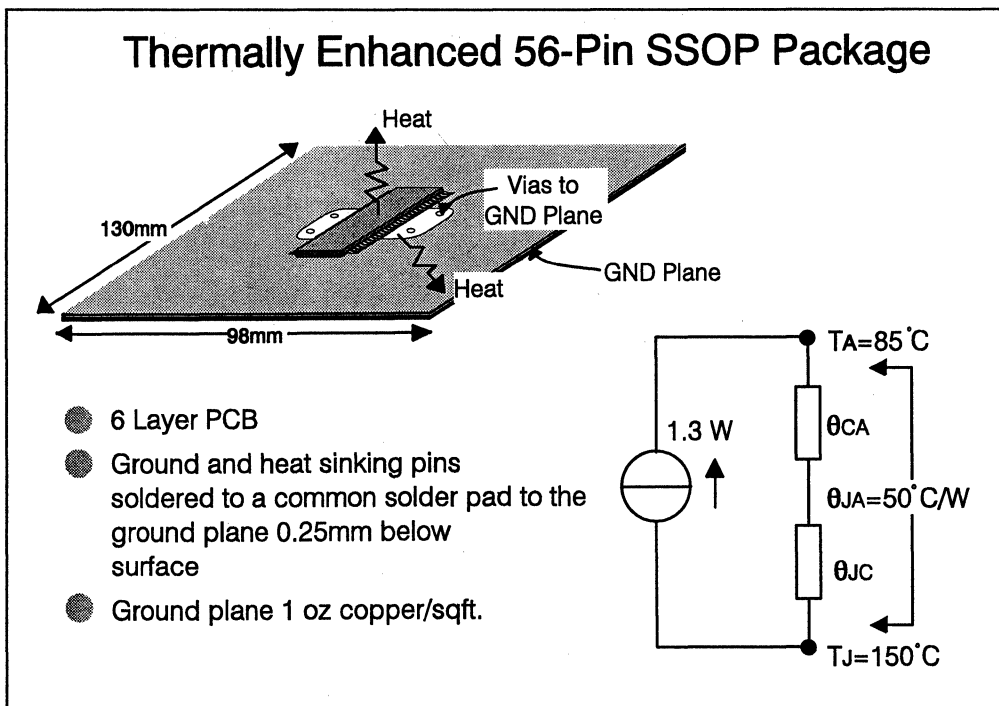
The total mean power dissipated in 9 enabled drivers transmitting over a SCSI bus for several package thermal time constants is then

$$\begin{aligned}
 P_{D(DEV)} &= P_{DO(DEV)} + P_{DAC(DEV)} + P_{DCC(DEV)} \\
 &= 967.5 \text{ mW} + 232.2 \text{ mW} + 23.85 \text{ mW} \\
 &= 1223.6 \text{ mW}
 \end{aligned}$$

#### 4.3.4. Junction Temperature and Layout Considerations

Measurements of the thermally enhanced 56-pin SSOP package and lead frame used on the SN75LBC976DL were performed on a 130 mm by 98 mm six layer printed circuit board with the ground and heat sinking pins soldered to a common solder pad and connected to a second layer ground plane through one via interconnect, see Figure 5.39. The ground plane was 0.254 mm below the surface of the board and was a 1 oz copper layer. The results of two tests resulted in  $\theta_{JA}$ s of 52.9 and 46.6°C/W with zero air flow.

The mean junction temperature rise above ambient when all drivers are enabled and transmitting data over the SCSI bus for several package thermal time constants can then be calculated.



**Figure 5.39 - Thermally Enhanced 56-pin SSOP Package**

Junction temperature rise above ambient:

$$\begin{aligned}
 T_J - T_A &= \theta_{JA} \times P_{D(DEV)} \\
 &= (52.9^\circ\text{C/W} + 46.6^\circ\text{C/W})/2 \times 1233.6\text{mW} \times 1 \text{ W}/1000 \text{ mW} \\
 &= 60.8^\circ\text{C}
 \end{aligned}$$

Most designs require two junction temperature conditions to be met. The junction operating temperature should not exceed 150°C under worst case operating conditions and the average operating junction temperature should be no more than 110°C.

Since the worst case condition of all nine channels transmitting data was used for analysis, the maximum ambient air temperature,  $T_A$ , with no air flow should be:

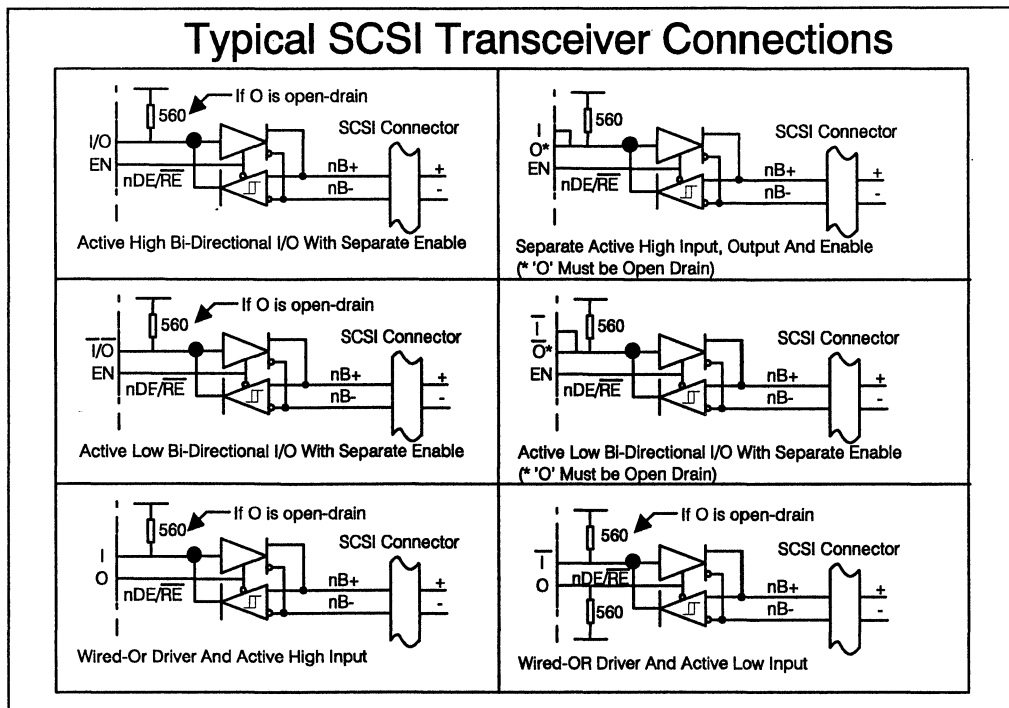
$$\begin{aligned}
 T_{A(\text{maximum})} &= 150^\circ\text{C} - 60.8^\circ\text{C} \\
 &= 89.2^\circ\text{C}
 \end{aligned}$$

When evaluating the average operating junction temperature the effects of transmit/receive duty cycle communication port activity must be taken into account.

## 4.4. Driving the 'Wired-Or' SCSI Lines with the SN75LBC976

The control lines of the SCSI bus have three 'Wired-Or' lines, these are BSY (busy), RST (reset), and SEL (select). These lines are wired-or in that the line drivers connected to these lines drive in one direction (assertion) only and are tri-stated (high impedance) when negated. This allows numerous drivers to be active at the same time without affecting the logic state of the line and requires that all drivers be released or off before the logical state can change. When tri-stated the bus termination network passively negates the signal.

The technique used for wired-or operation with differential transceivers is to input the signal into the driver enable pin and connect the driver input to a fixed logic level input. When the input signal to the driver enable is active (high), the driver becomes enabled and the outputs drive the SCSI bus to the state of the driver input. When the input signal at the driver enable pin goes low, the driver turns off and allows the bus termination to negate the signal on the bus after all other drivers on the bus are also shut off.



**Figure 5.40 - Typical SCSI Transceiver Connections**

Many communications controllers used for differential SCSI have separate inputs and outputs for these signals. When used with the SN75176 type RS-485 transceiver, these controller I/Os can be directly connected to separate driver enable inputs or receiver outputs. The SN75LBC976 device does not have a separate driver input and receiver input, these are tied together internally to save pins.

Controllers with separate I/Os can still be used with the LBC976 using the connections shown in Figure 5.40. The controller output will go high and enable the driver and disable the receiver. Upon disabling the receiver, the external pull-up or pull-down resistor will drive pin A of the LBC976 to the proper level for bus assertion and the driver will assert the SCSI signal line. When the controller output goes low, the driver disables and allows the termination to negate the bus signal. After a short delay, the receiver outputs are enabled and will reflect the logical state of the bus signal.





## 5. Summary and Further Information

### 5.1. EIA Standards

For copies of the EIA standards please contact the Electronic Industries Association. For details and a copy of the Catalog of EIA & Jedec Standards & Engineering Publications contact the EIA Standard Sales office at the following address and telephone number:

**EIA Standard Sales Office**  
2001 Pennsylvania Avenue, N.W.  
Washington, D.C. 20006  
United States  
Telephone Number + 1 (202) 457-4966

### 5.2. References

The following books were invaluable in producing the Section on data transmission:

1. **Digital, Analog, and Data Communication** - William Sinnema & Tom McGovern - Prentice-Hall International - ISBN 0-835-91313-9.
2. **Data Transmission** - D. Tugal and O. Tugal - McGraw Hill - 1980

### 5.3. Texas Instruments - Completing the Picture

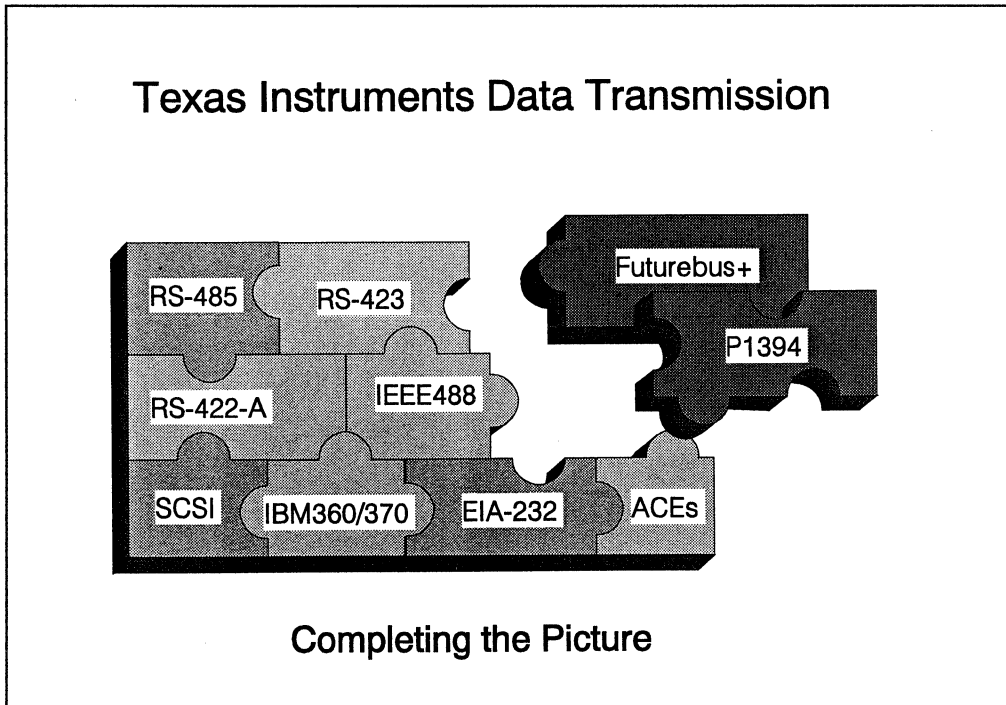
The range of products discussed throughout this section demonstrate the commitment made by Texas Instruments to the field of peripheral interfacing. Use of leadership technologies has enabled the production of high performance, reliable line drive/receive functions and highly integrated controllers.

Obviously in a seminar this short it is impossible to cover all TI's data transmission products. The reader is encouraged to contact a TI representative to obtain the latest data books on transmission ICs. As a minimum the designer should possess a copy of the Interface Circuits Data Book.

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Finally, although we have discussed the older standards such as EIA-232 and RS-485 TI is continuing to work and participate in the various standard definition committees to ensure new and emerging standards gain the full support of TI's semiconductor experience. Testimony to this fact has been the SN75LBC976 RS-485 transceiver, the specifications for which were closely aligned with the outcome of the work done by the SCSI-3 standards group.

Over the next 12 months TI will be releasing products in support of Futurebus+, and is actively supporting the P1394 high speed serial bus Working Group among others.



*Figure 5.41 - Texas Instruments Data Transmission - Completing the Picture*



<b>General Information</b>	<b>1</b>
<b>Line Drivers, Receivers, Transceivers</b>	<b>2</b>
<b>Universal Async Receivers/Transmitters</b>	<b>3</b>
<b>Explanation of Logic Symbols</b>	<b>4</b>
<b>Applications</b>	<b>5</b>
<b>Mechanical Data</b>	<b>6</b>

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**9**

**Mechanical Data**



# ORDERING INSTRUCTIONS

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## Military Data Transmission Nomenclature

Example:           SNJ           55           109A           J

**Prefix** \_\_\_\_\_

SN = Standard  
SNJ = Class B Processing

**Second-Source Prefix**

AM = AMD  
MC = Motorola

**Operating Temperature Range** \_\_\_\_\_

55 = Military  
      -55°C to 125°C  
95 = Nonstandard

**Unique Circuit Description** \_\_\_\_\_

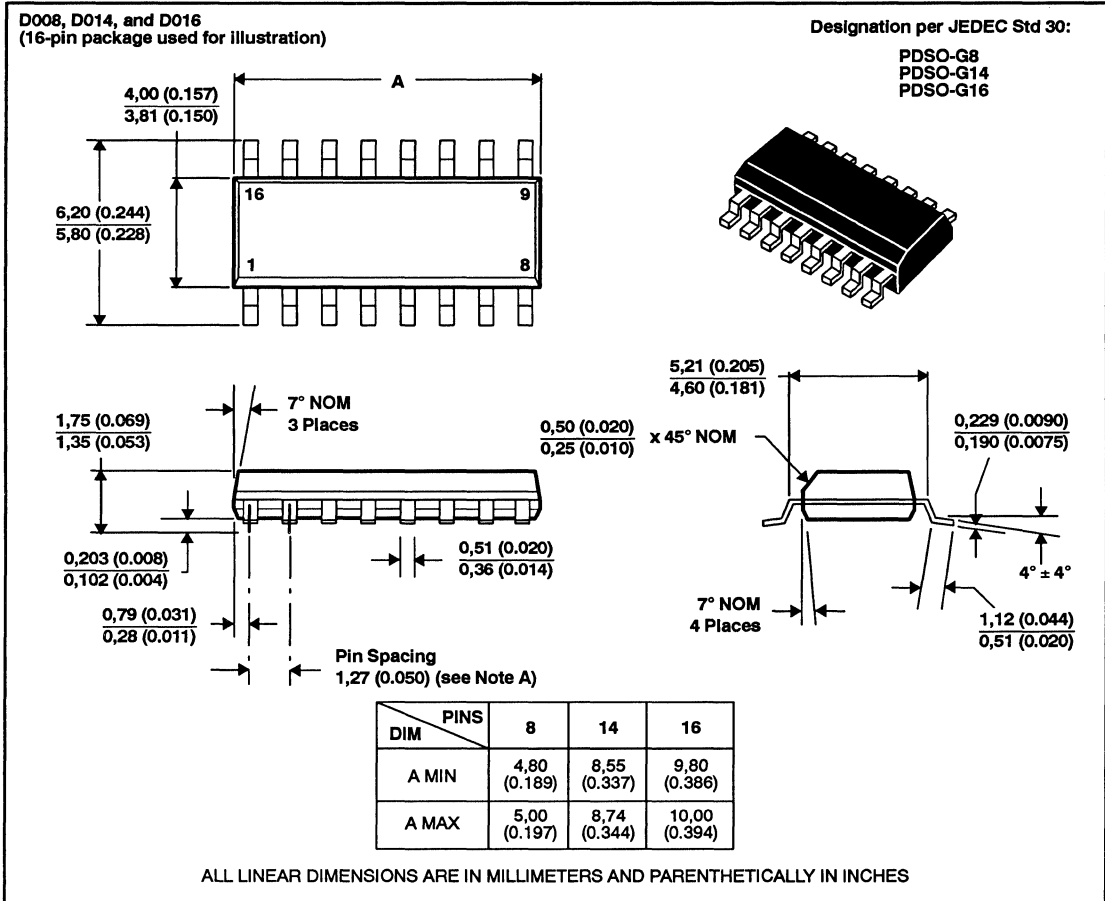
Possibly with A or B in Last Position

**Package Designation** \_\_\_\_\_

FK = LCC  
J = C-DIP  
JG = 8-Pin C-DIP  
W = Flatpack

**D008, D014, and D016**  
**plastic small-outline packages**

Each of these small-outline packages consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



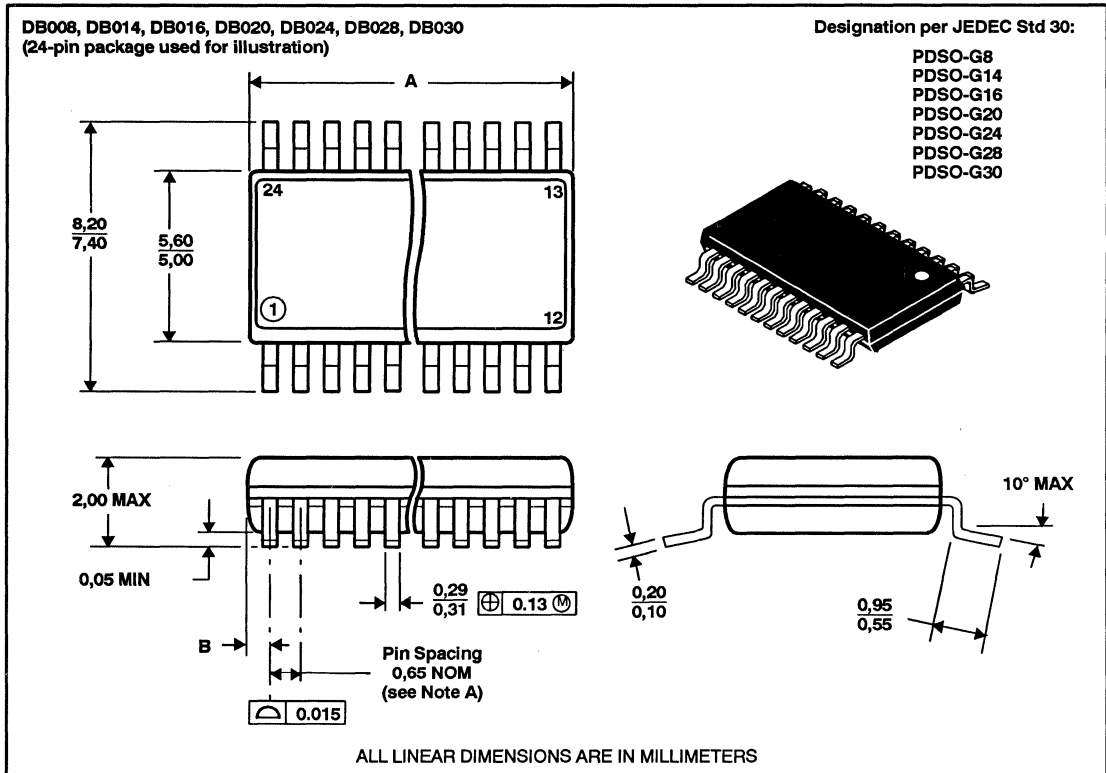
- NOTES: A. Leads are within 0,25 (0.010) radius of true position at maximum material condition.  
 B. Body dimensions do not include mold flash or protrusion.  
 C. Mold flash or protrusion shall not exceed 0,15 (0.006).  
 D. Lead tips to be planar within ±0,051 (0.002) exclusive of solder.



# MECHANICAL DATA

## DB008, DB014, DB016, DB020, DB024, DB028, DB030 plastic shrink small-outline packages

These shrink small-outline packages consist of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.

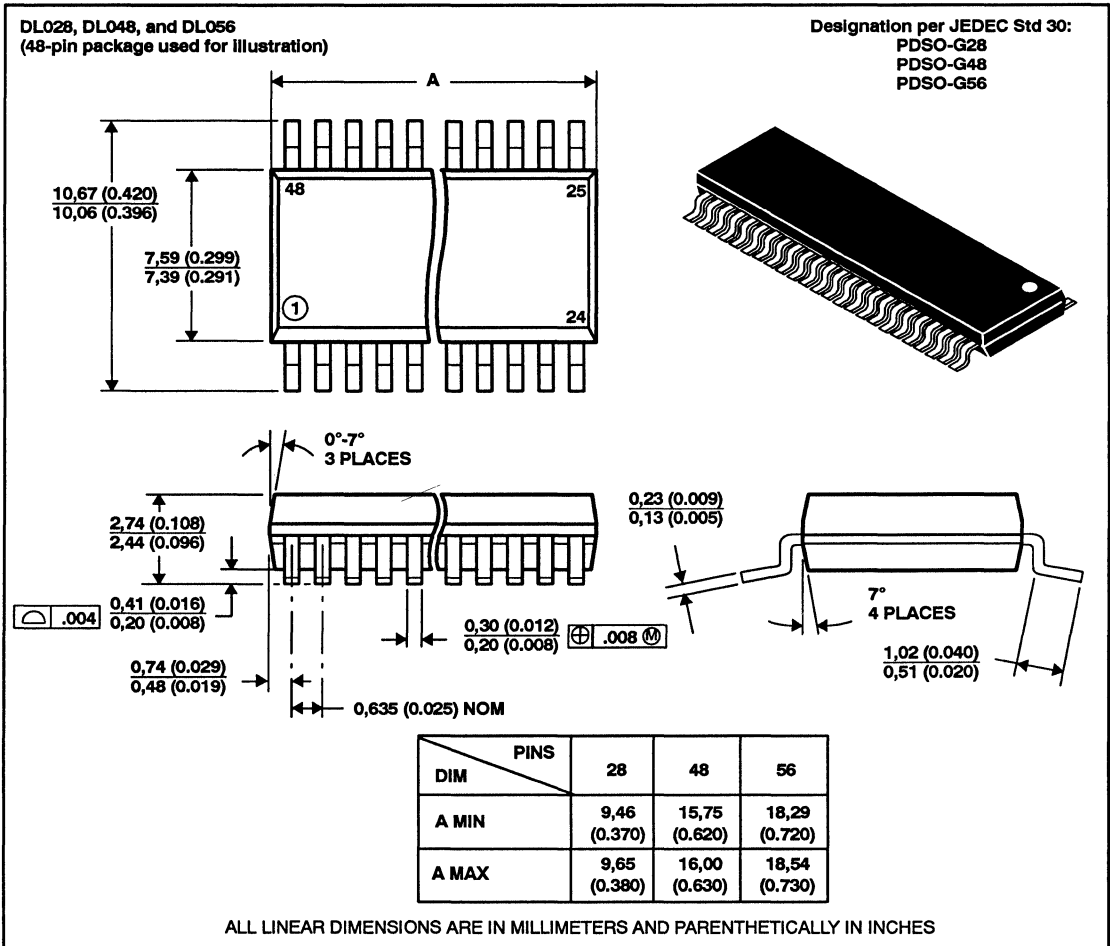


- NOTES: A. Leads are within 0,25 mm radius of true position at maximum material condition.  
 B. Body dimensions do not include mold flash or protrusion.  
 C. Mold or flash end protrusion shall not exceed 0,15 mm.  
 D. Interlead flash shall be controlled by TI statistical process control (additional information available through TI field office).  
 E. Lead tips to be planar within  $\pm 0,05$  mm exclusive of solder.

DIM \ PINS	PINS						
	8	14	16	20	24	28	30
A MIN	2,70	5,90	5,90	6,90	7,90	9,90	9,90
A MAX	3,30	6,50	6,50	7,50	8,50	10,50	10,50
B MAX	0,68	1,30	0,98	0,83	0,68	1,03	0,70

**DL028, DL048, and DL056**  
**plastic shrink small-outline packages**

Each of these shrink small-outline packages consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.

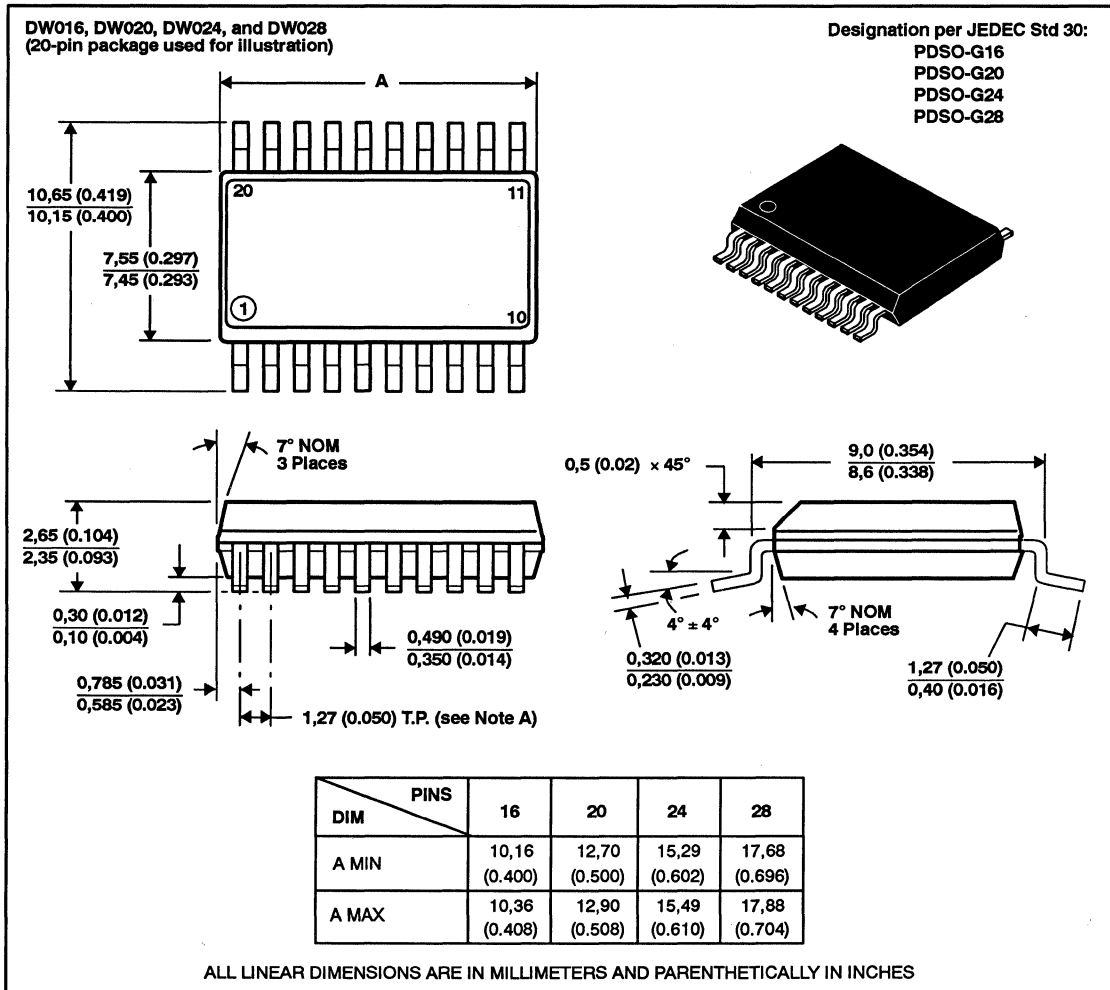


- NOTES: A. Body dimensions do not include mold flash or protrusion.  
 B. End protrusion shall not exceed 0,15 (0.006).  
 C. Interlead flash shall be controlled by TI statistical process control (additional information available through TI field office).  
 D. Lead tips to be planar within ±0,05 (0.002).

# MECHANICAL DATA

## DW016, DW020, DW024, and DW028 plastic small-outline packages

Each of these small-outline packages consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.

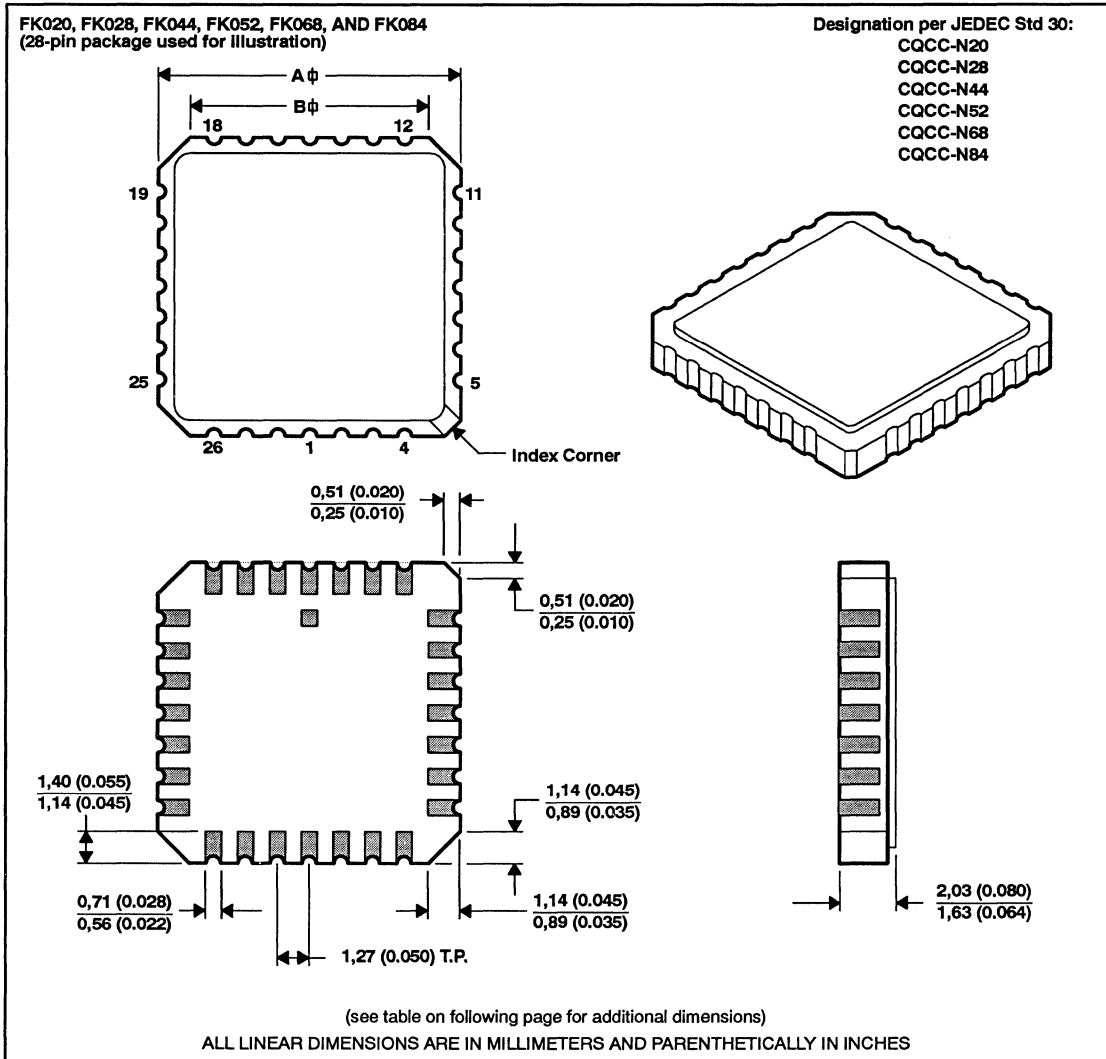


- NOTES: A. Leads are within 0,25 (0.010) radius of true position at maximum material condition.  
 B. Body dimensions do not include mold flash or protrusion.  
 C. Mold flash or protrusion shall not exceed, 0,15 (0.006).  
 D. Lead tips to be planar within  $\pm 0,051$  (0.002) exclusive of solder.

**FK020, FK028, FK044, FK052, FK068, and FK084**  
ceramic chip carrier

Each of these hermetically-sealed chip carrier packages has a three-layer ceramic base with a metal lid and braze seal. These packages are intended for surface mounting on solder leads on 1,27 (0.050) centers. Terminals require no additional cleaning or processing when used in soldered assembly.

FK package terminal assignments conform to JEDEC Standards 1 and 2.



# MECHANICAL DATA

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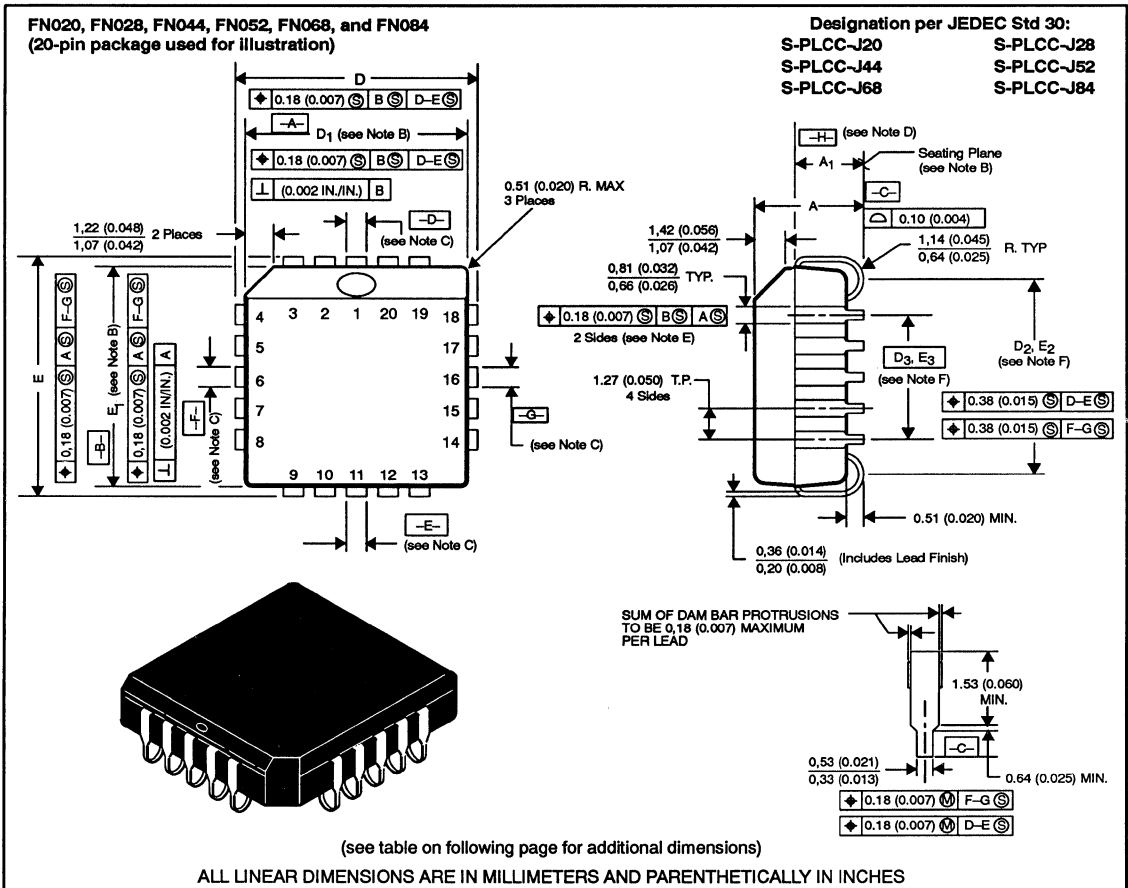
## FK020, FK028, FK044, FK052, FK068, and FK084 ceramic chip carrier (continued)

JEDEC OUTLINE DESIGNATION†	NUMBER OF TERMINALS	A		B	
		MIN	MAX	MIN	MAX
MS-004-CB	20	8,69 (0.342)	9,09 (0.358)	7,80 (0.307)	9,09 (0.358)
MS-004-CC	28	11,23 (0.442)	11,63 (0.458)	10,31 (0.406)	11,63 (0.458)
MS-004-CD	44	16,26 (0.640)	16,76 (0.660)	12,58 (0.495)	14,22 (0.560)
MS-004-CE	52	18,78 (0.740)	19,32 (0.760)	12,58 (0.495)	14,22 (0.560)
MS-004-CF	68	23,83 (0.938)	24,43 (0.962)	21,80 (0.850)	21,80 (0.858)
MS-004-CG	84	28,99 (1,141)	29,59 (1,164)	26,60 (1,047)	27,00 (1,063)

† All dimensions and notes for the specified JEDEC outline apply.

**FN020, FN028, FN044, FN052, FN068, and FN084  
plastic J-led chip carrier**

Each of these chip carrier packages consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound withstands soldering temperatures with no deformation, and circuit performance characteristics remain stable when the devices are operated in high-humidity conditions. The package is intended for surface mounting on 1,27 (0.050) centers. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. All dimensions conform to JEDEC Specification MO-047AA/AF. Dimensions and tolerancing are per ANSI Y14.5M-1982.  
 B. Dimensions  $D_1$  and  $E_1$  do not include mold flash protrusion. Protrusion shall not exceed 0,25 (0.010) on any side. Centerline of center pin each side is within 0,10 (0.004) of package centerline by dimension B. The lead contact points are planar within 0,10 (0.004).  
 C. Datums  $D-E$  and  $F-G$  for center leads are determined at datum  $-H-$ .  
 D. Datum  $-H-$  is located at top of leads where they exit plastic body.  
 E. Location of datums  $-A-$  and  $-B-$  to be determined at datum  $-H-$ .  
 F. Determined at seating plane  $-C-$ .

# MECHANICAL DATA

## FN020, FN028, FN044, FN052, FN068, and FN084 plastic J-leaded chip carrier (continued)

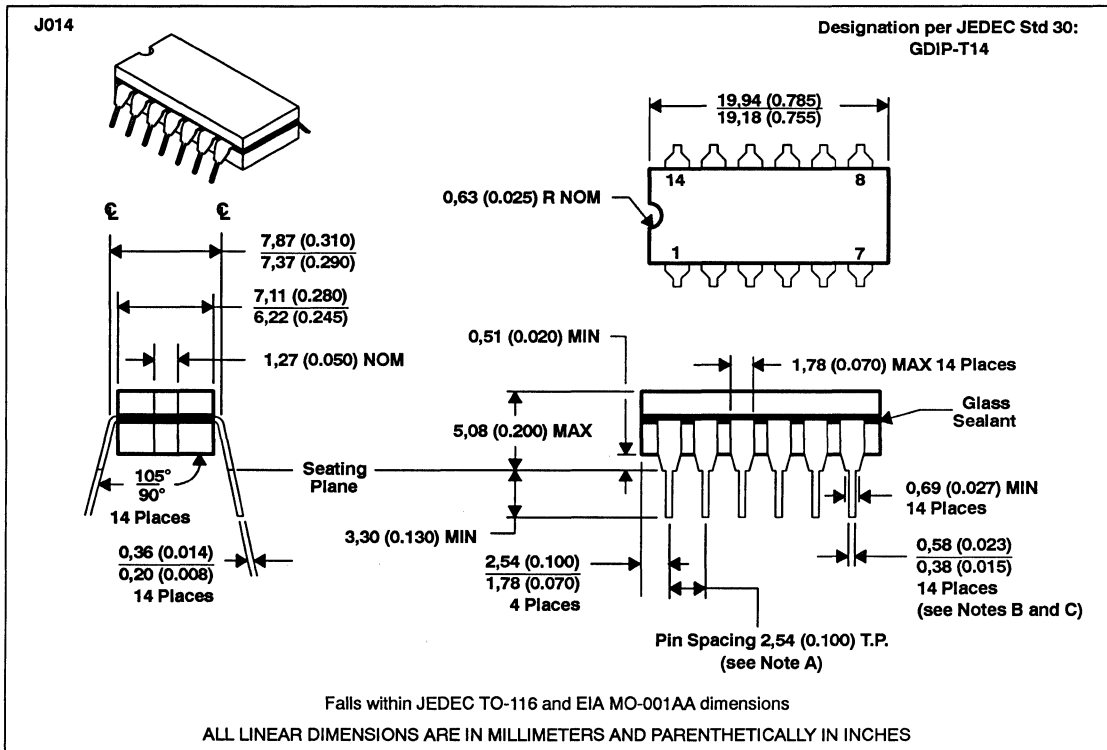
JEDEC OUTLINE	NO.OF PINS	A		A <sub>1</sub>		D, E		D <sub>1</sub> , E <sub>1</sub>		D <sub>2</sub> , E <sub>2</sub>		D <sub>3</sub> , E <sub>3</sub>
		MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	BASIC
MO-047AA	20	4,19 (0.165)	4,57 (0.180)	2,29 (0.090)	3,05 (0.120)	9,78 (0.385)	10,03 (0.395)	8,89 (0.350)	9,04 (0.356)	7,37 (0.290)	8,38 (0.330)	5,08 (0.200)
MO-047AB	28	4,19 (0.165)	4,57 (0.180)	2,29 (0.090)	3,05 (0.120)	12,32 (0.485)	12,57 (0.495)	11,43 (0.450)	11,58 (0.456)	9,91 (0.390)	10,92 (0.430)	7,62 (0.300)
MO-047AC	44	4,19 (0.165)	4,57 (0.180)	2,29 (0.090)	3,05 (0.120)	17,40 (0.685)	17,65 (0.695)	16,51 (0.650)	16,66 (0.656)	14,99 (0.590)	16,00 (0.630)	12,70 (0.500)
MO-047AD	52	4,19 (0.165)	5,08 (0.200)	2,29 (0.090)	3,30 (0.130)	19,94 (0.785)	20,19 (0.795)	19,05 (0.750)	19,20 (0.756)	17,53 (0.690)	18,54 (0.730)	15,24 (0.600)
MO-047AE	68	4,19 (0.165)	5,08 (0.200)	2,29 (0.090)	3,30 (0.130)	25,02 (0.985)	25,27 (0.995)	24,13 (0.950)	24,33 (0.956)	22,61 (0.890)	23,62 (0.930)	20,32 (0.800)
MO-047AF	84	4,19 (0.165)	5,08 (0.200)	2,29 (0.090)	3,30 (0.130)	30,10 (1.185)	30,35 (1.195)	29,21 (1.150)	29,41 (1.141)	27,69 (1.090)	28,70 (1.130)	25,40 (1.000)

NOTES A: All dimensions conform to JEDEC Specification MO-047AA/AF. Dimensions and tolerancing are per ANSI Y14.5M-1982.

F: Determined at seating plane -C-.

**J014**  
ceramic dual-in-line package

This hermetically-sealed dual-in-line package consists of a ceramic base, ceramic cap, and lead frame. Hermetic sealing is accomplished with glass. The package is intended for insertion in mounting hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Tin-plated ("bright-dipped") leads require no additional cleaning or processing when used in soldered assembly.



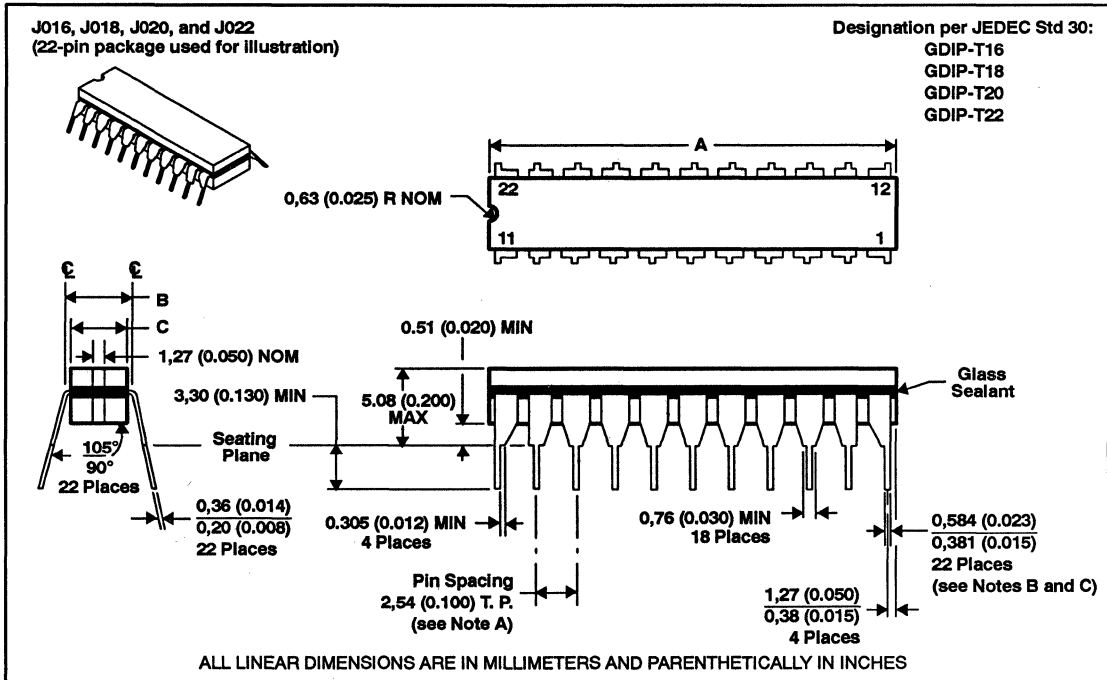
- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.  
 B. This dimension does not apply for solder-dipped leads.  
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.



# MECHANICAL DATA

## J016, J018, J020, and J022 ceramic dual-in-line

These hermetically-sealed dual-in-line packages consist of a ceramic base, ceramic cap, and a lead frame. Hermetic sealing is accomplished with glass. These packages are intended for insertion in mounting-hole rows of 7,62 (0.300) centers for the J016, J018, J020, and 10,16 (0.400) centers for the J022, respectively. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Tin-plated (bright-dipped) leads require no additional cleaning or processing when used in solder assembly.

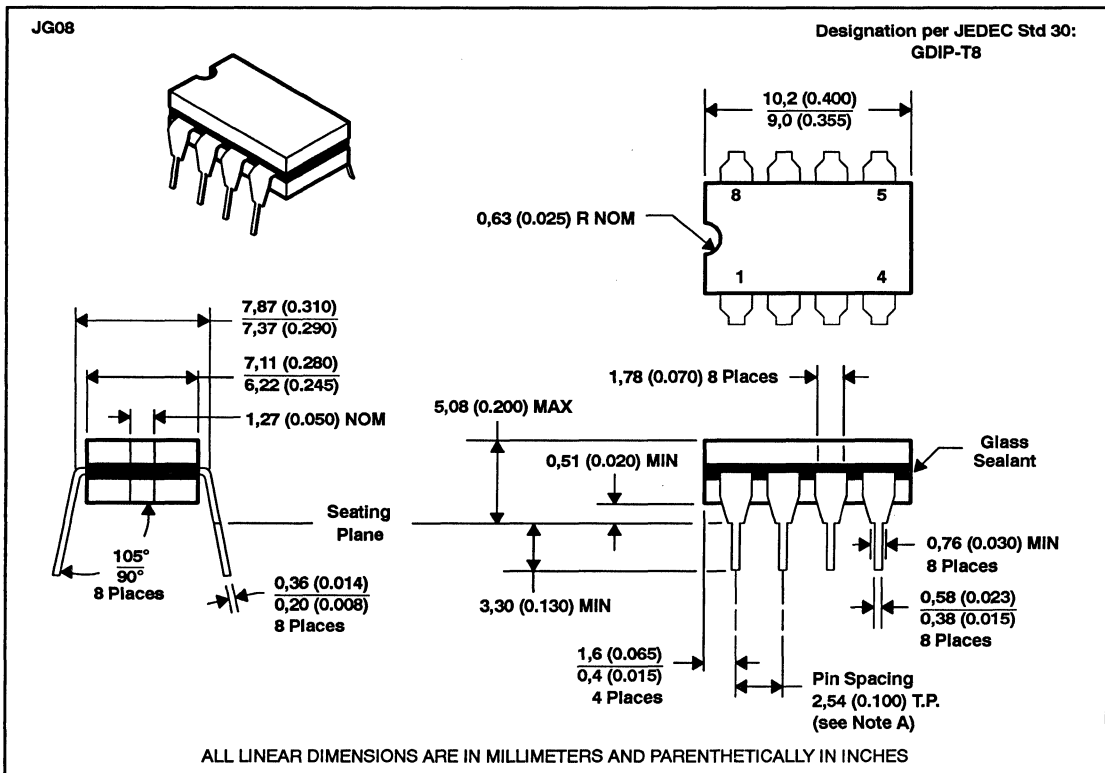


PINS	DIM	A		B		C	
		MIN	MAX	MIN	MAX	MIN	MAX
16		19,18 (0.755)	19,94 (0.785)	7,37 (0.290)	7,87 (0.310)	6,22 (0.245)	7,62 (0.300)
18			23,1 (0.910)	7,37 (0.290)	7,87 (0.310)	6,22 (0.245)	7,62 (0.300)
20		23,62 (0.930)	24,76 (0.975)	7,37 (0.290)	7,87 (0.310)	6,22 (0.245)	7,62 (0.300)
22			28,0 (1.100)	9,91 (0.390)	10,41 (0.410)		9,65 (0.388)

- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.  
B. This dimension does not apply for solder-dipped leads.  
C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above the seating plane.

**JG008**  
ceramic dual-in-line package

This hermetically-sealed dual-in-line package consists of a ceramic base, ceramic cap, and lead frame. Hermetic sealing is accomplished with glass. The package is intended for insertion in mounting hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Tin-plated ("bright-dipped") leads require no additional cleaning or processing when used in soldered assembly.

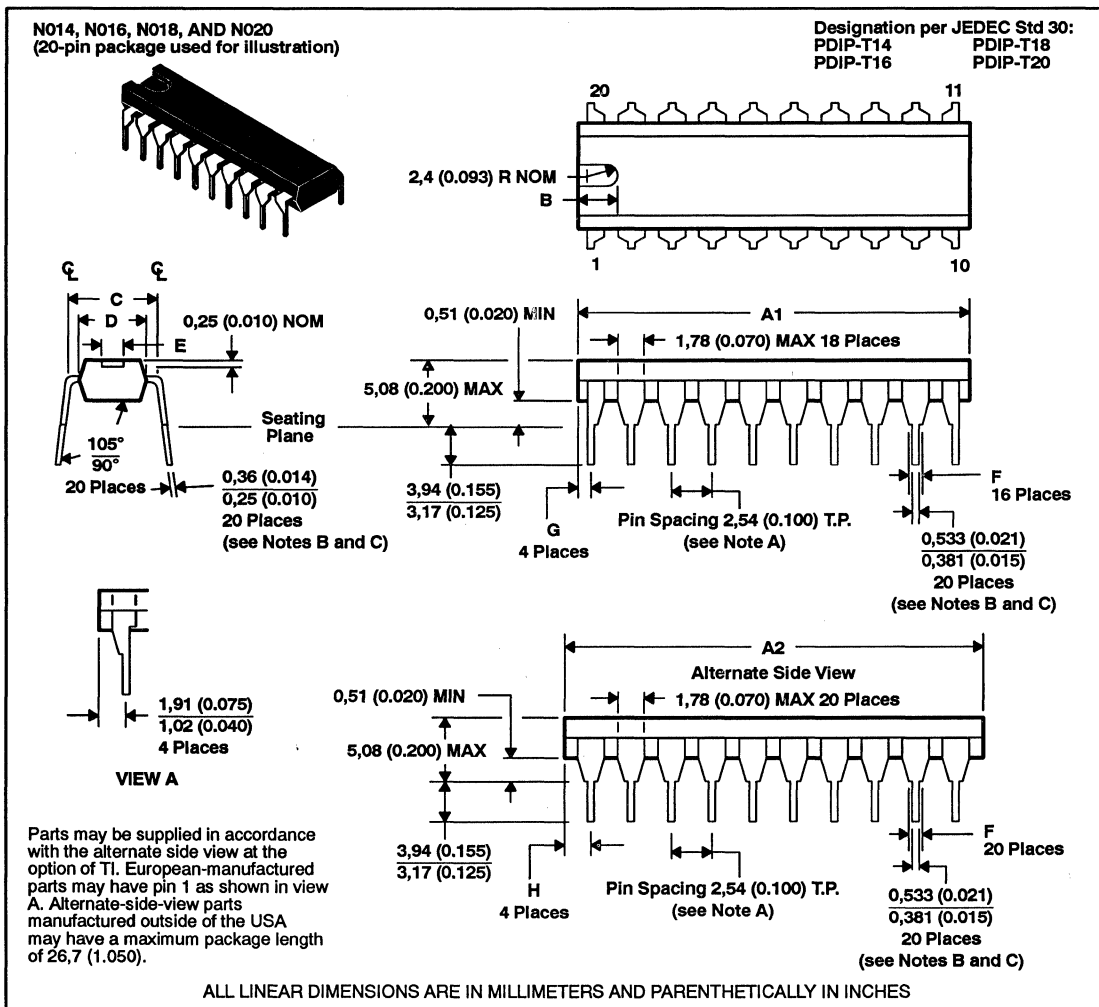


NOTE A: Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.

# MECHANICAL DATA

## N014, N016, N018, and N020 300-mil plastic dual-in-line packages

These dual-in-line packages consist of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation and circuit performance characteristics will remain stable when operated in high-humidity conditions. These packages are intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.  
B. This dimension does not apply for solder-dipped leads.  
C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

**N014, N016, N018, and N020**  
**300-mil plastic dual-in-line package (continued)**

DIM		PIN			
		14	16	18	20
A1	MIN	18,0 (0.710)			23,22 (0.914)
	MAX	19,8 (0.780)	19,8 (0.780)	23,4 (0.920)	24,77 (0.975)
A2	MIN	18,0 (0.710)			23,62 (0.930)
	MAX	19,8 (0.780)			25,4 (1.000)
B	NOM	2,8 (0.110)	2,8 (0.110)	4,06 (0.160)	2,80 (0.110)
C	MIN	7,37 (0.290)	7,37 (0.290)	7,37 (0.290)	7,37 (0.290)
	MAX	7,87 (0.310)	7,87 (0.310)	7,87 (0.310)	7,87 (0.310)
D	MIN	6,10 (0.240)	6,10 (0.240)		6,60 (0.240)
	MAX	6,60 (0.260)	6,60 (0.260)	6,99 (0.275)	7,11 (0.280)
E	NOM	2,0 (0.080)	2,0 (0.080)	2,03 (0.080)	2,0 (0.080)
F	MIN	0,84 (0.033)	0,84 (0.033)	0,89 (0.035)	0,84 (0.033)
G	MIN	See Note D	0,38 (0.015)	See Note D	1,68 (0.066)
	MAX	See Note D	1,65 (0.065)	See Note D	0,22 (0.009)
H	MIN	2,54 (0.100)	1,02 (0.040)	0,23 (0.009)	0,38 (0.015)
	MAX	1,52 (0.060)	2,41 (0.095)	1,91 (0.075)	1,27 (0.050)

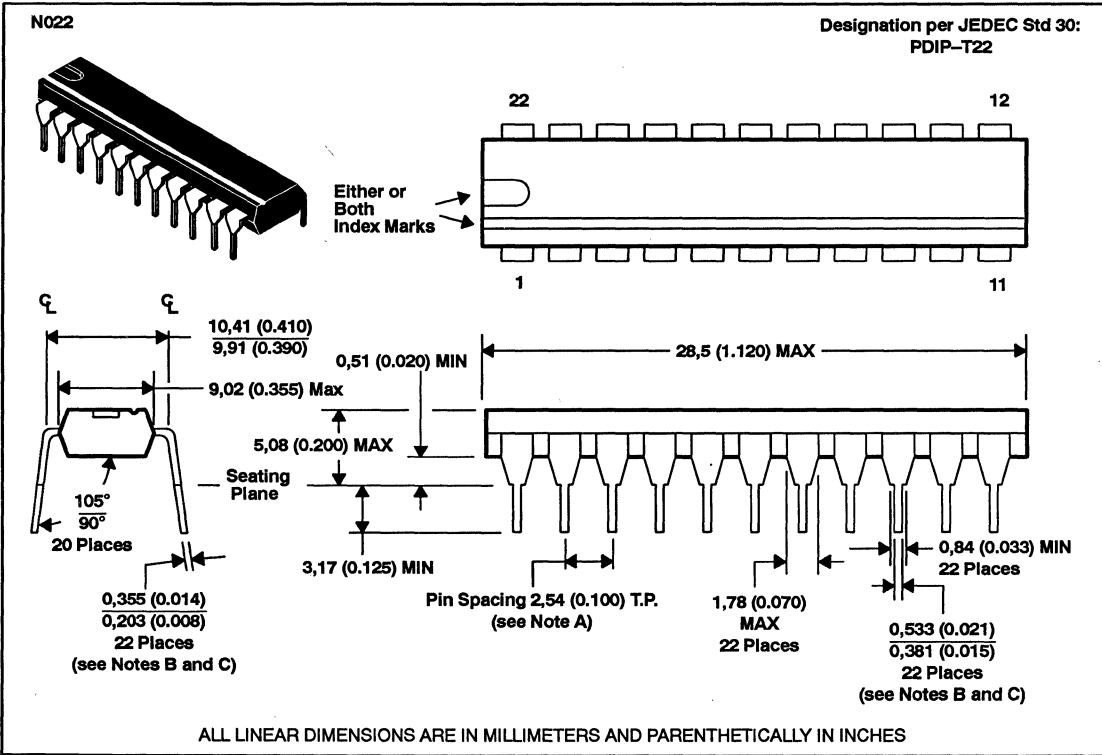
NOTE D: The 14-pin and 18-pin plastic dual-in-line package is only offered with the external pins shaped in their entirety, and do not have alternate side view dimensions.

# MECHANICAL DATA

## N022

### 400-mil plastic dual-in-line package

This dual-in-line package consist of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation and circuit performance characteristics will remain stable when operated in high-humidity conditions. This package is intended for insertion in mounting-hole rows on 10,16 (0.400) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



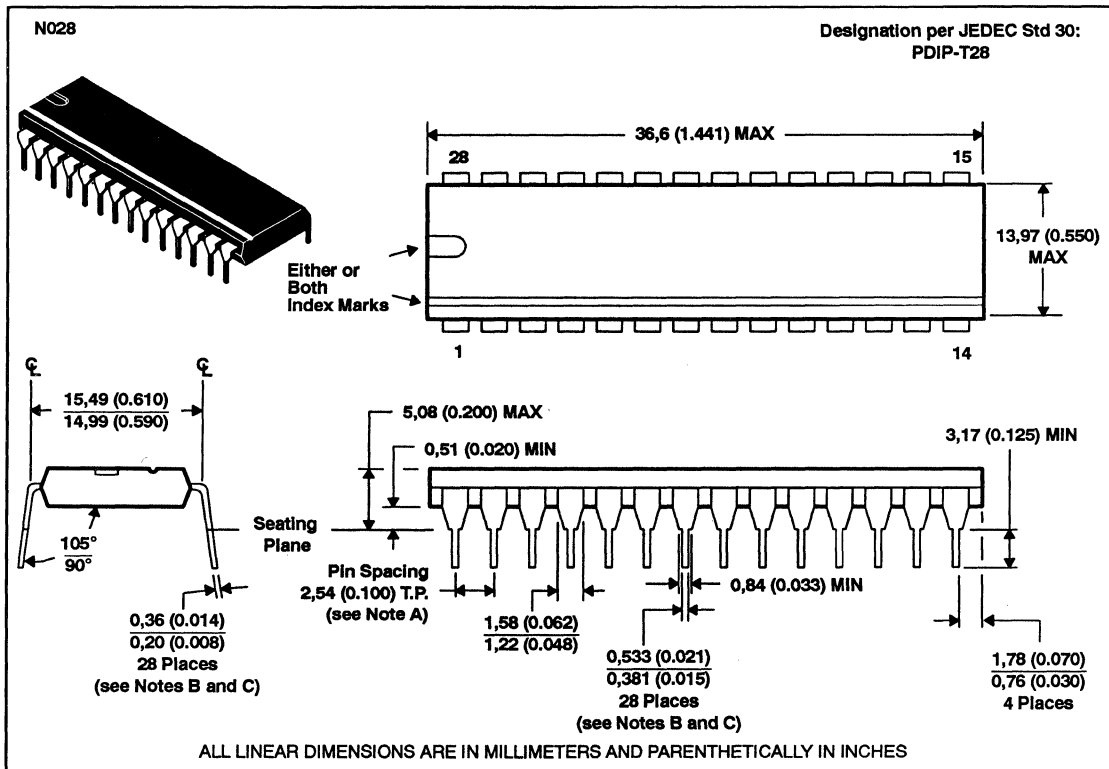
NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.

B. This dimension does not apply for solder-dipped leads.

C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

**N028**  
**600-mil plastic dual-in-line package**

This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation and circuit performance characteristics will remain stable when operated in high-humidity conditions. This package is intended for insertion in mounting-hole rows on 15,24 (0.600) centers (see Note A). Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.

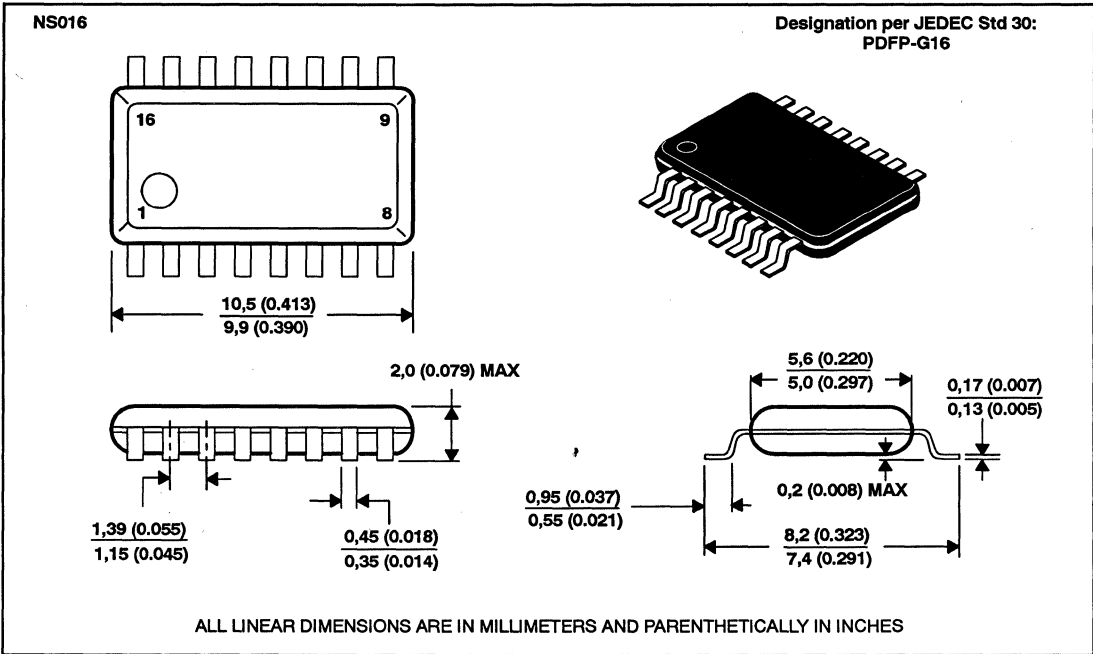


- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.  
 B. This dimension does not apply for solder-dipped leads.  
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

# MECHANICAL DATA

## NS016 plastic package

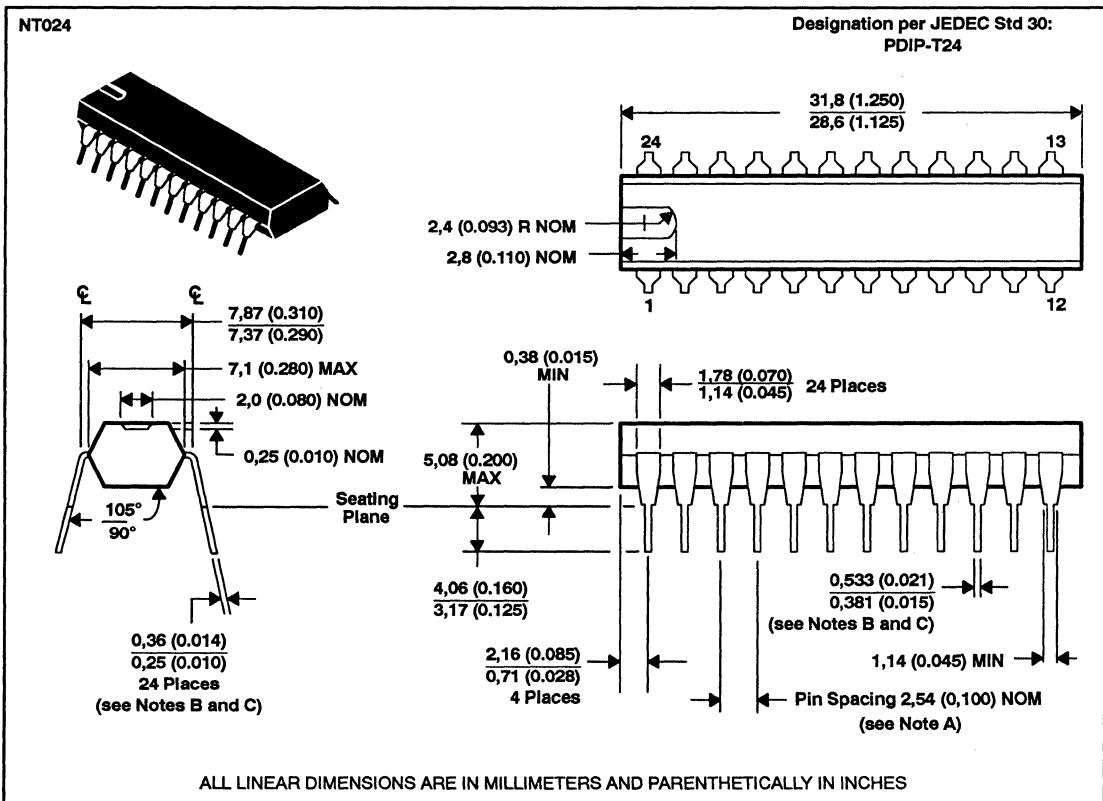
This package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound withstands soldering temperature with no deformation, and circuit performance characteristics remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



**NT024**  
**300-mil plastic dual-in-line packages**

This package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation and circuit performance characteristics will remain stable when operated in high-humidity conditions. This package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.

NOTE: For all except 24-pin package, the letter N is used by itself since the 24-pin package may be available in more than one row-spacing. For the 24-pin package, the 7,62 (0.300) version is designated NT; the 15,24 (0.600) version is designated NW. If no second letter or row-spacing is specified, the package is assumed to have 15,24 (0.600) row-spacing.



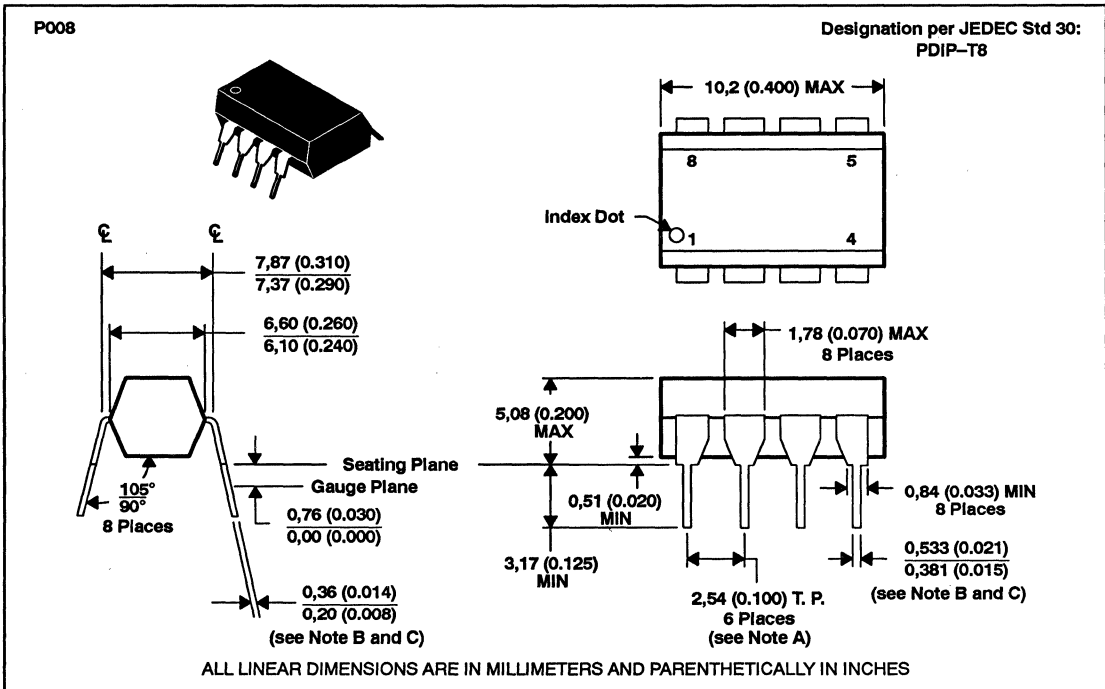
- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.  
 B. This dimension does not apply for solder-dipped leads.  
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.



# MECHANICAL DATA

## P008 plastic dual-in-line package

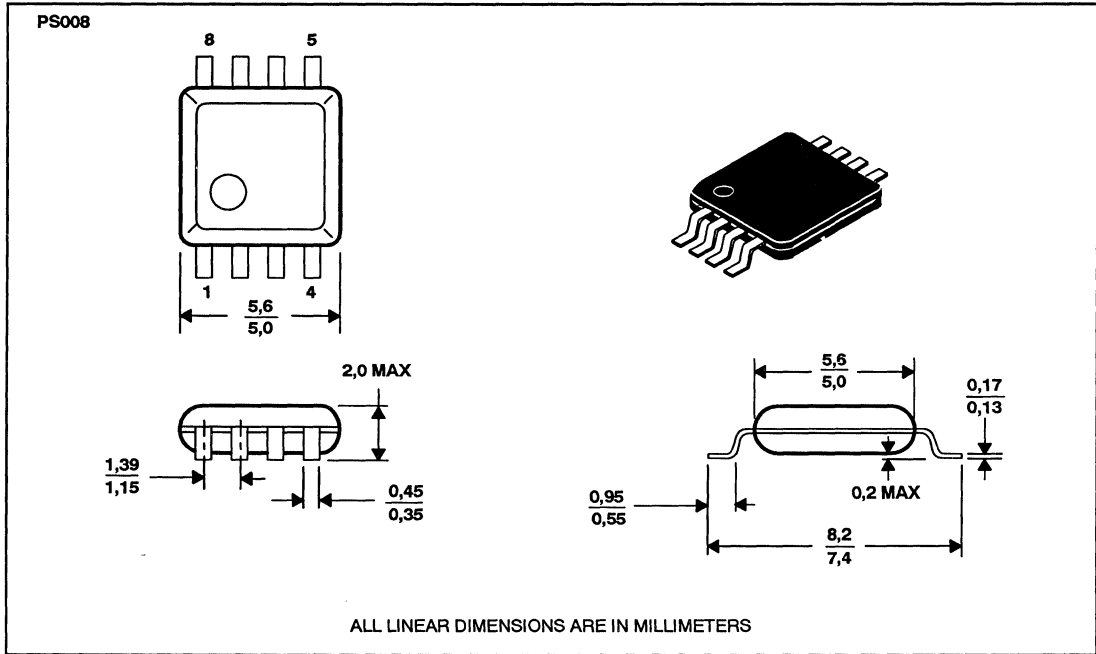
This package consists of a circuit mounted on an 8-pin lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Solder-plated lead require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. Each pin centerline is located within 0,25 (0.01) of its true longitudinal position.  
 B. This dimension does not apply for solder-dipped leads.  
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

**PS**  
**plastic dual-in-line package**

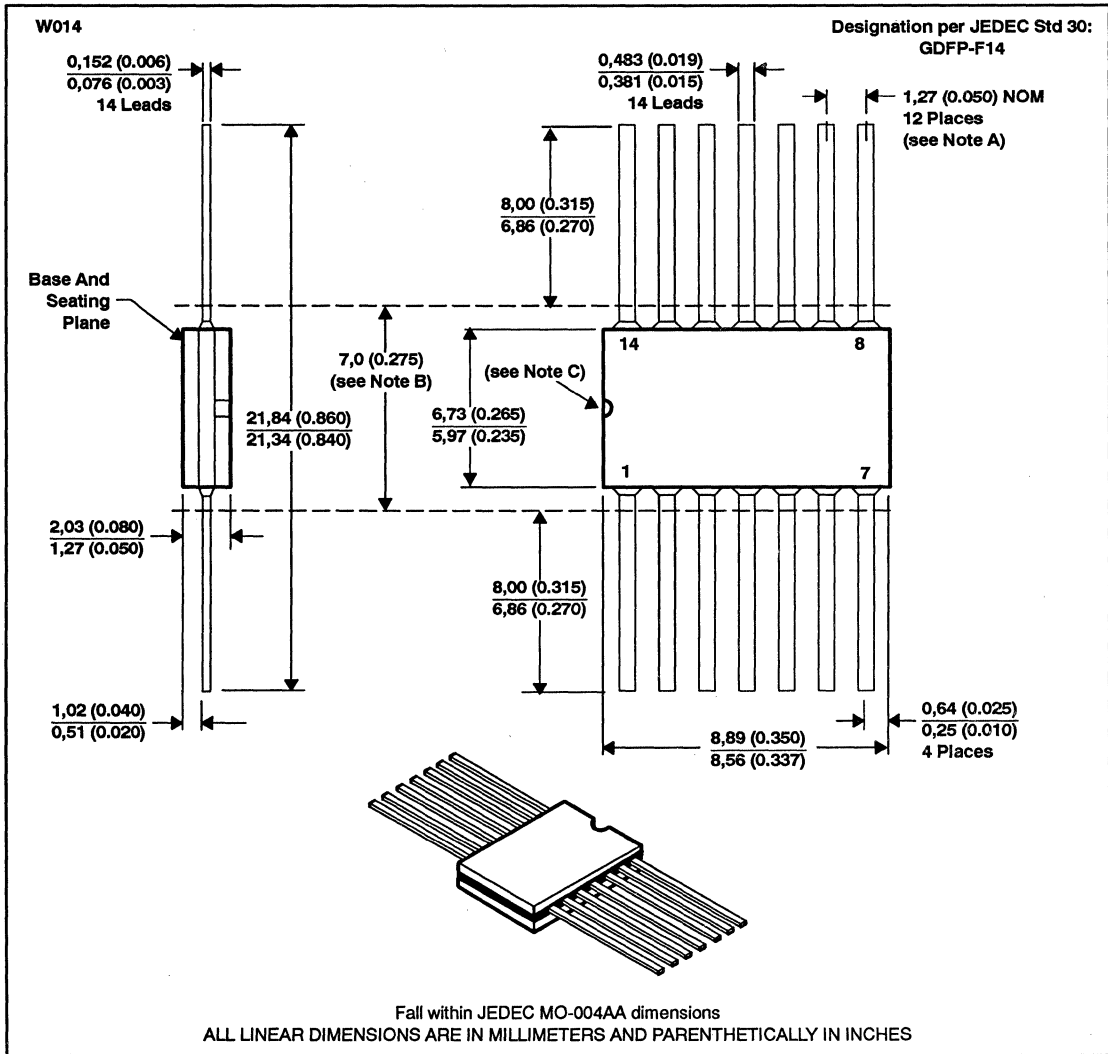
This package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound withstands soldering temperature with no deformation, and circuit performance characteristics remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



# MECHANICAL DATA

## W014 ceramic flat package

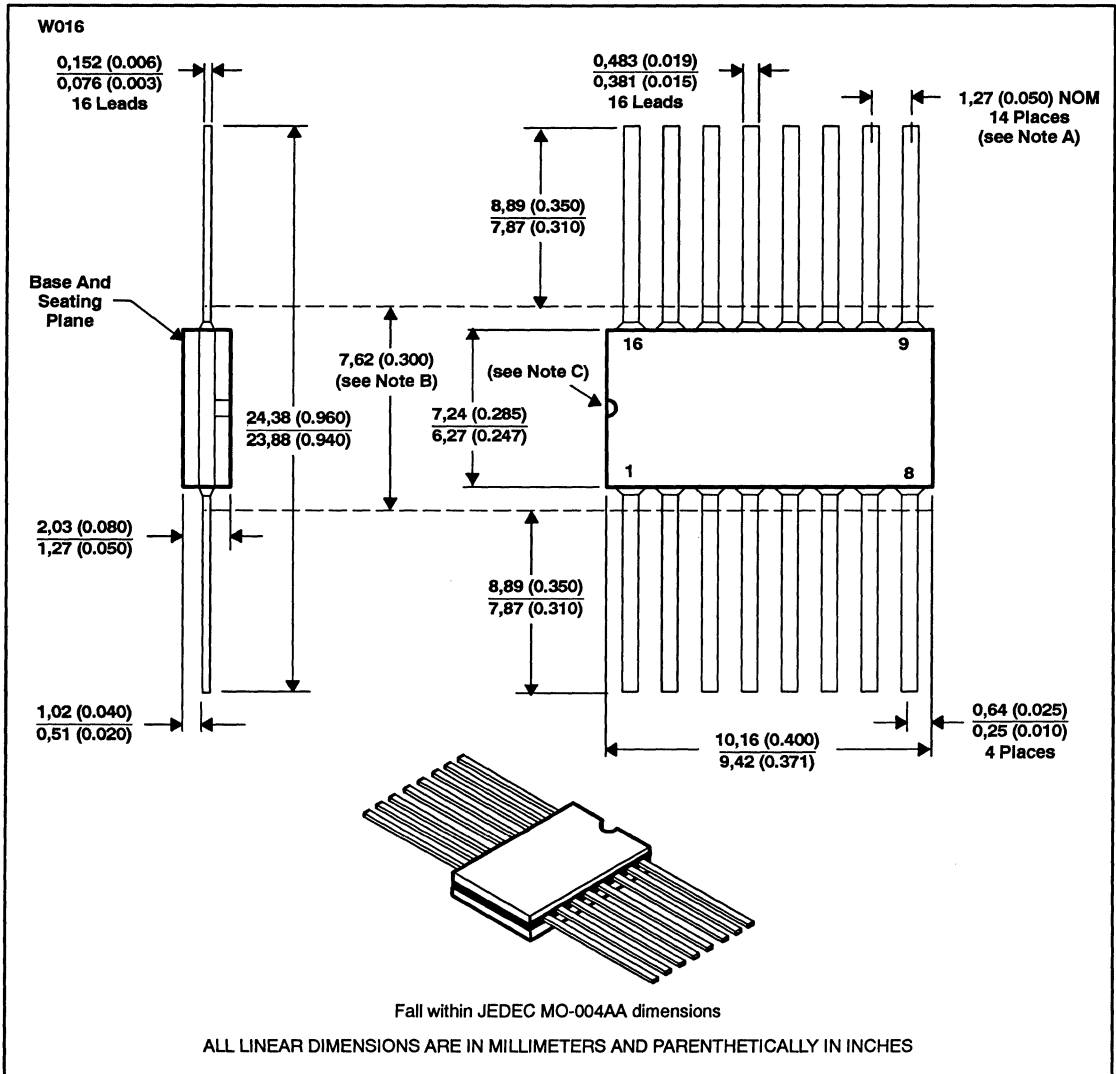
This hermetically sealed flat package consists of an electrically nonconductive ceramic base and cap and a lead frame. Hermetic sealing is accomplished with glass. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. Leads are within 0,13 (0.005) radius of true position (T.P.) at maximum material condition.  
B. This dimension determines a zone within which all body and lead irregularities lie.  
C. Index point is provided on cap for terminal identification only.

**W016**  
ceramic flat package

This hermetically sealed flat package consists of an electrically nonconductive ceramic base and cap and a lead frame. Hermetic sealing is accomplished with glass. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. Leads are within 0,13 (0.005) radius of true position (T.P.) at maximum material condition.  
 B. This dimension determines a zone within which all body and lead irregularities lie.  
 C. Index point is provided on cap for terminal identification only.



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 Quebec: Future (418) 897-6666;  
 Toronto: Arrow/Schweber (416) 670-7769;  
 Future (416) 612-9200; Marshall (416) 458-8046;  
 Vancouver: Arrow/Schweber (604) 421-2333;  
 Future (604) 294-1166.

## TI Die Processors

Chip Supply (407) 298-7100  
 Elmo Semiconductor (818) 768-7400  
 Minco Technology Labs (512) 834-2022



# TI Worldwide Sales Offices

**ALABAMA:** Huntsville: 4980 Corporate Drive, Suite 150, Huntsville, AL 35805, (205) 837-7530.

**ARIZONA:** Phoenix: 8825 N. 23rd Avenue, Suite 100, Phoenix, AZ 85021, (602) 995-1007.

**CALIFORNIA:** Irvine: 1920 Main Street, Suite 900, Irvine, CA 92714, (714) 660-1200; San Diego: 5625 Ruffin Road, Suite 100, San Diego, CA 92123, (619) 278-9600; Santa Clara: 5353 Betsy Ross Drive, Santa Clara, CA 95054, (408) 980-9000; Woodland Hills: 21550 Oxnard Street, Suite 700, Woodland Hills, CA 91367, (818) 704-8100.

**COLORADO:** Aurora: 1400 S. Potomac Street, Suite 101, Aurora, CO 80012, (303) 368-8000.

**CONNECTICUT:** Wallingford: 9 Barnes Industrial Park So., Wallingford, CT 06492, (203) 269-0074.

**FLORIDA:** Altamonte Springs: 370 S. North Lake Boulevard, Suite 1008, Altamonte Springs, FL 32701, (407) 260-2118; Fort Lauderdale: 2950 N.W. 62nd Street, Suite 100, Fort Lauderdale, FL 33309, (305) 973-8502;

Tampa: 4803 George Road, Suite 390, Tampa, FL 33634-6234, (813) 885-7588.

**GEORGIA:** Norcross: 5515 Spalding Drive, Norcross, GA 30092-2560, (404) 862-7967.

**ILLINOIS:** Arlington Heights: 515 West Algonquin, Arlington Heights, IL 60005, (708) 640-2925.

**INDIANA:** Carmel: 550 Congressional Drive, Suite 100, Carmel, IN 46032, (317) 573-6400; Fort Wayne: 103 Airport North Office Park., Fort Wayne, IN 46825, (219) 489-4697.

**KANSAS:** Overland Park: 7300 College Boulevard, Lighton Plaza, Suite 150, Overland Park, KS 66210, (913) 451-4511.

**MARYLAND:** Columbia: 8815 Centre Park Drive, Suite 100, Columbia, MD 21045, (410) 964-2003.

**MASSACHUSETTS:** Waltham: Bay Colony Corporate Center, 950 Winter Street, Suite 2800, Waltham, MA 02154, (617) 895-9100.

**MICHIGAN:** Farmington Hills: 33737 W. 12 Mile Road, Farmington Hills, MI 48018, (313) 553-1581;

**MINNESOTA:** Eden Prairie: 11000 W. 78th Street, Suite 100, Eden Prairie, MN 55344, (612) 828-9300.

**MISSOURI:** St. Louis: 12412 Powerscourt Drive, Suite 125, St. Louis, MO 63131, (314) 821-8400.

**NEW JERSEY:** Iselin: Metropolitan Corporate Plaza, 485 Bldg. E. U.S. 1 South, Iselin, NJ 08830, (908) 750-1050.

**NEW MEXICO:** Albuquerque: 2709 J. Pan American Freeway NE, Albuquerque, NM 87101, (505) 345-2555.

**NEW YORK:** East Syracuse: 6365 Collamer Drive, East Syracuse, NY 13057, (315) 463-9291; Fishkill: 300 Westage Business Center, Suite 140, Fishkill, NY 12524, (914) 897-2900;

Melville: 48 South Service Road, Suite 100, Melville, NY 11747, (516) 454-6601; Pittsford: 2851 Clover Street, Pittsford, NY 14534, (716) 385-6770.

**NORTH CAROLINA:** Charlotte: 8 Woodlawn Green, Suite 100, Charlotte, NC 28217, (704) 527-0930;

Raleigh: 2809 Highwoods Boulevard, Suite 100, Raleigh, NC 27625, (919) 876-2725.

**OHIO:** Beachwood: 23775 Commerce Park Road, Beachwood, OH 44122-5875, (216) 765-7528; Beavercreek: 4200 Colonel Glenn Highway, Suite 800, Beavercreek, OH 45431, (513) 427-6200.

**OREGON:** Beaverton: 6700 S.W. 105th Street, Suite 110, Beaverton, OR 97005, (503) 643-8758.

**PENNSYLVANIA:** Plymouth Meeting: 600 West Germantown Pike, Suite 200, Plymouth Meeting, PA 19462, (215) 825-9500.

**PUERTO RICO:** Hato Rey: 615 Mercantil Plaza Building, Suite 505, Hato Rey, PR 00919, (809) 753-8700.

**TEXAS:** Austin: 12501 Research Boulevard, Austin, TX 78759, (512) 250-6769;

Dallas: 7839 Churchill Way, Dallas, TX 75251, (214) 917-1264;

Houston: 9301 Southwest Freeway, Commerce Park, Suite 360, Houston, TX 77074, (713) 778-6592;

Midland: FM 1788 & I-20, Midland, TX 79711-0448, (915) 561-7137.

**UTAH:** Salt Lake City: 2180 South 1300 East, Suite 335, Salt Lake City, UT 84106, (801) 466-8972.

**WISCONSIN:** Waukeesa: 20825 Swenson Drive, Suite 900, Waukeesa, WI 53186, (414) 798-1001.

**CANADA:** Nepean: 301 Moodle Drive, Suite 102, Mallom Centre, Nepean, Ontario, Canada K2H 9C4, (613) 726-1970;

Richmond Hill: 1200 Centre Street East, Richmond Hill, Ontario, Canada L4C 1B1, (416) 884-9181;

St. Laurent: 9460 Trans Canada Highway, St. Laurent, Quebec, Canada H4S 1R7, (514) 335-8392.

**AUSTRALIA (& NEW ZEALAND):** Texas Instruments Australia Ltd., 6-10 Talavera Road, North Ryde (Sydney), New South Wales, Australia 2113, 2-878-9000; 14th Floor, 380 Street, Kilda Road, Melbourne, Victoria, Australia 3000, 3-696-1211.

**BELGIUM:** Texas Instruments Belgium S.A./N.V., Avenue Jules Bordetlaan 11, 1140 Brussels, Belgium, (02) 242 30 80.

**BRAZIL:** Texas Instrumentos Electronicos do Brasil Ltda., Av. Eng. Luiz Carlos Berrini, 1461-110, andar, 04571, Sao Paulo, SP, Brazil, 11-535-5133.

**DENMARK:** Texas Instruments A/S, Borupvang 2D, DK-2750 Ballerup, Denmark, (44) 68 74 00.

**FINLAND:** Texas Instruments OY, Ahertajantie 3, P.O. Box 88, 02321 Espoo, Finland, (0) 802 6517.

**FRANCE:** Texas Instruments France, 8-10 Avenue Morane-Saunier, B.P. 67, 78141 Vélizy-Villacoublay Cedex, France, (1) 30 70 1003.

**GERMANY:** Texas Instruments Deutschland GmbH, Haggertystraße 1, 8050 Freising, (08161) 80-0; Kurfürstendamm 195-196, 1000 Berlin 15, (030) 8 82 73 65; Düsseldorf: Straße 40, 6236 Eschborn 1, (06196) 80 70; Hollestraße 3, 4300 Essen 1, (0201) 23 66 40; Kirchhorster Straße 2, 3000 Hannover 51, (0511) 64 68-0; Maybachstraße II, 7302 Ostfildern 2 (Nellingen), (0711) 3003 257.

**HOLLAND:** Texas Instruments Holland B.V., Hogehilweg 19, Postbus 12995, 1100 AZ Amsterdam-Zuidoost, Holland, (020) 5602911.

**HONG KONG:** Texas Instruments Hong Kong Ltd., 8th Floor, World Shipping Centre, 7 Canton Road, Kowloon, Hong Kong, 737-0338.

**HUNGARY:** Texas Instruments Representation, Budaörsi u.42, H-1112 Budapest, Hungary, (1) 1 66 66 17.

**IRELAND:** Texas Instruments Ireland Ltd., 7/8 Harcourt Street, Dublin 2, Ireland, (01) 755233.

**ITALY:** Texas Instruments Italia S.p.A., Centro Direzionale Colleoni, Palazzo Perseo-Via Paracelso 12, 20041 Agrate Brianza (MI), Italy, (039) 63221; Via Castello della Magliana, 38, 00148 Roma, Italy (6) 657 2651.

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**KOREA:** Texas Instruments Korea Ltd., 28th Floor, Trade Tower, 159-1, Samsung-Dong, Kangnam-ku Seoul, Korea, 2-551-2800.

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**MEXICO:** Texas Instruments de Mexico S.A. de C.V., Alfonso Reyes 115, Col. Hipodromo Condesa, Mexico, D.F., 06170, 5-515-6081.

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**PHILIPPINES:** Texas Instruments Asia Ltd., Philippines Branch, 14th Floor, Ba-Lepanto Building, 8747 Paseo de Roxas, 1226 Makati, Metro Manila, Philippines, 2-817-6031.

**PORTUGAL:** Texas Instruments Equipamento Electronico (Portugal) LDA., Ing. Frederico Ulricho, 2650 Moreira Da Maia, 4470 Maia, Portugal (2) 948 1003.

**SINGAPORE (& INDIA, INDONESIA, THAILAND):** Texas Instruments Singapore (PTE) Ltd., Asia Pacific, 101 Thomson Road, #23-01, United Square, Singapore 1130, 350-8100.

**SPAIN:** Texas Instruments España S.A., c/Gobelias 9, Urbanizacion La Florida, 28023, Madrid, Spain, (91) 372 8051; c/Diputacion, 279-3-5, 08007 Barcelona, Spain, (93) 317 91 80.

**SWEDEN:** Texas Instruments International Trade Corporation (Sverigefilialen), Box 30, S-164 93 Kista, Sweden, (08) 752 58 00.

**SWITZERLAND:** Texas Instruments Switzerland AG, Riedstrasse 6, CH-8953 Dietikon, Switzerland, (01) 744 2811.

**TAIWAN:** Texas Instruments Taiwan Limited, Taipei Branch, 10th Floor, Bank Tower, 205 Tung Hua N. Road, Taipei, Taiwan, 10592, Republic of China, (02) 713 9311.

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